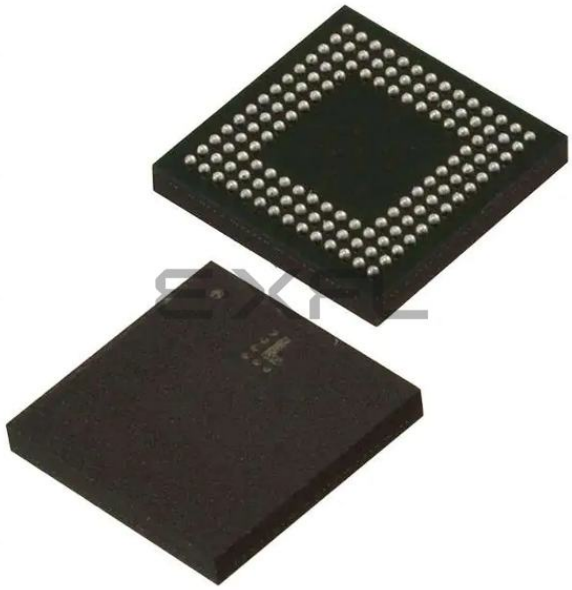

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	625
Number of Logic Elements/Cells	5000
Total RAM Bits	169984
Number of I/O	86
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 125°C (TJ)
Package / Case	132-LFBGA, CSPBGA
Supplier Device Package	132-CSBGA (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/laxp2-5e-5mn132e

Figure 2-4. General Purpose PLL (GPLL) Diagram

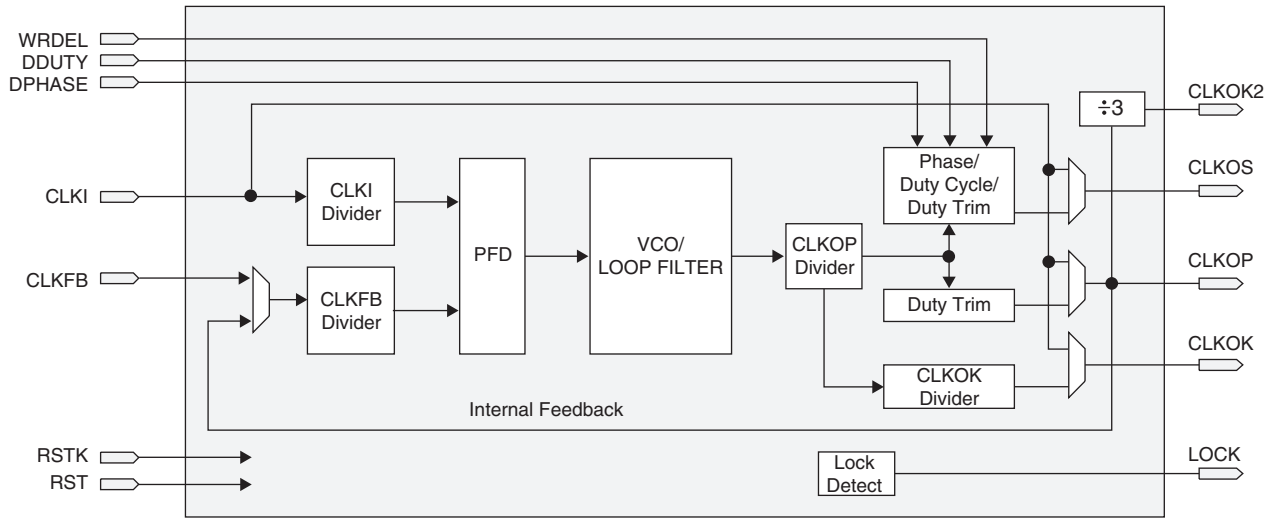


Table 2-4 provides a description of the signals in the GPLL blocks.

Table 2-4. GPLL Block Signal Descriptions

Signal	I/O	Description
CLKI	I	Clock input from external pin or routing
CLKFB	I	PLL feedback input from CLKOP (PLL internal), from clock net (CLKOP) or from a user clock (PIN or logic)
RST	I	"1" to reset PLL counters, VCO, charge pumps and M-dividers
RSTK	I	"1" to reset K-divider
DPHASE [3:0]	I	DPA Phase Adjust input
DDDUTY [3:0]	I	DPA Duty Cycle Select input
WRDEL	I	DPA Fine Delay Adjust input
CLKOS	O	PLL output clock to clock tree (phase shifted/duty cycle changed)
CLKOP	O	PLL output clock to clock tree (no phase shift)
CLKOK	O	PLL output to clock tree through secondary clock divider
CLKOK2	O	PLL output to clock tree (CLKOP divided by 3)
LOCK	O	"1" indicates PLL LOCK to CLKI

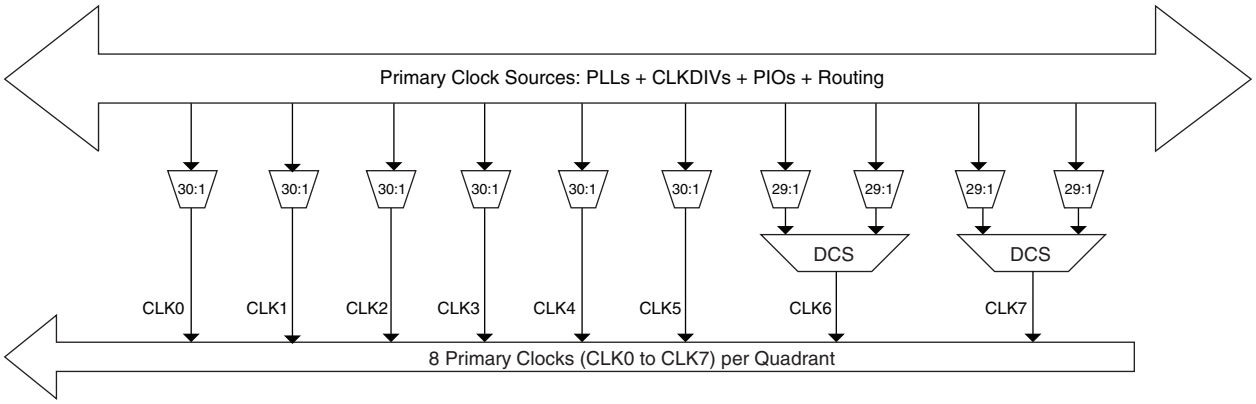
Clock Dividers

LA-LatticeXP2 devices have two clock dividers, one on the left side and one on the right side of the device. These are intended to generate a slower-speed system clock from a high-speed edge clock. The block operates in a ÷2, ÷4 or ÷8 mode and maintains a known phase relationship between the divided down clock and the high-speed clock based on the release of its reset signal. The clock dividers can be fed from the CLKOP output from the GPLLs or from the Edge Clocks (ECLK). The clock divider outputs serve as primary clock sources and feed into the clock distribution network. The Reset (RST) control signal resets the input and forces all outputs to low. The RELEASE signal releases outputs to the input clock. For further information on clock dividers, see TN1126, LatticeXP2 sysCLOCK PLL Design and Usage Guide. Figure 2-5 shows the clock divider connections.

Primary Clock Routing

The clock routing structure in LA-LatticeXP2 devices consists of a network of eight primary clock lines (CLK0 through CLK7) per quadrant. The primary clocks of each quadrant are generated from muxes located in the center of the device. All the clock sources are connected to these muxes. Figure 2-9 shows the clock routing for one quadrant. Each quadrant mux is identical. If desired, any clock can be routed globally.

Figure 2-9. Per Quadrant Primary Clock Selection

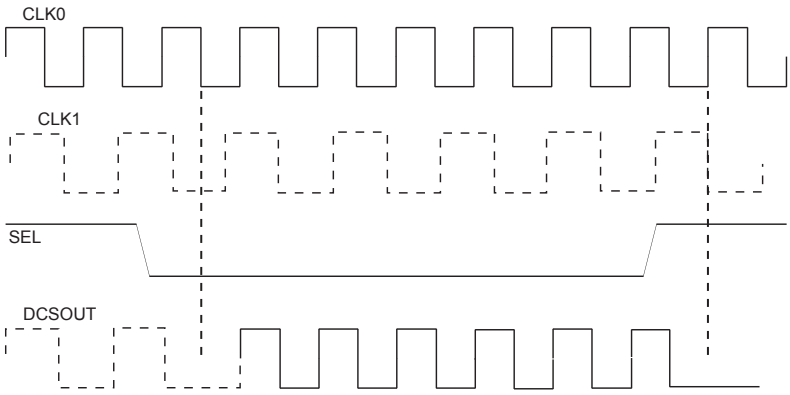


Dynamic Clock Select (DCS)

The DCS is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources without any glitches or runt pulses. This is achieved irrespective of when the select signal is toggled. There are two DCS blocks per quadrant; in total, eight DCS blocks per device. The inputs to the DCS block come from the center muxes. The output of the DCS is connected to primary clocks CLK6 and CLK7 (see Figure 2-9).

Figure 2-10 shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information on the DCS, see TN1126, LatticeXP2 sysCLOCK PLL Design and Usage Guide.

Figure 2-10. DCS Waveforms



Secondary Clock/Control Routing

Secondary clocks in the LA-LatticeXP2 devices are region-based resources. The benefit of region-based resources is the relatively low injection delay and skew within the region, as compared to primary clocks. EBR rows, DSP rows and a special vertical routing channel bound the secondary clock regions. This special vertical routing channel aligns with either the left edge of the center DSP block in the DSP row or the center of the DSP row. Figure 2-11 shows this special vertical routing channel and the six secondary clock regions for the LA-

Slice Clock Selection

Figure 2-13 shows the clock selections and Figure 2-14 shows the control selections for Slice 0 through Slice 2. All the primary clocks and the four secondary clocks are routed to this clock selection mux. Other signals, via routing, can be used as clock inputs to the slices. Slice controls are generated from the secondary clocks or other signals connected via routing.

If none of the signals are selected for both clock and control, then the default value of the mux output is 1. Slice 3 does not have any registers; therefore it does not have the clock or control muxes.

Figure 2-13. Slice 0 through Slice 2 Clock Selection

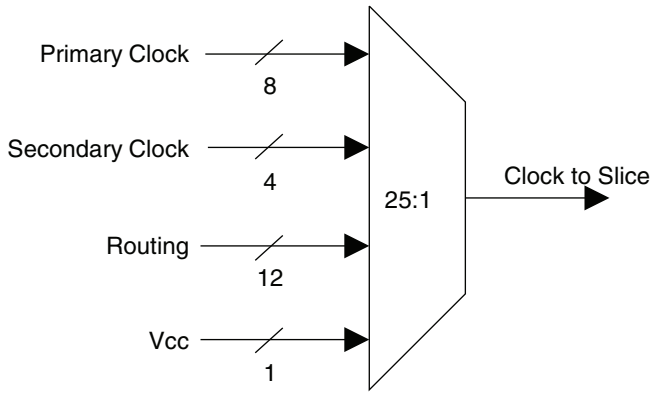
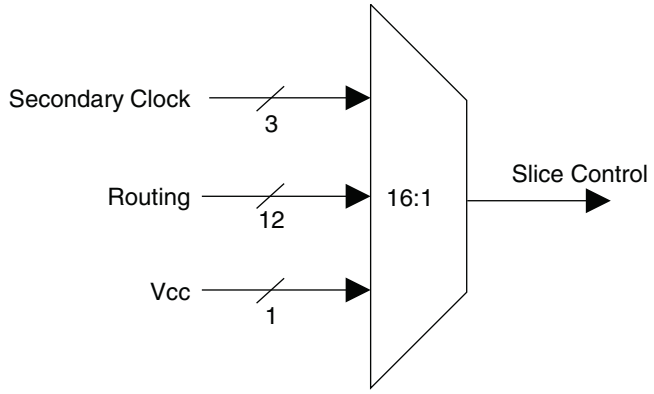


Figure 2-14. Slice 0 through Slice 2 Control Selection



Edge Clock Routing

LA-LatticeXP2 devices have eight high-speed edge clocks that are intended for use with the PIOs in the implementation of high-speed interfaces. Each device has two edge clocks per edge. Figure 2-15 shows the selection muxes for these clocks.

EBR memory supports two forms of write behavior for single port or dual port operation:

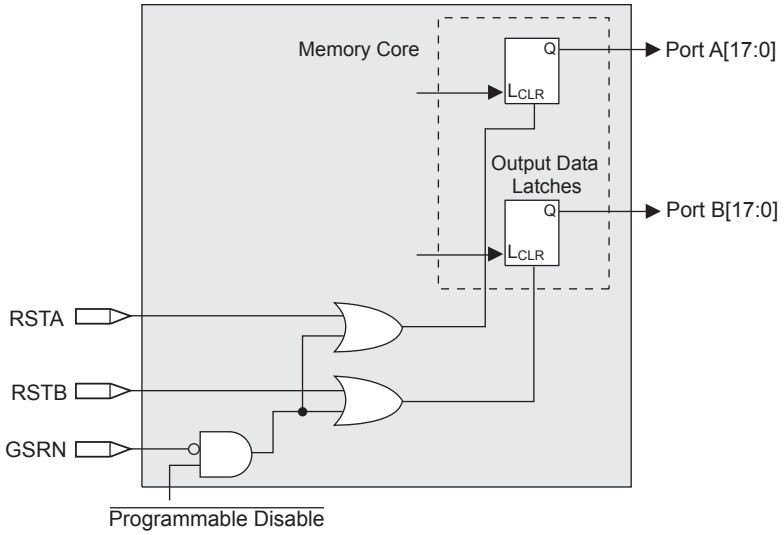
1. **Normal** – Data on the output appears only during a read cycle. During a write cycle, the data (at the current address) does not appear on the output. This mode is supported for all data widths.

2. **Write Through** – A copy of the input data appears at the output of the same port during a write cycle. This mode is supported for all data widths.

Memory Core Reset

The memory array in the EBR utilizes latches at the A and B output ports. These latches can be reset asynchronously or synchronously. RSTA and RSTB are local signals, which reset the output latches associated with Port A and Port B respectively. GSRN, the global reset signal, resets both ports. The output data latches and associated resets for both ports are as shown in Figure 2-17.

Figure 2-17. Memory Core Reset

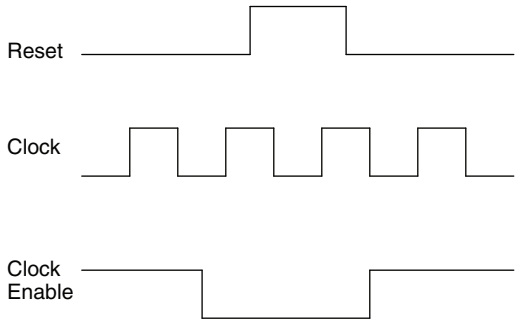


For further information on the sysMEM EBR block, see TN1137, LatticeXP2 Memory Usage Guide.

EBR Asynchronous Reset

EBR asynchronous reset or GSR (if used) can only be applied if all clock enables are low for a clock cycle before the reset is applied and released a clock cycle after the low-to-high transition of the reset signal, as shown in Figure 2-18. The GSR input to the EBR is always asynchronous.

Figure 2-18. EBR Asynchronous Reset (Including GSR) Timing Diagram



mixed within a function element. Similarly, the operand widths cannot be mixed within a block. DSP elements can be concatenated.

The resources in each sysDSP block can be con•gured to support the following four elements:

- MULT (Multiply)
- MAC (Multiply, Accumulate)
- MULTADDSUB (Multiply, Addition/Subtraction)
- MULTADDSUBSUM (Multiply, Addition/Subtraction, Accumulate)

The number of elements available in each block depends on the width selected from the three available options: x9, x18, and x36. A number of these elements are concatenated for highly parallel implementations of DSP functions. Table 2-6 shows the capabilities of the block.

Table 2-6. Maximum Number of Elements in a Block

Width of Multiply	x9	x18	x36
MULT	8	4	1
MAC	2	2	—
MULTADDSUB	4	2	—
MULTADDSUBSUM	2	1	—

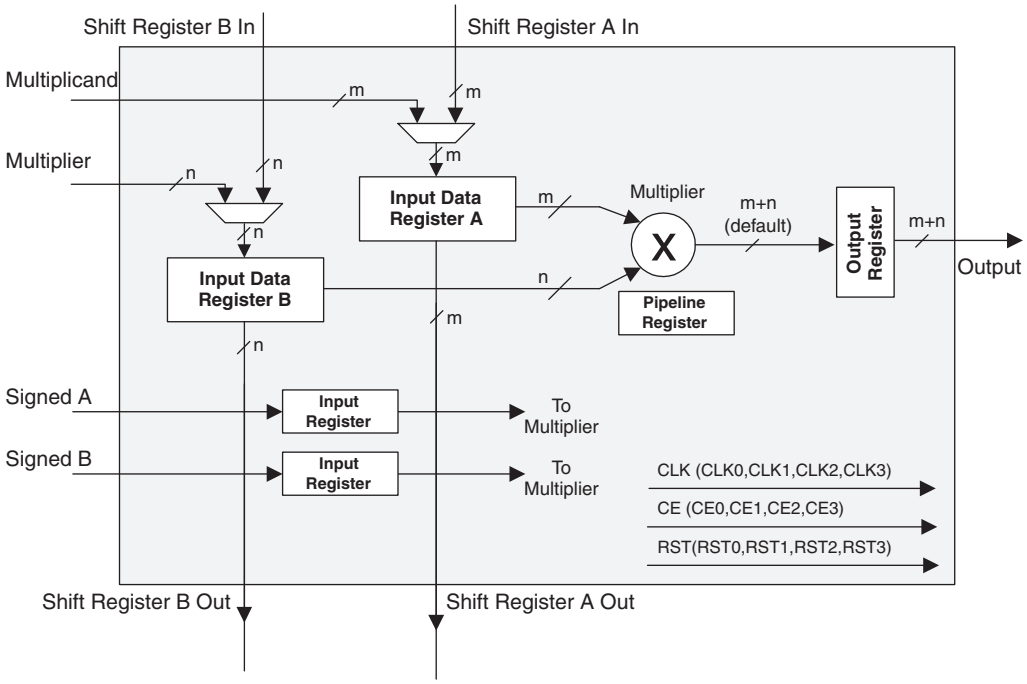
Some options are available in four elements. The input register in all the elements can be directly loaded or can be loaded as shift register from previous operand registers. By selecting ‘dynamic operation’ the following operations are possible:

- In the ‘Signed/Unsigned’ options the operands can be switched between signed and unsigned on every cycle.
- In the ‘Add/Sub’ option the Accumulator can be switched between addition and subtraction on every cycle.
- The loading of operands can switch between parallel and serial operations.

MULT sysDSP Element

This multiplier element implements a multiply with no addition or accumulator nodes. The two operands, A and B, are multiplied and the result is available at the output. The user can enable the input/output and pipeline registers. Figure 2-20 shows the MULT sysDSP element.

Figure 2-20. MULT sysDSP Element



IPexpress™

The user can access the sysDSP block via the Lattice IPexpress tool, which provides the option to configure each DSP module (or group of modules), or by direct HDL instantiation. In addition, Lattice has partnered with The MathWorks® to support instantiation in the Simulink® tool, a graphical simulation environment. Simulink works with Diamond to dramatically shorten the DSP design cycle in Lattice FPGAs.

Optimized DSP Functions

Lattice provides a library of optimized DSP IP functions. Some of the IP cores planned for the LA-LatticeXP2 DSP include the Bit Correlator, FFT functions, FIR Filter, Reed-Solomon Encoder/Decoder, Turbo Encoder/Decoder and Convolutional Encoder/Decoder. Please contact Lattice to obtain the latest list of available DSP IP cores.

Resources Available in the LA-LatticeXP2 Family

Table 2-8 shows the maximum number of multipliers for each member of the LA-LatticeXP2 family. Table 2-9 shows the maximum available EBR RAM Blocks and Serial TAG Memory bits in each LA-LatticeXP2 device. EBR blocks, together with Distributed RAM can be used to store variables locally for fast DSP operations.

Table 2-8. Maximum Number of DSP Blocks in the LA-LatticeXP2 Family

Device	DSP Block	9 x 9 Multiplier	18 x 18 Multiplier	36 x 36 Multiplier
LA-XP2-5	3	24	12	3
LA-XP2-8	4	32	16	4
LA-XP2-17	5	40	20	5

Table 2-9. Embedded SRAM/TAG Memory in the LA-LatticeXP2 Family

Device	EBR SRAM Block	Total EBR SRAM (kbits)	TAG Memory (Bits)
LA-XP2-5	9	166	632
LA-XP2-8	12	221	768
LA-XP2-17	15	276	2184

LA-LatticeXP2 DSP Performance

Table 2-10 lists the maximum performance in Millions of MAC (MMAC) operations per second for each member of the LA-LatticeXP2 family.

Table 2-10. DSP Performance

Device	DSP Block	DSP Performance MMAC
LA-XP2-5	3	3,900
LA-XP2-8	4	5,200
LA-XP2-17	5	6,500

For further information on the sysDSP block, see TN1140, LatticeXP2 sysDSP Usage Guide.

Programmable I/O Cells (PIC)

Each PIC contains two PIOs connected to their respective sysIO buffers as shown in Figure 2-25. The PIO Block supplies the output data (DO) and the tri-state control signal (TO) to the sysIO buffer and receives input from the buffer. Table 2-11 provides the PIO signal list.

Table 2-12. Supported Input Standards

Input Standard	V_{REF} (Nom.)	V_{CCIO}[1] (Nom.)
Single Ended Interfaces		
LVTTL	—	—
LVCMOS33	—	—
LVCMOS25	—	—
LVCMOS18	—	1.8
LVCMOS15	—	1.5
LVCMOS12	—	—
PCI33	—	—
HSTL18 Class I, II	0.9	—
HSTL15 Class I	0.75	—
SSTL33 Class I, II	1.5	—
SSTL25 Class I, II	1.25	—
SSTL18 Class I, II	0.9	—
Differential Interfaces		
Differential SSTL18 Class I, II	—	—
Differential SSTL25 Class I, II	—	—
Differential SSTL33 Class I, II	—	—
Differential HSTL15 Class I	—	—
Differential HSTL18 Class I, II	—	—
LVDS, MLVDS, LVPECL, BLVDS, RSDS	—	—

1. When not specified, V_{CCIO} can be set anywhere in the valid operating range (page 3-1).

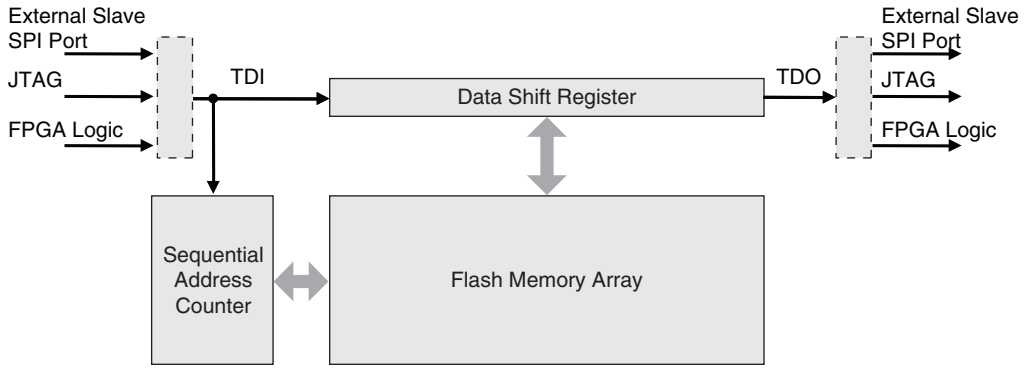
1. Unlocked
2. Key Locked – Presenting the key through the programming interface allows the device to be unlocked.
3. Permanently Locked – The device is permanently locked.

To further complement the security of the device a One Time Programmable (OTP) mode is available. Once the device is set in this mode it is not possible to erase or re-program the Flash portion of the device.

Serial TAG Memory

LA-LatticeXP2 devices offer 0.6 to 3.3kbits of Flash memory in the form of Serial TAG memory. The TAG memory is an area of the on-chip Flash that can be used for non-volatile storage including electronic ID codes, version codes, date stamps, asset IDs and calibration settings. A block diagram of the TAG memory is shown in Figure 2-34. The TAG memory is accessed in the same way as external SPI Flash and it can be read or programmed either through JTAG, an external Slave SPI Port, or directly from FPGA logic. To read the TAG memory, a start address is specified and the entire TAG memory contents are read sequentially in a first-in-first-out manner. The TAG memory is independent of the Flash used for device configuration and given its use for general-purpose storage functions is always accessible regardless of the device security settings. For more information, see TN1137, LatticeXP2 Memory Usage Guide and TN1141, LatticeXP2 sysCONFIG Usage Guide.

Figure 2-34. Serial TAG Memory Diagram



Live Update Technology

Many applications require field updates of the FPGA. LA-LatticeXP2 devices provide three features that enable this configuration to be done in a secure and failsafe manner while minimizing impact on system operation.

1. **Decryption Support**
   LA-LatticeXP2 devices provide on-chip, non-volatile key storage to support decryption of a 128-bit AES encrypted bitstream, securing designs and deterring design piracy.

2. **TransFR (Transparent Field Reconfiguration)**
   TransFR I/O (TFR) is a unique Lattice technology that allows users to update their logic in the field without interrupting system operation using a single ispVM command. TransFR I/O allows I/O states to be frozen during device configuration. This allows the device to be field updated with a minimum of system disruption and downtime. For more information please see TN1143, LatticeXP2 TransFR I/O.

3. **Dual Boot Image Support**
   Dual boot images are supported for applications requiring reliable remote updates of configuration data for the system FPGA. After the system is running with a basic configuration, a new boot image can be downloaded remotely and stored in a separate location in the configuration storage device. Any time after the update the LA-LatticeXP2 can be re-booted from this new configuration file. If there is a problem such as corrupt data during download or incorrect version number with this new boot image, the LA-LatticeXP2 device can revert back

Density Shifting

The LA-LatticeXP2 family is designed to ensure that different density devices in the same family and in the same package have the same pinout. Furthermore, the architecture ensures a high success rate when performing design migration from lower density devices to higher density devices. In many cases, it is also possible to shift a lower utilization design targeted for a high-density device to a lower density device. However, the exact details of the final resource utilization will impact the likely success in each case.

Programming and Erase Flash Supply Current[1, 2, 3, 4, 5]

Over Recommended Operating Conditions

Symbol	Parameter	Device	Typical (25 °C, Max. Supply)[6]	Units
I_{CC}	Core Power Supply Current	LA-XP2-5	17	mA
		LA-XP2-8	21	mA
		LA-XP2-17	28	mA
I_{CCAUX}	Auxiliary Power Supply Current[7]	LA-XP2-5	64	mA
		LA-XP2-8	66	mA
		LA-XP2-17	83	mA
I_{CCPLL}	PLL Power Supply Current (per PLL)		0.1	mA
I_{CCIO}	Bank Power Supply Current (per Bank)		5	mA
I_{CCJ}	V_{CCJ} Power Supply Current[8]		14	mA

1. For further information on supply current, see TN1139, Power Estimation and Management for LatticeXP2 Devices.
2. Assumes all outputs are tristated, all inputs are con•gured as LVCMOS and held at the V_{CCIO} or GND.
3. Frequency 0 MHz (excludes dynamic power from FPGA operation).
4. A specific configuration pattern is used that scales with the size of the device; consists of 75% PFU utilization, 50% EBR, and 25% I/O configuration.
5. Bypass or decoupling capacitor across the supply.
6. T_J = 25 °C, power supplies at nominal voltage.
7. In fpBGA packages the PLLs are connected to and powered from the auxiliary power supply. For these packages, the actual auxiliary supply current is the sum of I_{CCAUX} and I_{CCPLL}. For csBGA, PQFP and TQFP packages the PLLs are powered independent of the auxiliary power supply.
8. When programming via JTAG.

sysIO Single-Ended DC Electrical Characteristics

Over Recommended Operating Conditions

Input/Output Standard	V_{IL}		V_{IH}		V_{OL}	V_{OH}	I_{OL}[1] (mA)	I_{OH}[1] (mA)
	Min. (V)	Max. (V)	Min. (V)	Max. (V)	Max. (V)	Min. (V)		
LVCMOS33	–0.3	0.8	2.0	3.6	0.4	V_{CCIO} - 0.4	20, 16, 12, 8, 4	–20, –16, –12, –8, –4
					0.2	V_{CCIO} - 0.2	0.1	–0.1
LVTTL33	–0.3	0.8	2.0	3.6	0.4	V_{CCIO} - 0.4	20, 16, 12, 8, 4	–20, –16, –12, –8, –4
					0.2	V_{CCIO} - 0.2	0.1	–0.1
LVCMOS25	–0.3	0.7	1.7	3.6	0.4	V_{CCIO} - 0.4	20, 16, 12, 8, 4	–20, –16, –12, –8, –4
					0.2	V_{CCIO} - 0.2	0.1	–0.1
LVCMOS18	–0.3	0.35 V_{CCIO}	0.65 V_{CCIO}	3.6	0.4	V_{CCIO} - 0.4	16, 12, 8, 4	–16, –12, –8, –4
					0.2	V_{CCIO} - 0.2	0.1	–0.1
LVCMOS15	–0.3	0.35 V_{CCIO}	0.65 V_{CCIO}	3.6	0.4	V_{CCIO} - 0.4	8, 4	–8, –4
					0.2	V_{CCIO} - 0.2	0.1	–0.1
LVCMOS12	–0.3	0.35 V_{CC}	0.65 V_{CC}	3.6	0.4	V_{CCIO} - 0.4	6, 2	–6, –2
					0.2	V_{CCIO} - 0.2	0.1	–0.1
PCI33	–0.3	0.3 V_{CCIO}	0.5 V_{CCIO}	3.6	0.1 V_{CCIO}	0.9 V_{CCIO}	1.5	–0.5
SSTL33_I	–0.3	V_{REF} - 0.2	V_{REF} + 0.2	3.6	0.7	V_{CCIO} - 1.1	8	–8
SSTL33_II	–0.3	V_{REF} - 0.2	V_{REF} + 0.2	3.6	0.5	V_{CCIO} - 0.9	16	–16
SSTL25_I	–0.3	V_{REF} - 0.18	V_{REF} + 0.18	3.6	0.54	V_{CCIO} - 0.62	7.6	–7.6
							12	–12
SSTL25_II	–0.3	V_{REF} - 0.18	V_{REF} + 0.18	3.6	0.35	V_{CCIO} - 0.43	15.2	–15.2
							20	–20
SSTL18_I	–0.3	V_{REF} - 0.125	V_{REF} + 0.125	3.6	0.4	V_{CCIO} - 0.4	6.7	–6.7
SSTL18_II	–0.3	V_{REF} - 0.125	V_{REF} + 0.125	3.6	0.28	V_{CCIO} - 0.28	8	–8
							11	–11
HSTL15_I	–0.3	V_{REF} - 0.1	V_{REF} + 0.1	3.6	0.4	V_{CCIO} - 0.4	4	–4
							8	–8
HSTL18_I	–0.3	V_{REF} - 0.1	V_{REF} + 0.1	3.6	0.4	V_{CCIO} - 0.4	8	–8
							12	–12
HSTL18_II	–0.3	V_{REF} - 0.1	V_{REF} + 0.1	3.6	0.4	V_{CCIO} - 0.4	16	–16

1. The average DC current drawn by I/Os between GND connections, or between the last GND in an I/O bank and the end of an I/O bank, as shown in the logic signal connections table shall not exceed n * 8 mA, where n is the number of I/Os between bank GND connections or between the last GND in a bank and the end of a bank.

sysIO Differential Electrical Characteristics

LVDS

Over Recommended Operating Conditions

Parameter	Description	Test Conditions	Min.	Typ.	Max.	Units
V_{INP}, V_{INM}	Input Voltage		0	—	2.4	V
V_{CM}	Input Common Mode Voltage	Half the Sum of the Two Inputs	0.05	—	2.35	V
V_{THD}	Differential Input Threshold	Difference Between the Two Inputs	+/–100	—	—	mV
I_{IN}	Input Current	Power On or Power Off	—	—	+/–10	µA
V_{OH}	Output High Voltage for V_{OP} or V_{OM}	R_T = 100 Ohm	—	1.38	1.60	V
V_{OL}	Output Low Voltage for V_{OP} or V_{OM}	R_T = 100 Ohm	0.9 V	1.03	—	V
V_{OD}	Output Voltage Differential	(V_{OP} - V_{OM}), R_T = 100 Ohm	250	350	450	mV
ýV$_{OD}$	Change in V_{OD} Between High and Low		—	—	50	mV
V_{OS}	Output Voltage Offset	(V_{OP} + V_{OM})/2, R_T = 100 Ohm	1.125	1.20	1.375	V
ýV$_{OS}$	Change in V_{OS} Between H and L		—	—	50	mV
I_{SA}	Output Short Circuit Current	V_{OD} = 0 V Driver Outputs Shorted to Ground	—	—	24	mA
I_{SAB}	Output Short Circuit Current	V_{OD} = 0 V Driver Outputs Shorted to Each Other	—	—	12	mA

Differential HSTL and SSTL

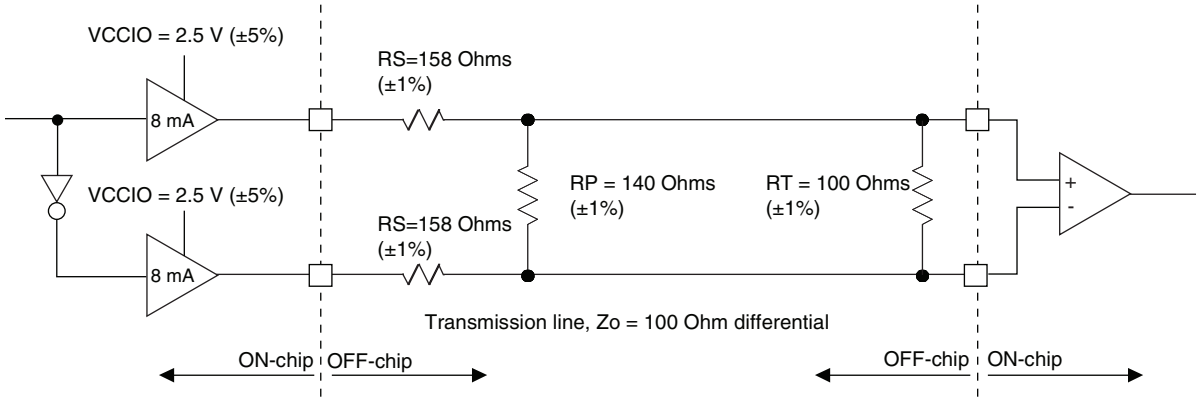
Differential HSTL and SSTL outputs are implemented as a pair of complementary single-ended outputs. All allowable single-ended output classes (class I and class II) are supported in this mode.

For further information on LVPECL, RSDS, MLVDS, BLVDS and other differential interfaces please see details in additional technical notes listed at the end of this data sheet.

LVDS25E

The top and bottom sides of LA-LatticeXP2 devices support LVDS outputs via emulated complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The scheme shown in Figure 3-1 is one possible solution for point-to-point signals.

Figure 3-1. LVDS25E Output Termination Example



LVPECL

The LA-LatticeXP2 devices support the differential LVPECL standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The LVPECL input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 3-3 is one possible solution for point-to-point signals.

Figure 3-3. Differential LVPECL

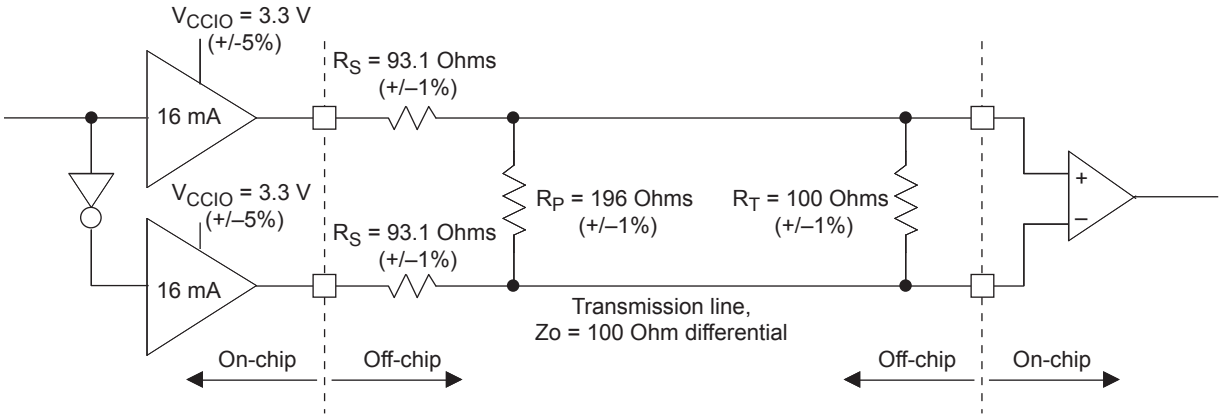


Table 3-3. LVPECL DC Conditions[1]

Over Recommended Operating Conditions

Parameter	Description	Typical	Units
V_{CCIO}	Output Driver Supply (+/–5%)	3.30	V
Z_{OUT}	Driver Impedance	10	Ω
R_S	Driver Series Resistor (+/–1%)	93	Ω
R_P	Driver Parallel Resistor (+/–1%)	196	Ω
R_T	Receiver Termination (+/–1%)	100	Ω
V_{OH}	Output High Voltage (After R_P)	2.05	V
V_{OL}	Output Low Voltage (After R_P)	1.25	V
V_{OD}	Output Differential Voltage (After R_P)	0.80	V
V_{CM}	Output Common Mode Voltage	1.65	V
Z_{BACK}	Back Impedance	100.5	Ω
I_{DC}	DC Output Current	12.11	mA

1. For input buffer, see LVDS table.

RSDS

The LA-LatticeXP2 devices support differential RSDS standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The RSDS input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 3-4 is one possible solution for RSDS standard implementation. Resistor values in Figure 3-4 are industry standard values for 1% resistors.

Figure 3-4. RSDS (Reduced Swing Differential Standard)

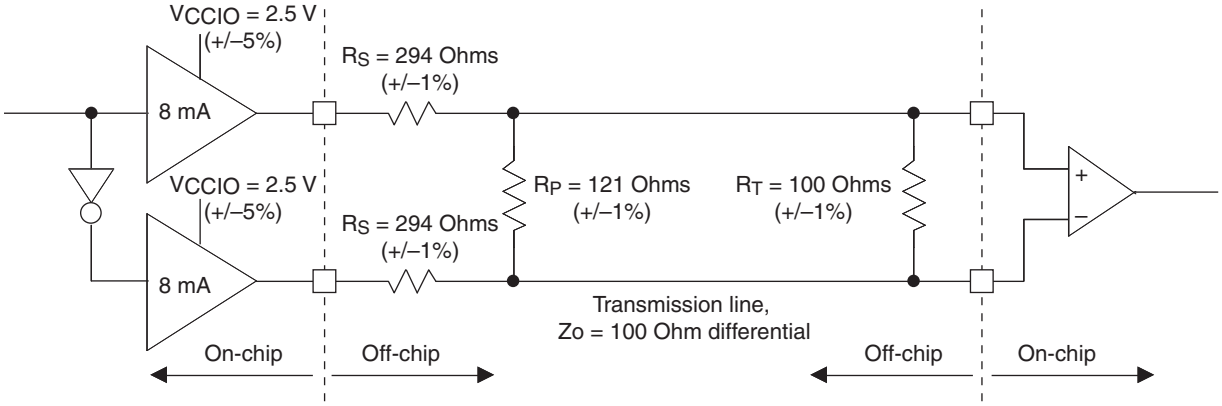


Table 3-4. RSDS DC Conditions[1]

Over Recommended Operating Conditions

Parameter	Description	Typical	Units
V_{CCIO}	Output Driver Supply (+/–5%)	2.50	V
Z_{OUT}	Driver Impedance	20	Ω
R_S	Driver Series Resistor (+/–1%)	294	Ω
R_P	Driver Parallel Resistor (+/–1%)	121	Ω
R_T	Receiver Termination (+/–1%)	100	Ω
V_{OH}	Output High Voltage (After R_P)	1.35	V
V_{OL}	Output Low Voltage (After R_P)	1.15	V
V_{OD}	Output Differential Voltage (After R_P)	0.20	V
V_{CM}	Output Common Mode Voltage	1.25	V
Z_{BACK}	Back Impedance	101.5	Ω
I_{DC}	DC Output Current	3.66	mA

1. For input buffer, see LVDS table.

Register-to-Register Performance (Continued)

Function	–5 Timing	Units
DSP IP Functions		
16-Tap Fully-Parallel FIR Filter	136	MHz
1024-pt FFT	152	MHz
8 X 8 Matrix Multiplication	137	MHz

1. These timing numbers were generated using the ispLEVER design tool. Exact performance may vary with device, design and tool version. The tool uses internal parameters that have been characterized but are not tested on every device.

Timing v. A 0.12

Derating Timing Tables

Logic timing provided in the following sections of this data sheet and the Diamond design tools are worst case numbers in the operating range. Actual delays at nominal temperature and voltage for best case process, can be much better than the values given in the tables. The Diamond design software can provide logic timing numbers at a particular temperature and voltage.

LA-LatticeXP2 Internal Switching Characteristics[1]

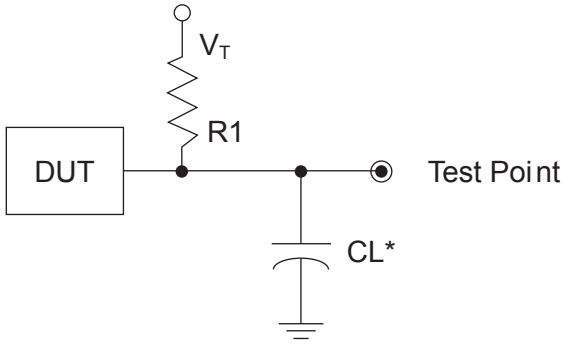
Over Recommended Operating Conditions

Parameter	Description	–5		Units
		Min.	Max.	
PFU/PFF Logic Mode Timing				
t_{LUT4_PFU}	LUT4 delay (A to D inputs to F output)	—	0.275	ns
t_{LUT6_PFU}	LUT6 delay (A to D inputs to OFX output)	—	0.522	ns
t_{LSR_PFU}	Set/Reset to output of PFU (Asynchronous)	—	0.865	ns
t_{SUM_PFU}	Clock to Mux (M0,M1) Input Setup Time	0.156	—	ns
t_{HM_PFU}	Clock to Mux (M0,M1) Input Hold Time	–0.056	—	ns
t_{SUD_PFU}	Clock to D input setup time	0.098	—	ns
t_{HD_PFU}	Clock to D input hold time	0.003	—	ns
t_{CK2Q_PFU}	Clock to Q delay, (D-type Register Configuration)	—	0.405	ns
t_{RSTREC_PFU}	Asynchronous reset recovery time for PFU Logic	—	0.791	ns
t_{RST_PFU}	Asynchronous reset time for PFU Logic	—	0.865	ns
PFU Dual Port Memory Mode Timing				
t_{CORAM_PFU}	Clock to Output (F Port)	—	1.535	ns
t_{SUDATA_PFU}	Data Setup Time	–0.290	—	ns
t_{HDATA_PFU}	Data Hold Time	0.330	—	ns
t_{SUADDR_PFU}	Address Setup Time	–0.392	—	ns
t_{HADDR_PFU}	Address Hold Time	0.392	—	ns
t_{SUWREN_PFU}	Write/Read Enable Setup Time	–0.204	—	ns
t_{HWREN_PFU}	Write/Read Enable Hold Time	0.219	—	ns
PIO Input/Output Buffer Timing				
t_{IN_PIO}	Input Buffer Delay (LVCMOS25)	—	0.708	ns
t_{OUT_PIO}	Output Buffer Delay (LVCMOS25)	—	1.308	ns
IOLOGIC Input/Output Timing				
t_{SUI_PIO}	Input Register Setup Time (Data Before Clock)	1.215	—	ns
t_{HI_PIO}	Input Register Hold Time (Data after Clock)	0.448	—	ns
t_{COO_PIO}	Output Register Clock to Output Delay	—	0.724	ns
t_{SUCE_PIO}	Input Register Clock Enable Setup Time	0.041	—	ns
t_{HCE_PIO}	Input Register Clock Enable Hold Time	–0.028	—	ns
t_{SULSR_PIO}	Set/Reset Setup Time	0.220	—	ns
t_{HLSR_PIO}	Set/Reset Hold Time	–0.094	—	ns
t_{RSTREC_PIO}	Asynchronous reset recovery time for IO Logic	0.269	—	ns
t_{RST_PIO}	Asynchronous reset time for PFU Logic	—	0.457	ns
t_{DEL}	Dynamic Delay Step Size	0.035	0.035	ns
EBR Timing				
t_{CO_EBR}	Clock (Read) to Output from Address or Data	—	3.552	ns
t_{COO_EBR}	Clock (Write) to Output from EBR Output Register	—	0.461	ns
t_{SUDATA_EBR}	Setup Data to EBR Memory (Write Clk)	–0.232	—	ns
t_{HDATA_EBR}	Hold Data to EBR Memory (Write Clk)	0.270	—	ns
t_{SUADDR_EBR}	Setup Address to EBR Memory (Write Clk)	–0.159	—	ns
t_{HADDR_EBR}	Hold Address to EBR Memory (Write Clk)	0.209	—	ns
t_{SUWREN_EBR}	Setup Write/Read Enable to EBR Memory (Write/Read Clk)	–0.184	—	ns

Switching Test Conditions

Figure 3-11 shows the output test load that is used for AC testing. The speci•c values for resistance, capacitance, voltage, and other test conditions are shown in Table 3-6.

Figure 3-11. Output Test Load, LVTTL and LVCMOS Standards



*CL Includes Test Fixture and Probe Capacitance

Table 3-6. Test Fixture Required Components, Non-Terminated Interfaces

Test Condition	R_1	R_2	C_L	Timing Ref.	V_T
LVTTL and other LVCMOS settings (L -> H, H -> L)	∞	∞	0pF	LVCMOS 3.3 = 1.5 V	—
				LVCMOS 2.5 = $V_{CCIO}/2$	—
				LVCMOS 1.8 = $V_{CCIO}/2$	—
				LVCMOS 1.5 = $V_{CCIO}/2$	—
				LVCMOS 1.2 = $V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z -> H)	∞	1MΩ		$V_{CCIO}/2$	—
LVCMOS 2.5 I/O (Z -> L)	1MΩ	∞		$V_{CCIO}/2$	V_{CCIO}
LVCMOS 2.5 I/O (H -> Z)	∞	100		V_{OH} – 0.10	—
LVCMOS 2.5 I/O (L -> Z)	100	∞		V_{OL} + 0.10	V_{CCIO}

Note: Output test conditions for all other interfaces are determined by the respective standards.

Signal Descriptions

Signal Name	I/O	Description
General Purpose		
P[Edge] [Row/Column Number*]_[A/B]	I/O	[Edge] indicates the edge of the device on which the pad is located. Valid edge designations are L (Left), B (Bottom), R (Right), T (Top). [Row/Column Number] indicates the PFU row or the column of the device on which the PIC exists. When Edge is T (Top) or B (Bottom), only need to specify Row Number. When Edge is L (Left) or R (Right), only need to specify Column Number. [A/B] indicates the PIO within the PIC to which the pad is connected. Some of these user-programmable pins are shared with special function pins. These pins, when not used as special purpose pins, can be programmed as I/Os for user logic. During configuration the user-programmable I/Os are tri-stated with an internal pull-up resistor enabled. If any pin is not used (or not bonded to a package pin), it is also tri-stated with an internal pull-up resistor enabled after configuration.
GSRN	I	Global RESET signal (active low). Any I/O pin can be GSRN.
NC	—	No connect.
GND	—	Ground. Dedicated pins.
V_{CC}	—	Power supply pins for core logic. Dedicated pins.
V_{CCAUX}	—	Auxiliary power supply pin. This dedicated pin powers all the differential and referenced input buffers.
V_{CCPLL}	—	PLL supply pins. csBGA, PQFP and TQFP packages only.
V_{CCIOx}	—	Dedicated power supply pins for I/O bank x.
V_{REF1_x}, V_{REF2_x}	—	Reference supply pins for I/O bank x. Pre-determined pins in each bank are assigned as V_{REF} inputs. When not used, they may be used as I/O pins.
PLL and Clock Functions (Used as user programmable I/O pins when not in use for PLL or clock pins)		
[LOC][num]_V_{CCPLL}	—	Power supply pin for PLL: LLC, LRC, URC, ULC, num = row from center.
[LOC][num]_GPLL[T, C]_IN_A	I	General Purpose PLL (GPLL) input pads: LLC, LRC, URC, ULC, num = row from center, T = true and C = complement, index A,B,C...at each side.
[LOC][num]_GPLL[T, C]_FB_A	I	Optional feedback GPLL input pads: LLC, LRC, URC, ULC, num = row from center, T = true and C = complement, index A,B,C...at each side.
PCLK[T, C]_[n:0]_[3:0]	I	Primary Clock pads, T = true and C = complement, n per side, indexed by bank and 0,1,2,3 within bank.
[LOC]DQS[num]	I	DQS input pads: T (Top), R (Right), B (Bottom), L (Left), DQS, num = ball function number. Any pad can be configured to be output.
Test and Programming (Dedicated Pins)		
TMS	I	Test Mode Select input, used to control the 1149.1 state machine. Pull-up is enabled during configuration.
TCK	I	Test Clock input pin, used to clock the 1149.1 state machine. No pull-up enabled.
TDI	I	Test Data in pin. Used to load data into device using 1149.1 state machine. After power-up, this TAP port can be activated for configuration by sending appropriate command. (Note: once a configuration port is selected it is locked. Another configuration port cannot be selected until the power-up sequence). Pull-up is enabled during configuration.

Pin Information Summary (Continued)

Pin Type		LA-XP2-5				LA-XP2-8				LA-XP2-17	
		132 csBGA	144 TQFP	208 PQFP	256 ftBGA	132 csBGA	144 TQFP	208 PQFP	256 ftBGA	208 PQFP	256 ftBGA
DDR Banks Bonding Out per I/O Bank[1]	Bank0	1	1	1	1	1	1	1	1	1	1
	Bank1	0	0	1	1	0	0	1	1	1	1
	Bank2	1	1	1	1	1	1	1	1	1	1
	Bank3	0	0	1	1	0	0	1	1	1	1
	Bank4	0	0	1	1	0	0	1	1	1	1
	Bank5	1	1	1	1	1	1	1	1	1	1
	Bank6	0	0	1	1	0	0	1	1	1	1
	Bank7	1	1	1	1	1	1	1	1	1	1
PCI capable I/Os Bonding Out per Bank	Bank0	18	20	20	26	18	20	20	28	20	28
	Bank1	4	6	18	18	4	6	18	22	18	22
	Bank2	0	0	0	0	0	0	0	0	0	0
	Bank3	0	0	0	0	0	0	0	0	0	0
	Bank4	8	8	18	18	8	8	18	26	18	26
	Bank5	14	18	20	24	14	18	20	24	20	24
	Bank6	0	0	0	0	0	0	0	0	0	0
	Bank7	0	0	0	0	0	0	0	0	0	0

1. Minimum requirement to implement a fully functional 8-bit wide DDR bus. Available DDR interface consists of at least 12 I/Os (1 DQS + 1 DQSB + 8 DQs + 1 DM + Bank VREF1).

Logic Signal Connections

Package pinout information can be found on the LatticeXP2 product pages on the Lattice website at www.latticesemi.com/products/fpga/xp2 and in the Lattice Diamond design software.

Thermal Management

Thermal management is recommended as part of any sound FPGA design methodology. To assess the thermal characteristics of a system, Lattice specifies a maximum allowable junction temperature in all device data sheets. Designers must complete a thermal analysis of their specific design to ensure that the device and package do not exceed the junction temperature limits. Refer to the Thermal Management document to find the device/package-specific thermal values.

For Further Information

- TN1139 - Power Estimation and Management for LatticeXP2 Devices
- Power Calculator tool included with Lattice Diamond design software or as a standalone download from www.latticesemi.com/software