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### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	33MHz
Connectivity	UART/USART
Peripherals	POR, WDT
Number of I/O	32
Program Memory Size	64KB (64K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 5.5V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	44-LCC (J-Lead)
Supplier Device Package	44-PLCC (16.59x16.59)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/p89c60x2ba-00-512">https://www.e-xfl.com/product-detail/nxp-semiconductors/p89c60x2ba-00-512</a>

## 80C51 8-bit Flash microcontroller family

64KB Flash, 512B/1024B RAM

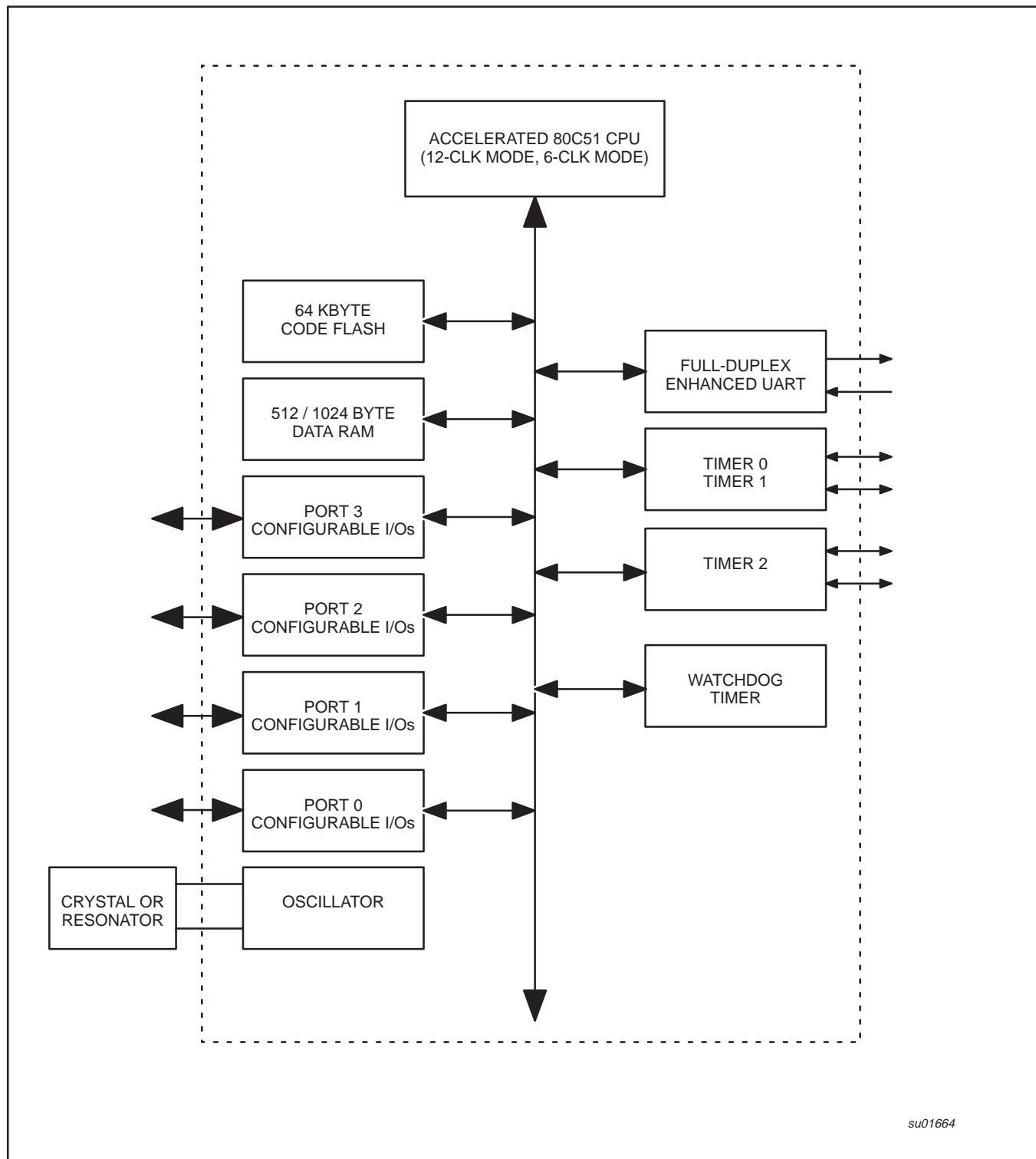
## P89C60X2/61X2

### FEATURES

- 80C51 Central Processing Unit
  - 64 kbytes Flash
  - 512 bytes RAM (P89C60X2)
  - 1024 bytes RAM (P89C61X2)
  - Boolean processor
  - Fully static operation
- In-System Programmable (ISP) Flash memory
- 12-clock operation with selectable 6-clock operation (via software or via parallel programmer)
- Memory addressing capability
  - Up to 64 kbytes ROM and 64 kbytes RAM
- Power control modes:
  - Clock can be stopped and resumed
  - Idle mode
  - Power-down mode
- Two speed ranges
  - 0 to 20 MHz with 6-clock operation
  - 0 to 33 MHz with 12-clock operation
- LQFP, PLCC, and DIP packages
- Dual Data Pointers
- Three security bits
- Four interrupt priority levels
- Six interrupt sources
- Four 8-bit I/O ports
- Full-duplex enhanced UART
  - Framing error detection
  - Automatic address recognition
- Three 16-bit timers/counters T0, T1 (standard 80C51) and additional T2 (capture and compare)
- Programmable clock-out pin
- Watchdog timer
- Asynchronous port reset
- Low EMI (inhibit ALE, 6-clock mode)
- Wake-up from Power Down by an external interrupt

80C51 8-bit Flash microcontroller family  
64KB Flash, 512B/1024B RAM

P89C60X2/61X2

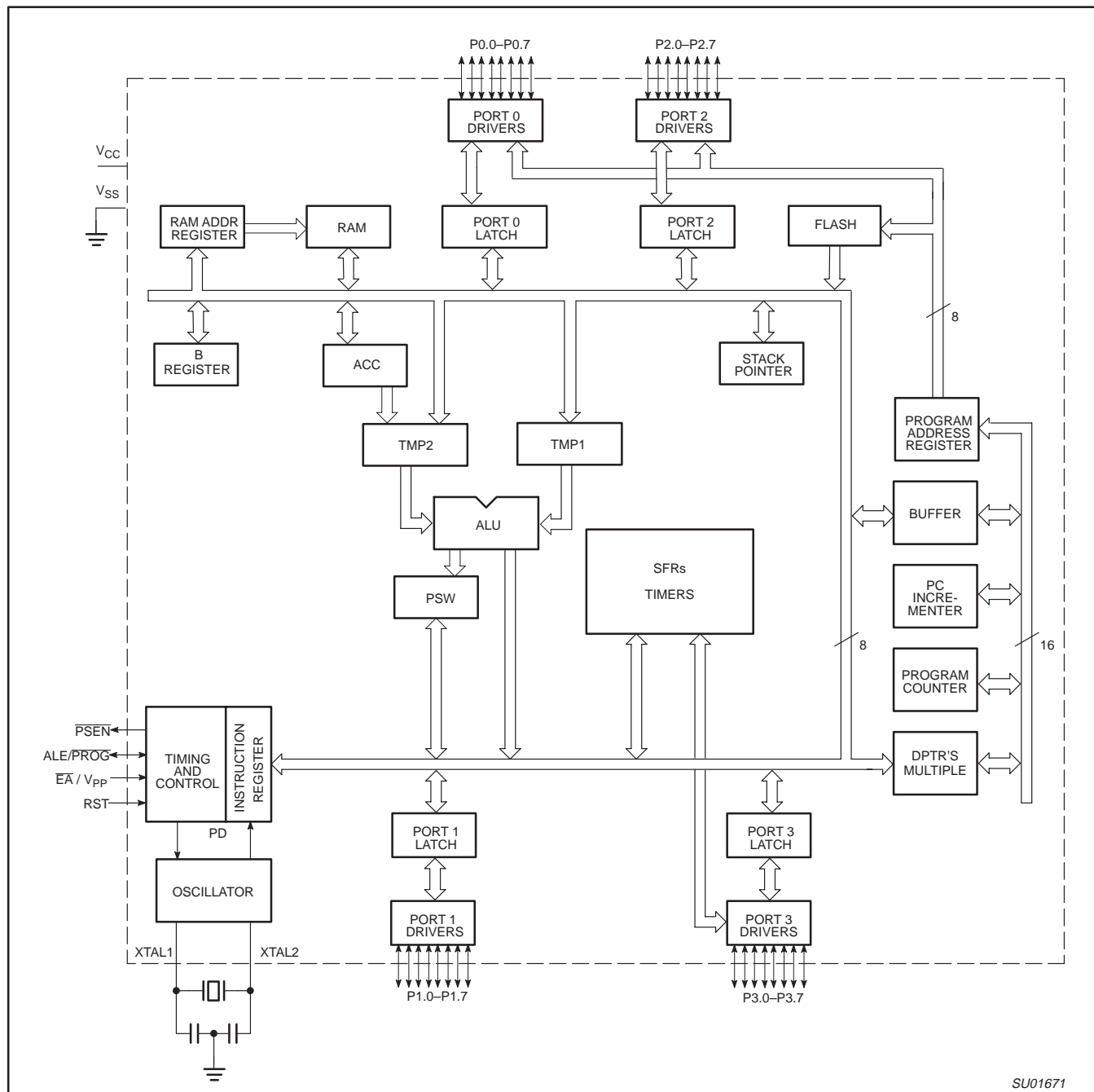
**BLOCK DIAGRAM 1**

# 80C51 8-bit Flash microcontroller family

64KB Flash, 512B/1024B RAM

P89C60X2/61X2

## BLOCK DIAGRAM 2 (CPU-ORIENTED)

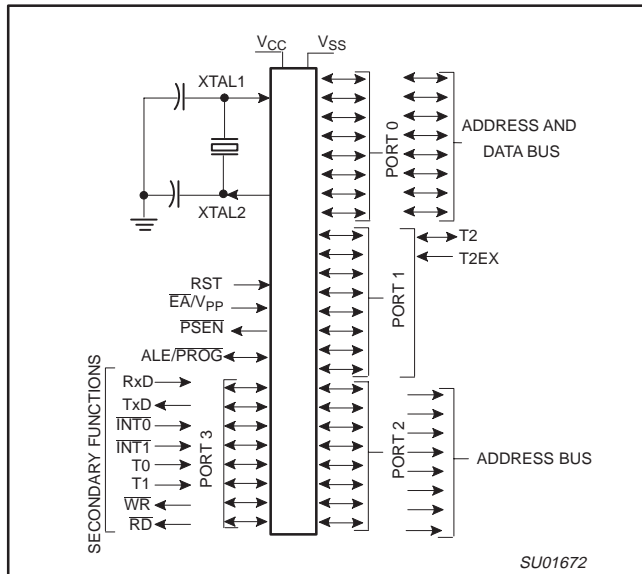


# 80C51 8-bit Flash microcontroller family

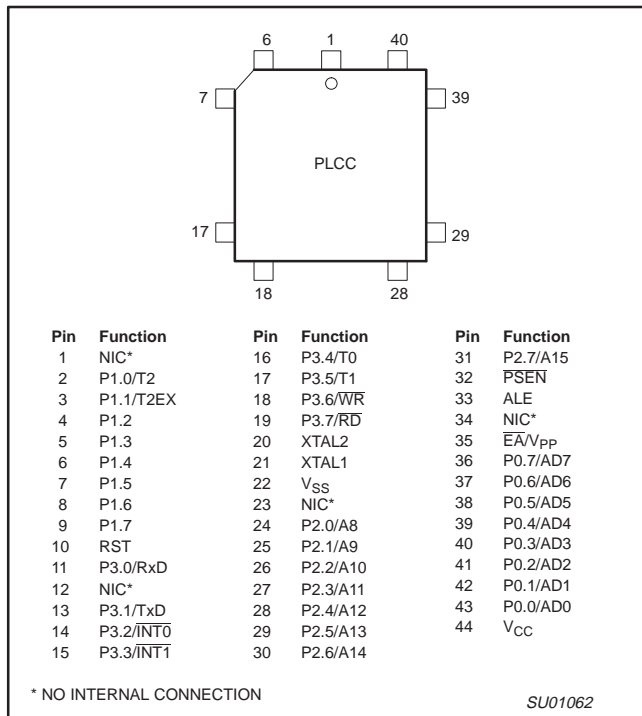
## 64KB Flash, 512B/1024B RAM

P89C60X2/61X2

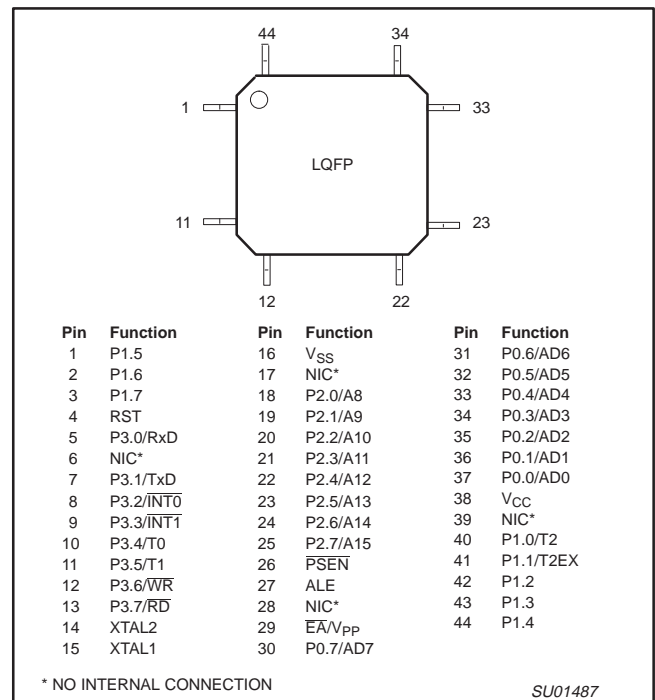
### LOGIC SYMBOL



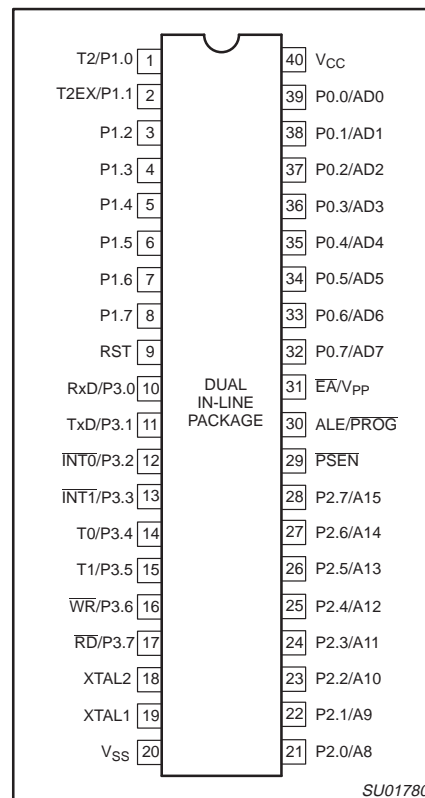
### PLASTIC LEADED CHIP CARRIER PIN FUNCTIONS



### LOW PROFILE QUAD FLAT PACK PIN FUNCTIONS



### PLASTIC DUAL IN-LINE PACKAGE PIN FUNCTIONS



# 80C51 8-bit Flash microcontroller family

## 64KB Flash, 512B/1024B RAM

P89C60X2/61X2

Table 1.

CLOCK MODE CONFIG BIT (FX2)	X2 bit in CKCON	DESCRIPTION
erased	0	12-clock mode (default)
erased	1	6-clock mode
programmed	x	6-clock mode

**NOTE:**

1. Default clock mode after ChipErase is set to 12-clock.

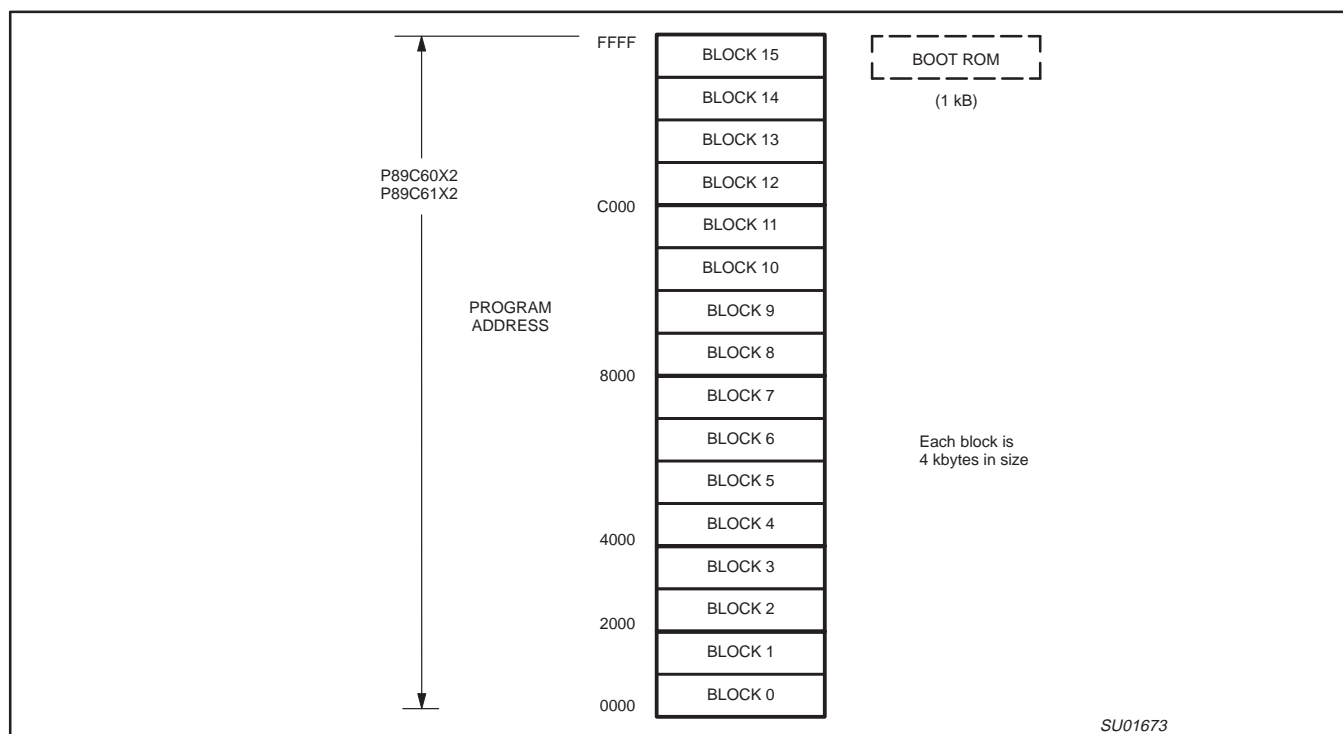


Figure 1. Flash Memory Configuration

### Power-On Reset Code Execution

The P89C60X2/61X2 contains a special Flash register, the STATUS BYTE. At the falling edge of reset, the P89C60X2/61X2 examines the contents of the Status Byte. If the Status Byte is set to zero, power-up execution starts at location 0000H, which is the normal start address of the user's application code. When the Status Byte is set to a value other than zero, the factory masked-ROM ISP boot loader is invoked. The factory default for the Status Byte is FFh. Once set to 00h, the Status Byte can only be changed back to FFh by a full-chip erase operation when using ISP.

### Hardware Activation of the Boot Loader

The boot loader can also be executed by holding  $\overline{\text{PSEN}}$  LOW,  $\overline{\text{EA}}$  greater than  $V_{IH}$  (such as +5 V), and ALE HIGH (or not connected) at the falling edge of RESET. This is the same effect as having a non-zero status byte. This allows an application to be built that will normally execute the end user's code but can be manually forced into ISP operation.

After programming the Flash, the status byte should be programmed to zero in order to allow execution of the user's application code beginning at address 0000H.

**80C51 8-bit Flash microcontroller family**  
 64KB Flash, 512B/1024B RAM

**P89C60X2/61X2**

**Table 2. Intel-Hex Records Used by In-System Programming**

RECORD TYPE	COMMAND/DATA FUNCTION
00	<p>Program Data            :nnaaaa0dd...ddcc</p> <p>Where:            nn = number of bytes (hex) in record            aaaa = memory address of first byte in record            dd...dd = data bytes            cc = checksum</p> <p>Example:            :10008000AF5F67F0602703E0322CFA92007780C3FD</p>
01	<p>End of File (EOF), no operation            :xxxxxx0lcc</p> <p>Where:            xxxxxx = required field, but value is a "don't care"            cc = checksum</p> <p>Example:            :00000001FF</p>
03	<p>Miscellaneous Write Functions            :nnxxxx03ffssddcc</p> <p>Where:            nn = number of bytes (hex) in record            xxxxx = required field, but value is a "don't care"            03 = Write Function            ff = subfunction code            ss = selection code            dd = data input (as needed)            cc = checksum</p> <p>Subfunction Code = 04 (Set Status Byte to 00h)            ff = 04            ss = don't care</p> <p>Example:            :020000030400F7 set status byte to 00h (device executes user code after Reset)</p> <p>Subfunction Code = 05 (Program Security Bits)            ff = 05            ss = 00 program security bit 1 (inhibit writing to Flash)                  01 program security bit 2 (inhibit Flash verify)                  02 program security bit 3 (disable external memory)</p> <p>Example:            :020000030501F5 program security bit 2</p> <p>Subfunction Code = 06 (Program Flash X2 bit)            ff = 06            ss = 02 program FX2 bit (dd = 80) ⇒ 6-clk. mode enabled            dd = data</p> <p>Example 1:            :0300000306028072 program FX2 bit (enable 6-clk. mode)</p>

# 80C51 8-bit Flash microcontroller family

## 64KB Flash, 512B/1024B RAM

P89C60X2/61X2

RECORD TYPE	COMMAND/DATA FUNCTION														
03 (cont.)	<p>Subfunction Code = 07 (Full Chip Erase) Erases all blocks, security bits, and sets status byte to default values ff = 07 ss = don't care dd = don't care Example: :0100000307F5 full chip erase</p> <p>Subfunction Code = 0C (Erase 4k blocks) ff = 0C ss = block code as shown below: block 0, 0k ~ 4k, 00H block 1, 4k ~ 8k, 10H block 2, 8k ~ 12k, 20H block 3, 12k ~ 16k, 30H block 4, 16k ~ 20k, 40H block 5, 20k ~ 24k, 50H block 6, 24k ~ 28k, 60H block 7, 28k ~ 32k, 70H block 8, 32k ~ 36k, 80H block 9, 36k ~ 40k, 90H block 10, 40k ~ 44k, A0H block 11, 44k ~ 48k, B0H block 12, 48k ~ 52k, C0H block 13, 52k ~ 56k, D0H block 14, 56k ~ 60k, E0H block 15, 60k ~ 64k, F0H</p> <p>Example: :020000030C20CF erase 4k block 2</p>														
04	<p>Display Device Data or Blank Check – Record type 04 causes the contents of the entire Flash array to be sent out the serial port in a formatted display. This display consists of an address and the contents of 16 bytes starting with that address. No display of the device contents will occur if security bit 2 has been programmed. Data to the serial port is initiated by the reception of any character and terminated by the reception of any character.</p> <p>General Format of Function 04 :05xxxx04sssseeeffcc</p> <p>Where:</p> <table><tr><td>05</td><td>= number of bytes (hex) in record</td></tr><tr><td>xxxx</td><td>= required field, but value is a "don't care"</td></tr><tr><td>04</td><td>= "Display Device Data or Blank Check" function code</td></tr><tr><td>ssss</td><td>= starting address</td></tr><tr><td>eeee</td><td>= ending address</td></tr><tr><td>ff</td><td>= subfunction 00 = display data 01 = blank check 02 = display data in data block (valid addresses: 0001 ~ 0FFFH)</td></tr><tr><td>cc</td><td>= checksum</td></tr></table> <p>Example 1: :0500000440004FFF0069 display 4000-4FFF</p> <p>Example 2: :0500000400000FFF02E7 display data in data block (the data at address 0000 is invalid)</p>	05	= number of bytes (hex) in record	xxxx	= required field, but value is a "don't care"	04	= "Display Device Data or Blank Check" function code	ssss	= starting address	eeee	= ending address	ff	= subfunction 00 = display data 01 = blank check 02 = display data in data block (valid addresses: 0001 ~ 0FFFH)	cc	= checksum
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**80C51 8-bit Flash microcontroller family**  
64KB Flash, 512B/1024B RAM

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**P89C60X2/61X2****Security**

The security feature protects against software piracy and prevents the contents of the FLASH from being read. The Security Lock bits are located in FLASH. The P89C60X2/61X2 has 3 programmable security lock bits that will provide different levels of protection for the on-chip code and data (see Table 3). Unlike the ROM and OTP versions, the security lock bits are independent. LB3 includes the security protection of LB1.

**Table 3.**

<b>SECURITY LOCK BITS<sup>1</sup></b>	<b>PROTECTION DESCRIPTION</b>
<b>Level</b>	
LB1	MOVC instructions executed from external program memory are disabled from fetching code bytes from internal memory.
LB2	Program verification is disabled
LB3	External execution is disabled.

**NOTE:**

1. The security lock bits are independent.

## 80C51 8-bit Flash microcontroller family

### 64KB Flash, 512B/1024B RAM

P89C60X2/61X2

### OSCILLATOR CHARACTERISTICS

Using the oscillator, XTAL1 and XTAL2 are the input and output, respectively, of an inverting amplifier. The pins can be configured for use as an on-chip oscillator, as shown in the logic symbol.

To drive the device from an external clock source, XTAL1 should be driven while XTAL2 is left unconnected. However, minimum and maximum high and low times specified in the data sheet must be observed.

### Clock Control Register (CKCON)

This device provides control of the 6-clock/12-clock mode by both an SFR bit (bit X2 in register CKCON) and a Flash bit (bit FX2, located in the Security Block). When X2 is 0, 12-clock mode is activated. By setting this bit to 1, the system is switching to 6-clock mode. Having this option implemented as SFR bit, it can be accessed anytime and changed to either value. Changing X2 from 0 to 1 will result in executing user code at twice the speed, since all system time intervals will be divided by 2. Changing back from 6-clock to 12-clock mode will slow down running code by a factor of 2.

The Flash clock control bit (FX2) activates the 6-clock mode when programmed using a parallel programmer, superceding the X2 bit (CKCON.0). Please also see Table 4 below.

**Table 4.**

FX2 clock mode bit (can only be set by parallel programmer)	X2 bit (CKCON.0)	CPU clock mode
erased	0	12-clock mode (default)
erased	1	6-clock mode
programmed	X	6-clock mode

### Programmable Clock-Out Pin

A 50% duty cycle clock can be programmed to be output on P1.0. This pin, besides being a regular I/O pin, has two alternate functions. It can be programmed:

1. to input the external clock for Timer/Counter 2, or
2. to output a 50% duty cycle clock ranging from 61 Hz to 4 MHz at a 16 MHz operating frequency in 12-clock mode (122 Hz to 8 MHz in 6-clock mode).

To configure the Timer/Counter 2 as a clock generator, bit C/T2 (in T2CON) must be cleared and bit T2OE in T2MOD must be set. Bit TR2 (T2CON.2) also must be set to start the timer.

The Clock-Out frequency depends on the oscillator frequency and the reload value of Timer 2 capture registers (RCAP2H, RCAP2L) as shown in this equation:

$$\frac{\text{Oscillator Frequency}}{n \times (65536 - \text{RCAP2H}, \text{RCAP2L})}$$

Where:

$n = 2$  in 6-clock mode, 4 in 12-clock mode.

(RCAP2H, RCAP2L) = the content of RCAP2H and RCAP2L taken as a 16-bit unsigned integer.

In the Clock-Out mode Timer 2 roll-overs will not generate an interrupt. This is similar to when it is used as a baud-rate generator. It is possible to use Timer 2 as a baud-rate generator and a clock generator simultaneously. Note, however, that the baud-rate and the Clock-Out frequency will be the same.

### RESET

A reset is accomplished by holding the RST pin HIGH for at least two machine cycles (24 oscillator periods in 12-clock and 12 oscillator periods in 6-clock mode), while the oscillator is running. To insure a reliable power-up reset, the RST pin must be high long enough to allow the oscillator time to start up (normally a few milliseconds) plus two machine cycles, unless it has been set to 6-clock operation using a parallel programmer.

### LOW POWER MODES

#### Stop Clock Mode

The static design enables the clock speed to be reduced down to 0 MHz (stopped). When the oscillator is stopped, the RAM and Special Function Registers retain their values. This mode allows step-by-step utilization and permits reduced system power consumption by lowering the clock frequency down to any value. For lowest power consumption the Power Down mode is suggested.

#### Idle Mode

In idle mode (see Table 5), the CPU puts itself to sleep while all of the on-chip peripherals stay active. The instruction to invoke the idle mode is the last instruction executed in the normal operating mode before the idle mode is activated. The CPU contents, the on-chip RAM, and all of the special function registers remain intact during this mode. The idle mode can be terminated either by any enabled interrupt (at which time the process is picked up at the interrupt service routine and continued), or by a hardware reset which starts the processor in the same manner as a power-on reset.

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64KB Flash, 512B/1024B RAM

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Mode 3 is provided for applications requiring an extra 8-bit timer on the counter. With Timer 0 in Mode 3, an 80C51 can look like it has three Timer/Counters. When Timer 0 is in Mode 3, Timer 1 can be

turned on and off by switching it out of and into its own Mode 3, or can still be used by the serial port as a baud rate generator, or in fact, in any application not requiring an interrupt.

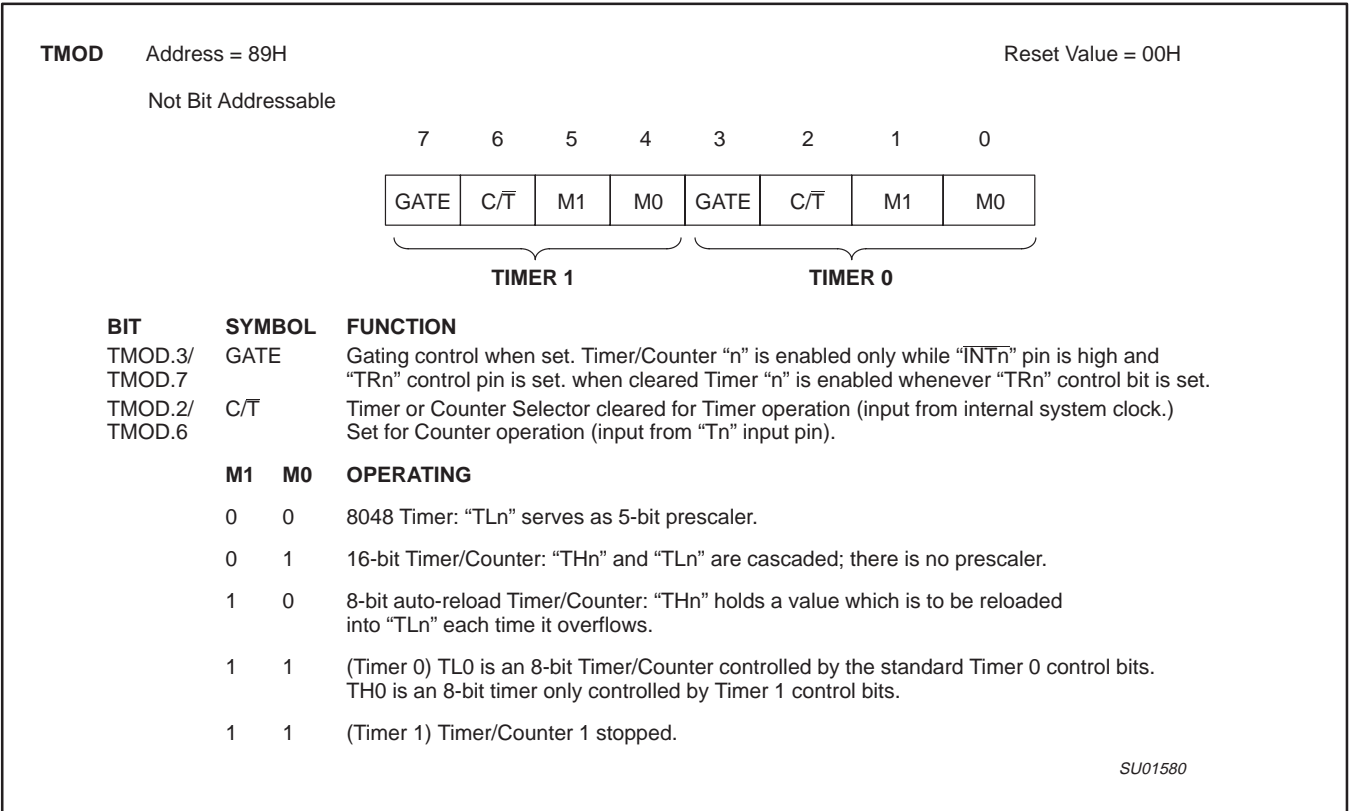


Figure 3. Timer/Counter 0/1 Mode Control (TMOD) Register

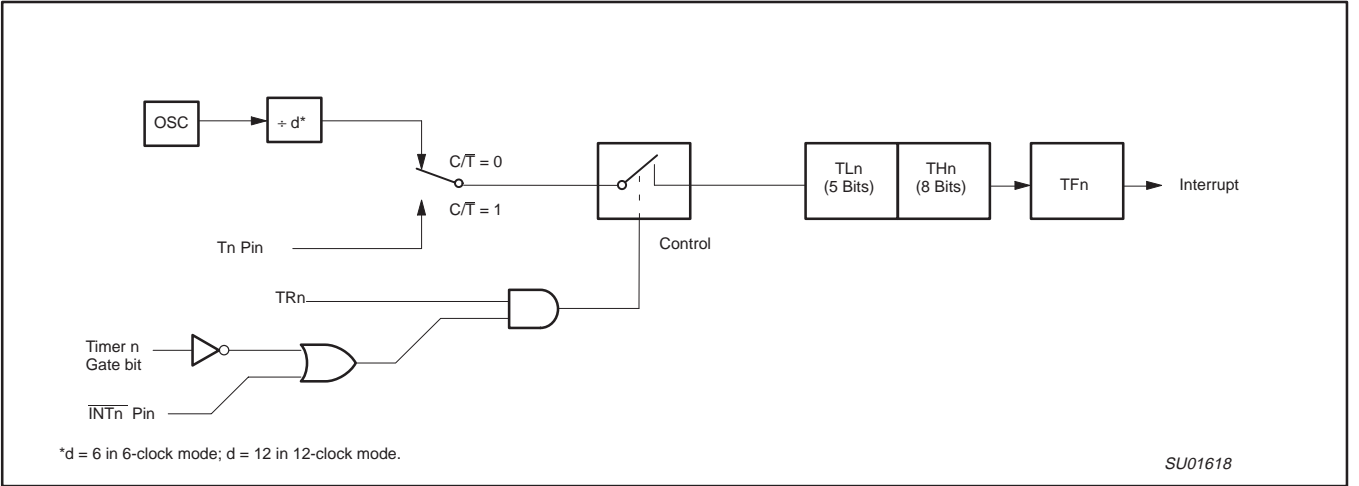


Figure 4. Timer/Counter 0/1 Mode 0: 13-Bit Timer/Counter

# 80C51 8-bit Flash microcontroller family

## 64KB Flash, 512B/1024B RAM

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**Table 6. Timer 2 Operating Modes**

RCLK + TCLK	CP/RL2	TR2	MODE
0	0	1	16-bit Auto-reload
0	1	1	16-bit Capture
1	X	1	Baud rate generator
X	X	0	(off)

T2CON

Address = C8H

Reset Value = 00H

Bit Addressable

7

6

5

4

3

2

1

0

TF2

EXF2

RCLK

TCLK

EXEN2

TR2

C/T2

CP/RL2

Symbol	Position	Name and Significance
TF2	T2CON.7	Timer 2 overflow flag set by a Timer 2 overflow and must be cleared by software. TF2 will not be set when either RCLK or TCLK = 1.
EXF2	T2CON.6	Timer 2 external flag set when either a capture or reload is caused by a negative transition on T2EX and EXEN2 = 1. When Timer 2 interrupt is enabled, EXF2 = 1 will cause the CPU to vector to the Timer 2 interrupt routine. EXF2 must be cleared by software. EXF2 does not cause an interrupt in up/down counter mode (DCEN = 1).
RCLK	T2CON.5	Receive clock flag. When set, causes the serial port to use Timer 2 overflow pulses for its receive clock in modes 1 and 3. RCLK = 0 causes Timer 1 overflow to be used for the receive clock.
TCLK	T2CON.4	Transmit clock flag. When set, causes the serial port to use Timer 2 overflow pulses for its transmit clock in modes 1 and 3. TCLK = 0 causes Timer 1 overflows to be used for the transmit clock.
EXEN2	T2CON.3	Timer 2 external enable flag. When set, allows a capture or reload to occur as a result of a negative transition on T2EX if Timer 2 is not being used to clock the serial port. EXEN2 = 0 causes Timer 2 to ignore events at T2EX.
TR2	T2CON.2	Start/stop control for Timer 2. A logic 1 starts the timer.
C/T2	T2CON.1	Timer or counter select. (Timer 2) 0 = Internal timer (OSC/12 in 12-clock mode or OSC/6 in 6-clock mode) 1 = External event counter (falling edge triggered).
CP/RL2	T2CON.0	Capture/Reload flag. When set, captures will occur on negative transitions at T2EX if EXEN2 = 1. When cleared, auto-reloads will occur either with Timer 2 overflows or negative transitions at T2EX when EXEN2 = 1. When either RCLK = 1 or TCLK = 1, this bit is ignored and the timer is forced to auto-reload on Timer 2 overflow.

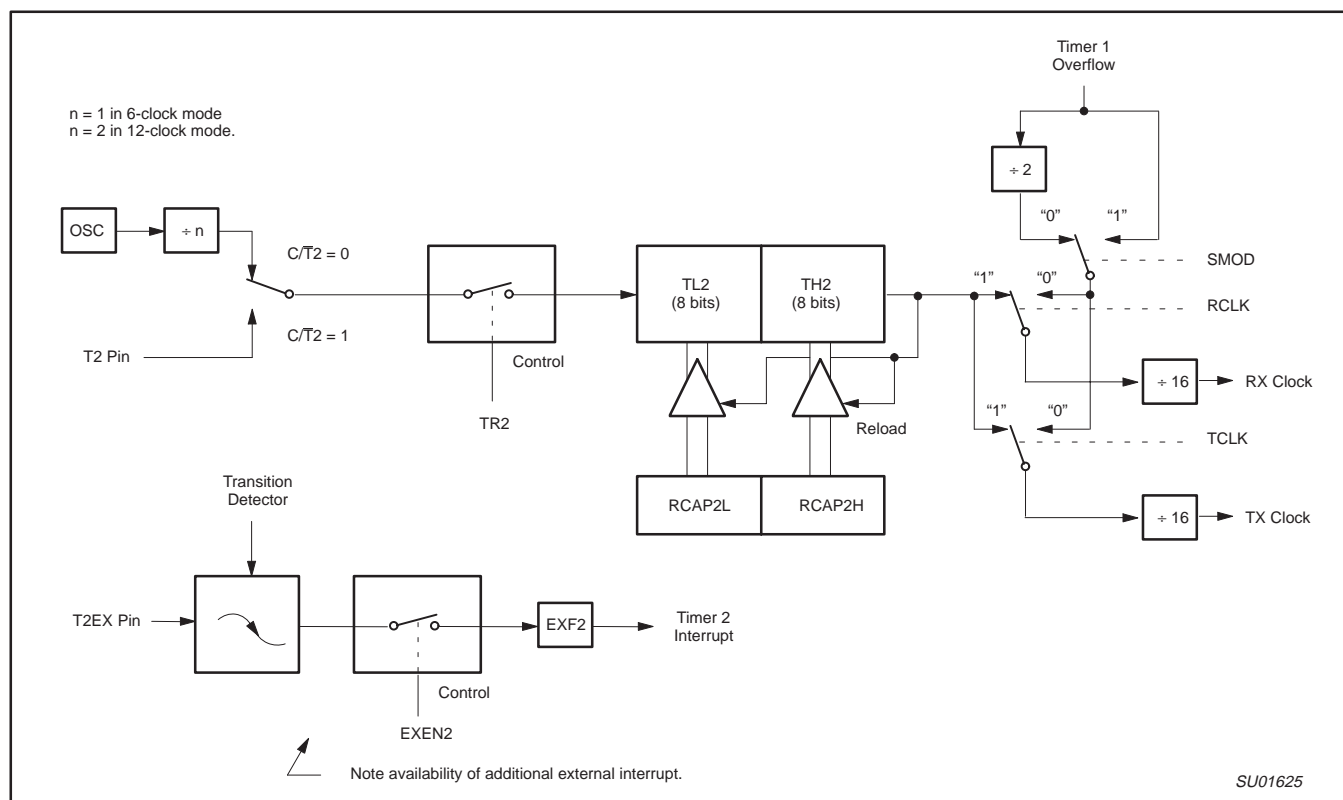
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**Figure 8. Timer/Counter 2 (T2CON) Control Register**

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**Figure 13. Timer 2 in Baud Rate Generator Mode**

## Baud Rate Generator Mode

Bits TCLK and/or RCLK in T2CON (Table 6) allow the serial port transmit and receive baud rates to be derived from either Timer 1 or Timer 2. When TCLK= 0, Timer 1 is used as the serial port transmit baud rate generator. When TCLK= 1, Timer 2 is used as the serial port transmit baud rate generator. RCLK has the same effect for the serial port receive baud rate. With these two bits, the serial port can have different receive and transmit baud rates – one generated by Timer 1, the other by Timer 2.

Figure 13 shows the Timer 2 in baud rate generation mode. The baud rate generation mode is like the auto-reload mode, in that a rollover in TH2 causes the Timer 2 registers to be reloaded with the 16-bit value in registers RCAP2H and RCAP2L, which are preset by software.

The baud rates in modes 1 and 3 are determined by Timer 2's overflow rate given below:

$$\text{Modes 1 and 3 Baud Rates} = \frac{\text{Timer 2 Overflow Rate}}{16}$$

The timer can be configured for either “timer” or “counter” operation. In many applications, it is configured for “timer” operation ( $C/T2=0$ ). Timer operation is different for Timer 2 when it is being used as a baud rate generator.

Usually, as a timer it would increment every machine cycle (i.e., 1/6 the oscillator frequency in 6-clock mode or 1/12 the oscillator frequency in 12-clock mode). As a baud rate generator, it increments at the oscillator frequency in 6-clock mode or at 1/2 the oscillator frequency in 12-clock mode. Thus the baud rate formula is as follows:

Modes 1 and 3 Baud Rates =

$$\frac{\text{Oscillator Frequency}}{[n \times [65536 - (\text{RCAP2H}, \text{RCAP2L})]]}$$

Where:

n = 16 in 6-clock mode, 32 in 12-clock mode.

(RCAP2H, RCAP2L)= The content of RCAP2H and RCAP2L taken as a 16-bit unsigned integer.

The Timer 2 as a baud rate generator mode shown in Figure 13 is valid only if RCLK and/or TCLK = 1 in T2CON register. Note that a rollover in TH2 does not set TF2, and will not generate an interrupt. Thus, the Timer 2 interrupt does not have to be disabled when Timer 2 is in the baud rate generator mode. Also if the EXEN2 (T2 external enable flag) is set, a 1-to-0 transition in T2EX (Timer/counter 2 trigger input) will set EXF2 (T2 external flag) but will not cause a reload from (RCAP2H, RCAP2L) to (TH2, TL2). Therefore when Timer 2 is in use as a baud rate generator, T2EX can be used as an additional external interrupt, if needed.

When Timer 2 is in the baud rate generator mode, one should not try to read or write TH2 and TL2. As a baud rate generator, Timer 2 is incremented every state time ( $\text{osc}/2$ ) or asynchronously from pin T2; under these conditions, a read or write of TH2 or TL2 may not be accurate. The RCAP2 registers may be read, but should not be written to, because a write might overlap a reload and cause write and/or reload errors. The timer should be turned off (clear TR2) before accessing the Timer 2 or RCAP2 registers.

Table 7 shows commonly used baud rates and how they can be obtained from Timer 2.

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### 64KB Flash, 512B/1024B RAM

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## FULL-DUPLEX ENHANCED UART

### Standard UART operation

The serial port is full duplex, meaning it can transmit and receive simultaneously. It is also receive-buffered, meaning it can commence reception of a second byte before a previously received byte has been read from the register. (However, if the first byte still hasn't been read by the time reception of the second byte is complete, one of the bytes will be lost.) The serial port receive and transmit registers are both accessed at Special Function Register SBUF. Writing to SBUF loads the transmit register, and reading SBUF accesses a physically separate receive register.

The serial port can operate in 4 modes:

- Mode 0:** Serial data enters and exits through RxD. TxD outputs the shift clock. 8 bits are transmitted/received (LSB first). The baud rate is fixed at 1/12 the oscillator frequency (in 12-clock mode) or 1/6 the oscillator frequency (in 6-clock mode).
- Mode 1:** 10 bits are transmitted (through TxD) or received (through RxD): a start bit (0), 8 data bits (LSB first), and a stop bit (1). On receive, the stop bit goes into RB8 in Special Function Register SCON. The baud rate is variable.
- Mode 2:** 11 bits are transmitted (through TxD) or received (through RxD): start bit (0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (1). On Transmit, the 9th data bit (TB8 in SCON) can be assigned the value of 0 or 1. Or, for example, the parity bit (P, in the PSW) could be moved into TB8. On receive, the 9th data bit goes into RB8 in Special Function Register SCON, while the stop bit is ignored. The baud rate is programmable to either 1/32 or 1/64 the oscillator frequency (in 12-clock mode) or 1/16 or 1/32 the oscillator frequency (in 6-clock mode).
- Mode 3:** 11 bits are transmitted (through TxD) or received (through RxD): a start bit (0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (1). In fact, Mode 3 is the same as Mode 2 in all respects except baud rate. The baud rate in Mode 3 is variable.

In all four modes, transmission is initiated by any instruction that uses SBUF as a destination register. Reception is initiated in Mode 0 by the condition RI = 0 and REN = 1. Reception is initiated in the other modes by the incoming start bit if REN = 1.

### Multiprocessor Communications

Modes 2 and 3 have a special provision for multiprocessor communications. In these modes, 9 data bits are received. The 9th one goes into RB8. Then comes a stop bit. The port can be programmed such that when the stop bit is received, the serial port interrupt will be activated only if RB8 = 1. This feature is enabled by setting bit SM2 in SCON. A way to use this feature in multiprocessor systems is as follows:

When the master processor wants to transmit a block of data to one of several slaves, it first sends out an address byte which identifies the target slave. An address byte differs from a data byte in that the 9th bit is 1 in an address byte and 0 in a data byte. With SM2 = 1, no slave will be interrupted by a data byte. An address byte, however, will interrupt all slaves, so that each slave can examine the received byte and see if it is being addressed. The addressed slave will clear its SM2 bit and prepare to receive the data bytes that will be coming.

The slaves that weren't being addressed leave their SM2s set and go on about their business, ignoring the coming data bytes.

SM2 has no effect in Mode 0, and in Mode 1 can be used to check the validity of the stop bit. In a Mode 1 reception, if SM2 = 1, the receive interrupt will not be activated unless a valid stop bit is received.

### Serial Port Control Register

The serial port control and status register is the Special Function Register SCON, shown in Figure 14. This register contains not only the mode selection bits, but also the 9th data bit for transmit and receive (TB8 and RB8), and the serial port interrupt bits (TI and RI).

### Baud Rates

The baud rate in Mode 0 is fixed: Mode 0 Baud Rate = Oscillator Frequency / 12 (in 12-clock mode) or / 6 (in 6-clock mode). The baud rate in Mode 2 depends on the value of bit SMOD in Special Function Register PCON. If SMOD = 0 (which is the value on reset), and the port pins in 12-clock mode, the baud rate is 1/64 the oscillator frequency. If SMOD = 1, the baud rate is 1/32 the oscillator frequency. In 6-clock mode, the baud rate is 1/32 or 1/16 the oscillator frequency, respectively.

Mode 2 Baud Rate =

$$\frac{2^{\text{SMOD}}}{n} \times (\text{Oscillator Frequency})$$

Where:

$$n = 64 \text{ in 12-clock mode, } 32 \text{ in 6-clock mode}$$

The baud rates in Modes 1 and 3 are determined by the Timer 1 or Timer 2 overflow rate.

### Using Timer 1 to Generate Baud Rates

When Timer 1 is used as the baud rate generator (T2CON.RCLK = 0, T2CON.TCLK = 0), the baud rates in Modes 1 and 3 are determined by the Timer 1 overflow rate and the value of SMOD as follows:

Mode 1, 3 Baud Rate =

$$\frac{2^{\text{SMOD}}}{n} \times (\text{Timer 1 Overflow Rate})$$

Where:

$$n = 32 \text{ in 12-clock mode, } 16 \text{ in 6-clock mode}$$

The Timer 1 interrupt should be disabled in this application. The Timer itself can be configured for either "timer" or "counter" operation, and in any of its 3 running modes. In the most typical applications, it is configured for "timer" operation, in the auto-reload mode (high nibble of TMOD = 0010B). In that case the baud rate is given by the formula:

Mode 1, 3 Baud Rate =

$$\frac{2^{\text{SMOD}}}{n} \times \frac{\text{Oscillator Frequency}}{12 \times [256 - (\text{TH1})]}$$

Where:

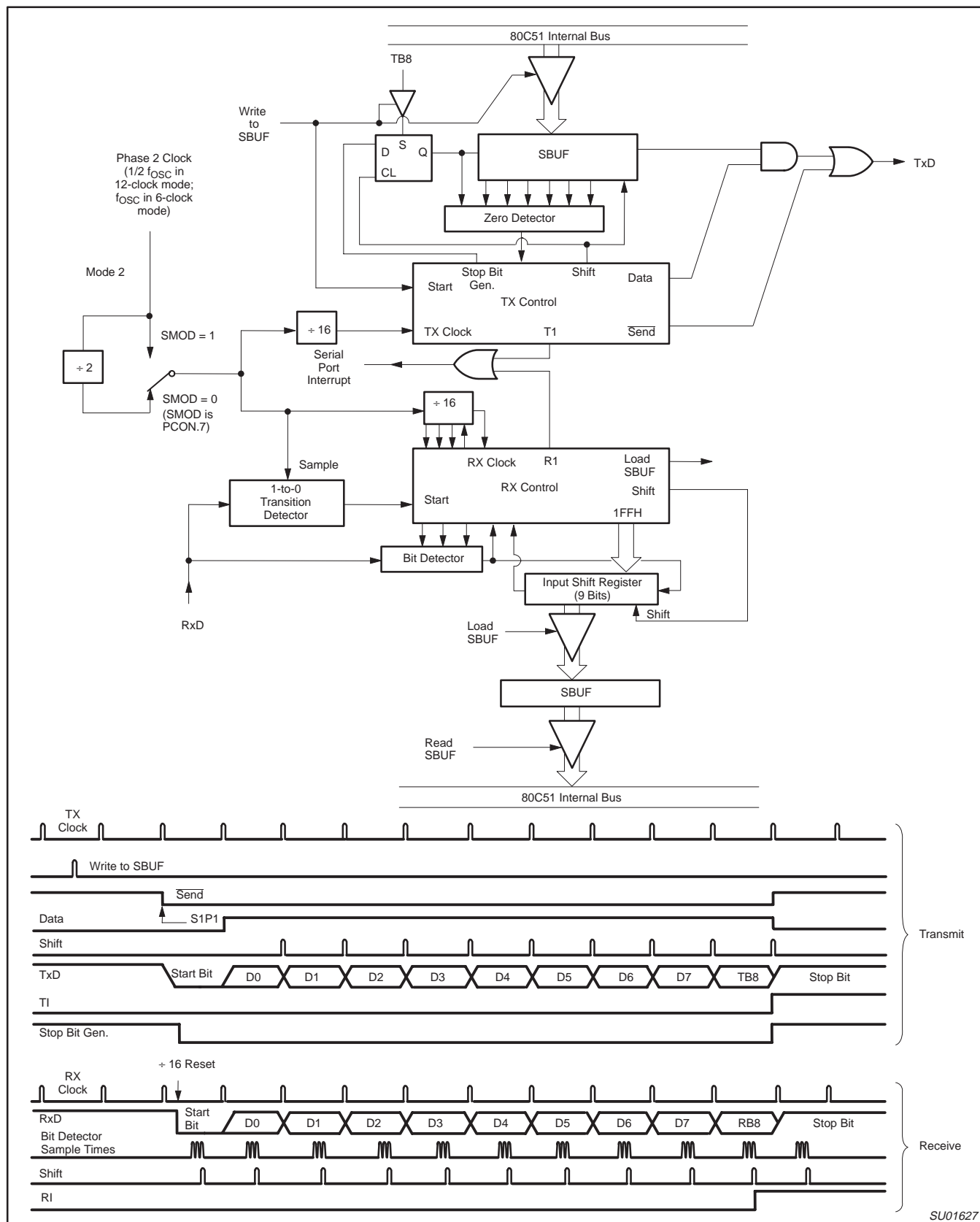
$$n = 32 \text{ in 12-clock mode, } 16 \text{ in 6-clock mode}$$

One can achieve very low baud rates with Timer 1 by leaving the Timer 1 interrupt enabled, and configuring the Timer to run as a 16-bit timer (high nibble of TMOD = 0001B), and using the Timer 1 interrupt to do a 16-bit software reload. Figure 15 lists various commonly used baud rates and how they can be obtained from Timer 1.

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SU01627

Figure 18. Serial Port Mode 2



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### Enhanced UART operation

In addition to the standard operation modes, the UART can perform framing error detect by looking for missing stop bits, and automatic address recognition. The UART also fully supports multiprocessor communication.

When used for framing error detect the UART looks for missing stop bits in the communication. A missing bit will set the FE bit in the SCON register. The FE bit shares the SCON.7 bit with SM0 and the function of SCON.7 is determined by PCON.6 (SMOD0) (see Figure 20). If SMOD0 is set then SCON.7 functions as FE. SCON.7 functions as SM0 when SMOD0 is cleared. When used as FE SCON.7 can only be cleared by software. Refer to Figure 21.

### Automatic Address Recognition

Automatic Address Recognition is a feature which allows the UART to recognize certain addresses in the serial bit stream by using hardware to make the comparisons. This feature saves a great deal of software overhead by eliminating the need for the software to examine every serial address which passes by the serial port. This feature is enabled by setting the SM2 bit in SCON. In the 9 bit UART modes, mode 2 and mode 3, the Receive Interrupt flag (RI) will be automatically set when the received byte contains either the "Given" address or the "Broadcast" address. The 9 bit mode requires that the 9th information bit is a 1 to indicate that the received information is an address and not data. Automatic address recognition is shown in Figure 22.

The 8 bit mode is called Mode 1. In this mode the RI flag will be set if SM2 is enabled and the information received has a valid stop bit following the 8 address bits and the information is either a Given or Broadcast address.

Mode 0 is the Shift Register mode and SM2 is ignored.

Using the Automatic Address Recognition feature allows a master to selectively communicate with one or more slaves by invoking the Given slave address or addresses. All of the slaves may be contacted by using the Broadcast address. Two special Function Registers are used to define the slave's address, SADDR, and the address mask, SADEN. SADEN is used to define which bits in the SADDR are to be used and which bits are "don't care". The SADEN mask can be logically ANDed with the SADDR to create the "Given" address which the master will use for addressing each of the slaves. Use of the Given address allows multiple slaves to be recognized while excluding others. The following examples will help to show the versatility of this scheme:

Slave 0	SADDR =	1100 0000
	SADEN =	<u>1111 1101</u>
	Given =	1100 00X0

Slave 1	SADDR =	1100 0000
	SADEN =	<u>1111 1110</u>
	Given =	1100 000X

In the above example SADDR is the same and the SADEN data is used to differentiate between the two slaves. Slave 0 requires a 0 in bit 0 and it ignores bit 1. Slave 1 requires a 0 in bit 1 and bit 0 is ignored. A unique address for Slave 0 would be 1100 0010 since slave 1 requires a 0 in bit 1. A unique address for slave 1 would be 1100 0001 since a 1 in bit 0 will exclude slave 0. Both slaves can be selected at the same time by an address which has bit 0 = 0 (for slave 0) and bit 1 = 0 (for slave 1). Thus, both could be addressed with 1100 0000.

In a more complex system the following could be used to select slaves 1 and 2 while excluding slave 0:

Slave 0	SADDR =	1100 0000
	SADEN =	<u>1111 1001</u>
	Given =	1100 0XX0
Slave 1	SADDR =	1110 0000
	SADEN =	<u>1111 1010</u>
	Given =	1110 0X0X
Slave 2	SADDR =	1110 0000
	SADEN =	<u>1111 1100</u>
	Given =	1110 00XX

In the above example the differentiation among the 3 slaves is in the lower 3 address bits. Slave 0 requires that bit 0 = 0 and it can be uniquely addressed by 1110 0110. Slave 1 requires that bit 1 = 0 and it can be uniquely addressed by 1110 and 0101. Slave 2 requires that bit 2 = 0 and its unique address is 1110 0011. To select Slaves 0 and 1 and exclude Slave 2 use address 1110 0100, since it is necessary to make bit 2 = 1 to exclude slave 2.

The Broadcast Address for each slave is created by taking the logical OR of SADDR and SADEN. Zeros in this result are trended as don't-cares. In most cases, interpreting the don't-cares as ones, the broadcast address will be FF hexadecimal.

Upon reset SADDR (SFR address 0A9H) and SADEN (SFR address 0B9H) are loaded with 0s. This produces a given address of all "don't cares" as well as a Broadcast address of all "don't cares". This effectively disables the Automatic Addressing mode and allows the microcontroller to use standard 80C51 type UART drivers which do not make use of this feature.

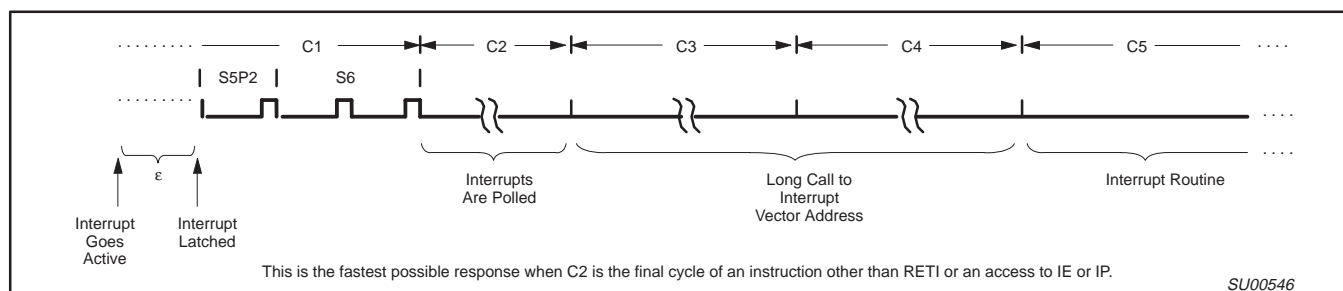




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**Figure 27. Interrupt Response Timing Diagram**

The polling cycle/LCALL sequence is illustrated in Figure 27.

Note that if an interrupt of higher priority level goes active prior to S5P2 of the machine cycle labeled C3 in Figure 27, then in accordance with the above rules it will be vectored to during C5 and C6, without any instruction of the lower priority routine having been executed.

Thus the processor acknowledges an interrupt request by executing a hardware-generated LCALL to the appropriate servicing routine. In some cases it also clears the flag that generated the interrupt, and in other cases it doesn't. It never clears the Serial Port flag. This has to be done in the user's software. It clears an external interrupt flag (IE0 or IE1) only if it was transition-activated. The hardware-generated LCALL pushes the contents of the Program Counter on to the stack (but it does not save the PSW) and reloads the PC with an address that depends on the source of the interrupt being vectored to, as shown in Table 10.

Execution proceeds from that location until the RETI instruction is encountered. The RETI instruction informs the processor that this interrupt routine is no longer in progress, then pops the top two bytes from the stack and reloads the Program Counter. Execution of the interrupted program continues from where it left off.

Note that a simple RET instruction would also have returned execution to the interrupted program, but it would have left the interrupt control system thinking an interrupt was still in progress, making future interrupts impossible.

### External Interrupts

The external sources can be programmed to be level-activated or transition-activated by setting or clearing bit IT1 or IT0 in Register TCON. If ITx = 0, external interrupt x is triggered by a detected low at the INTx pin. If ITx = 1, external interrupt x is edge triggered. In this mode if successive samples of the INTx pin show a high in one cycle and a low in the next cycle, interrupt request flag IEx in TCON is set. Flag bit IEx then requests the interrupt.

Since the external interrupt pins are sampled once each machine cycle, an input high or low should hold for at least 12 oscillator periods to ensure sampling. If the external interrupt is transition-activated, the external source has to hold the request pin high for at least one cycle, and then hold it low for at least one cycle. This is done to ensure that the transition is seen so that interrupt request flag IEx will be set. IEx will be automatically cleared by the CPU when the service routine is called.

If the external interrupt is level-activated, the external source has to hold the request active until the requested interrupt is actually generated. Then it has to deactivate the request before the interrupt

service routine is completed, or else another interrupt will be generated.

### Response Time

The INT0 and INT1 levels are inverted and latched into IE0 and IE1 at S5P2 of every machine cycle. The values are not actually polled by the circuitry until the next machine cycle. If a request is active and conditions are right for it to be acknowledged, a hardware subroutine call to the requested service routine will be the next instruction to be executed. The call itself takes two cycles. Thus, a minimum of three complete machine cycles elapse between activation of an external interrupt request and the beginning of execution of the first instruction of the service routine. Figure 27 shows interrupt response timings.

A longer response time would result if the request is blocked by one of the 3 previously listed conditions. If an interrupt of equal or higher priority level is already in progress, the additional wait time obviously depends on the nature of the other interrupt's service routine. If the instruction in progress is not in its final cycle, the additional wait time cannot be more than 3 cycles, since the longest instructions (MUL and DIV) are only 4 cycles long, and if the instruction in progress is RETI or an access to IE or IP, the additional wait time cannot be more than 5 cycles (a maximum of one more cycle to complete the instruction in progress, plus 4 cycles to complete the next instruction if the instruction is MUL or DIV).

Thus, in a single-interrupt system, the response time is always more than 3 cycles and less than 9 cycles.

As previously mentioned, the derivatives described in this data sheet have a four-level interrupt structure. The corresponding registers are IE, IP and IPH. (See Figures 24, 25, and 26.) The IPH (Interrupt Priority High) register makes the four-level interrupt structure possible.

The function of the IPH SFR is simple and when combined with the IP SFR determines the priority of each interrupt. The priority of each interrupt is determined as shown in the following table:

PRIORITY BITS		INTERRUPT PRIORITY LEVEL
IPH.x	IP.x	
0	0	Level 0 (lowest priority)
0	1	Level 1
1	0	Level 2
1	1	Level 3 (highest priority)

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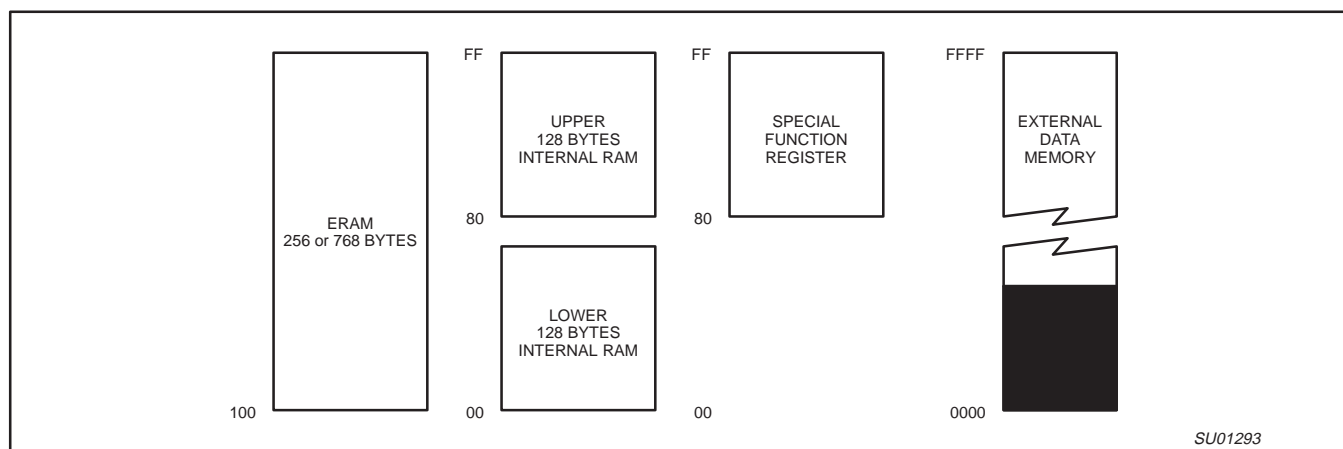


Figure 30. Internal and External Data Memory Address Space with EXTRAM = 0

### HARDWARE WATCHDOG TIMER (ONE-TIME ENABLED WITH RESET-OUT FOR P89C51RA2/RB2/RC2/RD2xx)

The WDT is intended as a recovery method in situations where the CPU may be subjected to software upset. The WDT consists of a 14-bit counter and the Watchdog Timer reset (WDTRST) SFR. The WDT is disabled at reset. To enable the WDT, the user must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When the WDT is enabled, it will increment every machine cycle while the oscillator is running and there is no way to disable the WDT except through reset (either hardware reset or WDT overflow reset). When the WDT overflows, it will drive an output reset HIGH pulse at the RST-pin (see the note below).

#### Using the WDT

To enable the WDT, the user must write 01EH and 0E1H in sequence to the WDTRST, SFR location 0A6H. When the WDT is

enabled, the user needs to service it by writing 01EH and 0E1H to WDTRST to avoid a WDT overflow. The 14-bit counter overflows when it reaches 16383 (3FFFH) and this will reset the device. When the WDT is enabled, it will increment every machine cycle while the oscillator is running. This means the user must reset the WDT at least every 16383 machine cycles. To reset the WDT, the user must write 01EH and 0E1h to WDTRST. WDTRST is a write only register. the WDT counter cannot be read or written. When the WDT overflows, it will generate an output RESET pulse at the reset pin (see note below). The RESET pulse duration is  $98 \times T_{OSC}$  (6-clock mode; 196 in 12-clock mode), where  $T_{OSC} = 1/f_{OSC}$ . To make the best use of the WDT, it should be serviced in those sections of code that will periodically be executed within the time required to prevent a WDT reset.

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### DC ELECTRICAL CHARACTERISTICS

 $T_{amb} = 0\text{ }^{\circ}\text{C to } +70\text{ }^{\circ}\text{C}; V_{CC} = 5\text{ V } \pm 10\%; V_{SS} = 0\text{ V (20/33 MHz max. CPU clock)}$ 

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP <sup>1</sup>	MAX	
$V_{IL}$	Input low voltage <sup>11</sup>	$4.5\text{ V} < V_{CC} < 5.5\text{ V}$	-0.5		$0.2 V_{CC} - 0.1$	V
$V_{IH}$	Input high voltage (ports 0, 1, 2, 3, $\overline{EA}$ )	—	$0.2 V_{CC} + 0.9$		$V_{CC} + 0.5$	V
$V_{IH1}$	Input high voltage, XTAL1, RST <sup>11</sup>	—	$0.7 V_{CC}$		$V_{CC} + 0.5$	V
$V_{OL}$	Output low voltage, ports 1, 2, 3 <sup>8</sup>	$V_{CC} = 4.5\text{ V}; I_{OL} = 1.6\text{ mA}^2$	—		0.4	V
$V_{OL1}$	Output low voltage, port 0, ALE, $\overline{PSEN}$ <sup>7, 8</sup>	$V_{CC} = 4.5\text{ V}; I_{OL} = 3.2\text{ mA}^2$	—		0.45	V
$V_{OH}$	Output high voltage, ports 1, 2, 3 <sup>3</sup>	$V_{CC} = 4.5\text{ V}; I_{OH} = -30\text{ }\mu\text{A}$	$V_{CC} - 0.7$		—	V
$V_{OH1}$	Output high voltage (port 0 in external bus mode), ALE <sup>9</sup> , $\overline{PSEN}$ <sup>3</sup>	$V_{CC} = 4.5\text{ V}; I_{OH} = -3.2\text{ mA}$	$V_{CC} - 0.7$		—	V
$I_{IL}$	Logical 0 input current, ports 1, 2, 3	$V_{IN} = 0.4\text{ V}$	-1		-75	$\mu\text{A}$
$I_{TL}$	Logical 1-to-0 transition current, ports 1, 2, 3 <sup>6</sup>	$V_{IN} = 2.0\text{ V}$ ; See note 4	—		-650	$\mu\text{A}$
$I_{LI}$	Input leakage current, port 0	$0.45 < V_{IN} < V_{CC} - 0.3$	—		$\pm 10$	$\mu\text{A}$
$I_{CC}$	Power supply current (see Figure 38): Active mode (see Note 5) Idle mode (see Note 5) Power-down mode or clock stopped (see Figure 42 for conditions) Programming and erase mode	See note 5  $T_{amb} = 0\text{ }^{\circ}\text{C to } 70\text{ }^{\circ}\text{C}$  $f_{OSC} = 20\text{ MHz}$				
				<30	100	$\mu\text{A}$
				60		mA
$R_{RST}$	Internal reset pull-down resistor	—	40		225	$\text{k}\Omega$
$C_{IO}$	Pin capacitance <sup>10</sup> (except $\overline{EA}$ )	—	—		15	pF

#### NOTES:

- Typical ratings are not guaranteed. The values listed are at room temperature, 5 V.
- Capacitive loading on ports 0 and 2 may cause spurious noise to be superimposed on the  $V_{OL}$ s of ALE and ports 1 and 3. The noise is due to external bus capacitance discharging into the port 0 and port 2 pins when these pins make 1-to-0 transitions during bus operations. In the worst cases (capacitive loading > 100 pF), the noise pulse on the ALE pin may exceed 0.8 V. In such cases, it may be desirable to qualify ALE with a Schmitt Trigger, or use an address latch with a Schmitt Trigger STROBE input.  $I_{OL}$  can exceed these conditions provided that no single output sinks more than 5 mA and no more than two outputs exceed the test conditions.
- Capacitive loading on ports 0 and 2 may cause the  $V_{OH}$  on ALE and  $\overline{PSEN}$  to momentarily fall below the  $V_{CC} - 0.7$  specification when the address bits are stabilizing.
- Pins of ports 1, 2 and 3 source a transition current when they are being externally driven from 1 to 0. The transition current reaches its maximum value when  $V_{IN}$  is approximately 2 V.
- See Figures 39 through 42 for  $I_{CC}$  test conditions and Figure 38 for  $I_{CC}$  vs. Frequency.  
12-clock mode characteristics:  
Active mode:  $I_{CC}(\text{MAX}) = (8.5 + 0.62 \times \text{FREQ. [MHz]})\text{mA}$   
Idle mode:  $I_{CC}(\text{MAX}) = (3.5 + 0.18 \times \text{FREQ. [MHz]})\text{mA}$
- This value applies to  $T_{amb} = 0\text{ }^{\circ}\text{C to } +70\text{ }^{\circ}\text{C}$ .
- Load capacitance for port 0, ALE, and  $\overline{PSEN} = 100\text{ pF}$ , load capacitance for all other outputs = 80 pF.
- Under steady state (non-transient) conditions,  $I_{OL}$  must be externally limited as follows:  
Maximum  $I_{OL}$  per port pin: 15 mA  
Maximum  $I_{OL}$  per 8-bit port: 26 mA  
Maximum total  $I_{OL}$  for all outputs: 71 mA  
If  $I_{OL}$  exceeds the test condition,  $V_{OL}$  may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.
- ALE is tested to  $V_{OH1}$ , except when ALE is off then  $V_{OH}$  is the voltage specification.
- Pin capacitance is characterized but not tested. Pin capacitance is less than 25 pF. Pin capacitance of ceramic package is less than 15 pF (except  $\overline{EA}$  is 25 pF).
- To improve noise rejection a nominal 100 ns glitch rejection circuitry has been added to the RST pin, and a nominal 15 ns glitch rejection circuitry has been added to the INT0 and INT1 pins. Previous devices provided only an inherent 5 ns of glitch rejection.

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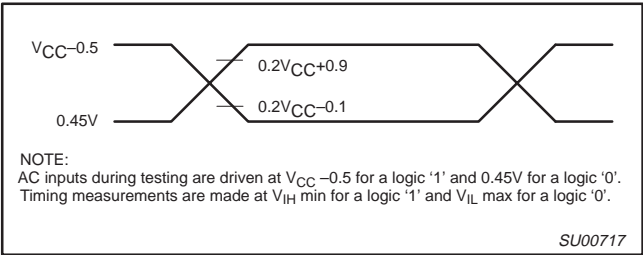


Figure 36. AC Testing Input/Output

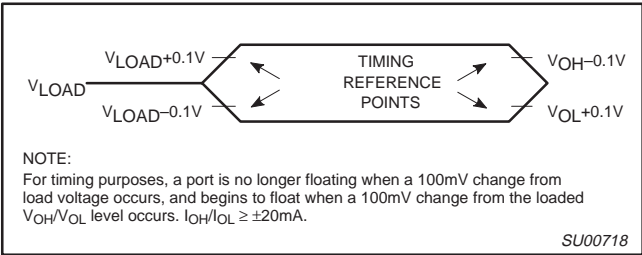


Figure 37. Float Waveform

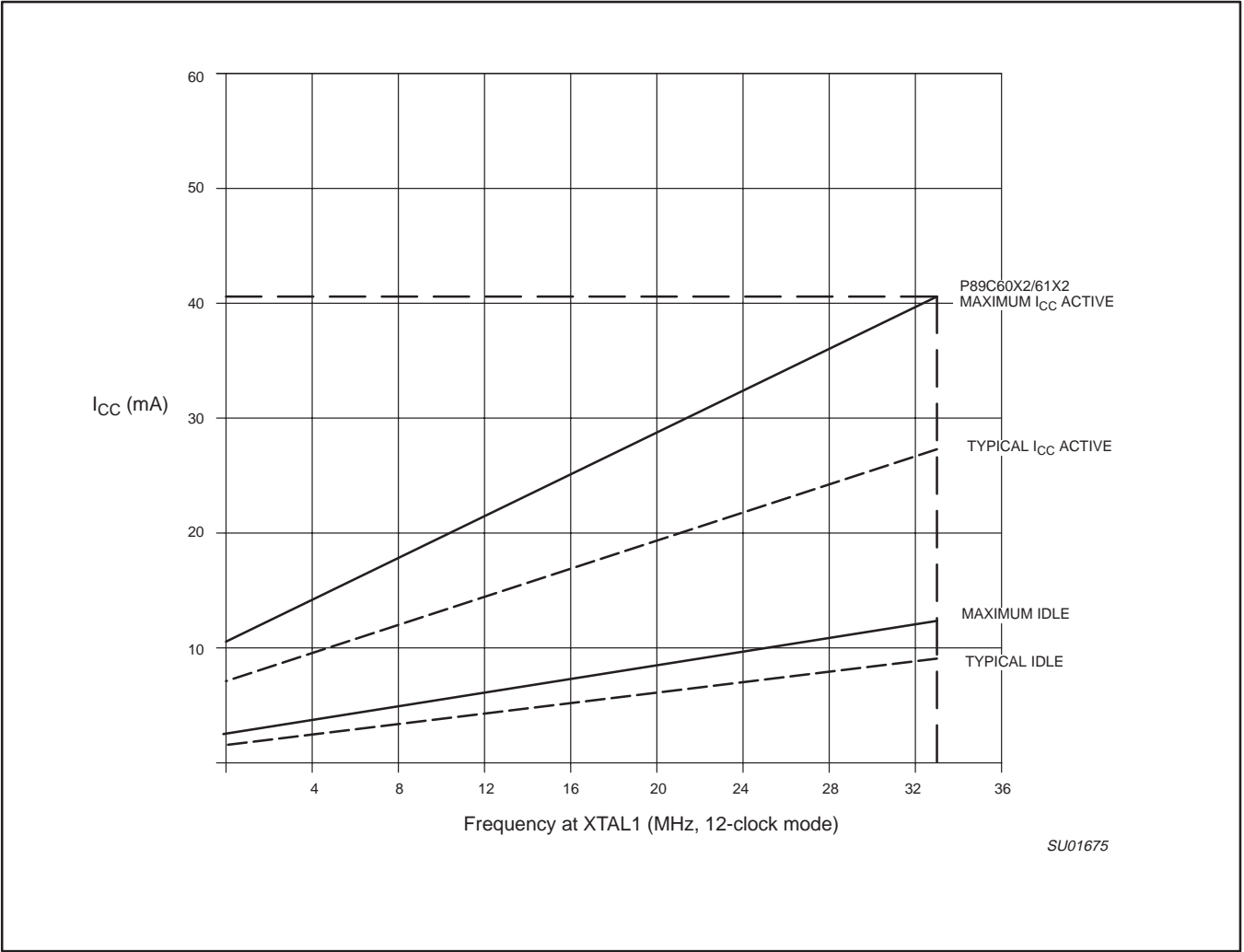


Figure 38.  $I_{CC}$  vs. FREQ for 12-clock operation  
Valid only within frequency specifications