

Welcome to [E-XFL.COM](#)

### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	HCS12
Core Size	16-Bit
Speed	32MHz
Connectivity	CANbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	34
Program Memory Size	96KB (96K x 8)
Program Memory Type	FLASH
EEPROM Size	4K x 8
RAM Size	6K x 8
Voltage - Supply (Vcc/Vdd)	1.72V ~ 5.5V
Data Converters	A/D 10x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	48-TFQFN Exposed Pad
Supplier Device Package	48-QFN-EP (7x7)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/s9s12p96j0mft">https://www.e-xfl.com/product-detail/nxp-semiconductors/s9s12p96j0mft</a>

- Programmable character length
- Programmable polarity for transmitter and receiver
- Active edge receive wakeup
- Break detect and transmit collision detect supporting LIN

### 1.3.12 Serial Peripheral Interface Module (SPI)

- Configurable 8- or 16-bit data size
- Full-duplex or single-wire bidirectional
- Double-buffered transmit and receive
- Master or slave mode
- MSB-first or LSB-first shifting
- Serial clock phase and polarity options

### 1.3.13 Analog-to-Digital Converter Module (ATD)

- 10-channel, 12-bit analog-to-digital converter
  - 3  $\mu$ s single conversion time
  - 8-/10-/12-bit resolution
  - Left or right justified result data
  - Internal oscillator for conversion in stop modes
  - Wakeup from low power modes on analog comparison  $>$  or  $\leq$  match
  - Continuous conversion mode
  - Multiple channel scans
- Pins can also be used as digital I/O

### 1.3.14 On-Chip Voltage Regulator (VREG)

- Linear voltage regulator with bandgap reference
- Low-voltage detect (LVD) with low-voltage interrupt (LVI)
- Power-on reset (POR) circuit
- Low-voltage reset (LVR)
- High temperature sensor

### 1.3.15 Background Debug (BDM)

- Non-intrusive memory access commands
- Supports in-circuit programming of on-chip nonvolatile memory

### 1.7.4.6 VSSPLL — Ground Pin for PLL

This pin provides ground for the oscillator and the phased-locked loop. The voltage supply of nominally 1.8V is derived from the internal voltage regulator.

### 1.7.4.7 Power and Ground Connection Summary

Table 1-9. Power and Ground Connection Summary

Mnemonic	Nominal Voltage	Description
VDDR	5.0 V	External power supply to internal voltage regulator
VDDX[2:1]	5.0 V	External power and ground, supply to pin drivers
VSSX[2:1]	0 V	
VDDA	5.0 V	Operating voltage and ground for the analog-to-digital converters and the reference for the internal voltage regulator, allows the supply voltage to the A/D to be bypassed independently.
VSSA	0 V	
VRL	0 V	Reference voltages for the analog-to-digital converter.
VRH	5.0 V	
VSS3	0V	Internal power and ground generated by internal regulator for the internal core.
VSSPLL	0V	Provides operating voltage and ground for the phased-locked loop. This allows the supply voltage to the PLL to be bypassed independently. Internal power and ground generated by internal regulator.

## 1.8 System Clock Description

For the system clock description please refer to chapter [Chapter 7, “S12 Clock, Reset and Power Management Unit \(S12CPMU\)”](#).

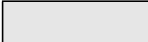
## 1.9 Modes of Operation

The MCU can operate in different modes. These are described in [1.9.1 Chip Configuration Summary](#).

The MCU can operate in different power modes to facilitate power saving when full system performance is not required. These are described in [1.9.2 Low Power Operation](#).

Some modules feature a software programmable option to freeze the module status whilst the background debug module is active to facilitate debugging.

Register Name		Bit 7	6	5	4	3	2	1	Bit 0
<u>0x027C</u>	R	0	0	0	0	0	0	0	0
Reserved	W								
<u>0x027D</u>	R	0	0	0	0	0	0	0	0
Reserved	W								
<u>0x027E</u>	R	0	0	0	0	0	0	0	0
Reserved	W								
<u>0x027F</u>	R	0	0	0	0	0	0	0	0
Reserved	W								

 = Unimplemented or Reserved

### 2.3.2 Register Descriptions

The following table summarizes the effect of the various configuration bits, i.e. data direction (DDR), output level (IO), reduced drive (RDR), pull enable (PE), pull select (PS) on the pin function and pull device activity.

The configuration bit PS is used for two purposes:

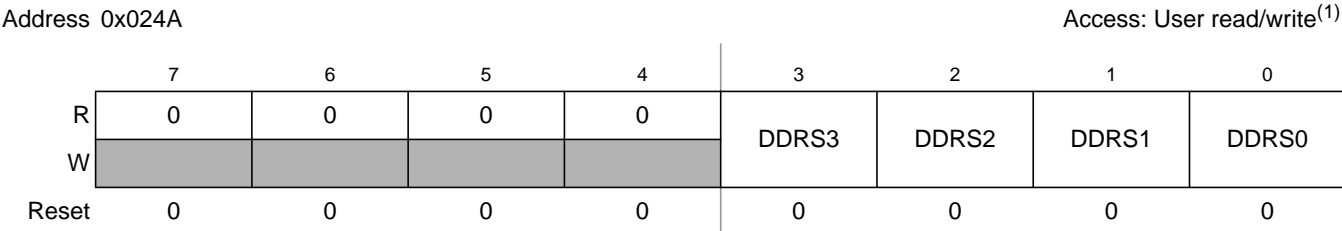
1. Configure the sensitive interrupt edge (rising or falling), if interrupt is enabled.
2. Select either a pull-up or pull-down device if PE is active.

1. Read: Anytime  
Write: Never, writes to this register have no effect.

**Table 2-22. PTIS Register Field Descriptions**

Field	Description
3-0 PTIS	<b>Port S input data—</b> A read always returns the buffered input state of the associated pin. It can be used to detect overload or short circuit conditions on output pins.

### 2.3.26 Port S Data Direction Register (DDRS)



**Figure 2-24. Port S Data Direction Register (DDRS)**

1. Read: Anytime  
Write: Anytime

**Table 2-23. DDRS Register Field Descriptions**

Field	Description
3-2 DDRS	<b>Port S data direction—</b> This bit determines whether the associated pin is an input or output.  1 Associated pin is configured as output 0 Associated pin is configured as input
1 DDRS	<b>Port S data direction—</b> This bit determines whether the associated pin is an input or output. Depending on the configuration of the enabled SCI the I/O state will be forced to be input or output. In this case the data direction bit will not change.  1 Associated pin is configured as output 0 Associated pin is configured as input
0 DDRS	<b>Port S data direction—</b> This bit determines whether the associated pin is an input or output. Depending on the configuration of the enabled SCI the I/O state will be forced to be input or output. In this case the data direction bit will not change.  1 Associated pin is configured as output 0 Associated pin is configured as input

CALL/RTC instructions should only be used when needed. The JSR and RTS instructions can be used to access subroutines that are already present in the local CPU memory map (i.e. in the same page in the program memory page window for example). However calling a function located in a different page requires usage of the CALL instruction. The function must be terminated by the RTC instruction. Because the RTC instruction restores contents of the PPAGE register from the stack, functions terminated with the RTC instruction must be called using the CALL instruction even when the correct page is already present in the memory map. This is to make sure that the correct PPAGE value will be present on stack at the time of the RTC instruction execution.

## NOTE

This information is being provided so that the MCU integrator will be aware that such a conflict could occur.

The hardware handshake protocol is enabled by the `ACK_ENABLE` and disabled by the `ACK_DISABLE` BDM commands. This provides backwards compatibility with the existing POD devices which are not able to execute the hardware handshake protocol. It also allows for new POD devices, that support the hardware handshake protocol, to freely communicate with the target device. If desired, without the need for waiting for the ACK pulse.

The commands are described as follows:

- `ACK_ENABLE` — enables the hardware handshake protocol. The target will issue the ACK pulse when a CPU command is executed by the CPU. The `ACK_ENABLE` command itself also has the ACK pulse as a response.
- `ACK_DISABLE` — disables the ACK pulse protocol. In this case, the host needs to use the worst case delay time at the appropriate places in the protocol.

The default state of the BDM after reset is hardware handshake protocol disabled.

All the read commands will ACK (if enabled) when the data bus cycle has completed and the data is then ready for reading out by the BKGD serial pin. All the write commands will ACK (if enabled) after the data has been received by the BDM through the BKGD serial pin and when the data bus cycle is complete. See [Section 5.4.3, “BDM Hardware Commands”](#) and [Section 5.4.4, “Standard BDM Firmware Commands”](#) for more information on the BDM commands.

The `ACK_ENABLE` sends an ACK pulse when the command has been completed. This feature could be used by the host to evaluate if the target supports the hardware handshake protocol. If an ACK pulse is issued in response to this command, the host knows that the target supports the hardware handshake protocol. If the target does not support the hardware handshake protocol the ACK pulse is not issued. In this case, the `ACK_ENABLE` command is ignored by the target since it is not recognized as a valid command.

The `BACKGROUND` command will issue an ACK pulse when the CPU changes from normal to background mode. The ACK pulse related to this command could be aborted using the `SYNC` command.

The `GO` command will issue an ACK pulse when the CPU exits from background mode. The ACK pulse related to this command could be aborted using the `SYNC` command.

The `GO_UNTIL` command is equivalent to a `GO` command with exception that the ACK pulse, in this case, is issued when the CPU enters into background mode. This command is an alternative to the `GO` command and should be used when the host wants to trace if a breakpoint match occurs and causes the CPU to enter active background mode. Note that the ACK is issued whenever the CPU enters BDM, which could be caused by a breakpoint match or by a `BGND` instruction being executed. The ACK pulse related to this command could be aborted using the `SYNC` command.

The `TRACE1` command has the related ACK pulse issued when the CPU enters background active mode after one instruction of the application program is executed. The ACK pulse related to this command could be aborted using the `SYNC` command.

Table 6-28. DBGADH Field Descriptions

Field	Description
7–0 Bits[15:8]	<b>Comparator Data High Compare Bits</b> — The Comparator data high compare bits control whether the selected comparator compares the data bus bits [15:8] to a logic one or logic zero. The comparator data compare bits are only used in comparison if the corresponding data mask bit is logic 1. This register is available only for comparator A. Data bus comparisons are only performed if the TAG bit in DBGACTL is clear. 0 Compare corresponding data bit to a logic zero 1 Compare corresponding data bit to a logic one

### 6.3.2.8.6 Debug Comparator Data Low Register (DBGADL)

Address: 0x002D

	7	6	5	4	3	2	1	0
R	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
W								
Reset	0	0	0	0	0	0	0	0

Figure 6-20. Debug Comparator Data Low Register (DBGADL)

Read: If COMRV[1:0] = 00

Write: If COMRV[1:0] = 00 and DBG not armed.

Table 6-29. DBGADL Field Descriptions

Field	Description
7–0 Bits[7:0]	<b>Comparator Data Low Compare Bits</b> — The Comparator data low compare bits control whether the selected comparator compares the data bus bits [7:0] to a logic one or logic zero. The comparator data compare bits are only used in comparison if the corresponding data mask bit is logic 1. This register is available only for comparator A. Data bus comparisons are only performed if the TAG bit in DBGACTL is clear 0 Compare corresponding data bit to a logic zero 1 Compare corresponding data bit to a logic one

### 6.3.2.8.7 Debug Comparator Data High Mask Register (DBGADHM)

Address: 0x002E

	7	6	5	4	3	2	1	0
R	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8
W								
Reset	0	0	0	0	0	0	0	0

Figure 6-21. Debug Comparator Data High Mask Register (DBGADHM)

Read: If COMRV[1:0] = 00

Write: If COMRV[1:0] = 00 and DBG not armed.





## 7.3 Memory Map and Registers

This section provides a detailed description of all registers accessible in the S12CPMU.

### 7.3.1 Module Memory Map

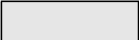
The S12CPMU registers are shown in [Figure 7-3](#).

Address s	Name		Bit 7	6	5	4	3	2	1	Bit 0
0x0034	CPMU SYNR	R W	VCOFRQ[1:0]		SYNDIV[5:0]					
0x0035	CPMU REFDIV	R W	REFFRQ[1:0]		0	0	REFDIV[3:0]			
0x0036	CPMU POSTDIV	R W			0	0	0	POSTDIV[4:0]		
0x0037	CPMUFLG	R W	RTIF	PORF	LVRF	LOCKIF	LOCK	ILAF	OSCIF	UPOSC
0x0038	CPMUINT	R W	RTIE	0	0	LOCKIE	0	0	OSCIE	0
0x0039	CPMUCLKS	R W	PLLSEL	PSTP	0	0	PRE	PCE	RTI OSCSEL	COP OSCSEL
0x003A	CPMUPLL	R W	0	0	FM1	FM0	0	0	0	0
0x003B	CPMURTI	R W	RTDEC	RTR6	RTR5	RTR4	RTR3	RTR2	RTR1	RTR0
0x003C	CPMUCOP	R W	WCOP	RSBCK	0	0	0	CR2	CR1	CR0
0x003D	RESERVED CPMUTEST0	R W	0	0	0	0	0	0	0	0
0x003E	RESERVED CPMUTEST1	R W	0	0	0	0	0	0	0	0
0x003F	CPMU ARMCOP	R W	0	0	0	0	0	0	0	0
0x02F0	CPMU HTCTL	R W	0	0	VSEL	0	HTE	HTDS	HTIE	HTIF
0x02F1	CPMU LVCTL	R W	0	0		0	0	0	LVDS	LVIE
0x02F2	CPMU APICTL	R W	APICLK	0	0	APIES	APIEA	APIFE	APIE	APIF
				= Unimplemented or Reserved						

= Unimplemented or Reserved

**Figure 7-3. CPMU Register Summary**

Register Name		Bit 7	6	5	4	3	2	1	Bit 0
0x000E CANRXERR	R	RXERR7	RXERR6	RXERR5	RXERR4	RXERR3	RXERR2	RXERR1	RXERR0
	W								
0x000F CANTXERR	R	TXERR7	TXERR6	TXERR5	TXERR4	TXERR3	TXERR2	TXERR1	TXERR0
	W								
0x0010–0x0013 CANIDAR0–3	R	AC7	AC6	AC5	AC4	AC3	AC2	AC1	AC0
	W								
0x0014–0x0017 CANIDMRx	R	AM7	AM6	AM5	AM4	AM3	AM2	AM1	AM0
	W								
0x0018–0x001B CANIDAR4–7	R	AC7	AC6	AC5	AC4	AC3	AC2	AC1	AC0
	W								
0x001C–0x001F CANIDMR4–7	R	AM7	AM6	AM5	AM4	AM3	AM2	AM1	AM0
	W								
0x0020–0x002F CANRXFG	R	See <a href="#">Section 8.3.3, “Programmer’s Model of Message Storage”</a>							
	W								
0x0030–0x003F CANTXFG	R	See <a href="#">Section 8.3.3, “Programmer’s Model of Message Storage”</a>							
	W								

 = Unimplemented or Reserved

**Figure 8-3. MSCAN Register Summary (continued)**

## 8.3.2 Register Descriptions


This section describes in detail all the registers and register bits in the MSCAN module. Each description includes a standard register diagram with an associated figure number. Details of register bit and field function follow the register diagrams, in bit order. All bits of all registers in this module are completely synchronous to internal clocks during a register read.

### 8.3.2.1 MSCAN Control Register 0 (CANCTL0)

The CANCTL0 register provides various control bits of the MSCAN module as described below.

**Figure 8-24. Receive/Transmit Message Buffer — Extended Identifier Mapping (continued)**

Register Name		Bit 7	6	5	4	3	2	1	Bit0
<u>0x00X4</u> DSR0	R	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
	W								
<u>0x00X5</u> DSR1	R	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
	W								
<u>0x00X6</u> DSR2	R	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
	W								
<u>0x00X7</u> DSR3	R	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
	W								
<u>0x00X8</u> DSR4	R	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
	W								
<u>0x00X9</u> DSR5	R	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
	W								
<u>0x00XA</u> DSR6	R	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
	W								
<u>0x00XB</u> DSR7	R	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
	W								
<u>0x00XC</u> DLR	R					DLC3	DLC2	DLC1	DLC0
	W								

 = Unused, always read 'x'

Read:

- For transmit buffers, anytime when TXEx flag is set (see [Section 8.3.2.7, “MSCAN Transmitter Flag Register \(CANTFLG\)”](#)) and the corresponding transmit buffer is selected in CANTBSEL (see [Section 8.3.2.11, “MSCAN Transmit Buffer Selection Register \(CANTBSEL\)”](#)).
- For receive buffers, only when RXF flag is set (see [Section 8.3.2.5, “MSCAN Receiver Flag Register \(CANRFLG\)”](#)).

Write:



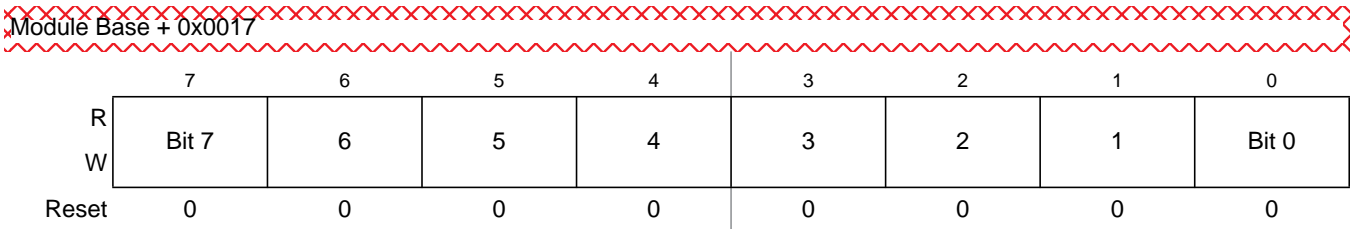


Figure 10-26. PWM Channel Period Registers (PWMPER5)

Read: anytime

Write: anytime

### 10.3.2.14 PWM Channel Duty Registers (PWMDTYx)

There is a dedicated duty register for each channel. The value in this register determines the duty of the associated PWM channel. The duty value is compared to the counter and if it is equal to the counter value a match occurs and the output changes state.

The duty registers for each channel are double buffered so that if they change while the channel is enabled, the change will NOT take effect until one of the following occurs:

- The effective period ends
- The counter is written (counter resets to 0x0000)
- The channel is disabled

In this way, the output of the PWM will always be either the old duty waveform or the new duty waveform, not some variation in between. If the channel is not enabled, then writes to the duty register will go directly to the latches as well as the buffer.

#### NOTE

Reads of this register return the most recent value written. Reads do not necessarily return the value of the currently active duty due to the double buffering scheme.

Reference [Section 10.4.2.3, “PWM Period and Duty,”](#) for more information.

#### NOTE

Depending on the polarity bit, the duty registers will contain the count of either the high time or the low time. If the polarity bit is 1, the output starts high and then goes low when the duty count is reached, so the duty registers contain a count of the high time. If the polarity bit is 0, the output starts low and then goes high when the duty count is reached, so the duty registers contain a count of the low time.

To calculate the output duty cycle (high time as a % of period) for a particular channel:

- Polarity = 0 (PPOLx = 0)  
Duty cycle =  $[(PWMPERx \text{ PWMDTYx}) / PWMPERx] * 100\%$
- Polarity = 1 (PPOLx = 1)  
Duty cycle =  $[PWMDTYx / PWMPERx] * 100\%$

## 11.4.6 Receiver

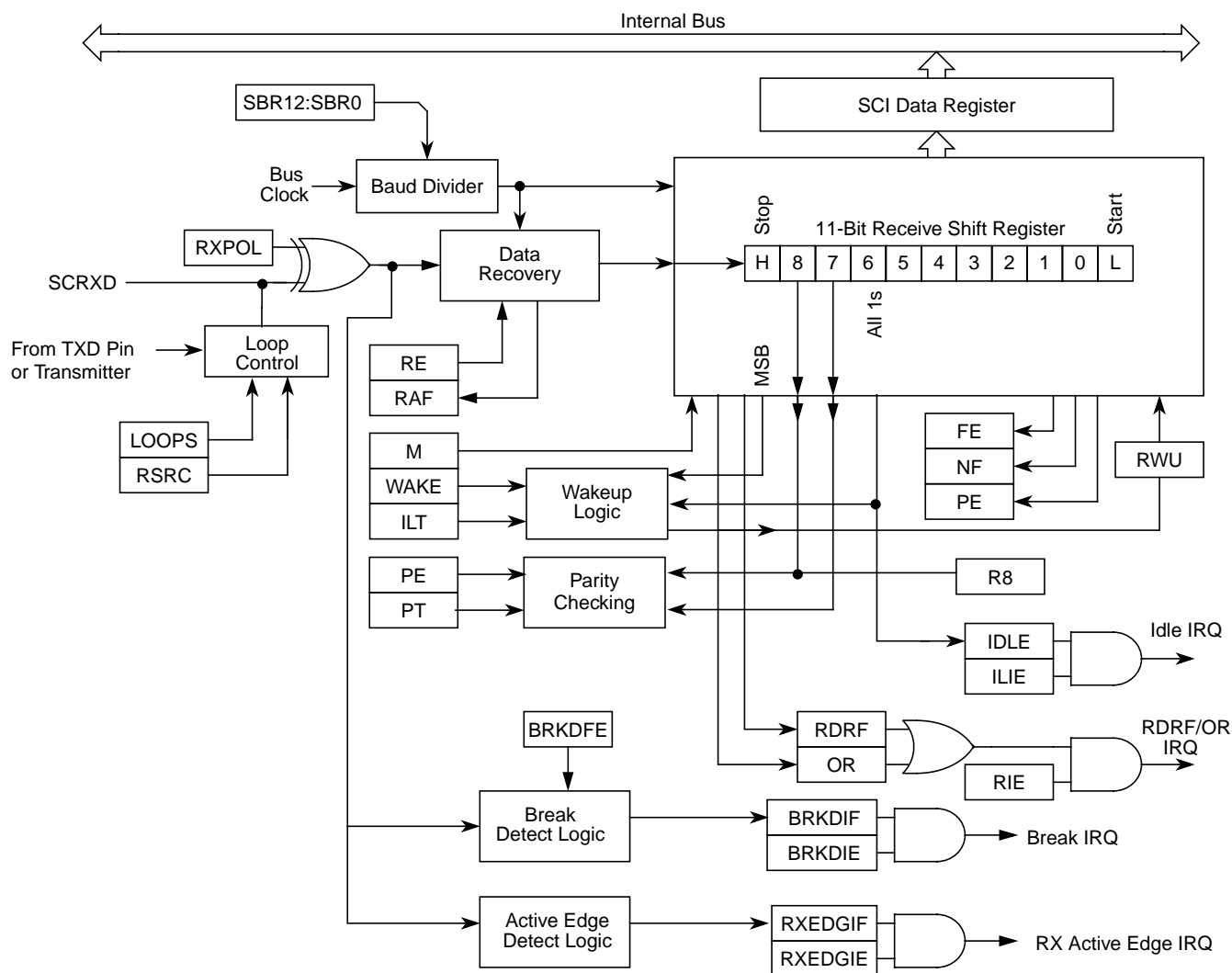


Figure 11-20. SCI Receiver Block Diagram

### 11.4.6.1 Receiver Character Length

The SCI receiver can accommodate either 8-bit or 9-bit data characters. The state of the M bit in SCI control register 1 (SCICR1) determines the length of data characters. When receiving 9-bit data, bit R8 in SCI data register high (SCIDRH) is the ninth bit (bit 8).

### 11.4.6.2 Character Reception

During an SCI reception, the receive shift register shifts a frame in from the RXD pin. The SCI data register is the read-only buffer between the internal data bus and the receive shift register.

After a complete frame shifts into the receive shift register, the data portion of the frame transfers to the SCI data register. The receive data register full flag, RDRF, in SCI status register 1 (SCISR1) becomes set,

Table 12-2. SPICR1 Field Descriptions (continued)

Field	Description
1 SSOE	<b>Slave Select Output Enable</b> — The $\overline{SS}$ output feature is enabled only in master mode, if MODFEN is set, by asserting the SSOE as shown in Table 12-3. In master mode, a change of this bit will abort a transmission in progress and force the SPI system into idle state.
0 LSBFE	<b>LSB-First Enable</b> — This bit does not affect the position of the MSB and LSB in the data register. Reads and writes of the data register always have the MSB in the highest bit position. In master mode, a change of this bit will abort a transmission in progress and force the SPI system into idle state. 0 Data is transferred most significant bit first. 1 Data is transferred least significant bit first.

Table 12-3.  $\overline{SS}$  Input / Output Selection

MODFEN	SSOE	Master Mode	Slave Mode
0	0	$\overline{SS}$ not used by SPI	$\overline{SS}$ input
0	1	$\overline{SS}$ not used by SPI	$\overline{SS}$ input
1	0	$\overline{SS}$ input with MODF feature	$\overline{SS}$ input
1	1	$\overline{SS}$ is slave select output	$\overline{SS}$ input

12.3.2.2 SPI Control Register 2 (SPICR2)

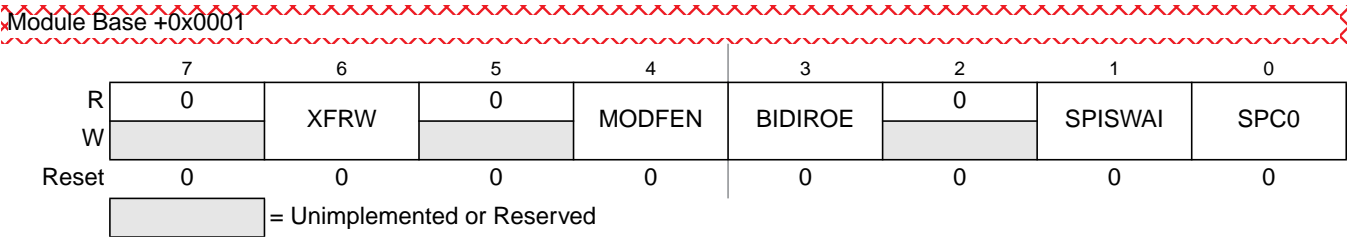


Figure 12-4. SPI Control Register 2 (SPICR2)

Read: Anytime

Write: Anytime; writes to the reserved bits have no effect



drive the MOSI and SCK lines. In this case, the SPI immediately switches to slave mode, by clearing the MSTR bit and also disables the slave output buffer MISO (or SISO in bidirectional mode). So the result is that all outputs are disabled and SCK, MOSI, and MISO are inputs. If a transmission is in progress when the mode fault occurs, the transmission is aborted and the SPI is forced into idle state.

This mode fault error also sets the mode fault (MODF) flag in the SPI status register (SPISR). If the SPI interrupt enable bit (SPIE) is set when the MODF flag becomes set, then an SPI interrupt sequence is also requested.

When a write to the SPI data register in the master occurs, there is a half SCK-cycle delay. After the delay, SCK is started within the master. The rest of the transfer operation differs slightly, depending on the clock format specified by the SPI clock phase bit, CPHA, in SPI control register 1 (see [Section 12.4.3, “Transmission Formats”](#)).

### NOTE

A change of the bits CPOL, CPHA, SSOE, LSBFE, XFRW, MODFEN, SPC0, or BIDIROE with SPC0 set, SPPR2-SPPR0 and SPR2-SPR0 in master mode will abort a transmission in progress and force the SPI into idle state. The remote slave cannot detect this, therefore the master must ensure that the remote slave is returned to idle state.

## 12.4.2 Slave Mode

The SPI operates in slave mode when the MSTR bit in SPI control register 1 is clear.

- Serial clock

In slave mode, SCK is the SPI clock input from the master.

- MISO, MOSI pin

In slave mode, the function of the serial data output pin (MISO) and serial data input pin (MOSI) is determined by the SPC0 bit and BIDIROE bit in SPI control register 2.

- $\overline{SS}$  pin

The  $\overline{SS}$  pin is the slave select input. Before a data transmission occurs, the  $\overline{SS}$  pin of the slave SPI must be low.  $\overline{SS}$  must remain low until the transmission is complete. If  $\overline{SS}$  goes high, the SPI is forced into idle state.

The  $\overline{SS}$  input also controls the serial data output pin, if  $\overline{SS}$  is high (not selected), the serial data output pin is high impedance, and, if  $\overline{SS}$  is low, the first bit in the SPI data register is driven out of the serial data output pin. Also, if the slave is not selected ( $\overline{SS}$  is high), then the SCK input is ignored and no internal shifting of the SPI shift register occurs.

Although the SPI is capable of duplex operation, some SPI peripherals are capable of only receiving SPI data in a slave mode. For these simpler devices, there is no serial data out pin.

### NOTE

When peripherals with duplex capability are used, take care not to simultaneously enable two receivers whose serial outputs drive the same system slave's serial data output line.

## A.1.10 Supply Currents

This section describes the current consumption characteristics of the device as well as the conditions for the measurements.

### A.1.10.1 Measurement Conditions

Run current is measured on VDDR pin. It does not include the current to drive external loads. Unless otherwise noted the currents are measured in special single chip mode and the CPU code is executed from RAM. For Run and Wait current measurements PLL is on and the reference clock is the IRC1M trimmed to 1MHz. The bus frequency is 32MHz and the CPU frequency is 64MHz. Table A-8., Table A-9. and Table A-10. show the configuration of the CPMU module and the peripherals for Run, Wait and Stop current measurement.

**Table A-8. CPMU Configuration for Pseudo Stop Current Measurement**

CPMU REGISTER	Bit settings/Conditions
CPMUCLKS	PLLSEL=0, PSTP=1, PRE=PCE=RTIOSCSEL=COPOSCSEL=1
CPMUOSC	OSCE=1, External Square wave on EXTAL $f_{EXTAL}=16\text{MHz}$ , $V_{IH}=1.8\text{V}$ , $V_{IL}=0\text{V}$
CPMURTI	RTDEC=0, RTR[6:4]=111, RTR[3:0]=1111;
CPMUCOP	WCOP=1, CR[2:0]=111

**Table A-9. CPUM Configuration for Run/Wait and Full Stop Current Measurement**

CPMU REGISTER	Bit settings/Conditions
CPMUSYNR	VCOFRQ[1:0]=01, SYNDIV[5:0] = 32
CPMUPOSTDIV	POSTDIV[4:0]=0,
CPMUCLKS	PLLSEL=1
CPMUOSC	OSCE=0, Reference clock for PLL is $f_{ref}=f_{irc1m}$ trimmed to 1MHz
API settings for STOP current measurement	
CPMUAPICTL	APIEA=0, APIFE=1, APIE=0
CPMUAPITR	trimmed to 10Khz
CPMUAPIRH/RL	set to \$FFFF

## NOTE

All values shown in [Table A-19](#) are preliminary and subject to further characterization.

**Table A-19. NVM Reliability Characteristics**

Conditions are shown in <a href="#">Table A-4</a> unless otherwise noted							
Num	C	Rating	Symbol	Min	Typ	Max	Unit
<b>Program Flash Arrays</b>							
1	C	Data retention at an average junction temperature of $T_{Javg} = 85^{\circ}\text{C}^{(1)}$ after up to 10,000 program/erase cycles	$t_{NVMRET}$	20	$100^{(2)}$	—	Years
2	C	Program Flash number of program/erase cycles ( $-40^{\circ}\text{C} \leq t_j \leq 150^{\circ}\text{C}$ )	$n_{FLPE}$	10K	$100\text{K}^{(3)}$	—	Cycles
<b>Data Flash Array</b>							
3	C	Data retention at an average junction temperature of $T_{Javg} = 85^{\circ}\text{C}^1$ after up to 50,000 program/erase cycles	$t_{NVMRET}$	5	$100^2$	—	Years
4	C	Data retention at an average junction temperature of $T_{Javg} = 85^{\circ}\text{C}^1$ after up to 10,000 program/erase cycles	$t_{NVMRET}$	10	$100^2$	—	Years
5	C	Data retention at an average junction temperature of $T_{Javg} = 85^{\circ}\text{C}^1$ after less than 100 program/erase cycles	$t_{NVMRET}$	20	$100^2$	—	Years
6	C	Data Flash number of program/erase cycles ( $-40^{\circ}\text{C} \leq t_j \leq 150^{\circ}\text{C}$ )	$n_{FLPE}$	50K	$500\text{K}^3$	—	Cycles

1.  $T_{Javg}$  does not exceed  $85^{\circ}\text{C}$  in a typical temperature profile over the lifetime of a consumer, industrial or automotive application.

2. Typical data retention values are based on intrinsic capability of the technology measured at high temperature and de-rated to  $25^{\circ}\text{C}$  using the Arrhenius equation. For additional information on how Freescale defines Typical Data Retention, please refer to Engineering Bulletin EB618

3. Spec table quotes typical endurance evaluated at  $25^{\circ}\text{C}$  for this product family. For additional information on how Freescale defines Typical Endurance, please refer to Engineering Bulletin EB619.

## A.4 Phase Locked Loop

### A.4.1 Jitter Definitions

With each transition of the feedback clock, the deviation from the reference clock is measured and input voltage to the VCO is adjusted accordingly. The adjustment is done continuously with no abrupt changes in the VCOCLK frequency. Noise, voltage, temperature and other factors cause slight variations in the control loop resulting in a clock jitter. This jitter affects the real minimum and maximum clock periods as illustrated in [Figure A-2](#).

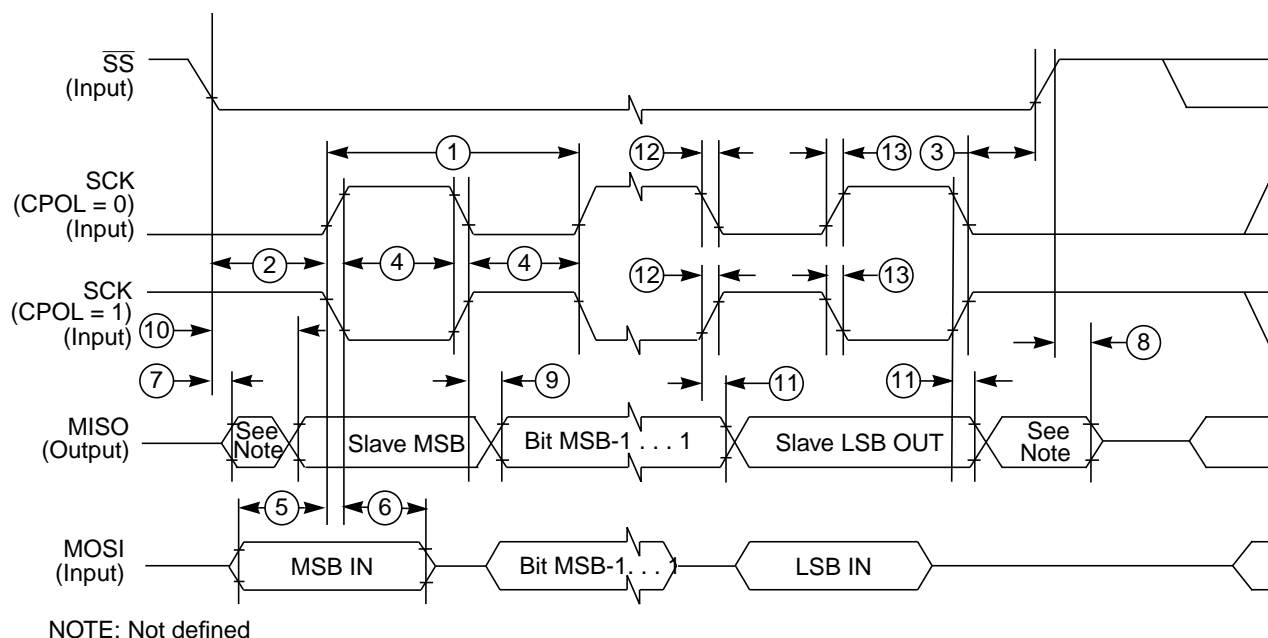
In Table A-27 the timing characteristics for master mode are listed.

**Table A-27. SPI Master Mode Timing Characteristics**

Num	C	Characteristic	Symbol	Min	Typ	Max	Unit
1	D	SCK frequency	$f_{sck}$	1/2048	—	1/2	$f_{bus}$
1	D	SCK period	$t_{sck}$	2	—	2048	$t_{bus}$
2	D	Enable lead time	$t_{lead}$	—	1/2	—	$t_{sck}$
3	D	Enable lag time	$t_{lag}$	—	1/2	—	$t_{sck}$
4	D	Clock (SCK) high or low time	$t_{wsck}$	—	1/2	—	$t_{sck}$
5	D	Data setup time (inputs)	$t_{su}$	8	—	—	ns
6	D	Data hold time (inputs)	$t_{hi}$	8	—	—	ns
9	D	Data valid after SCK edge	$t_{vsck}$	—	—	29	ns
10	D	Data valid after $\overline{SS}$ fall (CPHA = 0)	$t_{vss}$	—	—	15	ns
11	D	Data hold time (outputs)	$t_{ho}$	20	—	—	ns
12	D	Rise and fall time inputs	$t_{rfi}$	—	—	8	ns
13	D	Rise and fall time outputs	$t_{rfo}$	—	—	8	ns

## A.11.2 Slave Mode

In Figure A-7 the timing diagram for slave mode with transmission format CPHA = 0 is depicted.



**Figure A-7. SPI Slave Timing (CPHA = 0)**

## 0x0034-0x003F Clock Reset and Power Management (CPMU) Map

Address	Name		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x0038	CPMUIINT	R	RTIE	0	0	LOCKIE	0	0	OSCIE	0
		W								
0x0039	CPMUCLKS	R	PLLSEL	PSTP	0	0	PRE	PCE	RTIOSCS EL	COPOSC SEL
		W								
0x003A	CPMUPLL	R	0	0	FM1	FM0	0	0	0	0
		W								
0x003B	CPMURTI	R	RTDEC	RTR6	RTR5	RTR4	RTR3	RTR2	RTR1	RTR0
		W								
0x003C	CPMUCOP	R	WCOP	RSBCK	0	0	0	CR2	CR1	CR0
		W			WRTMAS K					
0x003D	Reserved	R	0	0	0	0	0	0	0	0
		W	Reserved For Factory Test							
0x003E	Reserved	R	0	0	0	0	0	0	0	0
		W	Reserved For Factory Test							
0x003F	CPMU ARMCOP	R	0	0	0	0	0	0	0	0
		W	Bit 7	6	5	4	3	2	1	Bit 0

## 0x0040-0x006F Timer Module (TIM) Map

Address	Name		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x0040	TIOS	R	IOS7	IOS6	IOS5	IOS4	IOS3	IOS2	IOS1	IOS0
		W								
0x0041	CFORC	R	0	0	0	0	0	0	0	0
		W	FOC7	FOC6	FOC5	FOC4	FOC3	FOC2	FOC1	FOC0
0x0042	OC7M	R	OC7M7	OC7M6	OC7M5	OC7M4	OC7M3	OC7M2	OC7M1	OC7M0
		W								
0x0043	OC7D	R	OC7D7	OC7D6	OC7D5	OC7D4	OC7D3	OC7D2	OC7D1	OC7D0
		W								
0x0044	TCNTH	R	Bit 15	14	13	12	11	10	9	Bit 8
		W								
0x0045	TCNTL	R	Bit 7	6	5	4	3	2	1	Bit 0
		W								
0x0046	TSCR1	R	TEN	TSWAI	TSFRZ	TFFCA	PRNT	0	0	0
		W								
0x0047	TTOV	R	TOV7	TOV6	TOV5	TOV4	TOV3	TOV2	TOV1	TOV0
		W								
0x0048	TCTL1	R	OM7	OL7	OM6	OL6	OM5	OL5	OM4	OL4
		W								
0x0049	TCTL2	R	OM3	OL3	OM2	OL2	OM1	OL1	OM0	OL0
		W								
0x004A	TCTL3	R	EDG7B	EDG7A	EDG6B	EDG6A	EDG5B	EDG5A	EDG4B	EDG4A
		W								
0x004B	TCTL4	R	EDG3B	EDG3A	EDG2B	EDG2A	EDG1B	EDG1A	EDG0B	EDG0A
		W								