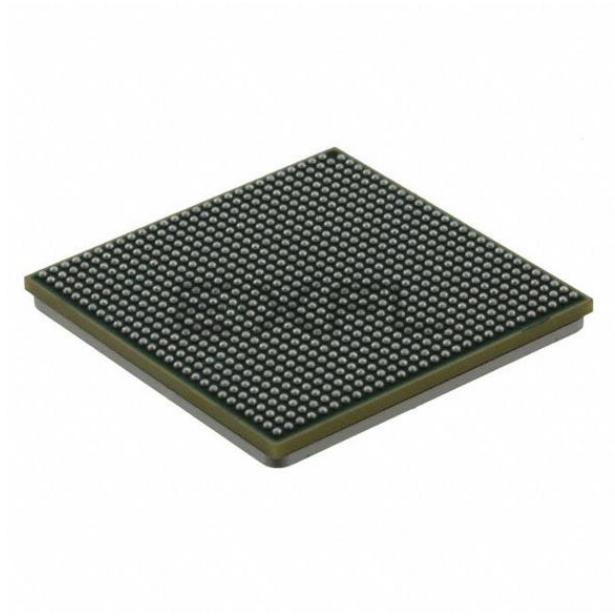

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.333GHz
Co-Processors/DSP	Signal Processing; SPE, Security; SEC
RAM Controllers	DDR, DDR2, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCBGA (29x29)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc8547evtaujb

2.1.3 Output Driver Characteristics

The following table provides information on the characteristics of the output driver strengths. The values are preliminary estimates.

Table 3. Output Drive Capability

Driver Type	Programmable Output Impedance (Ω)	Supply Voltage	Notes
Local bus interface utilities signals	25 25	BV_{DD} = 3.3 V BV_{DD} = 2.5 V	1
	45(default) 45(default)	BV_{DD} = 3.3 V BV_{DD} = 2.5 V	
PCI signals	25	OV_{DD} = 3.3 V	2
	45(default)		
DDR signal	18 36 (half strength mode)	GV_{DD} = 2.5 V	3
DDR2 signal	18 36 (half strength mode)	GV_{DD} = 1.8 V	3
TSEC/10/100 signals	45	L/TV_{DD} = 2.5/3.3 V	—
DUART, system control, JTAG	45	OV_{DD} = 3.3 V	—
I2C	150	OV_{DD} = 3.3 V	—

Notes:

1. The drive strength of the local bus interface is determined by the configuration of the appropriate bits in PORIMPSCR.
2. The drive strength of the PCI interface is determined by the setting of the $\overline{\text{PCI\_GNT1}}$ signal at reset.
3. The drive strength of the DDR interface in half-strength mode is at T_j = 105°C and at GV_{DD} (min).

2.2 Power Sequencing

The device requires its power rails to be applied in a specific sequence in order to ensure proper device operation. These requirements are as follows for power-up:

1. V_{DD}, AV_{DD}_n, BV_{DD}, LV_{DD}, OV_{DD}, SV_{DD}, TV_{DD}, XV_{DD}
2. GV_{DD}

All supplies must be at their stable values within 50 ms.

NOTE

Items on the same line have no ordering requirement with respect to one another. Items on separate lines must be ordered sequentially such that voltage rails on a previous step must reach 90% of their value before the voltage rails on the current step reach 10% of theirs.

NOTE

In order to guarantee MCKE low during power-up, the above sequencing for GV_{DD} is required. If there is no concern about any of the DDR signals being in an indeterminate state during power-up, then the sequencing for GV_{DD} is not required.

3 Power Characteristics

The estimated typical power dissipation for the core complex bus (CCB) versus the core frequency for this family of PowerQUICC III devices is shown in the following table.

Table 4. Device Power Dissipation

CCB Frequency[1]	Core Frequency	SLEEP[2]	Typical-65[3]	Typical-105[4]	Maximum[5]	Unit
400	800	2.7	4.6	7.5	8.1	W
	1000	2.7	5.0	7.9	8.5	W
	1200	2.7	5.4	8.3	8.9	
500	1500	11.5	13.6	16.5	18.6	W
533	1333	6.2	7.9	10.8	12.8	W

Notes:

1. CCB frequency is the SoC platform frequency, which corresponds to the DDR data rate.
2. SLEEP is based on V_{DD} = 1.1 V, T_j = 65°C.
3. Typical-65 is based on V_{DD} = 1.1 V, T_j = 65°C, running Dhrystone.
4. Typical-105 is based on V_{DD} = 1.1 V, T_j = 105°C, running Dhrystone.
5. Maximum is based on V_{DD} = 1.1 V, T_j = 105°C, running a smoke test.

4.3 eTSEC Gigabit Reference Clock Timing

The following table provides the eTSEC gigabit reference clocks (EC_GTX_CLK125) AC timing specifications for the device.

Table 6. EC_GTX_CLK125 AC Timing Specifications

Parameter/Condition	Symbol	Min	Typ	Max	Unit	Notes
EC_GTX_CLK125 frequency	f_{G125}	—	125	—	MHz	—
EC_GTX_CLK125 cycle time	t_{G125}	—	8	—	ns	
EC_GTX_CLK125 rise and fall time L/TVDD = 2.5 V L/TVDD = 3.3 V	t_{G125R}, t_{G125F}	—	—	 0.75 1.0	ns	1
EC_GTX_CLK125 duty cycle GMII, TBI 1000Base-T for RGMII, RTBI	t_{G125H}/t_{G125}	 45 47	—	 55 53	%	2, 3

Notes:

1. Rise and fall times for EC_GTX_CLK125 are measured from 0.5 and 2.0 V for L/TV_{DD} = 2.5 V, and from 0.6 and 2.7 V for L/TV_{DD} = 3.3 V.
2. Timing is guaranteed by design and characterization.
3. EC_GTX_CLK125 is used to generate the GTX clock TSEC*n*_GTX_CLK for the eTSEC transmitter with 2% degradation. EC_GTX_CLK125 duty cycle can be loosened from 47/53% as long as the PHY device can tolerate the duty cycle generated by the TSEC*n*_ GTX_CLK. See Section 8.2.6, "RGMII and RTBI AC Timing Specifications," for duty cycle for 10Base-T and 100Base-T reference clock.

4.4 PCI/PCI-X Reference Clock Timing

When the PCI/PCI-X controller is configured for asynchronous operation, the reference clock for the PCI/PCI-x controller is not the SYSCLK input, but instead the PCI*n*_CLK. The following table provides the PCI/PCI-X reference clock AC timing specifications for the device.

Table 7. PCI*n*_CLK AC Timing Specifications

At recommended operating conditions (see Table 2) with OV_{DD} = 3.3 V ± 165 mV.

Parameter/Condition	Symbol	Min	Typ	Max	Unit	Notes
PCI*n*_CLK frequency	f_{PCICLK}	16	—	133	MHz	—
PCI*n*_CLK cycle time	t_{PCICLK}	7.5	—	60	ns	—
PCI*n*_CLK rise and fall time	t_{PCIKH}, t_{PCIKL}	0.6	1.0	2.1	ns	1, 2
PCI*n*_CLK duty cycle	$t_{PCIKHKL}/t_{PCICLK}$	40	—	60	%	2

Notes:

1. Rise and fall times for SYSCLK are measured at 0.6 and 2.7 V.
2. Timing is guaranteed by design and characterization.

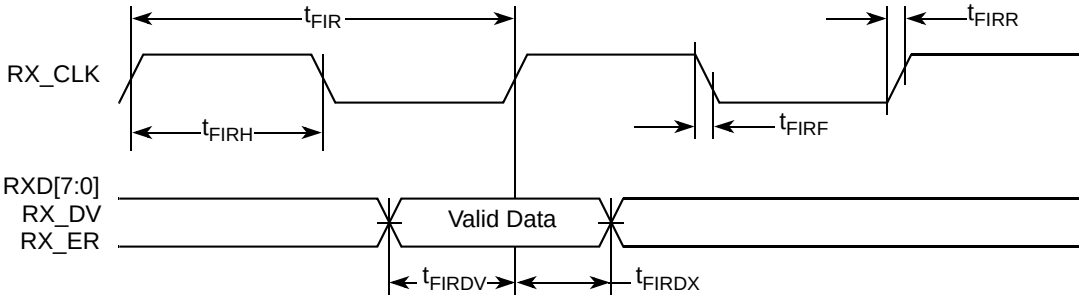

Figure 7. FIFO Receive AC Timing Diagram

8.2.2 GMII AC Timing Specifications

This section describes the GMII transmit and receive AC timing specifications.

8.2.2.1 GMII Transmit AC Timing Specifications

This table provides the GMII transmit AC timing specifications.

Table 26. GMII Transmit AC Timing Specifications

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
GMII data TXD[7:0], TX_ER, TX_EN setup time	t_{GTKHDV}	2.5	—	—	ns
GTX_CLK to GMII data TXD[7:0], TX_ER, TX_EN delay	t_{GTKHDX}	0.5	—	5.0	ns
GTX_CLK data clock rise time (20%–80%)	t_{GTXR}[2]	—	—	1.0	ns
GTX_CLK data clock fall time (80%–20%)	t_{GTXF}[2]	—	—	1.0	ns

Notes:

1. The symbols used for timing specifications follow the pattern $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GTKHDV} symbolizes GMII transmit timing (GT) with respect to the t_{GTX} clock reference (K) going to the high state (H) relative to the time date input signals (D) reaching the valid state (V) to state or setup time. Also, t_{GTKHDX} symbolizes GMII transmit timing (GT) with respect to the t_{GTX} clock reference (K) going to the high state (H) relative to the time date input signals (D) going invalid (X) or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GTX} represents the GMII(G) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. Guaranteed by design.

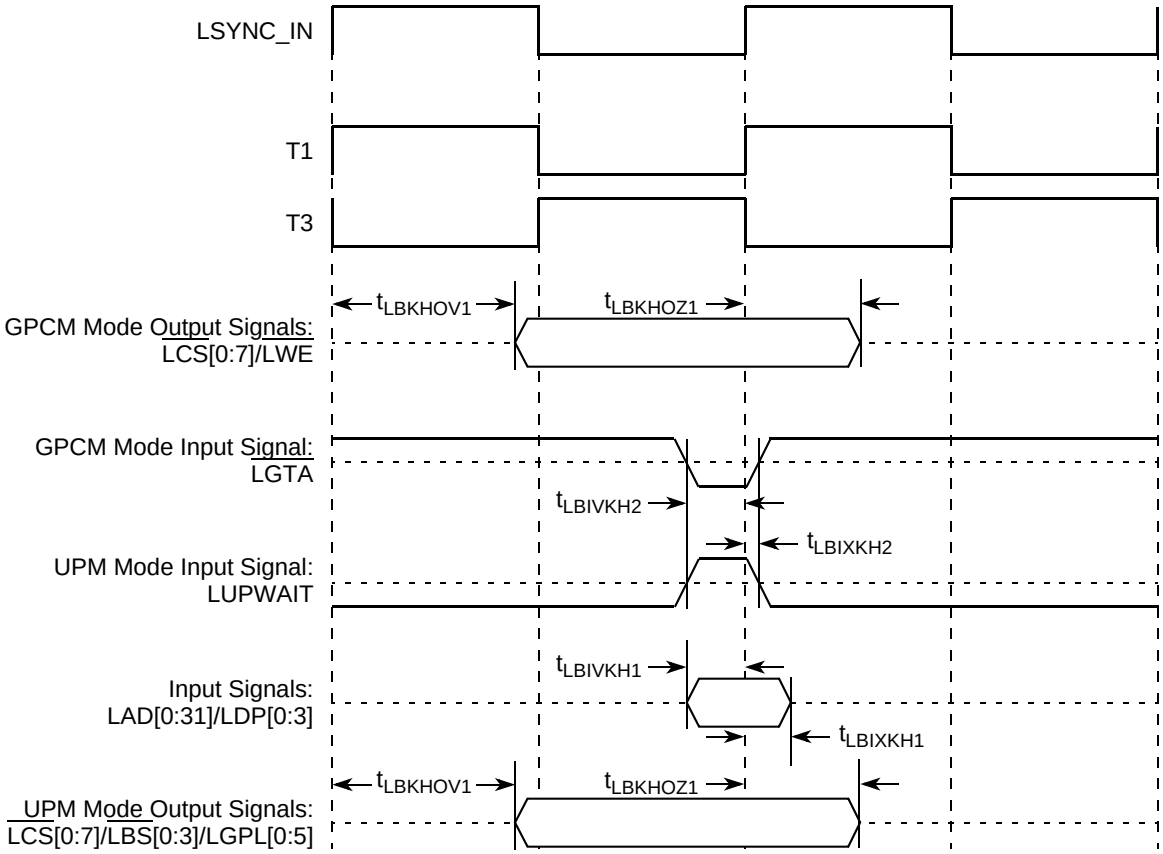

Figure 25. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 (PLL Enabled)

13 I²C

This section describes the DC and AC electrical characteristics for the I²C interfaces of the device.

13.1 I²C DC Electrical Characteristics

This table provides the DC electrical characteristics for the I²C interfaces.

Table 45. I²C DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage level	V_{IH}	$0.7 \times OV_{DD}$	$OV_{DD} + 0.3$	V	—
Input low voltage level	V_{IL}	–0.3	$0.3 \times OV_{DD}$	V	—
Low level output voltage	V_{OL}	0	$0.2 \times OV_{DD}$	V	1
Pulse width of spikes which must be suppressed by the input filter	t_{I2KHKL}	0	50	ns	2
Input current each I/O pin (input voltage is between $0.1 \times OV_{DD}$ and $0.9 \times OV_{DD}$(max)	I_I	–10	10	μA	3
Capacitance for each I/O pin	C_I	—	10	pF	—

Notes:

1. Output voltage (open drain or open collector) condition = 3 mA sink current.
2. See the *MPC8548E PowerQUICC™ III Integrated Processor Family Reference Manual*, for information on the digital filter used.
3. I/O pins obstruct the SDA and SCL lines if OV_{DD} is switched off.

13.2 I²C AC Electrical Specifications

This table provides the AC timing parameters for the I²C interfaces.

Table 46. I²C AC Electrical Specifications

Parameter	Symbol[1]	Min	Max	Unit	Notes
SCL clock frequency	f_{I2C}	0	400	kHz	—
Low period of the SCL clock	t_{I2CL}	1.3	—	μs	4
High period of the SCL clock	t_{I2CH}	0.6	—	μs	4
Setup time for a repeated START condition	t_{I2SVKH}	0.6	—	μs	4
Hold time (repeated) START condition (after this period, the first clock pulse is generated)	t_{I2SXKL}	0.6	—	μs	4
Data setup time	t_{I2DVKH}	100	—	ns	4
Data input hold time: CBUS compatible masters I²C bus devices	t_{I2DXKL}	 — 0	 — —	μs	2
Data output delay time:	t_{I2OVKL}	—	0.9	—	3
Set-up time for STOP condition	t_{I2PVKH}	0.6	—	μs	—
Bus free time between a STOP and START condition	t_{I2KHDX}	1.3	—	μs	—

Figure 36 shows the PCI/PCI-X input AC timing conditions.

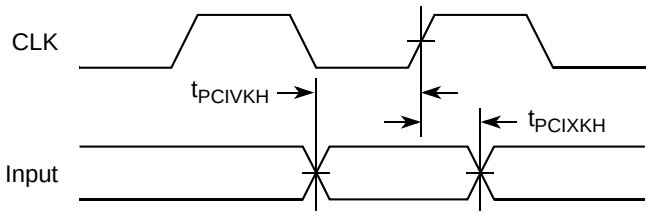

Figure 36. PCI/PCI-X Input AC Timing Measurement Conditions

Figure 37 shows the PCI/PCI-X output AC timing conditions.

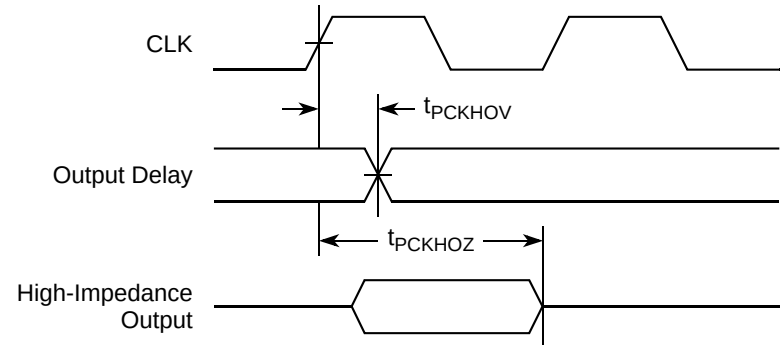

Figure 37. PCI/PCI-X Output AC Timing Measurement Condition

Table 53 provides the PCI-X AC timing specifications at 66 MHz.

Table 53. PCI-X AC Timing Specifications at 66 MHz

Parameter	Symbol	Min	Max	Unit	Notes
SYSCLK to signal valid delay	t_{PCKHOV}	—	3.8	ns	1, 2, 3, 7, 8
Output hold from SYSCLK	t_{PCKHOX}	0.7	—	ns	1, 10
SYSCLK to output high impedance	t_{PCKHOZ}	—	7	ns	1, 4, 8, 11
Input setup time to SYSCLK	t_{PCIVKH}	1.7	—	ns	3, 5
Input hold time from SYSCLK	t_{PCIXKH}	0.5	—	ns	10
$\overline{\text{REQ64}}$ to $\overline{\text{HRESET}}$ setup time	t_{PCRVRH}	10	—	clocks	11
$\overline{\text{HRESET}}$ to $\overline{\text{REQ64}}$ hold time	t_{PCRHRX}	0	50	ns	11
$\overline{\text{HRESET}}$ high to first $\overline{\text{FRAME}}$ assertion	t_{PCRHFV}	10	—	clocks	9, 11
PCI-X initialization pattern to $\overline{\text{HRESET}}$ setup time	t_{PCIVRH}	10	—	clocks	11

Table 54. PCI-X AC Timing Specifications at 133 MHz (continued)

Parameter	Symbol	Min	Max	Unit	Notes
HRESET to PCI-X initialization pattern hold time	t_{PCRHIX}	0	50	ns	6, 12

Notes:

1. See the timing measurement conditions in the *PCI-X 1.0a Specification*.
2. Minimum times are measured at the package pin (not the test point). Maximum times are measured with the test point and load circuit.
3. Setup time for point-to-point signals applies to REQ and GNT only. All other signals are bused.
4. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
5. Setup time applies only when the device is not driving the pin. Devices cannot drive and receive signals at the same time.
6. Maximum value is also limited by delay to the first transaction (time for HRESET high to first configuration access, t_{PCRHFV}). The PCI-X initialization pattern control signals after the rising edge of HRESET must be negated no later than two clocks before the first FRAME and must be floated no later than one clock before FRAME is asserted.
7. A PCI-X device is permitted to have the minimum values shown for t_{PCKHOV} and t_{CYC} only in PCI-X mode. In conventional mode, the device must meet the requirements specified in PCI 2.2 for the appropriate clock frequency.
8. Device must meet this specification independent of how many outputs switch simultaneously.
9. The timing parameter t_{PCIVKH} is a minimum of 1.4 ns rather than the minimum of 1.2 ns in the *PCI-X 1.0a Specification.*
10. The timing parameter t_{PCRHFV} is a minimum of 10 clocks rather than the minimum of 5 clocks in the *PCI-X 1.0a Specification.*
11. Guaranteed by characterization.
12. Guaranteed by design.

Table 57. Differential Receiver (RX) Input Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Unit	Comments
$L_{TX-SKEW}$	Total Skew	—	—	20	ns	Skew across all lanes on a Link. This includes variation in the length of SKP ordered set (for example, COM and one to five symbols) at the RX as well as any delay differences arising from the interconnect itself.

Notes:

1. No test load is necessarily associated with this value.
2. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in Figure 50 must be used as the RX device when taking measurements (also see the receiver compliance eye diagram shown in Figure 49). If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
3. A T_{RX-EYE} = 0.40 UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX-EYE-MEDIAN-to-MAX-JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive TX UIs. Note that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
4. The receiver input impedance shall result in a differential return loss greater than or equal to 15 dB with the D+ line biased to 300 mV and the D– line biased to –{300 mV and a common mode return loss greater than or equal to 6 dB (no bias required) over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements for is 50 Ω to ground for both the D+ and D– line (that is, as measured by a vector network analyzer with 50-Ω probes—see Figure 50). Note: that the series capacitors CTX is optional for the return loss measurement.
5. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM) there is a 5 ms transition time before receiver termination values must be met on all unconfigured lanes of a port.
6. The RX DC common mode Impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the RX ground.
7. It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

17.5 Receiver Compliance Eye Diagrams

The RX eye diagram in Figure 49 is specified using the passive compliance/test measurement load (see Figure 50) in place of any real PCI Express RX component.

Note: In general, the minimum receiver eye diagram measured with the compliance/test measurement load (see Figure 50) is larger than the minimum receiver eye diagram measured over a range of systems at the input receiver of any real PCI Express component. The degraded eye diagram at the input receiver is due to traces internal to the package as well as silicon parasitic characteristics which cause the real PCI Express component to vary in impedance from the compliance/test measurement load. The input receiver eye diagram is implementation specific and is not specified. RX component designer must provide additional margin to adequately compensate for the degraded minimum receiver eye diagram (shown in Figure 49) expected at the input receiver based on some adequate combination of system simulations and the return loss measured looking into the RX package and silicon. The RX eye diagram must be aligned in time using the jitter median to locate the center of the eye diagram.

18 Serial RapidIO

This section describes the DC and AC electrical specifications for the RapidIO interface of the MPC8548E, for the LP-Serial physical layer. The electrical specifications cover both single- and multiple-lane links. Two transmitters (short and long run) and a single receiver are specified for each of three baud rates, 1.25, 2.50, and 3.125 GBaud.

Two transmitter specifications allow for solutions ranging from simple board-to-board interconnect to driving two connectors across a backplane. A single receiver specification is given that accepts signals from both the short- and long-run transmitter specifications.

The short-run transmitter must be used mainly for chip-to-chip connections on either the same printed-circuit board or across a single connector. This covers the case where connections are made to a mezzanine (daughter) card. The minimum swings of the short-run specification reduce the overall power used by the transceivers.

The long-run transmitter specifications use larger voltage swings that are capable of driving signals across backplanes. This allows a user to drive signals across two connectors and a backplane. The specifications allow a distance of at least 50 cm at all baud rates.

All unit intervals are specified with a tolerance of ±100 ppm. The worst case frequency difference between any transmit and receive clock is 200 ppm.

To ensure interoperability between drivers and receivers of different vendors and technologies, AC coupling at the receiver input must be used.

18.1 DC Requirements for Serial RapidIO SD_REF_CLK and SD_REF_CLK

For more information, see Section 16.2, “SerDes Reference Clocks.”

18.2 AC Requirements for Serial RapidIO SD_REF_CLK and SD_REF_CLK

Table 58 lists the Serial RapidIO SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ AC requirements.

Table 58. SD_REF_CLK and $\overline{\text{SD\_REF\_CLK}}$ AC Requirements

Symbol	Parameter Description	Min	Typ	Max	Unit	Comments
t_{REF}	REFCLK cycle time	—	10(8)	—	ns	8 ns applies only to serial RapidIO with 125-MHz reference clock
t_{REFCJ}	REFCLK cycle-to-cycle jitter. Difference in the period of any two adjacent REFCLK cycles.	—	—	80	ps	—
t_{REFPJ}	Phase jitter. Deviation in edge location with respect to mean edge location.	–40	—	40	ps	—

18.3 Signal Definitions

LP-serial links use differential signaling. This section defines terms used in the description and specification of differential signals. Figure 51 shows how the signals are defined. The figures show waveforms for either a transmitter output (TD and $\overline{\text{TD}}$) or a receiver input (RD and $\overline{\text{RD}}$). Each signal swings between A volts and B volts where A > B. Using these waveforms, the definitions are as follows:

1. The transmitter output signals and the receiver input signals TD, $\overline{\text{TD}}$, RD, and $\overline{\text{RD}}$ each have a peak-to-peak swing of A – B volts.
2. The differential output signal of the transmitter, V_{OD}, is defined as $V_{TD} - V_{\overline{TD}}$.
3. The differential input signal of the receiver, V_{ID}, is defined as $V_{RD} - V_{\overline{RD}}$.
4. The differential output signal of the transmitter and the differential input signal of the receiver each range from A – B to –(A – B) volts.
5. The peak value of the differential transmitter output signal and the differential receiver input signal is A – B volts.
6. The peak-to-peak value of the differential transmitter output signal and the differential receiver input signal is 2 × (A – B) volts.

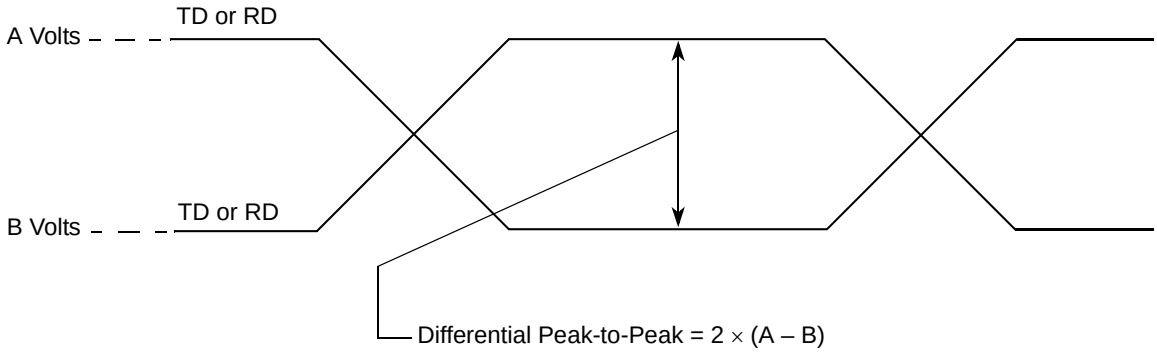

Figure 51. Differential Peak–Peak Voltage of Transmitter or Receiver

To illustrate these definitions using real values, consider the case of a CML (current mode logic) transmitter that has a common mode voltage of 2.25 V and each of its outputs, TD and $\overline{\text{TD}}$, has a swing that goes between 2.5 and 2.0 V. Using these values, the peak-to-peak voltage swing of the signals TD and $\overline{\text{TD}}$ is 500 mVp-p. The differential output signal ranges between 500 and –500 mV. The peak differential voltage is 500 mV. The peak-to-peak differential voltage is 1000 mVp-p.

18.4 Equalization

With the use of high-speed serial links, the interconnect media causes degradation of the signal at the receiver. Effects such as inter-symbol interference (ISI) or data dependent jitter are produced. This loss can be large enough to degrade the eye opening at the receiver beyond what is allowed in the specification. To negate a portion of these effects, equalization can be used. The most common equalization techniques that can be used are:

- A passive high pass filter network placed at the receiver. This is often referred to as passive equalization.
- The use of active circuits in the receiver. This is often referred to as adaptive equalization.

Table 63. Long Run Transmitter AC Timing Specifications—2.5 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Output voltage	V_O	–0.40	2.30	V	Voltage relative to COMMON of either signal comprising a differential pair
Differential output voltage	V_{DIFFPP}	800	1600	mVp-p	—
Deterministic jitter	J_D	—	0.17	UI p-p	—
Total jitter	J_T	—	0.35	UI p-p	—
Multiple output skew	S_{MO}	—	1000	ps	Skew at the transmitter output between lanes of a multilane link
Unit interval	UI	400	400	ps	±100 ppm

Table 64. Long Run Transmitter AC Timing Specifications—3.125 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Output voltage	V_O	–0.40	2.30	V	Voltage relative to COMMON of either signal comprising a differential pair
Differential output voltage	V_{DIFFPP}	800	1600	mVp-p	—
Deterministic jitter	J_D	—	0.17	UI p-p	—
Total jitter	J_T	—	0.35	UI p-p	—
Multiple output skew	S_{MO}	—	1000	ps	Skew at the transmitter output between lanes of a multilane link
Unit interval	UI	320	320	ps	±100 ppm

For each baud rate at which an LP-serial transmitter is specified to operate, the output eye pattern of the transmitter shall fall entirely within the unshaded portion of the transmitter output compliance mask shown in Figure 52 with the parameters specified in Table 65 when measured at the output pins of the device and the device is driving a 100-Ω ± 5% differential resistive load. The output eye pattern of an LP-serial

Notes:

1. All dimensions are in millimeters.
2. Dimensioning and tolerancing per ASME Y14.5M-1994.
3. Maximum solder ball diameter measured parallel to datum A.
4. Datum A, the seating plane, is determined by the spherical crowns of the solder balls.
5. Parallelism measurement shall exclude any effect of mark on top surface of package.
6. All dimensions are symmetric across the package center lines unless dimensioned otherwise.

Table 71. MPC8548E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes

25.These are test signals for factory use only and must be pulled up (100 Ω–1 kΩ) to OV_{DD} for normal machine operation.

26.Independent supplies derived from board V_{DD}.

27.Recommend a pull-up resistor (~1 kΩ) be placed on this pin to OV_{DD}.

29. The following pins must NOT be pulled down during power-on reset: TSEC3_TXD[3], TSEC4_TXD3/TSEC3_TXD7, HRESET_REQ, TRIG_OUT/READY/QUIESCE, MSRCID[2:4], ASLEEP.

30.This pin requires an external 4.7-kΩ pull-down resistor to prevent PHY from seeing a valid transmit enable before it is actively driven.

31.This pin is only an output in eTSEC3 FIFO mode when used as Rx flow control.

32.These pins must be connected to XV_{DD}.

33.TSEC2_TXD1, TSEC2_TX_ER are multiplexed as cfg_dram_type[0:1]. They must be valid at power-up, even before HRESET assertion.

34.These pins must be pulled to ground through a 300-Ω (±10%) resistor.

35.When a PCI block is disabled, either the POR config pin that selects between internal and external arbiter must be pulled down to select external arbiter if there is any other PCI device connected on the PCI bus, or leave the PCI*n*_AD pins as 'no connect' or terminated through 2–10 kΩ pull-up resistors with the default of internal arbiter if the PCI*n*_AD pins are not connected to any other PCI device. The PCI block drives the PCI*n*_AD pins if it is configured to be the PCI arbiter—through POR config pins—irrespective of whether it is disabled via the DEVDISR register or not. It may cause contention if there is any other PCI device connected on the bus.

36.MDIC0 is grounded through an 18.2-Ω precision 1% resistor and MDIC1 is connected to GV_{DD} through an 18.2-Ω precision 1% resistor. These pins are used for automatic calibration of the DDR IOs.

38.These pins must be left floating.

39. If PCI1 or PCI2 is configured as PCI asynchronous mode, a valid clock must be provided on pin PCI1_CLK or PCI2_CLK. Otherwise the processor will not boot up.

40.These pins must be connected to GND.

101.This pin requires an external 4.7-kΩ resistor to GND.

102.For Rev. 2.x silicon, DMA_DACK[0:1] must be 0b11 during POR configuration; for rev. 1.x silicon, the pin values during POR configuration are don't care.

103.If these pins are not used as GPIN*n* (general-purpose input), they must be pulled low (to GND) or high (to LV_{DD}) through 2–10 kΩ resistors.

104.These must be pulled low to GND through 2–10 kΩ resistors if they are not used.

105.These must be pulled low or high to LV_{DD} through 2–10 kΩ resistors if they are not used.

106.For rev. 2.x silicon, DMA_DACK[0:1] must be 0b10 during POR configuration; for rev. 1.x silicon, the pin values during POR configuration are don't care.

107.For rev. 2.x silicon, DMA_DACK[0:1] must be 0b01 during POR configuration; for rev. 1.x silicon, the pin values during POR configuration are don't care.

108.For rev. 2.x silicon, DMA_DACK[0:1] must be 0b11 during POR configuration; for rev. 1.x silicon, the pin values during POR configuration are don't care.

109.This is a test signal for factory use only and must be pulled down (100 Ω – 1 kΩ) to GND for normal machine operation.

110.These pins must be pulled high to OV_{DD} through 2–10 kΩ resistors.

111.If these pins are not used as GPIN*n* (general-purpose input), they must be pulled low (to GND) or high (to OV_{DD}) through 2–10 kΩ resistors.

112.This pin must not be pulled down during POR configuration.

113.These should be pulled low or high to OV_{DD} through 2–10 kΩ resistors.

Table 72. MPC8547E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
Reserved	AE26	—	—	2
cfg_pci1_clk	AG24	I	OV_{DD}	5
Reserved	AF25	—	—	101
Reserved	AE25	—	—	2
Reserved	AG25	—	—	2
Reserved	AD24	—	—	2
Reserved	AF24	—	—	2
Reserved	AD27	—	—	2
Reserved	AD28, AE27, W17, AF26	—	—	2
Reserved	AH25	—	—	2
DDR SDRAM Memory Interface				
MDQ[0:63]	L18, J18, K14, L13, L19, M18, L15, L14, A17, B17, A13, B12, C18, B18, B13, A12, H18, F18, J14, F15, K19, J19, H16, K15, D17, G16, K13, D14, D18, F17, F14, E14, A7, A6, D5, A4, C8, D7, B5, B4, A2, B1, D1, E4, A3, B2, D2, E3, F3, G4, J5, K5, F6, G5, J6, K4, J1, K2, M5, M3, J3, J2, L1, M6	I/O	GV_{DD}	—
MECC[0:7]	H13, F13, F11, C11, J13, G13, D12, M12	I/O	GV_{DD}	—
MDM[0:8]	M17, C16, K17, E16, B6, C4, H4, K1, E13	O	GV_{DD}	—
MDQS[0:8]	M15, A16, G17, G14, A5, D3, H1, L2, C13	I/O	GV_{DD}	—
$\overline{\text{MDQS}}$[0:8]	L17, B16, J16, H14, C6, C2, H3, L4, D13	I/O	GV_{DD}	—
MA[0:15]	A8, F9, D9, B9, A9, L10, M10, H10, K10, G10, B8, E10, B10, G6, A10, L11	O	GV_{DD}	—
MBA[0:2]	F7, J7, M11	O	GV_{DD}	—
$\overline{\text{MWE}}$	E7	O	GV_{DD}	—
$\overline{\text{MCAS}}$	H7	O	GV_{DD}	—
$\overline{\text{MRAS}}$	L8	O	GV_{DD}	—
MCKE[0:3]	F10, C10, J11, H11	O	GV_{DD}	11
$\overline{\text{MCS}}$[0:3]	K8, J8, G8, F8	O	GV_{DD}	—
MCK[0:5]	H9, B15, G2, M9, A14, F1	O	GV_{DD}	—
$\overline{\text{MCK}}$[0:5]	J9, A15, G1, L9, B14, F2	O	GV_{DD}	—
MODT[0:3]	E6, K6, L7, M7	O	GV_{DD}	—
MDIC[0:1]	A19, B19	I/O	GV_{DD}	36

Table 73. MPC8545E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$\overline{\text{UDE}}$	AH16	I	OV_{DD}	—
$\overline{\text{MCP}}$	AG19	I	OV_{DD}	—
IRQ[0:7]	AG23, AF18, AE18, AF20, AG18, AF17, AH24, AE20	I	OV_{DD}	—
IRQ[8]	AF19	I	OV_{DD}	—
IRQ[9]/$\overline{\text{DMA\_DREQ3}}$	AF21	I	OV_{DD}	1
IRQ[10]/$\overline{\text{DMA\_DACK3}}$	AE19	I/O	OV_{DD}	1
IRQ[11]/$\overline{\text{DMA\_DDONE3}}$	AD20	I/O	OV_{DD}	1
$\overline{\text{IRQ\_OUT}}$	AD18	O	OV_{DD}	2, 4
Ethernet Management Interface				
EC_MDC	AB9	O	OV_{DD}	5, 9
EC_MDIO	AC8	I/O	OV_{DD}	—
Gigabit Reference Clock				
EC_GTX_CLK125	V11	I	LV_{DD}	—
Three-Speed Ethernet Controller (Gigabit Ethernet 1)				
TSEC1_RXD[7:0]	R5, U1, R3, U2, V3, V1, T3, T2	I	LV_{DD}	—
TSEC1_TXD[7:0]	T10, V7, U10, U5, U4, V6, T5, T8	O	LV_{DD}	5, 9
TSEC1_COL	R4	I	LV_{DD}	—
TSEC1_CRS	V5	I/O	LV_{DD}	20
TSEC1_GTX_CLK	U7	O	LV_{DD}	—
TSEC1_RX_CLK	U3	I	LV_{DD}	—
TSEC1_RX_DV	V2	I	LV_{DD}	—
TSEC1_RX_ER	T1	I	LV_{DD}	—
TSEC1_TX_CLK	T6	I	LV_{DD}	—
TSEC1_TX_EN	U9	O	LV_{DD}	30
TSEC1_TX_ER	T7	O	LV_{DD}	—
GPIN[0:7]	P2, R2, N1, N2, P3, M2, M1, N3	I	LV_{DD}	103
GPOUT[0:5]	N9, N10, P8, N7, R9, N5	O	LV_{DD}	—
cfg_dram_type0/GPOUT6	R8	O	LV_{DD}	5, 9
GPOUT7	N6	O	LV_{DD}	—
Reserved	P1	—	—	104
Reserved	R6	—	—	104
Reserved	P6	—	—	15
Reserved	N4	—	—	105

Table 85. Package Thermal Characteristics for FC-PBGA (continued)

Characteristic	JEDEC Board	Symbol	Value	Unit	Notes
Die junction-to-board	N/A	$R_{\theta JB}$	5	°C/W	3
Die junction-to-case	N/A	$R_{\theta JC}$	0.8	°C/W	4

Notes:

1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, airflow, power dissipation of other components on the board, and board thermal resistance.
2. Per JEDEC JESD51-6 with the board (JESD51-7) horizontal.
3. Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
4. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1). The cold plate temperature is used for the case temperature, measured value includes the thermal resistance of the interface layer.

21.3 Heat Sink Solution

Every system application has different conditions that the thermal management solution must solve. As such, providing a recommended heat sink has not been found to be very useful. When a heat sink is chosen, give special consideration to the mounting technique. Mounting the heat sink to the printed-circuit board is the recommended procedure using a maximum of 10 lbs force (45 Newtons) perpendicular to the package and board. Clipping the heat sink to the package is not recommended.

22 System Design Information

This section provides electrical design recommendations for successful application of the device.

22.1 System Clocking

This device includes five PLLs, as follows:

1. The platform PLL generates the platform clock from the externally supplied SYSCLK input. The frequency ratio between the platform and SYSCLK is selected using the platform PLL ratio configuration bits as described in Section 20.2, “CCB/SYSCLK PLL Ratio.”
2. The e500 core PLL generates the core clock as a slave to the platform clock. The frequency ratio between the e500 core clock and the platform clock is selected using the e500 PLL ratio configuration bits as described in Section 20.3, “e500 Core PLL Ratio.”
3. The PCI PLL generates the clocking for the PCI bus.
4. The local bus PLL generates the clock for the local bus.
5. There is a PLL for the SerDes block.

22.2 PLL Power Supply Filtering

Each of the PLLs listed above is provided with power through independent power supply pins (AV_{DD}_PLAT, AV_{DD}_CORE, AV_{DD}_PCI, AV_{DD}_LBIU, and AV_{DD}_SRDS, respectively). The AV_{DD}

level must always be equivalent to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide independent filter circuits per PLL power supply as illustrated in Figure 57, one to each of the AV_{DD} pins. By providing independent filters to each PLL the opportunity to cause noise injection from one PLL to the other is reduced.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It must be built with surface mount capacitors with minimum Effective Series Inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit must be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It must be routed directly from the capacitors to the AV_{DD} pin, which is on the periphery of the footprint, without the inductance of vias.

Figure 57 through Figure 59 shows the PLL power supply filter circuits.

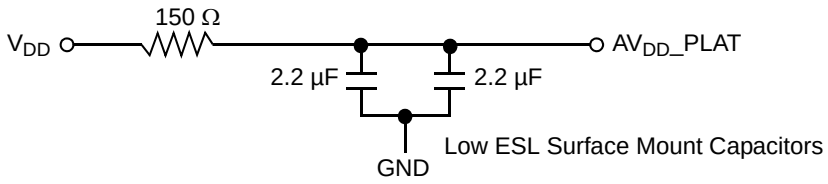

Figure 57. PLL Power Supply Filter Circuit with PLAT Pins

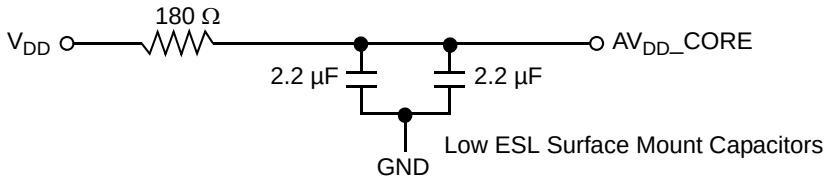

Figure 58. PLL Power Supply Filter Circuit with CORE Pins

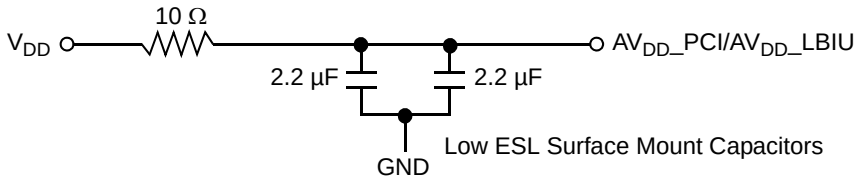

Figure 59. PLL Power Supply Filter Circuit with PCI/LBIU Pins

The AV_{DD}_SRDS signal provides power for the analog portions of the SerDes PLL. To ensure stability of the internal clock, the power supplied to the PLL is filtered using a circuit similar to the one shown in following figure. For maximum effectiveness, the filter circuit is placed as closely as possible to the AV_{DD}_SRDS ball to ensure it filters out as much noise as possible. The ground connection must be near the AV_{DD}_SRDS ball. The 0.003-µF capacitor is closest to the ball, followed by the two 2.2 µF capacitors, and finally the 1 Ω resistor to the board supply plane. The capacitors are connected from AV_{DD}_SRDS to

- $\overline{\text{SD\_REF\_CLK}}$

NOTE

It is recommended to power down the unused lane through SRDSCR1[0:7] register (offset = 0xE_0F08) (this prevents the oscillations and holds the receiver output in a fixed state) that maps to SERDES lane 0 to lane 7 accordingly.

Pins V28 and M26 must be tied to XV_{DD}. Pins V27 and M25 must be tied to GND through a 300-Ω resistor.

22.11 Guideline for PCI Interface Termination

PCI termination if PCI 1 or PCI 2 is not used at all.

Option 1

If PCI arbiter is enabled during POR:

- All AD pins are driven to the stable states after POR. Therefore, all ADs pins can be floating.
- All PCI control pins can be grouped together and tied to OV_{DD} through a single 10-kΩ resistor.
- It is optional to disable PCI block through DEVDISR register after POR reset.

Option 2

If PCI arbiter is disabled during POR:

- All AD pins are in the input state. Therefore, all ADs pins need to be grouped together and tied to OV_{DD} through a single (or multiple) 10-kΩ resistor(s).
- All PCI control pins can be grouped together and tied to OV_{DD} through a single 10-kΩ resistor.
- It is optional to disable PCI block through DEVDISR register after POR reset.

22.12 Guideline for LBIU Termination

If the LBIU parity pins are not used, the following is the termination recommendation:

- For LDP[0:3]—tie them to ground or the power supply rail via a 4.7-kΩ resistor.
- For LPBSE—tie it to the power supply rail via a 4.7-kΩ resistor (pull-up resistor).

24 Document Revision History

The following table provides a revision history for this hardware specification.

Table 88. Document Revision History

Rev. Number	Date	Substantive Change(s)
9	02/2012	• Updated Section 21.2, “Thermal for Version 2.1.1, 2.1.2, and 2.1.3 Silicon FC-PBGA with Full Lid and Version 3.1.x Silicon with Stamped Lid,” with version 3.0 silicon information. • Added Figure 56, “Mechanical Dimensions and Bottom Surface Nomenclature of the FC-PBGA with Stamped Lid.” • Updated Table 87, “Part Numbering Nomenclature,” with version 3.0 silicon information. • Removed Note from Section 5.1, “Power-On Ramp Rate”. • Changed the Table 10 title to “Power Supply Ramp Rate”. • Removed table 11. • Updated the title of Section 21.2, “Thermal for Version 2.1.1, 2.1.2, and 2.1.3 Silicon FC-PBGA with Full Lid and Version 3.1.x Silicon with Stamped Lid” to include Thermal Version 2.1.3 and Version 3.1.x Silicon. • Corrected the leaded Solder Ball composition in Table 70, “Package Parameters” • Updated Table 87, “Part Numbering Nomenclature,” with Version 3.1.x silicon information. • Updated the Min and Max value of TDO in the valid times row of Table 44, “JTAG AC Timing Specifications (Independent of SYSCLK)[1]” from 4 and 25 to 2 and 10 respectively .
8	04/2011	• Added Section 14.1, “GPOUT/GPIN Electrical Characteristics.” • Updated Table 71, “MPC8548E Pinout Listing,” Table 72, “MPC8547E Pinout Listing,” Table 73, “MPC8545E Pinout Listing,” and Table 74, “MPC8543E Pinout Listing,” to reflect that the TDO signal is not driven during HRSET* assertion. • Updated Table 87, “Part Numbering Nomenclature” with Ver. 2.1.3 silicon information.
7	09/2010	• In Table 37, “MII Management AC Timing Specifications, modified the fifth row from “MDC to MDIO delay tMDKHDX (16 × tptb_clk × 8) – 3 — (16 × tptb_clk × 8) + 3” to “MDC to MDIO delay tMDKHDX (16 × tCCB × 8) – 3 — (16 × tCCB × 8) + 3.” • Updated Figure 55, “Mechanical Dimensions and Bottom Surface Nomenclature of the HiCTE FC-CBGA and FC-PBGA with Full Lid and figure notes.
6	12/2009	• In Section 5.1, “Power-On Ramp Rate” added explanation that Power-On Ramp Rate is required to avoid falsely triggering ESD circuitry. • In Table 13 changed required ramp rate from 545 V/s for MVREF and VDD/XVDD/SVDD to 3500 V/s for MVREF and 4000 V/s for VDD. • In Table 13 deleted ramp rate requirement for XVDD/SVDD. • In Table 13 footnote 1 changed voltage range of concern from 0–400 mV to 20–500mV. • In Table 13 added footnote 2 explaining that VDD voltage ramp rate is intended to control ramp rate of AVDD pins.
5	10/2009	• In Table 27, “GMII Receive AC Timing Specifications,” changed duty cycle specification from 40/60 to 35/75 for RX_CLK duty cycle. • Updated tMDKHDX in Table 37, “MII Management AC Timing Specifications.” • Added a reference to Revision 2.1.2. • Updated Table 55, “MII Management AC Timing Specifications.” • Added Section 5.1, “Power-On Ramp Rate.”