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Details

Product Status	Active
Core Processor	AVR
Core Size	8-Bit
Speed	20MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	32
Program Memory Size	64KB (32K x 16)
Program Memory Type	FLASH
EEPROM Size	2K x 8
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-VFQFN Exposed Pad
Supplier Device Package	44-VQFN (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/atmega644p-20mur

ical ALU operation, two operands are output from the Register File, the operation is executed, and the result is stored back in the Register File – in one clock cycle.

Six of the 32 registers can be used as three 16-bit indirect address register pointers for Data Space addressing – enabling efficient address calculations. One of these address pointers can also be used as an address pointer for look up tables in Flash program memory. These added function registers are the 16-bit X-, Y-, and Z-register, described later in this section.

The ALU supports arithmetic and logic operations between registers or between a constant and a register. Single register operations can also be executed in the ALU. After an arithmetic operation, the Status Register is updated to reflect information about the result of the operation.

Program flow is provided by conditional and unconditional jump and call instructions, able to directly address the whole address space. Most AVR instructions have a single 16-bit word format. Every program memory address contains a 16- or 32-bit instruction.

Program Flash memory space is divided in two sections, the Boot Program section and the Application Program section. Both sections have dedicated Lock bits for write and read/write protection. The SPM instruction that writes into the Application Flash memory section must reside in the Boot Program section.

During interrupts and subroutine calls, the return address Program Counter (PC) is stored on the Stack. The Stack is effectively allocated in the general data SRAM, and consequently the Stack size is only limited by the total SRAM size and the usage of the SRAM. All user programs must initialize the SP in the Reset routine (before subroutines or interrupts are executed). The Stack Pointer (SP) is read/write accessible in the I/O space. The data SRAM can easily be accessed through the five different addressing modes supported in the AVR architecture.

The memory spaces in the AVR architecture are all linear and regular memory maps.

A flexible interrupt module has its control registers in the I/O space with an additional Global Interrupt Enable bit in the Status Register. All interrupts have a separate Interrupt Vector in the Interrupt Vector table. The interrupts have priority in accordance with their Interrupt Vector position. The lower the Interrupt Vector address, the higher the priority.

The I/O memory space contains 64 addresses for CPU peripheral functions as Control Registers, SPI, and other I/O functions. The I/O Memory can be accessed directly, or as the Data Space locations following those of the Register File, 0x20 - 0x5F. In addition, the ATmega164P/324P/644P has Extended I/O space from 0x60 - 0xFF in SRAM where only the ST/STS/STD and LD/LDS/LDD instructions can be used.

4.2 ALU – Arithmetic Logic Unit

The high-performance AVR ALU operates in direct connection with all the 32 general purpose working registers. Within a single clock cycle, arithmetic operations between general purpose registers or between a register and an immediate are executed. The ALU operations are divided into three main categories – arithmetic, logical, and bit-functions. Some implementations of the architecture also provide a powerful multiplier supporting both signed/unsigned multiplication and fractional format. See the “Instruction Set” section for a detailed description.

4.3 Status Register

The Status Register contains information about the result of the most recently executed arithmetic instruction. This information can be used for altering program flow in order to perform conditional operations. Note that the Status Register is updated after all ALU operations, as

5.4 EEPROM Data Memory

The ATmega164P/324P/644P contains 512B/1K/2K bytes of data EEPROM memory. It is organized as a separate data space, in which single bytes can be read and written. The EEPROM has an endurance of at least 100,000 write/erase cycles. The access between the EEPROM and the CPU is described in the following, specifying the EEPROM Address Registers, the EEPROM Data Register, and the EEPROM Control Register.

For a detailed description of SPI, JTAG and Parallel data downloading to the EEPROM, see page 308, page 312, and page 297 respectively.

5.4.1 EEPROM Read/Write Access

The EEPROM Access Registers are accessible in the I/O space. See "Register Description" on page 23 for details.

The write access time for the EEPROM is given in Table 5-2 on page 25. A self-timing function, however, lets the user software detect when the next byte can be written. If the user code contains instructions that write the EEPROM, some precautions must be taken. In heavily filtered power supplies, V_{CC} is likely to rise or fall slowly on power-up/down. This causes the device for some period of time to run at a voltage lower than specified as minimum for the clock frequency used. See Section "5.4.2" on page 21. for details on how to avoid problems in these situations.

In order to prevent unintentional EEPROM writes, a specific write procedure must be followed. Refer to the description of the EEPROM Control Register for details on this.

When the EEPROM is read, the CPU is halted for four clock cycles before the next instruction is executed. When the EEPROM is written, the CPU is halted for two clock cycles before the next instruction is executed.

5.4.2 Preventing EEPROM Corruption

During periods of low V_{CC} , the EEPROM data can be corrupted because the supply voltage is too low for the CPU and the EEPROM to operate properly. These issues are the same as for board level systems using EEPROM, and the same design solutions should be applied.

An EEPROM data corruption can be caused by two situations when the voltage is too low. First, a regular write sequence to the EEPROM requires a minimum voltage to operate correctly. Secondly, the CPU itself can execute instructions incorrectly, if the supply voltage is too low.

EEPROM data corruption can easily be avoided by following this design recommendation:

Keep the AVR RESET active (low) during periods of insufficient power supply voltage. This can be done by enabling the internal Brown-out Detector (BOD). If the detection level of the internal BOD does not match the needed detection level, an external low V_{CC} reset Protection circuit can be used. If a reset occurs while a write operation is in progress, the write operation will be completed provided that the power supply voltage is sufficient.

6.1.3 Flash Clock – $\text{clk}_{\text{FLASH}}$

The Flash clock controls operation of the Flash interface. The Flash clock is usually active simultaneously with the CPU clock.

6.1.4 Asynchronous Timer Clock – clk_{ASY}

The Asynchronous Timer clock allows the Asynchronous Timer/Counter to be clocked directly from an external clock or an external 32 kHz clock crystal. The dedicated clock domain allows using this Timer/Counter as a real-time counter even when the device is in sleep mode.

6.1.5 ADC Clock – clk_{ADC}

The ADC is provided with a dedicated clock domain. This allows halting the CPU and I/O clocks in order to reduce noise generated by digital circuitry. This gives more accurate ADC conversion results.

6.2 Clock Sources

The device has the following clock source options, selectable by Flash Fuse bits as shown below. The clock from the selected source is input to the AVR clock generator, and routed to the appropriate modules.

Table 6-1. Device Clocking Options Select⁽¹⁾

Device Clocking Option	CKSEL3..0
Low Power Crystal Oscillator	1111 - 1000
Full Swing Crystal Oscillator	0111 - 0110
Low Frequency Crystal Oscillator	0101 - 0100
Internal 128 kHz RC Oscillator	0011
Calibrated Internal RC Oscillator	0010
External Clock	0000
Reserved	0001

Note: 1. For all fuses “1” means unprogrammed while “0” means programmed.

6.2.1 Default Clock Source

The device is shipped with internal RC oscillator at 8.0MHz and with the fuse CKDIV8 programmed, resulting in 1.0MHz system clock. The startup time is set to maximum and time-out period enabled. (CKSEL = "0010", SUT = "10", CKDIV8 = "0"). The default setting ensures that all users can make their desired clock source setting using any available programming interface.

6.2.2 Clock Startup Sequence

Any clock source needs a sufficient V_{CC} to start oscillating and a minimum number of oscillating cycles before it can be considered stable.

To ensure sufficient V_{CC} , the device issues an internal reset with a time-out delay (t_{TOUT}) after the device reset is released by all other reset sources. "On-chip Debug System" on page 46 describes the start conditions for the internal reset. The delay (t_{TOUT}) is timed from the Watchdog Oscillator and the number of cycles in the delay is set by the SUTx and CKSELx fuse bits. The

1. In the same operation, write a logic one to the Watchdog change enable bit (WDCE) and WDE. A logic one must be written to WDE regardless of the previous value of the WDE bit.
2. Within the next four clock cycles, write the WDE and Watchdog prescaler bits (WDP) as desired, but with the WDCE bit cleared. This must be done in one operation.

The following code example shows one assembly and one C function for turning off the Watchdog Timer. The example assumes that interrupts are controlled (e.g. by disabling interrupts globally) so that no interrupts will occur during the execution of these functions.

Assembly Code Example⁽¹⁾

```
WDT_off:
    ; Turn off global interrupt
    cli
    ; Reset Watchdog Timer
    wdr
    ; Clear WDRF in MCUSR
    in    r16, MCUSR
    andi  r16, (0xff & (0<<WDRF))
    out   MCUSR, r16
    ; Write logical one to WDCE and WDE
    ; Keep old prescaler setting to prevent unintentional time-out
    in    r16, WDTCR
    ori   r16, (1<<WDCE) | (1<<WDE)
    out   WDTCR, r16
    ; Turn off WDT
    ldi   r16, (0<<WDE)
    out   WDTCR, r16
    ; Turn on global interrupt
    sei
    ret
```

C Code Example⁽¹⁾

```
void WDT_off(void)
{
    __disable_interrupt();
    __watchdog_reset();
    /* Clear WDRF in MCUSR */
    MCUSR &= ~(1<<WDRF);
    /* Write logical one to WDCE and WDE */
    /* Keep old prescaler setting to prevent unintentional time-out */
    WDTCR |= (1<<WDCE) | (1<<WDE);
    /* Turn off WDT */
    WDTCR = 0x00;
    __enable_interrupt();
}
```

Note: 1. The example code assumes that the part specific header file is included.

Table 9-1. Reset and Interrupt Vectors (Continued)

Vector No.	Program Address ⁽²⁾	Source	Interrupt Definition
28	\$0036	SPM_READY	Store Program Memory Ready
29	\$0038	USART1_RX	USART1 Rx Complete
30	\$003A	USART1_UDRE	USART1 Data Register Empty
31	\$003C	USART1_TX	USART1 Tx Complete

- Notes:
1. When the BOOTRST Fuse is programmed, the device will jump to the Boot Loader address at reset, see "Memory Programming" on page 293.
 2. When the IVSEL bit in MCUCR is set, Interrupt Vectors will be moved to the start of the Boot Flash Section. The address of each Interrupt Vector will then be the address in this table added to the start address of the Boot Flash Section.

Table 9-2 shows reset and Interrupt Vectors placement for the various combinations of BOOTRST and IVSEL settings. If the program never enables an interrupt source, the Interrupt Vectors are not used, and regular program code can be placed at these locations. This is also the case if the Reset Vector is in the Application section while the Interrupt Vectors are in the Boot section or vice versa.

Table 9-2. Reset and Interrupt Vectors Placement⁽¹⁾

BOOTRST	IVSEL	Reset Address	Interrupt Vectors Start Address
1	0	0x0000	0x0002
1	1	0x0000	Boot Reset Address + 0x0002
0	0	Boot Reset Address	0x0002
0	1	Boot Reset Address	Boot Reset Address + 0x0002

- Note:
1. The Boot Reset Address is shown in Table 23-7 on page 288. For the BOOTRST Fuse "1" means unprogrammed while "0" means programmed.

The most typical and general program setup for the Reset and Interrupt Vector Addresses in ATmega164P/324P/644P is:

Address	Labels	Code	Comments
0x0000		jmp RESET	; Reset
0x0002		jmp INT0	; IRQ0
0x0004		jmp INT1	; IRQ1
0x0006		jmp INT2	; IRQ2
0x0008		jmp PCINT0	; PCINT0
0x000A		jmp PCINT1	; PCINT1
0x000C		jmp PCINT2	; PCINT2
0x000E		jmp PCINT3	; PCINT3
0x0010		jmp WDT	; Watchdog Timeout
0x0012		jmp TIM2_COMPA	; Timer2 CompareA
0x0014		jmp TIM2_COMPB	; Timer2 CompareB
0x0016		jmp TIM2_OVF	; Timer2 Overflow
0x0018		jmp TIM1_CAPT	; Timer1 Capture
0x001A		jmp TIM1_COMPA	; Timer1 CompareA
0x001C		jmp TIM1_COMPB	; Timer1 CompareB
0x001E		jmp TIM1_OVF	; Timer1 Overflow
0x0020		jmp TIM0_COMPA	; Timer0 CompareA

- **MOSI/PCINT13 – Port B, Bit 5**

MOSI: SPI Master Data output, Slave Data input for SPI channel. When the SPI is enabled as a slave, this pin is configured as an input regardless of the setting of DDB5. When the SPI is enabled as a master, the data direction of this pin is controlled by DDB5. When the pin is forced to be an input, the pull-up can still be controlled by the PORTB5 bit.

PCINT13, Pin Change Interrupt source 13: The PB5 pin can serve as an external interrupt source.

- **\overline{SS} /OC0B/PCINT12 – Port B, Bit 4**

\overline{SS} : Slave Port Select input. When the SPI is enabled as a slave, this pin is configured as an input regardless of the setting of DDB4. As a slave, the SPI is activated when this pin is driven low. When the SPI is enabled as a master, the data direction of this pin is controlled by DDB4. When the pin is forced to be an input, the pull-up can still be controlled by the PORTB4 bit.

OC0B, Output Compare Match B output: The PB4 pin can serve as an external output for the Timer/Counter0 Output Compare. The pin has to be configured as an output (DDB4 set “one”) to serve this function. The OC0B pin is also the output pin for the PWM mode timer function.

PCINT12, Pin Change Interrupt source 12: The PB4 pin can serve as an external interrupt source.

- **AIN1/OC0A/PCINT11, Bit 3**

AIN1, Analog Comparator Negative input. This pin is directly connected to the negative input of the Analog Comparator.

OC0A, Output Compare Match A output: The PB3 pin can serve as an external output for the Timer/Counter0 Output Compare. The pin has to be configured as an output (DDB3 set “one”) to serve this function. The OC0A pin is also the output pin for the PWM mode timer function.

PCINT11, Pin Change Interrupt source 11: The PB3 pin can serve as an external interrupt source.

- **AIN0/INT2/PCINT10, Bit 2**

AIN0, Analog Comparator Positive input. This pin is directly connected to the positive input of the Analog Comparator.

INT2, External Interrupt source 2. The PB2 pin can serve as an External Interrupt source to the MCU.

PCINT10, Pin Change Interrupt source 10: The PB2 pin can serve as an external interrupt source.

- **T1/CLKO/PCINT9, Bit 1**

T1, Timer/Counter1 counter source.

CLKO, Divided System Clock: The divided system clock can be output on the PB1 pin. The divided system clock will be output if the CKOUT Fuse is programmed, regardless of the PORTB1 and DDB1 settings. It will also be output during reset.

PCINT9, Pin Change Interrupt source 9: The PB1 pin can serve as an external interrupt source.

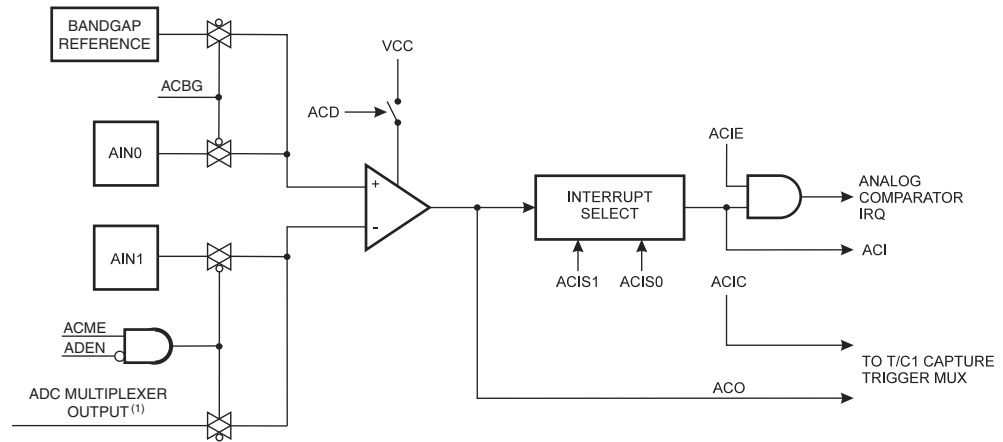
19. AC - Analog Comparator

19.1 Overview

The Analog Comparator compares the input values on the positive pin AIN0 and negative pin AIN1. When the voltage on the positive pin AIN0 is higher than the voltage on the negative pin AIN1, the Analog Comparator output, ACO, is set. The comparator's output can be set to trigger the Timer/Counter1 Input Capture function. In addition, the comparator can trigger a separate interrupt, exclusive to the Analog Comparator. The user can select Interrupt triggering on comparator output rise, fall or toggle. A block diagram of the comparator and its surrounding logic is shown in Figure 19-1.

The Power Reduction ADC bit, PRADC, in "PRR – Power Reduction Register" on page 48 must be disabled by writing a logical zero to be able to use the ADC input MUX.

Figure 19-1. Analog Comparator Block Diagram⁽²⁾



- Notes:
1. See Table 19-1 on page 237.
 2. Refer to Figure 1-1 on page 2 and Table 11-5 on page 81 for Analog Comparator pin placement.

19.2 Analog Comparator Multiplexed Input

It is possible to select any of the ADC7..0 pins to replace the negative input to the Analog Comparator. The ADC multiplexer is used to select this input, and consequently, the ADC must be switched off to utilize this feature. If the Analog Comparator Multiplexer Enable bit (ACME in ADCSRB) is set and the ADC is switched off (ADEN in ADCSRA is zero), MUX2..0 in ADMUX select the input pin to replace the negative input to the Analog Comparator, as shown in Table 19-1 on page 237. If ACME is cleared or ADEN is set, AIN1 is applied to the negative input to the Analog Comparator.

Table 19-1. Analog Comparator Multiplexed Input

ACME	ADEN	MUX2..0	Analog Comparator Negative Input
0	x	xxx	AIN1
1	1	xxx	AIN1
1	0	000	ADC0

in ADCSRA. The prescaler keeps running for as long as the ADEN bit is set, and is continuously reset when ADEN is low.

When initiating a single ended conversion by setting the ADSC bit in ADCSRA, the conversion starts at the following rising edge of the ADC clock cycle. See "Differential Gain Channels" on page 246 for details on differential conversion timing.

A normal conversion takes 13 ADC clock cycles. The first conversion after the ADC is switched on (ADEN in ADCSRA is set) takes 25 ADC clock cycles in order to initialize the analog circuitry.

When the bandgap reference voltage is used as input to the ADC, it will take a certain time for the voltage to stabilize. If not stabilized, the first value read after the first conversion may be wrong.

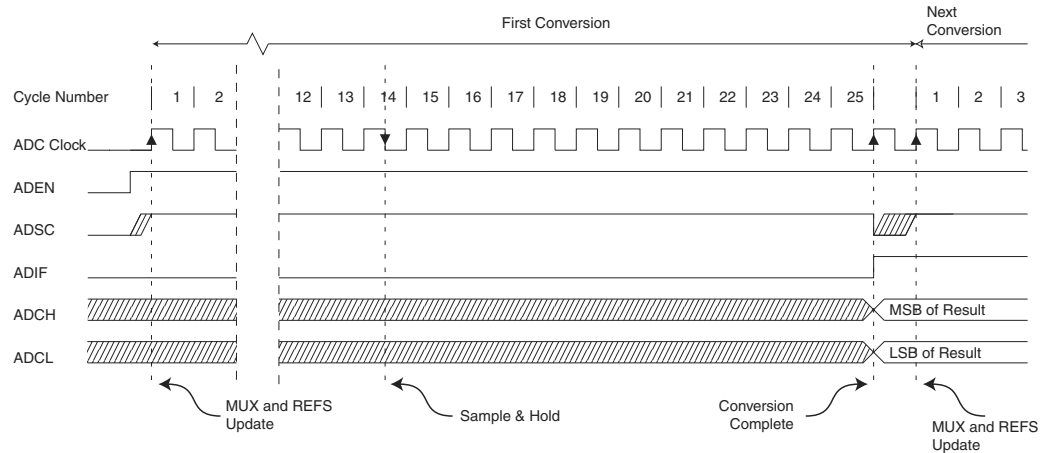
The actual sample-and-hold takes place 1.5 ADC clock cycles after the start of a normal conversion and 13.5 ADC clock cycles after the start of a first conversion. When a conversion is complete, the result is written to the ADC Data Registers, and ADIF is set. In single conversion mode, ADSC is cleared simultaneously. The software may then set ADSC again, and a new conversion will be initiated on the first rising ADC clock edge.

When Auto Triggering is used, the prescaler is reset when the trigger event occurs. This assures a fixed delay from the trigger event to the start of conversion. In this mode, the sample-and-hold takes place 2 ADC clock cycles after the rising edge on the trigger source signal. Three additional CPU clock cycles are used for synchronization logic.

When using Differential mode, along with Auto Triggering from a source other than the ADC Conversion Complete, each conversion will require 25 ADC clocks. This is because the ADC must be disabled and re-enabled after every conversion.

In Free Running mode, a new conversion will be started immediately after the conversion completes, while ADSC remains high. For a summary of conversion times, see Table 20-1 on page 246.

Figure 20-4. ADC Timing Diagram, First Conversion (Single Conversion Mode)



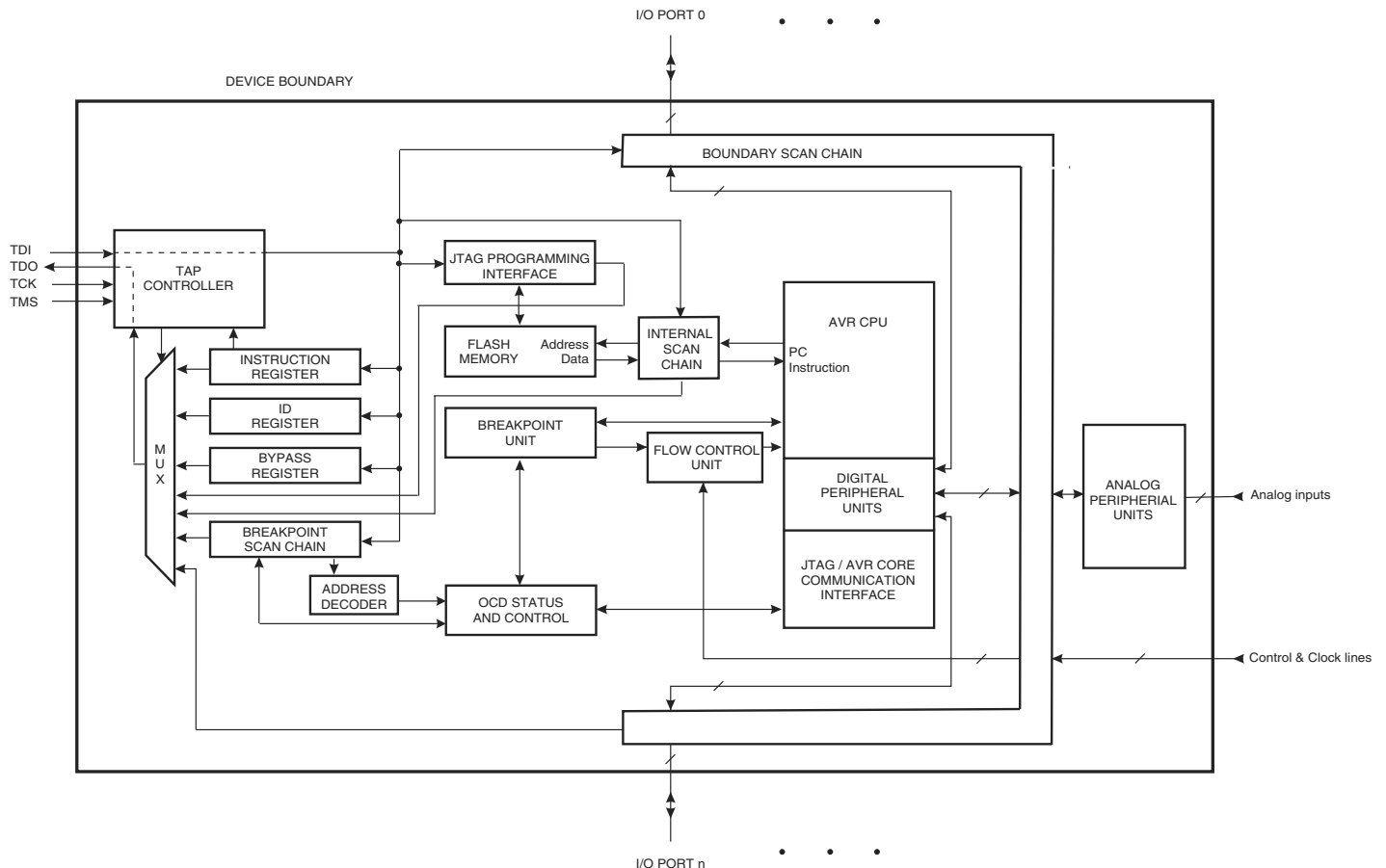
- TCK: Test Clock. JTAG operation is synchronous to TCK.
- TDI: Test Data In. Serial input data to be shifted in to the Instruction Register or Data Register (Scan Chains).
- TDO: Test Data Out. Serial output data from Instruction Register or Data Register.

The IEEE std. 1149.1 also specifies an optional TAP signal; TRST – Test ReSeT – which is not provided.

When the JTAGEN Fuse is unprogrammed, these four TAP pins are normal port pins, and the TAP controller is in reset. When programmed, the input TAP signals are internally pulled high and the JTAG is enabled for Boundary-scan and programming. The device is shipped with this fuse programmed.

For the On-chip Debug system, in addition to the JTAG interface pins, the $\overline{\text{RESET}}$ pin is monitored by the debugger to be able to detect external reset sources. The debugger can also pull the $\overline{\text{RESET}}$ pin low to reset the whole system, assuming only open collectors on the reset line are used in the application.

Figure 21-1. Block Diagram



23.8.13 Simple Assembly Code Example for a Boot Loader

```

; -the routine writes one page of data from RAM to Flash
; the first data location in RAM is pointed to by the Y pointer
; the first data location in Flash is pointed to by the Z-pointer
; -error handling is not included
; -the routine must be placed inside the Boot space
; (at least the Do_spm sub routine). Only code inside NRWW section can
; be read during Self-Programming (Page Erase and Page Write).
; -registers used: r0, r1, temp1 (r16), temp2 (r17), looplo (r24),
; loophi (r25), spmcval (r20)
; storing and restoring of registers is not included in the routine
; register usage can be optimized at the expense of code size
; -It is assumed that either the interrupt table is moved to the Boot
; loader section or that the interrupts are disabled.
.equ PAGESIZEB = PAGESIZE*2 ; PAGESIZEB is page size in BYTES, not words
.org SMALLBOOTSTART
Write_page:
; Page Erase
ldi spmcval, (1<<PGERS) | (1<<SPMEN)
call Do_spm

; re-enable the RWW section
ldi spmcval, (1<<RWWSRE) | (1<<SPMEN)
call Do_spm

; transfer data from RAM to Flash page buffer
ldi looplo, low(PAGESIZEB) ; init loop variable
ldi loophi, high(PAGESIZEB) ; not required for PAGESIZEB<=256
Wrloop:
ld r0, Y+
ld r1, Y+
ldi spmcval, (1<<SPMEN)
call Do_spm
adiw ZH:ZL, 2
sbiw loophi:looplo, 2 ; use subi for PAGESIZEB<=256
brne Wrloop

; execute Page Write
subi ZL, low(PAGESIZEB) ; restore pointer
sbci ZH, high(PAGESIZEB) ; not required for PAGESIZEB<=256
ldi spmcval, (1<<PGWRT) | (1<<SPMEN)
call Do_spm

; re-enable the RWW section
ldi spmcval, (1<<RWWSRE) | (1<<SPMEN)
call Do_spm

; read back and check, optional
ldi looplo, low(PAGESIZEB) ; init loop variable
ldi loophi, high(PAGESIZEB) ; not required for PAGESIZEB<=256
subi YL, low(PAGESIZEB) ; restore pointer
sbci YH, high(PAGESIZEB)
Rdloop:
elpm r0, Z+
ld r1, Y+
cpse r0, r1
jmp Error

```

1. Enter JTAG instruction PROG_COMMANDS.
2. Enable Flash read using programming instruction 3a.
3. Load the page address using programming instructions 3b, 3c and 3d. PCWORD (refer to Table 24-7 on page 296) is used to address within one page and must be written as 0.
4. Enter JTAG instruction PROG_PAGEREAD.
5. Read the entire page (or Flash) by shifting out all instruction words in the page (or Flash), starting with the LSB of the first instruction in the page (Flash) and ending with the MSB of the last instruction in the page (Flash). The Capture-DR state both captures the data from the Flash, and also auto-increments the program counter after each word is read. Note that Capture-DR comes before the shift-DR state. Hence, the first byte which is shifted out contains valid data.
6. Enter JTAG instruction PROG_COMMANDS.
7. Repeat steps 3 to 6 until all data have been read.

24.10.18 Programming the EEPROM

Before programming the EEPROM a Chip Erase must be performed, see “Performing Chip Erase” on page 322.

1. Enter JTAG instruction PROG_COMMANDS.
2. Enable EEPROM write using programming instruction 4a.
3. Load address High byte using programming instruction 4b.
4. Load address Low byte using programming instruction 4c.
5. Load data using programming instructions 4d and 4e.
6. Repeat steps 4 and 5 for all data bytes in the page.
7. Write the data using programming instruction 4f.
8. Poll for EEPROM write complete using programming instruction 4g, or wait for t_{WLRH} (refer to Table 24-14 on page 306).
9. Repeat steps 3 to 8 until all data have been programmed.

Note that the PROG_PAGELOAD instruction can not be used when programming the EEPROM.

24.10.19 Reading the EEPROM

1. Enter JTAG instruction PROG_COMMANDS.
2. Enable EEPROM read using programming instruction 5a.
3. Load address using programming instructions 5b and 5c.
4. Read data using programming instruction 5d.
5. Repeat steps 3 and 4 until all data have been read.

Note that the PROG_PAGEREAD instruction can not be used when reading the EEPROM.

24.10.20 Programming the Fuses

1. Enter JTAG instruction PROG_COMMANDS.
2. Enable Fuse write using programming instruction 6a.
3. Load data high byte using programming instructions 6b. A bit value of “0” will program the corresponding fuse, a “1” will unprogram the fuse.
4. Write Fuse High byte using programming instruction 6c.
5. Poll for Fuse write complete using programming instruction 6d, or wait for t_{WLRH} (refer to Table 24-14 on page 306).

25. Electrical Characteristics

Absolute Maximum Ratings*

Operating Temperature	-55°C to +125°C
Storage Temperature	-65°C to +150°C
Voltage on any Pin except $\overline{\text{RESET}}$ with respect to Ground	-0.5V to $V_{CC}+0.5V$
Voltage on $\overline{\text{RESET}}$ with respect to Ground	-0.5V to +13.0V
Maximum Operating Voltage	6.0V
DC Current per I/O Pin	40.0 mA
DC Current V_{CC} and GND Pins	200.0 mA

*NOTICE: Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

25.1 DC Characteristics

$T_A = -40^\circ\text{C}$ to 85°C , $V_{CC} = 1.8V$ to $5.5V$ (unless otherwise noted)

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
V_{IL}	Input Low Voltage, Except XTAL1 and Reset pin	$V_{CC} = 1.8V - 2.4V$ $V_{CC} = 2.4V - 5.5V$	-0.5 -0.5		$0.2V_{CC}^{(1)}$ $0.3V_{CC}^{(1)}$	V
V_{IL1}	Input Low Voltage, XTAL1 pin	$V_{CC} = 1.8V - 5.5V$	-0.5		$0.1V_{CC}^{(1)}$	V
V_{IL2}	Input Low Voltage, RESET pin	$V_{CC} = 1.8V - 5.5V$	-0.5		$0.1V_{CC}^{(1)}$	V
V_{IH}	Input High Voltage, Except XTAL1 and RESET pins	$V_{CC} = 1.8V - 2.4V$ $V_{CC} = 2.4V - 5.5V$	$0.7V_{CC}^{(2)}$ $0.6V_{CC}^{(2)}$		$V_{CC} + 0.5$ $V_{CC} + 0.5$	V
V_{IH1}	Input High Voltage, XTAL1 pin	$V_{CC} = 1.8V - 2.4V$ $V_{CC} = 2.4V - 5.5V$	$0.8V_{CC}^{(2)}$ $0.7V_{CC}^{(2)}$		$V_{CC} + 0.5$ $V_{CC} + 0.5$	V
V_{IH2}	Input High Voltage, RESET pin	$V_{CC} = 1.8V - 5.5V$	$0.9V_{CC}^{(2)}$		$V_{CC} + 0.5$	V
V_{OL}	Output Low Voltage ⁽³⁾ ,	$I_{OL} = 20\text{ mA}$, $V_{CC} = 5V$ $I_{OL} = 10\text{ mA}$, $V_{CC} = 3V$			0.9 0.6	V
V_{OH}	Output High Voltage ⁽⁴⁾ ,	$I_{OH} = -20\text{ mA}$, $V_{CC} = 5V$ $I_{OH} = -10\text{ mA}$, $V_{CC} = 3V$	4.2 2.3			V
I_{IL}	Input Leakage Current I/O Pin	$V_{CC} = 5.5V$, pin low (absolute value)			1	μA
I_{IH}	Input Leakage Current I/O Pin	$V_{CC} = 5.5V$, pin high (absolute value)			1	μA
R_{RST}	Reset Pull-up Resistor		30		60	$k\Omega$
R_{PU}	I/O Pin Pull-up Resistor		20		50	$k\Omega$

Figure 26-32. BOD Threshold vs. Temperature (Bodlevel is 2.7V).

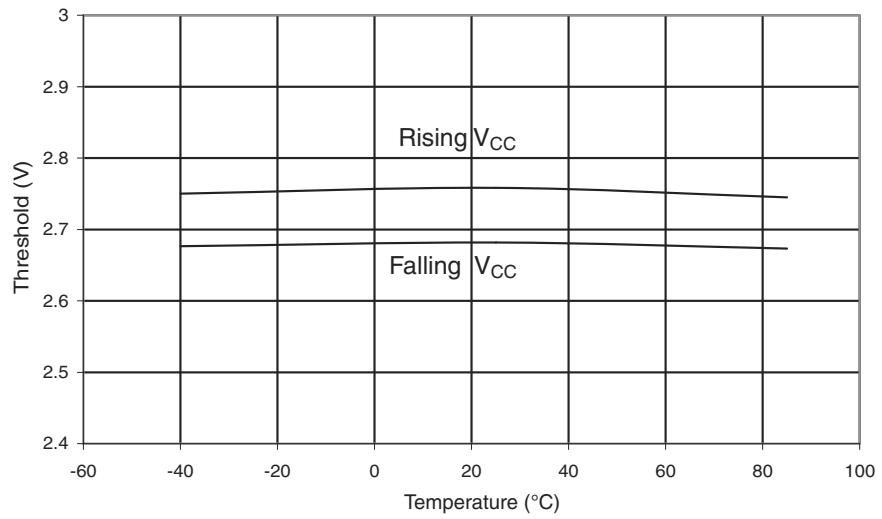


Figure 26-33. BOD Threshold vs. Temperature (Bodlevel is 1.8V).

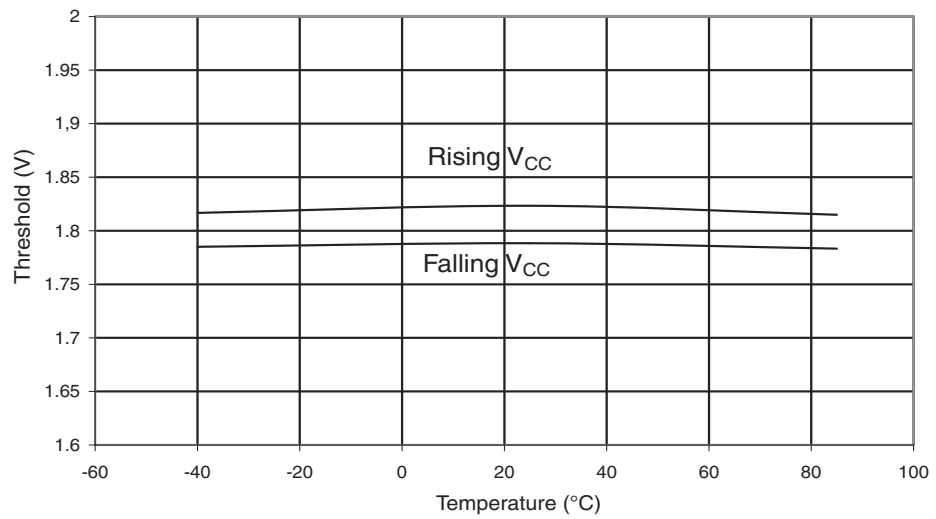


Figure 26-54. Idle Supply Current vs. V_{CC} (1 - 20 MHz).

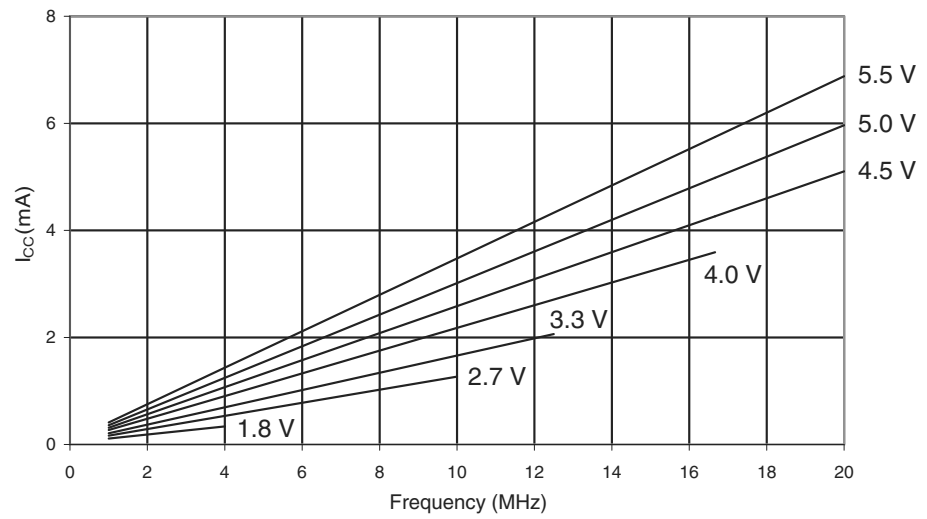
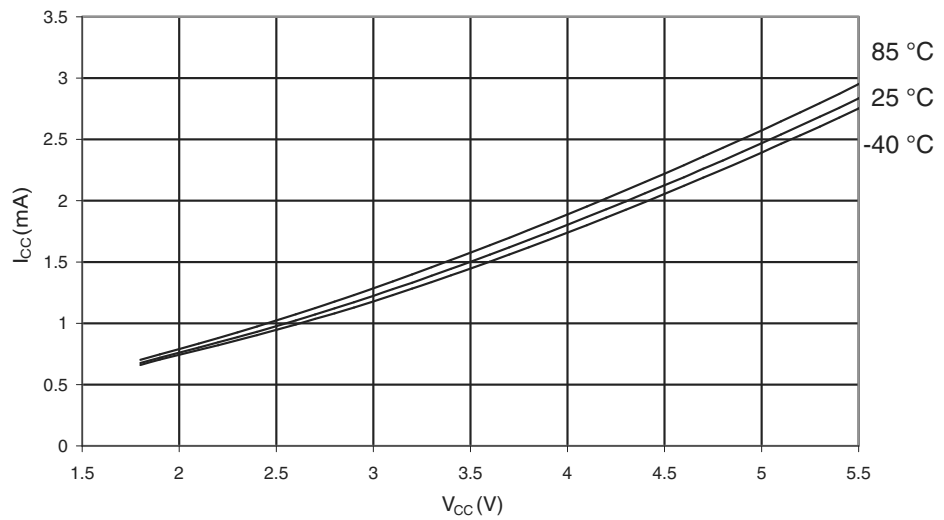


Figure 26-55. Idle Supply Current vs. V_{CC} (Internal RC Oscillator, 8 MHz).



26.2.9 Pin Threshold and Hysteresis

Figure 26-72. I/O Pin Input Threshold vs. V_{CC} (V_{IH} , I/O Pin Read as '1').

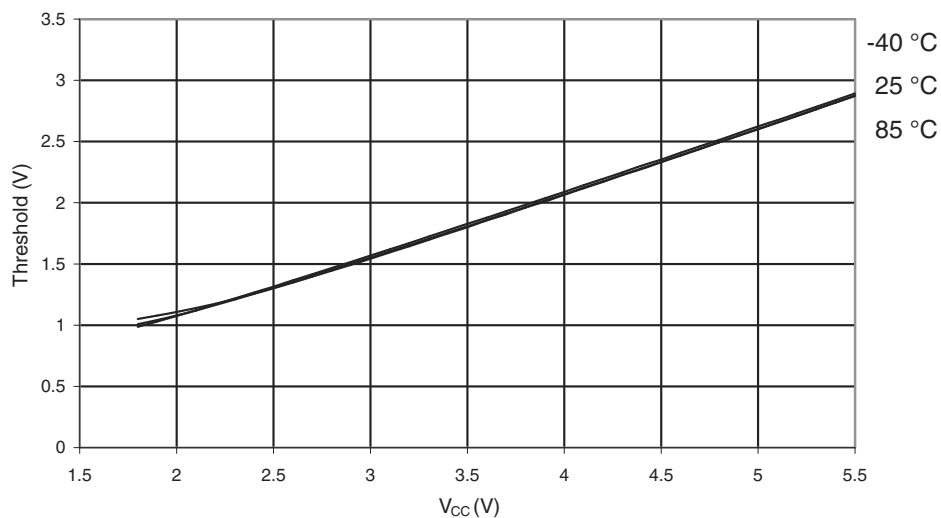


Figure 26-73. I/O Pin Input Threshold vs. V_{CC} (V_{IL} , I/O Pin Read as '0').

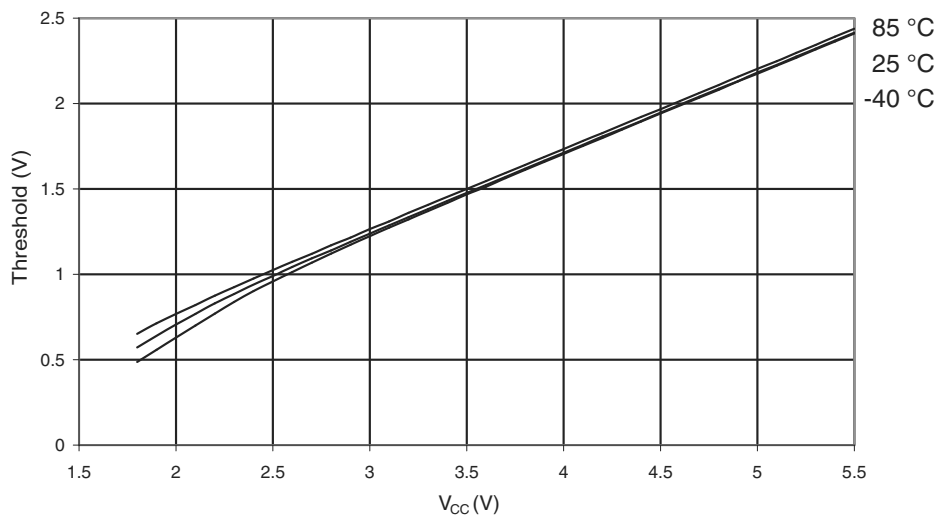


Figure 26-76. Reset Pin Input Threshold vs. V_{CC} (V_{IL} , I/O Pin Read as '0').

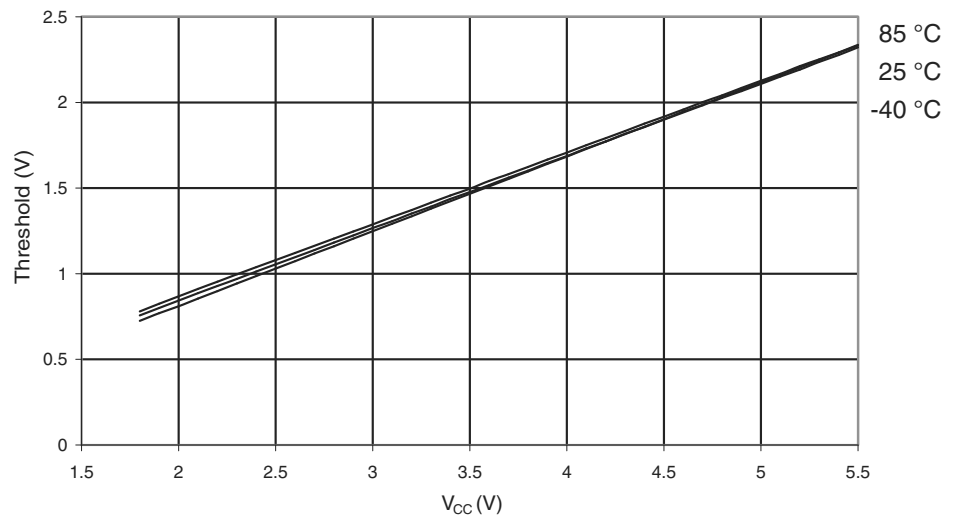
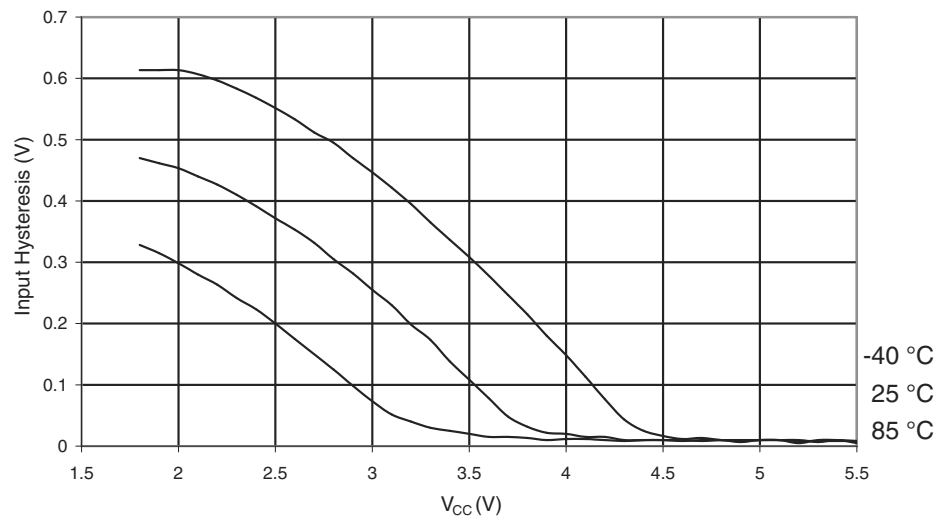
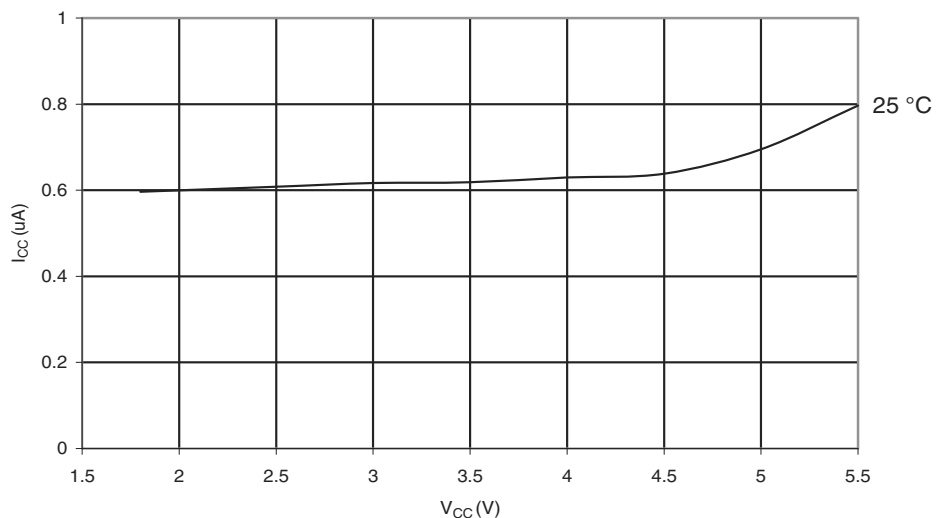


Figure 26-77. Reset Pin Input Hysteresis vs. V_{CC}



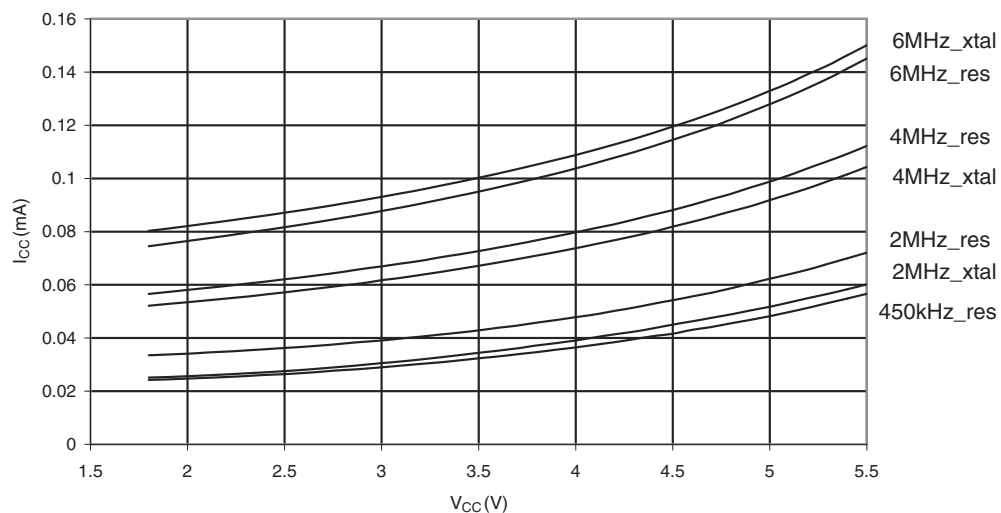
26.3.5 Power-save Supply Current

Figure 26-107.Power-save Supply Current vs. V_{CC} (Watchdog Timer Disabled).



26.3.6 Standby Supply Current

Figure 26-108.Standby Supply Current vs. V_{CC} (Watchdog Timer Disabled).



29. Ordering Information

29.1 ATmega164P

Speed (MHz) ⁽³⁾	Power Supply	Ordering Code	Package ⁽¹⁾	Operational Range
10	1.8 - 5.5V	ATmega164PV-10AU ⁽²⁾	44A	Industrial (-40°C to 85°C)
		ATmega164PV-10PU ⁽²⁾	40P6	
		ATmega164PV-10MU ⁽²⁾	44M1	
20	2.7 - 5.5V	ATmega164P-20AU ⁽²⁾	44A	
		ATmega164P-20PU ⁽²⁾	40P6	
		ATmega164P-20MU ⁽²⁾	44M1	

- Notes:
1. This device can also be supplied in wafer form. Please contact your local Atmel sales office for detailed ordering information and minimum quantities.
 2. Pb-free packaging, complies to the European Directive for Restriction of Hazardous Substances (RoHS directive). Also Halide free and fully Green.
 3. For Speed vs. V_{CC} see "Speed Grades" on page 329.

Package Type	
44A	44-lead, Thin (1.0 mm) Plastic Gull Wing Quad Flat Package (TQFP)
40P6	40-pin, 0.600" Wide, Plastic Dual Inline Package (PDIP)
44M1	44-pad, 7 x 7 x 1.0 mm body, lead pitch 0.50 mm, Thermally Enhanced Plastic Very Thin Quad Flat No-Lead (VQFN)

32.13 Rev. 8011B - 09/06

1. Updated "DC Characteristics" on page 325.

32.14 Rev. 8011A - 08/06

1. Initial revision.

32.12 Rev. 8011C - 10/06	429
32.13 Rev. 8011B - 09/06	430
32.14 Rev. 8011A - 08/06	430
<i>Table of Contents</i>	<i>i</i>