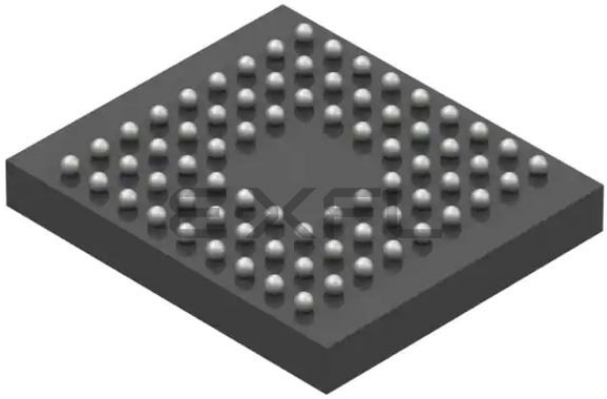

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	ARM® Cortex®-M4
Core Size	32-Bit Single-Core
Speed	80MHz
Connectivity	CANbus, I²C, IrDA, LINbus, MMC/SD, QSPI, SAI, SPI, SWPMI, UART/USART, USB OTG
Peripherals	Brown-out Detect/Reset, DMA, LCD, PWM, WDT
Number of I/O	57
Program Memory Size	512KB (512K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128K x 8
Voltage - Supply (Vcc/Vdd)	1.71V ~ 3.6V
Data Converters	A/D 16x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	72-UFBGA, WLCSP
Supplier Device Package	72-WLCSP (4.41x3.76)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/stm32l476jey6vtr

1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32L476xx microcontrollers.

This document should be read in conjunction with the STM32L4x6 reference manual (RM0351). The reference manual is available from the STMicroelectronics website *www.st.com*.

For information on the ARM® Cortex®-M4 core, please refer to the Cortex®-M4 Technical Reference Manual, available from the www.arm.com website.

3.7 Boot modes

At startup, BOOT0 pin and BOOT1 option bit are used to select one of three boot options:

- Boot from user Flash
- Boot from system memory
- Boot from embedded SRAM

The boot loader is located in system memory. It is used to reprogram the Flash memory by using USART, I2C, SPI, CAN and USB OTG FS in Device mode through DFU (device firmware upgrade).

3.8 Cyclic redundancy check calculation unit (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a configurable generator polynomial value and size.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

3.9 Power supply management

3.9.1 Power supply schemes

- V_{DD} = 1.71 to 3.6 V: external power supply for I/Os (V_{DDIO1}), the internal regulator and the system analog such as reset, power management and internal clocks. It is provided externally through V_{DD} pins.
- V_{DDA} = 1.62 V (ADCs/COMPs) / 1.8 (DACs/OPAMPs) to 3.6 V: external analog power supply for ADCs, DACs, OPAMPs, Comparators and Voltage reference buffer. The V_{DDA} voltage level is independent from the V_{DD} voltage.
- V_{DDUSB} = 3.0 to 3.6 V: external independent power supply for USB transceivers. The V_{DDUSB} voltage level is independent from the V_{DD} voltage.
- V_{DDIO2} = 1.08 to 3.6 V: external power supply for 14 I/Os (PG[15:2]). The V_{DDIO2} voltage level is independent from the V_{DD} voltage.
- V_{LCD} = 2.5 to 3.6 V: the LCD controller can be powered either externally through VLCD pin, or internally from an internal voltage generated by the embedded step-up converter.
- V_{BAT} = 1.55 to 3.6 V: power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V_{DD} is not present.

Note: When the functions supplied by V_{DDA}, V_{DDUSB} or V_{DDIO2} are not used, these supplies should preferably be shorted to V_{DD}.

Note: If these supplies are tied to ground, the I/Os supplied by these power supplies are not 5 V tolerant (refer to Table 19: Voltage characteristics).

Note: V_{DDIOx} is the I/Os general purpose digital functions supply. V_{DDIOx} represents V_{DDIO1} or V_{DDIO2}, with V_{DDIO1} = V_{DD}. V_{DDIO2} supply voltage level is independent from V_{DDIO1}.

3.11 Clocks and startup

The clock controller (see *Figure 3*) distributes the clocks coming from different oscillators to the core and the peripherals. It also manages clock gating for low-power modes and ensures clock robustness. It features:

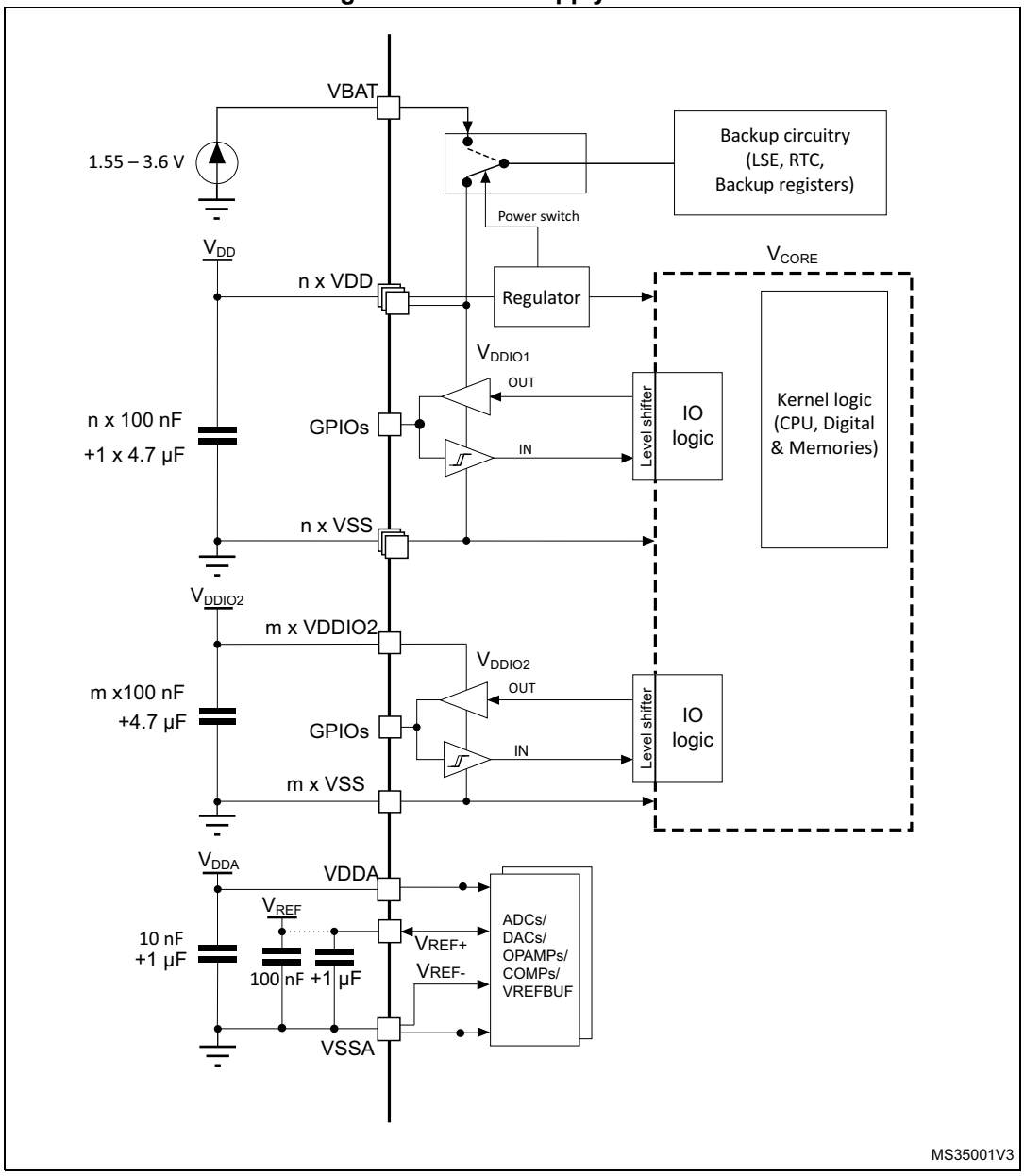
- **Clock prescaler:** to get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler
- **Safe clock switching:** clock sources can be changed safely on the fly in run mode through a configuration register.
- **Clock management:** to reduce power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.
- **System clock source:** four different clock sources can be used to drive the master clock SYSCLK:
  - 4-48 MHz high-speed external crystal or ceramic resonator (HSE), that can supply a PLL. The HSE can also be configured in bypass mode for an external clock.
  - 16 MHz high-speed internal RC oscillator (HSI16), trimmable by software, that can supply a PLL
  - Multispeed internal RC oscillator (MSI), trimmable by software, able to generate 12 frequencies from 100 kHz to 48 MHz. When a 32.768 kHz clock source is available in the system (LSE), the MSI frequency can be automatically trimmed by hardware to reach better than ±0.25% accuracy. In this mode the MSI can feed the USB device, saving the need of an external high-speed crystal (HSE). The MSI can supply a PLL.
  - System PLL which can be fed by HSE, HSI16 or MSI, with a maximum frequency at 80 MHz.
- **Auxiliary clock source:** two ultralow-power clock sources that can be used to drive the LCD controller and the real-time clock:
  - 32.768 kHz low-speed external crystal (LSE), supporting four drive capability modes. The LSE can also be configured in bypass mode for an external clock.
  - 32 kHz low-speed internal RC (LSI), also used to drive the independent watchdog. The LSI clock accuracy is ±5% accuracy.
- **Peripheral clock sources:** Several peripherals (USB, SDMMC, RNG, SAI, USARTs, I2Cs, LPTimers, ADC, SWPMI) have their own independent clock whatever the system clock. Three PLLs, each having three independent outputs allowing the highest flexibility, can generate independent clocks for the ADC, the USB/SDMMC/RNG and the two SAIs.
- **Startup clock:** after reset, the microcontroller restarts by default with an internal 4 MHz clock (MSI). The prescaler ratio and clock source can be changed by the application program as soon as the code execution starts.
- **Clock security system (CSS):** this feature can be enabled by software. If a HSE clock failure occurs, the master clock is automatically switched to HSI16 and a software

Table 16. Alternate function AF0 to AF7 (for AF8 to AF15 see *Table 17*) (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
		SYS_AF	TIM1/TIM2/ TIM5/TIM8/ LPTIM1	TIM1/TIM2/ TIM3/TIM4/ TIM5	TIM8	I2C1/I2C2/I2C3	SPI1/SPI2	SPI3/DFSDM	USART1/ USART2/ USART3
Port D	PD0	-	-	-	-	-	SPI2_NSS	DFSDM_DATIN7	-
	PD1	-	-	-	-	-	SPI2_SCK	DFSDM_CKIN7	-
	PD2	-	-	TIM3_ETR	-	-	-	-	USART3_RTS_ DE
	PD3	-	-	-	-	-	SPI2_MISO	DFSDM_DATIN0	USART2_CTS
	PD4	-	-	-	-	-	SPI2_MOSI	DFSDM_CKIN0	USART2_RTS_ DE
	PD5	-	-	-	-	-	-	-	USART2_TX
	PD6	-	-	-	-	-	-	DFSDM_DATIN1	USART2_RX
	PD7	-	-	-	-	-	-	DFSDM_CKIN1	USART2_CK
	PD8	-	-	-	-	-	-	-	USART3_TX
	PD9	-	-	-	-	-	-	-	USART3_RX
	PD10	-	-	-	-	-	-	-	USART3_CK
	PD11	-	-	-	-	-	-	-	USART3_CTS
	PD12	-	-	TIM4_CH1	-	-	-	-	USART3_RTS_ DE
	PD13	-	-	TIM4_CH2	-	-	-	-	-
	PD14	-	-	TIM4_CH3	-	-	-	-	-
	PD15	-	-	TIM4_CH4	-	-	-	-	-

6.1.6 Power supply scheme

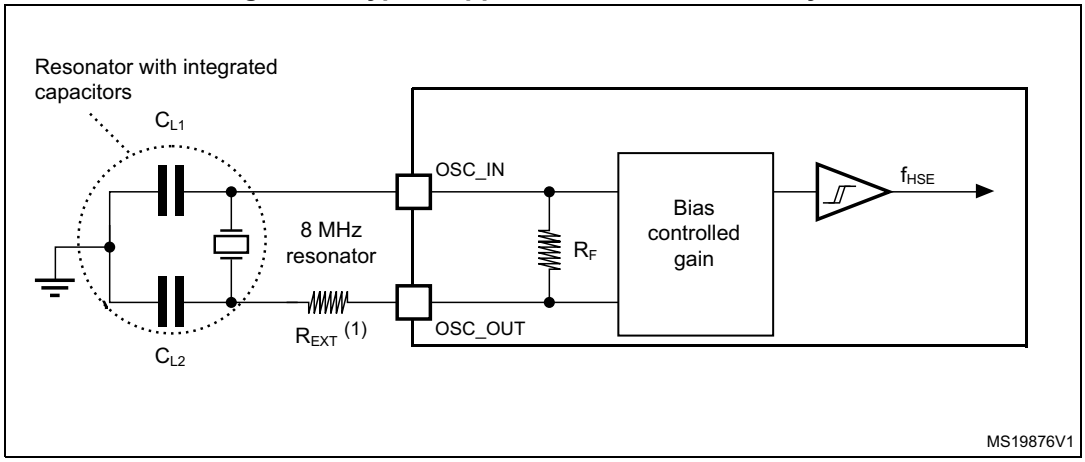
Figure 13. Power supply scheme



Caution: Each power supply pair (V_{DD}/V_{SS}, V_{DDA}/V_{SSA} etc.) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure the good functionality of the device.

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Figure 18. Typical application with an 8 MHz crystal



1. R_{EXT} value depends on the crystal characteristics.

Low-speed external clock generated from a crystal resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 46*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 46. LSE oscillator characteristics (f_{LSE} = 32.768 kHz)(1)

Symbol	Parameter	Conditions(2)	Min	Typ	Max	Unit
$I_{DD(LSE)}$	LSE current consumption	LSEDRV[1:0] = 00 Low drive capability	-	250	-	nA
		LSEDRV[1:0] = 01 Medium low drive capability	-	315	-	
		LSEDRV[1:0] = 10 Medium high drive capability	-	500	-	
		LSEDRV[1:0] = 11 High drive capability	-	630	-	
$Gm_{critmax}$	Maximum critical crystal gm	LSEDRV[1:0] = 00 Low drive capability	-	-	0.5	µA/V
		LSEDRV[1:0] = 01 Medium low drive capability	-	-	0.75	
		LSEDRV[1:0] = 10 Medium high drive capability	-	-	1.7	
		LSEDRV[1:0] = 11 High drive capability	-	-	2.7	
$t_{SU(LSE)}$(3)	Startup time	V_{DD} is stabilized	-	2	-	s

Multi-speed internal (MSI) RC oscillator

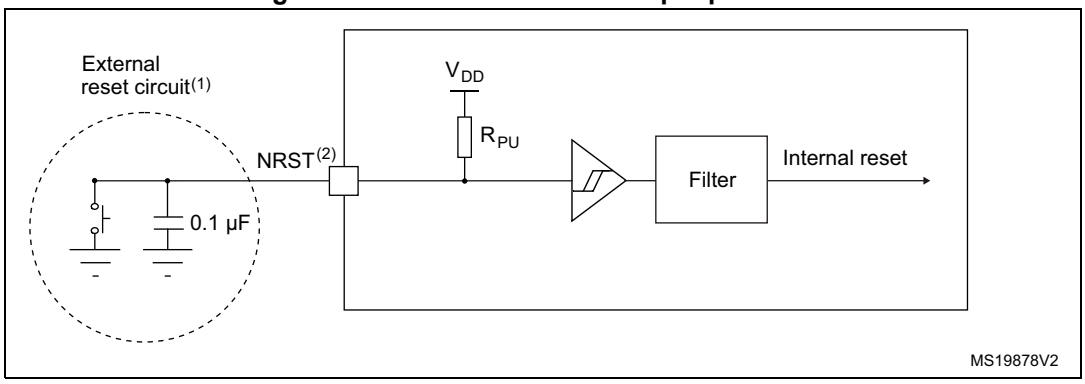
Table 48. MSI oscillator characteristics(1)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
f_{MSI}	MSI frequency after factory calibration, done at V_{DD}=3 V and T_A=30 °C	MSI mode	Range 0	99	100	101	kHz
			Range 1	198	200	202	
			Range 2	396	400	404	
			Range 3	792	800	808	
			Range 4	0.99	1	1.01	MHz
			Range 5	1.98	2	2.02	
			Range 6	3.96	4	4.04	
			Range 7	7.92	8	8.08	
			Range 8	15.8	16	16.16	
			Range 9	23.8	24	24.4	
			Range 10	31.7	32	32.32	
			Range 11	47.5	48	48.48	
		PLL mode XTAL= 32.768 kHz	Range 0	-	98.304	-	kHz
			Range 1	-	196.608	-	
			Range 2	-	393.216	-	
			Range 3	-	786.432	-	
			Range 4	-	1.016	-	MHz
			Range 5	-	1.999	-	
			Range 6	-	3.998	-	
			Range 7	-	7.995	-	
			Range 8	-	15.991	-	
			Range 9	-	23.986	-	
			Range 10	-	32.014	-	
			Range 11	-	48.005	-	
Δ_{TEMP}(MSI)(2)	MSI oscillator frequency drift over temperature	MSI mode	T_A= -0 to 85 °C	-3.5	-	3	%
			T_A= -40 to 125 °C	-8	-	6	

Table 48. MSI oscillator characteristics(1) (continued)

Symbol	Parameter	Conditions			Min	Typ	Max	Unit
Δ_{VDD}(MSI)(2)	MSI oscillator frequency drift over V_{DD} (reference is 3 V)	MSI mode	Range 0 to 3	V_{DD}=1.62 V to 3.6 V	-1.2	-	0.5	%
				V_{DD}=2.4 V to 3.6 V	-0.5	-		
			Range 4 to 7	V_{DD}=1.62 V to 3.6 V	-2.5	-	0.7	
				V_{DD}=2.4 V to 3.6 V	-0.8	-		
			Range 8 to 11	V_{DD}=1.62 V to 3.6 V	-5	-	1	
				V_{DD}=2.4 V to 3.6 V	-1.6	-		
$\Delta F_{SAMPLING}$ (MSI)(2)(6)	Frequency variation in sampling mode(3)	MSI mode	T_A= -40 to 85 °C		-	1	2	%
			T_A= -40 to 125 °C		-	2	4	
P_USB Jitter(MSI)(6)	Period jitter for USB clock(4)	PLL mode Range 11	for next transition	-	-	-	3.458	ns
			for paired transition	-	-	-	3.916	
MT_USB Jitter(MSI)(6)	Medium term jitter for USB clock(5)	PLL mode Range 11	for next transition	-	-	-	2	ns
			for paired transition	-	-	-	1	
CC jitter(MSI)(6)	RMS cycle-to-cycle jitter	PLL mode Range 11		-	-	60	-	ps
P jitter(MSI)(6)	RMS Period jitter	PLL mode Range 11		-	-	50	-	ps
t_{SU}(MSI)(6)	MSI oscillator start-up time	Range 0		-	-	10	20	us
		Range 1		-	-	5	10	
		Range 2		-	-	4	8	
		Range 3		-	-	3	7	
		Range 4 to 7		-	-	3	6	
		Range 8 to 11		-	-	2.5	6	
t_{STAB}(MSI)(6)	MSI oscillator stabilization time	PLL mode Range 11	10 % of final frequency	-	-	0.25	0.5	ms
			5 % of final frequency	-	-	0.5	1.25	
			1 % of final frequency	-	-	-	2.5	

Figure 24. Recommended NRST pin protection



1. The reset network protects the device against parasitic resets.
2. The user must ensure that the level on the NRST pin can go below the $V_{IL(NRST)}$ max level specified in *Table 61: NRST pin characteristics*. Otherwise the reset will not be taken into account by the device.

6.3.16 Analog switches booster

Table 62. Analog switches booster characteristics(1)

Symbol	Parameter	Min	Typ	Max	Unit
V_{DD}	Supply voltage	1.62	-	3.6	V
V_{BOOST}	Boost supply	2.7	-	4	
$t_{SU(BOOST)}$	Booster startup time	-	-	240	µs
$I_{DD(BOOST)}$	Booster consumption for 1.62 V ≤ V_{DD} ≤ 2.0 V	-	-	250	µA
	Booster consumption for 2.0 V ≤ V_{DD} ≤ 2.7 V	-	-	500	
	Booster consumption for 2.7 V ≤ V_{DD} ≤ 3.6 V	-	-	900	

1. Guaranteed by design.

Table 63. ADC characteristics(1) (2) (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{LATRINJ}$	Trigger conversion latency Injected channels aborting a regular conversion	CKMODE = 00	2.5	3	3.5	$1/f_{ADC}$
		CKMODE = 01	-	-	3.0	
		CKMODE = 10	-	-	3.25	
		CKMODE = 11	-	-	3.125	
t_s	Sampling time	f_{ADC} = 80 MHz	0.03125	-	8.00625	µs
		-	2.5	-	640.5	$1/f_{ADC}$
$t_{ADCVREG\_STUP}$	ADC voltage regulator start-up time	-	-	-	20	µs
t_{CONV}	Total conversion time (including sampling time)	f_{ADC} = 80 MHz Resolution = 12 bits	0.1875	-	8.1625	µs
		Resolution = 12 bits	ts + 12.5 cycles for successive approximation = 15 to 653			$1/f_{ADC}$
I_{DDA}(ADC)	ADC consumption from the V_{DDA} supply	fs = 5 Msps	-	730	830	µA
		fs = 1 Msps	-	160	220	
		fs = 10 ksps	-	16	50	
I_{DDV_S}(ADC)	ADC consumption from the V_{REF+} single ended mode	fs = 5 Msps	-	130	160	µA
		fs = 1 Msps	-	30	40	
		fs = 10 ksps	-	0.6	2	
I_{DDV_D}(ADC)	ADC consumption from the V_{REF+} differential mode	fs = 5 Msps	-	260	310	µA
		fs = 1 Msps	-	60	70	
		fs = 10 ksps	-	1.3	3	

1. Guaranteed by design
2. The I/O analog switch voltage booster is enable when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4V). It is disable when V_{DDA} ≥ 2.4 V.
3. V_{REF+} can be internally connected to V_{DDA} and V_{REF-} can be internally connected to V_{SSA}, depending on the package. Refer to *Section 4: Pinouts and pin description* for further details.

6.3.20 Comparator characteristics

Table 72. COMP characteristics(1)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
V_{DDA}	Analog supply voltage	-		1.62	-	3.6	V
V_{IN}	Comparator input voltage range	-		0	-	V_{DDA}	
V_{BG}(2)	Scaler input voltage	-		V_{REFINT}			
V_{SC}	Scaler offset voltage	-		-	±5	±10	mV
I_{DDA}(SCALER)	Scaler static consumption from V_{DDA}	BRG_EN=0 (bridge disable)		-	200	300	nA
		BRG_EN=1 (bridge enable)		-	0.8	1	µA
t_{START_SCALER}	Scaler startup time	-		-	100	200	µs
t_{START}	Comparator startup time to reach propagation delay specification	High-speed mode	$V_{DDA} \geq 2.7$ V	-	-	5	µs
			$V_{DDA} < 2.7$ V	-	-	7	
		Medium mode	$V_{DDA} \geq 2.7$ V	-	-	15	
			$V_{DDA} < 2.7$ V	-	-	25	
		Ultra-low-power mode		-	-	80	
t_D(3)	Propagation delay for 200 mV step with 100 mV overdrive	High-speed mode	$V_{DDA} \geq 2.7$ V	-	55	80	ns
			$V_{DDA} < 2.7$ V	-	65	100	
		Medium mode	$V_{DDA} \geq 2.7$ V	-	0.55	0.9	µs
			$V_{DDA} < 2.7$ V	-	0.65	1	
		Ultra-low-power mode		-	5	12	
V_{offset}	Comparator offset error	Full common mode range	-	-	±5	±20	mV
V_{hys}	Comparator hysteresis	No hysteresis		-	0	-	mV
		Low hysteresis		-	8	-	
		Medium hysteresis		-	15	-	
		High hysteresis		-	27	-	
I_{DDA}(COMP)	Comparator consumption from V_{DDA}	Ultra-low-power mode	Static	-	400	600	nA
			With 50 kHz ±100 mV overdrive square signal	-	1200	-	
		Medium mode	Static	-	5	7	µA
			With 50 kHz ±100 mV overdrive square signal	-	6	-	
		High-speed mode	Static	-	70	100	
			With 50 kHz ±100 mV overdrive square signal	-	75	-	

1. Guaranteed by design.
2. LCD enabled with 3 V internal step-up active, 1/8 duty, 1/4 bias, division ratio= 64, all pixels active, no LCD connected.

Table 83. SPI characteristics(1) (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{v(SO)}$	Data output valid time	Slave mode 2.7 < V_{DD} < 3.6 V Voltage Range 1	-	12.5	19	ns
		Slave mode 1.71 < V_{DD} < 3.6 V Voltage Range 1	-	12.5	30	
		Slave mode 1.71 < V_{DD} < 3.6 V Voltage Range 2	-	12.5	33	
-		Slave mode 1.08 < V_{DDIO2} < 1.32 V(3)	-	25	62.5	
$t_{v(MO)}$		Master mode	-	2.5	12.5	
$t_{h(SO)}$	Data output hold time	Slave mode	9	-	-	ns
-		Slave mode 1.08 < V_{DDIO2} < 1.32 V(3)	24	-	-	
$t_{h(MO)}$		Master mode	0	-	-	

1. Guaranteed by characterization results.
2. Maximum frequency in Slave transmitter mode is determined by the sum of $t_{v(SO)}$ and $t_{su(MI)}$ which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having $t_{su(MI)}$ = 0 while Duty(SCK) = 50 %.
3. SPI mapped on Port G.

Figure 28. SPI timing diagram - slave mode and CPHA = 0

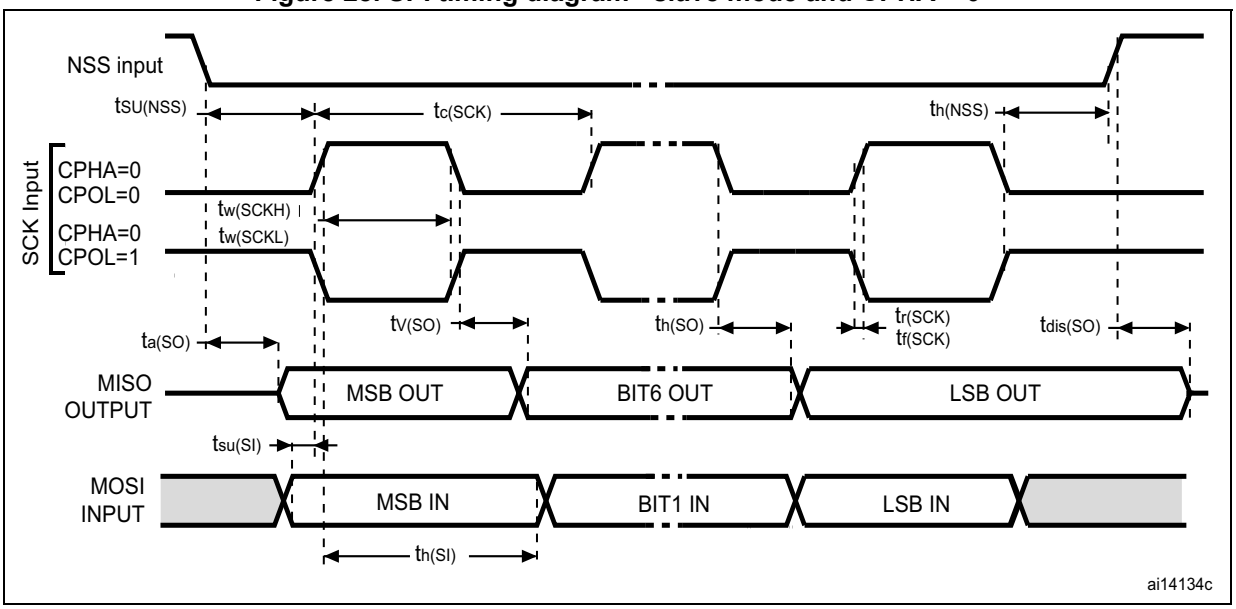


Table 86. SAI characteristics(1) (continued)

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{v(SD\_B\_ST)}$	Data output valid time	Slave transmitter (after enable edge) 2.7 ≤ V_{DD} ≤ 3.6	-	22	ns
		Slave transmitter (after enable edge) 1.71 ≤ V_{DD} ≤ 3.6	-	34	
$t_{h(SD\_B\_ST)}$	Data output hold time	Slave transmitter (after enable edge)	10	-	ns
$t_{v(SD\_A\_MT)}$	Data output valid time	Master transmitter (after enable edge) 2.7 ≤ V_{DD} ≤ 3.6	-	27	ns
		Master transmitter (after enable edge) 1.71 ≤ V_{DD} ≤ 3.6	-	40	
$t_{h(SD\_A\_MT)}$	Data output hold time	Master transmitter (after enable edge)	10	-	ns

1. Guaranteed by characterization results.
2. APB clock frequency must be at least twice SAI clock frequency.

Figure 33. SAI master timing waveforms

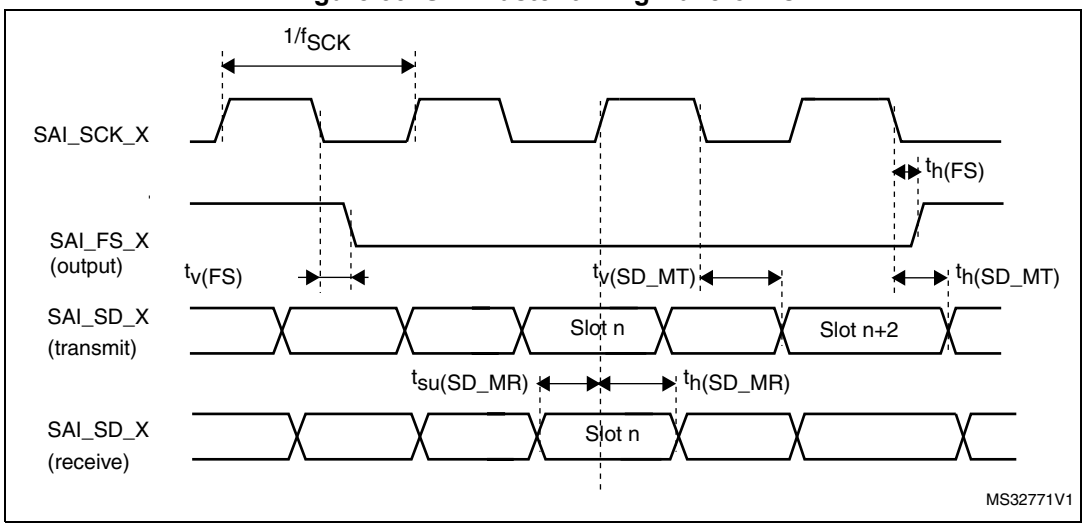


Table 93. Asynchronous non-multiplexed SRAM/PSRAM/NOR write-NWAIT timings(1)(2)

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$8T_{HCLK}+0.5$	$8T_{HCLK}+0.5$	ns
$t_{w(NWE)}$	FMC_NWE low time	$6T_{HCLK}-0.5$	$6T_{HCLK}+0.5$	
$t_{su(NWAIT\_NE)}$	FMC_NWAIT valid before FMC_NEx high	$6T_{HCLK}+2$	-	
$t_{h(NE\_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}+2$	-	

1. CL = 30 pF.
2. Guaranteed by characterization results.

Figure 39. Asynchronous multiplexed PSRAM/NOR read waveforms

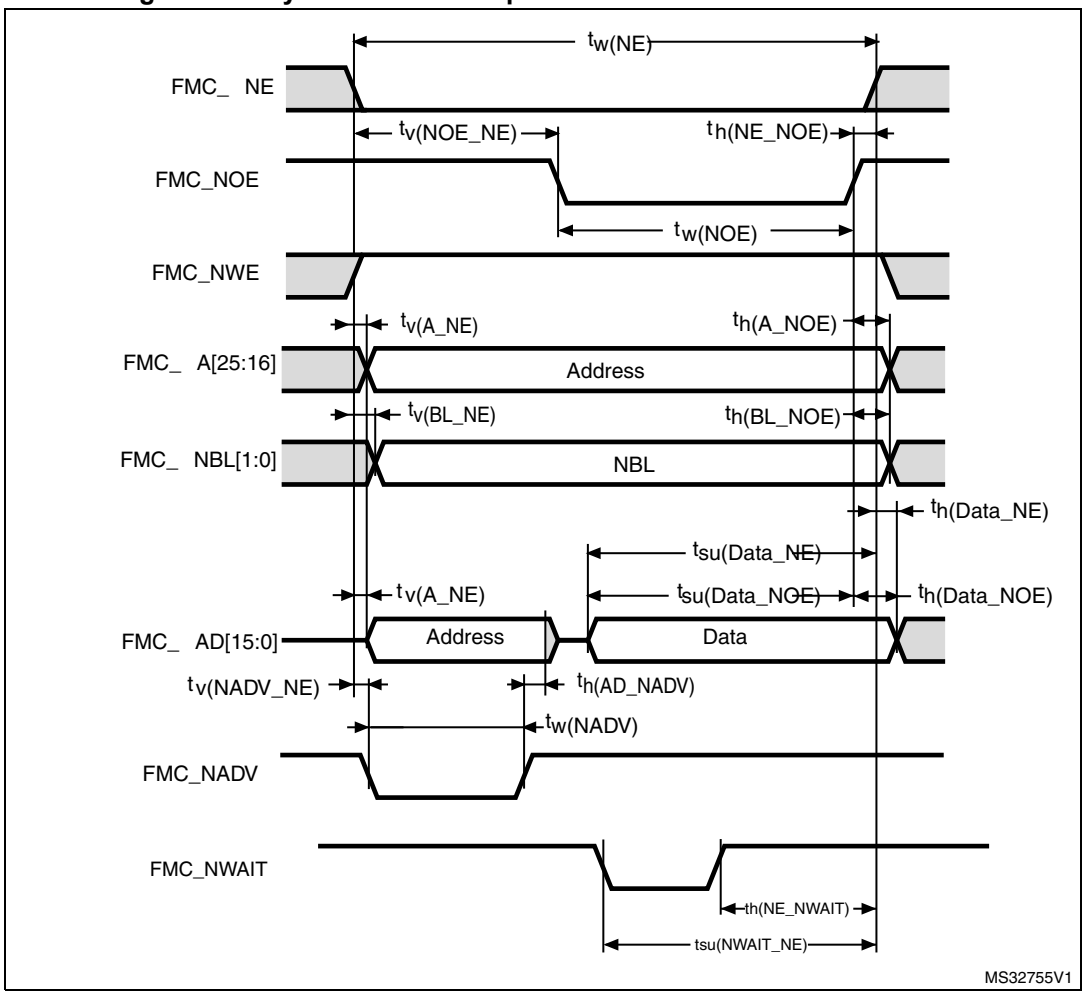


Figure 40. Asynchronous multiplexed PSRAM/NOR write waveforms

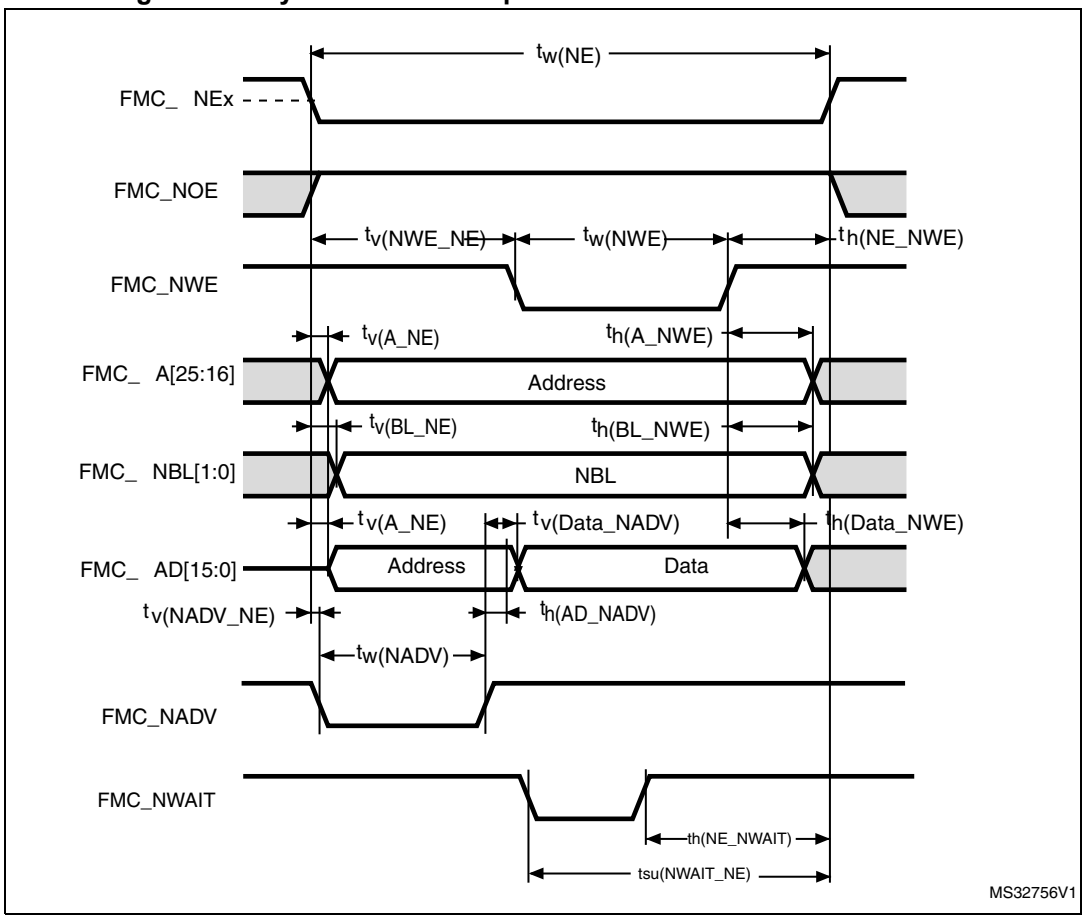


Table 96. Asynchronous multiplexed PSRAM/NOR write timings(1)(2)

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$4T_{HCLK}$-0.5	$4T_{HCLK}$+2	ns
$t_{v(NWE\_NE)}$	FMC_NEx low to FMC_NWE low	T_{HCLK}-0.5	T_{HCLK}+1	
$t_{w(NWE)}$	FMC_NWE low time	$2xT_{HCLK}$-1.5	$2xT_{HCLK}$+1.5	
$t_{h(NE\_NWE)}$	FMC_NWE high to FMC_NE high hold time	T_{HCLK}-0.5	-	
$t_{v(A\_NE)}$	FMC_NEx low to FMC_A valid	-	3	
$t_{v(NADV\_NE)}$	FMC_NEx low to FMC_NADV low	0	1	
$t_{w(NADV)}$	FMC_NADV low time	T_{HCLK}-0.5	T_{HCLK}+1	
$t_{h(AD\_NADV)}$	FMC_AD(adress) valid hold time after FMC_NADV high	T_{HCLK}-2	-	
$t_{h(A\_NWE)}$	Address hold time after FMC_NWE high	T_{HCLK}-1	-	
$t_{h(BL\_NWE)}$	FMC_BL hold time after FMC_NWE high	T_{HCLK}+0.5	-	
$t_{v(BL\_NE)}$	FMC_NEx low to FMC_BL valid	-	1.5	
$t_{v(Data\_NADV)}$	FMC_NADV high to Data valid	-	T_{HCLK} +4	
$t_{h(Data\_NWE)}$	Data hold time after FMC_NWE high	T_{HCLK} +0.5	-	

1. CL = 30 pF.
2. Guaranteed by characterization results.

Table 97. Asynchronous multiplexed PSRAM/NOR write-NWAIT timings(1)(2)

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$9T_{HCLK}$-0.5	$9T_{HCLK}$+2	ns
$t_{w(NWE)}$	FMC_NWE low time	$7T_{HCLK}$-1.5	$7T_{HCLK}$+1.5	
$t_{su(NWAIT\_NE)}$	FMC_NWAIT valid before FMC_NEx high	$6T_{HCLK}$+2	-	
$t_{h(NE\_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}$-3	-	

1. CL = 30 pF.
2. Guaranteed by characterization results.

Synchronous waveforms and timings

Figure 41 through *Figure 44* represent synchronous waveforms and *Table 98* through *Table 101* provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- BurstAccessMode = FMC_BurstAccessMode_Enable
- MemoryType = FMC_MemoryType_CRAM
- WriteBurst = FMC_WriteBurst_Enable
- CLKDivision = 1
- DataLatency = 1 for NOR Flash; DataLatency = 0 for PSRAM

In all timing tables, the T_{HCLK} is the HCLK clock period.

Figure 41. Synchronous multiplexed NOR/PSRAM read timings

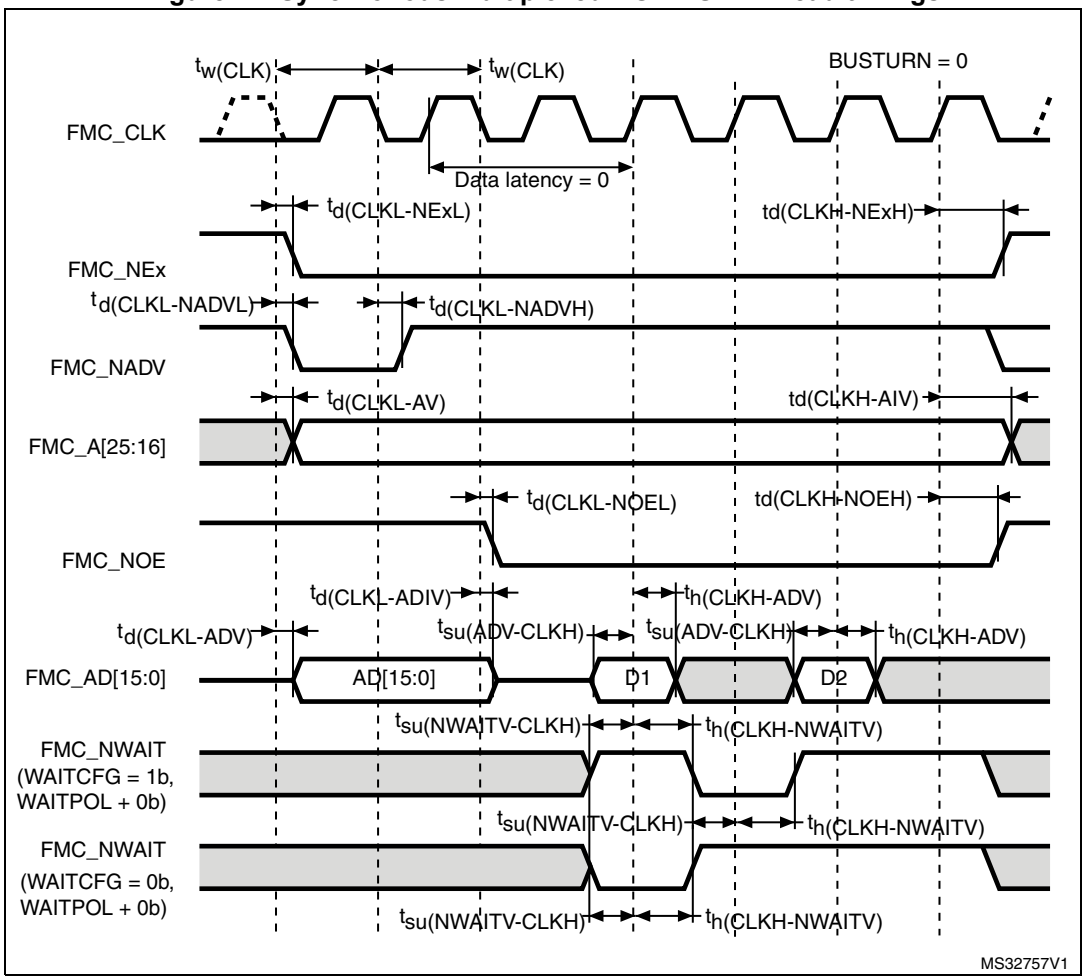
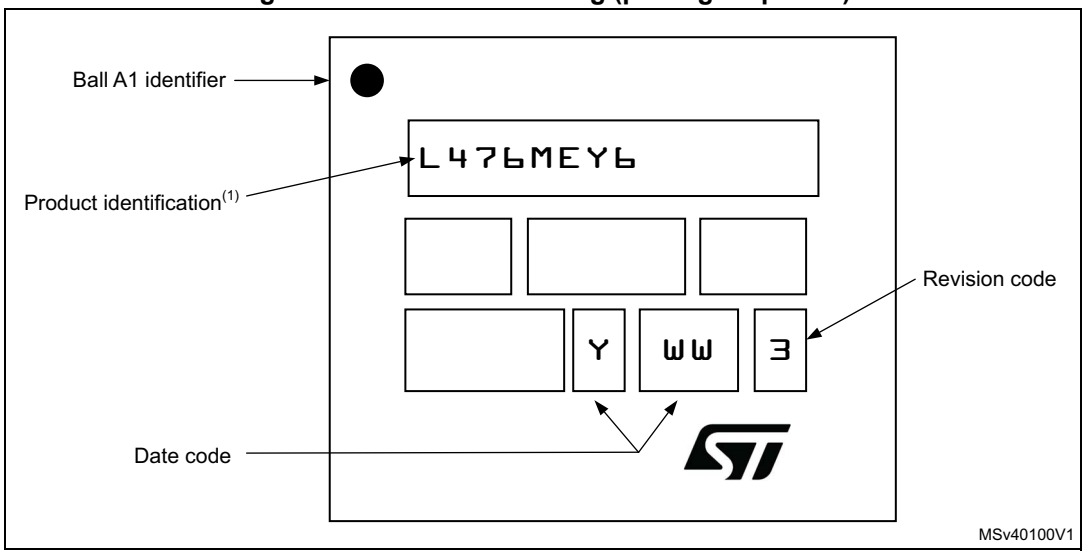


Table 98. Synchronous multiplexed NOR/PSRAM read timings(1)(2)

Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FMC_CLK period	$2T_{HCLK}$-1	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low (x=0..2)	-	2	
$t_{d(CLKH\_NExH)}$	FMC_CLK high to FMC_NEx high (x= 0…2)	T_{HCLK}+0.5	-	
$t_{d(CLKL-NADVL)}$	FMC_CLK low to FMC_NADV low	-	2.5	
$t_{d(CLKL-NADVH)}$	FMC_CLK low to FMC_NADV high	1	-	
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid (x=16…25)	-	3.5	
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid (x=16…25)	T_{HCLK}	-	
$t_{d(CLKL-NOEL)}$	FMC_CLK low to FMC_NOE low	-	1.5	
$t_{d(CLKH-NOEH)}$	FMC_CLK high to FMC_NOE high	T_{HCLK}+1	-	
$t_{d(CLKL-ADV)}$	FMC_CLK low to FMC_AD[15:0] valid	-	4	
$t_{d(CLKL-ADIV)}$	FMC_CLK low to FMC_AD[15:0] invalid	0	-	
$t_{su(ADV-CLKH)}$	FMC_A/D[15:0] valid data before FMC_CLK high	0	-	
$t_{h(CLKH-ADV)}$	FMC_A/D[15:0] valid data after FMC_CLK high	2.5	-	
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	0	-	
$t_{h(CLKH-NWAIT)}$	FMC_NWAIT valid after FMC_CLK high	4	-	

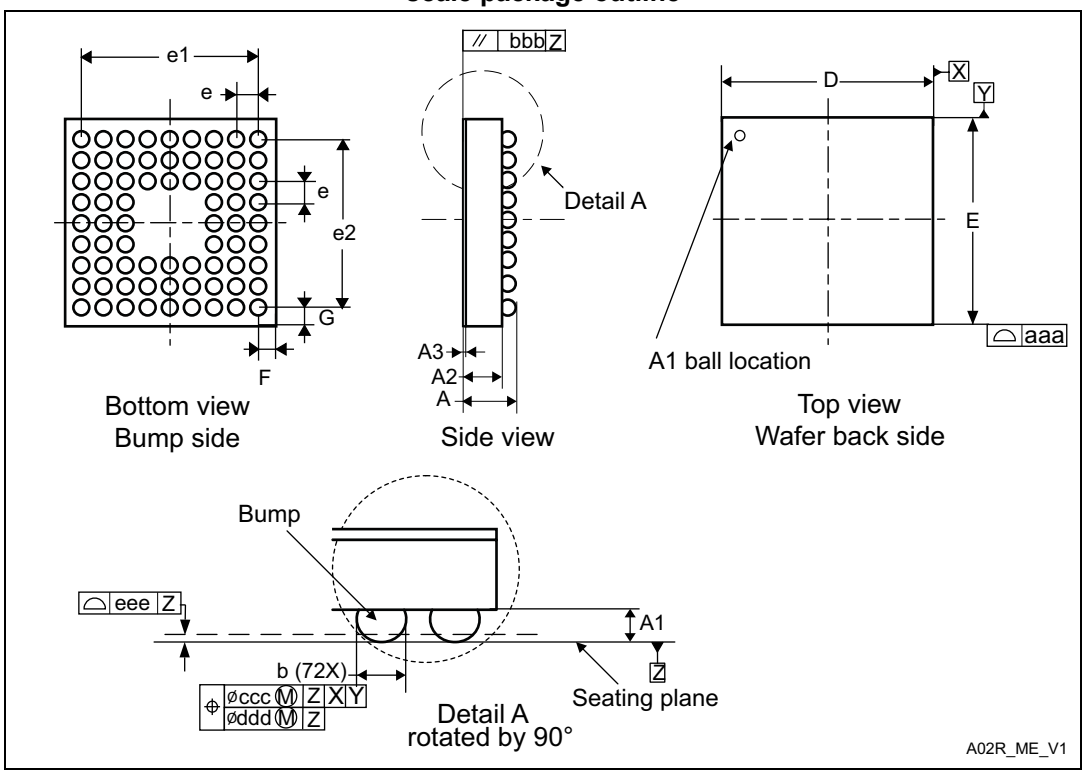
1. CL = 30 pF.
2. Guaranteed by characterization results.

Figure 60. WLCSP81 marking (package top view)



7.5 WLCSP72 package information

Figure 61. WLCSP72 - 72-ball, 4.4084 x 3.7594 mm, 0.4 mm pitch wafer level chip scale package outline



1. Drawing is not to scale.