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Details

Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	17
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 12x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	20-TSSOP (0.173", 4.40mm Width)
Supplier Device Package	20-TSSOP
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc9s08sh32ctj

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4.5.5 Access Errors

An access error occurs whenever the command execution protocol is violated.

Any of the following specific actions will cause the access error flag (FACCERR) in FSTAT to be set. FACCERR must be cleared by writing a 1 to FACCERR in FSTAT before any command can be processed.

- Writing to a FLASH address before the internal FLASH clock frequency has been set by writing to the FCDIV register
- Writing to a FLASH address while FCBEF is not set (A new command cannot be started until the command buffer is empty.)
- Writing a second time to a FLASH address before launching the previous command (There is only one write to FLASH for every command.)
- Writing a second time to FCMD before launching the previous command (There is only one write to FCMD for every command.)
- Writing to any FLASH control register other than FCMD after writing to a FLASH address
- Writing any command code other than the five allowed codes (0x05, 0x20, 0x25, 0x40, or 0x41) to FCMD
- Writing any FLASH control register other than the write to FSTAT (to clear FCBEF and launch the command) after writing the command to FCMD
- The MCU enters stop mode while a program or erase command is in progress (The command is aborted.)
- Writing the byte program, burst program, or page erase command code (0x20, 0x25, or 0x40) with a background debug command while the MCU is secured (The background debug controller can only do blank check and mass erase commands when the MCU is secure.)
- Writing 0 to FCBEF to cancel a partial command

4.5.6 FLASH Block Protection

The block protection feature prevents the protected region of FLASH from program or erase changes. Block protection is controlled through the FLASH protection register (FPROT). When enabled, block protection begins at any 512 byte boundary below the last address of FLASH, 0xFFFF. (See [Section 4.7.4, “FLASH Protection Register \(FPROT and NVPROT\)”](#)).

After exit from reset, FPROT is loaded with the contents of the NVPROT location, which is in the nonvolatile register block of the FLASH memory. FPROT cannot be changed directly from application software so a runaway program cannot alter the block protection settings. Because NVPROT is within the last 512 bytes of FLASH, if any amount of memory is protected, NVPROT is itself protected and cannot be altered (intentionally or unintentionally) by the application software. FPROT can be written through background debug commands, which allows a way to erase and reprogram a protected FLASH memory.

The block protection mechanism is illustrated in [Figure 4-4](#). The FPS bits are used as the upper bits of the last address of unprotected memory. This address is formed by concatenating FPS7:FPS1 with logic 1 bits as shown. For example, to protect the last 1536 bytes of memory (addresses 0xFA00 through 0xFFFF), the FPS bits must be set to 1111 100, which results in the value 0xF9FF as the last address of unprotected memory. In addition to programming the FPS bits to the appropriate value, FPDIS (bit 0 of NVPROT)

Table 4-7. FLASH Clock Divider Settings

f_{Bus}	PRDIV8 (Binary)	DIV (Decimal)	f_{CLK}	Program/Erase Timing Pulse (5 μs Min, 6.7 μs Max)
20 MHz	1	12	192.3 kHz	5.2 μs
10 MHz	0	49	200 kHz	5 μs
8 MHz	0	39	200 kHz	5 μs
4 MHz	0	19	200 kHz	5 μs
2 MHz	0	9	200 kHz	5 μs
1 MHz	0	4	200 kHz	5 μs
200 kHz	0	0	200 kHz	5 μs
150 kHz	0	0	150 kHz	6.7 μs

4.7.2 FLASH Options Register (FOPT and NVOPT)

During reset, the contents of the nonvolatile location NVOPT are copied from FLASH into FOPT. To change the value in this register, erase and reprogram the NVOPT location in FLASH memory as usual and then issue a new MCU reset.

	7	6	5	4	3	2	1	0
R	KEYEN	FNORED	—	—	—	—	SEC01	SEC00
W								

Reset This register is loaded from nonvolatile location NVOPT during reset.

= Unimplemented or Reserved

Figure 4-6. FLASH Options Register (FOPT)

Table 4-8. FOPT Register Field Descriptions

Field	Description
7 KEYEN	Backdoor Key Mechanism Enable — When this bit is 0, the backdoor key mechanism cannot be used to disengage security. The backdoor key mechanism is accessible only from user (secured) firmware. BDM commands cannot be used to write key comparison values that would unlock the backdoor key. For more detailed information about the backdoor key mechanism, refer to Section 4.6, “Security.” 0 No backdoor key access allowed. 1 If user firmware writes an 8-byte value that matches the nonvolatile backdoor key (NVBACKKEY through NVBACKKEY+7 in that order), security is temporarily disengaged until the next MCU reset.
6 FNORED	Vector Redirection Disable — When this bit is 1, then vector redirection is disabled. 0 Vector redirection enabled. 1 Vector redirection disabled.
1:0 SEC0[1:0]	Security State Code — This 2-bit field determines the security state of the MCU as shown in Table 4-9 . When the MCU is secure, the contents of RAM and FLASH memory cannot be accessed by instructions from any unsecured source including the background debug interface. SEC01:SEC00 changes to 1:0 after successful backdoor key entry or a successful blank check of FLASH. For more detailed information about security, refer to Section 4.6, “Security.”

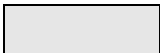
Table 5-4. SRS Register Field Descriptions

Field	Description
3 ILAD	Illegal Address — Reset was caused by an attempt to access either data or an instruction at an unimplemented memory address. 0 Reset not caused by an illegal address 1 Reset caused by an illegal address
1 LVD	Low Voltage Detect — If the LVDRE bit is set and the supply drops below the LVD trip voltage, an LVD reset will occur. This bit is also set by POR. 0 Reset not caused by LVD trip or POR. 1 Reset caused by LVD trip or POR.

5.7.3 System Background Debug Force Reset Register (SBDFR)

This high page register contains a single write-only control bit. A serial background command such as WRITE_BYTE must be used to write to SBDFR. Attempts to write this register from a user program are ignored. Reads always return 0x00.

	7	6	5	4	3	2	1	0
R	0	0	0	0	0	0	0	0
W								BDFR ¹
Reset:	0	0	0	0	0	0	0	0

 = Unimplemented or Reserved

¹ BDFR is writable only through serial background debug commands, not from user programs.

Figure 5-4. System Background Debug Force Reset Register (SBDFR)

Table 5-5. SBDFR Register Field Descriptions

Field	Description
0 BDFR	Background Debug Force Reset — A serial background command such as WRITE_BYTE can be used to allow an external debug host to force a target system reset. Writing 1 to this bit forces an MCU reset. This bit cannot be written from a user program.

Chapter 6

Parallel Input/Output Control

This section explains software controls related to parallel input/output (I/O) and pin control. The MC9S08SH32 has three parallel I/O ports which include a total of 23 I/O pins and one output-only pin. See [Chapter 2, “Pins and Connections,”](#) for more information about pin assignments and external hardware considerations of these pins.

Many of these pins are shared with on-chip peripherals such as timer systems, communication systems, or pin interrupts as shown in [Table 2-1](#). The peripheral modules have priority over the general-purpose I/O functions so that when a peripheral is enabled, the I/O functions associated with the shared pins are disabled.

After reset, the shared peripheral functions are disabled and the pins are configured as inputs ($PTxDDn = 0$). The pin control functions for each pin are configured as follows: slew rate disabled ($PTxSEn = 0$), low drive strength selected ($PTxDSn = 0$), and internal pull-ups disabled ($PTxPEN = 0$).

NOTE

Not all general-purpose I/O pins are available on all packages. To avoid extra current drain from floating input pins, the user's reset initialization routine in the application program must either enable on-chip pull-up devices or change the direction of unconnected pins to outputs so the pins do not float.

6.1 Port Data and Data Direction

Reading and writing of parallel I/Os are performed through the port data registers. The direction, either input or output, is controlled through the port data direction registers. The parallel I/O port function for an individual pin is illustrated in the block diagram shown in [Figure 6-1](#).

The data direction control bit ($PTxDDn$) determines whether the output buffer for the associated pin is enabled, and also controls the source for port data register reads. The input buffer for the associated pin is always enabled unless the pin is enabled as an analog function or is an output-only pin.

When a shared digital function is enabled for a pin, the output buffer is controlled by the shared function. However, the data direction register bit will continue to control the source for reads of the port data register.

When a shared analog function is enabled for a pin, both the input and output buffers are disabled. A value of 0 is read for any port data bit where the bit is an input ($PTxDDn = 0$) and the input buffer is disabled. In general, whenever a pin is shared with both an alternate digital function and an analog function, the analog function has priority such that if both the digital and analog functions are enabled, the analog function controls the pin.

6.6.2.7 Port B Interrupt Pin Select Register (PTBPS)

	7	6	5	4	3	2	1	0
R	0	0	0	0	PTBPS3	PTBPS2	PTBPS1	PTBPS0
W								
Reset:	0	0	0	0	0	0	0	0

Figure 6-17. Port B Interrupt Pin Select Register (PTBPS)

Table 6-16. PTBPS Register Field Descriptions

Field	Description
3:0 PTBPS[3:0]	Port B Interrupt Pin Selects — Each of the PTBPSn bits enable the corresponding port B interrupt pin. 0 Pin not enabled as interrupt. 1 Pin enabled as interrupt.

6.6.2.8 Port B Interrupt Edge Select Register (PTBES)

	7	6	5	4	3	2	1	0
R	0	0	0	0	PTBES3	PTBES2	PTBES1	PTBES0
W								
Reset:	0	0	0	0	0	0	0	0

Figure 6-18. Port B Edge Select Register (PTBES)

Table 6-17. PTBES Register Field Descriptions

Field	Description
3:0 PTBES[3:0]	Port B Edge Selects — Each of the PTBESn bits serves a dual purpose by selecting the polarity of the active interrupt edge as well as selecting a pull-up or pull-down device if enabled. 0 A pull-up device is connected to the associated pin and detects falling edge/low level for interrupt generation. 1 A pull-down device is connected to the associated pin and detects rising edge/high level for interrupt generation.

Table 7-3. Opcode Map (Sheet 1 of 2)

Bit-Manipulation			Branch		Read-Modify-Write										Control			Register/Memory													
00	5	10	5	20	3	30	5	40	1	50	1	60	5	70	4	80	9	90	3	A0	2	B0	3	C0	4	D0	4	E0	3	F0	3
BRSET0	DIR	BSET0	DIR	BRA	REL	NEG	DIR	NEGA	INH	NEGX	INH	NEG	IX1	NEG	IX	RTI	INH	BGE	REL	SUB	IMM	SUB	DIR	SUB	EXT	SUB	IX2	SUB	IX1	SUB	IX
01	5	11	5	21	3	31	5	41	4	51	4	61	5	71	5	81	6	91	3	A1	2	B1	3	C1	4	D1	4	E1	3	F1	3
BRCLR0	DIR	BCLR0	DIR	BRN	REL	CBEQ	DIR	CBEQA	IMM	CBEQX	IMM	CBEQ	IX1+	CBEQ	IX+	RTS	INH	BLT	REL	CMP	IMM	CMP	DIR	CMP	EXT	CMP	IX2	CMP	IX1	CMP	IX
02	5	12	5	22	3	32	5	42	5	52	6	62	1	72	1	82	5+	92	3	A2	2	B2	3	C2	4	D2	4	E2	3	F2	3
BRSET1	DIR	BSET1	DIR	BHI	REL	LDHX	EXT	MUL	INH	DIV	INH	NSA	INH	DAA	INH	BGND	INH	BGT	REL	SBC	IMM	SBC	DIR	SBC	EXT	SBC	IX2	SBC	IX1	SBC	IX
03	5	13	5	23	3	33	5	43	1	53	1	63	5	73	4	83	11	93	3	A3	2	B3	3	C3	4	D3	4	E3	3	F3	3
BRCLR1	DIR	BCLR1	DIR	BLS	REL	COM	DIR	COMA	INH	COMX	INH	COM	IX1	COM	IX	SWI	INH	BLE	REL	CPX	IMM	CPX	DIR	CPX	EXT	CPX	IX2	CPX	IX1	CPX	IX
04	5	14	5	24	3	34	5	44	1	54	1	64	5	74	4	84	1	94	2	A4	2	B4	3	C4	4	D4	4	E4	3	F4	3
BRSET2	DIR	BSET2	DIR	BCC	REL	LSR	DIR	LSRA	INH	LSRX	INH	LSR	IX1	LSR	IX	TAP	INH	TXS	INH	AND	IMM	AND	DIR	AND	EXT	AND	IX2	AND	IX1	AND	IX
05	5	15	5	25	3	35	4	45	3	55	4	65	3	75	5	85	1	95	2	A5	2	B5	3	C5	4	D5	4	E5	3	F5	3
BRCLR2	DIR	BCLR2	DIR	BCS	REL	STHX	DIR	LDHX	IMM	LDHX	DIR	CPHX	IMM	CPHX	DIR	TPA	INH	TSX	INH	BIT	IMM	BIT	DIR	BIT	EXT	BIT	IX2	BIT	IX1	BIT	IX
06	5	16	5	26	3	36	5	46	1	56	1	66	5	76	4	86	3	96	5	A6	2	B6	3	C6	4	D6	4	E6	3	F6	3
BRSET3	DIR	BSET3	DIR	BNE	REL	ROR	DIR	RORA	INH	RORX	INH	ROR	IX1	ROR	IX	PULA	INH	STHX	EXT	LDA	IMM	LDA	DIR	LDA	EXT	LDA	IX2	LDA	IX1	LDA	IX
07	5	17	5	27	3	37	5	47	1	57	1	67	5	77	4	87	2	97	1	A7	2	B7	3	C7	4	D7	4	E7	3	F7	2
BRCLR3	DIR	BCLR3	DIR	BEQ	REL	ASR	DIR	ASRA	INH	ASRX	INH	ASR	IX1	ASR	IX	PSHA	INH	TAX	INH	AIS	IMM	STA	DIR	STA	EXT	STA	IX2	STA	IX1	STA	IX
08	5	18	5	28	3	38	5	48	1	58	1	68	5	78	4	88	3	98	1	A8	2	B8	3	C8	4	D8	4	E8	3	F8	3
BRSET4	DIR	BSET4	DIR	BHCC	REL	LSL	DIR	LSLA	INH	LSLX	INH	LSL	IX1	LSL	IX	PULX	INH	CLC	INH	EOR	IMM	EOR	DIR	EOR	EXT	EOR	IX2	EOR	IX1	EOR	IX
09	5	19	5	29	3	39	5	49	1	59	1	69	5	79	4	89	2	99	1	A9	2	B9	3	C9	4	D9	4	E9	3	F9	3
BRCLR4	DIR	BCLR4	DIR	BHCS	REL	ROL	DIR	ROLA	INH	ROLX	INH	ROL	IX1	ROL	IX	PSHX	INH	SEC	INH	ADC	IMM	ADC	DIR	ADC	EXT	ADC	IX2	ADC	IX1	ADC	IX
0A	5	1A	5	2A	3	3A	5	4A	1	5A	1	6A	5	7A	4	8A	3	9A	1	AA	2	BA	3	CA	4	DA	4	EA	3	FA	3
BRSET5	DIR	BSET5	DIR	BPL	REL	DEC	DIR	DECA	INH	DECX	INH	DEC	IX1	DEC	IX	PULH	INH	CLI	INH	ORA	IMM	ORA	DIR	ORA	EXT	ORA	IX2	ORA	IX1	ORA	IX
0B	5	1B	5	2B	3	3B	7	4B	4	5B	4	6B	7	7B	6	8B	2	9B	1	AB	2	BB	3	CB	4	DB	4	EB	3	FB	3
BRCLR5	DIR	BCLR5	DIR	BMI	REL	DBNZ	DIR	DBNZA	INH	DBNZX	INH	DBNZ	IX1	DBNZ	IX	PSHH	INH	SEI	INH	ADD	IMM	ADD	DIR	ADD	EXT	ADD	IX2	ADD	IX1	ADD	IX
0C	5	1C	5	2C	3	3C	5	4C	1	5C	1	6C	5	7C	4	8C	1	9C	1			BC	3	CC	4	DC	4	EC	3	FC	3
BRSET6	DIR	BSET6	DIR	BMC	REL	INC	DIR	INCA	INH	INCX	INH	INC	IX1	INC	IX	CLRH	INH	RSP	INH			JMP	DIR	JMP	EXT	JMP	IX2	JMP	IX1	JMP	IX
0D	5	1D	5	2D	3	3D	4	4D	1	5D	1	6D	4	7D	3			9D	1	AD	5	BD	5	CD	6	DD	6	ED	5	FD	5
BRCLR6	DIR	BCLR6	DIR	BMS	REL	TST	DIR	TSTA	INH	TSTX	INH	TST	IX1	TST	IX			NOP	INH	BSR	REL	JSR	DIR	JSR	EXT	JSR	IX2	JSR	IX1	JSR	IX
0E	5	1E	5	2E	3	3E	6	4E	5	5E	5	6E	4	7E	5	8E	2+	9E	2+	AE	2	BE	3	CE	4	DE	4	EE	3	FE	3
BRSET7	DIR	BSET7	DIR	BIL	REL	CPHX	EXT	MOV	DD	MOV	DIX+	MOV	IMD	MOV	IX+D	STOP	INH	Page 2		LDX	IMM	LDX	DIR	LDX	EXT	LDX	IX2	LDX	IX1	LDX	IX
0F	5	1F	5	2F	3	3F	5	4F	1	5F	1	6F	5	7F	4	8F	2+	9F	1	AF	2	BF	3	CF	4	DF	4	EF	3	FF	2
BRCLR7	DIR	BCLR7	DIR	BIH	REL	CLR	DIR	CLRA	INH	CLRX	INH	CLR	IX1	CLR	IX	WAIT	INH	TXA	INH	AIX	IMM	STX	DIR	STX	EXT	STX	IX2	STX	IX1	STX	IX

INH Inherent
 IMM Immediate
 DIR Direct
 EXT Extended
 DD DIR to DIR
 IX+D IX+ to DIR
 REL Relative
 IX Indexed, No Offset
 IX1 Indexed, 8-Bit Offset
 IX2 Indexed, 16-Bit Offset
 IMM to DIR
 DIR to IX+
 SP1 Stack Pointer, 8-Bit Offset
 SP2 Stack Pointer, 16-Bit Offset
 IX+ Indexed, No Offset with Post Increment
 IX1+ Indexed, 1-Byte Offset with Post Increment

Opcode in Hexadecimal F0 3
 Number of Bytes 1 SUB IX
 HCS08 Cycles Instruction Mnemonic Addressing Mode

Table 9-8. Input Clock Select

ADICLK	Selected Clock Source
00	Bus clock
01	Bus clock divided by 2
10	Alternate clock (ALTCLK)
11	Asynchronous clock (ADACK)

9.3.8 Pin Control 1 Register (APCTL1)

The pin control registers are used to disable the I/O port control of MCU pins used as analog inputs. APCTL1 is used to control the pins associated with channels 0–7 of the ADC module.

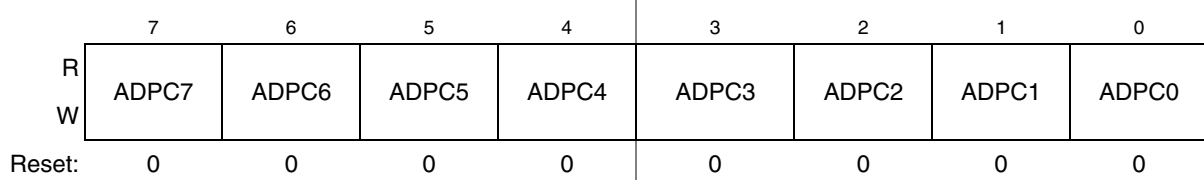


Figure 9-11. Pin Control 1 Register (APCTL1)

Table 9-9. APCTL1 Register Field Descriptions

Field	Description
7 ADPC7	ADC Pin Control 7 — ADPC7 is used to control the pin associated with channel AD7. 0 AD7 pin I/O control enabled 1 AD7 pin I/O control disabled
6 ADPC6	ADC Pin Control 6 — ADPC6 is used to control the pin associated with channel AD6. 0 AD6 pin I/O control enabled 1 AD6 pin I/O control disabled
5 ADPC5	ADC Pin Control 5 — ADPC5 is used to control the pin associated with channel AD5. 0 AD5 pin I/O control enabled 1 AD5 pin I/O control disabled
4 ADPC4	ADC Pin Control 4 — ADPC4 is used to control the pin associated with channel AD4. 0 AD4 pin I/O control enabled 1 AD4 pin I/O control disabled
3 ADPC3	ADC Pin Control 3 — ADPC3 is used to control the pin associated with channel AD3. 0 AD3 pin I/O control enabled 1 AD3 pin I/O control disabled
2 ADPC2	ADC Pin Control 2 — ADPC2 is used to control the pin associated with channel AD2. 0 AD2 pin I/O control enabled 1 AD2 pin I/O control disabled

result of the conversion is transferred to ADCRH and ADCRL upon completion of the conversion algorithm.

If the bus frequency is less than the f_{ADCK} frequency, precise sample time for continuous conversions cannot be guaranteed when short sample is enabled (ADLSMP=0). If the bus frequency is less than 1/11th of the f_{ADCK} frequency, precise sample time for continuous conversions cannot be guaranteed when long sample is enabled (ADLSMP=1).

The maximum total conversion time for different conditions is summarized in [Table 9-12](#).

Table 9-12. Total Conversion Time vs. Control Conditions

Conversion Type	ADICLK	ADLSMP	Max Total Conversion Time
Single or first continuous 8-bit	0x, 10	0	20 ADCK cycles + 5 bus clock cycles
Single or first continuous 10-bit	0x, 10	0	23 ADCK cycles + 5 bus clock cycles
Single or first continuous 8-bit	0x, 10	1	40 ADCK cycles + 5 bus clock cycles
Single or first continuous 10-bit	0x, 10	1	43 ADCK cycles + 5 bus clock cycles
Single or first continuous 8-bit	11	0	5 μ s + 20 ADCK + 5 bus clock cycles
Single or first continuous 10-bit	11	0	5 μ s + 23 ADCK + 5 bus clock cycles
Single or first continuous 8-bit	11	1	5 μ s + 40 ADCK + 5 bus clock cycles
Single or first continuous 10-bit	11	1	5 μ s + 43 ADCK + 5 bus clock cycles
Subsequent continuous 8-bit; $f_{BUS} \geq f_{ADCK}$	xx	0	17 ADCK cycles
Subsequent continuous 10-bit; $f_{BUS} \geq f_{ADCK}$	xx	0	20 ADCK cycles
Subsequent continuous 8-bit; $f_{BUS} \geq f_{ADCK}/11$	xx	1	37 ADCK cycles
Subsequent continuous 10-bit; $f_{BUS} \geq f_{ADCK}/11$	xx	1	40 ADCK cycles

The maximum total conversion time is determined by the clock source chosen and the divide ratio selected. The clock source is selectable by the ADICLK bits, and the divide ratio is specified by the ADIV bits. For example, in 10-bit mode, with the bus clock selected as the input clock source, the input clock divide-by-1 ratio selected, and a bus frequency of 8 MHz, then the conversion time for a single conversion is:

$$\text{Conversion time} = \frac{23 \text{ ADCK cyc}}{8 \text{ MHz}/1} + \frac{5 \text{ bus cyc}}{8 \text{ MHz}} = 3.5 \mu\text{s}$$

$$\text{Number of bus cycles} = 3.5 \mu\text{s} \times 8 \text{ MHz} = 28 \text{ cycles}$$

NOTE

The ADCK frequency must be between f_{ADCK} minimum and f_{ADCK} maximum to meet ADC specifications.

Table 10-7. IICS Field Descriptions

Field	Description
7 TCF	Transfer Complete Flag. This bit is set on the completion of a byte transfer. This bit is only valid during or immediately following a transfer to the IIC module or from the IIC module. The TCF bit is cleared by reading the IICD register in receive mode or writing to the IICD in transmit mode. 0 Transfer in progress 1 Transfer complete
6 IAAS	Addressed as a Slave. The IAAS bit is set when the calling address matches the programmed slave address or when the GCAEN bit is set and a general call is received. Writing the IICC register clears this bit. 0 Not addressed 1 Addressed as a slave
5 BUSY	Bus Busy. The BUSY bit indicates the status of the bus regardless of slave or master mode. The BUSY bit is set when a start signal is detected and cleared when a stop signal is detected. 0 Bus is idle 1 Bus is busy
4 ARBL	Arbitration Lost. This bit is set by hardware when the arbitration procedure is lost. The ARBL bit must be cleared by software by writing a 1 to it. 0 Standard bus operation 1 Loss of arbitration
2 SRW	Slave Read/Write. When addressed as a slave, the SRW bit indicates the value of the R/W command bit of the calling address sent to the master. 0 Slave receive, master writing to slave 1 Slave transmit, master reading from slave
1 IICIF	IIC Interrupt Flag. The IICIF bit is set when an interrupt is pending. This bit must be cleared by software, by writing a 1 to it in the interrupt routine. One of the following events can set the IICIF bit: <ul style="list-style-type: none"> One byte transfer completes Match of slave address to calling address Arbitration lost 0 No interrupt pending 1 Interrupt pending
0 RXAK	Receive Acknowledge. When the RXAK bit is low, it indicates an acknowledge signal has been received after the completion of one byte of data transmission on the bus. If the RXAK bit is high it means that no acknowledge signal is detected. 0 Acknowledge received 1 No acknowledge received

10.3.5 IIC Data I/O Register (IICD)

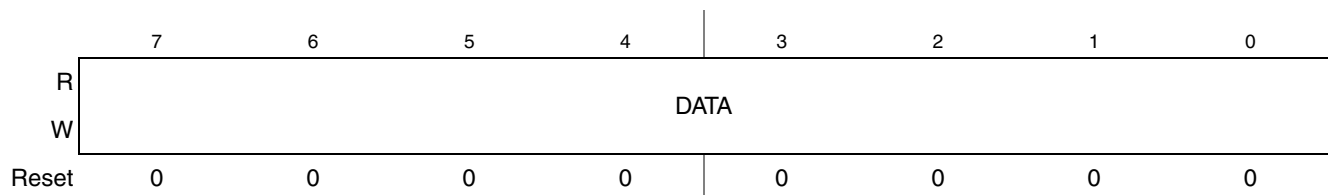


Figure 10-7. IIC Data I/O Register (IICD)

11.3.1 ICS Control Register 1 (ICSC1)

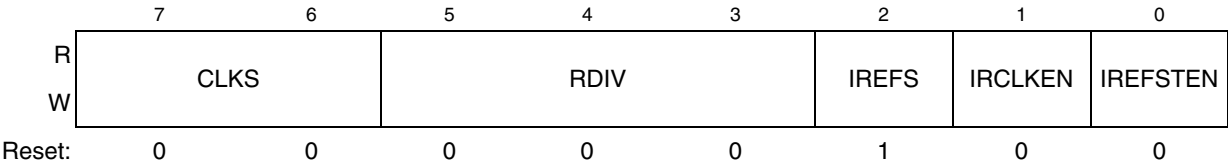


Figure 11-3. ICS Control Register 1 (ICSC1)

Table 11-2. ICS Control Register 1 Field Descriptions

Field	Description
7:6 CLKS	Clock Source Select — Selects the clock source that controls the bus frequency. The actual bus frequency depends on the value of the BDIV bits. 00 Output of FLL is selected. 01 Internal reference clock is selected. 10 External reference clock is selected. 11 Reserved, defaults to 00.
5:3 RDIV	Reference Divider — Selects the amount to divide down the FLL reference clock selected by the IREFS bits. Resulting frequency must be in the range 31.25 kHz to 39.0625 kHz. 000 Encoding 0 — Divides reference clock by 1 (reset default) 001 Encoding 1 — Divides reference clock by 2 010 Encoding 2 — Divides reference clock by 4 011 Encoding 3 — Divides reference clock by 8 100 Encoding 4 — Divides reference clock by 16 101 Encoding 5 — Divides reference clock by 32 110 Encoding 6 — Divides reference clock by 64 111 Encoding 7 — Divides reference clock by 128
2 IREFS	Internal Reference Select — The IREFS bit selects the reference clock source for the FLL. 1 Internal reference clock selected 0 External reference clock selected
1 IRCLKEN	Internal Reference Clock Enable — The IRCLKEN bit enables the internal reference clock for use as ICSIRCLK. 1 ICSIRCLK active 0 ICSIRCLK inactive
0 IREFSTEN	Internal Reference Stop Enable — The IREFSTEN bit controls whether or not the internal reference clock remains enabled when the ICS enters stop mode. 1 Internal reference clock stays enabled in stop if IRCLKEN is set or if ICS is in FEI, FBI, or FBILP mode before entering stop 0 Internal reference clock is disabled in stop

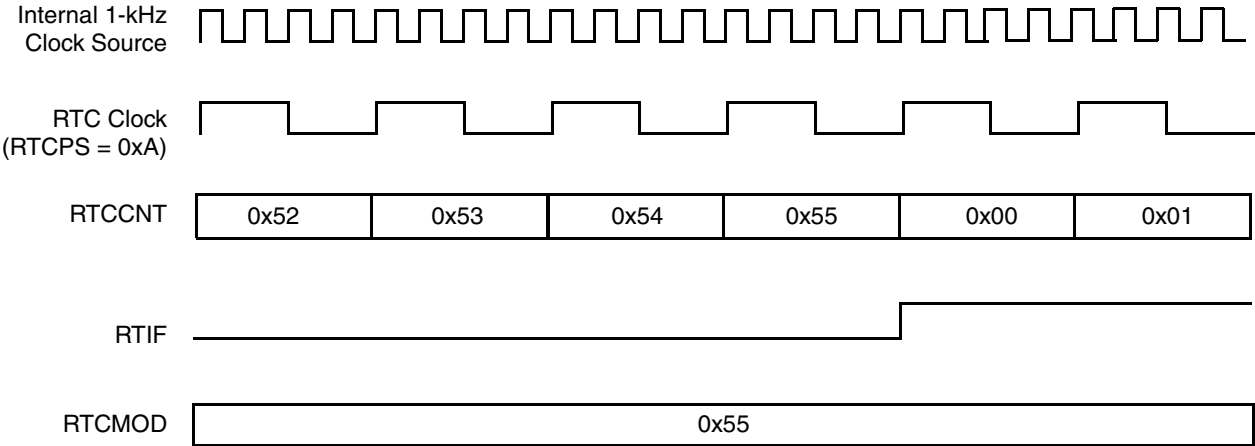


Figure 13-6. RTC Counter Overflow Example

In the example of [Figure 13-6](#), the selected clock source is the 1-kHz internal oscillator clock source. The prescaler (RTCPs) is set to 0xA or divide-by-4. The modulo value in the RTCMOD register is set to 0x55. When the counter, RTCCNT, reaches the modulo value of 0x55, the counter overflows to 0x00 and continues counting. The real-time interrupt flag, RTIF, sets when the counter value changes from 0x55 to 0x00. A real-time interrupt is generated when RTIF is set, if RTIE is set.

13.5 Initialization/Application Information

This section provides example code to give some basic direction to a user on how to initialize and configure the RTC module. The example software is implemented in C language.

The example below shows how to implement time of day with the RTC using the 1-kHz clock source to achieve the lowest possible power consumption. Because the 1-kHz clock source is not as accurate as a crystal, software can be added for any adjustments. For accuracy without adjustments at the expense of additional power consumption, the external clock (ERCLK) or the internal clock (IRCLK) can be selected with appropriate prescaler and modulo values.

```
/* Initialize the elapsed time counters */
Seconds = 0;
Minutes = 0;
Hours = 0;
Days=0;

/* Configure RTC to interrupt every 1 second from 1-kHz clock source */
RTCMOD.byte = 0x00;
RTCSC.byte = 0x1F;

/*****
Function Name : RTC_ISR
Notes : Interrupt service routine for RTC module.
*****/
```

Table 14-5. SCIS1 Field Descriptions

Field	Description
7 TDRE	Transmit Data Register Empty Flag — TDRE is set out of reset and when a transmit data value transfers from the transmit data buffer to the transmit shifter, leaving room for a new character in the buffer. To clear TDRE, read SCIS1 with TDRE = 1 and then write to the SCI data register (SCID). 0 Transmit data register (buffer) full. 1 Transmit data register (buffer) empty.
6 TC	Transmission Complete Flag — TC is set out of reset and when TDRE = 1 and no data, preamble, or break character is being transmitted. 0 Transmitter active (sending data, a preamble, or a break). 1 Transmitter idle (transmission activity complete). TC is cleared automatically by reading SCIS1 with TC = 1 and then doing one of the following three things: <ul style="list-style-type: none"> Write to the SCI data register (SCID) to transmit new data Queue a preamble by changing TE from 0 to 1 Queue a break character by writing 1 to SBK in SCIC2
5 RDRF	Receive Data Register Full Flag — RDRF becomes set when a character transfers from the receive shifter into the receive data register (SCID). To clear RDRF, read SCIS1 with RDRF = 1 and then read the SCI data register (SCID). 0 Receive data register empty. 1 Receive data register full.
4 IDLE	Idle Line Flag — IDLE is set when the SCI receive line becomes idle for a full character time after a period of activity. When ILT = 0, the receiver starts counting idle bit times after the start bit. So if the receive character is all 1s, these bit times and the stop bit time count toward the full character time of logic high (10 or 11 bit times depending on the M control bit) needed for the receiver to detect an idle line. When ILT = 1, the receiver doesn't start counting idle bit times until after the stop bit. So the stop bit and any logic high bit times at the end of the previous character do not count toward the full character time of logic high needed for the receiver to detect an idle line. To clear IDLE, read SCIS1 with IDLE = 1 and then read the SCI data register (SCID). After IDLE has been cleared, it cannot become set again until after a new character has been received and RDRF has been set. IDLE will get set only once even if the receive line remains idle for an extended period. 0 No idle line detected. 1 Idle line was detected.
3 OR	Receiver Overrun Flag — OR is set when a new serial character is ready to be transferred to the receive data register (buffer), but the previously received character has not been read from SCID yet. In this case, the new character (and all associated error information) is lost because there is no room to move it into SCID. To clear OR, read SCIS1 with OR = 1 and then read the SCI data register (SCID). 0 No overrun. 1 Receive overrun (new SCI data lost).
2 NF	Noise Flag — The advanced sampling technique used in the receiver takes seven samples during the start bit and three samples in each data bit and the stop bit. If any of these samples disagrees with the rest of the samples within any bit time in the frame, the flag NF will be set at the same time as the flag RDRF gets set for the character. To clear NF, read SCIS1 and then read the SCI data register (SCID). 0 No noise detected. 1 Noise detected in the received character in SCID.

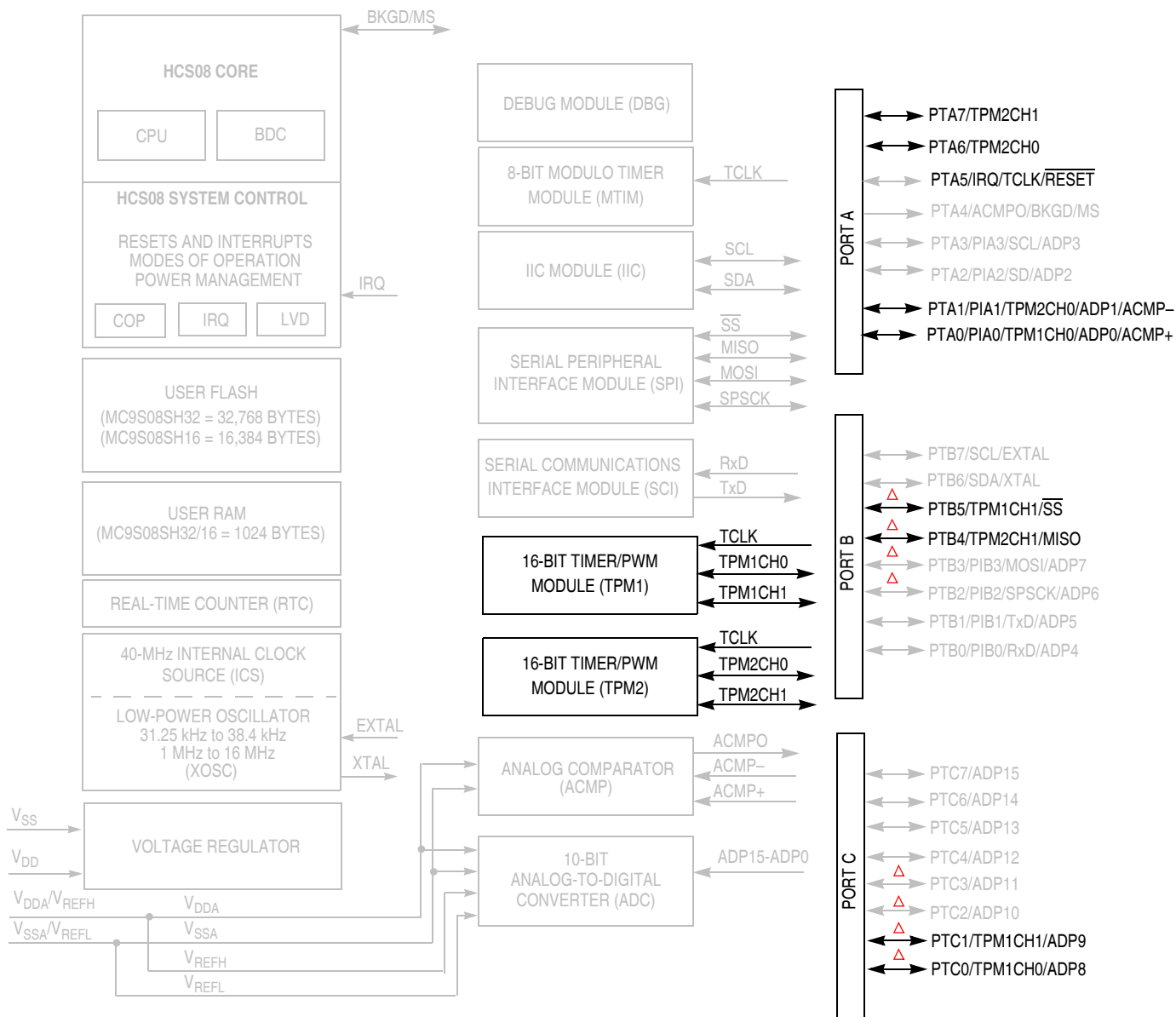


Figure 16-1. MC9S08SH32 Series Block Diagram Highlighting TPM Block and Pins

The TPM channels are programmable independently as input capture, output compare, or edge-aligned PWM channels. Alternately, the TPM can be configured to produce CPWM outputs on all channels. When the TPM is configured for CPWMs, the counter operates as an up/down counter; input capture, output compare, and EPWM functions are not practical.

If a channel is configured as input capture, an internal pullup device may be enabled for that channel. The details of how a module interacts with pin controls depends upon the chip implementation because the I/O pins and associated general purpose I/O controls are not part of the module. Refer to the discussion of the I/O port logic in a full-chip specification.

Because center-aligned PWMs are usually used to drive 3-phase AC-induction motors and brushless DC motors, they are typically used in sets of three or six channels.

16.2 Signal Description

Table 16-2 shows the user-accessible signals for the TPM. The number of channels may be varied from one to eight. When an external clock is included, it can be shared with the same pin as any TPM channel; however, it could be connected to a separate input pin. Refer to the I/O pin descriptions in full-chip specification for the specific chip implementation.

Table 16-2. Signal Properties

Name	Function
EXTCLK ¹	External clock source which may be selected to drive the TPM counter.
TPMxCHn ²	I/O pin associated with TPM channel n

¹ When preset, this signal can share any channel pin; however depending upon full-chip implementation, this signal could be connected to a separate external pin.

² n=channel number (1 to 8)

Refer to documentation for the full-chip for details about reset states, port connections, and whether there is any pullup device on these pins.

TPM channel pins can be associated with general purpose I/O pins and have passive pullup devices which can be enabled with a control bit when the TPM or general purpose I/O controls have configured the associated pin as an input. When no TPM function is enabled to use a corresponding pin, the pin reverts to being controlled by general purpose I/O controls, including the port-data and data-direction registers. Immediately after reset, no TPM functions are enabled, so all associated pins revert to general purpose I/O control.

16.2.1 Detailed Signal Descriptions

This section describes each user-accessible pin signal in detail. Although Table 16-2 grouped all channel pins together, any TPM pin can be shared with the external clock source signal. Since I/O pin logic is not part of the TPM, refer to full-chip documentation for a specific derivative for more details about the interaction of TPM pin functions and general purpose I/O controls including port data, data direction, and pullup controls.

16.6.2.1.2 Center-Aligned PWM Case

When CPWMS=1, TOF gets set when the timer counter changes direction from up-counting to down-counting at the end of the terminal count (the value in the modulo register). In this case the TOF corresponds to the end of a PWM period.

16.6.2.2 Channel Event Interrupt Description

The meaning of channel interrupts depends on the channel's current mode (input-capture, output-compare, edge-aligned PWM, or center-aligned PWM).

16.6.2.2.1 Input Capture Events

When a channel is configured as an input capture channel, the ELSnB:ELSnA control bits select no edge (off), rising edges, falling edges or any edge as the edge which triggers an input capture event. When the selected edge is detected, the interrupt flag is set. The flag is cleared by the two-step sequence described in [Section 16.6.2, "Description of Interrupt Operation."](#)

16.6.2.2.2 Output Compare Events

When a channel is configured as an output compare channel, the interrupt flag is set each time the main timer counter matches the 16-bit value in the channel value register. The flag is cleared by the two-step sequence described [Section 16.6.2, "Description of Interrupt Operation."](#)

16.6.2.2.3 PWM End-of-Duty-Cycle Events

For channels configured for PWM operation there are two possibilities. When the channel is configured for edge-aligned PWM, the channel flag gets set when the timer counter matches the channel value register which marks the end of the active duty cycle period. When the channel is configured for center-aligned PWM, the timer count matches the channel value register twice during each PWM cycle. In this CPWM case, the channel flag is set at the start and at the end of the active duty cycle period which are the times when the timer counter matches the channel value register. The flag is cleared by the two-step sequence described [Section 16.6.2, "Description of Interrupt Operation."](#)

17.4.3.8 Debug Trigger Register (DBGT)

This register can be read any time, but may be written only if ARM = 0, except bits 4 and 5 are hard-wired to 0s.

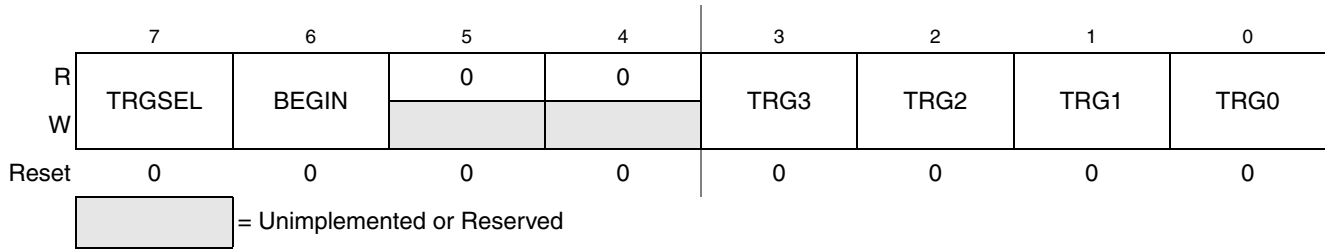


Figure 17-8. Debug Trigger Register (DBGT)

Table 17-5. DBGT Register Field Descriptions

Field	Description
7 TRGSEL	Trigger Type — Controls whether the match outputs from comparators A and B are qualified with the opcode tracking logic in the debug module. If TRGSEL is set, a match signal from comparator A or B must propagate through the opcode tracking logic and a trigger event is only signalled to the FIFO logic if the opcode at the match address is actually executed. 0 Trigger on access to compare address (force) 1 Trigger if opcode at compare address is executed (tag)
6 BEGIN	Begin/End Trigger Select — Controls whether the FIFO starts filling at a trigger or fills in a circular manner until a trigger ends the capture of information. In event-only trigger modes, this bit is ignored and all debug runs are assumed to be begin traces. 0 Data stored in FIFO until trigger (end trace) 1 Trigger initiates data storage (begin trace)
3:0 TRG[3:0]	Select Trigger Mode — Selects one of nine triggering modes, as described below. 0000 A-only 0001 A OR B 0010 A Then B 0011 Event-only B (store data) 0100 A then event-only B (store data) 0101 A AND B data (full mode) 0110 A AND NOT B data (full mode) 0111 Inside range: $A \leq \text{address} \leq B$ 1000 Outside range: $\text{address} < A$ or $\text{address} > B$ 1001 – 1111 (No trigger)

A.6 DC Characteristics

This section includes information about power supply requirements and I/O pin characteristics.

Table A-6. DC Characteristics

#	C	Characteristic	Symbol	Condition	Min	Typ ¹	Max	Unit
1	—	Operating Voltage	V _{DD}	—	2.7	—	5.5	V
2	C	Output high voltage	V _{OH}	5 V, I _{Load} = −4 mA	V _{DD} − 1.5	—	—	V
	P			5 V, I _{Load} = −2 mA	V _{DD} − 0.8	—	—	V
	C			3 V, I _{Load} = −1 mA	V _{DD} − 0.8	—	—	V
	C			5 V, I _{Load} = −20 mA	V _{DD} − 1.5	—	—	V
	P	All I/O pins, high-drive strength		5 V, I _{Load} = −10 mA	V _{DD} − 0.8	—	—	V
	C			3 V, I _{Load} = −5 mA	V _{DD} − 0.8	—	—	V
3	D	Output high current Max total I _{OH} for all ports	I _{OHT}	V _{OUT} < V _{DD}	0	—	−100	mA
4	C	All I/O pins low-drive strength	V _{OL}	5 V, I _{Load} = 4 mA	—	—	1.5	V
	P			5 V, I _{Load} = 2 mA	—	—	0.8	V
	C	Output low voltage		3 V, I _{Load} = 1 mA	—	—	0.8	V
	C			5 V, I _{Load} = 20 mA	—	—	1.5	V
	P			5 V, I _{Load} = 10 mA	—	—	0.8	V
	C			3 V, I _{Load} = 5 mA	—	—	0.8	V
5	D	Output low current Max total I _{OL} for all ports	I _{OLT}	V _{OUT} > V _{SS}	0	—	100	mA
6	P	Input high voltage; all digital inputs	V _{IH}	5V	0.65 x V _{DD}	—	—	V
	C			3V	0.7 x V _{DD}	—	—	V
7	P	Input low voltage; all digital inputs	V _{IL}	5V	—	—	0.35 x V _{DD}	V
	C			3V	—	—	0.35 x V _{DD}	V
8	C	Input hysteresis	V _{hys}	—	0.06 x V _{DD}	—	—	V
9	P	Input leakage current (per pin)	I _{In}	V _{In} = V _{DD} or V _{SS}	—	—	1	μA
10	P	Hi-Z (off-state) leakage current (per pin) input/output port pins	I _{OZ}	V _{In} = V _{DD} or V _{SS} ; temperature	—	—	1	μA
		PTA5/IRQ/TCLK/RESET, PTB6/SDA/XTAL pins		V _{In} = V _{DD} or V _{SS}	—	—	2	μA
11	P	Pullup or Pulldown ² resistors; when enabled I/O pins	R _{PU} , R _{PD}	—	17	37	52	kΩ
	C	PTA5/IRQ/TCLK/RESET ³	R _{PU}	—	17	37	52	kΩ

Table A-12. ADC Characteristics (continued)

#	Characteristic	Conditions	C	Symb	Min	Typ ¹	Max	Unit	Comment
	Full-scale error	28-pin packages only							
		10-bit mode	T	E _{FS}	0	±0.5	±1	LSB ²	
		8-bit mode			0	±0.5	±0.5	LSB ²	
		20-pin packages							
		10-bit mode	T	E _{FS}	0	±1.0	±1.5	LSB ²	
		8-bit mode			0	±0.5	±0.5	LSB ²	
		16-pin packages							
		10-bit mode	T	E _{FS}	0	±1.0	±1.5	LSB ²	
		8-bit mode			0	±0.5	±0.5	LSB ²	
	Quantization error	10-bit mode	D	E _Q	—	—	±0.5	LSB ²	
		8-bit mode			—	—	±0.5	LSB ²	
	Input leakage error	10-bit mode	D	E _{IL}	0	±0.2	±2.5	LSB ²	Pad leakage ³ * R _{AS}
		8-bit mode			0	±0.1	±1	LSB ²	
	Temp sensor slope	-40°C to 25°C	D	m	—	3.26 6	—	mV/°C	
		25°C to 125°C			—	3.63 8	—	mV/°C	
	Temp sensor voltage	25°C	D	V _{TEMP} 25	—	1.39 6	—	V	

¹ Typical values assume V_{DD} = 5.0 V, Temp = 25°C, f_{ADCK} = 1.0 MHz unless otherwise stated. Typical values are for reference only and are not tested in production.

² 1 LSB = (V_{REFH} - V_{REFL})/2^N

³ Based on input pad leakage current. Refer to pad electricals.

A.12 AC Characteristics

This section describes ac timing characteristics for each peripheral system.

A.12.1 Control Timing

Table A-13. Control Timing

Num	C	Rating		Symbol	Min	Typ ¹	Max	Unit
1	D	Bus frequency (t _{cyc} = 1/f _{Bus}) -40 °C to 125 °C		f _{Bus}	dc	—	20	MHz
2	D	Internal low power oscillator period -40 °C to 125 °C		t _{LPO}	700		1500	μs
3	D	External reset pulse width ²		t _{extrst}	100		—	ns
4	D	Reset low drive ³		t _{rstdrv}	66 x t _{cyc}		—	ns
5	D		IRQ pulse width Asynchronous path ² Synchronous path ⁴	t _{ILIH} , t _{IHIL}	100 1.5 x t _{cyc}	—	—	ns
6	D	Pin interrupt pulse width Asynchronous path ² Synchronous path ⁴		t _{ILIH} , t _{IHIL}	100 1.5 x t _{cyc}	—	—	ns
7	C	Port rise and fall time — Low output drive (PTxDS = 0) (load = 50 pF) ⁵						
		Slew rate control disabled (PTxSE = 0)		t _{Rise} , t _{Fall}	—	40	—	ns
		Slew rate control enabled (PTxSE = 1)			—	75	—	
		Port rise and fall time — High output drive (PTxDS = 1) (load = 50 pF) ⁵						
		Slew rate control disabled (PTxSE = 0)		t _{Rise} , t _{Fall}	—	11	—	ns
		Slew rate control enabled (PTxSE = 1)		t _{Rise} , t _{Fall}	—	35	—	

¹ Typical values are based on characterization data at $V_{DD} = 5.0V$, 25°C unless otherwise stated.

² This is the shortest pulse that is guaranteed to be recognized as a reset pin request.

³ When any reset is initiated, internal circuitry drives the reset pin low for about 66 cycles of t_{cyc} . After POR reset, the bus clock frequency changes to the untrimmed DCO frequency ($f_{reset} = (f_{dco_ut})/4$) because TRIM is reset to 0x80 and FTRIM is reset to 0, and there is an extra divide-by-two because BDIV is reset to 0:1. After other resets trim stays at the pre-reset value.

⁴ This is the minimum pulse width that is guaranteed to pass through the pin synchronization circuitry. Shorter pulses may or may not be recognized. In stop mode, the synchronizer is bypassed so shorter pulses can be recognized in that case.

⁵ Timing is shown with respect to 20% V_{DD} and 80% V_{DD} levels. Temperature range -40°C to 125°C.