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### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Obsolete
Core Processor	C166SV2
Core Size	16/32-Bit
Speed	66MHz
Connectivity	CANbus, EBI/EMI, I <sup>2</sup> C, LINbus, SPI, SSC, UART/USART, USI
Peripherals	I <sup>2</sup> S, POR, PWM, WDT
Number of I/O	38
Program Memory Size	320KB (320K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	42K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 9x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP Exposed Pad
Supplier Device Package	PG-LQFP-64-6
Purchase URL	<a href="https://www.e-xfl.com/product-detail/infineon-technologies/saf-xc2236n-40f66l-ab">https://www.e-xfl.com/product-detail/infineon-technologies/saf-xc2236n-40f66l-ab</a>



**Table 6 Pin Definitions and Functions (cont'd)**

<b>Pin</b>	<b>Symbol</b>	<b>Ctrl.</b>	<b>Type</b>	<b>Function</b>
7	P6.0	O0 / I	DA/A	<b>Bit 0 of Port 6, General Purpose Input/Output</b>
	EMUX0	O1	DA/A	<b>External Analog MUX Control Output 0 (ADC0)</b>
	TxDC2	O2	DA/A	<b>CAN Node 2 Transmit Data Output</b>
	BRKOUT	O3	DA/A	<b>OCDS Break Signal Output</b>
	ADCx_REQG TyG	I	DA/A	<b>External Request Gate Input for ADC0/1</b>
	U1C1_DX0E	I	DA/A	<b>USIC1 Channel 1 Shift Data Input</b>
8	P6.1	O0 / I	DA/A	<b>Bit 1 of Port 6, General Purpose Input/Output</b>
	EMUX1	O1	DA/A	<b>External Analog MUX Control Output 1 (ADC0)</b>
	T3OUT	O2	DA/A	<b>GPT12E Timer T3 Toggle Latch Output</b>
	U1C1_DOUT	O3	DA/A	<b>USIC1 Channel 1 Shift Data Output</b>
	ADCx_REQT RyE	I	DA/A	<b>External Request Trigger Input for ADC0/1</b>
	RxDC2E	I	DA/A	<b>CAN Node 2 Receive Data Input</b>
10	ESR1_6	I	DA/A	<b>ESR1 Trigger Input 6</b>
10	P15.0	I	In/A	<b>Bit 0 of Port 15, General Purpose Input</b>
	ADC1_CH0	I	In/A	<b>Analog Input Channel 0 for ADC1</b>
11	P15.4	I	In/A	<b>Bit 4 of Port 15, General Purpose Input</b>
	ADC1_CH4	I	In/A	<b>Analog Input Channel 4 for ADC1</b>
	T6INA	I	In/A	<b>GPT12E Timer T6 Count/Gate Input</b>
12	$V_{AREF}$	-	PS/A	<b>Reference Voltage for A/D Converters ADC0/1</b>
13	$V_{AGND}$	-	PS/A	<b>Reference Ground for A/D Converters ADC0/1</b>
14	P5.0	I	In/A	<b>Bit 0 of Port 5, General Purpose Input</b>
	ADC0_CH0	I	In/A	<b>Analog Input Channel 0 for ADC0</b>
15	P5.2	I	In/A	<b>Bit 2 of Port 5, General Purpose Input</b>
	ADC0_CH2	I	In/A	<b>Analog Input Channel 2 for ADC0</b>
	TDI_A	I	In/A	<b>JTAG Test Data Input</b>

**Table 6 Pin Definitions and Functions (cont'd)**

<b>Pin</b>	<b>Symbol</b>	<b>Ctrl.</b>	<b>Type</b>	<b>Function</b>
19	P5.4	I	In/A	<b>Bit 4 of Port 5, General Purpose Input</b>
	ADC0_CH4	I	In/A	<b>Analog Input Channel 4 for ADC0</b>
	T3EUDA	I	In/A	<b>GPT12E Timer T3 External Up/Down Control Input</b>
	TMS_A	I	In/A	<b>JTAG Test Mode Selection Input</b>
20	P5.8	I	In/A	<b>Bit 8 of Port 5, General Purpose Input</b>
	ADC0_CH8	I	In/A	<b>Analog Input Channel 8 for ADC0</b>
	ADC1_CH8	I	In/A	<b>Analog Input Channel 8 for ADC1</b>
	CCU6x_T12H RC	I	In/A	<b>External Run Control Input for T12 of CCU60/1</b>
	CCU6x_T13H RC	I	In/A	<b>External Run Control Input for T13 of CCU60/1</b>
	U2C0_DX0F	I	In/A	<b>USIC2 Channel 0 Shift Data Input</b>
21	P5.10	I	In/A	<b>Bit 10 of Port 5, General Purpose Input</b>
	ADC0_CH10	I	In/A	<b>Analog Input Channel 10 for ADC0</b>
	ADC1_CH10	I	In/A	<b>Analog Input Channel 10 for ADC1</b>
	BRKIN_A	I	In/A	<b>OCDS Break Signal Input</b>
	U2C1_DX0F	I	In/A	<b>USIC2 Channel 1 Shift Data Input</b>
	CCU61_T13 HRA	I	In/A	<b>External Run Control Input for T13 of CCU61</b>
22	P5.13	I	In/A	<b>Bit 13 of Port 5, General Purpose Input</b>
	ADC0_CH13	I	In/A	<b>Analog Input Channel 13 for ADC0</b>
23	P5.15	I	In/A	<b>Bit 15 of Port 5, General Purpose Input</b>
	ADC0_CH15	I	In/A	<b>Analog Input Channel 15 for ADC0</b>
	RxDC2F	I	In/A	<b>CAN Node 2 Receive Data Input</b>
25	P2.0	O0 / I	St/B	<b>Bit 0 of Port 2, General Purpose Input/Output</b>
	TxDC5	O1	St/B	<b>CAN Node 5 Transmit Data Output</b>
	RxDC0C	I	St/B	<b>CAN Node 0 Receive Data Input</b>
	T5INB	I	St/B	<b>GPT12E Timer T5 Count/Gate Input</b>

**Table 6 Pin Definitions and Functions (cont'd)**

<b>Pin</b>	<b>Symbol</b>	<b>Ctrl.</b>	<b>Type</b>	<b>Function</b>
40	P10.2	O0 / I	St/B	<b>Bit 2 of Port 10, General Purpose Input/Output</b>
	U0C0_SCLK OUT	O1	St/B	<b>USIC0 Channel 0 Shift Clock Output</b>
	CCU60_CC6 2	O2	St/B	<b>CCU60 Channel 2 Output</b>
	CCU60_CC6 2INA	I	St/B	<b>CCU60 Channel 2 Input</b>
	U0C0_DX1B	I	St/B	<b>USIC0 Channel 0 Shift Clock Input</b>
42	P2.10	O0 / I	St/B	<b>Bit 10 of Port 2, General Purpose Input/Output</b>
	U0C1_DOUT	O1	St/B	<b>USIC0 Channel 1 Shift Data Output</b>
	U0C0_SELO 3	O2	St/B	<b>USIC0 Channel 0 Select/Control 3 Output</b>
	CC2_CC23	O3 / I	St/B	<b>CAPCOM2 CC23IO Capture Inp./ Compare Out.</b>
	U0C1_DX0E	I	St/B	<b>USIC0 Channel 1 Shift Data Input</b>
	CAPINA	I	St/B	<b>GPT12E Register CAPREL Capture Input</b>
43	P10.3	O0 / I	St/B	<b>Bit 3 of Port 10, General Purpose Input/Output</b>
	CCU60_COU T60	O2	St/B	<b>CCU60 Channel 0 Output</b>
	U0C0_DX2A	I	St/B	<b>USIC0 Channel 0 Shift Control Input</b>
	U0C1_DX2A	I	St/B	<b>USIC0 Channel 1 Shift Control Input</b>
44	P10.4	O0 / I	St/B	<b>Bit 4 of Port 10, General Purpose Input/Output</b>
	U0C0_SELO 3	O1	St/B	<b>USIC0 Channel 0 Select/Control 3 Output</b>
	CCU60_COU T61	O2	St/B	<b>CCU60 Channel 1 Output</b>
	U0C0_DX2B	I	St/B	<b>USIC0 Channel 0 Shift Control Input</b>
	U0C1_DX2B	I	St/B	<b>USIC0 Channel 1 Shift Control Input</b>
	ESR1_9	I	St/B	<b>ESR1 Trigger Input 9</b>

**General Device Information**

**Table 6 Pin Definitions and Functions (cont'd)**

<b>Pin</b>	<b>Symbol</b>	<b>Ctrl.</b>	<b>Type</b>	<b>Function</b>
51	P10.8	O0 / I	St/B	<b>Bit 8 of Port 10, General Purpose Input/Output</b>
	U0C0_MCLK OUT	O1	St/B	<b>USIC0 Channel 0 Master Clock Output</b>
	U0C1_SELO 0	O2	St/B	<b>USIC0 Channel 1 Select/Control 0 Output</b>
	U2C1_DOUT	O3	St/B	<b>USIC2 Channel 1 Shift Data Output</b>
	CCU60_CCP OS1A	I	St/B	<b>CCU60 Position Input 1</b>
	U0C0_DX1C	I	St/B	<b>USIC0 Channel 0 Shift Clock Input</b>
	BRKIN_B	I	St/B	<b>OCDS Break Signal Input</b>
	T3EUDB	I	St/B	<b>GPT12E Timer T3 External Up/Down Control Input</b>
52	P10.9	O0 / I	St/B	<b>Bit 9 of Port 10, General Purpose Input/Output</b>
	U0C0_SELO 4	O1	St/B	<b>USIC0 Channel 0 Select/Control 4 Output</b>
	U0C1_MCLK OUT	O2	St/B	<b>USIC0 Channel 1 Master Clock Output</b>
	CCU60_CCP OS2A	I	St/B	<b>CCU60 Position Input 2</b>
	TCK_B	IH	St/B	<b>DAP0/JTAG Clock Input</b> If JTAG pos. B is selected during start-up, an internal pull-up device will hold this pin high when nothing is driving it. If DAP pos. 1 is selected during start-up, an internal pull-down device will hold this pin low when nothing is driving it.
	T3INB	I	St/B	<b>GPT12E Timer T3 Count/Gate Input</b>

### **3.8 Capture/Compare Units CCU6x**

The XC223xN types feature the CCU60, CCU61 unit(s).

CCU6 is a high-resolution capture and compare unit with application-specific modes. It provides inputs to start the timers synchronously, an important feature in devices with several CCU6 modules.

The module provides two independent timers (T12, T13), that can be used for PWM generation, especially for AC motor control. Additionally, special control modes for block commutation and multi-phase machines are supported.

#### **Timer 12 Features**

- Three capture/compare channels, where each channel can be used either as a capture or as a compare channel.
- Supports generation of a three-phase PWM (six outputs, individual signals for high-side and low-side switches)
- 16-bit resolution, maximum count frequency = peripheral clock
- Dead-time control for each channel to avoid short circuits in the power stage
- Concurrent update of the required T12/13 registers
- Center-aligned and edge-aligned PWM can be generated
- Single-shot mode supported
- Many interrupt request sources
- Hysteresis-like control mode
- Automatic start on a HW event (T12HR, for synchronization purposes)

#### **Timer 13 Features**

- One independent compare channel with one output
- 16-bit resolution, maximum count frequency = peripheral clock
- Can be synchronized to T12
- Interrupt generation at period match and compare match
- Single-shot mode supported
- Automatic start on a HW event (T13HR, for synchronization purposes)

#### **Additional Features**

- Block commutation for brushless DC drives implemented
- Position detection via Hall sensor pattern
- Automatic rotational speed measurement for block commutation
- Integrated error handling
- Fast emergency stop without CPU load via external signal ( $\overline{\text{CTRAP}}$ )
- Control modes for multi-channel AC drives
- Output levels can be selected and adapted to the power stage

### **3.11 A/D Converters**

For analog signal measurement, up to two 10-bit A/D converters (ADC0, ADC1) with 7 + 2 multiplexed input channels and a sample and hold circuit have been integrated on-chip. 2 inputs can be converted by both A/D converters. Conversions use the successive approximation method. The sample time (to charge the capacitors) and the conversion time are programmable so that they can be adjusted to the external circuit. The A/D converters can also operate in 8-bit conversion mode, further reducing the conversion time.

Several independent conversion result registers, selectable interrupt requests, and highly flexible conversion sequences provide a high degree of programmability to meet the application requirements. Both modules can be synchronized to allow parallel sampling of two input channels.

For applications that require more analog input channels, external analog multiplexers can be controlled automatically. For applications that require fewer analog input channels, the remaining channel inputs can be used as digital input port pins.

The A/D converters of the XC223xN support two types of request sources which can be triggered by several internal and external events.

- Parallel requests are activated at the same time and then executed in a predefined sequence.
- Queued requests are executed in a user-defined sequence.

In addition, the conversion of a specific channel can be inserted into a running sequence without disturbing that sequence. All requests are arbitrated according to the priority level assigned to them.

Data reduction features reduce the number of required CPU access operations allowing the precise evaluation of analog inputs (high conversion rate) even at a low CPU speed. Result data can be reduced by limit checking or accumulation of results.

The Peripheral Event Controller (PEC) can be used to control the A/D converters or to automatically store conversion results to a table in memory for later evaluation, without requiring the overhead of entering and exiting interrupt routines for each data transfer. Each A/D converter contains eight result registers which can be concatenated to build a result FIFO. Wait-for-read mode can be enabled for each result register to prevent the loss of conversion data.

In order to decouple analog inputs from digital noise and to avoid input trigger noise, those pins used for analog input can be disconnected from the digital input stages. This can be selected for each pin separately with the Port x Digital Input Disable registers.

The Auto-Power-Down feature of the A/D converters minimizes the power consumption when no conversion is in progress.

Broken wire detection for each channel and a multiplexer test mode provide information to verify the proper operation of the analog signal sources (e.g. a sensor system).



## Target Protocols

Each USIC channel can receive and transmit data frames with a selectable data word width from 1 to 16 bits in each of the following protocols:

- **UART** (asynchronous serial channel)
  - module capability: maximum baud rate =  $f_{\text{SYS}} / 4$
  - data frame length programmable from 1 to 63 bits
  - MSB or LSB first
- **LIN** Support (Local Interconnect Network)
  - module capability: maximum baud rate =  $f_{\text{SYS}} / 16$
  - checksum generation under software control
  - baud rate detection possible by built-in capture event of baud rate generator
- **SSC/SPI** (synchronous serial channel with or without data buffer)
  - module capability: maximum baud rate =  $f_{\text{SYS}} / 2$ , limited by loop delay
  - number of data bits programmable from 1 to 63, more with explicit stop condition
  - MSB or LSB first
  - optional control of slave select signals
- **IIC** (Inter-IC Bus)
  - supports baud rates of 100 kbit/s and 400 kbit/s
- **IIS** (Inter-IC Sound Bus)
  - module capability: maximum baud rate =  $f_{\text{SYS}} / 2$

*Note: Depending on the selected functions (such as digital filters, input synchronization stages, sample point adjustment, etc.), the maximum achievable baud rate can be limited. Please note that there may be additional delays, such as internal or external propagation delays and driver delays (e.g. for collision detection in UART mode, for IIC, etc.).*

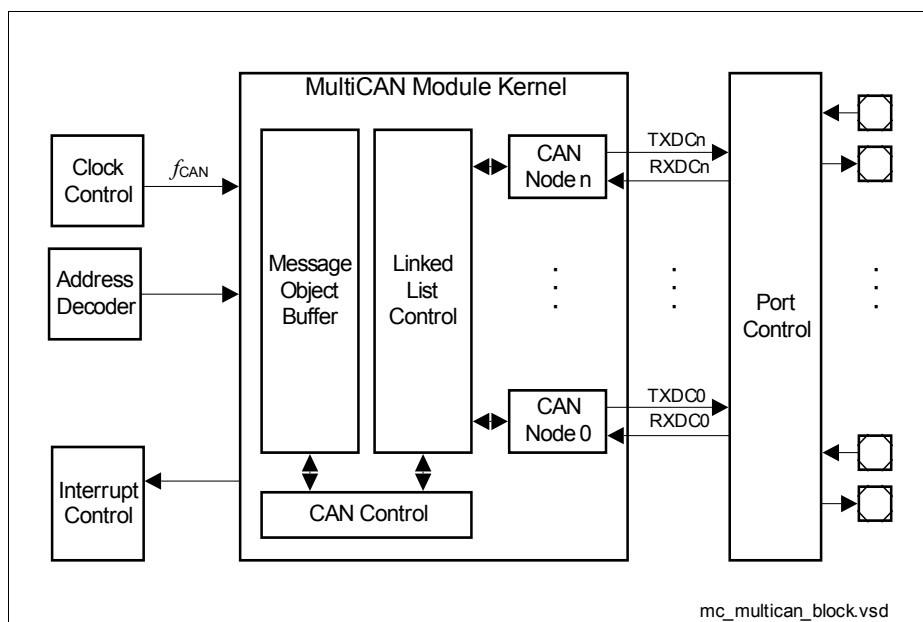
### 3.13 MultiCAN Module

The MultiCAN module contains independently operating CAN nodes with Full-CAN functionality which are able to exchange Data and Remote Frames using a gateway function. Transmission and reception of CAN frames is handled in accordance with CAN specification V2.0 B (active). Each CAN node can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers.

All CAN nodes share a common set of message objects. Each message object can be individually allocated to one of the CAN nodes. Besides serving as a storage container for incoming and outgoing frames, message objects can be combined to build gateways between the CAN nodes or to set up a FIFO buffer.

*Note: The number of CAN nodes and message objects depends on the selected device type.*

The message objects are organized in double-chained linked lists, where each CAN node has its own list of message objects. A CAN node stores frames only into message objects that are allocated to its own message object list and it transmits only messages belonging to this message object list. A powerful, command-driven list controller performs all message object list operations.



**Figure 12 Block Diagram of MultiCAN Module**

### 3.19 Instruction Set Summary

**Table 11** lists the instructions of the XC223xN.

The addressing modes that can be used with a specific instruction, the function of the instructions, parameters for conditional execution of instructions, and the opcodes for each instruction can be found in the **“Instruction Set Manual”**.

This document also provides a detailed description of each instruction.

**Table 11 Instruction Set Summary**

<b>Mnemonic</b>	<b>Description</b>	<b>Bytes</b>
ADD(B)	Add word (byte) operands	2 / 4
ADDC(B)	Add word (byte) operands with Carry	2 / 4
SUB(B)	Subtract word (byte) operands	2 / 4
SUBC(B)	Subtract word (byte) operands with Carry	2 / 4
MUL(U)	(Un)Signed multiply direct GPR by direct GPR (16- × 16-bit)	2
DIV(U)	(Un)Signed divide register MDL by direct GPR (16-/16-bit)	2
DIVL(U)	(Un)Signed long divide reg. MD by direct GPR (32-/16-bit)	2
CPL(B)	Complement direct word (byte) GPR	2
NEG(B)	Negate direct word (byte) GPR	2
AND(B)	Bitwise AND, (word/byte operands)	2 / 4
OR(B)	Bitwise OR, (word/byte operands)	2 / 4
XOR(B)	Bitwise exclusive OR, (word/byte operands)	2 / 4
BCLR/BSET	Clear/Set direct bit	2
BMOV(N)	Move (negated) direct bit to direct bit	4
BAND/BOR/BXOR	AND/OR/XOR direct bit with direct bit	4
BCMP	Compare direct bit to direct bit	4
BFLDH/BFLDL	Bitwise modify masked high/low byte of bit-addressable direct word memory with immediate data	4
CMP(B)	Compare word (byte) operands	2 / 4
CMPD1/2	Compare word data to GPR and decrement GPR by 1/2	2 / 4
CMPI1/2	Compare word data to GPR and increment GPR by 1/2	2 / 4
PRIOR	Determine number of shift cycles to normalize direct word GPR and store result in direct word GPR	2
SHL/SHR	Shift left/right direct word GPR	2

**Electrical Parameters**

**Table 13      Operating Conditions (cont'd)**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Overload current coupling factor for digital I/O pins	$K_{OVD}$ CC	—	$1.0 \times 10^{-2}$	$3.0 \times 10^{-2}$	-	$I_{OV} < 0$ mA; not subject to production test
		—	$1.0 \times 10^{-4}$	$5.0 \times 10^{-3}$	-	$I_{OV} > 0$ mA; not subject to production test
Absolute sum of overload currents	$\Sigma I_{OV} $ SR	—	—	50	mA	not subject to production test
Digital core supply voltage for domain M <sup>8)</sup>	$V_{DDIM}$ CC	—	1.5	—		
Digital core supply voltage for domain 1 <sup>8)</sup>	$V_{DDI1}$ CC	—	1.5	—		
Digital supply voltage for IO pads and voltage regulators	$V_{DDP}$ SR	3.0	—	5.5	V	
Digital ground voltage	$V_{SS}$ SR	—	0	—	V	

- 1) To ensure the stability of the voltage regulators the EVRs must be buffered with ceramic capacitors. Separate buffer capacitors with the recommended values shall be connected as close as possible to each  $V_{DDIM}$  and  $V_{DDI1}$  pin to keep the resistance of the board tracks below 2 Ohm. Connect all  $V_{DDI1}$  pins together. The minimum capacitance value is required for proper operation under all conditions (e.g. temperature). Higher values slightly increase the startup time.
- 2) Use one Capacitor for each pin.
- 3) This is the reference load. For bigger capacitive loads, use the derating factors listed in the pad properties section.
- 4) The timing is valid for pin drivers operating in default current mode (selected after reset). Reducing the output current may lead to increased delays or reduced driving capability ( $C_L$ ).
- 5) The operating frequency range may be reduced for specific device types. This is indicated in the device designation (...FxxL). 80 MHz devices are marked ...F80L.
- 6) Overload conditions occur if the standard operating conditions are exceeded, i.e. the voltage on any pin exceeds the specified range:  $V_{OV} > V_{IHmax}$  ( $I_{OV} > 0$ ) or  $V_{OV} < V_{ILmin}$  ( $I_{OV} < 0$ ). The absolute sum of input overload currents on all pins may not exceed 50 mA. The supply voltages must remain within the specified limits. Proper operation under overload conditions depends on the application. Overload conditions must not occur on pin XTAL1 (powered by  $V_{DDIM}$ ).

**Electrical Parameters**

**Table 16 DC Characteristics for Upper Voltage Range (cont'd)**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Output High voltage <sup>7)</sup>	$V_{OH}$ CC	$V_{DDP} - 1.0$	—	—	V	$I_{OH} \geq I_{OHmax}$
		$V_{DDP} - 0.4$	—	—	V	$I_{OH} \geq I_{OHnom}$ <sup>8)</sup>
Output Low Voltage <sup>7)</sup>	$V_{OL}$ CC	—	—	0.4	V	$I_{OL} \leq I_{OLnom}$ <sup>8)</sup>
		—	—	1.0	V	$I_{OL} \leq I_{OLmax}$

- 1) Because each double bond pin is connected to two pads (standard pad and high-speed pad), it has twice the normal value. For a list of affected pins refer to the pin definitions table in chapter 2.
- 2) Not subject to production test - verified by design/characterization. Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It cannot suppress switching due to external system noise under all conditions.
- 3) If the input voltage exceeds the respective supply voltage due to ground bouncing ( $V_{IN} < V_{SS}$ ) or supply ripple ( $V_{IN} > V_{DDP}$ ), a certain amount of current may flow through the protection diodes. This current adds to the leakage current. An additional error current ( $I_{INJ}$ ) will flow if an overload current flows through an adjacent pin. Please refer to the definition of the overload coupling factor  $K_{OV}$ .
- 4) The given values are worst-case values. In production test, this leakage current is only tested at 125 °C; other values are ensured by correlation. For derating, please refer to the following descriptions: Leakage derating depending on temperature ( $T_J$  = junction temperature [°C]):  $I_{OZ} = 0.05 \times e^{(1.5 + 0.028 \times T_J)}$  [μA]. For example, at a temperature of 95 °C the resulting leakage current is 3.2 μA. Leakage derating depending on voltage level ( $DV = V_{DDP} - V_{PIN}$  [V]):  $I_{OZ} = I_{OZtempmax} - (1.6 \times DV)$  (μA). This voltage derating formula is an approximation which applies for maximum temperature.
- 5) Drive the indicated minimum current through this pin to change the default pin level driven by the enabled pull device.
- 6) Limit the current through this pin to the indicated value so that the enabled pull device can keep the default pin level.
- 7) The maximum deliverable output current of a port driver depends on the selected output driver mode. This specification is not valid for outputs which are switched to open drain mode. In this case the respective output will float and the voltage is determined by the external circuit.
- 8) As a rule, with decreasing output current the output levels approach the respective supply level ( $V_{OL} \rightarrow V_{SS}$ ,  $V_{OH} \rightarrow V_{DDP}$ ). However, only the levels for nominal output currents are verified.

**Electrical Parameters**

**Table 17 DC Characteristics for Lower Voltage Range (cont'd)**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Output High voltage <sup>7)</sup>	$V_{OH}$ CC	$V_{DDP} - 1.0$	—	—	V	$I_{OH} \geq I_{OHmax}$
		$V_{DDP} - 0.4$	—	—	V	$I_{OH} \geq I_{OHnom}$ <sup>8)</sup>
Output Low Voltage <sup>7)</sup>	$V_{OL}$ CC	—	—	0.4	V	$I_{OL} \leq I_{OLnom}$ <sup>8)</sup>
		—	—	1.0	V	$I_{OL} \leq I_{OLmax}$

- 1) Because each double bond pin is connected to two pads (standard pad and high-speed pad), it has twice the normal value. For a list of affected pins refer to the pin definitions table in chapter 2.
- 2) Not subject to production test - verified by design/characterization. Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It cannot suppress switching due to external system noise under all conditions.
- 3) If the input voltage exceeds the respective supply voltage due to ground bouncing ( $V_{IN} < V_{SS}$ ) or supply ripple ( $V_{IN} > V_{DDP}$ ), a certain amount of current may flow through the protection diodes. This current adds to the leakage current. An additional error current ( $I_{INJ}$ ) will flow if an overload current flows through an adjacent pin. Please refer to the definition of the overload coupling factor  $K_{OV}$ .
- 4) The given values are worst-case values. In production test, this leakage current is only tested at 125 °C; other values are ensured by correlation. For derating, please refer to the following descriptions: Leakage derating depending on temperature ( $T_J$  = junction temperature [°C]):  $I_{OZ} = 0.05 \times e^{(1.5 + 0.028 \times T_J)}$  [μA]. For example, at a temperature of 95 °C the resulting leakage current is 3.2 μA. Leakage derating depending on voltage level ( $DV = V_{DDP} - V_{PIN}$  [V]):  $I_{OZ} = I_{OZtempmax} - (1.6 \times DV)$  (μA). This voltage derating formula is an approximation which applies for maximum temperature.
- 5) Drive the indicated minimum current through this pin to change the default pin level driven by the enabled pull device:  $V_{PIN} \leq V_{IL}$  for a pullup;  $V_{PIN} \geq V_{IH}$  for a pulldown.
- 6) Limit the current through this pin to the indicated value so that the enabled pull device can keep the default pin level:  $V_{PIN} \geq V_{IH}$  for a pullup;  $V_{PIN} \leq V_{IL}$  for a pulldown.
- 7) The maximum deliverable output current of a port driver depends on the selected output driver mode. This specification is not valid for outputs which are switched to open drain mode. In this case the respective output will float and the voltage is determined by the external circuit.
- 8) As a rule, with decreasing output current the output levels approach the respective supply level ( $V_{OL} \rightarrow V_{SS}$ ,  $V_{OH} \rightarrow V_{DDP}$ ). However, only the levels for nominal output currents are verified.

## Electrical Parameters

- 1) The short-term frequency deviation refers to a timeframe of a few hours and is measured relative to the current frequency at the beginning of the respective timeframe. This parameter is useful to determine a time span for re-triggering a LIN synchronization.
- 2) This parameter is tested for the fastest and the slowest selection. The medium selections are not subject to production test - verified by design/characterization
- 3)  $f_{WU}$  in MHz
- 4) This value includes a hysteresis of approximately 50 mV for rising voltage.
- 5)  $V_{LV}$  = selected SWD voltage level
- 6) The limit  $V_{LV} - 0.10$  V is valid for the OK1 level. The limit for the OK2 level is  $V_{LV} - 0.15$  V.

### Conditions for $t_{SPO}$ Timing Measurement

The time required for the transition from **Power-on** to **Base** mode is called  $t_{SPO}$ . It is measured under the following conditions:

Precondition: The pad supply is valid, i.e.  $V_{DDPB}$  is above 3.0V and remains above 3.0V even though the XC223xN is starting up. No debugger is attached.

Start condition: Power-on reset is removed ( $\overline{PORST} = 1$ ).

End condition: External pin toggle caused by first user instruction executed from FLASH after startup.

### Conditions for $t_{SSB}$ Timing Measurement

The time required for the transition from **Standby** to **Base** mode is called  $t_{SSB}$ . It is measured under the following conditions:

Precondition: The **Standby** mode has been entered using the procedure defined in the Programmer's Guide.

Start condition: Pin toggle on  $\overline{ESR}$  pin triggering the startup sequence.

End condition: External pin toggle caused by first user instruction executed from FLASH after startup.

### Conditions for $t_{SSO}$ Timing Measurement

The time required for the transition from **Stopover** to **Stopover Waked-Up** mode is called  $t_{SSO}$ . It is measured under the following conditions:

Precondition: The **Stopover** mode has been entered using the procedure defined in the Programmer's Guide.

Start condition: Pin toggle on  $\overline{ESR}$  pin triggering the startup sequence.

End condition: External pin toggle caused by first user instruction executed from PSRAM after startup.

**Electrical Parameters**

#### 4.6 Flash Memory Parameters

The XC223xN is delivered with all Flash sectors erased and with no protection installed. The data retention time of the XC223xN's Flash memory (i.e. the time after which stored data can still be retrieved) depends on the number of times the Flash memory has been erased and programmed.

*Note: These parameters are not subject to production test but verified by design and/or characterization.*

*Note: Operating Conditions apply.*

**Table 25 Flash Parameters**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Parallel Flash module program/erase limit depending on Flash read activity	$N_{PP}$ SR	–	–	2 <sup>1)</sup>		$N_{FL\_RD} \leq 1$
		–	–	1 <sup>2)</sup>		$N_{FL\_RD} > 1$
Flash erase endurance for security pages	$N_{SEC}$ SR	10	–	–	cycles	$t_{RET} \geq 20$ years
Flash wait states <sup>3)</sup>	$N_{WSFLASH}$ SR	1	–	–		$f_{SYS} \leq 8$ MHz
		2	–	–		$f_{SYS} \leq 13$ MHz
		3	–	–		$f_{SYS} \leq 17$ MHz
		4	–	–		$f_{SYS} > 17$ MHz
Erase time per sector/page	$t_{ER}$ CC	–	7 <sup>4)</sup>	8.0	ms	
Programming time per page	$t_{PR}$ CC	–	3 <sup>4)</sup>	3.5	ms	
Data retention time	$t_{RET}$ CC	20	–	–	years	$N_{ER} \leq 1,000$ cycles
Drain disturb limit	$N_{DD}$ SR	32	–	–	cycles	
Number of erase cycles	$N_{ER}$ SR	–	–	15.000	cycles	$t_{RET} \geq 5$ years; Valid for Flash module 1 (up to 64 kbytes)
		–	–	1.000	cycles	$t_{RET} \geq 20$ years

1) The unused Flash module(s) can be erased/programmed while code is executed and/or data is read from only one Flash module or from PSRAM. The Flash module that delivers code/data can, of course, not be erased/programmed.

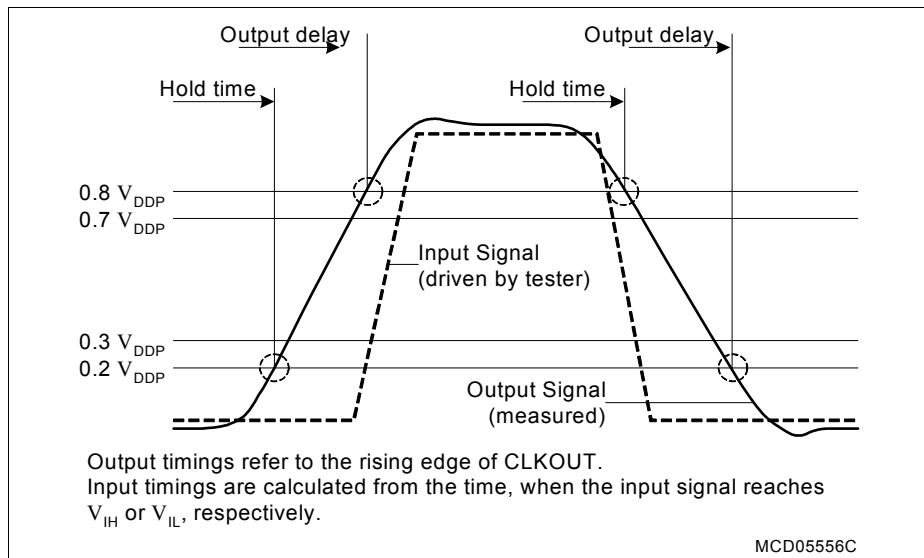


## 4.7 AC Parameters

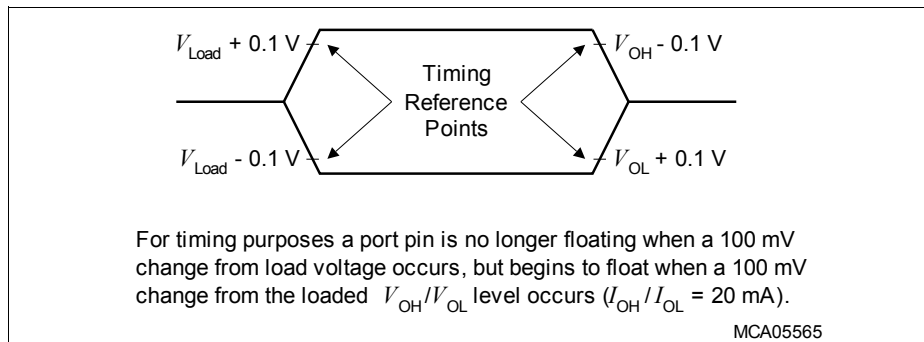
These parameters describe the dynamic behavior of the XC223xN.

### 4.7.1 Testing Waveforms

These values are used for characterization and production testing (except pin XTAL1).



**Figure 17 Input Output Waveforms**



**Figure 18 Floating Waveforms**

**Table 26      System PLL Parameters**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
VCO output frequency	$f_{VCO}$ CC	50	—	110	MHz	$VCOSEL=00b$ ; $VCOmode=$ controlled
		10	—	40	MHz	$VCOSEL=00b$ ; $VCOmode=$ free running
		100	—	160	MHz	$VCOSEL=01b$ ; $VCOmode=$ controlled
		20	—	80	MHz	$VCOSEL=01b$ ; $VCOmode=$ free running

#### **4.7.2.2      Wakeup Clock**

When wakeup operation is selected ( $SYSCON0.CLKSEL = 00_B$ ), the system clock is derived from the low-frequency wakeup clock source:

$$f_{SYS} = f_{WU}$$

In this mode, a basic functionality can be maintained without requiring an external clock source and while minimizing the power consumption.

#### **4.7.2.3      Selecting and Changing the Operating Frequency**

When selecting a clock source and the clock generation method, the required parameters must be carefully written to the respective bit fields, to avoid unintended intermediate states.

Many applications change the frequency of the system clock ( $f_{SYS}$ ) during operation in order to optimize system performance and power consumption. Changing the operating frequency also changes the switching currents, which influences the power supply.

To ensure proper operation of the on-chip EVRs while they generate the core voltage, the operating frequency shall only be changed in certain steps. This prevents overshoots and undershoots of the supply voltage.

To avoid the indicated problems, recommended sequences are provided which ensure the intended operation of the clock system interacting with the power system.

Please refer to the Programmer's Guide.

**Electrical Parameters**

**Table 28      Standard Pad Parameters for Upper Voltage Range (cont'd)**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Rise and Fall times (10% - 90%)	$t_{RF}$ CC	—	—	23 + 0.6 x $C_L$	ns	$C_L \geq 20$ pF; $C_L \leq 100$ pF; Driver_Strength = Medium
		—	—	11.6 + 0.22 x $C_L$	ns	$C_L \geq 20$ pF; $C_L \leq 100$ pF; Driver_Strength = Strong ; Driver_Edge= Medium
		—	—	4.2 + 0.14 x $C_L$	ns	$C_L \geq 20$ pF; $C_L \leq 100$ pF; Driver_Strength = Strong ; Driver_Edge= Sharp
		—	—	20.6 + 0.22 x $C_L$	ns	$C_L \geq 20$ pF; $C_L \leq 100$ pF; Driver_Strength = Strong ; Driver_Edge= Slow
		—	—	212 + 1.9 x $C_L$	ns	$C_L \geq 20$ pF; $C_L \leq 100$ pF; Driver_Strength = Weak

1) An output current above  $|I_{Oxnom}|$  may be drawn from up to three pins at the same time. For any group of 16 neighboring output pins, the total output current in each direction ( $\Sigma I_{OL}$  and  $\Sigma -I_{OH}$ ) must remain below 50 mA.

**Electrical Parameters**

**Table 29      Standard Pad Parameters for Lower Voltage Range**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Maximum output driver current (absolute value) <sup>1)</sup>	$I_{Omax}$ CC	–	–	2.5	mA	Driver_Strength = Medium
		–	–	10	mA	Driver_Strength = Strong
		–	–	0.5	mA	Driver_Strength = Weak
Nominal output driver current (absolute value)	$I_{Onom}$ CC	–	–	1.0	mA	Driver_Strength = Medium
		–	–	2.5	mA	Driver_Strength = Strong
		–	–	0.1	mA	Driver_Strength = Weak

**Electrical Parameters**

**Table 32 USIC SSC Slave Mode Timing for Upper Voltage Range (cont'd)**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Data input DX0 hold time from clock input DX1 receive edge <sup>1)</sup>	$t_{13}$ SR	5	—	—	ns	
Data output DOUT valid time	$t_{14}$ CC	7	—	33	ns	

1) These input timings are valid for asynchronous input signal handling of slave select input, shift clock input, and receive data input (bits DXnCR.DSEN = 0).

**Table 33** is valid under the following conditions:  $C_L = 20$  pF; SSC= slave ; voltage\_range= lower

**Table 33 USIC SSC Slave Mode Timing for Lower Voltage Range**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Select input DX2 setup to first clock input DX1 transmit edge <sup>1)</sup>	$t_{10}$ SR	7	—	—	ns	
Select input DX2 hold after last clock input DX1 receive edge <sup>1)</sup>	$t_{11}$ SR	7	—	—	ns	
Receive data input setup time to shift clock receive edge <sup>1)</sup>	$t_{12}$ SR	7	—	—	ns	
Data input DX0 hold time from clock input DX1 receive edge <sup>1)</sup>	$t_{13}$ SR	5	—	—	ns	
Data output DOUT valid time	$t_{14}$ CC	8	—	41	ns	

1) These input timings are valid for asynchronous input signal handling of slave select input, shift clock input, and receive data input (bits DXnCR.DSEN = 0).