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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of **Embedded - Microprocessors**

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

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Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.0GHz
Co-Processors/DSP	Signal Processing; SPE, Security; SEC
RAM Controllers	DDR, DDR2, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc8548ecvtaqgd

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

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Characteristic	Symbol	Max Value	Unit	Notes
Storage temperature range	T _{STG}	-55 to 150	°C	

Table 1. Absolute Maximum Ratings ¹ (continued)

Notes:

- 1. Functional and tested operating conditions are given in Table 2. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.
- 2. The -0.3 to 2.75 V range is for DDR and -0.3 to 1.98 V range is for DDR2.
- 3. The 3.63 V maximum is only supported when the port is configured in GMII, MII, RMII, or TBI modes; otherwise the 2.75 V maximum applies. See Section 8.2, "FIFO, GMII, MII, TBI, RGMII, RMII, and RTBI AC Timing Specifications," for details on the recommended operating conditions per protocol.
- 4. (M,L,O)V_{IN} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 2.

2.1.2 Recommended Operating Conditions

The following table provides the recommended operating conditions for this device. Note that the values in this table are the recommended and tested operating conditions. Proper device operation outside these conditions is not guaranteed.

	Characteristic	Symbol	Recommended Value	Unit	Notes
Core supply voltag	e	V _{DD}	1.1 V ± 55 mV	V	
PLL supply voltage		AV _{DD}	1.1 V ± 55 mV	V	1
Core power supply	for SerDes transceivers	SV _{DD}	1.1 V ± 55 mV	V	
Pad power supply	for SerDes transceivers	XV _{DD}	1.1 V ± 55 mV	V	
DDR and DDR2 DI	RAM I/O voltage	GV _{DD}	2.5 V ± 125 mV 1.8 V ± 90 mV	V	—
Three-speed Ethernet I/O voltage		LV _{DD}	3.3 V ± 165 mV 2.5 V ± 125 mV	V	4
		TV _{DD}	3.3 V ± 165 mV 2.5 V ± 125 mV	_	4
PCI/PCI-X, DUART Ethernet MII mana	F, system control and power management, I ² C, gement, and JTAG I/O voltage	OV _{DD}	3.3 V ± 165 mV	V	3
Local bus I/O volta	ge	BV _{DD}	3.3 V ± 165 mV 2.5 V ± 125 mV	V	—
Input voltage	DDR and DDR2 DRAM signals	MV _{IN}	GND to GV _{DD}	V	2
	DDR and DDR2 DRAM reference	MV _{REF}	GND to GV _{DD} /2	V	2
	Three-speed Ethernet signals	LV _{IN} TV _{IN}	GND to LV _{DD} GND to TV _{DD}	V	4
	Local bus signals	BV _{IN}	GND to BV _{DD}	V	
	PCI, DUART, SYSCLK, system control and power management, I ² C, Ethernet MII management, and JTAG signals	OV _{IN}	GND to OV _{DD}	V	3

Table 2. Recommended Operating Conditions

6.2.2 DDR SDRAM Output AC Timing Specifications

Table 19. DDR SDRAM Output AC Timing Specifications

At recommended operating conditions.

Parameter	Symbol ¹	Min	Мах	Unit	Notes
MCK[n] cycle time, MCK[<i>n</i>]/MCK[<i>n</i>] crossing	t _{MCK}	3.75	6	ns	2
ADDR/CMD output setup with respect to MCK 533 MHz 400 MHz 333 MHz	t _{ddkhas}	1.48 1.95 2.40		ns	3
ADDR/CMD output hold with respect to MCK 533 MHz 400 MHz 333 MHz	^t ddkhax	1.48 1.95 2.40		ns	3
MCS[<i>n</i>] output setup with respect to MCK 533 MHz 400 MHz 333 MHz	^t DDKHCS	1.48 1.95 2.40		ns	3
MCS[<i>n</i>] output hold with respect to MCK 533 MHz 400 MHz 333 MHz	^t DDKHCX	1.48 1.95 2.40		ns	3
MCK to MDQS Skew	t _{DDKHMH}	-0.6	0.6	ns	4
MDQ/MECC/MDM output setup with respect to MDQS 533 MHz 400 MHz 333 MHz	^t DDKHDS, ^t DDKLDS	538 700 900	 	ps	5
MDQ/MECC/MDM output hold with respect to MDQS 533 MHz 400 MHz 333 MHz	^t ddkhdx, ^t ddkldx	538 700 900		ps	5
MDQS preamble start	t _{DDKHMP}	$-0.5\times t_{MCK}-0.6$	$-0.5 \times t_{\text{MCK}} + 0.6$	ns	6



Figure 7. FIFO Receive AC Timing Diagram

8.2.2 GMII AC Timing Specifications

This section describes the GMII transmit and receive AC timing specifications.

8.2.2.1 GMII Transmit AC Timing Specifications

This table provides the GMII transmit AC timing specifications.

Table 26.	GMII	Transmit	AC	Timing	Specifications
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Parameter/Condition	Symbol ¹	Min	Тур	Max	Unit
GMII data TXD[7:0], TX_ER, TX_EN setup time	t _{GTKHDV}	2.5	_	_	ns
GTX_CLK to GMII data TXD[7:0], TX_ER, TX_EN delay	t _{GTKHDX}	0.5	_	5.0	ns
GTX_CLK data clock rise time (20%–80%)	t _{GTXR} ²	_	_	1.0	ns
GTX_CLK data clock fall time (80%–20%)	t _{GTXF} 2	—		1.0	ns

Notes:

The symbols used for timing specifications follow the pattern t<sub>(first two letters of functional block)(signal)(state)(reference)(state) for inputs and t_{(first two letters of functional block)(reference)(state)(signal)(state)} for outputs. For example, t_{GTKHDV} symbolizes GMII transmit timing (GT) with respect to the t_{GTX} clock reference (K) going to the high state (H) relative to the time date input signals (D) reaching the valid state (V) to state or setup time. Also, t_{GTKHDX} symbolizes GMII transmit timing (GT) with respect to the high state (H) relative to the time date input signals (D) reaching the clock reference (K) going to the high state (H) relative to the time date input signals (D) going invalid (X) or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GTX} represents the GMII(G) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
</sub>

2. Guaranteed by design.

Figure 11 shows the MII transmit AC timing diagram.



Figure 11. MII Transmit AC Timing Diagram

8.2.3.2 MII Receive AC Timing Specifications

This table provides the MII receive AC timing specifications.

Table 29. MII Receive A	C Timing Specifications
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Parameter/Condition	Symbol ¹	Min	Тур	Max	Unit
RX_CLK clock period 10 Mbps	t _{MRX} ²	_	400	—	ns
RX_CLK clock period 100 Mbps	t _{MRX}	—	40	—	ns
RX_CLK duty cycle	t _{MRXH} /t _{MRX}	35	_	65	%
RXD[3:0], RX_DV, RX_ER setup time to RX_CLK	t _{MRDVKH}	10.0	—	—	ns
RXD[3:0], RX_DV, RX_ER hold time to RX_CLK	t _{MRDXKH}	10.0	—	—	ns
RX_CLK clock rise (20%–80%)	t _{MRXR} ²	1.0	—	4.0	ns
RX_CLK clock fall time (80%–20%)	t _{MRXF} ²	1.0	_	4.0	ns

Notes:

1. The symbols used for timing specifications follow the pattern of t_{(first two letters of functional block)(signal)(state)(reference)(state) for inputs and t_{(first two letters of functional block)(reference)(state)(signal)(state)} for outputs. For example, t_{MRDVKH} symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MRX} clock reference (K) going to the high (H) state or setup time. Also, t_{MRDXKL} symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{MRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{MRX} represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).}

2. Guaranteed by design.

Figure 12 provides the AC test load for eTSEC.



Figure 12. eTSEC AC Test Load

10.2 Local Bus AC Electrical Specifications

This table describes the timing parameters of the local bus interface at $BV_{DD} = 3.3$ V. For information about the frequency range of local bus, see Section 20.1, "Clock Ranges."

Parameter	Symbol ¹	Min	Max	Unit	Notes
Local bus cycle time	t _{LBK}	7.5	12	ns	2
Local bus duty cycle	t _{LBKH/} t _{LBK}	43	57	%	—
LCLK[n] skew to LCLK[m] or LSYNC_OUT	t _{LBKSKEW}	—	150	ps	7, 8
Input setup to local bus clock (except LGTA/LUPWAIT)	t _{LBIVKH1}	1.8	—	ns	3, 4
LGTA/LUPWAIT input setup to local bus clock	t _{LBIVKH2}	1.7	—	ns	3, 4
Input hold from local bus clock (except LGTA/LUPWAIT)	t _{LBIXKH1}	1.0	—	ns	3, 4
LGTA/LUPWAIT input hold from local bus clock	t _{LBIXKH2}	1.0	—	ns	3, 4
LALE output transition to LAD/LDP output transition (LATCH hold time)	t _{LBOTOT}	1.5	—	ns	6
Local bus clock to output valid (except LAD/LDP and LALE)	t _{LBKHOV1}	—	2.0	ns	—
Local bus clock to data valid for LAD/LDP	t _{LBKHOV2}	—	2.2	ns	3
Local bus clock to address valid for LAD	t _{LBKHOV3}	—	2.3	ns	3
Local bus clock to LALE assertion	t _{LBKHOV4}	—	2.3	ns	3
Output hold from local bus clock (except LAD/LDP and LALE)	t _{LBKHOX1}	0.7	—	ns	3
Output hold from local bus clock for LAD/LDP	t _{LBKHOX2}	0.7	—	ns	3
Local bus clock to output high Impedance (except LAD/LDP and LALE)	t _{LBKHOZ1}	_	2.5	ns	5
Local bus clock to output high impedance for LAD/LDP	t _{LBKHOZ2}		2.5	ns	5

Table 40. Local Bus Timing Parameters (BV_{DD} = 3.3 V)—PLL Enabled

Notes:

- The symbols used for timing specifications follow the pattern of t_{(first two letters of functional block)(signal)(state)(reference)(state)} for inputs and t_{(first two letters of functional block)(reference)(state)(signal)(state)} for outputs. For example, t_{LBIXKH1} symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKH0X} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
- 2. All timings are in reference to LSYNC_IN for PLL enabled and internal local bus clock for PLL bypass mode.
- 3. All signals are measured from $BV_{DD}/2$ of the rising edge of LSYNC_IN for PLL enabled or internal local bus clock for PLL bypass mode to $0.4 \times BV_{DD}$ of the signal in question for 3.3-V signaling levels.
- 4. Input timings are measured at the pin.
- 5. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
- 6. t_{LBOTOT} is a measurement of the minimum time between the negation of LALE and any change in LAD. t_{LBOTOT} is programmed with the LBCR[AHD] parameter.
- 7. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[n]. Skew measured between complementary signals at BV_{DD}/2.
- 8. Guaranteed by design.

NOTE

PLL bypass mode is required when LBIU frequency is at or below 83 MHz. When LBIU operates above 83 MHz, LBIU PLL is recommended to be enabled.

Figure 23 through Figure 28 show the local bus signals.



This table describes the timing parameters of the local bus interface at $BV_{DD} = 3.3$ V with PLL disabled.

Table 42. Local Bus Timing	Parameters—PLL Bypassed
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Parameter	Symbol ¹	Min	Max	Unit	Notes
Local bus cycle time	t _{LBK}	12	—	ns	2
Local bus duty cycle	t _{LBKH/} t _{LBK}	43	57	%	—
Internal launch/capture clock to LCLK delay	t _{lbkhkt}	2.3	4.4	ns	8
Input setup to local bus clock (except LGTA/LUPWAIT)	t _{LBIVKH1}	6.2	—	ns	4, 5
LGTA/LUPWAIT input setup to local bus clock	t _{LBIVKL2}	6.1	—	ns	4, 5
Input hold from local bus clock (except LGTA/LUPWAIT)	t _{LBIXKH1}	-1.8	—	ns	4, 5





Figure 25. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 4 (PLL Enabled)

Table 53. PCI-X AC Timing Specifications at 66 MHz (continued)

Parameter	Symbol	Min	Max	Unit	Notes
HRESET to PCI-X initialization pattern hold time	t _{PCRHIX}	0	50	ns	6, 11

Notes:

- 1. See the timing measurement conditions in the PCI-X 1.0a Specification.
- 2. Minimum times are measured at the package pin (not the test point). Maximum times are measured with the test point and load circuit.
- 3. Setup time for point-to-point signals applies to REQ and GNT only. All other signals are bused.
- 4. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
- 5. Setup time applies only when the device is not driving the pin. Devices cannot drive and receive signals at the same time.
- 6. Maximum value is also limited by delay to the first transaction (time for HRESET high to first configuration access, t_{PCRHFV}). The PCI-X initialization pattern control signals after the rising edge of HRESET must be negated no later than two clocks before the first FRAME and must be floated no later than one clock before FRAME is asserted.
- 7. A PCI-X device is permitted to have the minimum values shown for t_{PCKHOV} and t_{CYC} only in PCI-X mode. In conventional mode, the device must meet the requirements specified in PCI 2.2 for the appropriate clock frequency.
- 8. Device must meet this specification independent of how many outputs switch simultaneously.

9. The timing parameter t_{PCRHFV} is a minimum of 10 clocks rather than the minimum of 5 clocks in the PCI-X 1.0a Specification.

10.Guaranteed by characterization.

11.Guaranteed by design.

This table provides the PCI-X AC timing specifications at 133 MHz. Note that the maximum PCI-X frequency in synchronous mode is 110 MHz.

Parameter	Symbol	Min	Max	Unit	Notes
SYSCLK to signal valid delay	^t PCKHOV		3.8	ns	1, 2, 3, 7, 8
Output hold from SYSCLK	t _{PCKHOX}	0.7	_	ns	1, 11
SYSCLK to output high impedance	t _{PCKHOZ}		7	ns	1, 4, 8, 12
Input setup time to SYSCLK	t _{PCIVKH}	1.2	_	ns	3, 5, 9, 11
Input hold time from SYSCLK	t _{PCIXKH}	0.5	_	ns	11
REQ64 to HRESET setup time	t _{PCRVRH}	10	_	clocks	12
HRESET to REQ64 hold time	t _{PCRHRX}	0	50	ns	12
HRESET high to first FRAME assertion	t _{PCRHFV}	10	_	clocks	10, 12
PCI-X initialization pattern to HRESET setup time	^t PCIVRH	10	_	clocks	12

Table 54. PCI-X AC Timing Specifications at 133 MHz

16 High-Speed Serial Interfaces (HSSI)

The device features one Serializer/Deserializer (SerDes) interface to be used for high-speed serial interconnect applications. The SerDes interface can be used for PCI Express and/or serial RapidIO data transfers.

This section describes the common portion of SerDes DC electrical specifications, which is the DC requirement for SerDes reference clocks. The SerDes data lane's transmitter and receiver reference circuits are also shown.

16.1 Signal Terms Definition

The SerDes utilizes differential signaling to transfer data across the serial link. This section defines terms used in the description and specification of differential signals.

Figure 38 shows how the signals are defined. For illustration purpose, only one SerDes lane is used for the description. The figure shows a waveform for either a transmitter output (SD_TX and \overline{SD}_TX) or a receiver input (SD_RX and \overline{SD}_RX). Each signal swings between A volts and B volts where A > B.

Using this waveform, the definitions are as follows. To simplify the illustration, the following definitions assume that the SerDes transmitter and receiver operate in a fully symmetrical differential signaling environment.

• Single-ended swing

The transmitter output signals and the receiver input signals SD_TX, \overline{SD}_TX , \overline{SD}_RX and \overline{SD}_RX each have a peak-to-peak swing of A – B volts. This is also referred as each signal wire's single-ended swing.

- Differential output voltage, V_{OD} (or differential output swing): The differential output voltage (or swing) of the transmitter, V_{OD} , is defined as the difference of the two complimentary output voltages: $V_{SD_TX} - V_{\overline{SD_TX}}$. The V_{OD} value can be either positive or negative.
- Differential input voltage, V_{ID} (or differential input swing): The differential input voltage (or swing) of the receiver, V_{ID}, is defined as the difference of the two complimentary input voltages: V_{SD_RX} – V_{SD_RX}. The V_{ID} value can be either positive or negative.
- Differential peak voltage, V_{DIFFp} The peak value of the differential transmitter output signal or the differential receiver input signal is defined as differential peak voltage, $V_{DIFFp} = |A - B|$ volts.
- Differential peak-to-peak, $V_{DIFFp-p}$ Because the differential output signal of the transmitter and the differential input signal of the receiver each range from A – B to –(A – B) volts, the peak-to-peak value of the differential transmitter output signal or the differential receiver input signal is defined as differential peak-to-peak voltage, $V_{DIFFp-p} = 2 \times V_{DIFFp} = 2 \times |(A - B)|$ volts, which is twice of differential swing in amplitude, or twice of the differential peak. For example, the output differential peak-to-peak voltage can also be calculated as $V_{TX-DIFFp-p} = 2 \times |V_{OD}|$.
- Common mode voltage, V_{cm} The common mode voltage is equal to one half of the sum of the voltages between each conductor

PCI Express

Symbol	Parameter	Min	Nom	Max	Unit	Comments
V _{RX-CM-ACp}	AC peak common mode input voltage	_	_	150	mV	$\begin{split} & V_{RX\text{-}CM\text{-}ACp} = V_{RXD\text{+}} - V_{RXD\text{-}} /2 + V_{RX\text{-}CM\text{-}DC} \\ & V_{RX\text{-}CM\text{-}DC} = DC_{(avg)} \text{ of } V_{RX\text{-}D\text{+}} + V_{RX\text{-}D\text{-}} \div 2. \\ & See Note 2. \end{split}$
RL _{RX-DIFF}	Differential return loss	15	—	_	dB	Measured over 50 MHz to 1.25 GHz with the D+ and D– lines biased at +300 mV and –300 mV, respectively. See Note 4.
RL _{RX-CM}	Common mode return loss	6	—	—	dB	Measured over 50 MHz to 1.25 GHz with the D+ and D– lines biased at 0 V. See Note 4.
Z _{RX-DIFF-DC}	DC differential input impedance	80	100	120	Ω	RX DC differential mode impedance. See Note 5.
Z _{RX-DC}	DC input impedance	40	50	60	Ω	Required RX D+ as well as D– DC impedance (50 \pm 20% tolerance). See Notes 2 and 5.
Z _{RX-HIGH-IMP-DC}	Powered down DC input impedance	200 k	—	_	Ω	Required RX D+ as well as D– DC impedance when the receiver terminations do not have power. See Note 6.
V _{RX-IDLE-DET-DIFFp-p}	Electrical idle detect threshold	65	—	175	mV	$V_{RX-IDLE-DET-DIFF_{p-p}} = 2 \times V_{RX-D+} - V_{RX-D-} .$ Measured at the package pins of the receiver
T _{RX-IDLE-DET-DIFF-} ENTERTIME	Unexpected electrical idle enter detect threshold integration time			10	ms	An unexpected electrical idle ($V_{RX-DIFFp-p} < V_{RX-IDLE-DET-DIFFp-p}$) must be recognized no longer than $T_{RX-IDLE-DET-DIFF-ENTERING}$ to signal an unexpected idle condition.

Table 57. Differential Receiver (RX) Input Specifications (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes		
I ² C interface						
IIC1_SCL	AG22	I/O	OV _{DD}	4, 27		
IIC1_SDA	AG21	I/O	OV _{DD}	4, 27		
IIC2_SCL	AG15	I/O	OV _{DD}	4, 27		
IIC2_SDA	AG14	I/O	OV _{DD}	4, 27		
	SerDes			•		
SD_RX[0:7]	M28, N26, P28, R26, W26, Y28, AA26, AB28	I	XV _{DD}			
SD_RX[0:7]	M27, N25, P27, R25, W25, Y27, AA25, AB27	I	XV _{DD}	—		
SD_TX[0:7]	M22, N20, P22, R20, U20, V22, W20, Y22	0	XV _{DD}	—		
SD_TX[0:7]	M23, N21, P23, R21, U21, V23, W21, Y23	0	XV _{DD}	—		
SD_PLL_TPD	U28	0	XV _{DD}	24		
SD_REF_CLK	T28	I	XV _{DD}	3		
SD_REF_CLK	T27	I	XV _{DD}	3		
Reserved	AC1, AC3	—	—	2		
Reserved	M26, V28	_	_	32		
Reserved	M25, V27	—	_	34		
Reserved	M20, M21, T22, T23	—	—	38		
	General-Purpose Output					
GPOUT[24:31]	K26, K25, H27, G28, H25, J26, K24, K23	0	BV _{DD}	—		
	System Control					
HRESET	AG17	I	OV _{DD}	_		
HRESET_REQ	AG16	0	OV _{DD}	29		
SRESET	AG20	I	OV _{DD}			
CKSTP_IN	AA9	I	OV _{DD}	—		
CKSTP_OUT	AA8	0	OV _{DD}	2, 4		
	Debug					
TRIG_IN	AB2	I	OV _{DD}	—		
TRIG_OUT/READY/QUIESCE	AB1	0	OV _{DD}	6, 9, 19, 29		
MSRCID[0:1]	AE4, AG2	0	OV _{DD}	5, 6, 9		
MSRCID[2:4]	AF3, AF1, AF2	0	OV _{DD}	6, 19, 29		
MDVAL	AE5	0	OV _{DD}	6		
CLK_OUT	AE21	0	OV _{DD}	11		

Signal	Package Pin Number	Pin Type	Power Supply	Notes
SENSEVSS	M16	—	—	13
	Analog Signals			
MVREF	A18	I Reference voltage signal for DDR	MVREF	
SD_IMP_CAL_RX	L28	I	200Ω to GND	_
SD_IMP_CAL_TX	AB26	I	100Ω to GND	
SD_PLL_TPA	U26	0	—	24

Table 71. MPC8548E Pinout Listing (continued)

Notes:

1. All multiplexed signals are listed only once and do not re-occur. For example, LCS5/DMA_REQ2 is listed only once in the local bus controller section, and is not mentioned in the DMA section even though the pin also functions as DMA_REQ2.

- 2. Recommend a weak pull-up resistor (2-10 kΩ) be placed on this pin to OV_{DD}.
- 3. A valid clock must be provided at POR if TSEC4_TXD[2] is set = 1.
- 4. This pin is an open drain signal.
- 5. This pin is a reset configuration pin. It has a weak internal pull-up P-FET which is enabled only when the processor is in the reset state. This pull-up is designed such that it can be overpowered by an external 4.7-kΩ pull-down resistor. However, if the signal is intended to be high after reset, and if there is any device on the net which might pull down the value of the net at reset, then a pullup or active driver is needed.
- 6. Treat these pins as no connects (NC) unless using debug address functionality.
- The value of LA[28:31] during reset sets the CCB clock to SYSCLK PLL ratio. These pins require 4.7-kΩ pull-up or pull-down resistors. See Section 20.2, "CCB/SYSCLK PLL Ratio."
- 8. The value of LALE, LGPL2, and LBCTL at reset set the e500 core clock to CCB clock PLL ratio. These pins require 4.7-kΩ pull-up or pull-down resistors. See the Section 20.3, "e500 Core PLL Ratio."
- 9. Functionally, this pin is an output, but structurally it is an I/O because it either samples configuration input during reset or because it has other manufacturing test functions. This pin therefore is described as an I/O for boundary scan.
- 10. This pin functionally requires a pull-up resistor, but during reset it is a configuration input that controls 32- vs. 64-bit PCI operation. Therefore, it must be actively driven low during reset by reset logic if the device is to be configured to be a 64-bit PCI device. See the *PCI Specification*.
- 11. This output is actively driven during reset rather than being three-stated during reset.
- 12. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
- 13. These pins are connected to the V_{DD}/GND planes internally and may be used by the core power supply to improve tracking and regulation.
- 14.Internal thermally sensitive resistor.
- 15.No connections must be made to these pins if they are not used.
- 16. These pins are not connected for any use.
- 17.PCI specifications recommend that a weak pull-up resistor (2–10 kΩ) be placed on the higher order pins to OV_{DD} when using 64-bit buffer mode (pins PCI_AD[63:32] and PCI1_C_BE[7:4]).
- 19.If this pin is connected to a device that pulls down during reset, an external pull-up is required to drive this pin to a safe state during reset.
- 20. This pin is only an output in FIFO mode when used as Rx flow control.

24.Do not connect.

Table 72	. MPC8547E	Pinout Listing	(continued)
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Signal	Package Pin Number	Pin Type	Power Supply	Notes
Reserved	AE26	_		2
cfg_pci1_clk	AG24	I	OV _{DD}	5
Reserved	AF25	_		101
Reserved	AE25	_	_	2
Reserved	AG25	_	_	2
Reserved	AD24	_	_	2
Reserved	AF24	_		2
Reserved	AD27	_		2
Reserved	AD28, AE27, W17, AF26	_		2
Reserved	AH25	_		2
	DDR SDRAM Memory Interface			
MDQ[0:63]	L18, J18, K14, L13, L19, M18, L15, L14, A17, B17, A13, B12, C18, B18, B13, A12, H18, F18, J14, F15, K19, J19, H16, K15, D17, G16, K13, D14, D18, F17, F14, E14, A7, A6, D5, A4, C8, D7, B5, B4, A2, B1, D1, E4, A3, B2, D2, E3, F3, G4, J5, K5, F6, G5, J6, K4, J1, K2, M5, M3, J3, J2, L1, M6	I/O	GV _{DD}	_
MECC[0:7]	H13, F13, F11, C11, J13, G13, D12, M12	I/O	GV _{DD}	—
MDM[0:8]	M17, C16, K17, E16, B6, C4, H4, K1, E13	0	GV _{DD}	—
MDQS[0:8]	M15, A16, G17, G14, A5, D3, H1, L2, C13	I/O	GV _{DD}	—
MDQS[0:8]	L17, B16, J16, H14, C6, C2, H3, L4, D13	I/O	GV _{DD}	—
MA[0:15]	A8, F9, D9, B9, A9, L10, M10, H10, K10, G10, B8, E10, B10, G6, A10, L11	0	GV _{DD}	_
MBA[0:2]	F7, J7, M11	0	GV _{DD}	—
MWE	E7	0	GV _{DD}	—
MCAS	H7	0	GV _{DD}	—
MRAS	L8	0	GV _{DD}	—
MCKE[0:3]	F10, C10, J11, H11	0	GV _{DD}	11
MCS[0:3]	K8, J8, G8, F8	0	GV _{DD}	—
MCK[0:5]	H9, B15, G2, M9, A14, F1	0	GV _{DD}	—
MCK[0:5]	J9, A15, G1, L9, B14, F2	0	GV _{DD}	_
MODT[0:3]	E6, K6, L7, M7	0	GV _{DD}	_
MDIC[0:1]	A19, B19	I/O	GV _{DD}	36

Signal	Package Pin Number	Pin Type	Power Supply	Notes
TSEC2_TX_ER	R10	0	LV _{DD}	5, 9, 33
Three	e-Speed Ethernet Controller (Gigabit Ethe	ernet 3)		
TSEC3_TXD[3:0]	V8, W10, Y10, W7	0	TV _{DD}	5, 9, 29
TSEC3_RXD[3:0]	Y1, W3, W5, W4	I	TV _{DD}	_
TSEC3_GTX_CLK	W8	0	TV _{DD}	_
TSEC3_RX_CLK	W2	I	TV _{DD}	_
TSEC3_RX_DV	W1	I	TV _{DD}	_
TSEC3_RX_ER	Y2	I	TV _{DD}	_
TSEC3_TX_CLK	V10	I	TV _{DD}	_
TSEC3_TX_EN	V9	0	TV _{DD}	30
Three	-Speed Ethernet Controller (Gigabit Ethe	ernet 4)		
TSEC4_TXD[3:0]/TSEC3_TXD[7:4]	AB8, Y7, AA7, Y8	0	TV _{DD}	1, 5, 9, 29
TSEC4_RXD[3:0]/TSEC3_RXD[7:4]	AA1, Y3, AA2, AA4	I	TV _{DD}	1
TSEC4_GTX_CLK	AA5	0	TV _{DD}	
TSEC4_RX_CLK/TSEC3_COL	Y5	I	TV _{DD}	1
TSEC4_RX_DV/TSEC3_CRS	AA3	I/O	TV _{DD}	1, 31
TSEC4_TX_EN/TSEC3_TX_ER	AB6	0	TV _{DD}	1, 30
· · ·	DUART			
UART_CTS[0:1]	AB3, AC5	I	OV _{DD}	—
UART_RTS[0:1]	AC6, AD7	0	OV _{DD}	_
UART_SIN[0:1]	AB5, AC7	I	OV _{DD}	_
UART_SOUT[0:1]	AB7, AD8	0	OV _{DD}	_
· · ·	I ² C Interface			
IIC1_SCL	AG22	I/O	OV _{DD}	4, 27
IIC1_SDA	AG21	I/O	OV _{DD}	4, 27
IIC2_SCL	AG15	I/O	OV _{DD}	4, 27
IIC2_SDA	AG14	I/O	OV _{DD}	4, 27
· · · · · ·	SerDes	·		
SD_RX[0:3]	M28, N26, P28, R26	I	XV _{DD}	_
SD_RX[0:3]	M27, N25, P27, R25	I	XV _{DD}	—
SD_TX[0:3]	M22, N20, P22, R20	0	XV _{DD}	_
SD_TX[0:3]	M23, N21, P23, R21	0	XV _{DD}	—
Reserved	W26, Y28, AA26, AB28	—	_	40
Reserved	W25, Y27, AA25, AB27	—	-	40

Table 72. MPC8547E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes		
MDIC[0:1]	A19, B19	I/O	GV _{DD}	36		
	Local Bus Controller Interface					
LAD[0:31]	E27, B20, H19, F25, A20, C19, E28, J23, A25, K22, B28, D27, D19, J22, K20, D28, D25, B25, E22, F22, F21, C25, C22, B23, F20, A23, A22, E19, A21, D21, F19, B21	I/O	BV _{DD}			
LDP[0:3]	K21, C28, B26, B22	I/O	BV _{DD}	_		
LA[27]	H21	0	BV _{DD}	5, 9		
LA[28:31]	H20, A27, D26, A28	0	BV _{DD}	5, 7, 9		
LCS[0:4]	J25, C20, J24, G26, A26	0	BV _{DD}	—		
LCS5/DMA_DREQ2	D23	I/O	BV _{DD}	1		
LCS6/DMA_DACK2	G20	0	BV _{DD}	1		
LCS7/DMA_DDONE2	E21	0	BV _{DD}	1		
LWE0/LBS0/LSDDQM[0]	G25	0	BV _{DD}	5, 9		
LWE1/LBS1/LSDDQM[1]	C23	0	BV _{DD}	5, 9		
LWE2/LBS2/LSDDQM[2]	J21	0	BV _{DD}	5, 9		
LWE3/LBS3/LSDDQM[3]	A24	0	BV _{DD}	5, 9		
LALE	H24	0	BV _{DD}	5, 8, 9		
LBCTL	G27	0	BV _{DD}	5, 8, 9		
LGPL0/LSDA10	F23	0	BV _{DD}	5, 9		
LGPL1/LSDWE	G22	0	BV _{DD}	5, 9		
LGPL2/LOE/LSDRAS	B27	0	BV _{DD}	5, 8, 9		
LGPL3/LSDCAS	F24	0	BV _{DD}	5, 9		
LGPL4/LGTA/LUPWAIT/LPBSE	H23	I/O	BV _{DD}	—		
LGPL5	E26	0	BV _{DD}	5, 9		
LCKE	E24	0	BV _{DD}	—		
LCLK[0:2]	E23, D24, H22	0	BV _{DD}	—		
LSYNC_IN	F27	I	BV _{DD}	—		
LSYNC_OUT	F28	0	BV _{DD}	—		
	DMA		I			
DMA_DACK[0:1]	AD3, AE1	0	OV _{DD}	5, 9, 106		
DMA_DREQ[0:1]	AD4, AE2	I	OV _{DD}	-		
DMA_DDONE[0:1]	AD2, AD1	0	OV _{DD}	-		
Programmable Interrupt Controller						

Signal	Package Pin Number	Pin Type	Power Supply	Notes
UDE	AH16	I	OV _{DD}	—
MCP	AG19	I	OV _{DD}	—
IRQ[0:7]	AG23, AF18, AE18, AF20, AG18, AF17, AH24, AE20	I	OV _{DD}	_
IRQ[8]	AF19	I	OV _{DD}	—
IRQ[9]/DMA_DREQ3	AF21	I	OV _{DD}	1
IRQ[10]/DMA_DACK3	AE19	I/O	OV _{DD}	1
IRQ[11]/DMA_DDONE3	AD20	I/O	OV _{DD}	1
IRQ_OUT	AD18	0	OV _{DD}	2, 4
	Ethernet Management Interface			
EC_MDC	AB9	0	OV _{DD}	5, 9
EC_MDIO	AC8	I/O	OV _{DD}	—
	Gigabit Reference Clock		•	•
EC_GTX_CLK125	V11	I	LV _{DD}	—
Tł	ree-Speed Ethernet Controller (Gigabit Ethern	et 1)	•	
TSEC1_RXD[7:0]	R5, U1, R3, U2, V3, V1, T3, T2	I	LV _{DD}	—
TSEC1_TXD[7:0]	T10, V7, U10, U5, U4, V6, T5, T8	0	LV _{DD}	5, 9
TSEC1_COL	R4	I	LV _{DD}	—
TSEC1_CRS	V5	I/O	LV _{DD}	20
TSEC1_GTX_CLK	U7	0	LV _{DD}	—
TSEC1_RX_CLK	U3	I	LV _{DD}	—
TSEC1_RX_DV	V2	I	LV _{DD}	—
TSEC1_RX_ER	T1	I	LV _{DD}	—
TSEC1_TX_CLK	Т6	I	LV _{DD}	—
TSEC1_TX_EN	U9	0	LV _{DD}	30
TSEC1_TX_ER	Τ7	0	LV _{DD}	—
GPIN[0:7]	P2, R2, N1, N2, P3, M2, M1, N3	I	LV _{DD}	103
GPOUT[0:5]	N9, N10, P8, N7, R9, N5	0	LV _{DD}	—
cfg_dram_type0/GPOUT6	R8	0	LV _{DD}	5, 9
GPOUT7	N6	0	LV _{DD}	-
Reserved	P1	_	—	104
Reserved	R6	_	—	104
Reserved	P6	_	-	15
Reserved	N4			105

Signal	Package Pin Number	Pin Type	Power Supply	Notes			
LSYNC_IN	F27	I	BV _{DD}	—			
LSYNC_OUT	F28	0	BV _{DD}	—			
DMA							
DMA_DACK[0:1]	AD3, AE1	0	OV _{DD}	5, 9, 108			
DMA_DREQ[0:1]	AD4, AE2	I	OV _{DD}	—			
DMA_DDONE[0:1]	AD2, AD1	0	OV _{DD}	—			
	Programmable Interrupt Controller		I				
UDE	AH16	I	OV _{DD}	_			
MCP	AG19	I	OV _{DD}	—			
IRQ[0:7]	AG23, AF18, AE18, AF20, AG18, AF17, AH24, AE20	Ι	OV _{DD}	—			
IRQ[8]	AF19	I	OV _{DD}	—			
IRQ[9]/DMA_DREQ3	AF21	I	OV _{DD}	1			
IRQ[10]/DMA_DACK3	AE19	I/O	OV _{DD}	1			
IRQ[11]/DMA_DDONE3	AD20	I/O	OV _{DD}	1			
IRQ_OUT	AD18	0	OV _{DD}	2, 4			
	Ethernet Management Interface						
EC_MDC	AB9	0	OV _{DD}	5, 9			
EC_MDIO	AC8	I/O	OV _{DD}	—			
Gigabit Reference Clock							
EC_GTX_CLK125	V11	I	LV _{DD}	—			
	Three-Speed Ethernet Controller (Gigabit Ether	rnet 1)					
TSEC1_RXD[7:0]	R5, U1, R3, U2, V3, V1, T3, T2	I	LV _{DD}	—			
TSEC1_TXD[7:0]	T10, V7, U10, U5, U4, V6, T5, T8	0	LV _{DD}	5, 9			
TSEC1_COL	R4	I	LV _{DD}	—			
TSEC1_CRS	V5	I/O	LV _{DD}	20			
TSEC1_GTX_CLK	U7	0	LV _{DD}	—			
TSEC1_RX_CLK	U3	I	LV _{DD}	—			
TSEC1_RX_DV	V2	I	LV _{DD}	—			
TSEC1_RX_ER	T1	I	LV _{DD}	_			
TSEC1_TX_CLK	Т6	I	LV _{DD}	—			
TSEC1_TX_EN	U9	0	LV _{DD}	30			
TSEC1_TX_ER	Τ7	0	LV _{DD}	—			
GPIN[0:7]	P2, R2, N1, N2, P3, M2, M1, N3	I	LV _{DD}	103			

Signal	Package Pin Number	Pin Type	Power Supply	Notes				
SENSEVSS	M16	—	—	13				
Analog Signals								
MVREF	A18	l Reference voltage signal for DDR	MVREF					
SD_IMP_CAL_RX	L28	I	200 Ω (±1%) to GND	_				
SD_IMP_CAL_TX	AB26	I	100 Ω (±1%) to GND	—				
SD_PLL_TPA	U26	0	AVDD_SRDS	24				

Table 74. MPC8543E Pinout Listing (continued)

Note: All note references in this table use the same numbers as those for Table 71. See Table 71 for the meanings of these notes.

System Design Information

level must always be equivalent to V_{DD} , and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide independent filter circuits per PLL power supply as illustrated in Figure 57, one to each of the AV_{DD} pins. By providing independent filters to each PLL the opportunity to cause noise injection from one PLL to the other is reduced.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It must be built with surface mount capacitors with minimum Effective Series Inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit must be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It must be routed directly from the capacitors to the AV_{DD} pin, which is on the periphery of the footprint, without the inductance of vias.

Figure 57 through Figure 59 shows the PLL power supply filter circuits.



Figure 57. PLL Power Supply Filter Circuit with PLAT Pins



Figure 58. PLL Power Supply Filter Circuit with CORE Pins



Figure 59. PLL Power Supply Filter Circuit with PCI/LBIU Pins

The AV_{DD}_SRDS signal provides power for the analog portions of the SerDes PLL. To ensure stability of the internal clock, the power supplied to the PLL is filtered using a circuit similar to the one shown in following figure. For maximum effectiveness, the filter circuit is placed as closely as possible to the AV_{DD}_SRDS ball to ensure it filters out as much noise as possible. The ground connection must be near the AV_{DD}_SRDS ball. The 0.003- μ F capacitor is closest to the ball, followed by the two 2.2 μ F capacitors, and finally the 1 Ω resistor to the board supply plane. The capacitors are connected from AV_{DD}_SRDS to

System Design Information



Figure 61. Driver Impedance Measurement

This table summarizes the signal impedance targets. The driver impedances are targeted at minimum V_{DD} , nominal OV_{DD} , 105°C.

Table 86. Impedance Characteristics

Impedance	Local Bus, Ethernet, DUART, Control, Configuration, Power Management	PCI	DDR DRAM	Symbol	Unit
R _N	43 Target	25 Target	20 Target	Z ₀	W
R _P	43 Target	25 Target	20 Target	Z ₀	W

Note: Nominal supply voltages. See Table 1, $T_i = 105^{\circ}C$.

22.8 Configuration Pin Muxing

The device provides the user with power-on configuration options which can be set through the use of external pull-up or pull-down resistors of $4.7 \text{ k}\Omega$ on certain output pins (see customer visible configuration pins). These pins are generally used as output only pins in normal operation.

While $\overline{\text{HRESET}}$ is asserted however, these pins are treated as inputs. The value presented on these pins while $\overline{\text{HRESET}}$ is asserted, is latched when $\overline{\text{HRESET}}$ deasserts, at which time the input receiver is disabled and the I/O circuit takes on its normal function. Most of these sampled configuration pins are equipped with an on-chip gated resistor of approximately 20 k Ω . This value must permit the 4.7-k Ω resistor to pull the configuration pin to a valid logic low level. The pull-up resistor is enabled only during $\overline{\text{HRESET}}$ (and for platform/system clocks after $\overline{\text{HRESET}}$ deassertion to ensure capture of the reset value). When the input receiver is disabled the pull-up is also, thus allowing functional operation of the pin as an output with minimal signal quality or delay disruption. The default value for all configuration bits treated this way has been encoded such that a high voltage level puts the device into the default state and external resistors are needed only when non-default settings are required by the user.

Careful board layout with stubless connections to these pull-down resistors coupled with the large value of the pull-down resistor minimizes the disruption of signal quality or speed for output pins thus configured.