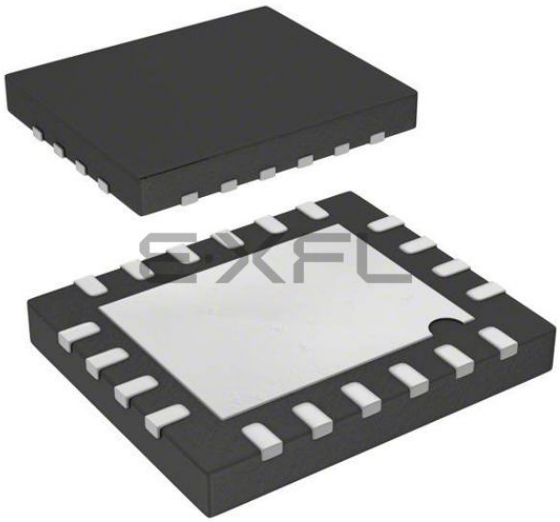

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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	ST7
Core Size	8-Bit
Speed	8MHz
Connectivity	SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	17
Program Memory Size	4KB (4K x 8)
Program Memory Type	FLASH
EEPROM Size	128 x 8
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 7x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	20-VQFN Exposed Pad
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/st7flit19bf1u6

DATA EEPROM (Cont'd)

5.4 POWER SAVING MODES

Wait mode

The DATA EEPROM can enter WAIT mode on execution of the WFI instruction of the microcontroller or when the microcontroller enters Active-HALT mode.The DATA EEPROM will immediately enter this mode if there is no programming in progress, otherwise the DATA EEPROM will finish the cycle and then enter WAIT mode.

Active-Halt mode

Refer to Wait mode.

Halt mode

The DATA EEPROM immediately enters HALT mode if the microcontroller executes the HALT instruction. Therefore the EEPROM will stop the function in progress, and data may be corrupted.

5.5 ACCESS ERROR HANDLING

If a read access occurs while E2LAT=1, then the data bus will not be driven.

If a write access occurs while E2LAT=0, then the data on the bus will not be latched.

If a programming cycle is interrupted (by a RESET action), the integrity of the data in memory will not be guaranteed.

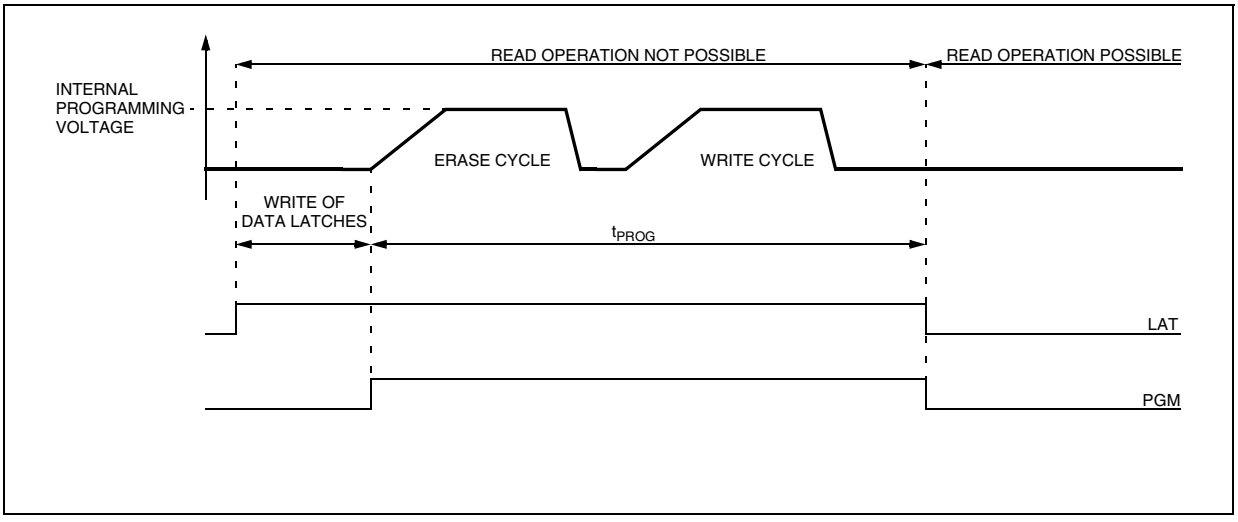
5.6 Data EEPROM Read-out Protection

The read-out protection is enabled through an option bit (see option byte section).

When this option is selected, the programs and data stored in the EEPROM memory are protected against read-out (including a re-write protection). In Flash devices, when this protection is removed by reprogramming the Option Byte, the entire Program memory and EEPROM is first automatically erased.

Note: Both Program Memory and data EEPROM are protected using the same option bit.

Figure 10. Data EEPROM Programming Cycle



CPU REGISTERS (Cont'd)

STACK POINTER (SP)

Read/Write

Reset Value: 01FFh

15							8
0	0	0	0	0	0	0	1

7							0
1	SP6	SP5	SP4	SP3	SP2	SP1	SP0

The Stack Pointer is a 16-bit register which is always pointing to the next free location in the stack. It is then decremented after data has been pushed onto the stack and incremented before data is popped from the stack (see Figure 12).

Since the stack is 128 bytes deep, the 9 most significant bits are forced by hardware. Following an MCU Reset, or after a Reset Stack Pointer instruction (RSP), the Stack Pointer contains its reset value (the SP6 to SP0 bits are set) which is the stack higher address.

The least significant byte of the Stack Pointer (called S) can be directly accessed by a LD instruction.

Note: When the lower limit is exceeded, the Stack Pointer wraps around to the stack upper limit, without indicating the stack overflow. The previously stored information is then overwritten and therefore lost. The stack also wraps in case of an underflow.

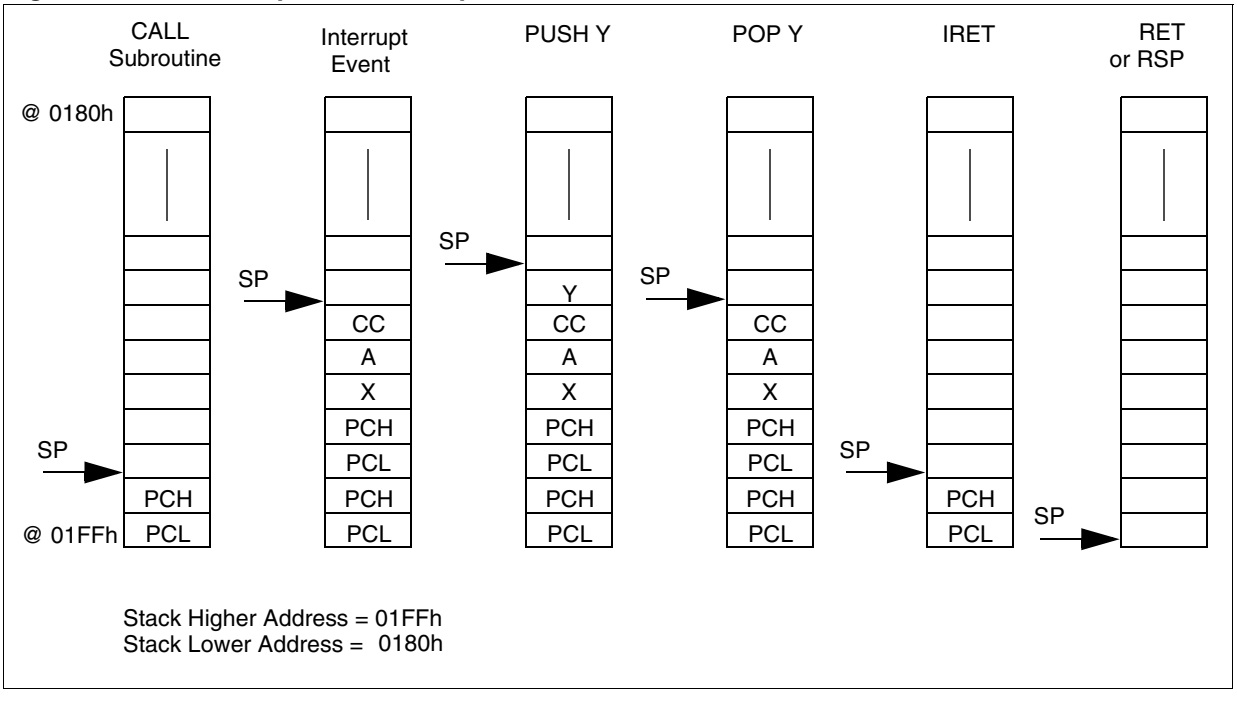
The stack is used to save the return address during a subroutine call and the CPU context during an interrupt. The user may also directly manipulate the stack by means of the PUSH and POP instructions. In the case of an interrupt, the PCL is stored at the first location pointed to by the SP. Then the other registers are stored in the next locations as shown in Figure 12.

– When an interrupt is received, the SP is decremented and the context is pushed on the stack.

– On return from interrupt, the SP is incremented and the context is popped from the stack.

A subroutine call occupies two locations and an interrupt five locations in the stack area.

Figure 12. Stack Manipulation Example



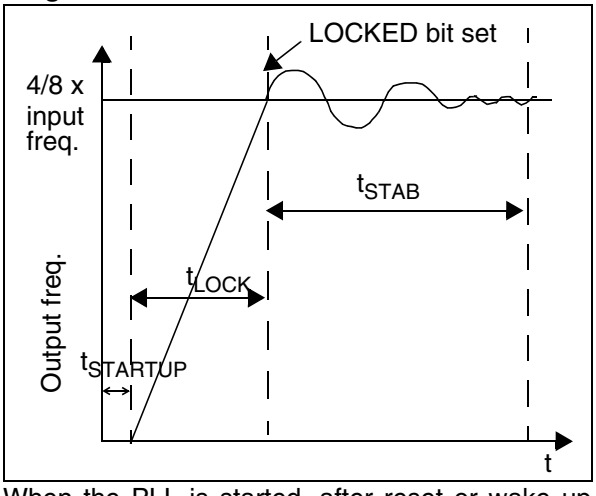
– The x8 PLL is intended for operation with V_{DD} in the 3.3V to 5.5V range [1)]

Refer to Section 15.1 for the option byte description.

If the PLL is disabled and the RC oscillator is enabled, then f_{OSC} = 1MHz.

If both the RC oscillator and the PLL are disabled, f_{OSC} is driven by the external clock.

Figure 13. PLL Output Frequency Timing Diagram



When the PLL is started, after reset or wake up from Halt mode or AWUFH mode, it outputs the clock after a delay of $t_{STARTUP}$.

When the PLL output signal reaches the operating frequency, the LOCKED bit in the SICSCR register is set. Full PLL accuracy (ACC_{PLL}) is reached after a stabilization time of t_{STAB} (see Figure 13 and 13.3.5 Internal RC Oscillator and PLL)

Refer to section 7.6.4 on page 35 for a description of the LOCKED bit in the SICSR register.

Note 1:

It is possible to obtain f_{OSC} = 4MHz in the 3.3V to 5.5V range with internal RC and PLL enabled by selecting 1MHz RC and x8 PLL and setting the PLLdiv2 bit in the PLLTST register (see section 7.6.4 on page 35).

7.6 SYSTEM INTEGRITY MANAGEMENT (SI)

The System Integrity Management block contains the Low voltage Detector (LVD) and Auxiliary Voltage Detector (AVD) functions. It is managed by the SICSR register.

Note: A reset can also be triggered following the detection of an illegal opcode or prebyte code. Refer to section 12.2.1 on page 107 for further details.

7.6.1 Low Voltage Detector (LVD)

The Low Voltage Detector function (LVD) generates a static reset when the V_{DD} supply voltage is below a $V_{IT-(LVD)}$ reference value. This means that it secures the power-up as well as the power-down keeping the ST7 in reset.

The $V_{IT-(LVD)}$ reference value for a voltage drop is lower than the $V_{IT+(LVD)}$ reference value for power-on in order to avoid a parasitic reset when the MCU starts running and sinks current on the supply (hysteresis).

The LVD Reset circuitry generates a reset when V_{DD} is below:

– $V_{IT+(LVD)}$when V_{DD} is rising

– $V_{IT-(LVD)}$ when V_{DD} is falling

The LVD function is illustrated in Figure 18.

The voltage threshold can be configured by option byte to be low, medium or high.

Provided the minimum V_{DD} value (guaranteed for the oscillator frequency) is above $V_{IT-(LVD)}$, the MCU can only be in two modes:

– under full software control

– in static safe reset

In these conditions, secure operation is always ensured for the application without the need for external reset hardware.

During a Low Voltage Detector Reset, the $\overline{\text{RESET}}$ pin is held low, thus permitting the MCU to reset other devices.

Notes:

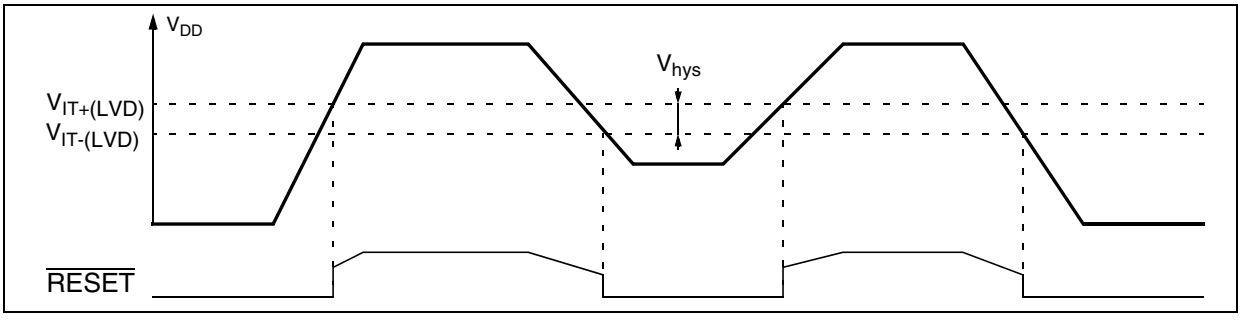
The LVD allows the device to be used without any external RESET circuitry.

The LVD is an optional function which can be selected by option byte.

Use of LVD with capacitive power supply: with this type of power supply, if power cuts occur in the application, it is recommended to pull V_{DD} down to 0V to ensure optimum restart conditions. Refer to circuit example in Figure 106 on page 136 and note 4.

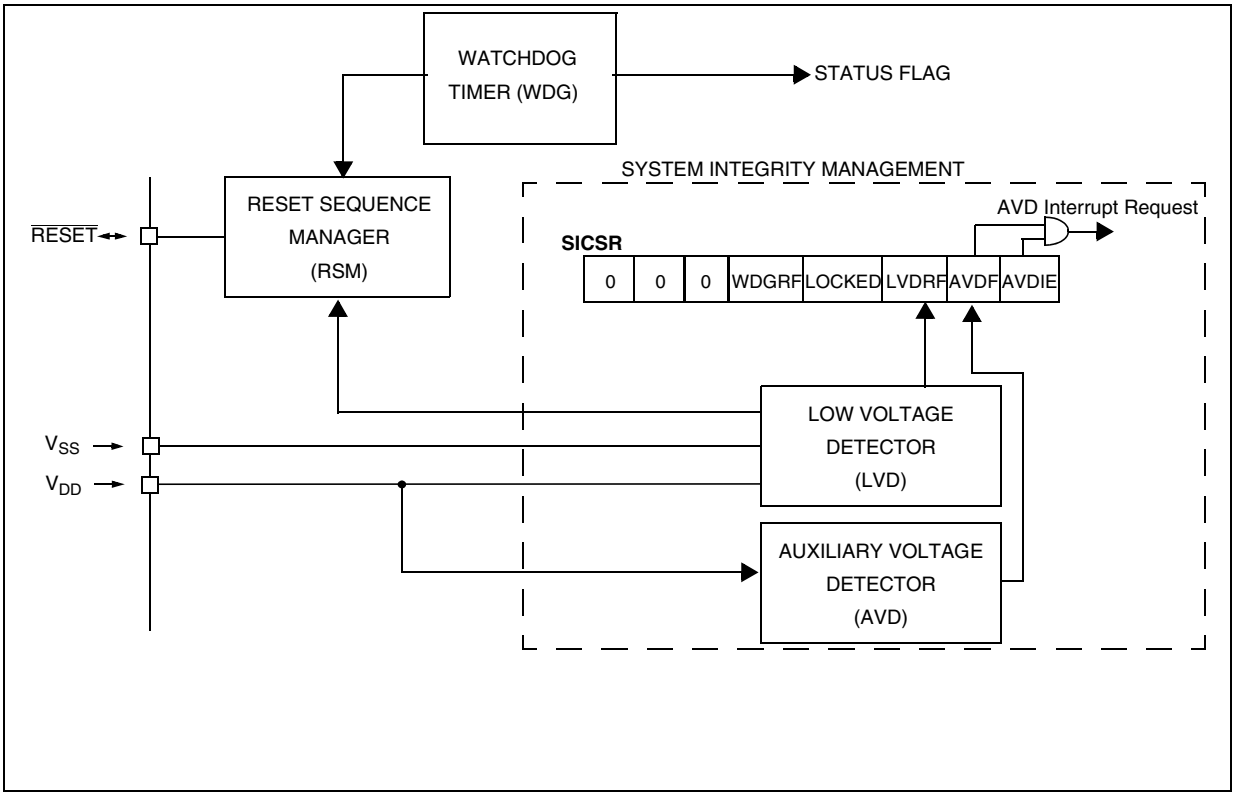
It is recommended to make sure that the V_{DD} supply voltage rises monotonously when the device is exiting from Reset, to ensure the application functions properly.

Figure 18. Low Voltage Detector vs Reset



SYSTEM INTEGRITY MANAGEMENT (Cont'd)

Figure 19. Reset and Supply Management Block Diagram



POWER SAVING MODES (Cont'd)

Similarities with Halt mode

The following AWUFH mode behaviour is the same as normal Halt mode:

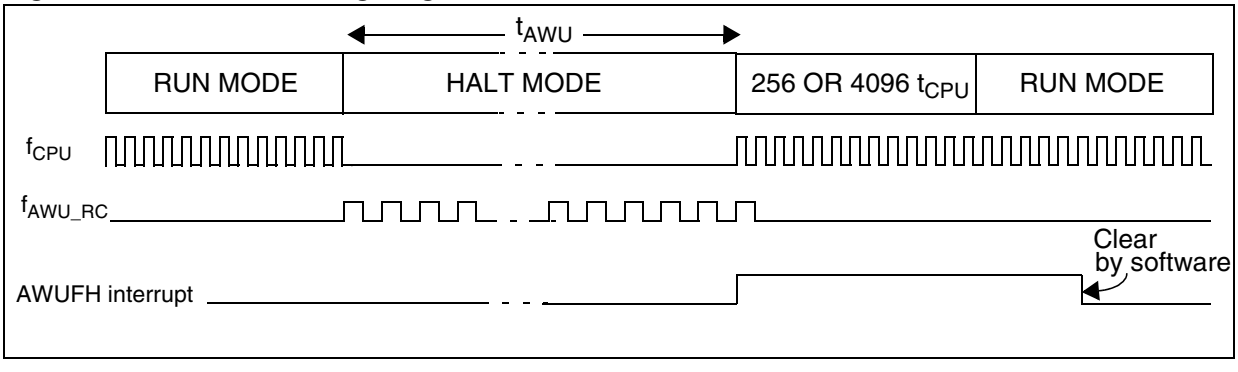
– The MCU can exit AWUFH mode by means of any interrupt with exit from Halt capability or a reset (see Section 9.4 HALT MODE).

– When entering AWUFH mode, the I bit in the CC register is forced to 0 to enable interrupts. Therefore, if an interrupt is pending, the MCU wakes up immediately.

– In AWUFH mode, the main oscillator is turned off causing all internal processing to be stopped, including the operation of the on-chip peripherals. None of the peripherals are clocked except those which get their clock supply from another clock generator (such as an external or auxiliary oscillator like the AWU oscillator).

– The compatibility of Watchdog operation with AWUFH mode is configured by the WDGHALT option bit in the option byte. Depending on this setting, the HALT instruction when executed while the Watchdog system is enabled, can generate a Watchdog RESET.

Figure 30. AWUF Halt Timing Diagram



I/O PORTS (Cont'd)

10.7 DEVICE-SPECIFIC I/O PORT CONFIGURATION

The I/O port register configurations are summarised as follows.

Standard Ports

PA7:0, PB6:0

MODE	DDR	OR
floating input	0	0
pull-up input	0	1
open drain output	1	0
push-pull output	1	1

PC1:0 (multiplexed with OSC1,OSC2)

MODE	DDR
floating input	0
push-pull output	1

The selection between OSC1 or PC0 and OSC2 or PC1 is done by option byte. Refer to section 15.1 on page 149. Interrupt capability is not available on PC1:0.

Note: PCOR not implemented but p-transistor always active in output mode (refer to Figure 32 on page 50)

Interrupt Ports

Ports where the external interrupt capability is selected using the EISR register

MODE	DDR	OR
floating input	0	0
pull-up interrupt input	0	1
open drain output	1	0
push-pull output	1	1

Table 10. Port Configuration (Standard ports)

Port	Pin name	Input		Output	
		OR = 0	OR = 1	OR = 0	OR = 1
Port A	PA7:0	floating	pull-up	open drain	push-pull
Port B	PB6:0	floating	pull-up	open drain	push-pull

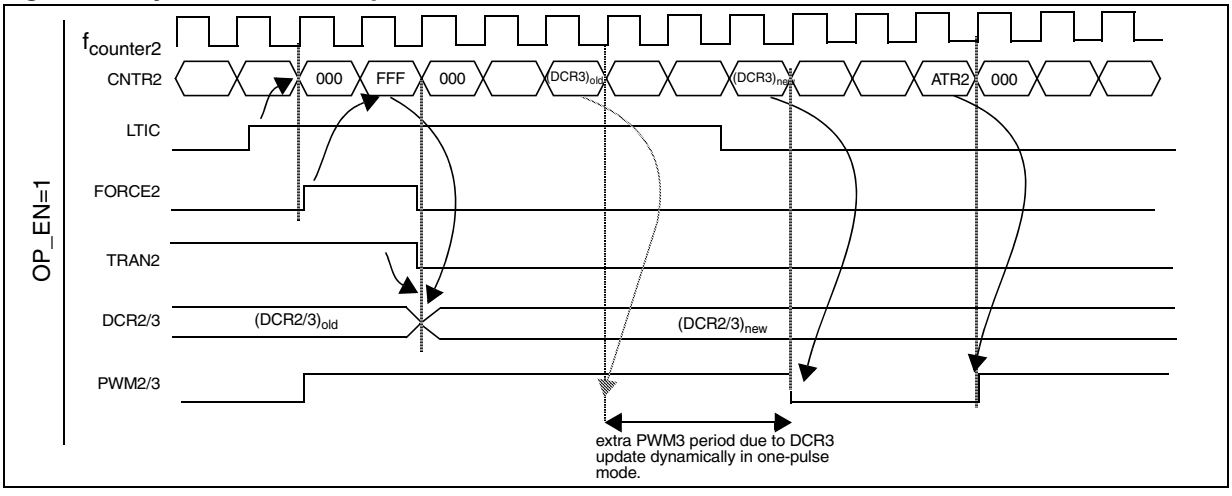
Note: On ports where the external interrupt capability is selected using the EISR register, the configuration will be as follows:

Port	Pin name	Input		Output	
		OR = 0	OR = 1	OR = 0	OR = 1
Port A	PA7:0	floating	pull-up interrupt	open drain	push-pull
Port B	PB6:0	floating	pull-up interrupt	open drain	push-pull

Table 11. I/O Port Register Map and Reset Values

Address (Hex.)	Register Label	7	6	5	4	3	2	1	0
0000h	**PADR**	MSB							LSB
	Reset Value	1	1	1	1	1	1	1	1
0001h	**PADDR**	MSB							LSB
	Reset Value	0	0	0	0	0	0	0	0

Figure 49. Dynamic DCR2/3 update in One Pulse Mode



11.3 LITE TIMER 2 (LT2)

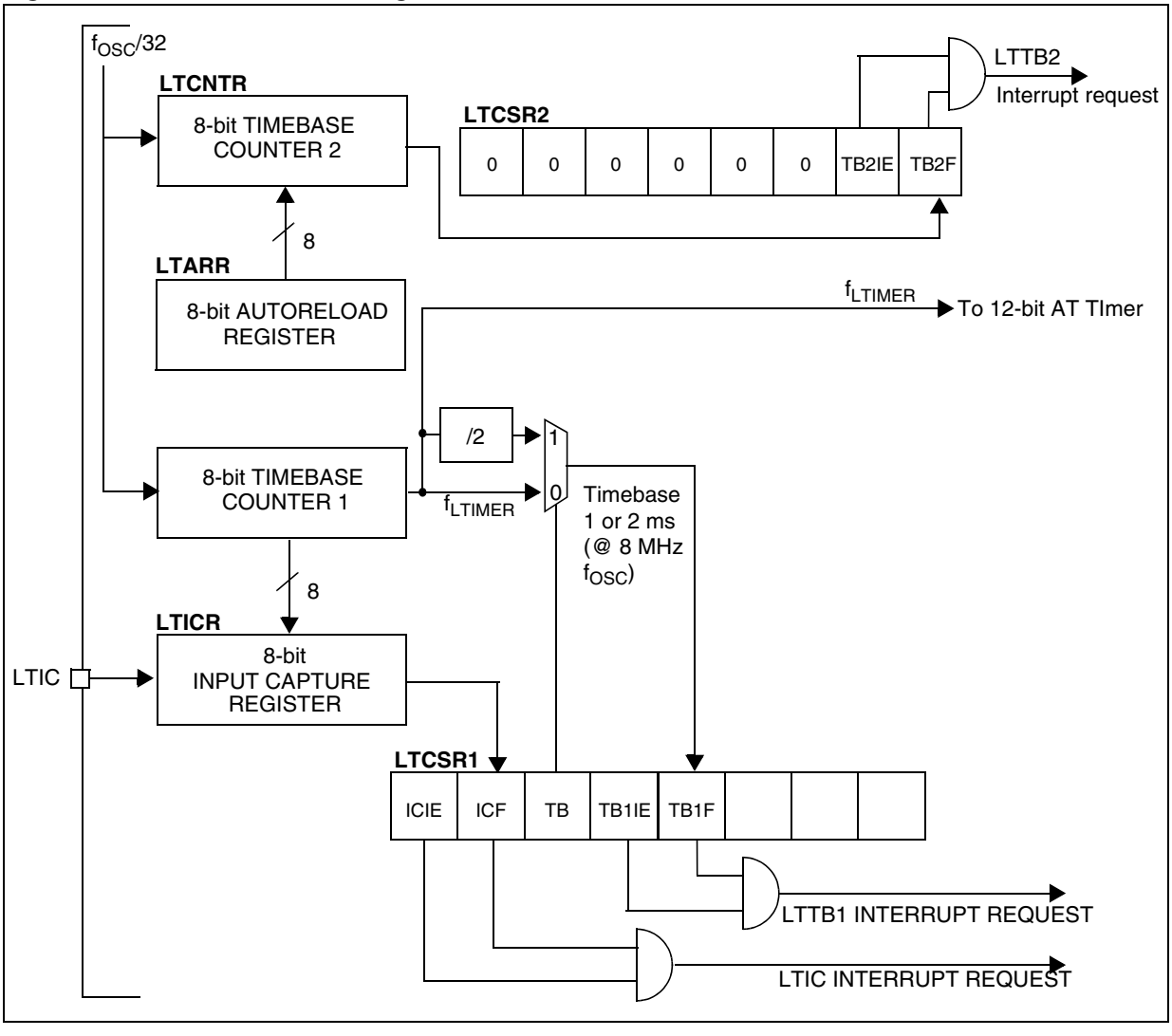
11.3.1 Introduction

The Lite Timer can be used for general-purpose timing functions. It is based on two free-running 8-bit upcounters and an 8-bit input capture register.

11.3.2 Main Features

- Realtime Clock
  - One 8-bit upcounter 1 ms or 2 ms timebase period (@ 8 MHz f_{OSC})
  - One 8-bit upcounter with autoreload and programmable timebase period from 4µs to 1.024ms in 4µs increments (@ 8 MHz f_{OSC})
  - 2 Maskable timebase interrupts
- Input Capture
  - 8-bit input capture register (LTICR)
  - Maskable interrupt with wake-up from Halt mode capability

Figure 51. Lite Timer 2 Block Diagram



ON-CHIP PERIPHERALS (cont'd)

11.4 SERIAL PERIPHERAL INTERFACE (SPI)

11.4.1 Introduction

The Serial Peripheral Interface (SPI) allows full-duplex, synchronous, serial communication with external devices. An SPI system may consist of a master and one or more slaves or a system in which devices may be either masters or slaves.

11.4.2 Main Features

- Full duplex synchronous transfers (on three lines)
- Simplex synchronous transfers (on two lines)
- Master or slave operation
- 6 master mode frequencies (f_{CPU}/4 max.)
- f_{CPU}/2 max. slave mode frequency (see note)
- $\overline{SS}$ Management by software or hardware
- Programmable clock polarity and phase
- End of transfer interrupt flag
- Write collision, Master Mode Fault and Overrun flags

Note: In slave mode, continuous transmission is not possible at maximum frequency due to the software overhead for clearing status flags and to initiate the next transmission sequence.

11.4.3 General Description

Figure 1 on page 3 shows the serial peripheral interface (SPI) block diagram. There are three registers:

- SPI Control Register (SPICR)
- SPI Control/Status Register (SPICSR)
- SPI Data Register (SPIDR)

The SPI is connected to external devices through four pins:

- MISO: Master In / Slave Out data
- MOSI: Master Out / Slave In data
- SCK: Serial Clock out by SPI masters and input by SPI slaves
- $\overline{SS}$: Slave select:
  This input signal acts as a 'chip select' to let the SPI master communicate with slaves individually and to avoid contention on the data lines. Slave $\overline{SS}$ inputs can be driven by standard I/O ports on the master Device.

SERIAL PERIPHERAL INTERFACE (cont'd)

11.4.3.3 Master Mode Operation

In master mode, the serial clock is output on the SCK pin. The clock frequency, polarity and phase are configured by software (refer to the description of the SPICSR register).

Note: The idle state of SCK must correspond to the polarity selected in the SPICSR register (by pulling up SCK if CPOL = 1 or pulling down SCK if CPOL = 0).

How to operate the SPI in master mode

To operate the SPI in master mode, perform the following steps in order:

1. Write to the SPICR register:
   - Select the clock frequency by configuring the SPR[2:0] bits.
   - Select the clock polarity and clock phase by configuring the CPOL and CPHA bits. Figure 5 shows the four possible configurations.
     Note: The slave must have the same CPOL and CPHA settings as the master.
2. Write to the SPICSR register:
   - Either set the SSM bit and set the SSI bit or clear the SSM bit and tie the SS pin high for the complete byte transmit sequence.
3. Write to the SPICR register:
   - Set the MSTR and SPE bits
     Note: MSTR and SPE bits remain set only if SS is high).

Important note: if the SPICSR register is not written first, the SPICR register setting (MSTR bit) may be not taken into account.

The transmit sequence begins when software writes a byte in the SPIDR register.

11.4.3.4 Master Mode Transmit Sequence

When software writes to the SPIDR register, the data byte is loaded into the 8-bit shift register and then shifted out serially to the MOSI pin most significant bit first.

When data transfer is complete:

- The SPIF bit is set by hardware.
- An interrupt request is generated if the SPIE bit is set and the interrupt mask in the CCR register is cleared.

Clearing the SPIF bit is performed by the following software sequence:

1. An access to the SPICSR register while the SPIF bit is set
2. A read to the SPIDR register

Note: While the SPIF bit is set, all writes to the SPIDR register are inhibited until the SPICSR register is read.

11.4.3.5 Slave Mode Operation

In slave mode, the serial clock is received on the SCK pin from the master device.

To operate the SPI in slave mode:

1. Write to the SPICSR register to perform the following actions:
   - Select the clock polarity and clock phase by configuring the CPOL and CPHA bits (see Figure 5).
     Note: The slave must have the same CPOL and CPHA settings as the master.
   - Manage the $\overline{SS}$ pin as described in Section 0.1.3.2 and Figure 3. If CPHA = 1 $\overline{SS}$ must be held low continuously. If CPHA = 0 $\overline{SS}$ must be held low during byte transmission and pulled up between each byte to let the slave write in the shift register.
2. Write to the SPICR register to clear the MSTR bit and set the SPE bit to enable the SPI I/O functions.

11.4.3.6 Slave Mode Transmit Sequence

When software writes to the SPIDR register, the data byte is loaded into the 8-bit shift register and then shifted out serially to the MISO pin most significant bit first.

The transmit sequence begins when the slave device receives the clock signal and the most significant bit of the data on its MOSI pin.

When data transfer is complete:

- The SPIF bit is set by hardware.
- An interrupt request is generated if SPIE bit is set and interrupt mask in the CCR register is cleared.

Clearing the SPIF bit is performed by the following software sequence:

1. An access to the SPICSR register while the SPIF bit is set
2. A write or a read to the SPIDR register

Notes: While the SPIF bit is set, all writes to the SPIDR register are inhibited until the SPICSR register is read.

The SPIF bit can be cleared during a second transmission; however, it must be cleared before the second SPIF bit in order to prevent an Overrun condition (see Section 0.1.5.2).

SERIAL PERIPHERAL INTERFACE (cont'd)

11.4.5 Error Flags

11.4.5.1 Master Mode Fault (MODF)

Master mode fault occurs when the master device's $\overline{SS}$ pin is pulled low.

When a Master mode fault occurs:

– The MODF bit is set and an SPI interrupt request is generated if the SPIE bit is set.

– The SPE bit is reset. This blocks all output from the device and disables the SPI peripheral.

– The MSTR bit is reset, thus forcing the device into slave mode.

Clearing the MODF bit is done through a software sequence:

1. A read access to the SPICSR register while the MODF bit is set.

2. A write to the SPICR register.

Notes: To avoid any conflicts in an application with multiple slaves, the $\overline{SS}$ pin must be pulled high during the MODF bit clearing sequence. The SPE and MSTR bits may be restored to their original state during or after this clearing sequence.

Hardware does not allow the user to set the SPE and MSTR bits while the MODF bit is set except in the MODF bit clearing sequence.

In a slave device, the MODF bit can not be set, but in a multimaster configuration the device can be in slave mode with the MODF bit set.

The MODF bit indicates that there might have been a multimaster conflict and allows software to handle this using an interrupt routine and either perform a reset or return to an application default state.

11.4.5.2 Overrun Condition (OVR)

An overrun condition occurs when the master device has sent a data byte and the slave device has not cleared the SPIF bit issued from the previously transmitted byte.

When an Overrun occurs:

– The OVR bit is set and an interrupt request is generated if the SPIE bit is set.

In this case, the receiver buffer contains the byte sent after the SPIF bit was last cleared. A read to the SPIDR register returns this byte. All other bytes are lost.

The OVR bit is cleared by reading the SPICSR register.

11.4.5.3 Write Collision Error (WCOL)

A write collision occurs when the software tries to write to the SPIDR register while a data transfer is taking place with an external device. When this happens, the transfer continues uninterrupted and the software write will be unsuccessful.

Write collisions can occur both in master and slave mode. See also Section 0.1.3.2 Slave Select Management.

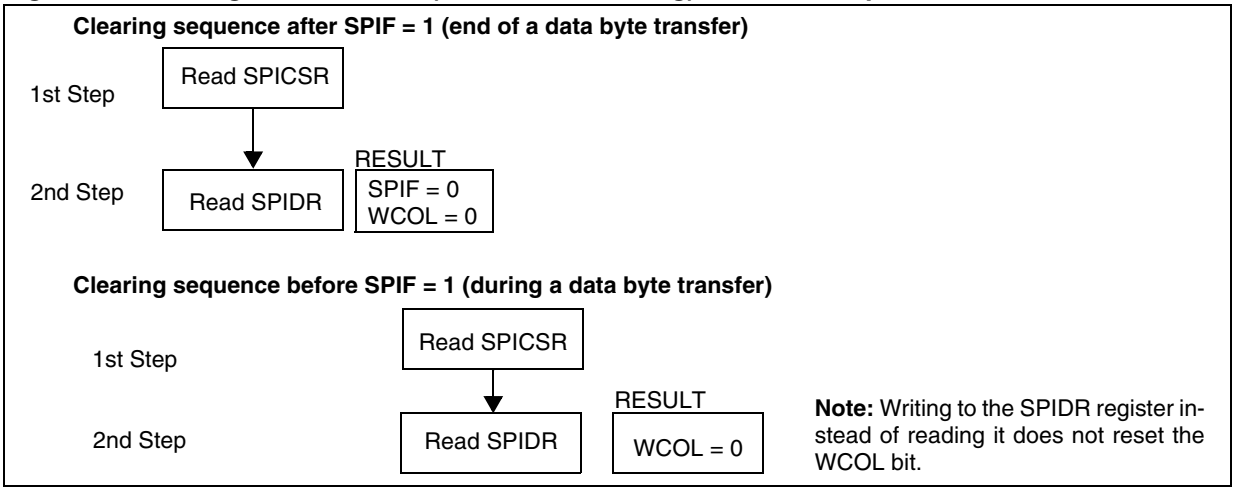
Note: A "read collision" will never occur since the received data byte is placed in a buffer in which access is always synchronous with the CPU operation.

The WCOL bit in the SPICSR register is set if a write collision occurs.

No SPI interrupt is generated when the WCOL bit is set (the WCOL bit is a status flag only).

Clearing the WCOL bit is done through a software sequence (see Figure 6).

Figure 58. Clearing the WCOL Bit (Write Collision Flag) Software Sequence



13 ELECTRICAL CHARACTERISTICS

13.1 PARAMETER CONDITIONS

Unless otherwise specified, all voltages are referred to V_{SS}.

13.1.1 Minimum and Maximum values

Unless otherwise specified the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at T_A=25°C and T_A=$T_A$max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean±3Σ).

13.1.2 Typical values

Unless otherwise specified, typical data are based on T_A=25°C, V_{DD}=5V (for the 4.5V≤V_{DD}≤5.5V voltage range) and V_{DD}=3.3V (for the 3V≤V_{DD}≤3.6V voltage range). They are given only as design guidelines and are not tested.

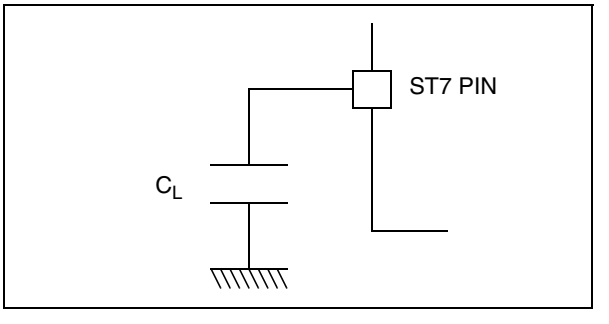
13.1.3 Typical curves

Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

13.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in Figure 63.

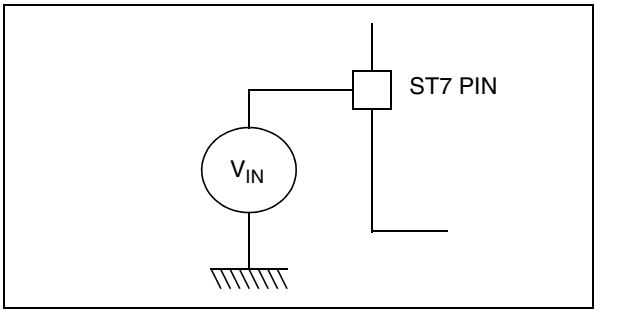
Figure 63. Pin loading conditions



13.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in Figure 64.

Figure 64. Pin input voltage



13.3.6 Operating conditions with ADC

T_A = -40 to 125°C, unless otherwise specified

Symbol	Parameter	Typ	Unit
$I_{INJ(ANA)}$ [1]	Injected current on any analog pin	0	mA

Note:
1. Current injection (negative or positive) not allowed on any analog pin.

13.7 EMC CHARACTERISTICS

Susceptibility tests are performed on a sample basis during product characterization.

13.7.1 Functional EMS (Electro Magnetic Susceptibility)

Based on a simple running application on the product (toggling 2 LEDs through I/O ports), the product is stressed by two electro magnetic events until a failure occurs (indicated by the LEDs).

- **ESD**: Electro-Static Discharge (positive and negative) is applied on all pins of the device until a functional disturbance occurs. This test conforms with the IEC 1000-4-2 standard.
- **FTB**: A Burst of Fast Transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100pF capacitor, until a functional disturbance occurs. This test conforms with the IEC 1000-4-4 standard.

A device reset allows normal operations to be resumed. The test results are given in the table below based on the EMS levels and classes defined in application note AN1709.

13.7.1.1 Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

Software recommendations:

The software flowchart must include the management of runaway conditions such as:

– Corrupted program counter

– Unexpected reset

– Critical Data corruption (control registers...)

Prequalification trials:

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the RESET pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behaviour is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

Symbol	Parameter	Conditions	Level/ Class
V_{FESD}	Voltage limits to be applied on any I/O pin to induce a functional disturbance	V_{DD}=5V, T_A=+25°C, f_{OSC}=8MHz conforms to IEC 1000-4-2	2B
V_{FFTB}	Fast transient voltage burst limits to be applied through 100pF on V_{DD} and V_{SS} pins to induce a functional disturbance	V_{DD}=5V, T_A=+25°C, f_{OSC}=8MHz conforms to IEC 1000-4-4	3B

13.7.2 Electro Magnetic Interference (EMI)

Based on a simple application running on the product (toggling 2 LEDs through the I/O ports), the product is monitored in terms of emission. This emission test is in line with the norm SAE J 1752/3 which specifies the board and the loading of each pin.

Symbol	Parameter	Conditions	Monitored Frequency Band	Max vs. [f_{OSC}/f_{CPU}]		Unit
				8/4MHz	16/8MHz	
S_{EMI}	Peak level	V_{DD}=5V, T_A=+25°C, SO20 package, conforming to SAE J 1752/3	0.1MHz to 30MHz	15	21	dBµV
			30MHz to 130MHz	22	29	
			130MHz to 1GHz	17	22	
			SAE EMI Level	3.5	3.5	-

Note:

1. Data based on characterization results, not tested in production.

COMMUNICATION INTERFACE CHARACTERISTICS (Cont'd)

Figure 108. SPI Slave Timing Diagram with CPHA=1 [1)]

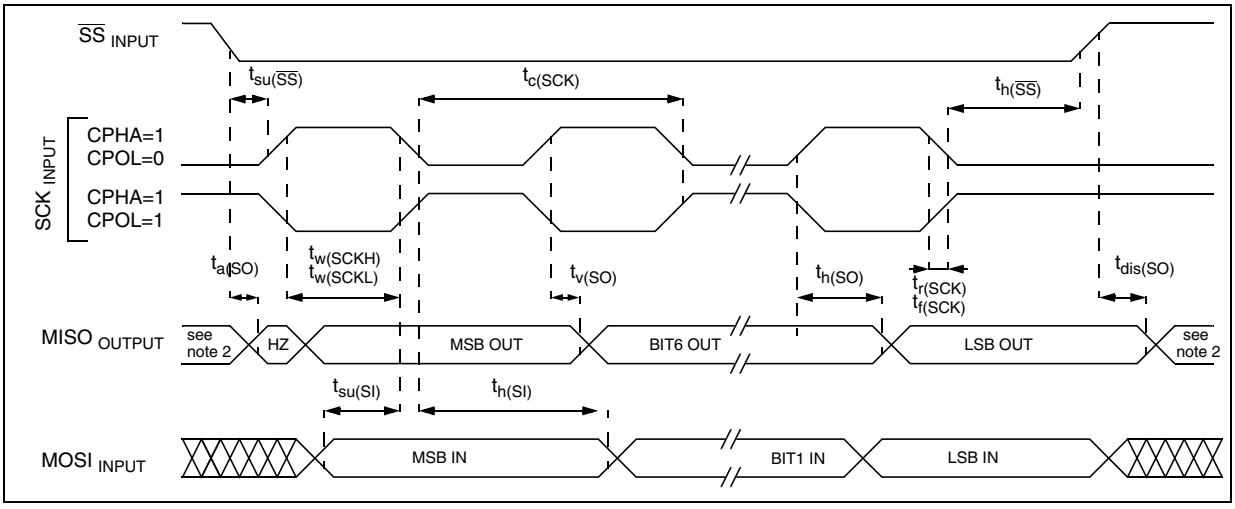
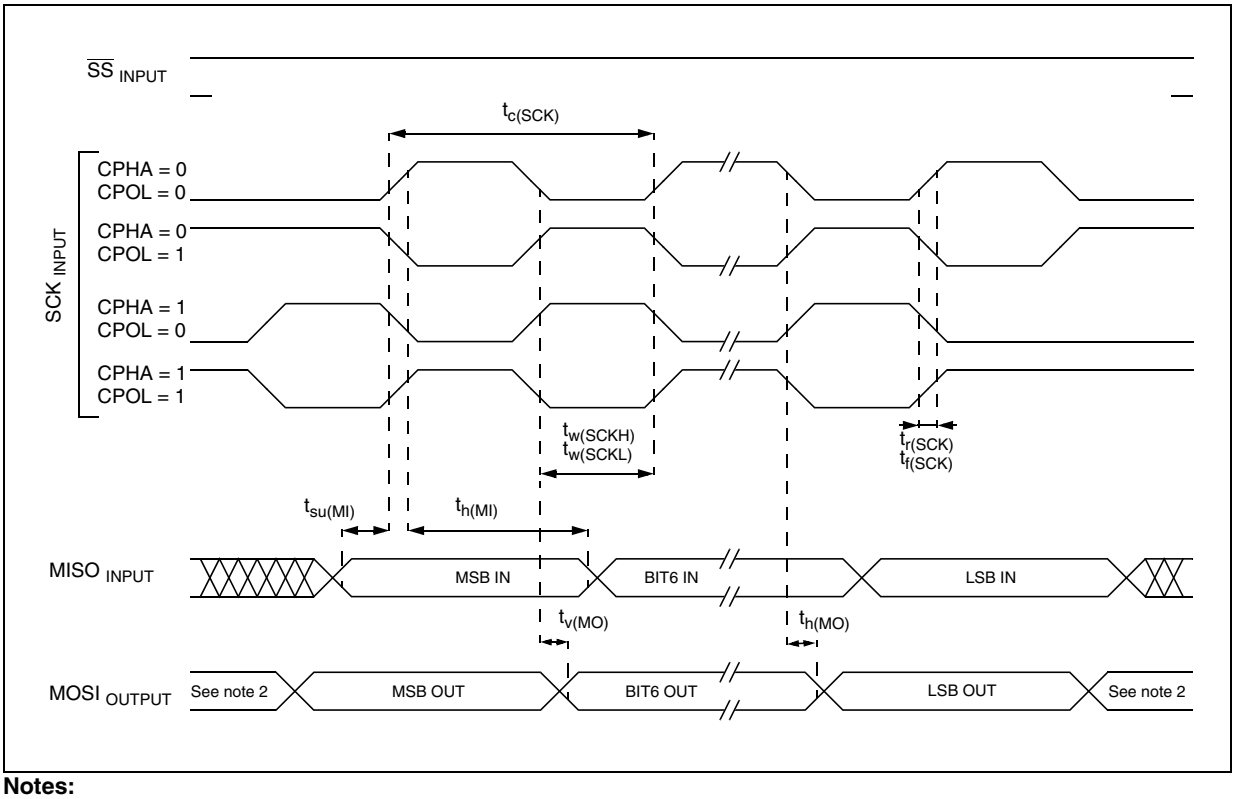


Figure 109. SPI Master Timing Diagram [1)]



Notes:

1. Measurement points are done at CMOS levels: $0.3xV_{DD}$ and $0.7xV_{DD}$.

2. When no communication is on-going the data output line of the SPI (MOSI in master mode, MISO in slave mode) has its alternate function capability released. In this case, the pin status depends of the I/O port configuration.

ADC CHARACTERISTICS (Cont'd)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{DD(AMP)}$	Amplifier operating voltage		3.6		5.5	V
V_{IN}	Amplifier input voltage[4]	V_{DD}=3.6V	0		350	mV
		V_{DD}=5V	0		500	
V_{OFFSET} [1]	Amplifier output offset voltage[5]	V_{DD}=5V		200		mV
V_{STEP} [1]	Step size for monotonicity[3]	V_{DD}=3.6V	3.5			mV
		V_{DD}=5V	4.89			
Linearity [1]	Output Voltage Response		Linear			
Gain factor [1]	Amplified Analog input Gain[2]			8		
Vmax [1]	Output Linearity Max Voltage	V_{INmax} = 430mV, V_{DD}=5V		3.65		V
Vmin [1]	Output Linearity Min Voltage			200		mV

Notes:

1. Data based on characterization results over the whole temperature range, not tested in production.

2. For precise conversion results it is recommended to calibrate the amplifier at the following two points:

– offset at V_{INmin} = 0V

– gain at full scale (for example V_{IN}=430mV)

3. Monotonicity guaranteed if V_{IN} increases or decreases in steps of min. 5mV.

4. Please refer to the Application Note AN1830 for details of TE% vs Vin.

5. Refer to the offset variation in temperature below

Amplifier output offset variation

The offset is quite sensitive to temperature variations. In order to ensure a good reliability in measurements, the offset must be recalibrated periodically i.e. during power on or whenever the device is reset depending on the customer application and during temperature variation. The table below gives the typical offset variation over temperature:

Typical Offset Variation (LSB)				UNIT
-45	-20	+25	+90	°C
-12	-7	-	+13	LSB

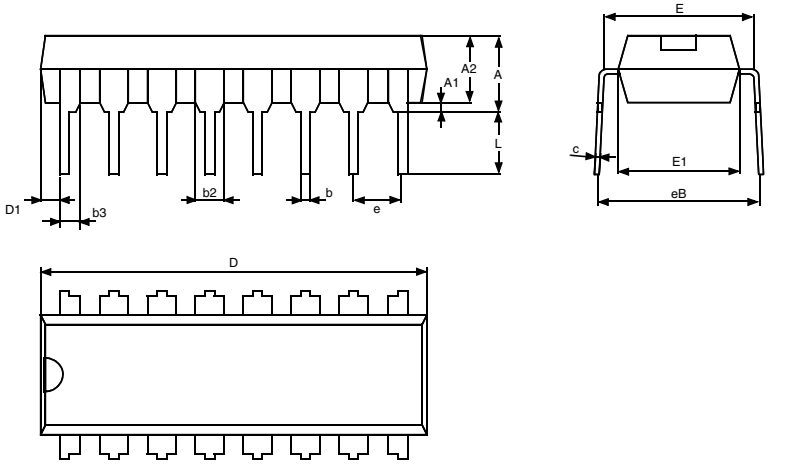
14 PACKAGE CHARACTERISTICS

In order to meet environmental requirements, ST offers these devices in ECOPACK® packages. These packages have a Lead-free second level interconnect. The category of second Level Interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label.

ECOPACK is an ST trademark. ECOPACK specifications are available at: www.st.com.

14.1 PACKAGE MECHANICAL DATA

Figure 113. 16-Pin Plastic Dual In-Line Package, 300-mil Width



Dim.	mm			inches[1]		
	Min	Typ	Max	Min	Typ	Max
A			5.33			0.210
A1	0.38			0.015		
A2	2.92	3.30	4.95	0.115	0.130	0.195
b	0.36	0.46	0.56	0.014	0.018	0.022
b2	1.14	1.52	1.78	0.045	0.060	0.070
b3	0.76	0.99	1.14	0.030	0.039	0.045
c	0.20	0.25	0.36	0.008	0.010	0.014
D	18.67	19.18	19.69	0.735	0.755	0.775
D1	0.13			0.005		
e		2.54			0.100	
E	7.62	7.87	8.26	0.300	0.310	0.325
E1	6.10	6.35	7.11	0.240	0.250	0.280
L	2.92	3.30	3.81	0.115	0.130	0.150
eB			10.92			0.430
	Number of Pins					
N	16					

Note . Values in inches are converted from mm and rounded to 3 decimal digits.

OPTION BYTES (Cont'd)

OPTION BYTE 1

OPT7 = **PLLx4x8** *PLL Factor selection.*
0: PLLx4
1: PLLx8

OPT6 = **PLLOFF** *PLL disable.*
0: PLL enabled
1: PLL disabled (by-passed)

OPT5 = **PLL32OFF** *32MHz PLL disable.*
0: PLL32 enabled
1: PLL32 disabled (by-passed)

OPT4 = **OSC** *RC Oscillator selection*
0: RC oscillator on
1: RC oscillator off

Notes:

– 1% RC oscillator available on ST7LITE15B and ST7LITE19B devices only

– If the RC oscillator is selected, then to improve clock stability and frequency accuracy, it is recommended to place a decoupling capacitor, typically 100nF, between the V_{DD} and V_{SS} pins as close as possible to the ST7 device.

OPT3:2 = **LVD[1:0]** *Low voltage detection selection*
These option bits enable the LVD block with a selected threshold as shown in Table 26.

Table 26. LVD Threshold Configuration

Configuration	LVD1	LVD0
LVD Off	1	1
Highest Voltage Threshold (~4.1V)	1	0
Medium Voltage Threshold (~3.5V)	0	1
Lowest Voltage Threshold (~2.8V)	0	0

OPT1 = **WDG SW** *Hardware or Software Watchdog*
This option bit selects the watchdog type.
0: Hardware (watchdog always enabled)
1: Software (watchdog to be enabled by software)

OPT0 = **WDG HALT** *Watchdog Reset on Halt*
This option bit determines if a RESET is generated when entering HALT mode while the Watchdog is active.
0: No Reset generation when entering Halt mode
1: Reset generation when entering Halt mode

Table 27. List of valid option combinations

Operating conditions				Option Bits		
V_{DD} range	Clock Source	PLL	Typ f_{CPU}	OSC	PLLOFF	PLLx4x8
2.7V - 3.3V	Internal RC 1%[1)]	off	1MHz @3.3V	0	1	1
		x4	4MHz @3.3V	0	0	0
		x8	-	-	-	-
	External clock	off	0-4MHz	1	1	1
		x4	4MHz	1	0	0
		x8	-	-	-	-
3.3V - 5.5V	Internal RC 1% [1)]	off	1MHz @5V	0	1	1
		x4	-	-	-	-
		x8	8MHz @5V	0	0	1
	External clock	off	0-8MHz	1	1	1
		x4	-	-	-	-
		x8	8 MHz	1	0	1

Note 1: Configuration available on ST7LITE15B and ST7LITE19B devices only

Note: see Clock Management Block diagram in Figure 14

Table 29. ST7 Application Notes

IDENTIFICATION	DESCRIPTION
AN1039	ST7 MATH UTILITY ROUTINES
AN1064	WRITING OPTIMIZED HIWARE C LANGUAGE FOR ST7
AN1071	HALF DUPLEX USB-TO-SERIAL BRIDGE USING THE ST72611 USB MICROCONTROLLER
AN1106	TRANSLATING ASSEMBLY CODE FROM HC05 TO ST7
AN1179	PROGRAMMING ST7 FLASH MICROCONTROLLERS IN REMOTE ISP MODE (IN-SITU PROGRAMMING)
AN1446	USING THE ST72521 EMULATOR TO DEBUG A ST72324 TARGET APPLICATION
AN1477	EMULATED DATA EEPROM WITH XFLASH MEMORY
AN1478	PORTING AN ST7 PANTA PROJECT TO CODEWARRIOR IDE
AN1527	DEVELOPING A USB SMARTCARD READER WITH ST7SCR
AN1575	ON-BOARD PROGRAMMING METHODS FOR XFLASH AND HDFLASH ST7 MCUS
AN1576	IN-APPLICATION PROGRAMMING (IAP) DRIVERS FOR ST7 HDFLASH OR XFLASH MCUS
AN1577	DEVICE FIRMWARE UPGRADE (DFU) IMPLEMENTATION FOR ST7 USB APPLICATIONS
AN1601	SOFTWARE IMPLEMENTATION FOR ST7DALI-EVAL
AN1603	USING THE ST7 USB DEVICE FIRMWARE UPGRADE DEVELOPMENT KIT (DFU-DK)
AN1635	ST7 CUSTOMER ROM CODE RELEASE INFORMATION
AN1754	DATA LOGGING PROGRAM FOR TESTING ST7 APPLICATIONS VIA ICC
AN1796	FIELD UPDATES FOR FLASH BASED ST7 APPLICATIONS USING A PC COMM PORT
AN1900	HARDWARE IMPLEMENTATION FOR ST7DALI-EVAL
AN1904	ST7MC THREE-PHASE AC INDUCTION MOTOR CONTROL SOFTWARE LIBRARY
AN1905	ST7MC THREE-PHASE BLDC MOTOR CONTROL SOFTWARE LIBRARY
SYSTEM OPTIMIZATION	
AN1711	SOFTWARE TECHNIQUES FOR COMPENSATING ST7 ADC ERRORS
AN1827	IMPLEMENTATION OF SIGMA-DELTA ADC WITH ST7FLITE05/09
AN2009	PWM MANAGEMENT FOR 3-PHASE BLDC MOTOR DRIVES USING THE ST7FMC
AN2030	BACK EMF DETECTION DURING PWM ON TIME BY ST7MC