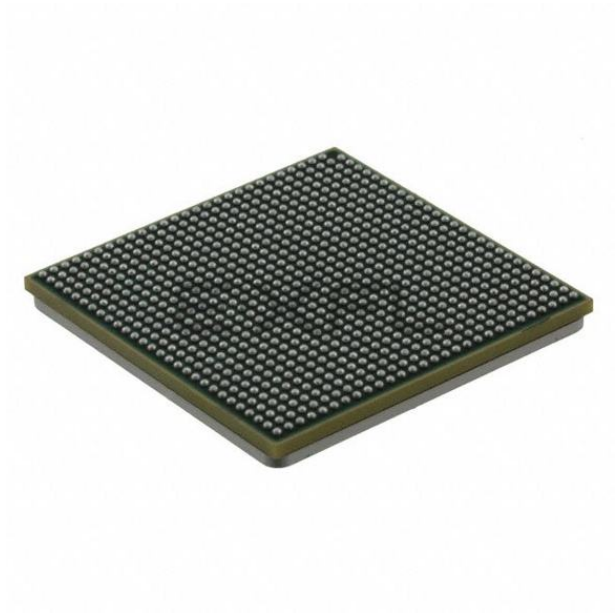

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Understanding Embedded - Microprocessors

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Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.2GHz
Co-Processors/DSP	Signal Processing; SPE, Security; SEC
RAM Controllers	DDR, DDR2, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc8547ehxatg

7 DUART

This section describes the DC and AC electrical specifications for the DUART interface of the device.

7.1 DUART DC Electrical Characteristics

This table provides the DC electrical characteristics for the DUART interface.

Table 20. DUART DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.8	V
Input current (V_{IN}[1] = 0 V or V_{IN} = V_{DD})	I_{IN}	—	±5	μA
High-level output voltage (OV_{DD} = min, I_{OH} = –2 mA)	V_{OH}	2.4	—	V
Low-level output voltage (OV_{DD} = min, I_{OL} = 2 mA)	V_{OL}	—	0.4	V

Note:
1. Note that the symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

7.2 DUART AC Electrical Specifications

This table provides the AC timing parameters for the DUART interface.

Table 21. DUART AC Timing Specifications

Parameter	Value	Unit	Notes
Minimum baud rate	f_{CCB}/1,048,576	baud	1, 2
Maximum baud rate	f_{CCB}/16	baud	1, 2, 3
Oversample rate	16	—	1, 4

Notes:
1. Guaranteed by design.
2. f_{CCB} refers to the internal platform clock.
3. Actual attainable baud rate is limited by the latency of interrupt processing.
4. The middle of a start bit is detected as the 8^{th} sampled 0 after the 1-to-0 transition of the start bit. Subsequent bit values are sampled each 16^{th} sample.

Table 34. RMII Transmit AC Timing Specifications (continued)

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
TSEC*n*_TX_CLK to RMII data TXD[1:0], TX_EN delay	t_{RMTDX}	1.0	—	10.0	ns

Note:
1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MTKHDX} symbolizes MII transmit timing (MT) for the time t_{MTX} clock reference (K) going high (H) until data outputs (D) are invalid (X). Note that, in general, the clock reference symbol representation is based on two to three letters representing the clock of a particular functional. For example, the subscript of t_{MTX} represents the MII(M) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

Figure 18 shows the RMII transmit AC timing diagram.

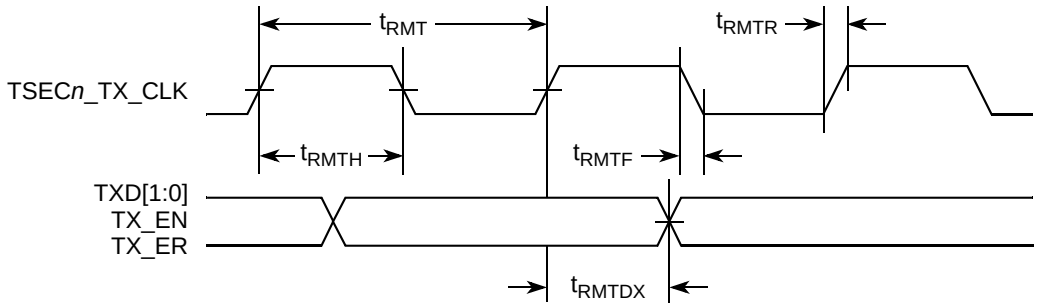

Figure 18. RMII Transmit AC Timing Diagram

8.2.7.2 RMII Receive AC Timing Specifications

Table 35. RMII Receive AC Timing Specifications

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
TSEC*n*_TX_CLK clock period	t_{RMR}	15.0	20.0	25.0	ns
TSEC*n*_TX_CLK duty cycle	t_{RMRH}	35	50	65	%
TSEC*n*_TX_CLK peak-to-peak jitter	t_{RMRJ}	—	—	250	ps
Rise time TSEC*n*_TX_CLK(20%–80%)	t_{RMRR}	1.0	—	2.0	ns
Fall time TSEC*n*_TX_CLK (80%–20%)	t_{RMRF}	1.0	—	2.0	ns
RXD[1:0], CRS_DV, RX_ER setup time to REF_CLK rising edge	t_{RMRDV}	4.0	—	—	ns
RXD[1:0], CRS_DV, RX_ER hold time to REF_CLK rising edge	t_{RMRDX}	2.0	—	—	ns

Note:
1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MRDVKH} symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MRX} clock reference (K) going to the high (H) state or setup time. Also, t_{MRDXKL} symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{MRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{MRX} represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

NOTE

PLL bypass mode is required when LBIU frequency is at or below 83 MHz. When LBIU operates above 83 MHz, LBIU PLL is recommended to be enabled.

Figure 23 through Figure 28 show the local bus signals.

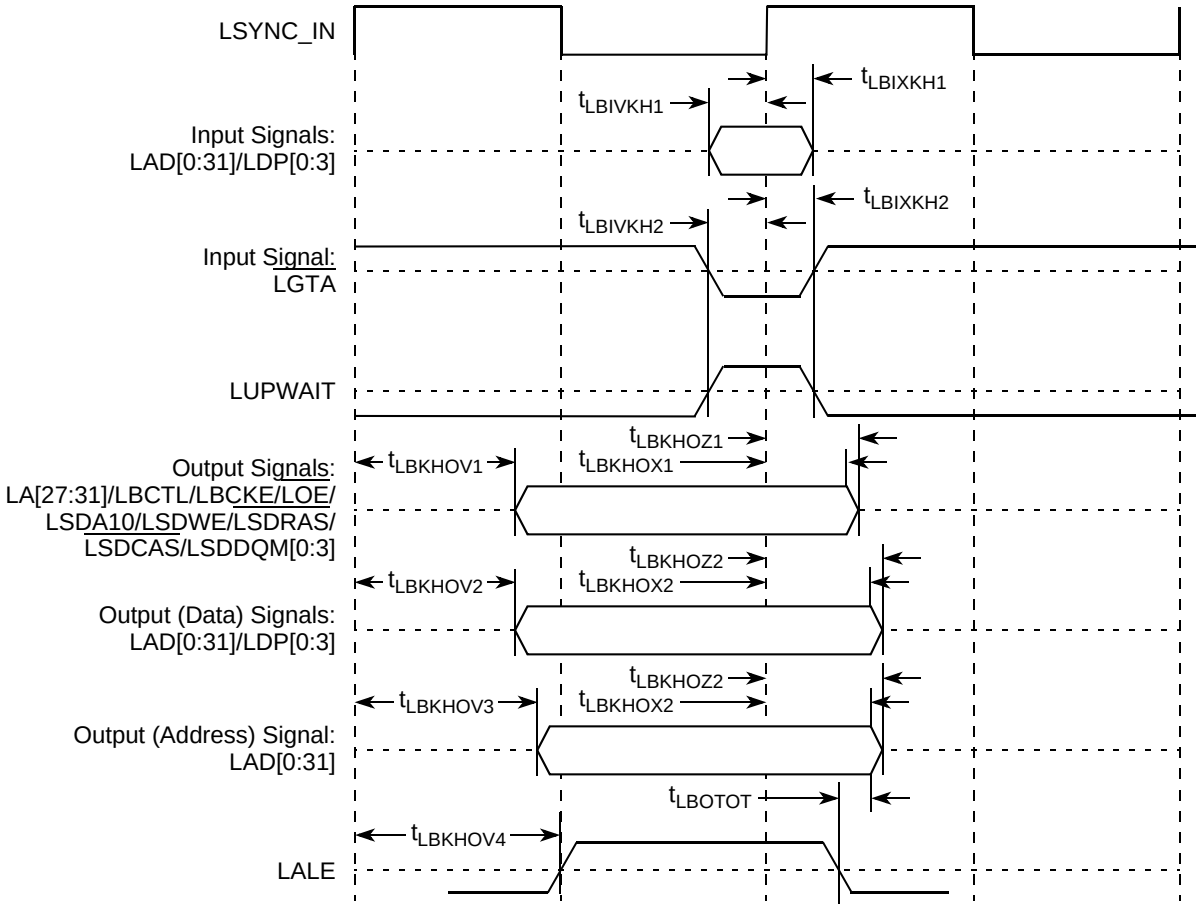

Figure 23. Local Bus Signals (PLL Enabled)

This table describes the timing parameters of the local bus interface at BV_{DD} = 3.3 V with PLL disabled.

Table 42. Local Bus Timing Parameters—PLL Bypassed

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	12	—	ns	2
Local bus duty cycle	t_{LBKH}/t_{LBK}	43	57	%	—
Internal launch/capture clock to LCLK delay	t_{LBKHKT}	2.3	4.4	ns	8
Input setup to local bus clock (except $\overline{\text{LGTA}}$/LUPWAIT)	$t_{LBIVKH1}$	6.2	—	ns	4, 5
$\overline{\text{LGTA}}$/LUPWAIT input setup to local bus clock	$t_{LBIVKL2}$	6.1	—	ns	4, 5
Input hold from local bus clock (except $\overline{\text{LGTA}}$/LUPWAIT)	$t_{LBIXKH1}$	–1.8	—	ns	4, 5

Table 42. Local Bus Timing Parameters—PLL Bypassed (continued)

Parameter	Symbol[1]	Min	Max	Unit	Notes
$\overline{\text{LGTA}}$/LUPWAIT input hold from local bus clock	$t_{LBIXKL2}$	–1.3	—	ns	4, 5
LALE output transition to LAD/LDP output transition (LATCH hold time)	t_{LBOTOT}	1.5	—	ns	6
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKLOV1}$	—	–0.3	ns	—
Local bus clock to data valid for LAD/LDP	$t_{LBKLOV2}$	—	–0.1	ns	4
Local bus clock to address valid for LAD	$t_{LBKLOV3}$	—	0	ns	4
Local bus clock to LALE assertion	$t_{LBKLOV4}$	—	0	ns	4
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKLOX1}$	–3.7	—	ns	4
Output hold from local bus clock for LAD/LDP	$t_{LBKLOX2}$	–3.7	—	ns	4
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$t_{LBKLOZ1}$	—	0.2	ns	7
Local bus clock to output high impedance for LAD/LDP	$t_{LBKLOZ2}$	—	0.2	ns	7

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to local bus clock for PLL bypass mode. Timings may be negative with respect to the local bus clock because the actual launch and capture of signals is done with the internal launch/capture clock, which precedes LCLK by t_{LBKHKT}.
3. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[*n*]. Skew measured between complementary signals at $BV_{DD}/2$.
4. All signals are measured from $BV_{DD}/2$ of the rising edge of local bus clock for PLL bypass mode to $0.4 \times BV_{DD}$ of the signal in question for 3.3-V signaling levels.
5. Input timings are measured at the pin.
6. The value of t_{LBOTOT} is the measurement of the minimum time between the negation of LALE and any change in LAD.
7. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
8. Guaranteed by characterization.
9. Guaranteed by design.

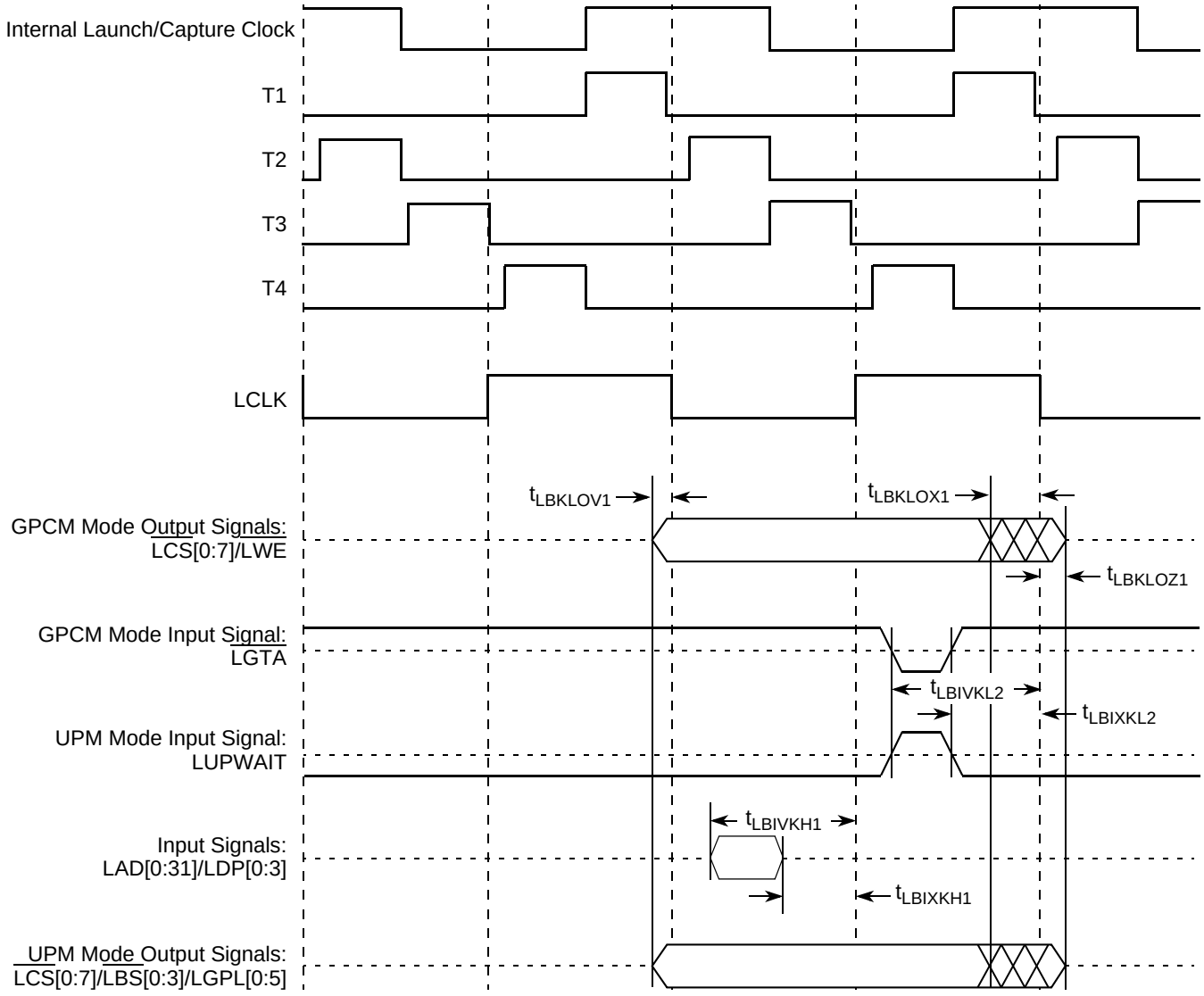

Figure 28. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 8 or 16 (PLL Bypass Mode)

Table 50. GP_{IN} DC Electrical Characteristics (2.5 V DC)

Parameter	Symbol	Min	Max	Unit
Supply voltage 2.5 V	BV_{DD}	2.37	2.63	V
High-level input voltage	V_{IH}	1.70	BV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.7	V
Input current (BV_{IN} [1] = 0 V or BV_{IN} = BV_{DD})	I_{IH}	—	10	μA

Note:
1. The symbol BV_{IN}, in this case, represents the BV_{IN} symbol referenced in Table 1.

15 PCI/PCI-X

This section describes the DC and AC electrical specifications for the PCI/PCI-X bus of the device.

Note that the maximum PCI-X frequency in synchronous mode is 110 MHz.

15.1 PCI/PCI-X DC Electrical Characteristics

This table provides the DC electrical characteristics for the PCI/PCI-X interface.

Table 51. PCI/PCI-X DC Electrical Characteristics[1]

Parameter	Symbol	Min	Max	Unit	Notes
High-level input voltage	V_{IH}	2	OV_{DD} + 0.3	V	—
Low-level input voltage	V_{IL}	–0.3	0.8	V	—
Input current (V_{IN} = 0 V or V_{IN} = V_{DD})	I_{IN}	—	±5	μA	2
High-level output voltage (OV_{DD} = min, I_{OH} = –2 mA)	V_{OH}	2.4	—	V	—
Low-level output voltage (OV_{DD} = min, I_{OL} = 2 mA)	V_{OL}	—	0.4	V	—

Notes:
1. Ranges listed do not meet the full range of the DC specifications of the *PCI 2.2 Local Bus Specifications*.
2. The symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

15.2 PCI/PCI-X AC Electrical Specifications

This section describes the general AC timing parameters of the PCI/PCI-X bus. Note that the clock reference CLK is represented by SYSCLK when the PCI controller is configured for synchronous mode and by PCI*n*_CLK when it is configured for asynchronous mode.

Table 53. PCI-X AC Timing Specifications at 66 MHz (continued)

Parameter	Symbol	Min	Max	Unit	Notes
$\overline{\text{HRESET}}$ to PCI-X initialization pattern hold time	t_{PCRHIX}	0	50	ns	6, 11

Notes:

1. See the timing measurement conditions in the *PCI-X 1.0a Specification*.
2. Minimum times are measured at the package pin (not the test point). Maximum times are measured with the test point and load circuit.
3. Setup time for point-to-point signals applies to $\overline{\text{REQ}}$ and $\overline{\text{GNT}}$ only. All other signals are bused.
4. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
5. Setup time applies only when the device is not driving the pin. Devices cannot drive and receive signals at the same time.
6. Maximum value is also limited by delay to the first transaction (time for $\overline{\text{HRESET}}$ high to first configuration access, t_{PCRHFV}). The PCI-X initialization pattern control signals after the rising edge of $\overline{\text{HRESET}}$ must be negated no later than two clocks before the first $\overline{\text{FRAME}}$ and must be floated no later than one clock before $\overline{\text{FRAME}}$ is asserted.
7. A PCI-X device is permitted to have the minimum values shown for t_{PCKHOV} and t_{CYC} only in PCI-X mode. In conventional mode, the device must meet the requirements specified in PCI 2.2 for the appropriate clock frequency.
8. Device must meet this specification independent of how many outputs switch simultaneously.
9. The timing parameter t_{PCRHFV} is a minimum of 10 clocks rather than the minimum of 5 clocks in the *PCI-X 1.0a Specification.*
10. Guaranteed by characterization.
11. Guaranteed by design.

This table provides the PCI-X AC timing specifications at 133 MHz. Note that the maximum PCI-X frequency in synchronous mode is 110 MHz.

Table 54. PCI-X AC Timing Specifications at 133 MHz

Parameter	Symbol	Min	Max	Unit	Notes
SYSCLK to signal valid delay	t_{PCKHOV}	—	3.8	ns	1, 2, 3, 7, 8
Output hold from SYSCLK	t_{PCKHOX}	0.7	—	ns	1, 11
SYSCLK to output high impedance	t_{PCKHOZ}	—	7	ns	1, 4, 8, 12
Input setup time to SYSCLK	t_{PCIVKH}	1.2	—	ns	3, 5, 9, 11
Input hold time from SYSCLK	t_{PCIXKH}	0.5	—	ns	11
$\overline{\text{REQ64}}$ to $\overline{\text{HRESET}}$ setup time	t_{PCRVRH}	10	—	clocks	12
$\overline{\text{HRESET}}$ to $\overline{\text{REQ64}}$ hold time	t_{PCRHRX}	0	50	ns	12
$\overline{\text{HRESET}}$ high to first $\overline{\text{FRAME}}$ assertion	t_{PCRHFV}	10	—	clocks	10, 12
PCI-X initialization pattern to $\overline{\text{HRESET}}$ setup time	t_{PCIVRH}	10	—	clocks	12

Table 54. PCI-X AC Timing Specifications at 133 MHz (continued)

Parameter	Symbol	Min	Max	Unit	Notes
$\overline{\text{HRESET}}$ to PCI-X initialization pattern hold time	t_{PCRHIX}	0	50	ns	6, 12

Notes:

1. See the timing measurement conditions in the *PCI-X 1.0a Specification*.
2. Minimum times are measured at the package pin (not the test point). Maximum times are measured with the test point and load circuit.
3. Setup time for point-to-point signals applies to $\overline{\text{REQ}}$ and $\overline{\text{GNT}}$ only. All other signals are bused.
4. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
5. Setup time applies only when the device is not driving the pin. Devices cannot drive and receive signals at the same time.
6. Maximum value is also limited by delay to the first transaction (time for $\overline{\text{HRESET}}$ high to first configuration access, t_{PCRHFV}). The PCI-X initialization pattern control signals after the rising edge of $\overline{\text{HRESET}}$ must be negated no later than two clocks before the first $\overline{\text{FRAME}}$ and must be floated no later than one clock before $\overline{\text{FRAME}}$ is asserted.
7. A PCI-X device is permitted to have the minimum values shown for t_{PCKHOV} and t_{CYC} only in PCI-X mode. In conventional mode, the device must meet the requirements specified in PCI 2.2 for the appropriate clock frequency.
8. Device must meet this specification independent of how many outputs switch simultaneously.
9. The timing parameter t_{PCIVKH} is a minimum of 1.4 ns rather than the minimum of 1.2 ns in the *PCI-X 1.0a Specification.*
10. The timing parameter t_{PCRHFV} is a minimum of 10 clocks rather than the minimum of 5 clocks in the *PCI-X 1.0a Specification.*
11. Guaranteed by characterization.
12. Guaranteed by design.

— The input amplitude of the differential clock must be between 400 and 1600 mV differential peak-peak (or between 200 and 800 mV differential peak). In other words, each signal wire of the differential pair must have a single-ended swing less than 800 mV and greater than 200 mV. This requirement is the same for both external DC- or AC-coupled connection.

— For external DC-coupled connection, as described in Section 16.2.1, "SerDes Reference Clock Receiver Characteristics," the maximum average current requirements sets the requirement for average voltage (common mode voltage) to be between 100 and 400 mV. Figure 40 shows the SerDes reference clock input requirement for DC-coupled connection scheme.

— For external AC-coupled connection, there is no common mode voltage requirement for the clock driver. Since the external AC-coupling capacitor blocks the DC level, the clock driver and the SerDes reference clock receiver operate in different command mode voltages. The SerDes reference clock receiver in this connection scheme has its common mode voltage set to SGND_SRDS*n*. Each signal wire of the differential inputs is allowed to swing below and above the command mode voltage (SGND_SRDS*n*). Figure 41 shows the SerDes reference clock input requirement for AC-coupled connection scheme.

- Single-ended mode
  — The reference clock can also be single-ended. The SD_REF_CLK input amplitude (single-ended swing) must be between 400 and 800 mV peak-to-peak (from V_{min} to V_{max}) with $\overline{\text{SD\_REF\_CLK}}$ either left unconnected or tied to ground.
  — The SD_REF_CLK input average voltage must be between 200 and 400 mV. Figure 42 shows the SerDes reference clock input requirement for single-ended signaling mode.
  — To meet the input amplitude requirement, the reference clock inputs might need to be DC- or AC-coupled externally. For the best noise performance, the reference of the clock could be DC- or AC-coupled into the unused phase ($\overline{\text{SD\_REF\_CLK}}$) through the same source impedance as the clock input (SD_REF_CLK) in use.

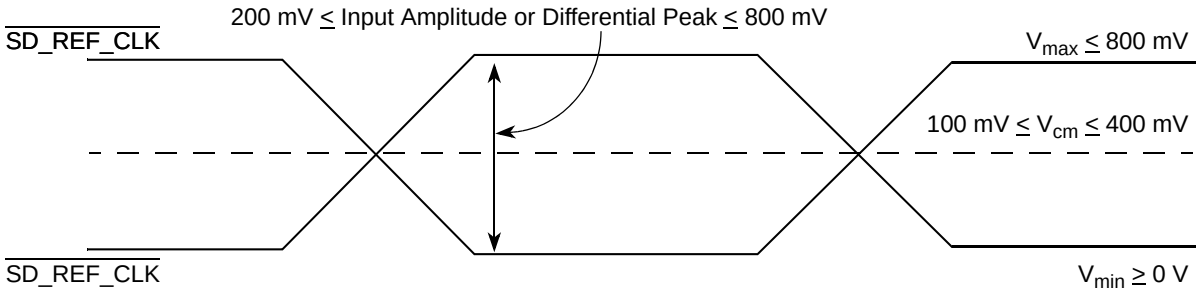

Figure 40. Differential Reference Clock Input DC Requirements (External DC-Coupled)

17 PCI Express

This section describes the DC and AC electrical specifications for the PCI Express bus of the MPC8548E.

17.1 DC Requirements for PCI Express SD_REF_CLK and SD_REF_CLK

For more information, see Section 16.2, "SerDes Reference Clocks."

17.2 AC Requirements for PCI Express SerDes Clocks

Table 55 lists the AC requirements for the PCI Express SerDes clocks.

Table 55. SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ AC Requirements

Symbol	Parameter Description	Min	Typ	Max	Unit	Notes
t_{REF}	REFCLK cycle time	—	10	—	ns	1
t_{REFCJ}	REFCLK cycle-to-cycle jitter. Difference in the period of any two adjacent REFCLK cycles.	—	—	100	ps	—
t_{REFPJ}	Phase jitter. Deviation in edge location with respect to mean edge location.	–50	—	50	ps	—

Note:
1. Typical based on *PCI Express Specification 2.0*.

17.3 Clocking Dependencies

The ports on the two ends of a link must transmit data at a rate that is within 600 parts per million (ppm) of each other at all times. This is specified to allow bit rate clock sources with a ±300 ppm tolerance.

17.4 Physical Layer Specifications

The following is a summary of the specifications for the physical layer of PCI Express on this device. For further details as well as the specifications of the transport and data link layer see *PCI Express Base Specification. Rev. 1.0a*.

17.4.1 Differential Transmitter (TX) Output

Table 56 defines the specifications for the differential output at all transmitters (TXs). The parameters are specified at the component pins.

Table 57. Differential Receiver (RX) Input Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Unit	Comments
$L_{TX-SKEW}$	Total Skew	—	—	20	ns	Skew across all lanes on a Link. This includes variation in the length of SKP ordered set (for example, COM and one to five symbols) at the RX as well as any delay differences arising from the interconnect itself.

Notes:

1. No test load is necessarily associated with this value.
2. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in Figure 50 must be used as the RX device when taking measurements (also see the receiver compliance eye diagram shown in Figure 49). If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
3. A T_{RX-EYE} = 0.40 UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX-EYE-MEDIAN-to-MAX-JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive TX UIs. Note that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
4. The receiver input impedance shall result in a differential return loss greater than or equal to 15 dB with the D+ line biased to 300 mV and the D– line biased to –{300 mV and a common mode return loss greater than or equal to 6 dB (no bias required) over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements for is 50 Ω to ground for both the D+ and D– line (that is, as measured by a vector network analyzer with 50-Ω probes—see Figure 50). Note: that the series capacitors CTX is optional for the return loss measurement.
5. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM) there is a 5 ms transition time before receiver termination values must be met on all unconfigured lanes of a port.
6. The RX DC common mode Impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the RX ground.
7. It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

17.5 Receiver Compliance Eye Diagrams

The RX eye diagram in Figure 49 is specified using the passive compliance/test measurement load (see Figure 50) in place of any real PCI Express RX component.

Note: In general, the minimum receiver eye diagram measured with the compliance/test measurement load (see Figure 50) is larger than the minimum receiver eye diagram measured over a range of systems at the input receiver of any real PCI Express component. The degraded eye diagram at the input receiver is due to traces internal to the package as well as silicon parasitic characteristics which cause the real PCI Express component to vary in impedance from the compliance/test measurement load. The input receiver eye diagram is implementation specific and is not specified. RX component designer must provide additional margin to adequately compensate for the degraded minimum receiver eye diagram (shown in Figure 49) expected at the input receiver based on some adequate combination of system simulations and the return loss measured looking into the RX package and silicon. The RX eye diagram must be aligned in time using the jitter median to locate the center of the eye diagram.

18.3 Signal Definitions

LP-serial links use differential signaling. This section defines terms used in the description and specification of differential signals. Figure 51 shows how the signals are defined. The figures show waveforms for either a transmitter output (TD and $\overline{\text{TD}}$) or a receiver input (RD and $\overline{\text{RD}}$). Each signal swings between A volts and B volts where A > B. Using these waveforms, the definitions are as follows:

1. The transmitter output signals and the receiver input signals TD, $\overline{\text{TD}}$, RD, and $\overline{\text{RD}}$ each have a peak-to-peak swing of A – B volts.
2. The differential output signal of the transmitter, V_{OD}, is defined as $V_{TD} - V_{\overline{TD}}$.
3. The differential input signal of the receiver, V_{ID}, is defined as $V_{RD} - V_{\overline{RD}}$.
4. The differential output signal of the transmitter and the differential input signal of the receiver each range from A – B to –(A – B) volts.
5. The peak value of the differential transmitter output signal and the differential receiver input signal is A – B volts.
6. The peak-to-peak value of the differential transmitter output signal and the differential receiver input signal is 2 × (A – B) volts.

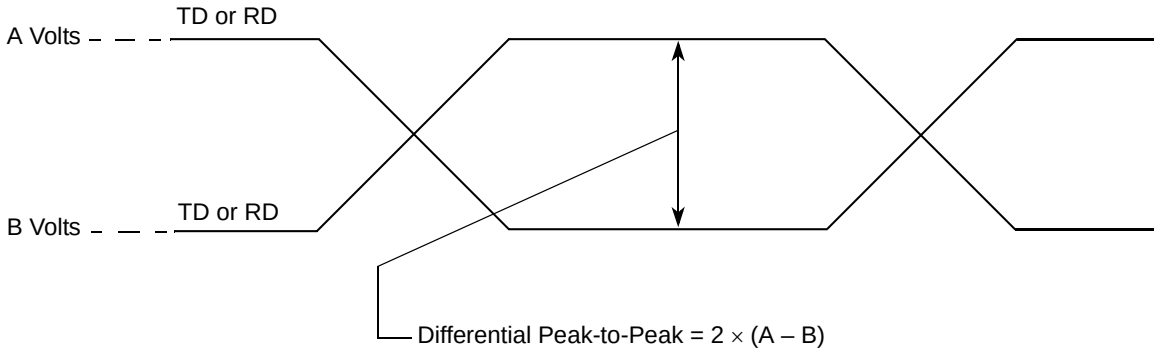

Figure 51. Differential Peak–Peak Voltage of Transmitter or Receiver

To illustrate these definitions using real values, consider the case of a CML (current mode logic) transmitter that has a common mode voltage of 2.25 V and each of its outputs, TD and $\overline{\text{TD}}$, has a swing that goes between 2.5 and 2.0 V. Using these values, the peak-to-peak voltage swing of the signals TD and $\overline{\text{TD}}$ is 500 mVp-p. The differential output signal ranges between 500 and –500 mV. The peak differential voltage is 500 mV. The peak-to-peak differential voltage is 1000 mVp-p.

18.4 Equalization

With the use of high-speed serial links, the interconnect media causes degradation of the signal at the receiver. Effects such as inter-symbol interference (ISI) or data dependent jitter are produced. This loss can be large enough to degrade the eye opening at the receiver beyond what is allowed in the specification. To negate a portion of these effects, equalization can be used. The most common equalization techniques that can be used are:

- A passive high pass filter network placed at the receiver. This is often referred to as passive equalization.
- The use of active circuits in the receiver. This is often referred to as adaptive equalization.

Table 71. MPC8548E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
SENSEVSS	M16	—	—	13
Analog Signals				
MVREF	A18	I Reference voltage signal for DDR	MVREF	—
SD_IMP_CAL_RX	L28	I	200Ω to GND	—
SD_IMP_CAL_TX	AB26	I	100Ω to GND	—
SD_PLL_TPA	U26	O	—	24

Notes:

1. All multiplexed signals are listed only once and do not re-occur. For example, $\overline{\text{LCS5}}$/$\overline{\text{DMA\_REQ2}}$ is listed only once in the local bus controller section, and is not mentioned in the DMA section even though the pin also functions as $\overline{\text{DMA\_REQ2}}$.
2. Recommend a weak pull-up resistor (2–10 kΩ) be placed on this pin to OV_{DD}.
3. A valid clock must be provided at POR if TSEC4_TXD[2] is set = 1.
4. This pin is an open drain signal.
5. This pin is a reset configuration pin. It has a weak internal pull-up P-FET which is enabled only when the processor is in the reset state. This pull-up is designed such that it can be overpowered by an external 4.7-kΩ pull-down resistor. However, if the signal is intended to be high after reset, and if there is any device on the net which might pull down the value of the net at reset, then a pullup or active driver is needed.
6. Treat these pins as no connects (NC) unless using debug address functionality.
7. The value of LA[28:31] during reset sets the CCB clock to SYSCLK PLL ratio. These pins require 4.7-kΩ pull-up or pull-down resistors. See Section 20.2, “CCB/SYSCLK PLL Ratio.”
8. The value of LALE, LGPL2, and LBCTL at reset set the e500 core clock to CCB clock PLL ratio. These pins require 4.7-kΩ pull-up or pull-down resistors. See the Section 20.3, “e500 Core PLL Ratio.”
9. Functionally, this pin is an output, but structurally it is an I/O because it either samples configuration input during reset or because it has other manufacturing test functions. This pin therefore is described as an I/O for boundary scan.
10. This pin functionally requires a pull-up resistor, but during reset it is a configuration input that controls 32- vs. 64-bit PCI operation. Therefore, it must be actively driven low during reset by reset logic if the device is to be configured to be a 64-bit PCI device. See the *PCI Specification*.
11. This output is actively driven during reset rather than being three-stated during reset.
12. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
13. These pins are connected to the V_{DD}/GND planes internally and may be used by the core power supply to improve tracking and regulation.
14. Internal thermally sensitive resistor.
15. No connections must be made to these pins if they are not used.
16. These pins are not connected for any use.
17. PCI specifications recommend that a weak pull-up resistor (2–10 kΩ) be placed on the higher order pins to OV_{DD} when using 64-bit buffer mode (pins PCI_AD[63:32] and $\overline{\text{PCI1\_C\_BE}}$[7:4]).
19. If this pin is connected to a device that pulls down during reset, an external pull-up is required to drive this pin to a safe state during reset.
20. This pin is only an output in FIFO mode when used as Rx flow control.
24. Do not connect.

Table 72. MPC8547E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
Local Bus Controller Interface				
LAD[0:31]	E27, B20, H19, F25, A20, C19, E28, J23, A25, K22, B28, D27, D19, J22, K20, D28, D25, B25, E22, F22, F21, C25, C22, B23, F20, A23, A22, E19, A21, D21, F19, B21	I/O	BV_{DD}	—
LDP[0:3]	K21, C28, B26, B22	I/O	BV_{DD}	—
LA[27]	H21	O	BV_{DD}	5, 9
LA[28:31]	H20, A27, D26, A28	O	BV_{DD}	5, 7, 9
$\overline{\text{LCS}}$[0:4]	J25, C20, J24, G26, A26	O	BV_{DD}	—
$\overline{\text{LCS5}}$/$\overline{\text{DMA\_DREQ2}}$	D23	I/O	BV_{DD}	1
$\overline{\text{LCS6}}$/$\overline{\text{DMA\_DACK2}}$	G20	O	BV_{DD}	1
$\overline{\text{LCS7}}$/$\overline{\text{DMA\_DDONE2}}$	E21	O	BV_{DD}	1
$\overline{\text{LWE0}}$/$\overline{\text{LBS0}}$/LSDDQM[0]	G25	O	BV_{DD}	5, 9
$\overline{\text{LWE1}}$/$\overline{\text{LBS1}}$/LSDDQM[1]	C23	O	BV_{DD}	5, 9
$\overline{\text{LWE2}}$/$\overline{\text{LBS2}}$/LSDDQM[2]	J21	O	BV_{DD}	5, 9
$\overline{\text{LWE3}}$/$\overline{\text{LBS3}}$/LSDDQM[3]	A24	O	BV_{DD}	5, 9
LALE	H24	O	BV_{DD}	5, 8, 9
LBCTL	G27	O	BV_{DD}	5, 8, 9
LGPL0/LSDA10	F23	O	BV_{DD}	5, 9
LGPL1/$\overline{\text{LSDWE}}$	G22	O	BV_{DD}	5, 9
LGPL2/$\overline{\text{LOE}}$/$\overline{\text{LSDRAS}}$	B27	O	BV_{DD}	5, 8, 9
LGPL3/$\overline{\text{LSDCAS}}$	F24	O	BV_{DD}	5, 9
LGPL4/$\overline{\text{LGTA}}$/LUPWAIT/LPBSE	H23	I/O	BV_{DD}	—
LGPL5	E26	O	BV_{DD}	5, 9
LCKE	E24	O	BV_{DD}	—
LCLK[0:2]	E23, D24, H22	O	BV_{DD}	—
LSYNC_IN	F27	I	BV_{DD}	—
LSYNC_OUT	F28	O	BV_{DD}	—
DMA				
$\overline{\text{DMA\_DACK}}$[0:1]	AD3, AE1	O	OV_{DD}	5, 9, 107
$\overline{\text{DMA\_DREQ}}$[0:1]	AD4, AE2	I	OV_{DD}	—
$\overline{\text{DMA\_DDONE}}$[0:1]	AD2, AD1	O	OV_{DD}	—
Programmable Interrupt Controller				
$\overline{\text{UDE}}$	AH16	I	OV_{DD}	—
$\overline{\text{MCP}}$	AG19	I	OV_{DD}	—

Table 85. Package Thermal Characteristics for FC-PBGA (continued)

Characteristic	JEDEC Board	Symbol	Value	Unit	Notes
Die junction-to-board	N/A	$R_{\theta JB}$	5	°C/W	3
Die junction-to-case	N/A	$R_{\theta JC}$	0.8	°C/W	4

Notes:

1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, airflow, power dissipation of other components on the board, and board thermal resistance.
2. Per JEDEC JESD51-6 with the board (JESD51-7) horizontal.
3. Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
4. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1). The cold plate temperature is used for the case temperature, measured value includes the thermal resistance of the interface layer.

21.3 Heat Sink Solution

Every system application has different conditions that the thermal management solution must solve. As such, providing a recommended heat sink has not been found to be very useful. When a heat sink is chosen, give special consideration to the mounting technique. Mounting the heat sink to the printed-circuit board is the recommended procedure using a maximum of 10 lbs force (45 Newtons) perpendicular to the package and board. Clipping the heat sink to the package is not recommended.

22 System Design Information

This section provides electrical design recommendations for successful application of the device.

22.1 System Clocking

This device includes five PLLs, as follows:

1. The platform PLL generates the platform clock from the externally supplied SYSCLK input. The frequency ratio between the platform and SYSCLK is selected using the platform PLL ratio configuration bits as described in Section 20.2, “CCB/SYSCLK PLL Ratio.”
2. The e500 core PLL generates the core clock as a slave to the platform clock. The frequency ratio between the e500 core clock and the platform clock is selected using the e500 PLL ratio configuration bits as described in Section 20.3, “e500 Core PLL Ratio.”
3. The PCI PLL generates the clocking for the PCI bus.
4. The local bus PLL generates the clock for the local bus.
5. There is a PLL for the SerDes block.

22.2 PLL Power Supply Filtering

Each of the PLLs listed above is provided with power through independent power supply pins (AV_{DD}_PLAT, AV_{DD}_CORE, AV_{DD}_PCI, AV_{DD}_LBIU, and AV_{DD}_SRDS, respectively). The AV_{DD}

level must always be equivalent to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide independent filter circuits per PLL power supply as illustrated in Figure 57, one to each of the AV_{DD} pins. By providing independent filters to each PLL the opportunity to cause noise injection from one PLL to the other is reduced.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It must be built with surface mount capacitors with minimum Effective Series Inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit must be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It must be routed directly from the capacitors to the AV_{DD} pin, which is on the periphery of the footprint, without the inductance of vias.

Figure 57 through Figure 59 shows the PLL power supply filter circuits.

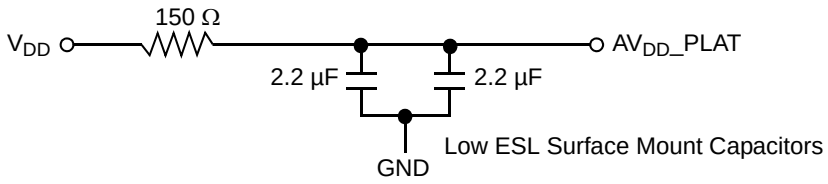

Figure 57. PLL Power Supply Filter Circuit with PLAT Pins

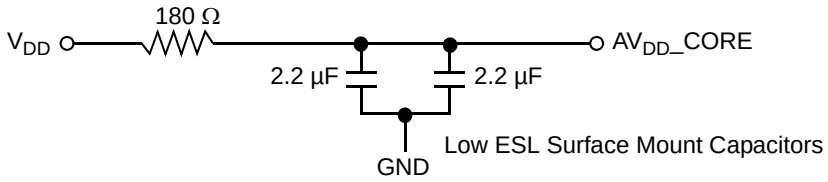

Figure 58. PLL Power Supply Filter Circuit with CORE Pins

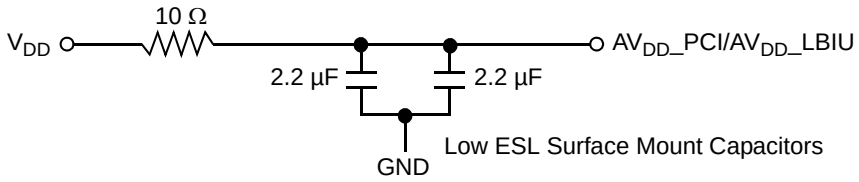

Figure 59. PLL Power Supply Filter Circuit with PCI/LBIU Pins

The AV_{DD}_SRDS signal provides power for the analog portions of the SerDes PLL. To ensure stability of the internal clock, the power supplied to the PLL is filtered using a circuit similar to the one shown in following figure. For maximum effectiveness, the filter circuit is placed as closely as possible to the AV_{DD}_SRDS ball to ensure it filters out as much noise as possible. The ground connection must be near the AV_{DD}_SRDS ball. The 0.003-µF capacitor is closest to the ball, followed by the two 2.2 µF capacitors, and finally the 1 Ω resistor to the board supply plane. The capacitors are connected from AV_{DD}_SRDS to

as shown in Figure 63. If this is not possible, the isolation resistor allows future access to $\overline{\text{TRST}}$ in case a JTAG interface may need to be wired onto the system in future debug situations.

- No pull-up/pull-down is required for TDI, TMS, TDO, or TCK.

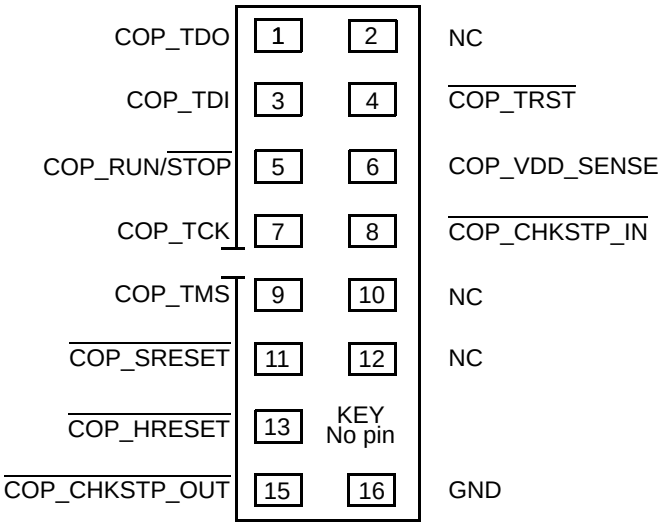

Figure 62. COP Connector Physical Pinout

Table 87. Part Numbering Nomenclature (continued)

MPC	*nnnn*	*t*	*pp*	*ff*	*c*	*r*
Product Code	**Part Identifier**	**Temperature**	**Package[1, 2, 3]**	**Processor Frequency[4]**	**Core Frequency**	**Silicon Version**
MPC	8545E	Blank = 0 to 105°C C = –40° to 105°C	HX = CBGA VU = Pb-free CBGA PX = PBGA VT = Pb-free PBGA	AT = 1200 AQ = 1000 AN = 800	G = 400	Blank = Ver. 2.0 (SVR = 0x80390220) A = Ver. 2.1.1 B = Ver. 2.1.2 D = Ver. 3.1.x (SVR = 0x80390231)
	8545					Blank = Ver. 2.0 (SVR = 0x80310220) A = Ver. 2.1.1 B = Ver. 2.1.2 D = Ver. 3.1.x (SVR = 0x80310231)
	8543E			AQ = 1000 AN = 800		Blank = Ver. 2.0 (SVR = 0x803A0020) A = Ver. 2.1.1 B = Ver. 2.1.2 D = Ver. 3.1.x (SVR = 0x803A0031)
	8543					Blank = Ver. 2.0 (SVR = 0x80320020) A = Ver. 2.1.1 B = Ver. 2.1.2 D = Ver. 3.1.x (SVR = 0x80320031)

Notes:

1. See Section 19, “Package Description,” for more information on available package types.
2. The HiCTE FC-CBGA package is available on only Version 2.0 of the device.
3. The FC-PBGA package is available on only Version 2.1.1, 2.1.2, and 2.1.3 of the device.
4. Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by part number specifications may support other maximum core frequencies.
5. This speed available only for silicon Version 2.1.1, 2.1.2, and 2.1.3.

24 Document Revision History

The following table provides a revision history for this hardware specification.

Table 88. Document Revision History

Rev. Number	Date	Substantive Change(s)
9	02/2012	• Updated Section 21.2, “Thermal for Version 2.1.1, 2.1.2, and 2.1.3 Silicon FC-PBGA with Full Lid and Version 3.1.x Silicon with Stamped Lid,” with version 3.0 silicon information. • Added Figure 56, “Mechanical Dimensions and Bottom Surface Nomenclature of the FC-PBGA with Stamped Lid.” • Updated Table 87, “Part Numbering Nomenclature,” with version 3.0 silicon information. • Removed Note from Section 5.1, “Power-On Ramp Rate”. • Changed the Table 10 title to “Power Supply Ramp Rate”. • Removed table 11. • Updated the title of Section 21.2, “Thermal for Version 2.1.1, 2.1.2, and 2.1.3 Silicon FC-PBGA with Full Lid and Version 3.1.x Silicon with Stamped Lid” to include Thermal Version 2.1.3 and Version 3.1.x Silicon. • Corrected the leaded Solder Ball composition in Table 70, “Package Parameters” • Updated Table 87, “Part Numbering Nomenclature,” with Version 3.1.x silicon information. • Updated the Min and Max value of TDO in the valid times row of Table 44, “JTAG AC Timing Specifications (Independent of SYSCLK)[1]” from 4 and 25 to 2 and 10 respectively .
8	04/2011	• Added Section 14.1, “GPOUT/GPIN Electrical Characteristics.” • Updated Table 71, “MPC8548E Pinout Listing,” Table 72, “MPC8547E Pinout Listing,” Table 73, “MPC8545E Pinout Listing,” and Table 74, “MPC8543E Pinout Listing,” to reflect that the TDO signal is not driven during HRSET* assertion. • Updated Table 87, “Part Numbering Nomenclature” with Ver. 2.1.3 silicon information.
7	09/2010	• In Table 37, “MII Management AC Timing Specifications, modified the fifth row from “MDC to MDIO delay tMDKHDX (16 × tptb_clk × 8) – 3 — (16 × tptb_clk × 8) + 3” to “MDC to MDIO delay tMDKHDX (16 × tCCB × 8) – 3 — (16 × tCCB × 8) + 3.” • Updated Figure 55, “Mechanical Dimensions and Bottom Surface Nomenclature of the HiCTE FC-CBGA and FC-PBGA with Full Lid and figure notes.
6	12/2009	• In Section 5.1, “Power-On Ramp Rate” added explanation that Power-On Ramp Rate is required to avoid falsely triggering ESD circuitry. • In Table 13 changed required ramp rate from 545 V/s for MVREF and VDD/XVDD/SVDD to 3500 V/s for MVREF and 4000 V/s for VDD. • In Table 13 deleted ramp rate requirement for XVDD/SVDD. • In Table 13 footnote 1 changed voltage range of concern from 0–400 mV to 20–500mV. • In Table 13 added footnote 2 explaining that VDD voltage ramp rate is intended to control ramp rate of AVDD pins.
5	10/2009	• In Table 27, “GMII Receive AC Timing Specifications,” changed duty cycle specification from 40/60 to 35/75 for RX_CLK duty cycle. • Updated tMDKHDX in Table 37, “MII Management AC Timing Specifications.” • Added a reference to Revision 2.1.2. • Updated Table 55, “MII Management AC Timing Specifications.” • Added Section 5.1, “Power-On Ramp Rate.”

Table 88. Document Revision History (continued)

Rev. Number	Date	Substantive Change(s)
2	04/2008	• Removed 1:1 support on Table 82, “e500 Core to CCB Clock Ratio.” • Removed MDM from Table 18, “DDR SDRAM Input AC Timing Specifications.” MDM is an Output. • Figure 57, “PLL Power Supply Filter Circuit with PLAT Pins” (AVDD_PLAT). • Figure 58, “PLL Power Supply Filter Circuit with CORE Pins” (AVDD_CORE). • Split Figure 59, “PLL Power Supply Filter Circuit with PCI/LBIU Pins,” (formerly called just “PLL Power Supply Filter Circuit”) into three figures: the original (now specific for AVDD_PCI/AVDD_LBIU) and two new ones.
1	10/2007	• Adjusted maximum SYSCLK frequency down in Table 5, “SYSCLK AC Timing Specifications” per device erratum GEN-13. • Clarified notes to Table 6, “EC_GTX_CLK125 AC Timing Specifications.” • Added Section 4.4, “PCI/PCI-X Reference Clock Timing.” • Clarified descriptions and added PCI/PCI-X to Table 9, “PLL Lock Times.” • Removed support for 266 and 200 Mbps data rates per device erratum GEN-13 in Section 6, “DDR and DDR2 SDRAM.” • Clarified Note 4 of Table 19, “DDR SDRAM Output AC Timing Specifications.” • Clarified the reference clock used in Section 7.2, “DUART AC Electrical Specifications.” • Corrected V_{IH}(min) in Table 22, “GMII, MII, RMII, and TBI DC Electrical Characteristics.” • Corrected V_{IL}(max) in Table 23, “GMII, MII, RMII, TBI, RGMII, RTBI, and FIFO DC Electrical Characteristics.” • Removed DC parameters from Table 24, Table 25, Table 26, Table 27, Table 28, Table 29, Table 32, Table 34, and Table 35. • Corrected V_{IH}(min) in Table 36, “MII Management DC Electrical Characteristics.” • Corrected t_{MDC}(min) in Table 37, “MII Management AC Timing Specifications.” • Updated parameter descriptions for $t_{LBIVKH1}$, $t_{LBIVKH2}$, $t_{LBIXKH1}$, and $t_{LBIXKH2}$ in Table 40, “Local Bus Timing Parameters (BV_{DD} = 3.3 V)—PLL Enabled” and Table 40, “Local Bus Timing Parameters (BV_{DD} = 2.5 V)—PLL Enabled.” • Updated parameter descriptions for $t_{LBIVKH1}$, $t_{LBIVKL2}$, $t_{LBIXKH1}$, and $t_{LBIXKL2}$ in Table 42, “Local Bus Timing Parameters—PLL Bypassed.” Note that $t_{LBIVKL2}$ and $t_{LBIXKL2}$ were previously labeled $t_{LBIVKH2}$ and $t_{LBIXKH2}$. • Added LUPWAIT signal to Figure 23, “Local Bus Signals (PLL Enabled)” and Figure 24, “Local Bus Signals (PLL Bypass Mode).” • Added $\overline{\text{LGTA}}$ signal to Figure 25, Figure 26, Figure 27 and Figure 28. • Corrected LUPWAIT assertion in Figure 26 and Figure 28. • Clarified the PCI reference clock in Section 15.2, “PCI/PCI-X AC Electrical Specifications” • Added Section 17.1, “Package Parameters.” • Added PBGA thermal information in Section 21.2, “Thermal for Version 2.1.1, 2.1.2, and 2.1.3 Silicon FC-PBGA with Full Lid and Version 3.1.x Silicon with Stamped Lid.” • Updated.” • Updated Table 87, “Part Numbering Nomenclature.”
0	07/2007	• Initial Release