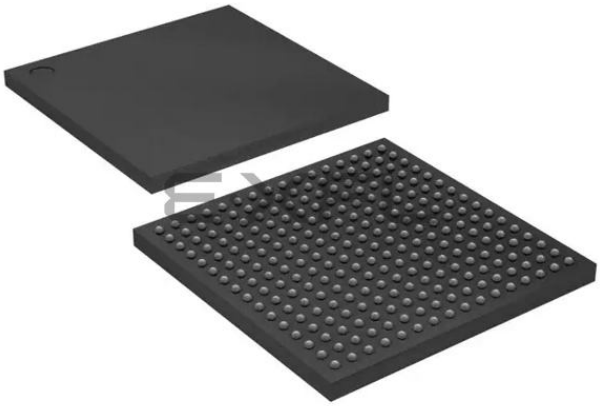

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

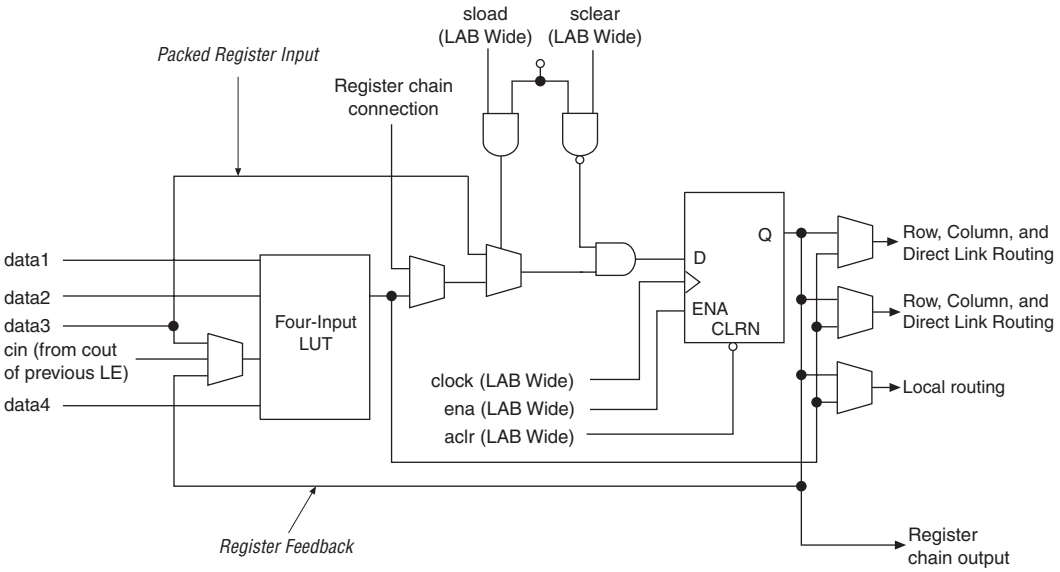
Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	288
Number of Logic Elements/Cells	4608
Total RAM Bits	119808
Number of I/O	158
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep2c5f256i8

Figure 2–3. LE in Normal Mode



Arithmetic Mode

The arithmetic mode is ideal for implementing adders, counters, accumulators, and comparators. An LE in arithmetic mode implements a 2-bit full adder and basic carry chain (see Figure 2–4). LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output. Register feedback and register packing are supported when LEs are used in arithmetic mode.

Table 2–4 describes the PLL features in Cyclone II devices.

Table 2–4. Cyclone II PLL Features

Feature	Description
Clock multiplication and division	*m* / (*n* × post-scale counter) *m* and post-scale counter values (C0 to C2) range from 1 to 32. *n* ranges from 1 to 4.
Phase shift	Cyclone II PLLs have an advanced clock shift capability that enables programmable phase shifts in increments of at least 45°. The finest resolution of phase shifting is determined by the voltage control oscillator (VCO) period divided by 8 (for example, 1/1000 MHz/8 = down to 125-ps increments).
Programmable duty cycle	The programmable duty cycle allows PLLs to generate clock outputs with a variable duty cycle. This feature is supported on each PLL post-scale counter (C0-C2).
Number of internal clock outputs	The Cyclone II PLL has three outputs which can drive the global clock network. One of these outputs (C2) can also drive a dedicated PLL<#>_OUT pin (single ended or differential).
Number of external clock outputs	The C2 output drives a dedicated PLL<#>_OUT pin. If the C2 output is not used to drive an external clock output, it can be used to drive the internal global clock network. The C2 output can concurrently drive the external clock output and internal global clock network.
Manual clock switchover	The Cyclone II PLLs support manual switchover of the reference clock through internal logic. This enables you to switch between two reference input clocks during user mode for applications that may require clock redundancy or support for clocks with two different frequencies.
Gated lock signal	The lock output indicates that there is a stable clock output signal in phase with the reference clock. Cyclone II PLLs include a programmable counter that holds the lock signal low for a user-selected number of input clock transitions, allowing the PLL to lock before enabling the locked signal. Either a gated locked signal or an ungated locked signal from the locked port can drive internal logic or an output pin.
Clock feedback modes	In zero delay buffer mode, the external clock output pin is phase-aligned with the clock input pin for zero delay. In normal mode, the PLL compensates for the internal global clock network delay from the input clock pin to the clock port of the IOE output registers or registers in the logic array. In no compensation mode, the PLL does not compensate for any clock networks.
Control signals	The pllenable signal enables and disables the PLLs. The areset signal resets/resynchronizes the inputs for each PLL. The pfdena signal controls the phase frequency detector (PFD) output with a programmable gate.

Table 5–11 specifies the bus hold parameters for general I/O pins.

Table 5–11. Bus Hold Parameters *Note (1)*

Parameter	Conditions	V_{CCIO} Level						Unit
		1.8 V		2.5 V		3.3 V		
		Min	Max	Min	Max	Min	Max	
Bus-hold low, sustaining current	$V_{IN} > V_{IL}$(maximum)	30	—	50	—	70	—	µA
Bus-hold high, sustaining current	$V_{IN} < V_{IL}$(minimum)	–30	—	–50	—	–70	—	µA
Bus-hold low, overdrive current	$0 V < V_{IN} < V_{CCIO}$	—	200	—	300	—	500	µA
Bus-hold high, overdrive current	$0 V < V_{IN} < V_{CCIO}$	—	–200	—	–300	—	–500	µA
Bus-hold trip point *(2)*	—	0.68	1.07	0.7	1.7	0.8	2.0	V

Notes to Table 5–11:
(1) There is no specification for bus-hold at V_{CCIO} = 1.5 V for the HSTL I/O standard.
(2) The bus-hold trip points are based on calculated input voltages from the JEDEC standard.

On-Chip Termination Specifications

Table 5–12 defines the specifications for internal termination resistance tolerance when using series or differential on-chip termination.

Table 5–12. Series On-Chip Termination Specifications

Symbol	Description	Conditions	Resistance Tolerance			
			Commercial Max	Industrial Max	Extended/ Automotive Temp Max	Unit
25-Ω R_S	Internal series termination without calibration (25-Ω setting)	V_{CCIO} = 3.3V	±30	±30	±40	%
50-Ω R_S	Internal series termination without calibration (50-Ω setting)	V_{CCIO} = 2.5V	±30	±30	±40	%
50-Ω R_S	Internal series termination without calibration (50-Ω setting)	V_{CCIO} = 1.8V	±30 *(1)*	±40	±50	%

Note to Table 5–12:
(1) For commercial –8 devices, the tolerance is ±40%.

Table 5–18. DSP Block Internal Timing Microparameters (Part 2 of 2)

Parameter	–6 Speed Grade (1)		–7 Speed Grade (2)		–8 Speed Grade (3)		Unit
	Min	Max	Min	Max	Min	Max	
TPIPE2OUTREG	47	104	45	142	45	185	ps
	—	—	47	—	47	—	ps
TPD9	529	2470	505	3353	505	4370	ps
	—	—	529	—	529	—	ps
TPD18	425	2903	406	3941	406	5136	ps
	—	—	425	—	425	—	ps
TCLR	2686	—	3572	—	3572	—	ps
	—	—	3129	—	3572	—	ps
TCLKL	1923	—	2769	—	2769	—	ps
	—	—	2307	—	2769	—	ps
TCLKH	1923	—	2769	—	2769	—	ps
	—	—	2307	—	2769	—	ps

Notes to Table 5–18:

(1) For the –6 speed grades, the minimum timing is for the commercial temperature grade. The –7 speed grade devices offer the automotive temperature grade. The –8 speed grade devices offer the industrial temperature grade.

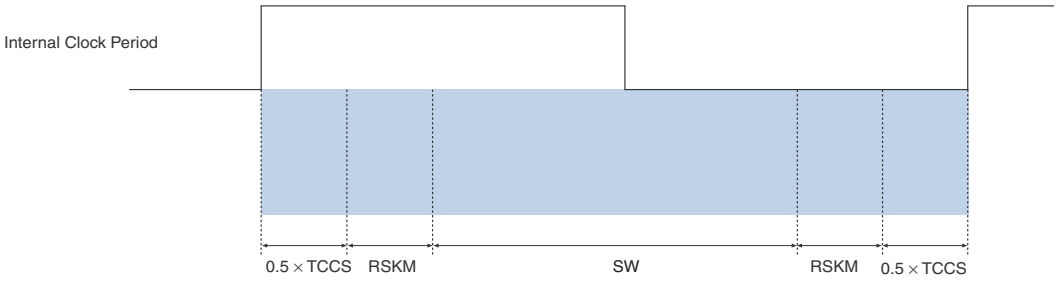
(2) For each parameter of the –7 speed grade columns, the value in the first row represents the minimum timing parameter for automotive devices. The second row represents the minimum timing parameter for commercial devices.

(3) For each parameter of the –8 speed grade columns, the value in the first row represents the minimum timing parameter for industrial devices. The second row represents the minimum timing parameter for commercial devices.

Table 5–19. M4K Block Internal Timing Microparameters (Part 1 of 3)

Parameter	–6 Speed Grade (1)		–7 Speed Grade (2)		–8 Speed Grade (3)		Unit
	Min	Max	Min	Max	Min	Max	
TM4KRC	2387	3764	2275	4248	2275	4736	ps
	—	—	2387	—	2387	—	ps
TM4KWERESU	35	—	46	—	46	—	ps
	—	—	40	—	46	—	ps
TM4KWEREH	234	—	267	—	267	—	ps
	—	—	250	—	267	—	ps
TM4KBESU	35	—	46	—	46	—	ps
	—	—	40	—	46	—	ps

Figure 5–4. High-Speed I/O Timing Budget *Note (1)*



Note to Figure 5–4:
(1) The equation for the high-speed I/O timing budget is: period = TCCS + RSKM + SW + RSKM.

Table 5–48 shows the RSDS timing budget for Cyclone II devices at 311 Mbps. RSDS is supported for transmitting from Cyclone II devices. Cyclone II devices cannot receive RSDS data because the devices are intended for applications where they will be driving display drivers. Cyclone II devices support a maximum RSDS data rate of 311 Mbps using DDIO registers. Cyclone II devices support RSDS only in the commercial temperature range.

Table 5–48. RSDS Transmitter Timing Specification (Part 1 of 2)

Symbol	Conditions	–6 Speed Grade			–7 Speed Grade			–8 Speed Grade			Unit
		Min	Typ	Max (1)	Min	Typ	Max (1)	Min	Typ	Max (1)	
f_{HSCLK} (input clock frequency)	×10	10	—	155.5	10	—	155.5	10	—	155.5	MHz
	×8	10	—	155.5	10	—	155.5	10	—	155.5	MHz
	×7	10	—	155.5	10	—	155.5	10	—	155.5	MHz
	×4	10	—	155.5	10	—	155.5	10	—	155.5	MHz
	×2	10	—	155.5	10	—	155.5	10	—	155.5	MHz
	×1	10	—	311	10	—	311	10	—	311	MHz
Device operation in Mbps	×10	100	—	311	100	—	311	100	—	311	Mbps
	×8	80	—	311	80	—	311	80	—	311	Mbps
	×7	70	—	311	70	—	311	70	—	311	Mbps
	×4	40	—	311	40	—	311	40	—	311	Mbps
	×2	20	—	311	20	—	311	20	—	311	Mbps
	×1	10	—	311	10	—	311	10	—	311	Mbps
t_{DUTY}	—	45	—	55	45	—	55	45	—	55	%

Table 5–56. Maximum DCD for SDR Output on Column I/O Notes (1), (2) *(Part 2 of 2)*

Column I/O Output Standard	C6	C7	C8	Unit
2.5-V	140	140	155	ps
1.8-V	115	115	165	ps
1.5-V	745	745	770	ps
SSTL-2 Class I	60	60	75	ps
SSTL-2 Class II	60	60	80	ps
SSTL-18 Class I	60	130	130	ps
SSTL-18 Class II	60	135	135	ps
HSTL-18 Class I	60	115	115	ps
HSTL-18 Class II	75	75	100	ps
HSTL-15 Class I	150	150	150	ps
HSTL-15 Class II	135	135	155	ps
Differential SSTL-2 Class I	60	60	75	ps
Differential SSTL-2 Class II	60	60	80	ps
Differential SSTL-18 Class I	60	130	130	ps
Differential SSTL-18 Class II	60	135	135	ps
Differential HSTL-18 Class I	60	115	115	ps
Differential HSTL-18 Class II	75	75	100	ps
Differential HSTL-15 Class I	150	150	150	ps
Differential HSTL-15 Class II	135	135	155	ps
LVDS	60	60	60	ps
Simple RSDS	60	70	70	ps
Mini-LVDS	60	60	60	ps

Notes to Table 5–56:

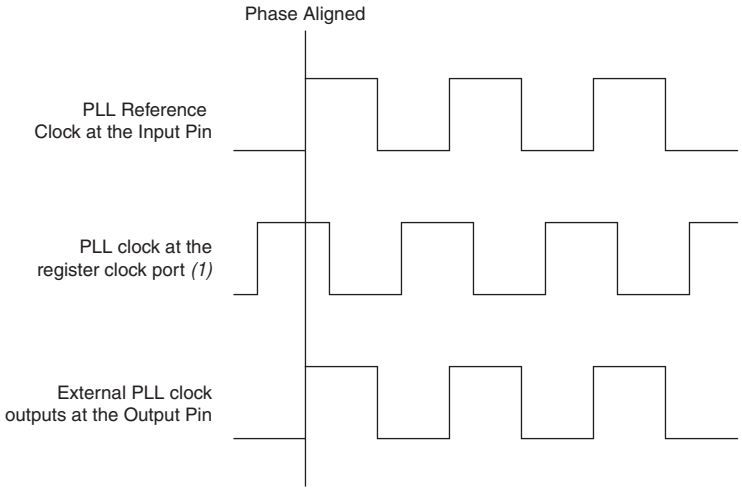
(1) The DCD specification is characterized using the maximum drive strength available for each I/O standard.

(2) Numbers are applicable for commercial, industrial, and automotive devices.

Table 5–57. Maximum for DDIO Output on Row Pins with PLL in the Clock Path Notes (1), (2) *(Part 1 of 2)*

Row Pins with PLL in the Clock Path	C6	C7	C8	Unit
LVCMOS	270	310	310	ps
LVTTL	285	305	335	ps
2.5-V	180	180	220	ps
1.8-V	165	175	205	ps

Figure 7–5. Phase Relationship between Cyclone II PLL Clocks in Zero Delay Buffer Mode



Note to Figure 7–5:
(1) The internal clock output(s) can lead or lag the external PLL clock output (PLL<#>_OUT) signals.

Altera recommends using the same I/O standard on the input and output clocks when using the Cyclone II PLL in zero delay buffer mode.

No Compensation Mode

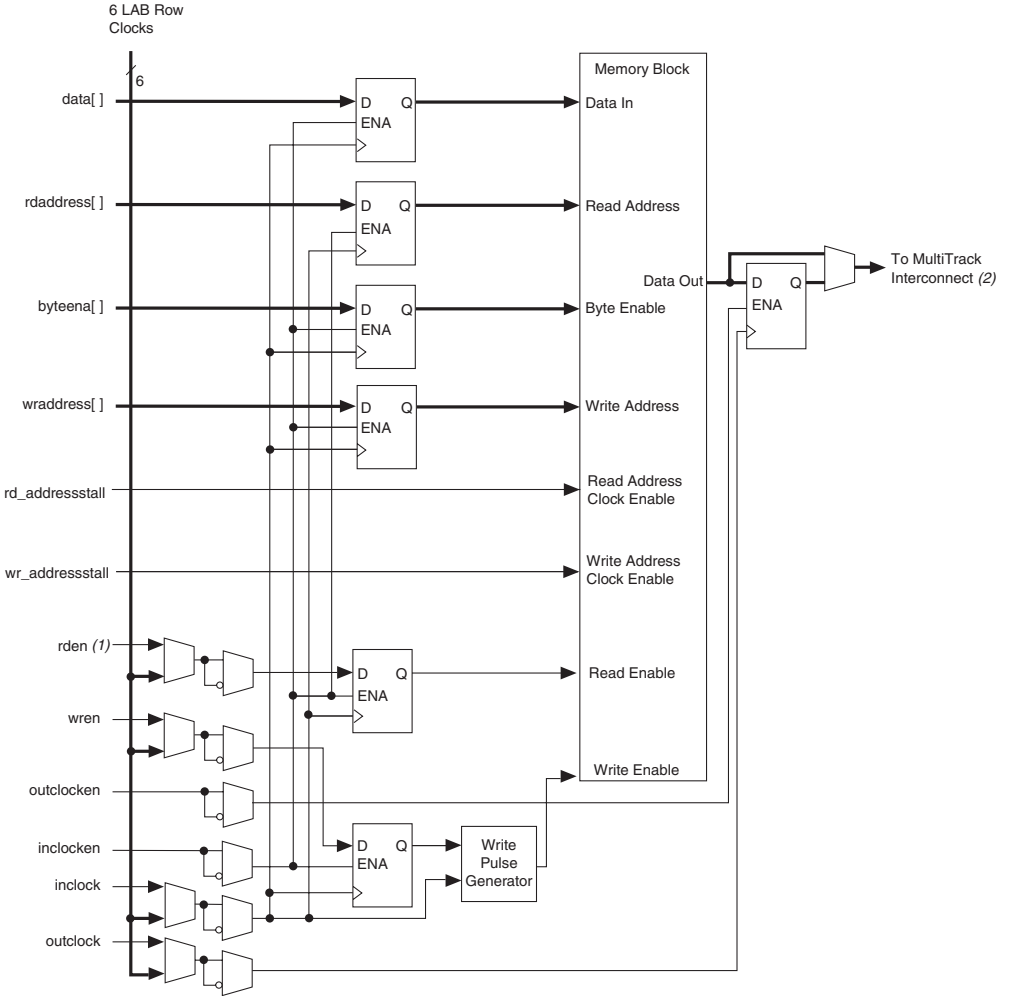
In no compensation mode, the PLL does not compensate for any clock networks, which leads to better jitter performance. Because the clock feedback into the PFD does not pass through as much circuitry, both the PLL internal clock outputs and external clock outputs are phase shifted with respect to the PLL clock input. Figure 7–6 shows an example waveform of the PLL clocks' phase relationship in this mode.

Table 7–8 shows the clock sources connectivity to the global clock networks.

Table 7–8. Global Clock Network Connections (Part 1 of 3)

Global Clock Network Clock Sources	Global Clock Networks															
	All Cyclone II Devices								EP2C15 through EP2C70 Devices Only							
	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
CLK0/LVDSCLK0p	✓		✓													
CLK1/LVDSCLK0n		✓	✓													
CLK2/LVDSCLK1p	✓			✓												
CLK3/LVDSCLK1n		✓		✓												
CLK4/LVDSCLK2p					✓		✓									
CLK5/LVDSCLK2n						✓	✓									
CLK6/LVDSCLK3p					✓			✓								
CLK7/LVDSCLK3n						✓		✓								
CLK8/LVDSCLK4n									✓		✓					
CLK9/LVDSCLK4p										✓	✓					
CLK10/LVDSCLK5n									✓			✓				
CLK11/LVDSCLK5p										✓		✓				
CLK12/LVDSCLK6n													✓		✓	
CLK13/LVDSCLK6p														✓	✓	
CLK14/LVDSCLK7n													✓			✓
CLK15/LVDSCLK7p														✓		✓
PLL1_c0	✓	✓		✓												
PLL1_c1	✓		✓	✓												
PLL1_c2		✓	✓													
PLL2_c0					✓	✓		✓								
PLL2_c1					✓		✓	✓								
PLL2_c2						✓	✓									
PLL3_c0									✓	✓		✓				
PLL3_c1									✓		✓	✓				
PLL3_c2										✓	✓					

Figure 8–15. Cyclone II Input/Output Clock Mode in Simple Dual-Port Mode *Notes (1), (2)*



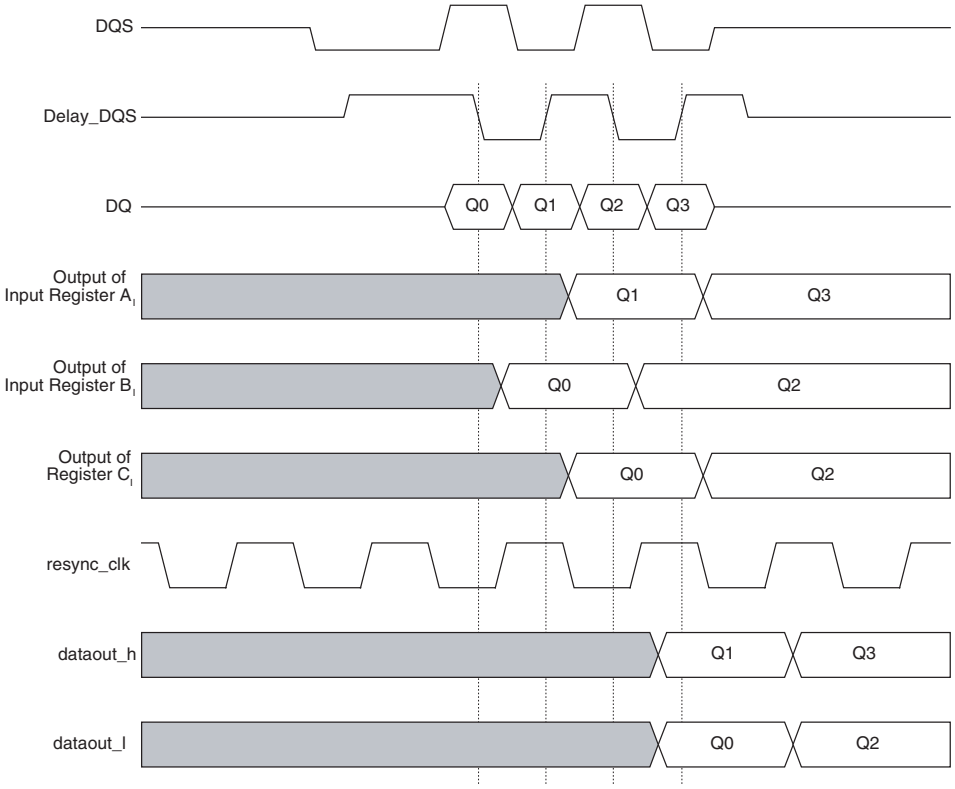
Notes to Figure 8–15:

(1) Violating the setup or hold time on the memory block address registers could corrupt memory contents. This applies to both read and write operations.

(2) See the *Cyclone II Device Family Data Sheet* in volume 1 of the *Cyclone II Device Handbook* for more information on the MultiTrack™ interconnect.

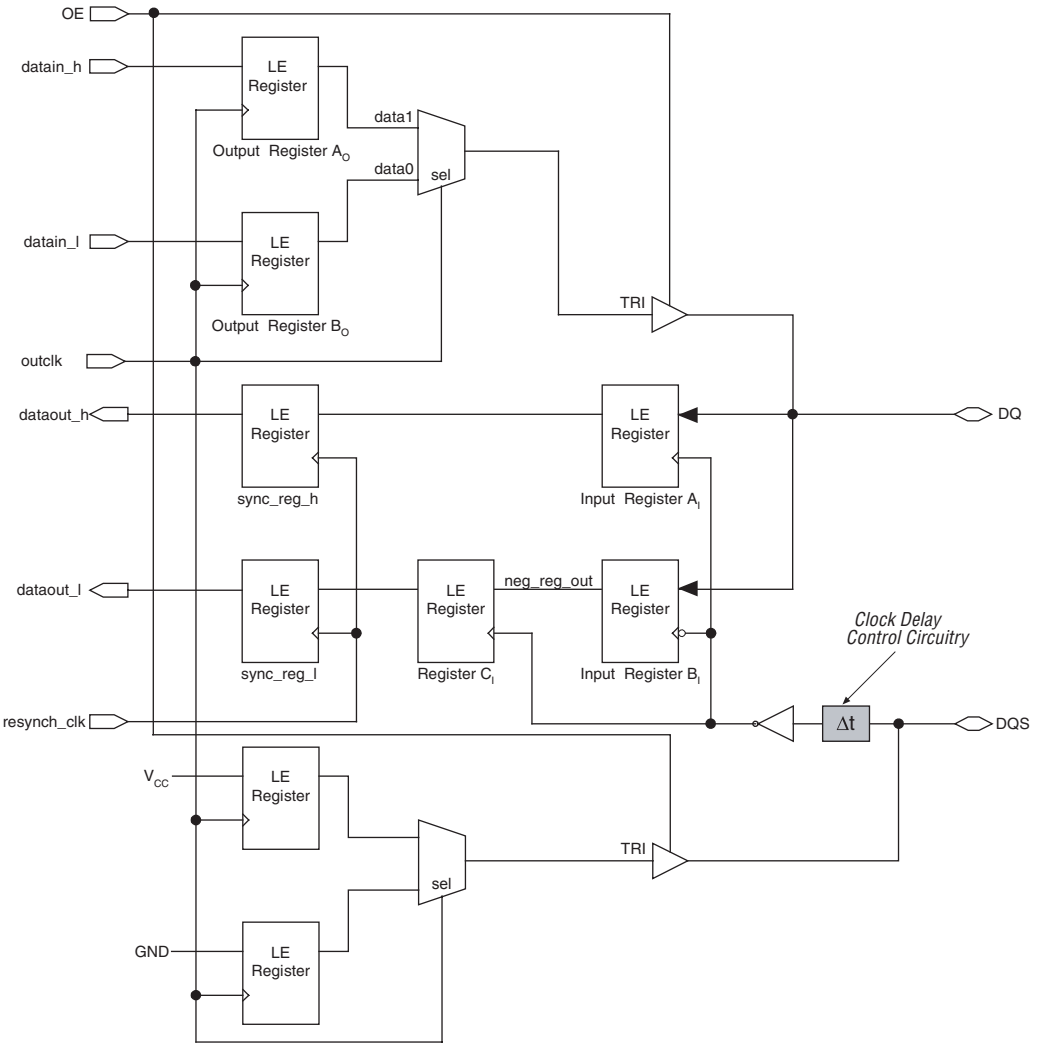
Registers sync_reg_h and sync_reg_l synchronize the two data streams to the rising edge of the resynchronization clock. Figure 9–12 shows examples of functional waveforms from a double data rate input implementation.

Figure 9–12. DDR Input Functional Waveforms



The Cyclone II DDR input registers require you to invert the incoming DQS signal to ensure proper data transfer. The altdq megafunction automatically adds the inverter on the clock port of the DQ signals. As shown in Figure 9–11, the inverted DQS signal's rising edge clocks register A_I, its falling edge clocks register B_I, and register C_I aligns the data clocked by register B_I with register A_I on the inverted DQS signal's rising edge. In a DDR memory read operation, the last data coincides with the falling edge of DQS signal. If you do not invert the DQS pin, you do not get this last data because the register does not latch until the next rising edge of the DQS signal.

Figure 9–16. Bidirectional DDR Implementation for DDR Memory Interfaces *Note (1)*



Note to Figure 9–16:

(1) You can use the `altdq` and `altdqs` megafunctions to generate the DQ and DQS signals.

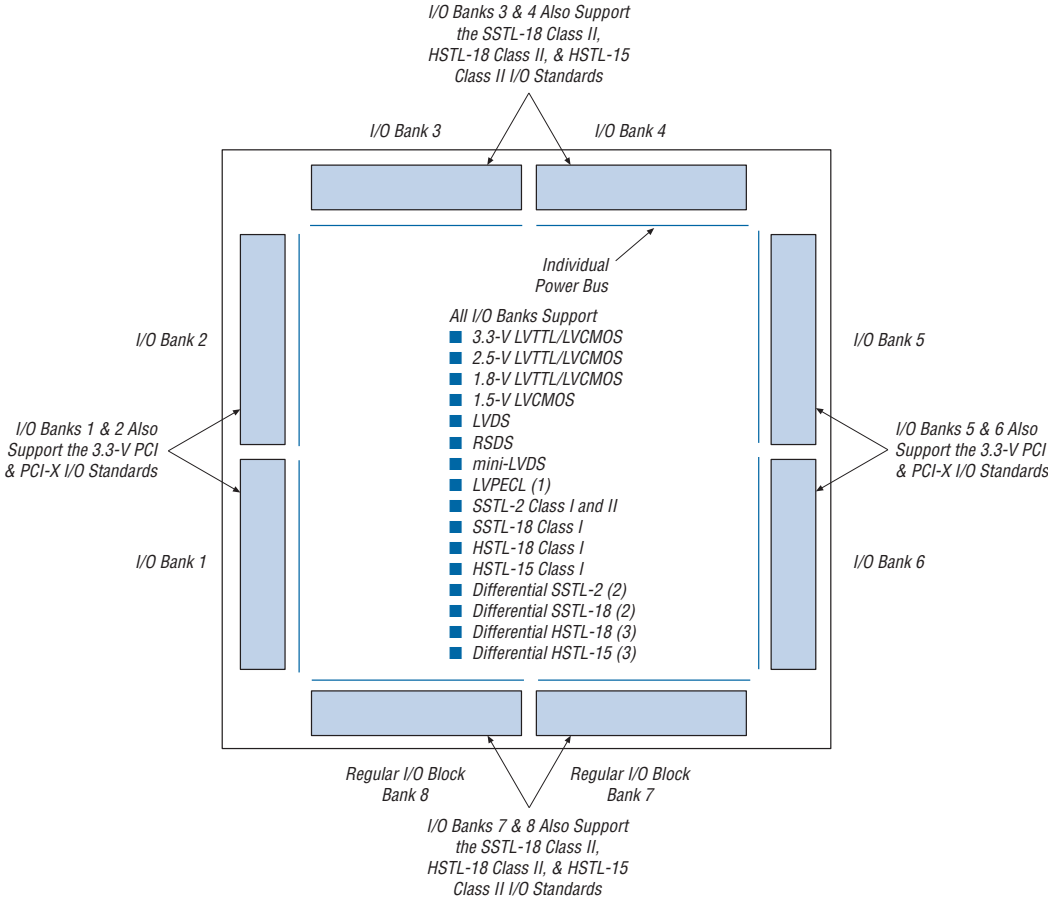
Figure 9–17 shows example waveforms from a bidirectional DDR implementation.

Table 10–6. Programmable Drive Strength (Part 2 of 2)

I/O Standard	I_{OH}/I_{OL} Current Strength Setting (mA)	
	Top and Bottom I/O Pins	Side I/O Pins
SSTL-2 class I	8	8
	12	12
SSTL-2 class II	16	16
	20	—
	24	—
SSTL-18 class I	6	6
	8	8
	10	10
	12	—
SSTL-18 class II	16	—
	18	—
HSTL-18 class I	8	8
	10	10
	12	12
HSTL-18 class II	16	N/A
	18	—
	20	—
HSTL-15 class I	8	8
	10	—
	12	—
HSTL-15 class II	16	N/A

These drive-strength settings are programmable on a per-pin basis using the Quartus II software.

Figure 11–2. I/O Banks in EP2C15, EP2C20, EP2C35, EP2C50 & EP2C70 Devices



Notes to Figure 11–2:

(1) The LVPECL I/O standard is only supported on clock input pins. This I/O standard is not supported on output pins.

(2) The differential SSTL-18 and SSTL-2 I/O standards are only supported on clock input pins and PLL output clock pins.

(3) The differential 1.8-V and 1.5-V HSTL I/O standards are only supported on clock input pins and PLL output clock pins.

Cyclone II High-Speed I/O Interface

Cyclone II devices provide a multi-protocol interface that allows communication between a variety of I/O standards, including LVDS, LVPECL, RSDS, mini-LVDS, differential HSTL, and differential SSTL. This feature makes the Cyclone II device family ideal for applications that require multiple I/O standards, such as protocol translation.

Figure 11–12. Differential SSTL Class I Interface

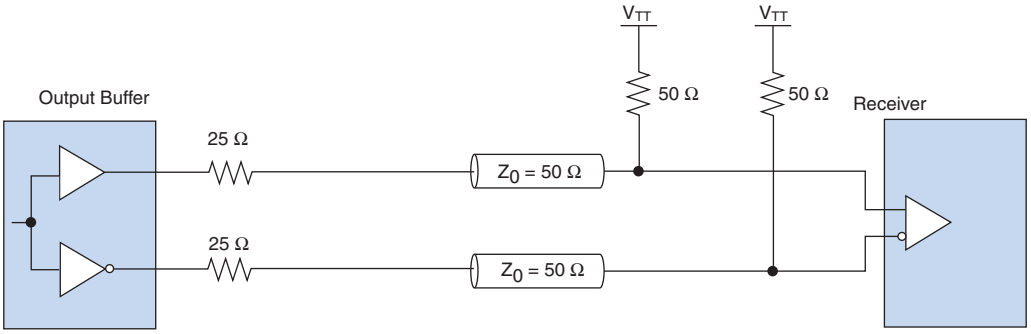
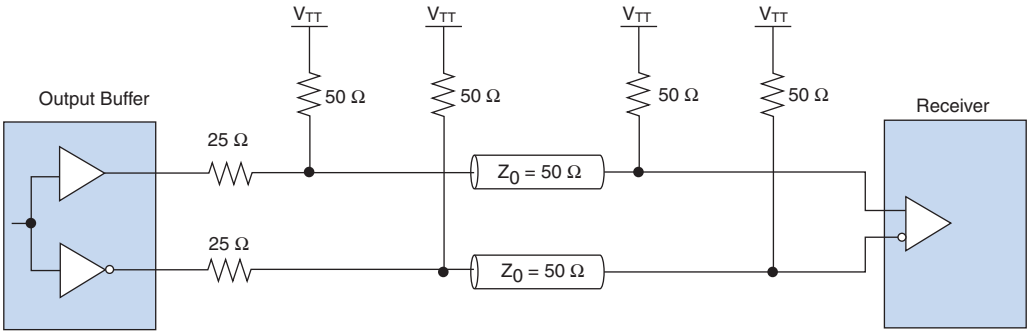


Figure 11–13. Differential SSTL Class II Interface



Differential HSTL Support in Cyclone II Devices

The differential HSTL AC and DC specifications are the same as the HSTL single-ended specifications. The differential HSTL I/O standard is available on the GCLK pins only, treating differential inputs as two single-ended HSTL, and only decoding one of them. The differential HSTL output I/O standard is only supported at the PLLCLKOUT pins using two single-ended HSTL output buffers with the second output programmed as inverted. The standard requires two differential inputs with an external termination voltage (V_{TT}) of 0.5 × V_{CCIO} to which termination resistors are connected.

For the HSTL signaling characteristics, see the *DC Characteristics & Timing Specifications* chapter and the *Selectable I/O Standards in Cyclone II Devices* chapter in Volume 1 of the *Cyclone II Device Handbook.*

When the `signa` and `signb` signals are unused, the Quartus® II software sets the multiplier to perform unsigned multiplication by default.

Output Registers

You can choose to register the embedded multiplier output using the output registers in 18- or 36-bit sections depending on the operational mode of the multiplier. The following control signals are available to each output register within the embedded multiplier:

- clock
- clock enable
- asynchronous clear

All input and output registers within a single embedded multiplier are fed by the same clock, clock enable, or asynchronous clear signal.

See the *Cyclone II Architecture* chapter in Volume 1 of the ***Cyclone II Device Handbook*** for more information on the embedded multiplier routing and interface.

Operational Modes

The embedded multiplier can be used in one of two operational modes, depending on the application needs:

- One 18-bit multiplier
- Up to two 9-bit independent multipliers

The Quartus II software includes megafunctions used to control the mode of operation of the multipliers. After you have made the appropriate parameter settings using the megafunction's MegaWizard® Plug-In Manager, the Quartus II software automatically configures the embedded multiplier.

The Cyclone II embedded multipliers can also be used to implement multiplier adder and multiplier accumulator functions where the multiplier portion of the function is implemented using embedded multipliers and the adder or accumulator function is implemented in logic elements (LEs).

For more information on megafunction and Quartus II support for Cyclone II embedded multipliers, see the "Software Support" section.

The **Enable user-supplied start-up clock (CLKUSR)** option can be turned on in the Quartus II software from the **General** tab of the **Device & Pin Options** dialog box. Supplying a clock on CLKUSR does not affect the configuration process. After all configuration data has been accepted and CONF_DONE goes high, Cyclone II devices require 299 clock cycles to initialize properly and support a CLKUSR f_{MAX} of 100 MHz.

If the optional CLKUSR pin is being used and nCONFIG is pulled low to restart configuration during device initialization, you need to ensure that CLKUSR continues toggling during the time nSTATUS is low (maximum of 40 µs).

An optional INIT_DONE pin signals the end of initialization and the start of user mode with a low-to-high transition. By default, the INIT_DONE output is disabled. You can enable the INIT_DONE output by turning on the **Enable INIT_DONE output** option in the Quartus II software. If you use the INIT_DONE pin, an external 10-kΩ pull-up resistor pulls the pin high when nCONFIG is low and during the beginning of configuration. Once the optional bit to enable INIT_DONE is programmed into the device (during the first frame of configuration data), the INIT_DONE pin transitions low. When initialization is complete, the INIT_DONE pin is released and pulled high. The MAX II device must be able to detect this low-to-high transition, which signals the FPGA has entered user mode.

If you want to use the INIT_DONE pin as a user I/O pin, you should wait for the maximum value of t_{CD2UM} (see Table 13–7) after the CONF_DONE signal transitions high so to ensure the Cyclone II device has been initialized properly and is in user mode.

Make sure the MAX II device does not drive the CONF_DONE signal low during configuration, initialization, and before the device enters user mode.

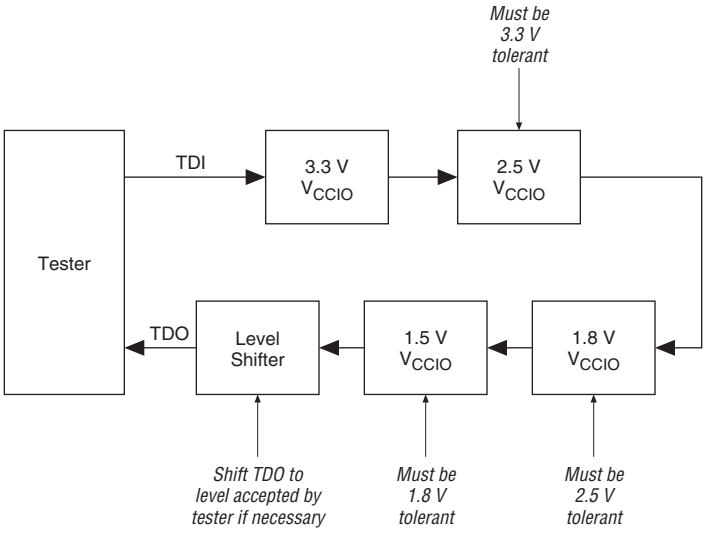
User Mode

When initialization is complete, the Cyclone II device enters user mode. In user mode, the user I/O pins no longer have pull-up resistors and function as assigned in your design.

To ensure DCLK and DATA0 are not left floating at the end of configuration, the MAX II device must drive them either high or low, which ever is convenient on your PCB. The Cyclone II device DATA0 pin is not available as a user I/O pin after configuration.

When the FPGA is in user mode, you can initiate a reconfiguration by transitioning the nCONFIG pin low-to-high. The nCONFIG pin must be low for at least 2 µs. When the nCONFIG transitions low, the Cyclone II

Figure 14–13. JTAG Chain of Mixed Voltages



Using IEEE Std. 1149.1 BST Circuitry

Cyclone II devices have dedicated JTAG pins, and the IEEE Std. 1149.1 BST circuitry is enabled upon device power-up. You can perform BST on Cyclone II FPGAs not only before and after configuration, but also during configuration. Cyclone II FPGAs support the BYPASS, IDCODE, and SAMPLE instructions during configuration without interrupting configuration. To send all other JTAG instructions, you must interrupt configuration using the CONFIG_IO instruction.

The CONFIG_IO instruction allows you to configure I/O buffers via the JTAG port, and when issued, interrupts configuration. This instruction allows you to perform board-level testing prior to configuring the Cyclone II FPGA or waiting for a configuration device to complete configuration. Once configuration has been interrupted and JTAG BST is complete, the part must be reconfigured via JTAG (PULSE_CONFIG instruction) or by pulsing nCONFIG low.

When you perform JTAG boundary-scan testing before configuration, the nCONFIG pin must be held low.

The device-wide reset (DEV_CLRn) and device-wide output enable (DEV_OE) pins on Cyclone II devices do not affect JTAG boundary-scan or configuration operations. Toggling these pins does not disrupt BST operation any more than usual.

Maunder, C. M., and R. E. Tulloss. *The Test Access Port and Boundary-Scan Architecture*. Los Alamitos: IEEE Computer Society Press, 1990.

Document Revision History

Table 14–4 shows the revision history for this document.

Table 14–4. Document Revision History

Date & Document Version	Changes Made	Summary of Changes
February 2007 v2.1	• Added document revision history. • Added new section "BST for Configured Devices".	• Added infomation about 'Always Enable Input Buffer' option.
July 2005 v2.0	Moved the "JTAG Timing Specifications" section to the *DC Characteristics & Timing Specifications* chapter.	
June 2004 v1.0	Added document to the Cyclone II Device Handbook.	

240-Pin Plastic Quad Flat Pack (PQFP)

- All dimensions and tolerances conform to ASME Y14.5M – 1994.
- Controlling dimension is in millimeters.
- Pin 1 may be indicated by an ID dot, or a special feature, in its proximity on package surface.

Tables 15–9 and 15–10 show the package information and package outline figure references, respectively, for the 240-pin PQFP package.

Table 15–9. 240-Pin PQFP Package Information

Description	Specification
Ordering Code Reference	Q
Package Acronym	PQFP
Leadframe Material	Copper
Lead Finish (Plating)	Regular: 85Sn:15Pb (Typ.) Pb-free: Matte Sn
JEDEC Outline Reference	MS-029 Variation: GA
Maximum Lead Coplanarity	0.003 inches (0.08mm)
Weight	7.0 g
Moisture Sensitivity Level	Printed on moisture barrier bag

Table 15–10. 240-Pin PQFP Package Outline Dimensions (Part 1 of 2)

Symbol	Millimeter		
	Min.	Nom.	Max.
A	–	–	4.10
A1	0.25	–	0.50
A2	3.20	3.40	3.60
D	34.60 BSC		
D1	32.00 BSC		
E	34.60 BSC		
E1	32.00 BSC		
L	0.45	0.60	0.75
L1	1.30 REF		
S	0.20	–	–
b	0.17	–	0.27
c	0.09	–	0.20