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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	833MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DDR, SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	-
USB	-
Voltage - I/O	2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	783-BFBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8560px833lb

  - Contiguous or discontiguous memory mapping
  - Read-modify-write support for RapidIO atomic increment, decrement, set, and clear transactions
  - Sleep mode support for self refresh SDRAM
  - Supports auto refreshing
  - On-the-fly power management using CKE signal
  - Registered DIMM support
  - Fast memory access via JTAG port
  - 2.5-V SSTL2 compatible I/O
- RapidIO interface unit
  - 8-bit RapidIO I/O and messaging protocols
  - Source-synchronous double data rate (DDR) interfaces
  - Supports small type systems (small domain, 8-bit device ID)
  - Supports four priority levels (ordering within a level)
  - Reordering across priority levels
  - Maximum data payload of 256 bytes per packet
  - Packet pacing support at the physical layer
  - CRC protection for packets
  - Supports atomic operations increment, decrement, set, and clear
  - LVDS signaling
- RapidIO–compliant message unit
  - One inbound data message structure (inbox)
  - One outbound data message structure (outbox)
  - Supports chaining and direct modes in the outbox
  - Support of up to 16 packets per message
  - Support of up to 256 bytes per packet and up to 4 Kbytes of data per message
  - Supports one inbound doorbell message structure
- Programmable interrupt controller (PIC)
  - Programming model is compliant with the OpenPIC architecture
  - Supports 16 programmable interrupt and processor task priority levels
  - Supports 12 discrete external interrupts
  - Supports 4 message interrupts with 32-bit messages
  - Supports connection of an external interrupt controller such as the 8259 programmable interrupt controller
  - Four global high resolution timers/counters that can generate interrupts
  - Supports 22 other internal interrupt sources
  - Supports fully nested interrupt delivery
  - Interrupts can be routed to external pin for external processing

  - MII management interface for control and status
  - Programmable CRC generation and checking
  - Ability to force allocation of header information and buffer descriptors into L2 cache.
- OCeaN switch fabric
  - Four-port crossbar packet switch
  - Reorders packets from a source based on priorities
  - Reorders packets to bypass blocked packets
  - Implements starvation avoidance algorithms
  - Supports packets with payloads of up to 256 bytes
- Integrated DMA controller
  - Four-channel controller
  - All channels accessible by both the local and remote masters
  - Extended DMA functions (advanced chaining and striding capability)
  - Support for scatter and gather transfers
  - Misaligned transfer capability
  - Interrupt on completed segment, link, list, and error
  - Supports transfers to or from any local memory or I/O port
  - Selectable hardware-enforced coherency (snoop/no-snoop)
  - Ability to start and flow control each DMA channel from external 3-pin interface
  - Ability to launch DMA from single write transaction
- PCI/PCI-X controller
  - PCI 2.2 and PCI-X 1.0 compatible
  - 64- or 32-bit PCI port supports at 16 to 66 MHz
  - 64-bit PCI-X support up to 133 MHz
  - Host and agent mode support
  - 64-bit dual address cycle (DAC) support
  - PCI-X supports multiple split transactions
  - Supports PCI-to-memory and memory-to-PCI streaming
  - Memory prefetching of PCI read accesses
  - Supports posting of processor-to-PCI and PCI-to-memory writes
  - PCI 3.3-V compatible
  - Selectable hardware-enforced coherency
- Power management
  - Fully static 1.2-V CMOS design with 3.3- and 2.5-V I/O
  - Supports power saving modes: doze, nap, and sleep
  - Employs dynamic power management, which automatically minimizes power consumption of blocks when they are idle.

Table 6 provides estimated I/O power numbers for each block: DDR, PCI, Local Bus, RapidIO, TSEC, and CPM.

Table 6. Estimated Typical I/O Power Consumption

Interface	Parameter	GV_{DD} (2.5 V)	OV_{DD} (3.3 V)	LV_{DD} (3.3 V)	LV_{DD} (2.5 V)	Units	Notes
DDR I/O	CCB = 200 MHz	0.46	—	—	—	W	1
	CCB = 266 MHz	0.59	—	—	—		
	CCB = 300 MHz	0.66	—	—	—		
	CCB = 333 MHz	0.73	—	—	—		
PCI/PCI-X I/O	32-bit, 33 MHz	—	0.04	—	—	W	2
	32-bit 66 MHz	—	0.07	—	—		
	64-bit, 66 MHz	—	0.14	—	—		
	64-bit, 133 MHz	—	0.25	—	—		
Local Bus I/O	32-bit, 33 MHz	—	0.07	—	—	W	3
	32-bit, 66 MHz	—	0.13	—	—		
	32-bit, 133 MHz	—	0.24	—	—		
	32-bit, 167 MHz	—	0.30	—	—		
RapidIO I/O	500 MHz data rate	—	0.96	—	—	W	4
TSEC I/O	MII	—	—	10	—	mW	5, 6
	GMII, TBI (2.5 V)	—	—	—	40		
	GMII, TBI (3.3 V)	—	—	70	—		
	RGMII, RTBI	—	—	—	40		
CPM-FCC	MII	—	15	—	—	mW	7
	RMII	—	13	—	—		
	HDLC 16 Mbps	—	9	—	—		
	UTOPIA-8 SPHY	—	60	—	—		
	UTOPIA-8 MPHY	—	100	—	—		
	UTOPIA-16 SPHY	—	94	—	—		
	UTOPIA-16 MPHY	—	135	—	—		
CPM-SCC	HDLC 16 Mbps	—	4	—	—	mW	7

4.2 TSEC Gigabit Reference Clock Timing

Table 7 provides the TSEC gigabit reference clock (EC_GTX_CLK125) AC timing specifications for the MPC8560.

Table 8. EC_GTX_CLK125 AC Timing Specifications

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
EC_GTX_CLK125 frequency	f_{G125}	—	125	—	MHz	—
EC_GTX_CLK125 cycle time	t_{G125}	—	8	—	ns	—
EC_GTX_CLK125 rise and fall time LV_{DD}=2.5 LV_{DD}=3.3	t_{G125R}, t_{G125F}	—	—	 0.75 1	ns	2
EC_GTX_CLK125 duty cycle GMII, TBI RGMII, RTBI	t_{G125H}/t_{G125}	 45 47	—	 55 53	%	1, 3

Notes:

1. Timing is guaranteed by design and characterization.
2. Rise and fall times for EC_GTX_CLK125 are measured from 0.5V and 2.0V for LV_{DD}=2.5V, and from 0.6 and 2.7V for LV_{DD}=3.3V.
3. EC_GTX_CLK125 is used to generate GTX clock for TSEC transmitter with 2% degradation EC_GTX_CLK125 duty cycle can be loosened from 47/53% as long as PHY device can tolerate the duty cycle generated by GTX_CLK of TSEC.

4.3 RapidIO Transmit Clock Input Timing

Table 9 provides the RapidIO transmit clock input (RIO_TX_CLK_IN) AC timing specifications for the MPC8560.

Table 9. RIO_TX_CLK_IN AC Timing Specifications

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
RIO_TX_CLK_IN frequency	f_{RCLK}	125	—	—	MHz	—
RIO_TX_CLK_IN cycle time	t_{RCLK}	—	—	8	ns	—
RIO_TX_CLK_IN duty cycle	t_{RCLKH}/t_{RCLK}	48	—	52	%	1

Notes:

1. Requires ±100 ppm long term frequency stability. Timing is guaranteed by design and characterization.

7.2.3 TBI AC Timing Specifications

This section describes the TBI transmit and receive AC timing specifications.

7.2.3.1 TBI Transmit AC Timing Specifications

Table 25 provides the TBI transmit AC timing specifications.

Table 25. TBI Transmit AC Timing Specifications

At recommended operating conditions with LV_{DD} of 3.3 V ± 5%, or LV_{DD}=2.5V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
GTX_CLK clock period	t_{TTX}	—	8.0	—	ns
GTX_CLK duty cycle	t_{TTXH}/t_{TTX}	40	—	60	%
TCG[9:0] setup time GTX_CLK going high	t_{TTKHDV}	2.0	—	—	ns
TCG[9:0] hold time from GTX_CLK going high	t_{TTKHDX}	1.0	—	—	ns
GTX_CLK clock rise and fall time	t_{TTXR}, t_{TTXF} [2,3]	—	—	1.0	ns

Notes:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{TTKHDV} symbolizes the TBI transmit timing (TT) with respect to the time from t_{TTX} (K) going high (H) until the referenced data signals (D) reach the valid state (V) or setup time. Also, t_{TTKHDX} symbolizes the TBI transmit timing (TT) with respect to the time from t_{TTX} (K) going high (H) until the referenced data signals (D) reach the invalid state (X) or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{TTX} represents the TBI (T) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. Signal timings are measured at 0.7 V and 1.9 V voltage levels.
3. Guaranteed by design.

Figure 12 shows the TBI transmit AC timing diagram.

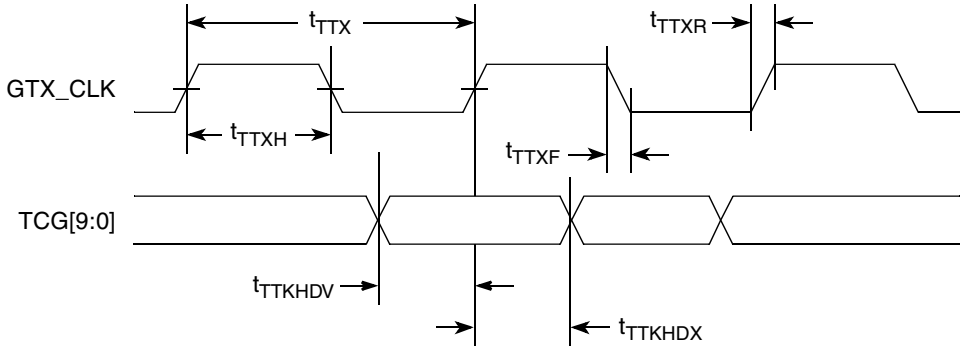

Figure 12. TBI Transmit AC Timing Diagram

Table 28. MII Management DC Electrical Characteristics (continued)

Parameter	Symbol	Min	Max	Unit
Input high current (OV_{DD} = Max, V_{IN} [1] = 2.1 V)	I_{IH}	—	40	μA
Input low current (OV_{DD} = Max, V_{IN} = 0.5 V)	I_{IL}	–600	—	μA

Note:

1.Note that the symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

7.3.2 MII Management AC Electrical Specifications

Table 29 provides the MII management AC timing specifications.

Table 29. MII Management AC Timing Specifications

At recommended operating conditions with OV_{DD} is 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit	Notes
MDC frequency	f_{MDC}	0.893	—	10.4	MHz	2, 4
MDC period	t_{MDC}	96	—	1120	ns	
MDC clock pulse width high	t_{MDCH}	32	—	—	ns	
MDC to MDIO valid	t_{MDKHDV}			$2*[1/(f_{ccb\_clk}/8)]$	ns	3
MDC to MDIO delay	t_{MDKHDX}	10	—	$2*[1/(f_{ccb\_clk}/8)]$	ns	3
MDIO to MDC setup time	t_{MDDVKH}	5	—	—	ns	
MDIO to MDC hold time	t_{MDDXKH}	0	—	—	ns	
MDC rise time	t_{MDCR}	—	—	10	ns	4
MDC fall time	t_{MDHF}	—	—	10	ns	4

Notes:

1.The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MDKHDX} symbolizes management data timing (MD) for the time t_{MDC} from clock reference (K) high (H) until data outputs (D) are invalid (X) or data hold time. Also, t_{MDDVKH} symbolizes management data timing (MD) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MDC} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

2.This parameter is dependent on the CCB clock speed (that is, for a CCB clock of 267 MHz, the maximum frequency is 8.3 MHz and the minimum frequency is 1.2 MHz; for a CCB clock of 333 MHz, the maximum frequency is 10.4 MHz and the minimum frequency is 1.5 MHz).

3.This parameter is dependent on the CCB clock speed (that is, for a CCB clock of 267 MHz, the delay is 60 ns and for a CCB clock of 333 MHz, the delay is 48 ns).

4.Guaranteed by design.

Table 33. CPM DC Electrical Characteristics (continued)

Characteristic	Symbol	Min	Max	Unit	Notes
Output high voltage (I_{OH} = –2.0 mA)	V_{OH}	2.4	—	V	1
Output low voltage (I_{OL} = 3.2 mA)	V_{OL}	—	0.4	V	1

Note:
1. This specification applies to the following pins: PA[0–31], PB[4–31], PC[0–31], and PD[4–31].
2. V_{IL}(max) for the IIC interface is 0.8 V rather than the 1.5 V specified in the IIC standard

9.2 CPM AC Timing Specifications

Table 34 and Table 35 provide the CPM input and output AC timing specifications, respectively.

NOTE: Rise/Fall Time on CPM Input Pins

It is recommended that the rise/fall time on CPM input pins should not exceed 5 ns. This should be enforced especially on clock signals. Rise time refers to signal transitions from 10% to 90% of VCC; fall time refers to transitions from 90% to 10% of VCC.

Table 34. CPM Input AC Timing Specifications [1]

Characteristic	Symbol [2]	Min [3]	Unit
FCC inputs—internal clock (NMSI) input setup time	t_{FIIVKH}	6	ns
FCC inputs—internal clock (NMSI) hold time	t_{FIIXKH}	0	ns
FCC inputs—external clock (NMSI) input setup time	t_{FEIVKH}	2.5	ns
FCC inputs—external clock (NMSI) hold time	t_{FEIXKH}b	2	ns
SCC/SPI inputs—internal clock (NMSI) input setup time	t_{NIIVKH}	6	ns
SCC/SPI inputs—internal clock (NMSI) input hold time	t_{NIIXKH}	0	ns
SCC/SPI inputs—external clock (NMSI) input setup time	t_{NEIVKH}	4	ns
SCC/SPI inputs—external clock (NMSI) input hold time	t_{NEIXKH}	2	ns
TDM inputs/SI—input setup time	t_{TDIVKH}	4	ns
TDM inputs/SI—hold time	t_{TDIXKH}	3	ns

Figure 24 shows the FCC internal clock.

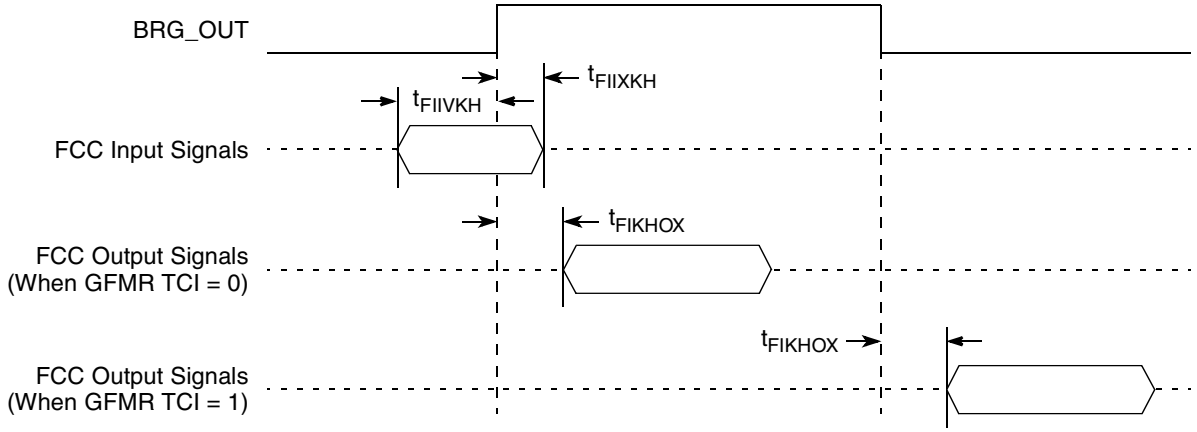

Figure 24. FCC Internal AC Timing Clock Diagram

Figure 25 shows the FCC external clock.

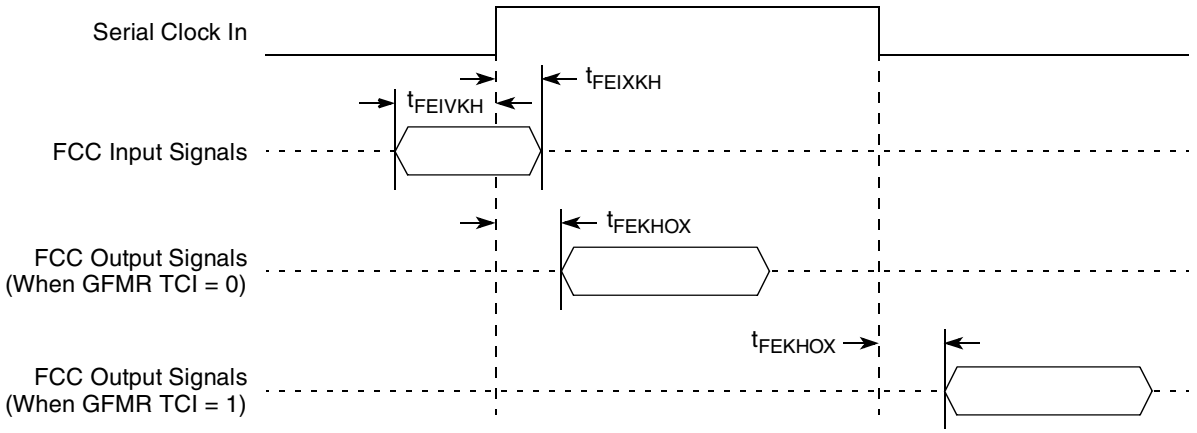

Figure 25. FCC External AC Timing Clock Diagram

Figure 26 shows Ethernet collision timing on FCCs.

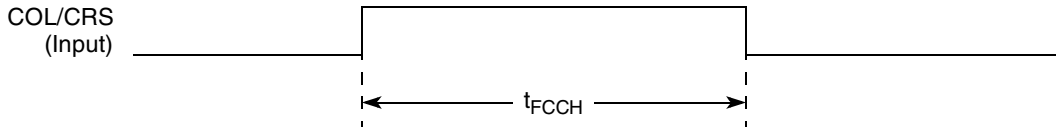

Figure 26. Ethernet Collision AC Timing Diagram (FCC)

Figure 31 provides the AC test load for TDO and the boundary-scan outputs of the MPC8560.

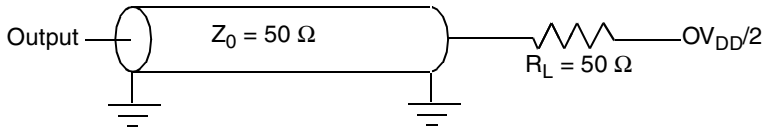

Figure 31. AC Test Load for the JTAG Interface

Figure 32 provides the JTAG clock input timing diagram.

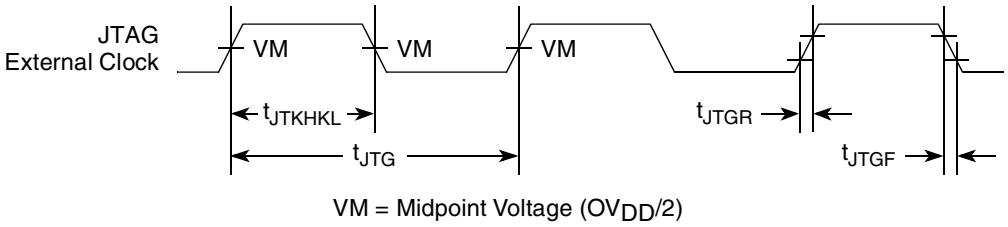

Figure 32. JTAG Clock Input Timing Diagram

Figure 33 provides the $\overline{\text{TRST}}$ timing diagram.

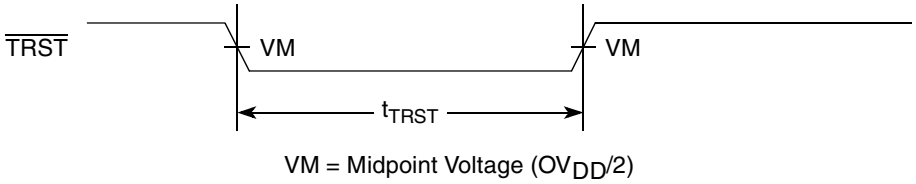

Figure 33. $\overline{\text{TRST}}$ Timing Diagram

Figure 34 provides the boundary-scan timing diagram.

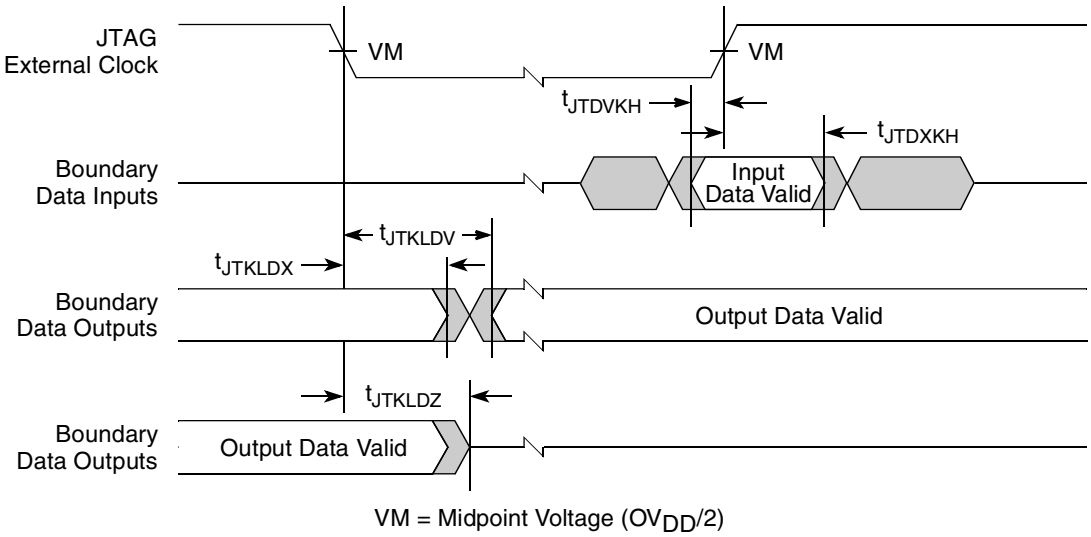

Figure 34. Boundary-Scan Timing Diagram

11.2 I²C AC Electrical Specifications

Table 41 provides the AC timing parameters for the I²C interface of the MPC8560.

Table 41. I²C AC Electrical Specifications

All values refer to V_{IH} (min) and V_{IL} (max) levels (see Table 40).

Parameter	Symbol [1]	Min	Max	Unit
SCL clock frequency	f_{I2C}	0	400	kHz
Low period of the SCL clock	t_{I2CL} [6]	1.3	—	μs
High period of the SCL clock	t_{I2CH} [6]	0.6	—	μs
Setup time for a repeated START condition	t_{I2SVKH} [6]	0.6	—	μs
Hold time (repeated) START condition (after this period, the first clock pulse is generated)	t_{I2SXKL} [6]	0.6	—	μs
Data setup time	t_{I2DVKH} [6]	100	—	ns
Data hold time: CBUS compatible masters I²C bus devices	t_{I2DXKL}	— 0 [2]	— 0.9 [3]	μs
Set-up time for STOP condition	t_{I2PVKH}	0.6	—	μs
Bus free time between a STOP and START condition	t_{I2KHDX}	1.3	—	μs
Noise margin at the LOW level for each connected device (including hysteresis)	V_{NL}	$0.1 \times OV_{DD}$	—	V
Noise margin at the HIGH level for each connected device (including hysteresis)	V_{NH}	$0.2 \times OV_{DD}$	—	V

Notes:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{I2DVKH} symbolizes I²C timing (I2) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SXKL} symbolizes I²C timing (I2) for the time that the data with respect to the start condition (S) went invalid (X) relative to the t_{I2C} clock reference (K) going to the low (L) state or hold time. Also, t_{I2PVKH} symbolizes I²C timing (I2) for the time that the data with respect to the stop condition (P) reaching the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. MPC8560 provides a hold time of at least 300 ns for the SDA signal (referred to the V_{IHmin} of the SCL signal) to bridge the undefined region of the falling edge of SCL.
3. The maximum t_{I2DVKH} has only to be met if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.
4. C_B = capacitance of one bus line in pF.
6. Guaranteed by design.

Figure 16 provides the AC test load for the I²C.

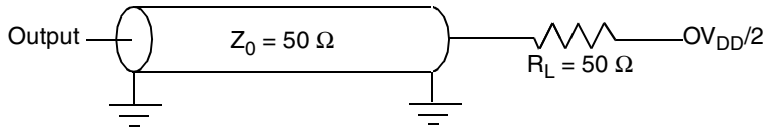

Figure 36. I²C AC Test Load

Table 44. PCI-X AC Timing Specifications at 66 MHz (continued)

Parameter	Symbol	Min	Max	Unit	Notes
PCI-X initialization pattern to $\overline{\text{HRESET}}$ setup time	t_{PCIVRH}	10	—	clocks	11
$\overline{\text{HRESET}}$ to PCI-X initialization pattern hold time	t_{PCRHIX}	0	50	ns	6, 11

Notes:

1.See the timing measurement conditions in the *PCI-X 1.0a Specification*.

2.Minimum times are measured at the package pin (not the test point). Maximum times are measured with the test point and load circuit.

3.Setup time for point-to-point signals applies to $\overline{\text{REQ}}$ and $\overline{\text{GNT}}$ only. All other signals are bused.

4.For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.

5.Setup time applies only when the device is not driving the pin. Devices cannot drive and receive signals at the same time.

6.Maximum value is also limited by delay to the first transaction (time for $\overline{\text{HRESET}}$ high to first configuration access, t_{PCRHFV}). The PCI-X initialization pattern control signals after the rising edge of $\overline{\text{HRESET}}$ must be negated no later than two clocks before the first $\overline{\text{FRAME}}$ and must be floated no later than one clock before $\overline{\text{FRAME}}$ is asserted.

7.A PCI-X device is permitted to have the minimum values shown for t_{PCKHOV} and t_{CYC} only in PCI-X mode. In conventional mode, the device must meet the requirements specified in PCI 2.2 for the appropriate clock frequency.

8.Device must meet this specification independent of how many outputs switch simultaneously.

9.The timing parameter t_{PCRHFV} is a minimum of 10 clocks rather than the minimum of 5 clocks in the *PCI-X 1.0a Specification.*

10.Guaranteed by characterization.

11.Guaranteed by design.

Table 45 provides the PCI-X AC timing specifications at 133 MHz.

Table 45. PCI-X AC Timing Specifications at 133 MHz

Parameter	Symbol	Min	Max	Unit	Notes
SYSCLK to signal valid delay	t_{PCKHOV}	—	3.8	ns	1, 2, 3, 7, 8
Output hold from SYSCLK	t_{PCKHOX}	0.7	—	ns	1, 11
SYSCLK to output high impedance	t_{PCKHOZ}	—	7	ns	1, 4, 8, 12
Input setup time to SYSCLK	t_{PCIVKH}	1.4	—	ns	3, 5, 9, 11
Input hold time from SYSCLK	t_{PCIXKH}	0.5	—	ns	11
$\overline{\text{REQ64}}$ to $\overline{\text{HRESET}}$ setup time	t_{PCRVRH}	10	—	clocks	12
$\overline{\text{HRESET}}$ to $\overline{\text{REQ64}}$ hold time	t_{PCRHRX}	0	50	ns	12
$\overline{\text{HRESET}}$ high to first $\overline{\text{FRAME}}$ assertion	t_{PCRHFV}	10	—	clocks	10, 12
PCI-X initialization pattern to $\overline{\text{HRESET}}$ setup time	t_{PCIVRH}	10	—	clocks	12

- The peak differential signal of the transmitter output or receiver input, is A – B volts.
- The peak-to-peak differential signal of the transmitter output or receiver input, is 2 × (A – B) volts.

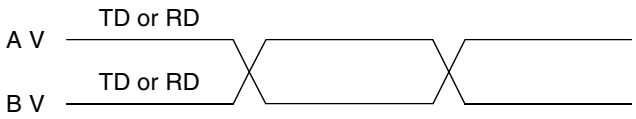

Figure 42. Differential Peak-to-Peak Voltage of Transmitter or Receiver

To illustrate these definitions using numerical values, consider the case where a LVDS transmitter has a common mode voltage of 1.2 V and each signal has a swing that goes between 1.4 and 1.0 V. Using these values, the peak-to-peak voltage swing of the signals TD, $\overline{\text{TD}}$, RD, and $\overline{\text{RD}}$ is 400 mV. The differential signal ranges between 400 and –400 mV. The peak differential signal is 400 mV, and the peak-to-peak differential signal is 800 mV.

A timing edge is the zero-crossing of a differential signal. Each skew timing parameter on a parallel bus is synchronously measured on two signals relative to each other in the same cycle, such as data to data, data to clock, or clock to clock. A skew timing parameter may be relative to the edge of a signal or to the middle of two sequential edges.

Static skew represents the timing difference between signals that does not vary over time regardless of system activity or data pattern. Path length differences are a primary source of static skew.

Dynamic skew represents the amount of timing difference between signals that is dependent on the activity of other signals and varies over time. Crosstalk between signals is a source of dynamic skew.

Eye diagrams and compliance masks are a useful way to visualize and specify driver and receiver performance. This technique is used in several serial bus specifications. An example compliance mask is shown in Figure 43. The key difference in the application of this technique for a parallel bus is that the data is source synchronous to its bus clock while serial data is referenced to its embedded clock. Eye diagrams reveal the quality (cleanness, openness, goodness) of a driver output or receiver input. An advantage of using an eye diagram and a compliance mask is that it allows specifying the quality of a signal without requiring separate specifications for effects such as rise time, duty cycle distortion, data dependent dynamic skew, random dynamic skew, etc. This allows the individual semiconductor manufacturer maximum flexibility to trade off various performance criteria while keeping the system performance constant.

In using the eye pattern and compliance mask approach, the quality of the signal is specified by the compliance mask. The mask specifies the maximum permissible magnitude of the signal and the minimum permissible eye opening. The eye diagram for the signal under test is generated according to the specification. Compliance is determined by whether the compliance mask can be positioned over the eye diagram such that the eye pattern falls entirely within the unshaded portion of the mask.

Serial specifications have clock encoded with the data, but the LP-LVDS physical layer defined by RapidIO is a source synchronous parallel port so additional specifications to include effects that are not found in serial links are required. Specifications for the effect of bit to bit timing differences caused by static skew have been added and the eye diagrams specified are measured relative to the associated clock in order to include clock to data effects. With the transmit output (or receiver input) eye diagram, the user can determine if the transmitter output (or receiver input) is compliant with an oscilloscope with the appropriate software.

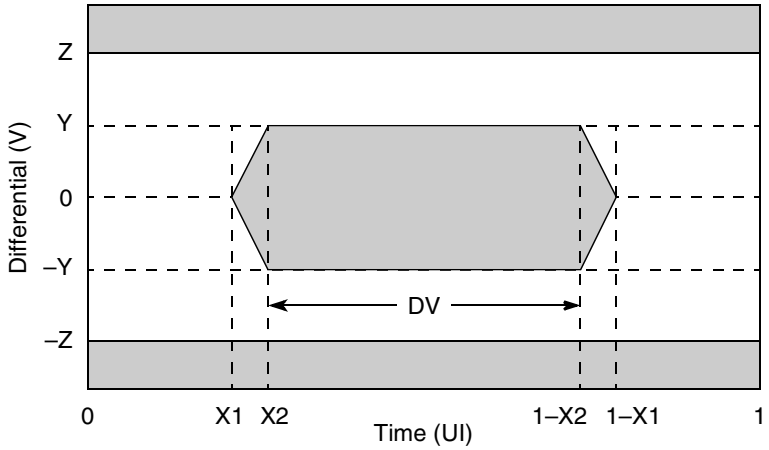

Figure 43. Example Compliance Mask

Y = minimum data valid amplitude

Z = maximum amplitude

1 UI = 1 unit interval = 1/baud rate

X1 = end of zero crossing region

X2 = beginning of data valid window

DV = data valid window = $1 - 2 \times X2$

The waveform of the signal under test must fall within the unshaded area of the mask to be compliant. Different masks are used for the driver output and the receiver input allowing each to be separately specified.

13.3.1 RapidIO Driver AC Timing Specifications

Driver AC timing specifications are provided in Table 48, Table 49, and Table 50. A driver shall comply with the specifications for each data rate/frequency for which operation of the driver is specified. Unless otherwise specified, these specifications are subject to the following conditions.

- The specifications apply over the supply voltage and ambient temperature ranges specified by the device vendor.
- The specifications apply for any combination of data patterns on the data signals.
- The output of a driver shall be connected to a 100 Ω, ±1%, differential (bridged) resistive load.
- Clock specifications apply only to clock signals.
- Data specifications apply only to data signals (FRAME, D[0:7]).

Table 48. RapidIO Driver AC Timing Specifications—500 Mbps Data Rate

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Differential output high voltage	V_{OHD}	200	540	mV	1
Differential output low voltage	V_{OLD}	–540	–200	mV	1

The eye pattern for a data signal is generated by making a large number of recordings of the signal and then overlaying the recordings. The number of recordings used to generate the eye shall be large enough that further increasing the number of recordings used does not cause the resulting eye pattern to change from one that complies with the RapidIO transmit mask to one that does not. Each data signal in the interface shall be carrying random or pseudo-random data when the recordings are made. If pseudo-random data is used, the length of the pseudo-random sequence (repeat length) shall be long enough that increasing the length of the sequence does not cause the resulting eye pattern to change from one that complies with the RapidIO transmit mask to one that does not comply with the mask. The data carried by any given data signal in the interface may not be correlated with the data carried by any other data signal in the interface. The zero-crossings of the clock associated with a data signal shall be used as the timing reference for aligning the multiple recordings of the data signal when the recordings are overlaid.

While the method used to make the recordings and overlay them to form the eye pattern is not specified, the method used shall be demonstrably equivalent to the following method. The signal under test is repeatedly recorded with a digital oscilloscope in infinite persistence mode. Each recording is triggered by a zero-crossing of the clock associated with the data signal under test. Roughly half of the recordings are triggered by positive-going clock zero-crossings and roughly half are triggered by negative-going clock zero-crossings. Each recording is at least 1.9 UI in length (to ensure that at least one complete eye is formed) and begins 0.5 UI before the trigger point (0.5 UI before the associated clock zero-crossing). Depending on the length of the individual recordings used to generate the eye pattern, one or more complete eyes will be formed. Regardless of the number of eyes, the eye whose center is immediately to the right of the trigger point is the eye used for compliance testing.

An example of an eye pattern generated using the above method with recordings 3 UI in length is shown in Figure 45. In this example, there is no skew between the signal under test and the associated clock used to trigger the recordings. If skew was present, the eye pattern would be shifted to the left or right relative to the oscilloscope trigger point.

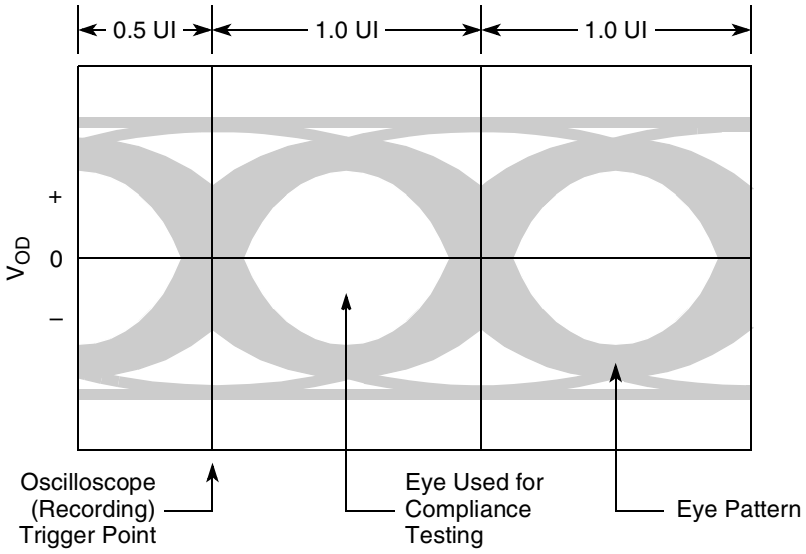

Figure 45. Example Driver Output Eye Pattern

Table 54. MPC8560 Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
DDR SDRAM Memory Interface				
MDQ[0:63]	M26, L27, L22, K24, M24, M23, K27, K26, K22, J28, F26, E27, J26, J23, H26, G26, C26, E25, C24, E23, D26, C25, A24, D23, B23, F22, J21, G21, G22, D22, H21, E21, N18, J18, D18, L17, M18, L18, C18, A18, K17, K16, C16, B16, G17, L16, A16, L15, G15, E15, C14, K13, C15, D15, E14, D14, D13, E13, D12, A11, F13, H13, A13, B12	I/O	GV_{DD}	—
MECC[0:7]	N20, M20, L19, E19, C21, A21, G19, A19	I/O	GV_{DD}	—
MDM[0:8]	L24, H28, F24, L21, E18, E16, G14, B13, M19	O	GV_{DD}	—
MDQS[0:8]	L26, J25, D25, A22, H18, F16, F14, C13, C20	I/O	GV_{DD}	—
MBA[0:1]	B18, B19	O	GV_{DD}	—
MA[0:14]	N19, B21, F21, K21, M21, C23, A23, B24, H23, G24, K19, B25, D27, J14, J13	O	GV_{DD}	—
$\overline{\text{MWE}}$	D17	O	GV_{DD}	—
$\overline{\text{MRAS}}$	F17	O	GV_{DD}	—
$\overline{\text{MCAS}}$	J16	O	GV_{DD}	—
$\overline{\text{MCS}}$[0:3]	H16, G16, J15, H15	O	GV_{DD}	—
MCKE[0:1]	E26, E28	O	GV_{DD}	11
MCK[0:5]	J20, H25, A15, D20, F28, K14	O	GV_{DD}	—
$\overline{\text{MCK}}$[0:5]	F20, G27, B15, E20, F27, L14	O	GV_{DD}	—
MSYNC_IN	M28	I	GV_{DD}	—
MSYNC_OUT	N28	O	GV_{DD}	—
Local Bus Controller Interface				
LA[27]	U18	O	OV_{DD}	5, 9
LA[28:31]	T18, T19, T20, T21	O	OV_{DD}	7, 9
LAD[0:31]	AD26, AD27, AD28, AC26, AC27, AC28, AA22, AA23, AA26, Y21, Y22, Y26, W20, W22, W26, V19, T22, R24, R23, R22, R21, R18, P26, P25, P20, P19, P18, N22, N23, N24, N25, N26	I/O	OV_{DD}	—
LALE	V21	O	OV_{DD}	8, 9
LBCTL	V20	O	OV_{DD}	9
LCKE	U23	O	OV_{DD}	—
LCLK[0:2]	U27, U28, V18	O	OV_{DD}	—
$\overline{\text{LCS}}$[0:4]	Y27, Y28, W27, W28, R27	O	OV_{DD}	18
$\overline{\text{LCS5}}$/$\overline{\text{DMA\_DREQ2}}$	R28	I/O	OV_{DD}	1

Table 54. MPC8560 Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
LV_{DD}	A4, C5, E7, H10	Reference Voltage; Three-Speed Ethernet I/O (2.5 V, 3.3 V)	LV_{DD}	—
MV_{REF}	N27	Reference Voltage Signal; DDR	MV_{REF}	—
No Connects	AH26, AH27, AH28, AG28, AF28, AE28, AH1, AG1, AH2, B1, B2, A2, A3, AH25	—	—	16
OV_{DD}	D1, E4, H3, K4, K10, L7, M5, N3, P22, R19, R25, T2, T7, U5, U20, U26, V8, W4, W13, W19, W21, Y7, Y23, AA5, AA12, AA16, AA20, AB7, AB9, AB26, AC5, AC11, AC17, AD4, AE1, AE8, AE10, AE15, AF7, AF12, AG27, AH4	PCI/PCI-X, RapidIO, 10/100 Ethernet, and other Standard (3.3 V)	OV_{DD}	—
RESERVED	C1, T11, U11, AF1	—	—	15
SENSEVDD	L12	Power for Core (1.2 V)	V_{DD}	13
SENSEVSS	K12	—	—	13
V_{DD}	M13, M15, M17, N14, N16, P13, P15, P17, R12, R14, R16, T13, T15, T17, U12, U14	Power for Core (1.2 V)	V_{DD}	—
CPM				
PA[0:31]	H1, H2, J1, J2, J3, J4, J5, J6, J7, J8, K8, K7, K6, K3, K2, K1, L1, L2, L3, L4, L5, L8, L9, L10, L11, M10, M9, M8, M7, M6, M3, M2	I/0	OV_{DD}	—
PB[4:31]	M1, N1, N4, N5, N6, N7, N8, N9, N10, N11, P11, P10, P9, P8, P7, P6, P5, P4, P3, P2, P1, R1, R2, R3, R4, R5, R6, R7	I/0	OV_{DD}	—
PC[0:31]	R8, R9, R10, R11, T9, T6, T5, T4, T1, U1, U2, U3, U4, U7, U8, U9, U10, V9, V6, V5, V4, V3, V2, V1, W1, W2, W3, W6, W7, W8, W9, Y9	I/0	OV_{DD}	—

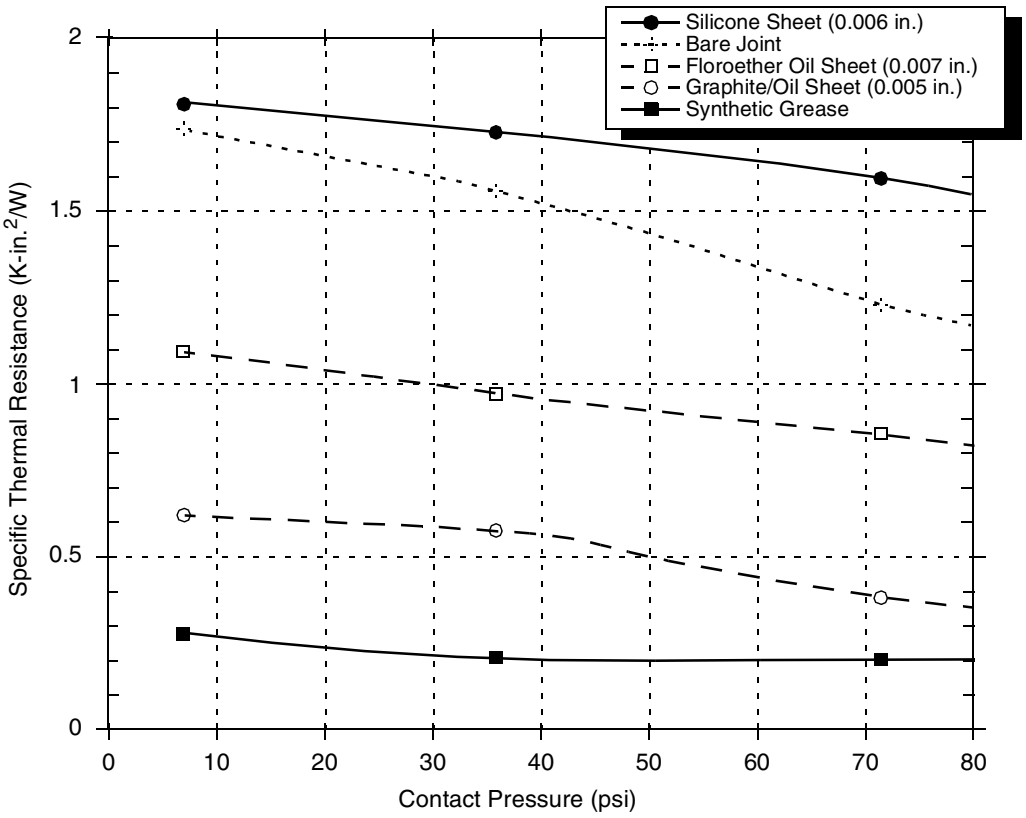

Figure 54. Thermal Performance of Select Thermal Interface Materials

The system board designer can choose between several types of thermal interface. There are several commercially-available thermal interfaces provided by the following vendors:

Chomerics, Inc. 77 Dragon Ct. Woburn, MA 01888-4014 Internet: www.chomerics.com	781-935-4850
Dow-Corning Corporation Dow-Corning Electronic Materials 2200 W. Salzburg Rd. Midland, MI 48686-0997 Internet: www.dowcorning.com	800-248-2481
Shin-Etsu MicroSi, Inc. 10028 S. 51st St. Phoenix, AZ 85044 Internet: www.microsi.com	888-642-7674
The Bergquist Company 18930 West 78th St. Chanhassen, MN 55317 Internet: www.bergquistcompany.com	800-347-4572

When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

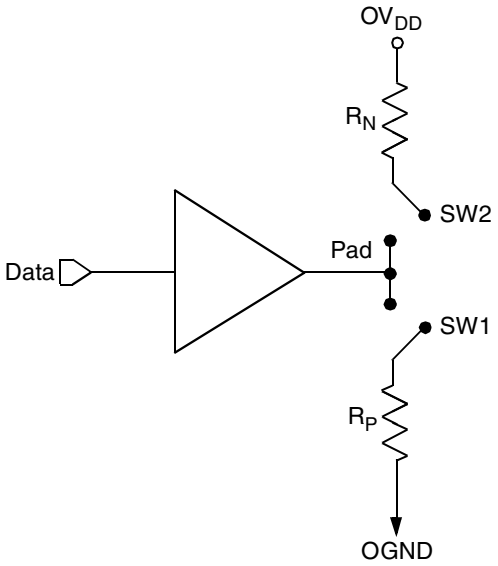

Figure 59. Driver Impedance Measurement

The output impedance of the RapidIO port drivers targets 200-Ω differential resistance. The value of this resistance and the strength of the driver's current source can be found by making two measurements. First, the output voltage is measured while driving logic 1 without an external differential termination resistor. The measured voltage is $V_1 = R_{source} \times I_{source}$. Second, the output voltage is measured while driving logic 1 with an external precision differential termination resistor of value R_{term}. The measured voltage is $V_2 = 1/(1/R_1 + 1/R_2)) \times I_{source}$. Solving for the output impedance gives $R_{source} = R_{term} \times (V_1/V_2 - 1)$. The drive current is then $I_{source} = V_1/R_{source}$.

Table 61 summarizes the signal impedance targets. The driver impedance are targeted at minimum V_{DD}, nominal OV_{DD}, 105°C.

Table 61. Impedance Characteristics

Impedance	Local Bus, Ethernet, DUART, Control, Configuration, Power Management	PCI/PCI-X	DDR DRAM	RapidIO	Symbol	Unit
R_N	43 Target	25 Target	20 Target	NA	Z_0	W
R_P	43 Target	25 Target	20 Target	NA	Z_0	W
Differential	NA	NA	NA	200 Target	Z_{DIFF}	W

Note: Nominal supply voltages. See Table 1, T_j = 105°C.

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