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Understanding [Embedded - Microprocessors](#)

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	1.0GHz
Co-Processors/DSP	Signal Processing; SPE
RAM Controllers	DDR, DDR2, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc8548vtaqgd

- Performance monitor facility that is similar to, but separate from, the device performance monitor

The e500 defines features that are not implemented on this device. It also generally defines some features that this device implements more specifically. An understanding of these differences can be critical to ensure proper operations.

- 512-Kbyte L2 cache/SRAM
 - Flexible configuration.
 - Full ECC support on 64-bit boundary in both cache and SRAM modes
 - Cache mode supports instruction caching, data caching, or both.
 - External masters can force data to be allocated into the cache through programmed memory ranges or special transaction types (stashing).
 - 1, 2, or 4 ways can be configured for stashing only.
 - Eight-way set-associative cache organization (32-byte cache lines)
 - Supports locking entire cache or selected lines. Individual line locks are set and cleared through Book E instructions or by externally mastered transactions.
 - Global locking and Flash clearing done through writes to L2 configuration registers
 - Instruction and data locks can be Flash cleared separately.
 - SRAM features include the following:
 - I/O devices access SRAM regions by marking transactions as snoopable (global).
 - Regions can reside at any aligned location in the memory map.
 - Byte-accessible ECC is protected using read-modify-write transaction accesses for smaller-than-cache-line accesses.
- Address translation and mapping unit (ATMU)
 - Eight local access windows define mapping within local 36-bit address space.
 - Inbound and outbound ATMUs map to larger external address spaces.
 - Three inbound windows plus a configuration window on PCI/PCI-X and PCI Express
 - Four inbound windows plus a default window on RapidIO™
 - Four outbound windows plus default translation for PCI/PCI-X and PCI Express
 - Eight outbound windows plus default translation for RapidIO with segmentation and sub-segmentation support
- DDR/DDR2 memory controller
 - Programmable timing supporting DDR and DDR2 SDRAM
 - 64-bit data interface
 - Four banks of memory supported, each up to 4 Gbytes, to a maximum of 16 Gbytes
 - DRAM chip configurations from 64 Mbits to 4 Gbits with $\times 8/\times 16$ data ports
 - Full ECC support
 - Page mode support
 - Up to 16 simultaneous open pages for DDR

- Single inbound doorbell message structure
- Facility to accept port-write messages
- PCI Express interface
 - PCI Express 1.0a compatible
 - Supports x8,x4,x2, and x1 link widths
 - Auto-detection of number of connected lanes
 - Selectable operation as root complex or endpoint
 - Both 32- and 64-bit addressing
 - 256-byte maximum payload size
 - Virtual channel 0 only
 - Traffic class 0 only
 - Full 64-bit decode with 32-bit wide windows
- Pin multiplexing for the high-speed I/O interfaces supports one of the following configurations:
 - 8 PCI Express
 - 4 PCI Express and 4 serial RapidIO
- Power management
 - Supports power saving modes: doze, nap, and sleep
 - Employs dynamic power management, which automatically minimizes power consumption of blocks when they are idle
- System performance monitor
 - Supports eight 32-bit counters that count the occurrence of selected events
 - Ability to count up to 512 counter-specific events
 - Supports 64 reference events that can be counted on any of the eight counters
 - Supports duration and quantity threshold counting
 - Burstiness feature that permits counting of burst events with a programmable time between bursts
 - Triggering and chaining capability
 - Ability to generate an interrupt on overflow
- System access port
 - Uses JTAG interface and a TAP controller to access entire system memory map
 - Supports 32-bit accesses to configuration registers
 - Supports cache-line burst accesses to main memory
 - Supports large block (4-Kbyte) uploads and downloads
 - Supports continuous bit streaming of entire block for fast upload and download
- JTAG boundary scan, designed to comply with IEEE Std. 1149.1™

4.5 Platform to FIFO Restrictions

Note the following FIFO maximum speed restrictions based on platform speed.

For FIFO GMII mode:

$$\text{FIFO TX/RX clock frequency} \leq \text{platform clock frequency}/4.2$$

For example, if the platform frequency is 533 MHz, the FIFO TX/RX clock frequency must be no more than 127 MHz.

For FIFO encoded mode:

$$\text{FIFO TX/RX clock frequency} \leq \text{platform clock frequency}/4.2$$

For example, if the platform frequency is 533 MHz, the FIFO TX/RX clock frequency must be no more than 167 MHz.

4.6 Platform Frequency Requirements for PCI-Express and Serial RapidIO

The CCB clock frequency must be considered for proper operation of the high-speed PCI-Express and Serial RapidIO interfaces as described below.

For proper PCI Express operation, the CCB clock frequency must be greater than:

$$\frac{527 \text{ MHz} \times (\text{PCI-Express link width})}{8}$$

See *MPC8548ERM, Rev. 2, PowerQUICC III Integrated Processor Family Reference Manual*, Section 18.1.3.2, “Link Width,” for PCI Express interface width details.

For proper serial RapidIO operation, the CCB clock frequency must be greater than:

$$\frac{2 \times (0.80) \times (\text{Serial RapidIO interface frequency}) \times (\text{Serial RapidIO link width})}{64}$$

See *MPC8548ERM, Rev. 2, PowerQUICC III Integrated Processor Family Reference Manual*, Section 17.4, “1x/4x LP-Serial Signal Descriptions,” for serial RapidIO interface width and frequency details.

4.7 Other Input Clocks

For information on the input clocks of other functional blocks of the platform see the specific section of this document.

Table 13 provides the recommended operating conditions for the DDR SDRAM controller when $GV_{DD}(typ) = 2.5\text{ V}$.

Table 13. DDR SDRAM DC Electrical Characteristics for $GV_{DD}(typ) = 2.5\text{ V}$

Parameter/Condition	Symbol	Min	Max	Unit	Notes
I/O supply voltage	GV_{DD}	2.375	2.625	V	1
I/O reference voltage	MV_{REF}	$0.49 \times GV_{DD}$	$0.51 \times GV_{DD}$	V	2
I/O termination voltage	V_{TT}	$MV_{REF} - 0.04$	$MV_{REF} + 0.04$	V	3
Input high voltage	V_{IH}	$MV_{REF} + 0.15$	$GV_{DD} + 0.3$	V	—
Input low voltage	V_{IL}	-0.3	$MV_{REF} - 0.15$	V	—
Output leakage current	I_{OZ}	-50	50	μA	4
Output high current ($V_{OUT} = 1.95\text{ V}$)	I_{OH}	-16.2	—	mA	—
Output low current ($V_{OUT} = 0.35\text{ V}$)	I_{OL}	16.2	—	mA	—

Notes:

- GV_{DD} is expected to be within 50 mV of the DRAM V_{DD} at all times.
- MV_{REF} is expected to be equal to $0.5 \times GV_{DD}$, and to track GV_{DD} DC variations as measured at the receiver. Peak-to-peak noise on MV_{REF} may not exceed $\pm 2\%$ of the DC value.
- V_{TT} is not applied directly to the device. It is the supply to which far end signal termination is made and is expected to be equal to MV_{REF} . This rail must track variations in the DC level of MV_{REF} .
- Output leakage is measured with all outputs disabled, $0\text{ V} \leq V_{OUT} \leq GV_{DD}$.

Table 14 provides the DDR I/O capacitance when $GV_{DD}(typ) = 2.5\text{ V}$.

Table 14. DDR SDRAM Capacitance for $GV_{DD}(typ) = 2.5\text{ V}$

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Input/output capacitance: DQ, DQS	C_{IO}	6	8	pF	1
Delta input/output capacitance: DQ, DQS	C_{DIO}	—	0.5	pF	1

Note:

- This parameter is sampled. $GV_{DD} = 2.5\text{ V} \pm 0.125\text{ V}$, $f = 1\text{ MHz}$, $T_A = 25^\circ\text{C}$, $V_{OUT} = GV_{DD}/2$, V_{OUT} (peak-to-peak) = 0.2 V.

This table provides the current draw characteristics for MV_{REF} .

Table 15. Current Draw Characteristics for MV_{REF}

Parameter/Condition	Symbol	Min	Max	Unit	Notes
Current draw for MV_{REF}	I_{MVREF}	—	500	μA	1

Note:

- The voltage regulator for MV_{REF} must be able to supply up to 500 μA current.

Figure 4 shows the DDR SDRAM output timing diagram.

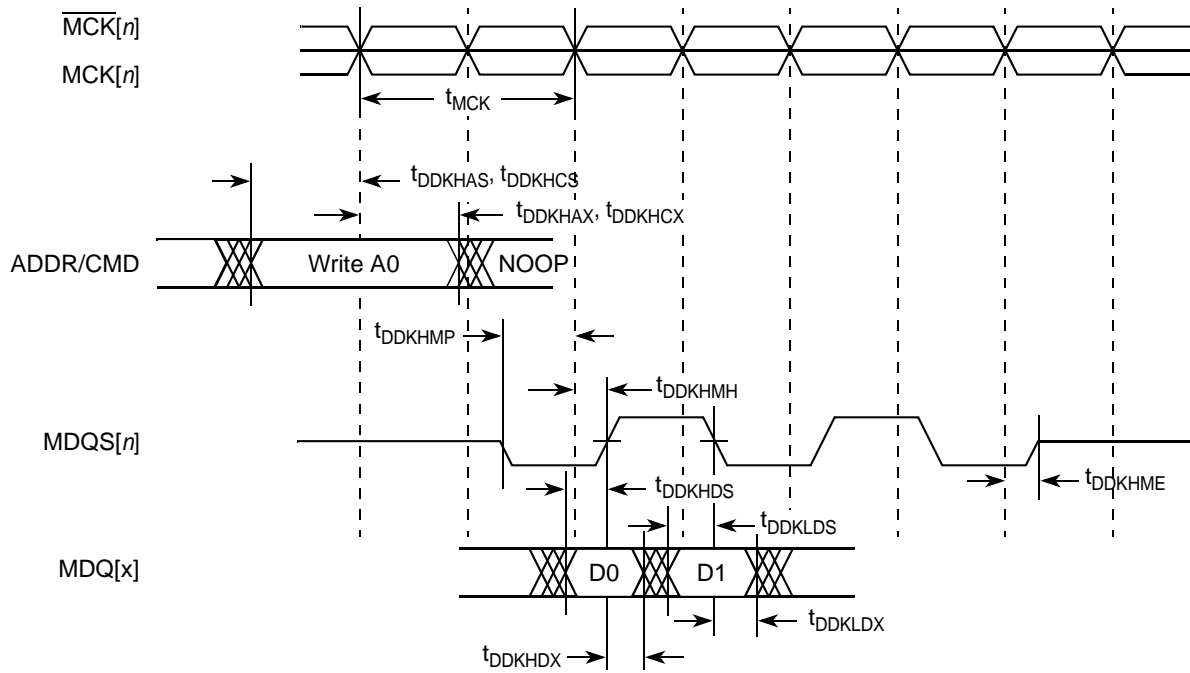


Figure 4. DDR SDRAM Output Timing Diagram

Figure 5 provides the AC test load for the DDR bus.

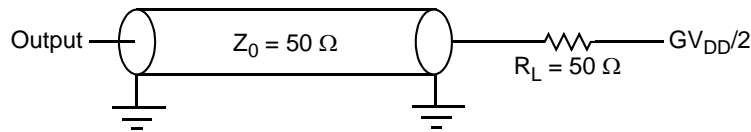


Figure 5. DDR AC Test Load

Figure 10 shows the GMII receive AC timing diagram.

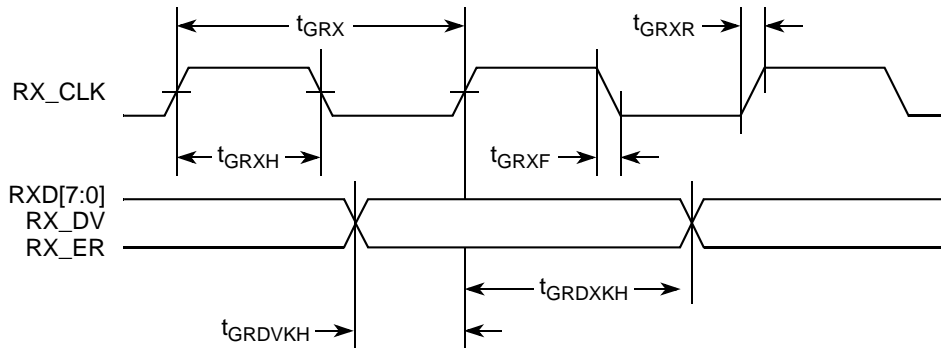


Figure 10. GMII Receive AC Timing Diagram

8.2.3 MII AC Timing Specifications

This section describes the MII transmit and receive AC timing specifications.

8.2.3.1 MII Transmit AC Timing Specifications

This table provides the MII transmit AC timing specifications.

Table 28. MII Transmit AC Timing Specifications

Parameter/Condition	Symbol ¹	Min	Typ	Max	Unit
TX_CLK clock period 10 Mbps	t_{MTX}^2	—	400	—	ns
TX_CLK clock period 100 Mbps	t_{MTX}	—	40	—	ns
TX_CLK duty cycle	t_{MTXH}/t_{MTX}	35	—	65	%
TX_CLK to MII data TXD[3:0], TX_ER, TX_EN delay	t_{MTKHDX}	1	5	15	ns
TX_CLK data clock rise (20%–80%)	t_{MTXR}^2	1.0	—	4.0	ns
TX_CLK data clock fall (80%–20%)	t_{MTXF}^2	1.0	—	4.0	ns

Notes:

- The symbols used for timing specifications follow the pattern of $t_{(first\ two\ letters\ of\ functional\ block)(signal)(state)(reference)(state)}$ for inputs and $t_{(first\ two\ letters\ of\ functional\ block)(reference)(state)(signal)(state)}$ for outputs. For example, t_{MTKHDX} symbolizes MII transmit timing (MT) for the time t_{MTX} clock reference (K) going high (H) until data outputs (D) are invalid (X). Note that, in general, the clock reference symbol representation is based on two to three letters representing the clock of a particular functional. For example, the subscript of t_{MTX} represents the MII(M) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
- Guaranteed by design.

Figure 11 shows the MII transmit AC timing diagram.

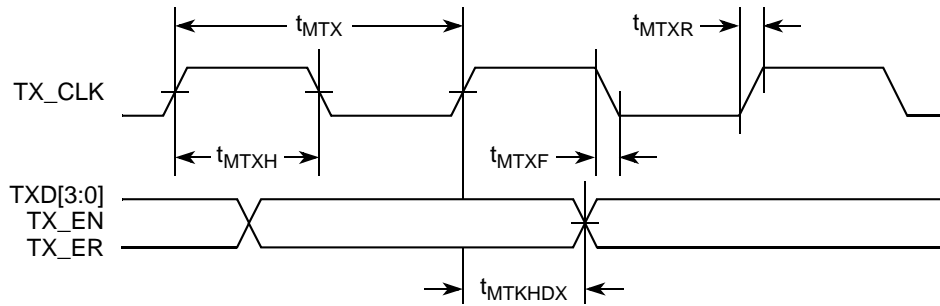


Figure 11. MII Transmit AC Timing Diagram

8.2.3.2 MII Receive AC Timing Specifications

This table provides the MII receive AC timing specifications.

Table 29. MII Receive AC Timing Specifications

Parameter/Condition	Symbol ¹	Min	Typ	Max	Unit
RX_CLK clock period 10 Mbps	t_{MRX}^2	—	400	—	ns
RX_CLK clock period 100 Mbps	t_{MRX}	—	40	—	ns
RX_CLK duty cycle	t_{MRXH}/t_{MRX}	35	—	65	%
RXD[3:0], RX_DV, RX_ER setup time to RX_CLK	t_{MRDVKH}	10.0	—	—	ns
RXD[3:0], RX_DV, RX_ER hold time to RX_CLK	t_{MRDXKH}	10.0	—	—	ns
RX_CLK clock rise (20%–80%)	t_{MRXR}^2	1.0	—	4.0	ns
RX_CLK clock fall time (80%–20%)	t_{MRXF}^2	1.0	—	4.0	ns

Notes:

- The symbols used for timing specifications follow the pattern of $t_{(\text{first two letters of functional block})(\text{signal})(\text{state})(\text{reference})(\text{state})}$ for inputs and $t_{(\text{first two letters of functional block})(\text{reference})(\text{state})(\text{signal})(\text{state})}$ for outputs. For example, t_{MRDVKH} symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MRX} clock reference (K) going to the high (H) state or setup time. Also, t_{MRDXKL} symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{MRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{MRX} represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
- Guaranteed by design.

Figure 12 provides the AC test load for eTSEC.

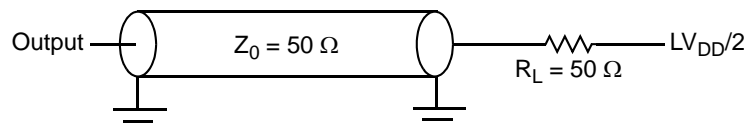


Figure 12. eTSEC AC Test Load

9 Ethernet Management Interface Electrical Characteristics

The electrical characteristics specified here apply to MII management interface signals MDIO (management data input/output) and MDC (management data clock). The electrical characteristics for GMII, RGMII, RMII, TBI, and RTBI are specified in “[Section 8, “Enhanced Three-Speed Ethernet \(eTSEC\).”](#)”

9.1 MII Management DC Electrical Characteristics

The MDC and MDIO are defined to operate at a supply voltage of 3.3 V. The DC electrical characteristics for MDIO and MDC are provided in this table.

Table 36. MII Management DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
Supply voltage (3.3 V)	OV_{DD}	3.13	3.47	V
Output high voltage ($OV_{DD} = \text{Min}$, $I_{OH} = -1.0 \text{ mA}$)	V_{OH}	2.10	$OV_{DD} + 0.3$	V
Output low voltage ($OV_{DD} = \text{Min}$, $I_{OL} = 1.0 \text{ mA}$)	V_{OL}	GND	0.50	V
Input high voltage	V_{IH}	2.0	—	V
Input low voltage	V_{IL}	—	0.90	V
Input high current ($OV_{DD} = \text{Max}$, $V_{IN}^1 = 2.1 \text{ V}$)	I_{IH}	—	40	μA
Input low current ($OV_{DD} = \text{Max}$, $V_{IN} = 0.5 \text{ V}$)	I_{IL}	-600	—	μA

Note:

1. Note that the symbol V_{IN} , in this case, represents the OV_{IN} symbol referenced in [Table 1](#) and [Table 2](#).

9.2 MII Management AC Electrical Specifications

This table provides the MII management AC timing specifications.

Table 37. MII Management AC Timing Specifications

At recommended operating conditions with OV_{DD} is 3.3 V \pm 5%.

Parameter	Symbol ¹	Min	Typ	Max	Unit	Notes
MDC frequency	f_{MDC}	0.72	2.5	8.3	MHz	2, 3, 4
MDC period	t_{MDC}	120.5	—	1389	ns	—
MDC clock pulse width high	t_{MDCH}	32	—	—	ns	—
MDC to MDIO valid	t_{MDKHDV}	$16 \times t_{CCB}$	—	—	ns	5
MDC to MDIO delay	t_{MDKHDX}	$(16 \times t_{CCB} \times 8) - 3$	—	$(16 \times t_{CCB} \times 8) + 3$	ns	5
MDIO to MDC setup time	t_{MDDVKH}	5	—	—	ns	—
MDIO to MDC hold time	t_{MDDXKH}	0	—	—	ns	—
MDC rise time	t_{MDCR}	—	—	10	ns	4

14 GP_{OUT}/GP_{IN}

This section describes the DC and AC electrical specifications for the GP_{OUT}/GP_{IN} bus of the device.

14.1 GP_{OUT}/GP_{IN} Electrical Characteristics

Table 47 and Table 48 provide the DC electrical characteristics for the GP_{OUT} interface.

Table 47. GP_{OUT} DC Electrical Characteristics (3.3 V DC)

Parameter	Symbol	Min	Max	Unit
Supply voltage 3.3 V	BV _{DD}	3.13	3.47	V
High-level output voltage (BV _{DD} = min, I _{OH} = -2 mA)	V _{OH}	BV _{DD} - 0.2	—	V
Low-level output voltage (BV _{DD} = min, I _{OL} = 2 mA)	V _{OL}	—	0.2	V

Table 48. GP_{OUT} DC Electrical Characteristics (2.5 V DC)

Parameter	Symbol	Min	Max	Unit
Supply voltage 2.5 V	BV _{DD}	2.37	2.63	V
High-level output voltage (BV _{DD} = min, I _{OH} = -1 mA)	V _{OH}	2.0	BV _{DD} + 0.3	V
Low-level output voltage (BV _{DD} min, I _{OL} = 1 mA)	V _{OL}	GND - 0.3	0.4	V

Table 49 and Table 50 provide the DC electrical characteristics for the GP_{IN} interface.

Table 49. GP_{IN} DC Electrical Characteristics (3.3 V DC)

Parameter	Symbol	Min	Max	Unit
Supply voltage 3.3 V	BV _{DD}	3.13	3.47	V
High-level input voltage	V _{IH}	2	BV _{DD} + 0.3	V
Low-level input voltage	V _{IL}	-0.3	0.8	V
Input current (BV _{IN} ¹ = 0 V or BV _{IN} = BV _{DD})	I _{IN}	—	±5	μA

Note:

1. The symbol BV_{IN}, in this case, represents the BV_{IN} symbol referenced in Table 1.

Figure 36 shows the PCI/PCI-X input AC timing conditions.

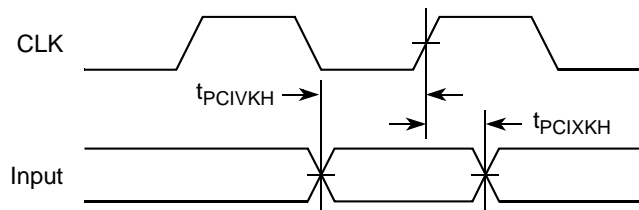


Figure 36. PCI/PCI-X Input AC Timing Measurement Conditions

Figure 37 shows the PCI/PCI-X output AC timing conditions.

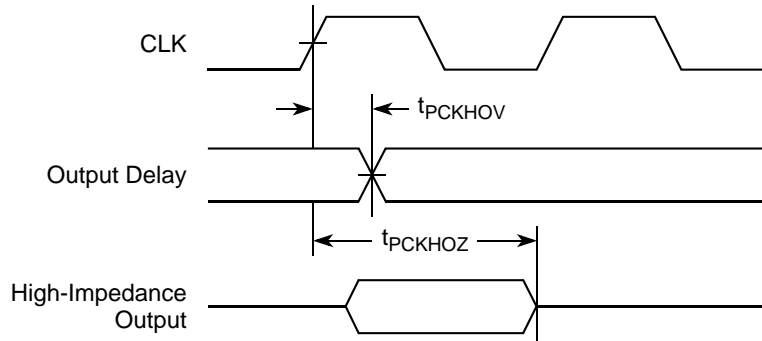


Figure 37. PCI/PCI-X Output AC Timing Measurement Condition

Table 53 provides the PCI-X AC timing specifications at 66 MHz.

Table 53. PCI-X AC Timing Specifications at 66 MHz

Parameter	Symbol	Min	Max	Unit	Notes
SYSCLK to signal valid delay	t_{PCKHOV}	—	3.8	ns	1, 2, 3, 7, 8
Output hold from SYSCLK	t_{PCKHOX}	0.7	—	ns	1, 10
SYSCLK to output high impedance	t_{PCKHOZ}	—	7	ns	1, 4, 8, 11
Input setup time to SYSCLK	t_{PCIVKH}	1.7	—	ns	3, 5
Input hold time from SYSCLK	t_{PCIXKH}	0.5	—	ns	10
$\overline{REQ64}$ to \overline{HRESET} setup time	t_{PCRVRH}	10	—	clocks	11
\overline{HRESET} to $\overline{REQ64}$ hold time	t_{PCRHRX}	0	50	ns	11
\overline{HRESET} high to first \overline{FRAME} assertion	t_{PCRHFV}	10	—	clocks	9, 11
PCI-X initialization pattern to \overline{HRESET} setup time	t_{PCIVRH}	10	—	clocks	11

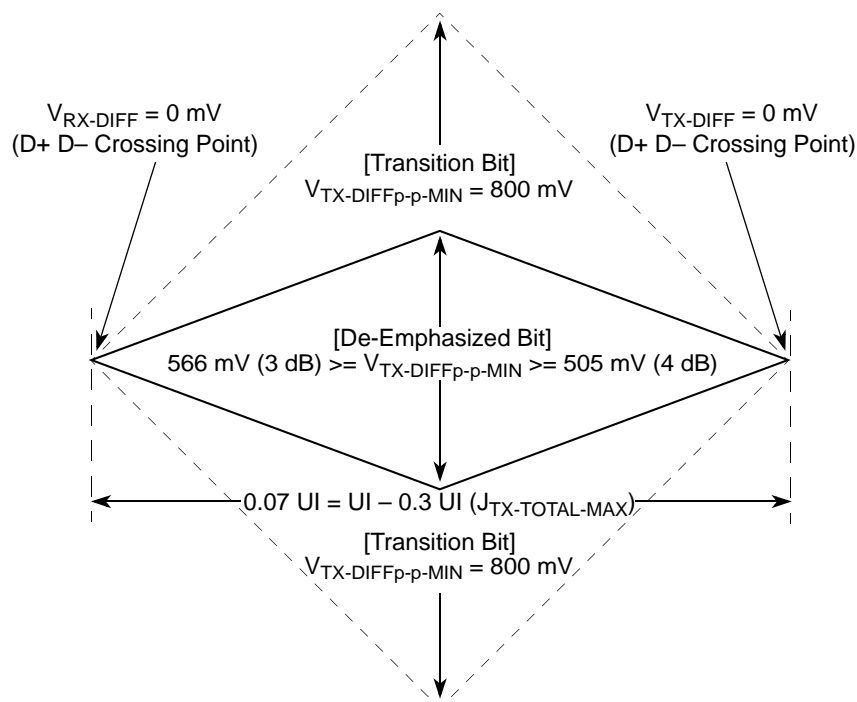


Figure 48. Minimum Transmitter Timing and Voltage Output Compliance Specifications

17.4.3 Differential Receiver (RX) Input Specifications

Table 57 defines the specifications for the differential input at all receivers (RXs). The parameters are specified at the component pins.

Table 57. Differential Receiver (RX) Input Specifications

Symbol	Parameter	Min	Nom	Max	Unit	Comments
UI	Unit interval	399.88	400	400.12	ps	Each UI is 400 ps \pm 300 ppm. UI does not account for spread spectrum clock dictated variations. See Note 1.
$V_{RX-DIFFp-p}$	Differential peak-to-peak input voltage	0.175	—	1.200	V	$V_{RX-DIFFp-p} = 2 \times V_{RX-D+} - V_{RX-D-} $. See Note 2.
T_{RX-EYE}	Minimum receiver eye width	0.4	—	—	UI	The maximum interconnect media and transmitter jitter that can be tolerated by the receiver can be derived as $T_{RX-MAX-JITTER} = 1 - T_{RX-EYE} = 0.6$ UI. See Notes 2 and 3.
$T_{RX-EYE-MEDIAN-to-MAX-JITTER}$	Maximum time between the jitter median and maximum deviation from the median	—	—	0.3	UI	Jitter is defined as the measurement variation of the crossing points ($V_{RX-DIFFp-p} = 0$ V) in relation to a recovered TX UI. A recovered TX UI is calculated over 3500 consecutive unit intervals of sample data. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the TX UI. See Notes 2, 3, and 7.

Table 57. Differential Receiver (RX) Input Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Unit	Comments
$L_{TX-SKEW}$	Total Skew	—	—	20	ns	Skew across all lanes on a Link. This includes variation in the length of SKP ordered set (for example, COM and one to five symbols) at the RX as well as any delay differences arising from the interconnect itself.

Notes:

- No test load is necessarily associated with this value.
- Specified at the measurement point and measured over any 250 consecutive UIs. The test load in [Figure 50](#) must be used as the RX device when taking measurements (also see the receiver compliance eye diagram shown in [Figure 49](#)). If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
- A $T_{RX-EYE} = 0.40$ UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX-EYE-MEDIAN-to-MAX-JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive TX UIs. Note that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
- The receiver input impedance shall result in a differential return loss greater than or equal to 15 dB with the D+ line biased to 300 mV and the D– line biased to –300 mV and a common mode return loss greater than or equal to 6 dB (no bias required) over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements for is 50 Ω to ground for both the D+ and D– line (that is, as measured by a vector network analyzer with 50- Ω probes—see [Figure 50](#)). Note: that the series capacitors CTX is optional for the return loss measurement.
- Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM) there is a 5 ms transition time before receiver termination values must be met on all unconfigured lanes of a port.
- The RX DC common mode Impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the RX ground.
- It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

17.5 Receiver Compliance Eye Diagrams

The RX eye diagram in [Figure 49](#) is specified using the passive compliance/test measurement load (see [Figure 50](#)) in place of any real PCI Express RX component.

Note: In general, the minimum receiver eye diagram measured with the compliance/test measurement load (see [Figure 50](#)) is larger than the minimum receiver eye diagram measured over a range of systems at the input receiver of any real PCI Express component. The degraded eye diagram at the input receiver is due to traces internal to the package as well as silicon parasitic characteristics which cause the real PCI Express component to vary in impedance from the compliance/test measurement load. The input receiver eye diagram is implementation specific and is not specified. RX component designer must provide additional margin to adequately compensate for the degraded minimum receiver eye diagram (shown in [Figure 49](#)) expected at the input receiver based on some adequate combination of system simulations and the return loss measured looking into the RX package and silicon. The RX eye diagram must be aligned in time using the jitter median to locate the center of the eye diagram.

Table 71. MPC8548E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$\overline{\text{PCI1_REQ}}[4:1]$	AH2, AG4, AG3, AH4	I	OV_{DD}	— — — — —
$\overline{\text{PCI1_REQ0}}$	AH3	I/O	OV_{DD}	—
PCI1_CLK	AH26	I	OV_{DD}	39
$\overline{\text{PCI1_DEVSEL}}$	AH11	I/O	OV_{DD}	2
$\overline{\text{PCI1_FRAME}}$	AE11	I/O	OV_{DD}	2
PCI1_IDSEL	AG9	I	OV_{DD}	—
$\overline{\text{PCI1_REQ64/PCI2_FRAME}}$	AF14	I/O	OV_{DD}	2, 5, 10
$\overline{\text{PCI1_ACK64/PCI2_DEVSEL}}$	V15	I/O	OV_{DD}	2
PCI2_CLK	AE28	I	OV_{DD}	39
$\overline{\text{PCI2_IRDY}}$	AD26	I/O	OV_{DD}	2
$\overline{\text{PCI2_PERR}}$	AD25	I/O	OV_{DD}	2
$\overline{\text{PCI2_GNT}}[4:1]$	AE26, AG24, AF25, AE25	O	OV_{DD}	5, 9, 35
$\overline{\text{PCI2_GNT0}}$	AG25	I/O	OV_{DD}	—
$\overline{\text{PCI2_SERR}}$	AD24	I/O	OV_{DD}	2, 4
$\overline{\text{PCI2_STOP}}$	AF24	I/O	OV_{DD}	2
$\overline{\text{PCI2_TRDY}}$	AD27	I/O	OV_{DD}	2
$\overline{\text{PCI2_REQ}}[4:1]$	AD28, AE27, W17, AF26	I	OV_{DD}	—
$\overline{\text{PCI2_REQ0}}$	AH25	I/O	OV_{DD}	—
DDR SDRAM Memory Interface				
MDQ[0:63]	L18, J18, K14, L13, L19, M18, L15, L14, A17, B17, A13, B12, C18, B18, B13, A12, H18, F18, J14, F15, K19, J19, H16, K15, D17, G16, K13, D14, D18, F17, F14, E14, A7, A6, D5, A4, C8, D7, B5, B4, A2, B1, D1, E4, A3, B2, D2, E3, F3, G4, J5, K5, F6, G5, J6, K4, J1, K2, M5, M3, J3, J2, L1, M6	I/O	GV_{DD}	—
MECC[0:7]	H13, F13, F11, C11, J13, G13, D12, M12	I/O	GV_{DD}	—
MDM[0:8]	M17, C16, K17, E16, B6, C4, H4, K1, E13	O	GV_{DD}	—
MDQS[0:8]	M15, A16, G17, G14, A5, D3, H1, L2, C13	I/O	GV_{DD}	—
$\overline{\text{MDQS}}[0:8]$	L17, B16, J16, H14, C6, C2, H3, L4, D13	I/O	GV_{DD}	—
MA[0:15]	A8, F9, D9, B9, A9, L10, M10, H10, K10, G10, B8, E10, B10, G6, A10, L11	O	GV_{DD}	—
MBA[0:2]	F7, J7, M11	O	GV_{DD}	—

Table 71. MPC8548E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
Clock				
RTC	AF16	I	OV _{DD}	—
SYSCLK	AH17	I	OV _{DD}	—
JTAG				
TCK	AG28	I	OV _{DD}	—
TDI	AH28	I	OV _{DD}	12
TDO	AF28	O	OV _{DD}	—
TMS	AH27	I	OV _{DD}	12
$\overline{\text{TRST}}$	AH23	I	OV _{DD}	12
DFT				
L1_TSTCLK	AC25	I	OV _{DD}	25
L2_TSTCLK	AE22	I	OV _{DD}	25
$\overline{\text{LSSD_MODE}}$	AH20	I	OV _{DD}	25
$\overline{\text{TEST_SEL}}$	AH14	I	OV _{DD}	25
Thermal Management				
THERM0	AG1	—	—	14
THERM1	AH1	—	—	14
Power Management				
ASLEEP	AH18	O	OV _{DD}	9, 19, 29
Power and Ground Signals				
GND	A11, B7, B24, C1, C3, C5, C12, C15, C26, D8, D11, D16, D20, D22, E1, E5, E9, E12, E15, E17, F4, F26, G12, G15, G18, G21, G24, H2, H6, H8, H28, J4, J12, J15, J17, J27, K7, K9, K11, K27, L3, L5, L12, L16, N11, N13, N15, N17, N19, P4, P9, P12, P14, P16, P18, R11, R13, R15, R17, R19, T4, T12, T14, T16, T18, U8, U11, U13, U15, U17, U19, V4, V12, V18, W6, W19, Y4, Y9, Y11, Y19, AA6, AA14, AA17, AA22, AA23, AB4, AC2, AC11, AC19, AC26, AD5, AD9, AD22, AE3, AE14, AF6, AF10, AF13, AG8, AG27, K28, L24, L26, N24, N27, P25, R28, T24, T26, U24, V25, W28, Y24, Y26, AA24, AA27, AB25, AC28, L21, L23, N22, P20, R23, T21, U22, V20, W23, Y21, U27	—	—	—
OV _{DD}	V16, W11, W14, Y18, AA13, AA21, AB11, AB17, AB24, AC4, AC9, AC21, AD6, AD13, AD17, AD19, AE10, AE8, AE24, AF4, AF12, AF22, AF27, AG26	Power for PCI and other standards (3.3 V)	OV _{DD}	—

Table 72. MPC8547E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
SD_PLL_TPA	U26	O	—	24

Note: All note references in this table use the same numbers as those for Table 71. See Table 71 for the meanings of these notes.

Table 73 provides the pin-out listing for the MPC8545E 783 FC-PBGA package.

NOTE

All note references in the following table use the same numbers as those for Table 71. See Table 71 for the meanings of these notes.

Table 73. MPC8545E Pinout Listing

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI1 and PCI2 (One 64-Bit or Two 32-Bit)				
PCI1_AD[63:32]/PCI2_AD[31:0]	AB14, AC15, AA15, Y16, W16, AB16, AC16, AA16, AE17, AA18, W18, AC17, AD16, AE16, Y17, AC18, AB18, AA19, AB19, AB21, AA20, AC20, AB20, AB22, AC22, AD21, AB23, AF23, AD23, AE23, AC23, AC24	I/O	OV _{DD}	17
PCI1_AD[31:0]	AH6, AE7, AF7, AG7, AH7, AF8, AH8, AE9, AH9, AC10, AB10, AD10, AG10, AA10, AH10, AA11, AB12, AE12, AG12, AH12, AB13, AA12, AC13, AE13, Y14, W13, AG13, V14, AH13, AC14, Y15, AB15	I/O	OV _{DD}	17
PCI1_C_BE[7:4]/PCI2_C_BE[3:0]	AF15, AD14, AE15, AD15	I/O	OV _{DD}	17
PCI1_C_BE[3:0]	AF9, AD11, Y12, Y13	I/O	OV _{DD}	17
PCI1_PAR64/PCI2_PAR	W15	I/O	OV _{DD}	—
PCI1_GNT[4:1]	AG6, AE6, AF5, AH5	O	OV _{DD}	5, 9, 35
PCI1_GNT0	AG5	I/O	OV _{DD}	—
PCI1_IRDY	AF11	I/O	OV _{DD}	2
PCI1_PAR	AD12	I/O	OV _{DD}	—
PCI1_PERR	AC12	I/O	OV _{DD}	2
PCI1_SERR	V13	I/O	OV _{DD}	2, 4
PCI1_STOP	W12	I/O	OV _{DD}	2
PCI1_TRDY	AG11	I/O	OV _{DD}	2
PCI1_REQ[4:1]	AH2, AG4, AG3, AH4	I	OV _{DD}	—
PCI1_REQ0	AH3	I/O	OV _{DD}	—
PCI1_CLK	AH26	I	OV _{DD}	39
PCI1_DEVSEL	AH11	I/O	OV _{DD}	2

Table 73. MPC8545E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
\overline{UDE}	AH16	I	OV _{DD}	—
\overline{MCP}	AG19	I	OV _{DD}	—
IRQ[0:7]	AG23, AF18, AE18, AF20, AG18, AF17, AH24, AE20	I	OV _{DD}	—
IRQ[8]	AF19	I	OV _{DD}	—
IRQ[9]/ $\overline{DMA_DREQ3}$	AF21	I	OV _{DD}	1
IRQ[10]/ $\overline{DMA_DACK3}$	AE19	I/O	OV _{DD}	1
IRQ[11]/ $\overline{DMA_DDONE3}$	AD20	I/O	OV _{DD}	1
$\overline{IRQ_OUT}$	AD18	O	OV _{DD}	2, 4
Ethernet Management Interface				
EC_MDC	AB9	O	OV _{DD}	5, 9
EC_MDIO	AC8	I/O	OV _{DD}	—
Gigabit Reference Clock				
EC_GTX_CLK125	V11	I	LV _{DD}	—
Three-Speed Ethernet Controller (Gigabit Ethernet 1)				
TSEC1_RXD[7:0]	R5, U1, R3, U2, V3, V1, T3, T2	I	LV _{DD}	—
TSEC1_TXD[7:0]	T10, V7, U10, U5, U4, V6, T5, T8	O	LV _{DD}	5, 9
TSEC1_COL	R4	I	LV _{DD}	—
TSEC1_CRS	V5	I/O	LV _{DD}	20
TSEC1_GTX_CLK	U7	O	LV _{DD}	—
TSEC1_RX_CLK	U3	I	LV _{DD}	—
TSEC1_RX_DV	V2	I	LV _{DD}	—
TSEC1_RX_ER	T1	I	LV _{DD}	—
TSEC1_TX_CLK	T6	I	LV _{DD}	—
TSEC1_TX_EN	U9	O	LV _{DD}	30
TSEC1_TX_ER	T7	O	LV _{DD}	—
GPIN[0:7]	P2, R2, N1, N2, P3, M2, M1, N3	I	LV _{DD}	103
GPOUT[0:5]	N9, N10, P8, N7, R9, N5	O	LV _{DD}	—
cfg_dram_type0/GPOUT6	R8	O	LV _{DD}	5, 9
GPOUT7	N6	O	LV _{DD}	—
Reserved	P1	—	—	104
Reserved	R6	—	—	104
Reserved	P6	—	—	15
Reserved	N4	—	—	105

Table 73. MPC8545E Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
GV _{DD}	B3, B11, C7, C9, C14, C17, D4, D6, D10, D15, E2, E8, E11, E18, F5, F12, F16, G3, G7, G9, G11, H5, H12, H15, H17, J10, K3, K12, K16, K18, L6, M4, M8, M13	Power for DDR1 and DDR2 DRAM I/O voltage (1.8 V, 2.5 V)	GV _{DD}	—
BV _{DD}	C21, C24, C27, E20, E25, G19, G23, H26, J20	Power for local bus (1.8 V, 2.5 V, 3.3 V)	BV _{DD}	—
V _{DD}	M19, N12, N14, N16, N18, P11, P13, P15, P17, P19, R12, R14, R16, R18, T11, T13, T15, T17, T19, U12, U14, U16, U18, V17, V19	Power for core (1.1 V)	V _{DD}	—
SV _{DD}	L25, L27, M24, N28, P24, P26, R24, R27, T25, V24, V26, W24, W27, Y25, AA28, AC27	Core power for SerDes transceivers (1.1 V)	SV _{DD}	—
XV _{DD}	L20, L22, N23, P21, R22, T20, U23, V21, W22, Y20	Pad power for SerDes transceivers (1.1 V)	XV _{DD}	—
AVDD_LBIU	J28	Power for local bus PLL (1.1 V)	—	26
AVDD_PCI1	AH21	Power for PCI1 PLL (1.1 V)	—	26
AVDD_PCI2	AH22	Power for PCI2 PLL (1.1 V)	—	26
AVDD_CORE	AH15	Power for e500 PLL (1.1 V)	—	26
AVDD_PLAT	AH19	Power for CCB PLL (1.1 V)	—	26
AVDD_SRDS	U25	Power for SRDSPLL (1.1 V)	—	26
SENSEVDD	M14	O	V _{DD}	13
SENSEVSS	M16	—	—	13
Analog Signals				
MVREF	A18	I Reference voltage signal for DDR	MVREF	—

20 Clocking

This section describes the PLL configuration of the device. Note that the platform clock is identical to the core complex bus (CCB) clock.

20.1 Clock Ranges

Table 75 through Table 77 provide the clocking specifications for the processor cores and Table 78, through Table 80 provide the clocking specifications for the memory bus.

Table 75. Processor Core Clocking Specifications (MPC8548E and MPC8547E)

Characteristic	Maximum Processor Core Frequency						Unit	Notes
	1000 MHz		1200 MHz		1333 MHz			
	Min	Max	Min	Max	Min	Max		
e500 core processor frequency	800	1000	800	1200	800	1333	MHz	1, 2

Notes:

- Caution:** The CCB to SYSCLK ratio and e500 core to CCB ratio settings must be chosen such that the resulting SYSCLK frequency, e500 (core) frequency, and CCB frequency do not exceed their respective maximum or minimum operating frequencies. See Section 20.2, "CCB/SYSCLK PLL Ratio," and Section 20.3, "e500 Core PLL Ratio," for ratio settings.
- The minimum e500 core frequency is based on the minimum platform frequency of 333 MHz.

Table 76. Processor Core Clocking Specifications (MPC8545E)

Characteristic	Maximum Processor Core Frequency						Unit	Notes
	800 MHz		1000 MHz		1200 MHz			
	Min	Max	Min	Max	Min	Max		
e500 core processor frequency	800	800	800	1000	800	1200	MHz	1, 2

Notes:

- Caution:** The CCB to SYSCLK ratio and e500 core to CCB ratio settings must be chosen such that the resulting SYSCLK frequency, e500 (core) frequency, and CCB frequency do not exceed their respective maximum or minimum operating frequencies. See Section 20.2, "CCB/SYSCLK PLL Ratio," and Section 20.3, "e500 Core PLL Ratio," for ratio settings.
- The minimum e500 core frequency is based on the minimum platform frequency of 333 MHz.

level must always be equivalent to V_{DD} , and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide independent filter circuits per PLL power supply as illustrated in [Figure 57](#), one to each of the AV_{DD} pins. By providing independent filters to each PLL the opportunity to cause noise injection from one PLL to the other is reduced.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It must be built with surface mount capacitors with minimum Effective Series Inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit must be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It must be routed directly from the capacitors to the AV_{DD} pin, which is on the periphery of the footprint, without the inductance of vias.

[Figure 57](#) through [Figure 59](#) shows the PLL power supply filter circuits.

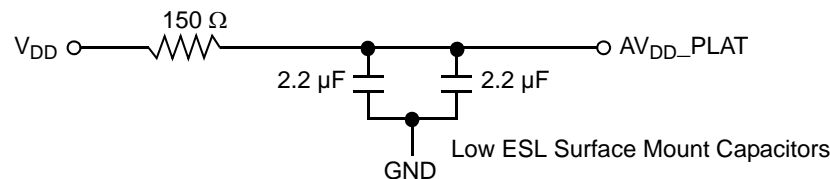


Figure 57. PLL Power Supply Filter Circuit with PLAT Pins

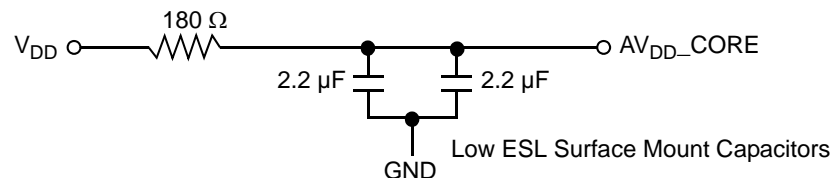


Figure 58. PLL Power Supply Filter Circuit with CORE Pins

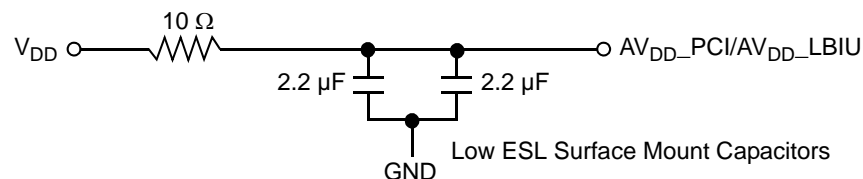


Figure 59. PLL Power Supply Filter Circuit with PCI/LBIU Pins

The AV_{DD_SRDS} signal provides power for the analog portions of the SerDes PLL. To ensure stability of the internal clock, the power supplied to the PLL is filtered using a circuit similar to the one shown in following figure. For maximum effectiveness, the filter circuit is placed as closely as possible to the AV_{DD_SRDS} ball to ensure it filters out as much noise as possible. The ground connection must be near the AV_{DD_SRDS} ball. The 0.003- μF capacitor is closest to the ball, followed by the two 2.2 μF capacitors, and finally the 1 Ω resistor to the board supply plane. The capacitors are connected from AV_{DD_SRDS} to

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