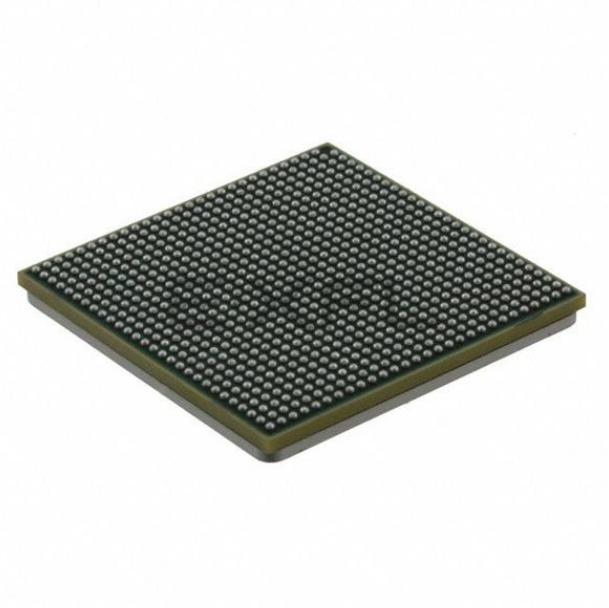
NXP USA Inc. - MPC8543EVTAQGD Datasheet



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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.0GHz
Co-Processors/DSP	Signal Processing; SPE, Security; SEC
RAM Controllers	DDR, DDR2, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8543evtaqgd

    - Dedicated single data rate SDRAM controller
  - Parity support
  - Default boot ROM chip select with configurable bus width (8, 16, or 32 bits)
- Four enhanced three-speed Ethernet controllers (eTSECs)
  - Three-speed support (10/100/1000 Mbps)
  - Four controllers designed to comply with IEEE Std. 802.3®, 802.3u, 802.3x, 802.3z, 802.3ac, and 802.3ab
  - Support for various Ethernet physical interfaces:
    - 1000 Mbps full-duplex IEEE 802.3 GMII, IEEE 802.3z TBI, RTBI, and RGMII
    - 10/100 Mbps full and half-duplex IEEE 802.3 MII, IEEE 802.3 RGMII, and RMII
  - Flexible configuration for multiple PHY interface configurations. See Section 8.1, "Enhanced Three-Speed Ethernet Controller (eTSEC) (10/100/1Gb Mbps)—GMII/MII/TBI/RGMII/RTBI/RMII Electrical Characteristics," for more information.
  - TCP/IP acceleration and QoS features available
    - IP v4 and IP v6 header recognition on receive
    - IP v4 header checksum verification and generation
    - TCP and UDP checksum verification and generation
    - Per-packet configurable acceleration
    - Recognition of VLAN, stacked (queue in queue) VLAN, IEEE Std 802.2™, PPPoE session, MPLS stacks, and ESP/AH IP-security headers
    - Supported in all FIFO modes
  - Quality of service support:
    - Transmission from up to eight physical queues
    - Reception to up to eight physical queues
  - Full- and half-duplex Ethernet support (1000 Mbps supports only full duplex):
    - IEEE 802.3 full-duplex flow control (automatic PAUSE frame generation or software-programmed PAUSE frame generation and recognition)
  - Programmable maximum frame length supports jumbo frames (up to 9.6 Kbytes) and IEEE Std. 802.1™ virtual local area network (VLAN) tags and priority
  - VLAN insertion and deletion
    - Per-frame VLAN control word or default VLAN for each eTSEC
    - Extracted VLAN control word passed to software separately
  - Retransmission following a collision
  - CRC generation and verification of inbound/outbound frames
  - Programmable Ethernet preamble insertion and extraction of up to 7 bytes
  - MAC address recognition:
    - Exact match on primary and virtual 48-bit unicast addresses

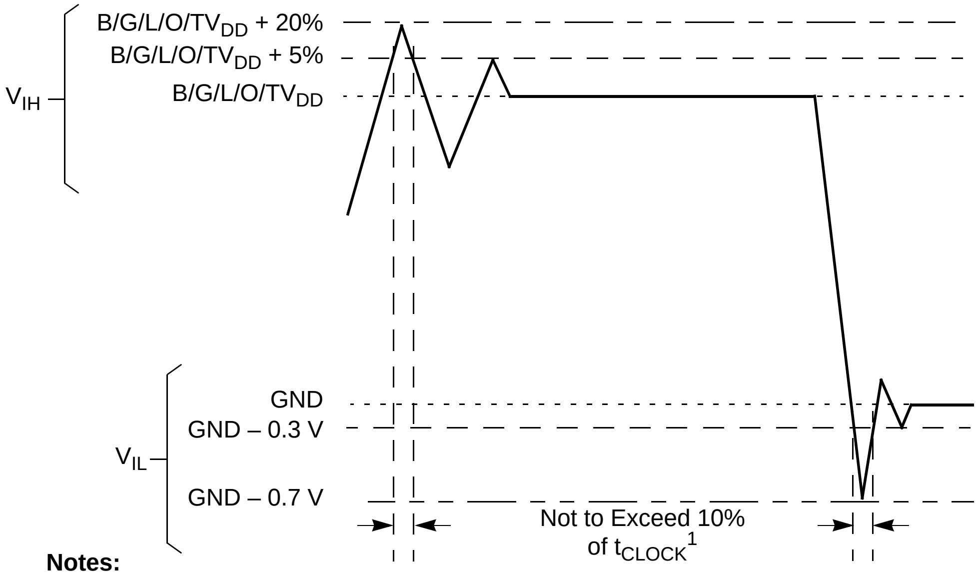
Table 2. Recommended Operating Conditions (continued)

Characteristic	Symbol	Recommended Value	Unit	Notes
Junction temperature range	Tj	0 to 105	°C	—

Notes:

1. This voltage is the input to the filter discussed in Section 22.2, “PLL Power Supply Filtering,” and not necessarily the voltage at the AV_{DD} pin, which may be reduced from V_{DD} by the filter.
2. **Caution:** MV_{IN} must not exceed GV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
3. **Caution:** OV_{IN} must not exceed OV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
4. **Caution:** L/TV_{IN} must not exceed L/TV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.

The following figure shows the undershoot and overshoot voltages at the interfaces of this device.



Notes:

1. t_{CLOCK} refers to the clock period associated with the respective interface:
   For I^2C and JTAG, t_{CLOCK} references SYSCLK.
   For DDR, t_{CLOCK} references MCLK.
   For eTSEC, t_{CLOCK} references EC_GTX_CLK125.
   For LBIU, t_{CLOCK} references LCLK.
   For PCI, t_{CLOCK} references PCI*n*_CLK or SYSCLK.
   For SerDes, t_{CLOCK} references SD_REF_CLK.
2. Note that with the PCI overshoot allowed (as specified above), the device does not fully comply with the maximum AC ratings and device protection guideline outlined in the PCI rev. 2.2 standard (section 4.2.2.3).

Figure 2. Overshoot/Undershoot Voltage for $GV_{DD}/OV_{DD}/LV_{DD}/BV_{DD}/TV_{DD}$

The core voltage must always be provided at nominal 1.1 V. Voltage to the processor interface I/Os are provided through separate sets of supply pins and must be provided at the voltages shown in Table 2. The input voltage threshold scales with respect to the associated I/O supply voltage. OV_{DD} and LV_{DD} based receivers are simple CMOS I/O circuits and satisfy appropriate LVCMOS type specifications. The DDR SDRAM interface uses a single-ended differential receiver referenced the externally supplied MV_{REF} signal (nominally set to $GV_{DD}/2$) as is appropriate for the SSTL2 electrical signaling standard.

5 RESET Initialization

This section describes the AC electrical specifications for the RESET initialization timing requirements of the device. The following table provides the RESET initialization AC timing specifications for the DDR SDRAM component(s).

Table 8. RESET Initialization Timing Specifications

Parameter/Condition	Min	Max	Unit	Notes
Required assertion time of $\overline{\text{HRESET}}$	100	—	µs	—
Minimum assertion time for $\overline{\text{SRESET}}$	3	—	SYSCLKs	1
PLL input setup time with stable SYSCLK before $\overline{\text{HRESET}}$ negation	100	—	µs	—
Input setup time for POR configs (other than PLL config) with respect to negation of $\overline{\text{HRESET}}$	4	—	SYSCLKs	1
Input hold time for all POR configs (including PLL config) with respect to negation of $\overline{\text{HRESET}}$	2	—	SYSCLKs	1
Maximum valid-to-high impedance time for actively driven POR configs with respect to negation of $\overline{\text{HRESET}}$	—	5	SYSCLKs	1

Note:

1. SYSCLK is the primary clock input for the device.

The following table provides the PLL lock times.

Table 9. PLL Lock Times

Parameter/Condition	Min	Max	Unit
Core and platform PLL lock times	—	100	µs
Local bus PLL lock time	—	50	µs
PCI/PCI-X bus PLL lock time	—	50	µs

5.1 Power-On Ramp Rate

This section describes the AC electrical specifications for the power-on ramp rate requirements. Controlling the maximum power-on ramp rate is required to avoid falsely triggering the ESD circuitry. The following table provides the power supply ramp rate specifications.

Table 10. Power Supply Ramp Rate

Parameter	Min	Max	Unit	Notes
Required ramp rate for MVREF	—	3500	V/s	1
Required ramp rate for VDD	—	4000	V/s	1, 2

Note:

1. Maximum ramp rate from 200 to 500 mV is most critical as this range may falsely trigger the ESD circuitry.
2. VDD itself is not vulnerable to false ESD triggering; however, as per Section 22.2, “PLL Power Supply Filtering,” the recommended AVDD_CORE, AVDD_PLAT, AVDD_LBIU, AVDD_PCI1 and AVDD_PCI2 filters are all connected to VDD. Their ramp rates must be equal to or less than the VDD ramp rate.

Table 19. DDR SDRAM Output AC Timing Specifications (continued)

At recommended operating conditions.

Parameter	Symbol[1]	Min	Max	Unit	Notes
MDQS epilogue end	t_{DDKHME}	–0.6	0.6	ns	6

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{(\text{first two letters of functional block})(\text{signal})(\text{state})(\text{reference})(\text{state})}$ for inputs and $t_{(\text{first two letters of functional block})(\text{reference})(\text{state})(\text{signal})(\text{state})}$ for outputs. Output hold time can be read as DDR timing (DD) from the rising or falling edge of the reference clock (KH or KL) until the output went invalid (AX or DX). For example, t_{DDKHAS} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes from the high (H) state until outputs (A) are setup (S) or output valid time. Also, t_{DDKLDX} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes low (L) until data outputs (D) are invalid (X) or data output hold time.
2. All MCK/$\overline{\text{MCK}}$ referenced measurements are made from the crossing of the two signals ±0.1 V.
3. ADDR/CMD includes all DDR SDRAM output signals except MCK/$\overline{\text{MCK}}$, $\overline{\text{MCS}}$, and MDQ/MECC/MDM/MDQS.
4. Note that t_{DDKHMH} follows the symbol conventions described in note 1. For example, t_{DDKHMH} describes the DDR timing (DD) from the rising edge of the MCK[*n*] clock (KH) until the MDQS signal is valid (MH). t_{DDKHMH} can be modified through control of the MDQS override bits (called WR_DATA_DELAY) in the TIMING_CFG_2 register. This is typically set to the same delay as in DDR_SDRAM_CLK_CNTL[CLK_ADJUST]. The timing parameters listed in the table assume that these 2 parameters have been set to the same adjustment value. See the *MPC8548E PowerQUICC III Integrated Processor Reference Manual* for a description and understanding of the timing modifications enabled by use of these bits.
5. Determined by maximum possible skew between a data strobe (MDQS) and any corresponding bit of data (MDQ), ECC (MECC), or data mask (MDM). The data strobe must be centered inside of the data eye at the pins of the microprocessor.
6. All outputs are referenced to the rising edge of MCK[*n*] at the pins of the microprocessor. Note that t_{DDKHMP} follows the symbol conventions described in note 1.

NOTE

For the ADDR/CMD setup and hold specifications in Table 19, it is assumed that the clock control register is set to adjust the memory clocks by 1/2 applied cycle.

Figure 3 shows the DDR SDRAM output timing for the MCK to MDQS skew measurement (t_{DDKHMH}).

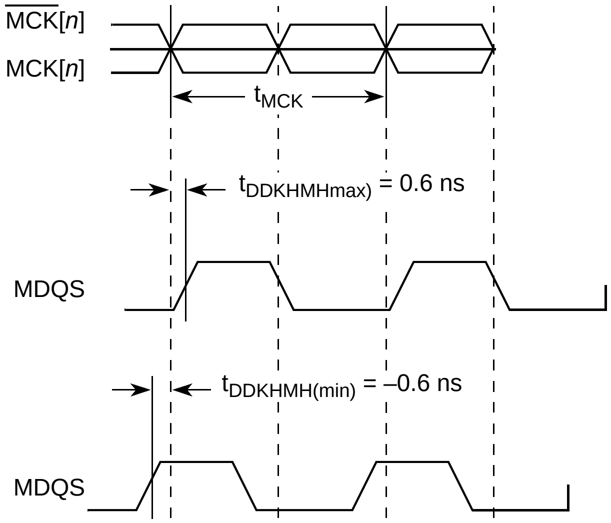


Figure 3. Timing Diagram for tDDKHMH

Table 23. GMII, MII, RMII, TBI, RGMII, RTBI, and FIFO DC Electrical Characteristics

Parameters	Symbol	Min	Max	Unit	Notes
Supply voltage 2.5 V	LV_{DD}/TV_{DD}	2.37	2.63	V	1, 2
Output high voltage (LV_{DD}/TV_{DD} = Min, I_{OH} = –1.0 mA)	V_{OH}	2.00	LV_{DD}/TV_{DD} + 0.3	V	—
Output low voltage (LV_{DD}/TV_{DD} = Min, I_{OL} = 1.0 mA)	V_{OL}	GND –0.3	0.40	V	—
Input high voltage	V_{IH}	1.70	LV_{DD}/TV_{DD} + 0.3	V	—
Input low voltage	V_{IL}	–0.3	0.90	V	—
Input high current ($V_{IN} = LV_{DD}$, $V_{IN} = TV_{DD}$)	I_{IH}	—	10	μA	1, 2, 3
Input low current (V_{IN} = GND)	I_{IL}	–15	—	μA	3

Notes:
1. LV_{DD} supports eTSECs 1 and 2.
2. TV_{DD} supports eTSECs 3 and 4.
3. Note that the symbol V_{IN}, in this case, represents the LV_{IN} and TV_{IN} symbols referenced in Table 1 and Table 2.

8.2 FIFO, GMII, MII, TBI, RGMII, RMII, and RTBI AC Timing Specifications

The AC timing specifications for FIFO, GMII, MII, TBI, RGMII, RMII, and RTBI are presented in this section.

8.2.1 FIFO AC Specifications

The basis for the AC specifications for the eTSEC's FIFO modes is the double data rate RGMII and RTBI specifications, since they have similar performances and are described in a source-synchronous fashion like FIFO modes. However, the FIFO interface provides deliberate skew between the transmitted data and source clock in GMII fashion.

When the eTSEC is configured for FIFO modes, all clocks are supplied from external sources to the relevant eTSEC interface. That is, the transmit clock must be applied to the eTSEC*n*'s TSEC*n*_TX_CLK, while the receive clock must be applied to pin TSEC*n*_RX_CLK. The eTSEC internally uses the transmit clock to synchronously generate transmit data and outputs an echoed copy of the transmit clock back out onto the TSEC*n*_GTX_CLK pin (while transmit data appears on TSEC*n*_TXD[7:0], for example). It is intended that external receivers capture eTSEC transmit data using the clock on TSEC*n*_GTX_CLK as a source- synchronous timing reference. Typically, the clock edge that launched the data can be used, since the clock is delayed by the eTSEC to allow acceptable set-up margin at the receiver. Note that there is relationship between the maximum FIFO speed and the platform speed. For more information see Section 4.5, "Platform to FIFO Restrictions."

A summary of the FIFO AC specifications appears in Table 24 and Table 25.

Table 24. FIFO Mode Transmit AC Timing Specification

Parameter/Condition	Symbol	Min	Typ	Max	Unit
TX_CLK, GTX_CLK clock period	t_{FIT}	5.3	8.0	100	ns
TX_CLK, GTX_CLK duty cycle	t_{FITH}/t_{FIT}	45	50	55	%
TX_CLK, GTX_CLK peak-to-peak jitter	t_{FITJ}	—	—	250	ps
Rise time TX_CLK (20%–80%)	t_{FITR}	—	—	0.75	ns
Fall time TX_CLK (80%–20%)	t_{FITF}	—	—	0.75	ns
FIFO data TXD[7:0], TX_ER, TX_EN setup time to GTX_CLK	t_{FITDV}	2.0	—	—	ns
GTX_CLK to FIFO data TXD[7:0], TX_ER, TX_EN hold time	t_{FITDX}	0.5	—	3.0	ns

Table 25. FIFO Mode Receive AC Timing Specification

Parameter/Condition	Symbol	Min	Typ	Max	Unit
RX_CLK clock period	t_{FIR}	5.3	8.0	100	ns
RX_CLK duty cycle	t_{FIRH}/t_{FIR}	45	50	55	%
RX_CLK peak-to-peak jitter	t_{FIRJ}	—	—	250	ps
Rise time RX_CLK (20%–80%)	t_{FIRR}	—	—	0.75	ns
Fall time RX_CLK (80%–20%)	t_{FIRF}	—	—	0.75	ns
RXD[7:0], RX_DV, RX_ER setup time to RX_CLK	t_{FIRDV}	1.5	—	—	ns
RXD[7:0], RX_DV, RX_ER hold time to RX_CLK	t_{FIRDX}	0.5	—	—	ns

Note:

1. The minimum cycle period of the TX_CLK and RX_CLK is dependent on the maximum platform frequency of the speed bins the part belongs to as well as the FIFO mode under operation. See Section 4.5, “Platform to FIFO Restrictions.”

Timing diagrams for FIFO appear in Figure 6 and Figure 7.

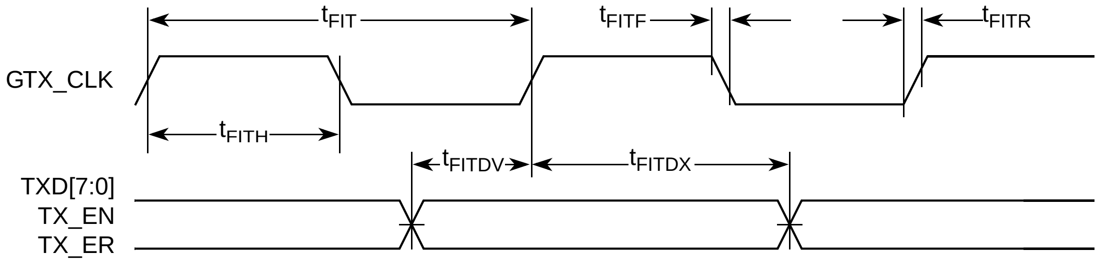


Figure 6. FIFO Transmit AC Timing Diagram

Figure 8 shows the GMII transmit AC timing diagram.

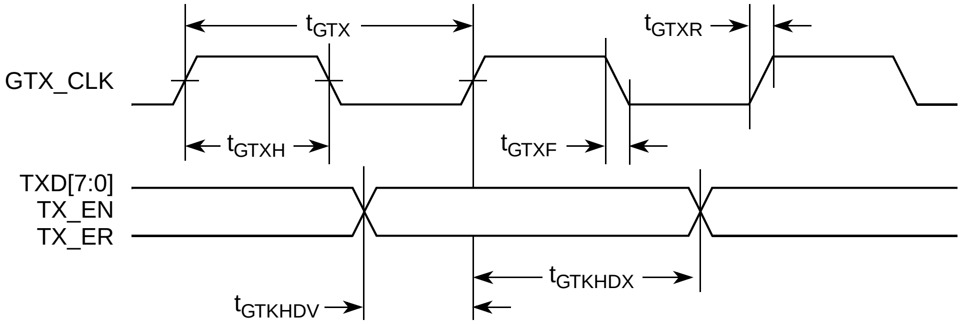


Figure 8. GMII Transmit AC Timing Diagram

8.2.2.2 GMII Receive AC Timing Specifications

This table provides the GMII receive AC timing specifications.

Table 27. GMII Receive AC Timing Specifications

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
RX_CLK clock period	t_{GRX}	—	8.0	—	ns
RX_CLK duty cycle	t_{GRXH}/t_{GRX}	35	—	75	ns
RXD[7:0], RX_DV, RX_ER setup time to RX_CLK	t_{GRDVKH}	2.0	—	—	ns
RXD[7:0], RX_DV, RX_ER hold time to RX_CLK	t_{GRDXKH}	0	—	—	ns
RX_CLK clock rise (20%-80%)	t_{GRXR}[2]	—	—	1.0	ns
RX_CLK clock fall time (80%-20%)	t_{GRXF}[2]	—	—	1.0	ns

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GRDVKH} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{RX} clock reference (K) going to the high state (H) or setup time. Also, t_{GRDXKL} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{GRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GRX} represents the GMII (G) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. Guaranteed by design.

Figure 9 provides the AC test load for eTSEC.

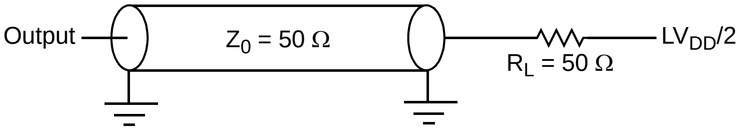


Figure 9. eTSEC AC Test Load

Table 42. Local Bus Timing Parameters—PLL Bypassed (continued)

Parameter	Symbol[1]	Min	Max	Unit	Notes
$\overline{\text{LGTA}}$/LUPWAIT input hold from local bus clock	$t_{LBIXKL2}$	–1.3	—	ns	4, 5
LALE output transition to LAD/LDP output transition (LATCH hold time)	t_{LBOTOT}	1.5	—	ns	6
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKLOV1}$	—	–0.3	ns	—
Local bus clock to data valid for LAD/LDP	$t_{LBKLOV2}$	—	–0.1	ns	4
Local bus clock to address valid for LAD	$t_{LBKLOV3}$	—	0	ns	4
Local bus clock to LALE assertion	$t_{LBKLOV4}$	—	0	ns	4
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKLOX1}$	–3.7	—	ns	4
Output hold from local bus clock for LAD/LDP	$t_{LBKLOX2}$	–3.7	—	ns	4
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$t_{LBKLOZ1}$	—	0.2	ns	7
Local bus clock to output high impedance for LAD/LDP	$t_{LBKLOZ2}$	—	0.2	ns	7

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to local bus clock for PLL bypass mode. Timings may be negative with respect to the local bus clock because the actual launch and capture of signals is done with the internal launch/capture clock, which precedes LCLK by t_{LBKHKT}.
3. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[*n*]. Skew measured between complementary signals at $BV_{DD}/2$.
4. All signals are measured from $BV_{DD}/2$ of the rising edge of local bus clock for PLL bypass mode to $0.4 \times BV_{DD}$ of the signal in question for 3.3-V signaling levels.
5. Input timings are measured at the pin.
6. The value of t_{LBOTOT} is the measurement of the minimum time between the negation of LALE and any change in LAD.
7. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
8. Guaranteed by characterization.
9. Guaranteed by design.

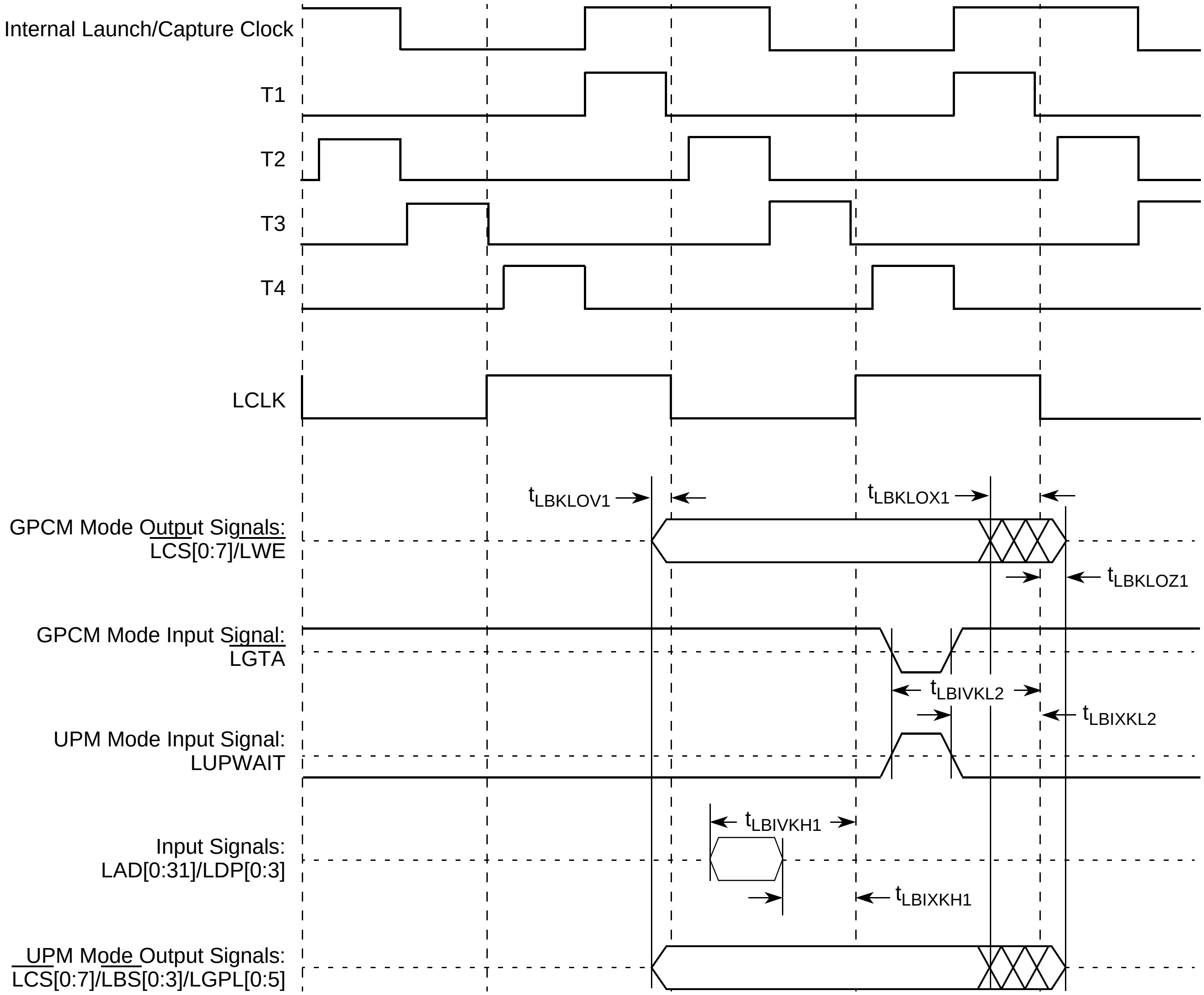


Figure 28. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 8 or 16 (PLL Bypass Mode)

to AC-coupling. Its value could be ranged from 140 to 240 Ω depending on the clock driver vendor's requirement. R2 is used together with the SerDes reference clock receiver's 50-Ω termination resistor to attenuate the LVPECL output's differential peak level such that it meets the SerDes reference clock's differential input amplitude requirement (between 200 and 800 mV differential peak). For example, if the LVPECL output's differential peak is 900 mV and the desired SerDes reference clock input amplitude is selected as 600 mV, the attenuation factor is 0.67, which requires R2 = 25 Ω. Consult a clock driver chip manufacturer to verify whether this connection scheme is compatible with a particular clock driver chip.

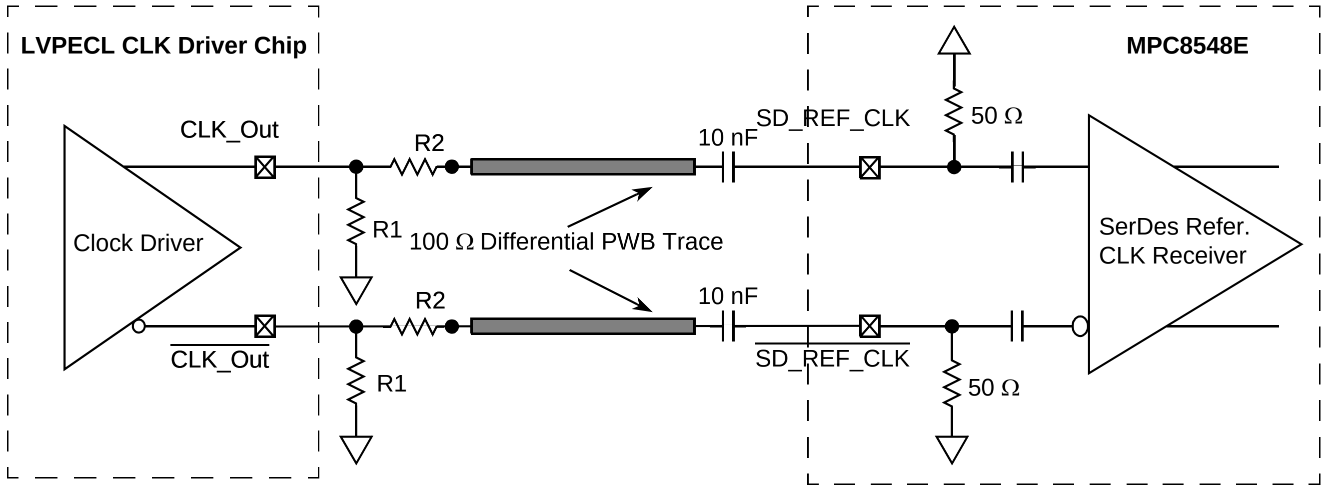


Figure 45. AC-Coupled Differential Connection with LVPECL Clock Driver (Reference Only)

Figure 46 shows the SerDes reference clock connection reference circuits for a single-ended clock driver. It assumes the DC levels of the clock driver are compatible with the SerDes reference clock input's DC requirement.

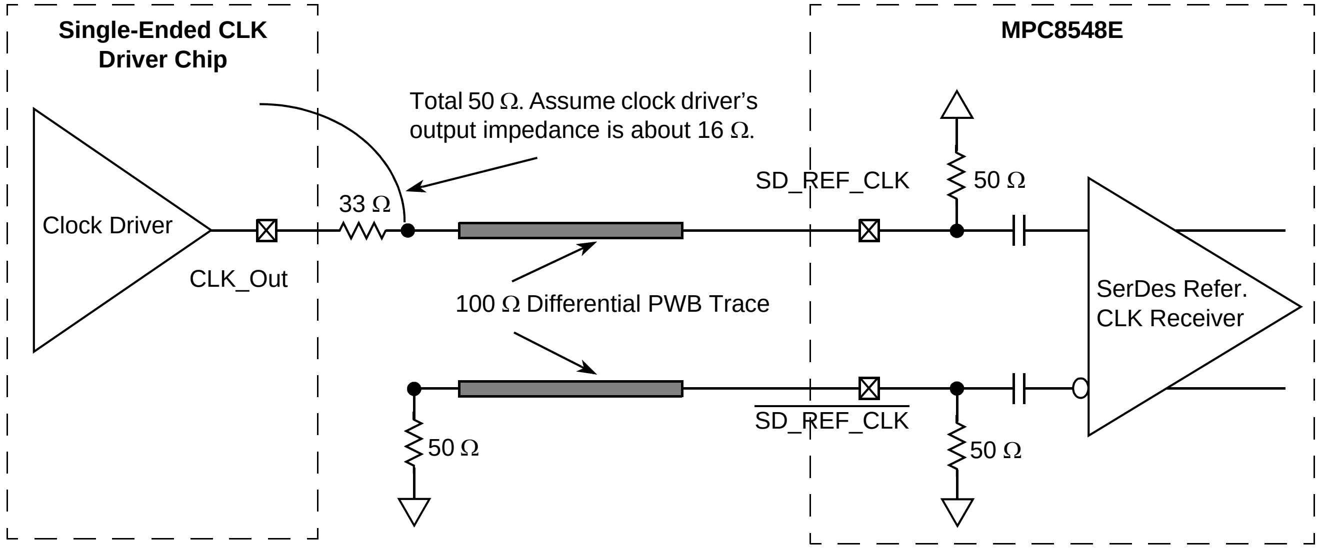


Figure 46. Single-Ended Connection (Reference Only)

Table 56. Differential Transmitter (TX) Output Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Unit	Comments
$V_{TX\text{-}DC\text{-}CM}$	The TX DC common mode voltage	0	—	3.6	V	The allowed DC common mode voltage under any conditions. See Note 6.
$I_{TX\text{-}SHORT}$	TX short circuit current limit	—	—	90	mA	The total current the transmitter can provide when shorted to its ground
$T_{TX\text{-}IDLE\text{-}MIN}$	Minimum time spent in electrical idle	50	—		UI	Minimum time a transmitter must be in electrical idle utilized by the receiver to start looking for an electrical idle exit after successfully receiving an electrical idle ordered set
$T_{TX\text{-}IDLE\text{-}SET\text{-}TO\text{-}IDLE}$	Maximum time to transition to a valid electrical idle after sending an electrical idle ordered set	—	—	20	UI	After sending an electrical idle ordered set, the transmitter must meet all electrical idle specifications within this time. This is considered a debounce time for the transmitter to meet electrical idle after transitioning from L0.
$T_{TX\text{-}IDLE\text{-}TO\text{-}DIFF\text{-}DATA}$	Maximum time to transition to valid TX specifications after leaving an electrical idle condition	—	—	20	UI	Maximum time to meet all TX specifications when transitioning from electrical idle to sending differential data. This is considered a debounce time for the TX to meet all TX specifications after leaving electrical idle
$RL_{TX\text{-}DIFF}$	Differential return loss	12	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4.
$RL_{TX\text{-}CM}$	Common mode return loss	6	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4.
$Z_{TX\text{-}DIFF\text{-}DC}$	DC differential TX impedance	80	100	120	Ω	TX DC differential mode low impedance
$Z_{TX\text{-}DC}$	Transmitter DC impedance	40	—	—	Ω	Required TX D+ as well as D– DC impedance during all states
$L_{TX\text{-}SKEW}$	Lane-to-lane output skew	—	—	500 + 2 UI	ps	Static skew between any two transmitter lanes within a single Link
C_{TX}	AC coupling capacitor	75	—	200	nF	All transmitters shall be AC coupled. The AC coupling is required either within the media or within the transmitting component itself. See note 8.

Table 57. Differential Receiver (RX) Input Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Unit	Comments
$L_{TX\text{-}SKEW}$	Total Skew	—	—	20	ns	Skew across all lanes on a Link. This includes variation in the length of SKP ordered set (for example, COM and one to five symbols) at the RX as well as any delay differences arising from the interconnect itself.

Notes:

1. No test load is necessarily associated with this value.
2. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in Figure 50 must be used as the RX device when taking measurements (also see the receiver compliance eye diagram shown in Figure 49). If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
3. A $T_{RX\text{-}EYE}$ = 0.40 UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX\text{-}EYE\text{-}MEDIAN\text{-}to\text{-}MAX\text{-}JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive TX UIs. Note that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
4. The receiver input impedance shall result in a differential return loss greater than or equal to 15 dB with the D+ line biased to 300 mV and the D– line biased to –{300 mV and a common mode return loss greater than or equal to 6 dB (no bias required) over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements for is 50 Ω to ground for both the D+ and D– line (that is, as measured by a vector network analyzer with 50-Ω probes—see Figure 50). Note: that the series capacitors CTX is optional for the return loss measurement.
5. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM) there is a 5 ms transition time before receiver termination values must be met on all unconfigured lanes of a port.
6. The RX DC common mode Impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the RX ground.
7. It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

17.5 Receiver Compliance Eye Diagrams

The RX eye diagram in Figure 49 is specified using the passive compliance/test measurement load (see Figure 50) in place of any real PCI Express RX component.

Note: In general, the minimum receiver eye diagram measured with the compliance/test measurement load (see Figure 50) is larger than the minimum receiver eye diagram measured over a range of systems at the input receiver of any real PCI Express component. The degraded eye diagram at the input receiver is due to traces internal to the package as well as silicon parasitic characteristics which cause the real PCI Express component to vary in impedance from the compliance/test measurement load. The input receiver eye diagram is implementation specific and is not specified. RX component designer must provide additional margin to adequately compensate for the degraded minimum receiver eye diagram (shown in Figure 49) expected at the input receiver based on some adequate combination of system simulations and the return loss measured looking into the RX package and silicon. The RX eye diagram must be aligned in time using the jitter median to locate the center of the eye diagram.

Notes:

1. All dimensions are in millimeters.
2. Dimensioning and tolerancing per ASME Y14.5M-1994.
3. Maximum solder ball diameter measured parallel to datum A.
4. Datum A, the seating plane, is determined by the spherical crowns of the solder balls.
5. Parallelism measurement shall exclude any effect of mark on top surface of package.
6. All dimensions are symmetric across the package center lines unless dimensioned otherwise.

Table 72. MPC8547E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
Reserved	AE26	—	—	2
cfg_pci1_clk	AG24	I	OV$_{DD}$	5
Reserved	AF25	—	—	101
Reserved	AE25	—	—	2
Reserved	AG25	—	—	2
Reserved	AD24	—	—	2
Reserved	AF24	—	—	2
Reserved	AD27	—	—	2
Reserved	AD28, AE27, W17, AF26	—	—	2
Reserved	AH25	—	—	2
DDR SDRAM Memory Interface				
MDQ[0:63]	L18, J18, K14, L13, L19, M18, L15, L14, A17, B17, A13, B12, C18, B18, B13, A12, H18, F18, J14, F15, K19, J19, H16, K15, D17, G16, K13, D14, D18, F17, F14, E14, A7, A6, D5, A4, C8, D7, B5, B4, A2, B1, D1, E4, A3, B2, D2, E3, F3, G4, J5, K5, F6, G5, J6, K4, J1, K2, M5, M3, J3, J2, L1, M6	I/O	GV$_{DD}$	—
MECC[0:7]	H13, F13, F11, C11, J13, G13, D12, M12	I/O	GV$_{DD}$	—
MDM[0:8]	M17, C16, K17, E16, B6, C4, H4, K1, E13	O	GV$_{DD}$	—
MDQS[0:8]	M15, A16, G17, G14, A5, D3, H1, L2, C13	I/O	GV$_{DD}$	—
$\overline{\text{MDQS}}$[0:8]	L17, B16, J16, H14, C6, C2, H3, L4, D13	I/O	GV$_{DD}$	—
MA[0:15]	A8, F9, D9, B9, A9, L10, M10, H10, K10, G10, B8, E10, B10, G6, A10, L11	O	GV$_{DD}$	—
MBA[0:2]	F7, J7, M11	O	GV$_{DD}$	—
$\overline{\text{MWE}}$	E7	O	GV$_{DD}$	—
$\overline{\text{MCAS}}$	H7	O	GV$_{DD}$	—
$\overline{\text{MRAS}}$	L8	O	GV$_{DD}$	—
MCKE[0:3]	F10, C10, J11, H11	O	GV$_{DD}$	11
$\overline{\text{MCS}}$[0:3]	K8, J8, G8, F8	O	GV$_{DD}$	—
MCK[0:5]	H9, B15, G2, M9, A14, F1	O	GV$_{DD}$	—
$\overline{\text{MCK}}$[0:5]	J9, A15, G1, L9, B14, F2	O	GV$_{DD}$	—
MODT[0:3]	E6, K6, L7, M7	O	GV$_{DD}$	—
MDIC[0:1]	A19, B19	I/O	GV$_{DD}$	36

Table 72. MPC8547E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
Local Bus Controller Interface				
LAD[0:31]	E27, B20, H19, F25, A20, C19, E28, J23, A25, K22, B28, D27, D19, J22, K20, D28, D25, B25, E22, F22, F21, C25, C22, B23, F20, A23, A22, E19, A21, D21, F19, B21	I/O	BV_{DD}	—
LDP[0:3]	K21, C28, B26, B22	I/O	BV_{DD}	—
LA[27]	H21	O	BV_{DD}	5, 9
LA[28:31]	H20, A27, D26, A28	O	BV_{DD}	5, 7, 9
$\overline{\text{LCS}}$[0:4]	J25, C20, J24, G26, A26	O	BV_{DD}	—
$\overline{\text{LCS5}}$/$\overline{\text{DMA\_DREQ2}}$	D23	I/O	BV_{DD}	1
$\overline{\text{LCS6}}$/$\overline{\text{DMA\_DACK2}}$	G20	O	BV_{DD}	1
$\overline{\text{LCS7}}$/$\overline{\text{DMA\_DDONE2}}$	E21	O	BV_{DD}	1
$\overline{\text{LWE0}}$/$\overline{\text{LBS0}}$/LSDDQM[0]	G25	O	BV_{DD}	5, 9
$\overline{\text{LWE1}}$/$\overline{\text{LBS1}}$/LSDDQM[1]	C23	O	BV_{DD}	5, 9
$\overline{\text{LWE2}}$/$\overline{\text{LBS2}}$/LSDDQM[2]	J21	O	BV_{DD}	5, 9
$\overline{\text{LWE3}}$/$\overline{\text{LBS3}}$/LSDDQM[3]	A24	O	BV_{DD}	5, 9
LALE	H24	O	BV_{DD}	5, 8, 9
LBCTL	G27	O	BV_{DD}	5, 8, 9
LGPL0/LSDA10	F23	O	BV_{DD}	5, 9
LGPL1/$\overline{\text{LSDWE}}$	G22	O	BV_{DD}	5, 9
LGPL2/$\overline{\text{LOE}}$/$\overline{\text{LSDRAS}}$	B27	O	BV_{DD}	5, 8, 9
LGPL3/$\overline{\text{LSDCAS}}$	F24	O	BV_{DD}	5, 9
LGPL4/$\overline{\text{LGTA}}$/LUPWAIT/LPBSE	H23	I/O	BV_{DD}	—
LGPL5	E26	O	BV_{DD}	5, 9
LCKE	E24	O	BV_{DD}	—
LCLK[0:2]	E23, D24, H22	O	BV_{DD}	—
LSYNC_IN	F27	I	BV_{DD}	—
LSYNC_OUT	F28	O	BV_{DD}	—
DMA				
$\overline{\text{DMA\_DACK}}$[0:1]	AD3, AE1	O	OV_{DD}	5, 9, 107
$\overline{\text{DMA\_DREQ}}$[0:1]	AD4, AE2	I	OV_{DD}	—
$\overline{\text{DMA\_DDONE}}$[0:1]	AD2, AD1	O	OV_{DD}	—
Programmable Interrupt Controller				
$\overline{\text{UDE}}$	AH16	I	OV_{DD}	—
$\overline{\text{MCP}}$	AG19	I	OV_{DD}	—

Table 72. MPC8547E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
SD_PLL_TPA	U26	O	—	24

Note: All note references in this table use the same numbers as those for Table 71. See Table 71 for the meanings of these notes.

Table 73 provides the pin-out listing for the MPC8545E 783 FC-PBGA package.

NOTE

All note references in the following table use the same numbers as those for Table 71. See Table 71 for the meanings of these notes.

Table 73. MPC8545E Pinout Listing

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI1 and PCI2 (One 64-Bit or Two 32-Bit)				
PCI1_AD[63:32]/PCI2_AD[31:0]	AB14, AC15, AA15, Y16, W16, AB16, AC16, AA16, AE17, AA18, W18, AC17, AD16, AE16, Y17, AC18, AB18, AA19, AB19, AB21, AA20, AC20, AB20, AB22, AC22, AD21, AB23, AF23, AD23, AE23, AC23, AC24	I/O	OV$_{DD}$	17
PCI1_AD[31:0]	AH6, AE7, AF7, AG7, AH7, AF8, AH8, AE9, AH9, AC10, AB10, AD10, AG10, AA10, AH10, AA11, AB12, AE12, AG12, AH12, AB13, AA12, AC13, AE13, Y14, W13, AG13, V14, AH13, AC14, Y15, AB15	I/O	OV$_{DD}$	17
PCI1_C_$\overline{\text{BE}}$[7:4]/PCI2_C_$\overline{\text{BE}}$[3:0]	AF15, AD14, AE15, AD15	I/O	OV$_{DD}$	17
PCI1_C_$\overline{\text{BE}}$[3:0]	AF9, AD11, Y12, Y13	I/O	OV$_{DD}$	17
PCI1_PAR64/PCI2_PAR	W15	I/O	OV$_{DD}$	—
$\overline{\text{PCI1\_GNT}}$[4:1]	AG6, AE6, AF5, AH5	O	OV$_{DD}$	5, 9, 35
$\overline{\text{PCI1\_GNT0}}$	AG5	I/O	OV$_{DD}$	—
$\overline{\text{PCI1\_IRDY}}$	AF11	I/O	OV$_{DD}$	2
PCI1_PAR	AD12	I/O	OV$_{DD}$	—
$\overline{\text{PCI1\_PERR}}$	AC12	I/O	OV$_{DD}$	2
$\overline{\text{PCI1\_SERR}}$	V13	I/O	OV$_{DD}$	2, 4
$\overline{\text{PCI1\_STOP}}$	W12	I/O	OV$_{DD}$	2
$\overline{\text{PCI1\_TRDY}}$	AG11	I/O	OV$_{DD}$	2
$\overline{\text{PCI1\_REQ}}$[4:1]	AH2, AG4, AG3, AH4	I	OV$_{DD}$	—
$\overline{\text{PCI1\_REQ0}}$	AH3	I/O	OV$_{DD}$	—
PCI1_CLK	AH26	I	OV$_{DD}$	39
$\overline{\text{PCI1\_DEVSEL}}$	AH11	I/O	OV$_{DD}$	2

Table 74. MPC8543E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$\overline{\text{MWE}}$	E7	O	GV$_{DD}$	—
$\overline{\text{MCAS}}$	H7	O	GV$_{DD}$	—
$\overline{\text{MRAS}}$	L8	O	GV$_{DD}$	—
MCKE[0:3]	F10, C10, J11, H11	O	GV$_{DD}$	11
$\overline{\text{MCS}}$[0:3]	K8, J8, G8, F8	O	GV$_{DD}$	—
MCK[0:5]	H9, B15, G2, M9, A14, F1	O	GV$_{DD}$	—
$\overline{\text{MCK}}$[0:5]	J9, A15, G1, L9, B14, F2	O	GV$_{DD}$	—
MODT[0:3]	E6, K6, L7, M7	O	GV$_{DD}$	—
MDIC[0:1]	A19, B19	I/O	GV$_{DD}$	36
Local Bus Controller Interface				
LAD[0:31]	E27, B20, H19, F25, A20, C19, E28, J23, A25, K22, B28, D27, D19, J22, K20, D28, D25, B25, E22, F22, F21, C25, C22, B23, F20, A23, A22, E19, A21, D21, F19, B21	I/O	BV$_{DD}$	—
LDP[0:3]	K21, C28, B26, B22	I/O	BV$_{DD}$	—
LA[27]	H21	O	BV$_{DD}$	5, 9
LA[28:31]	H20, A27, D26, A28	O	BV$_{DD}$	5, 7, 9
$\overline{\text{LCS}}$[0:4]	J25, C20, J24, G26, A26	O	BV$_{DD}$	—
$\overline{\text{LCS5}}$/$\overline{\text{DMA\_DREQ2}}$	D23	I/O	BV$_{DD}$	1
$\overline{\text{LCS6}}$/$\overline{\text{DMA\_DACK2}}$	G20	O	BV$_{DD}$	1
$\overline{\text{LCS7}}$/$\overline{\text{DMA\_DDONE2}}$	E21	O	BV$_{DD}$	1
$\overline{\text{LWE0}}$/$\overline{\text{LBS0}}$/LSDDQM[0]	G25	O	BV$_{DD}$	5, 9
$\overline{\text{LWE1}}$/$\overline{\text{LBS1}}$/LSDDQM[1]	C23	O	BV$_{DD}$	5, 9
$\overline{\text{LWE2}}$/$\overline{\text{LBS2}}$/LSDDQM[2]	J21	O	BV$_{DD}$	5, 9
$\overline{\text{LWE3}}$/$\overline{\text{LBS3}}$/LSDDQM[3]	A24	O	BV$_{DD}$	5, 9
LALE	H24	O	BV$_{DD}$	5, 8, 9
LBCTL	G27	O	BV$_{DD}$	5, 8, 9
LGPL0/LSDA10	F23	O	BV$_{DD}$	5, 9
LGPL1/$\overline{\text{LSDWE}}$	G22	O	BV$_{DD}$	5, 9
LGPL2/$\overline{\text{LOE}}$/$\overline{\text{LSDRAS}}$	B27	O	BV$_{DD}$	5, 8, 9
LGPL3/$\overline{\text{LSDCAS}}$	F24	O	BV$_{DD}$	5, 9
LGPL4/$\overline{\text{LGTA}}$/LUPWAIT/LPBSE	H23	I/O	BV$_{DD}$	—
LGPL5	E26	O	BV$_{DD}$	5, 9
LCKE	E24	O	BV$_{DD}$	—
LCLK[0:2]	E23, D24, H22	O	BV$_{DD}$	—

Table 74. MPC8543E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
IIC1_SDA	AG21	I/O	OV_{DD}	4, 27
IIC2_SCL	AG15	I/O	OV_{DD}	4, 27
IIC2_SDA	AG14	I/O	OV_{DD}	4, 27
SerDes				
SD_RX[0:7]	M28, N26, P28, R26, W26, Y28, AA26, AB28	I	XV_{DD}	—
$\overline{\text{SD\_RX}}$[0:7]	M27, N25, P27, R25, W25, Y27, AA25, AB27	I	XV_{DD}	—
SD_TX[0:7]	M22, N20, P22, R20, U20, V22, W20, Y22	O	XV_{DD}	—
$\overline{\text{SD\_TX}}$[0:7]	M23, N21, P23, R21, U21, V23, W21, Y23	O	XV_{DD}	—
SD_PLL_TPD	U28	O	XV_{DD}	24
SD_REF_CLK	T28	I	XV_{DD}	—
$\overline{\text{SD\_REF\_CLK}}$	T27	I	XV_{DD}	—
Reserved	AC1, AC3	—	—	2
Reserved	M26, V28	—	—	32
Reserved	M25, V27	—	—	34
Reserved	M20, M21, T22, T23	—	—	38
General-Purpose Output				
GPOUT[24:31]	K26, K25, H27, G28, H25, J26, K24, K23	O	BV_{DD}	—
System Control				
$\overline{\text{HRESET}}$	AG17	I	OV_{DD}	—
$\overline{\text{HRESET\_REQ}}$	AG16	O	OV_{DD}	29
$\overline{\text{SRESET}}$	AG20	I	OV_{DD}	—
$\overline{\text{CKSTP\_IN}}$	AA9	I	OV_{DD}	—
$\overline{\text{CKSTP\_OUT}}$	AA8	O	OV_{DD}	2, 4
Debug				
TRIG_IN	AB2	I	OV_{DD}	—
TRIG_OUT/READY/$\overline{\text{QUIESCE}}$	AB1	O	OV_{DD}	6, 9, 19, 29
MSRCID[0:1]	AE4, AG2	O	OV_{DD}	5, 6, 9
MSRCID[2:4]	AF3, AF1, AF2	O	OV_{DD}	6, 19, 29
MDVAL	AE5	O	OV_{DD}	6
CLK_OUT	AE21	O	OV_{DD}	11
Clock				
RTC	AF16	I	OV_{DD}	—
SYSCLK	AH17	I	OV_{DD}	—

Table 74. MPC8543E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
JTAG				
TCK	AG28	I	OV_{DD}	—
TDI	AH28	I	OV_{DD}	12
TDO	AF28	O	OV_{DD}	—
TMS	AH27	I	OV_{DD}	12
$\overline{\text{TRST}}$	AH23	I	OV_{DD}	12
DFT				
L1_TSTCLK	AC25	I	OV_{DD}	25
L2_TSTCLK	AE22	I	OV_{DD}	25
$\overline{\text{LSSD\_MODE}}$	AH20	I	OV_{DD}	25
TEST_SEL	AH14	I	OV_{DD}	109
Thermal Management				
THERM0	AG1	—	—	14
THERM1	AH1	—	—	14
Power Management				
ASLEEP	AH18	O	OV_{DD}	9, 19, 29
Power and Ground Signals				
GND	A11, B7, B24, C1, C3, C5, C12, C15, C26, D8, D11, D16, D20, D22, E1, E5, E9, E12, E15, E17, F4, F26, G12, G15, G18, G21, G24, H2, H6, H8, H28, J4, J12, J15, J17, J27, K7, K9, K11, K27, L3, L5, L12, L16, N11, N13, N15, N17, N19, P4, P9, P12, P14, P16, P18, R11, R13, R15, R17, R19, T4, T12, T14, T16, T18, U8, U11, U13, U15, U17, U19, V4, V12, V18, W6, W19, Y4, Y9, Y11, Y19, AA6, AA14, AA17, AA22, AA23, AB4, AC2, AC11, AC19, AC26, AD5, AD9, AD22, AE3, AE14, AF6, AF10, AF13, AG8, AG27, K28, L24, L26, N24, N27, P25, R28, T24, T26, U24, V25, W28, Y24, Y26, AA24, AA27, AB25, AC28, L21, L23, N22, P20, R23, T21, U22, V20, W23, Y21, U27	—	—	—
OV_{DD}	V16, W11, W14, Y18, AA13, AA21, AB11, AB17, AB24, AC4, AC9, AC21, AD6, AD13, AD17, AD19, AE10, AE8, AE24, AF4, AF12, AF22, AF27, AG26	Power for PCI and other standards (3.3 V)	OV_{DD}	—
LV_{DD}	N8, R7, T9, U6	Power for TSEC1 and TSEC2 (2.5 V, 3.3 V)	LV_{DD}	—

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