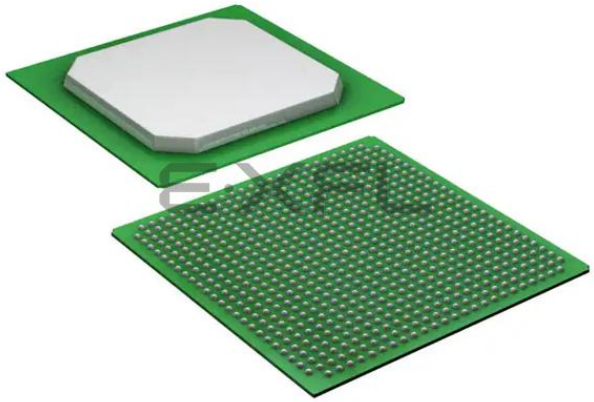

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	2076
Number of Logic Elements/Cells	33216
Total RAM Bits	483840
Number of I/O	475
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	672-BGA
Supplier Device Package	672-FBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep2c35f672c6

Table 1–1. Cyclone II FPGA Family Features (Part 2 of 2)

Feature	EP2C5 (2)	EP2C8 (2)	EP2C15 (1)	EP2C20 (2)	EP2C35	EP2C50	EP2C70
Maximum user I/O pins	158	182	315	315	475	450	622

Notes to Table 1–1:

(1) The EP2C15A is only available with the Fast On feature, which offers a faster POR time. This device is available in both commercial and industrial grade.

(2) The EP2C5, EP2C8, and EP2C20 optionally support the Fast On feature, which is designated with an "A" in the device ordering code. The EP2C5A is only available in the automotive speed grade. The EP2C8A and EP2C20A devices are only available in industrial grade.

(3) This is the total number of 18 × 18 multipliers. For the total number of 9 × 9 multipliers per device, multiply the total number of 18 × 18 multipliers by 2.

This gives a maximum of seven control signals at a time. When using the LAB-wide synchronous load, the `clkena` of `labclk1` is not available. Additionally, register packing and synchronous load cannot be used simultaneously.

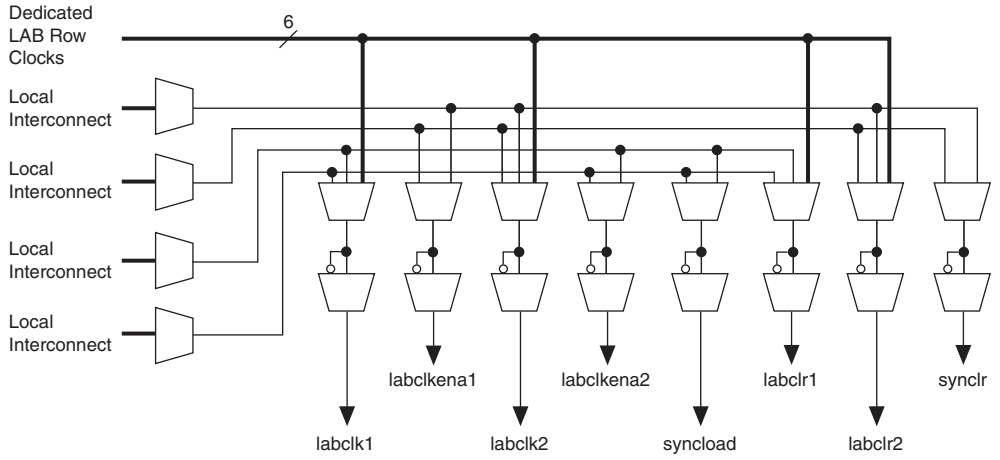
Each LAB can have up to four non-global control signals. Additional LAB control signals can be used as long as they are global signals.

Synchronous clear and load signals are useful for implementing counters and other functions. The synchronous clear and synchronous load signals are LAB-wide signals that affect all registers in the LAB.

Each LAB can use two clocks and two clock enable signals. Each LAB's clock and clock enable signals are linked. For example, any LE in a particular LAB using the `labclk1` signal also uses `labclkena1`. If the LAB uses both the rising and falling edges of a clock, it also uses both LAB-wide clock signals. De-asserting the clock enable signal turns off the LAB-wide clock.

The LAB row clocks [5..0] and LAB local interconnect generate the LAB-wide control signals. The MultiTrack™ interconnect's inherent low skew allows clock and control signal distribution in addition to data. Figure 2–7 shows the LAB control signal generation circuit.

Figure 2–7. LAB-Wide Control Signals



LAB-wide signals control the logic for the register's clear signal. The LE directly supports an asynchronous clear function. Each LAB supports up to two asynchronous clear signals (`labclr1` and `labclr2`).

Global Clock Network

The 16 or 8 global clock networks drive throughout the entire device. Dedicated clock pins (CLK[]), PLL outputs, the logic array, and dual-purpose clock (DPCLK[]) pins can also drive the global clock network.

The global clock network can provide clocks for all resources within the device, such as IOEs, LEs, memory blocks, and embedded multipliers. The global clock lines can also be used for control signals, such as clock enables and synchronous or asynchronous clears fed from the external pin, or DQS signals for DDR SDRAM or QDRII SRAM interfaces. Internal logic can also drive the global clock network for internally generated global clocks and asynchronous clears, clock enables, or other control signals with large fan-out.

Clock Control Block

There is a clock control block for each global clock network available in Cyclone II devices. The clock control blocks are arranged on the device periphery and there are a maximum of 16 clock control blocks available per Cyclone II device. The larger Cyclone II devices (EP2C15 devices and larger) have 16 clock control blocks, four on each side of the device. The smaller Cyclone II devices (EP2C5 and EP2C8 devices) have eight clock control blocks, four on the left and right sides of the device.

The control block has these functions:

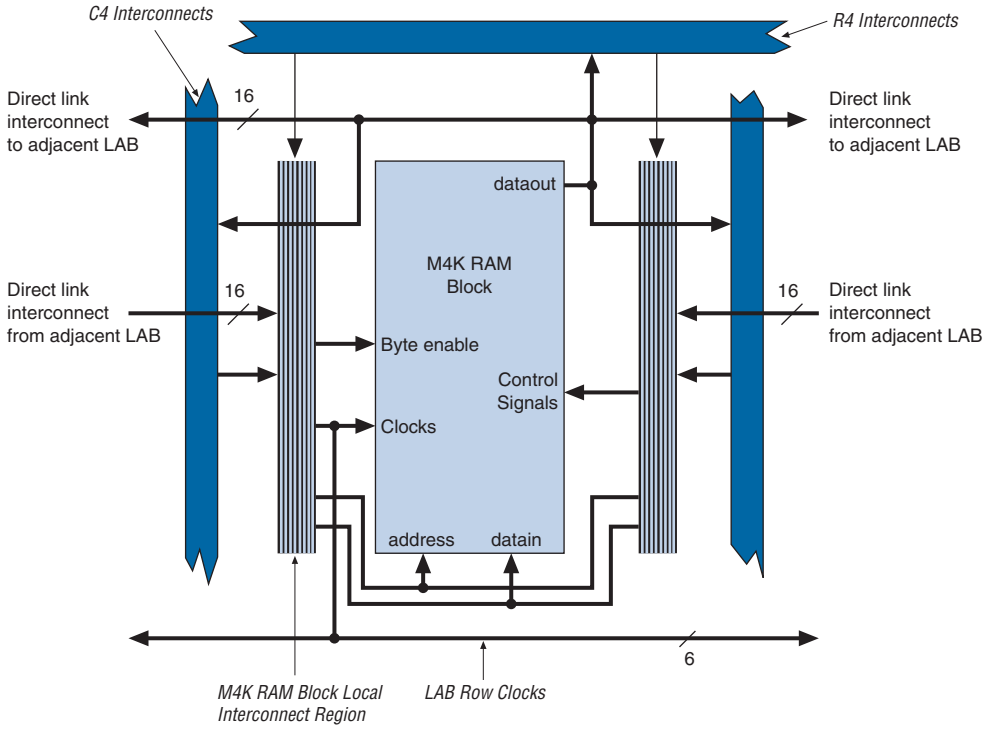
- Dynamic global clock network clock source selection
- Dynamic enable/disable of the global clock network

In Cyclone II devices, the dedicated CLK[] pins, PLL counter outputs, DPCLK[] pins, and internal logic can all feed the clock control block. The output from the clock control block in turn feeds the corresponding global clock network.

The following sources can be inputs to a given clock control block:

- Four clock pins on the same side as the clock control block
- Three PLL clock outputs from a PLL
- Four DPCLK pins (including CDPCLK pins) on the same side as the clock control block
- Four internally-generated signals

Figure 2–17. M4K RAM Block LAB Row Interface



For more information on Cyclone II embedded memory, see the *Cyclone II Memory Blocks* chapter in Volume 1 of the *Cyclone II Device Handbook*.

Embedded Multipliers

Cyclone II devices have embedded multiplier blocks optimized for multiplier-intensive digital signal processing (DSP) functions, such as finite impulse response (FIR) filters, fast Fourier transform (FFT) functions, and discrete cosine transform (DCT) functions. You can use the embedded multiplier in one of two basic operational modes, depending on the application needs:

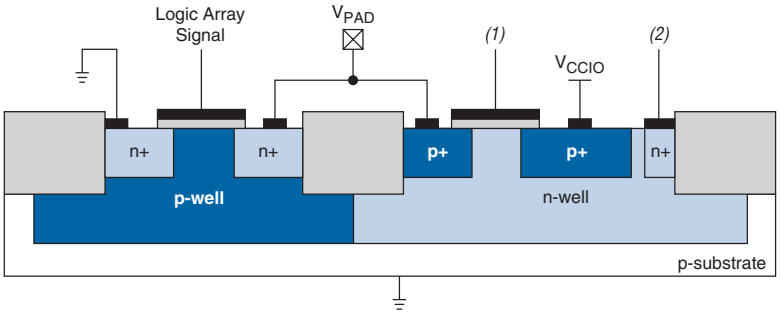
- One 18-bit multiplier
- Up to two independent 9-bit multipliers

Cyclone II devices also use the JTAG port to monitor the logic operation of the device with the SignalTap® II embedded logic analyzer. Cyclone II devices support the JTAG instructions shown in Table 3–1.

Table 3–1. Cyclone II JTAG Instructions (Part 1 of 2)

JTAG Instruction	Instruction Code	Description
SAMPLE/PRELOAD	00 0000 0101	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins. Also used by the SignalTap II embedded logic analyzer.
EXTEST *(1)*	00 0000 1111	Allows the external circuitry and board-level interconnects to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	11 1111 1111	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through selected devices to adjacent devices during normal device operation.
USERCODE	00 0000 0111	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO.
IDCODE	00 0000 0110	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.
HIGHZ *(1)*	00 0000 1011	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through selected devices to adjacent devices during normal device operation, while tri-stating all of the I/O pins.
CLAMP *(1)*	00 0000 1010	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through selected devices to adjacent devices during normal device operation while holding I/O pins to a state defined by the data in the boundary-scan register.
ICR instructions		Used when configuring a Cyclone II device via the JTAG port with a USB Blaster™, ByteBlaster™ II, MasterBlaster™ or ByteBlasterMV™ download cable, or when using a Jam File or JBC File via an embedded processor.
PULSE_NCONFIG	00 0000 0001	Emulates pulsing the nCONFIG pin low to trigger reconfiguration even though the physical pin is unaffected.

Figure 4–2. Transistor Level Diagram of FPGA Device I/O Buffers



Notes to Figure 4–2:
(1) This is the logic array signal or the larger of either the V_{CCIO} or V_{PAD} signal.
(2) This is the larger of either the V_{CCIO} or V_{PAD} signal.

Power-On Reset Circuitry

Cyclone II devices contain POR circuitry to keep the device in a reset state until the power supply voltage levels have stabilized during power-up. The POR circuit monitors the V_{CCINT} voltage levels and tri-states all user I/O pins until the V_{CC} reaches the recommended operating levels. In addition, the POR circuitry also monitors the V_{CCIO} level of the two I/O banks that contains configuration pins (I/O banks 1 and 3 for EP2C5 and EP2C8, I/O banks 2 and 6 for EP2C15A, EP2C20, EP2C35, EP2C50, and EP2C70) and tri-states all user I/O pins until the V_{CC} reaches the recommended operating levels.

After the Cyclone II device enters user mode, the POR circuit continues to monitor the V_{CCINT} voltage level so that a brown-out condition during user mode can be detected. If the V_{CCINT} voltage sags below the POR trip point during user mode, the POR circuit resets the device. If the V_{CCIO} voltage sags during user mode, the POR circuit does not reset the device.

"Wake-up" Time for Cyclone II Devices

In some applications, it may be necessary for a device to wake up very quickly in order to begin operation. The Cyclone II device family offers the Fast-On feature to support fast wake-up time applications. Devices that support the Fast-On feature are designated with an "A" in the ordering code and have stricter power up requirements compared to non-A devices.

Table 5–8 shows the recommended operating conditions for user I/O pins with differential I/O standards.

Table 5–8. Recommended Operating Conditions for User I/O Pins Using Differential Signal I/O Standards

I/O Standard	V_{CCIO} (V)			V_{ID} (V) *(1)*			V_{ICM} (V)			V_{IL} (V)		V_{IH} (V)	
	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Max	Min	Max
LVDS	2.375	2.5	2.625	0.1	—	0.65	0.1	—	2.0	—	—	—	—
Mini-LVDS *(2)*	2.375	2.5	2.625	—	—	—	—	—	—	—	—	—	—
RSDS *(2)*	2.375	2.5	2.625	—	—	—	—	—	—	—	—	—	—
LVPECL *(3) (6)*	3.135	3.3	3.465	0.1	0.6	0.95	—	—	—	0	2.2	2.1	2.88
Differential 1.5-V HSTL class I and II *(4)*	1.425	1.5	1.575	0.2	—	V_{CCIO} + 0.6	0.68	—	0.9	—	V_{REF} – 0.20	V_{REF} + 0.20	—
Differential 1.8-V HSTL class I and II *(4)*	1.71	1.8	1.89	—	—	—	—	—	—	—	V_{REF} – 0.20	V_{REF} + 0.20	—
Differential SSTL-2 class I and II *(5)*	2.375	2.5	2.625	0.36	—	V_{CCIO} + 0.6	0.5 × V_{CCIO} – 0.2	0.5 × V_{CCIO}	0.5 × V_{CCIO} + 0.2	—	V_{REF} – 0.35	V_{REF} + 0.35	—
Differential SSTL-18 class I and II *(5)*	1.7	1.8	1.9	0.25	—	V_{CCIO} + 0.6	0.5 × V_{CCIO} – 0.2	0.5 × V_{CCIO}	0.5 × V_{CCIO} + 0.2	—	V_{REF} – 0.25	V_{REF} + 0.25	—

Notes to Table 5–8:

(1) Refer to the *High-Speed Differential Interfaces in Cyclone II Devices* chapter of the *Cyclone II Device Handbook* for measurement conditions on V_{ID}.

(2) The RSDS and mini-LVDS I/O standards are only supported on output pins.

(3) The LVPECL I/O standard is only supported on clock input pins. This I/O standard is not supported on output pins.

(4) The differential 1.8-V and 1.5-V HSTL I/O standards are only supported on clock input pins and PLL output clock pins.

(5) The differential SSTL-18 and SSTL-2 I/O standards are only supported on clock input pins and PLL output clock pins.

(6) The LVPECL clock inputs are powered by V_{CCINT} and support all V_{CCIO} settings. However, it is recommended to connect V_{CCIO} to typical value of 3.3V.

Table 5–16. LE_FF Internal Timing Microparameters (Part 2 of 2)

Parameter	–6 Speed Grade (1)		–7 Speed Grade (2)		–8 Speed Grade (3)		Unit
	Min	Max	Min	Max	Min	Max	
TPRE	191	—	244	—	244	—	ps
	—	—	217	—	244	—	ps
TCLKL	1000	—	1242	—	1242	—	ps
	—	—	1111	—	1242	—	ps
TCLKH	1000	—	1242	—	1242	—	ps
	—	—	1111	—	1242	—	ps
tLUT	180	438	172	545	172	651	ps
	—	—	180	—	180	—	ps

Notes to Table 5–16:

(1) For the –6 speed grades, the minimum timing is for the commercial temperature grade. The –7 speed grade devices offer the automotive temperature grade. The –8 speed grade devices offer the industrial temperature grade.

(2) For each parameter of the –7 speed grade columns, the value in the first row represents the minimum timing parameter for automotive devices. The second row represents the minimum timing parameter for commercial devices.

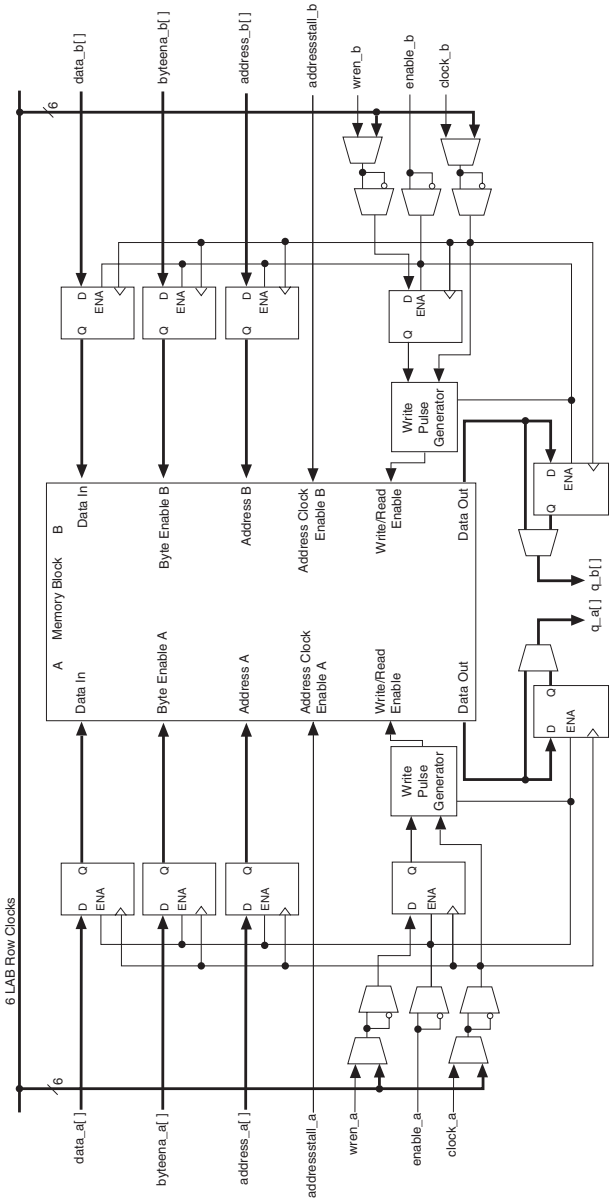
(3) For each parameter of the –8 speed grade columns, the value in the first row represents the minimum timing parameter for industrial devices. The second row represents the minimum timing parameter for commercial devices.

Table 5–17. IOE Internal Timing Microparameters (Part 1 of 2)

Parameter	–6 Speed Grade (1)		–7 Speed Grade (2)		–8 Speed Grade (3)		Unit
	Min	Max	Min	Max	Min	Max	
TSU	76	—	101	—	101	—	ps
	—	—	89	—	101	—	ps
TH	88	—	106	—	106	—	ps
	—	—	97	—	106	—	ps
TCO	99	155	95	171	95	187	ps
	—	—	99	—	99	—	ps
TPIN2COMBOUT_R	384	762	366	784	366	855	ps
	—	—	384	—	384	—	ps
TPIN2COMBOUT_C	385	760	367	783	367	854	ps
	—	—	385	—	385	—	ps
TCOMBIN2PIN_R	1344	2490	1280	2689	1280	2887	ps
	—	—	1344	—	1344	—	ps

In addition, the clock networks in the Cyclone II device support dynamic selection of the clock source and also support a power-down mode where clock networks that are not being used can easily be turned off, reducing the overall power consumption of the device.

Figure 8–13. Cyclone II Memory Block in Independent Clock Mode *Note (1)*



Note to Figure 8–13:

(1) Violating the setup or hold time on the memory block address registers could corrupt memory contents. This applies to both read and write operations.

9. External Memory Interfaces

CII51009-3.1

Introduction

Improving data bandwidth is an important design consideration when trying to enhance system performance without complicating board design. Traditionally, doubling the data bandwidth of a system required either doubling the system frequency or doubling the number of data I/O pins. Both methods are undesirable because they complicate the overall system design and increase the number of I/O pins. Using double data rate (DDR) I/O pins to transmit and receive data doubles the data bandwidth while keeping I/O counts low. The DDR architecture uses both edges of a clock to transmit data, which facilitates data transmission at twice the rate of a single data rate (SDR) architecture using the same clock speed while maintaining the same number of I/O pins. DDR transmission should be used where fast data transmission is required for a broad range of applications such as networking, communications, storage, and image processing.

Cyclone® II devices support a broad range of external memory interfaces, such as SDR SDRAM, DDR SDRAM, DDR2 SDRAM, and QDRII SRAM. Dedicated clock delay control circuitry allows Cyclone II devices to interface with an external memory device at clock speeds up to 167 MHz/333 Mbps for DDR and DDR2 SDRAM devices and 167 MHz/667 Mbps for QDRII SRAM devices. Although Cyclone II devices also support SDR SDRAM, this chapter focuses on the implementations of a double data rate I/O interface using the hardware features available in Cyclone II devices and explains briefly how each memory standard uses the Cyclone II features.

The easiest way to interface to external memory devices is by using one of the Altera® external memory IP cores listed below.

- DDR2 SDRAM Controller MegaCore® Function
- DDR SDRAM Controller MegaCore Function
- QDRII SRAM Controller MegaCore Function

OpenCore® Plus evaluations of these cores are available for free to Quartus® II Web Edition software users. In addition, Altera software subscription customers now receive full licenses to these MegaCore functions as part of the IP-BASE suite.

Section IV. I/O Standards

This section provides information on Cyclone® II single-ended, voltage referenced, and differential I/O standards.

This section includes the following chapters:

- Chapter 10, Selectable I/O Standards in Cyclone II Devices
- Chapter 11, High-Speed Differential Interfaces in Cyclone II Devices

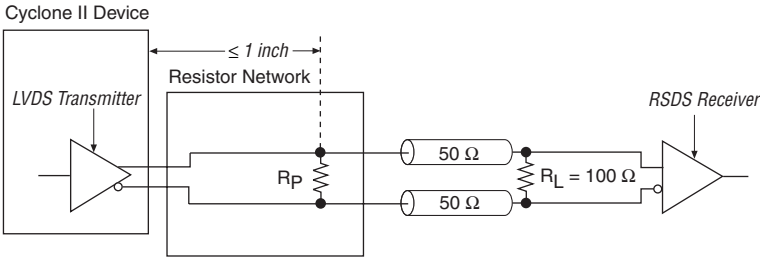
Revision History

Refer to each chapter for its own specific revision history. For information on when each chapter was updated, refer to the Chapter Revision Dates section, which appears in the complete handbook.

Designing with RSDS

Cyclone II devices support the RSDS output standard using the LVDS I/O buffer types. For transmitters, the LVDS output buffer can be used with the external resistor network shown in Figure 11–7.

Figure 11–7. RSDS Resistor Network *Note (1)*



Note to Figure 11–7:
(1) R_S = 120 Ω and R_P = 170 Ω

For more information on the RSDS I/O standard, see the RSDS specification from the National Semiconductor web site **(www.national.com)**.

A resistor network is required to attenuate the LVDS output voltage swing to meet the RSDS specifications. The resistor network values can be modified to reduce power or improve the noise margin. The resistor values chosen should satisfy the following equation:

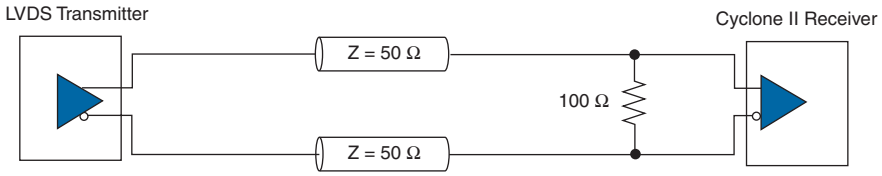
$$\frac{R_S \times \frac{R_P}{2}}{R_S + \frac{R_P}{2}} = 50\,\Omega$$

Additional simulations using the IBIS models should be performed to validate that custom resistor values meet the RSDS requirements.

Single Resistor RSDS Solution

The external single resistor solution reduces the external resistor count while still achieving the required signaling level for RSDS. To transmit the RSDS signal, an external resistor (R_P) is connected in parallel between the two adjacent I/O pins on the board as shown in Figure 11–8. The recommended value of the resistor R_P is 100 Ω

Figure 11–11. LVPECL I/O Interface



Differential SSTL Support in Cyclone II Devices

The differential SSTL I/O standard is a memory bus standard used for applications such as high-speed double data rate (DDR) SDRAM interfaces. The differential SSTL I/O standard is similar to voltage referenced SSTL and requires two differential inputs with an external termination voltage (V_{TT}) of 0.5 × V_{CCIO} to which termination resistors are connected. A 2.5-V output source voltage is required for differential SSTL-2, while a 1.8-V output source voltage is required for differential SSTL-18. The differential SSTL output standard is only supported at PLLCLKOUT pins using two single-ended SSTL output buffers programmed to have opposite polarity.

The differential SSTL input standard is supported at the global clock (GCLK) pins only, treating differential inputs as two single-ended SSTL, and only decoding one of them.

For SSTL signaling characteristics, see the *DC Characteristics & Timing Specification* chapter and the *Selectable I/O Standards in Cyclone II Devices* chapter in Volume 1 of the *Cyclone II Device Handbook*.

Figures 11–12 and 11–13 show the differential SSTL class I and II interfaces, respectively.

PS Configuration

You can use an Altera configuration device, a download cable, or an intelligent host, such as a MAX® II device or microprocessor to configure a Cyclone II device with the PS scheme. In the PS scheme, an external host (configuration device, MAX II device, embedded processor, or host PC) controls configuration. Configuration data is input to the target Cyclone II devices via the DATA0 pin at each rising edge of DCLK.

The Cyclone II decompression feature is fully available when configuring your Cyclone II device using PS mode.

Table 13–6 shows the MSEL pin settings when using the PS configuration scheme.

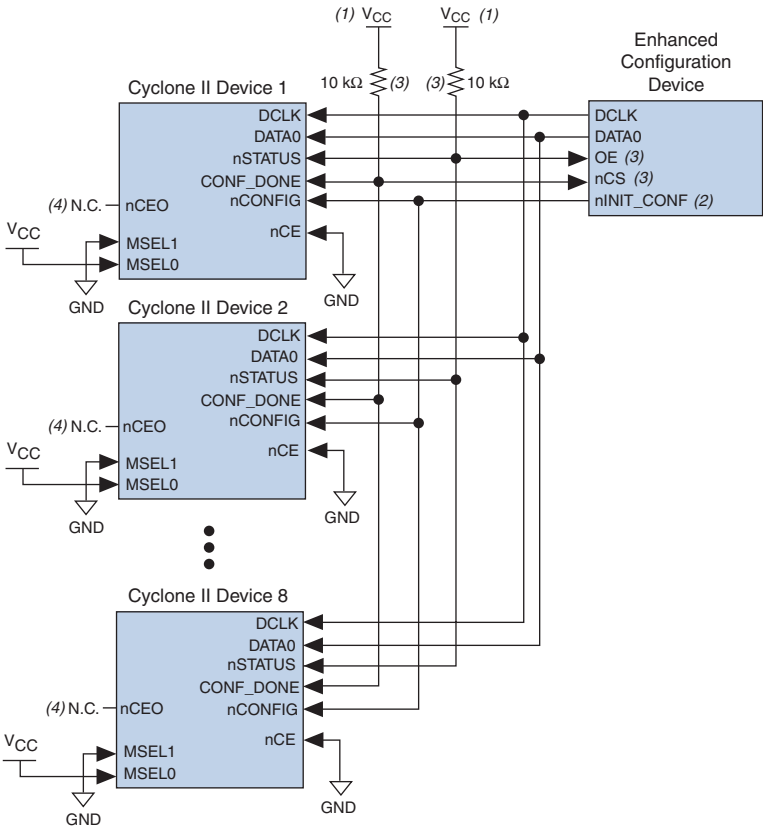
Table 13–6. Cyclone II MSEL Pin Settings for PS Configuration Schemes

Configuration Scheme	MSEL1	MSEL0
PS	0	1

Single Device PS Configuration Using a MAX II Device as an External Host

In the PS configuration scheme, you can use a MAX II device as an intelligent host that controls the transfer of configuration data from a storage device, such as flash memory, to the target Cyclone II device. Configuration data can be stored in RBF, HEX, or TTF format. Figure 13–9 shows the configuration interface connections between the Cyclone II device and a MAX II device for single device configuration.

Figure 13–16. Multiple Device PS Configuration Using an Enhanced Configuration Device When FPGAs Receive the Same Data



Notes to Figure 13–16:

(1) The pull-up resistor should be connected to the same supply voltage as the configuration device.

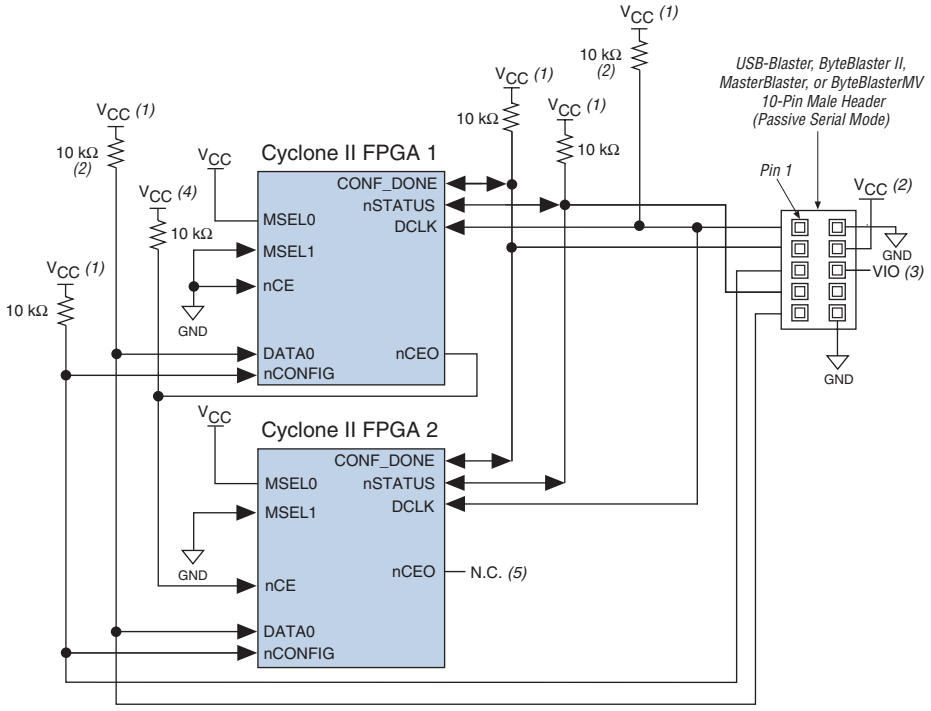
(2) The nINIT_CONF pin is available on enhanced configuration devices and has an internal pull-up resistor that is always active, meaning an external pull-up resistor should not be used on the nINIT_CONF to nCONFIG line. The nINIT_CONF pin does not need to be connected if its functionality is not used. If nINIT_CONF is not used, nCONFIG must be pulled to V_{CC} either directly or through a resistor (if reconfiguration is required, a resistor is necessary).

(3) The enhanced configuration devices' OE and nCS pins have internal programmable pull-up resistors. If internal pull-up resistors are used, external pull-up resistors should not be used on these pins. The internal pull-up resistors are used by default in the Quartus II software. To turn off the internal pull-up resistors, check the **Disable nCS and OE pull-ups on configuration device** option when generating programming files.

(4) The nCEO pin can be left unconnected or used as a user I/O pin when it does not feed other device's nCE pin.

You can cascade several EPC2 or EPC1 devices to configure multiple Cyclone II devices. The first configuration device in the chain is the master configuration device, and the subsequent devices are the slave devices. The master configuration device sends DCLK to the Cyclone II

Figure 13–20. Multiple Device PS Configuration Using a USB-Blaster, MasterBlaster, ByteBlaster II or ByteBlasterMV Cable



Notes to Figure 13–20:

(1) The pull-up resistor should be connected to the same supply voltage as the USB-Blaster, MasterBlaster (VIO pin), ByteBlaster II, or ByteBlasterMV cable.

(2) The pull-up resistors on DATA0 and DCLK are only needed if the download cable is the only configuration scheme used on your board. This is to ensure that DATA0 and DCLK are not left floating after configuration. For example, if you are also using a configuration device, the pull-up resistors on DATA0 and DCLK are not needed.

(3) Pin 6 of the header is a V_{IO} reference voltage for the MasterBlaster output driver. V_{IO} should match the device's V_{CCIO}. Refer to the MasterBlaster Serial/USB Communications Cable Data Sheet for this value. In the ByteBlasterMV, this pin is a no connect. In the USB-Blaster and ByteBlaster II, this pin is connected to nCE when it is used for AS programming, otherwise it is a no connect.

(4) Connect the pull-up resistor to the V_{CCIO} supply voltage of I/O bank that the nCEO pin resides in.

(5) The nCEO pin of the last device in chain can be left unconnected or used as a user I/O pin.

If you are using a download cable to configure Cyclone II devices on a PCB that also has configuration devices, you should electrically isolate the configuration devices from the target Cyclone II devices and cable. One way to isolate the configuration device is to add logic, such as a multiplexer, that can select between the configuration device and the cable. The multiplexer should allow bidirectional transfers on the nSTATUS and CONF_DONE signals. Additionally, you can add switches to

240-Pin Plastic Quad Flat Pack (PQFP)

- All dimensions and tolerances conform to ASME Y14.5M – 1994.
- Controlling dimension is in millimeters.
- Pin 1 may be indicated by an ID dot, or a special feature, in its proximity on package surface.

Tables 15–9 and 15–10 show the package information and package outline figure references, respectively, for the 240-pin PQFP package.

Table 15–9. 240-Pin PQFP Package Information

Description	Specification
Ordering Code Reference	Q
Package Acronym	PQFP
Leadframe Material	Copper
Lead Finish (Plating)	Regular: 85Sn:15Pb (Typ.) Pb-free: Matte Sn
JEDEC Outline Reference	MS-029 Variation: GA
Maximum Lead Coplanarity	0.003 inches (0.08mm)
Weight	7.0 g
Moisture Sensitivity Level	Printed on moisture barrier bag

Table 15–10. 240-Pin PQFP Package Outline Dimensions (Part 1 of 2)

Symbol	Millimeter		
	Min.	Nom.	Max.
A	–	–	4.10
A1	0.25	–	0.50
A2	3.20	3.40	3.60
D	34.60 BSC		
D1	32.00 BSC		
E	34.60 BSC		
E1	32.00 BSC		
L	0.45	0.60	0.75
L1	1.30 REF		
S	0.20	–	–
b	0.17	–	0.27
c	0.09	–	0.20

Figure 15–7 shows a 672-pin FineLine BGA package outline.

Figure 15–7. 672-Pin FineLine BGA Package Outline