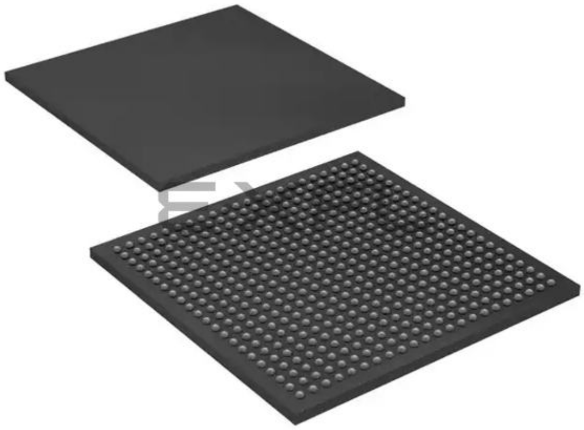
Intel - EP1S10F484I6 Datasheet



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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1057
Number of Logic Elements/Cells	10570
Total RAM Bits	920448
Number of I/O	335
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	484-BBGA, FCBGA
Supplier Device Package	484-FBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1s10f484i6

Section I. Stratix Device Family Data Sheet

This section provides the data sheet specifications for Stratix® devices. They contain feature definitions of the internal architecture, configuration and JTAG boundary-scan testing information, DC operating conditions, AC timing parameters, a reference to power consumption, and ordering information for Stratix devices.

This section contains the following chapters:

Revision History

The table below shows the revision history for Chapters 1 through 5.

Chapter	Date/Version	Changes Made
1	July 2005, v3.2	● Minor content changes.
	September 2004, v3.1	● Updated Table 1–6 on page 1–5.
	April 2004, v3.0	● Main section page numbers changed on first page. ● Changed PCI-X to PCI-X 1.0 in "Features" on page 1–2. ● Global change from SignalTap to SignalTap II. ● The DSP blocks in "Features" on page 1–2 provide dedicated implementation of multipliers that are now "faster than 300 MHz."
	January 2004, v2.2	● Updated -5 speed grade device information in Table 1-6.
	October 2003, v2.1	● Add -8 speed grade device information.
	July 2003, v2.0	● Format changes throughout chapter.

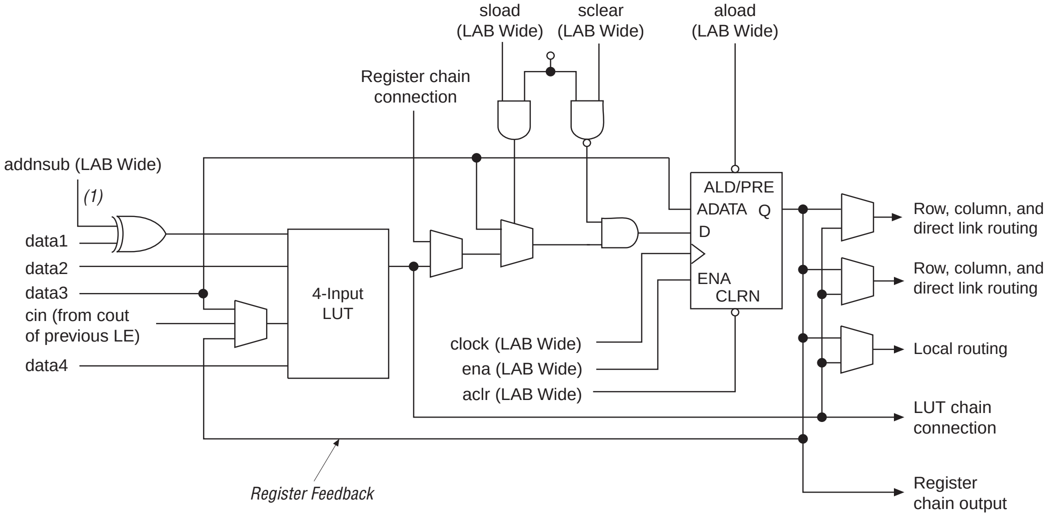
asynchronous preset load, synchronous clear, synchronous load, and clock enable control for the register. These LAB-wide signals are available in all LE modes. The `addnsub` control signal is allowed in arithmetic mode.

The Quartus II software, in conjunction with parameterized functions such as library of parameterized modules (LPM) functions, automatically chooses the appropriate mode for common functions such as counters, adders, subtractors, and arithmetic functions. If required, you can also create special-purpose functions that specify which LE operating mode to use for optimal performance.

Normal Mode

The normal mode is suitable for general logic applications and combinatorial functions. In normal mode, four data inputs from the LAB local interconnect are inputs to a four-input LUT (see Figure 2–6). The Quartus II Compiler automatically selects the carry-in or the `data3` signal as one of the inputs to the LUT. Each LE can use LUT chain connections to drive its combinatorial output directly to the next LE in the LAB. Asynchronous load data for the register comes from the `data3` input of the LE. LEs in normal mode support packed registers.

Figure 2–6. LE in Normal Mode



Note to Figure 2–6:

(1) This signal is only allowed in normal mode if the LE is at the end of an adder/subtractor chain.

C8 interconnects span eight LABs, M512, or M4K blocks up or down from a source LAB. Every LAB has its own set of C8 interconnects to drive either up or down. C8 interconnect connections between the LABs in a column are similar to the C4 connections shown in Figure 2–11 with the exception that they connect to eight LABs above and below. The C8 interconnects can drive and be driven by all types of architecture blocks similar to C4 interconnects. C8 interconnects can drive each other to extend their range as well as R8 interconnects for column-to-column connections. C8 interconnects are faster than two C4 interconnects.

C16 column interconnects span a length of 16 LABs and provide the fastest resource for long column connections between LABs, TriMatrix memory blocks, DSP blocks, and IOEs. C16 interconnects can cross M-RAM blocks and also drive to row and column interconnects at every fourth LAB. C16 interconnects drive LAB local interconnects via C4 and R4 interconnects and do not drive LAB local interconnects directly.

All embedded blocks communicate with the logic array similar to LAB-to-LAB interfaces. Each block (i.e., TriMatrix memory and DSP blocks) connects to row and column interconnects and has local interconnect regions driven by row and column interconnects. These blocks also have direct link interconnects for fast connections to and from a neighboring LAB. All blocks are fed by the row LAB clocks, `labclk[7..0]`.

Table 2–9. M-RAM Block Configurations (True Dual-Port)

Port A	Port B			
	64K × 9	32K × 18	16K × 36	8K × 72
64K × 9	✓	✓	✓	✓
32K × 18	✓	✓	✓	✓
16K × 36	✓	✓	✓	✓
8K × 72	✓	✓	✓	✓

The read and write operation of the memory is controlled by the `WREN` signal, which sets the ports into either read or write modes. There is no separate read enable (`RE`) signal.

Writing into RAM is controlled by both the `WREN` and byte enable (`byteena`) signals for each port. The default value for the `byteena` signal is high, in which case writing is controlled only by the `WREN` signal. The byte enables are available for the ×18, ×36, and ×72 modes. In the ×144 simple dual-port mode, the two sets of `byteena` signals (`byteena_a` and `byteena_b`) are combined to form the necessary 16 byte enables. Tables 2–10 and 2–11 summarize the byte selection.

Table 2–10. Byte Enable for M-RAM Blocks Notes (1), (2)

byteena[3..0]	datain ×18	datain ×36	datain ×72
[0] = 1	[8..0]	[8..0]	[8..0]
[1] = 1	[17..9]	[17..9]	[17..9]
[2] = 1	–	[26..18]	[26..18]
[3] = 1	–	[35..27]	[35..27]
[4] = 1	–	–	[44..36]
[5] = 1	–	–	[53..45]
[6] = 1	–	–	[62..54]
[7] = 1	–	–	[71..63]

Table 2–11. M-RAM Combined Byte Selection for ×144 Mode Notes (1), (2)

byteena[15..0]	datain ×144
[0] = 1	[8..0]
[1] = 1	[17..9]
[2] = 1	[26..18]
[3] = 1	[35..27]
[4] = 1	[44..36]
[5] = 1	[53..45]
[6] = 1	[62..54]
[7] = 1	[71..63]
[8] = 1	[80..72]
[9] = 1	[89..81]
[10] = 1	[98..90]
[11] = 1	[107..99]
[12] = 1	[116..108]
[13] = 1	[125..117]
[14] = 1	[134..126]
[15] = 1	[143..135]

Notes to Tables 2–10 and 2–11:

(1) Any combination of byte enables is possible.

(2) Byte enables can be used in the same manner with 8-bit words, i.e., in ×16, ×32, ×64, and ×128 modes.

Similar to all RAM blocks, M-RAM blocks can have different clocks on their inputs and outputs. All input registers—`renwe`, `datain`, address, and byte enable registers—are clocked together from either of the two clocks feeding the block. The output register can be bypassed. The eight `labclk` signals or local interconnect can drive the control signals for the A and B ports of the M-RAM block. LEs can also control the `clock_a`, `clock_b`, `renwe_a`, `renwe_b`, `clr_a`, clr_b, `clocken_a`, and `clocken_b` signals as shown in Figure 2–19.

The variation due to process, voltage, and temperature is about ±15% on the delay settings. PLL reconfiguration can control the clock delay shift elements, but not the VCO phase shift multiplexers, during system operation.

Spread-Spectrum Clocking

Stratix device enhanced PLLs use spread-spectrum technology to reduce electromagnetic interference generation from a system by distributing the energy over a broader frequency range. The enhanced PLL typically provides 0.5% down spread modulation using a triangular profile. The modulation frequency is programmable. Enabling spread-spectrum for a PLL affects all of its outputs.

Lock Detect

The lock output indicates that there is a stable clock output signal in phase with the reference clock. Without any additional circuitry, the lock signal may toggle as the PLL begins tracking the reference clock. You may need to gate the lock signal for use as a system control. The lock signal from the locked port can drive the logic array or an output pin.

Whenever the PLL loses lock (for example, `inclk` jitter, clock switchover, PLL reconfiguration, power supply noise, and so on), the PLL must be reset with the `areset` signal to guarantee correct phase relationship between the PLL output clocks. If the phase relationship between the input clock versus output clock, and between different output clocks from the PLL is not important in the design, then the PLL need not be reset.

See the *Stratix FPGA Errata Sheet* for more information on implementing the gated lock signal in a design.

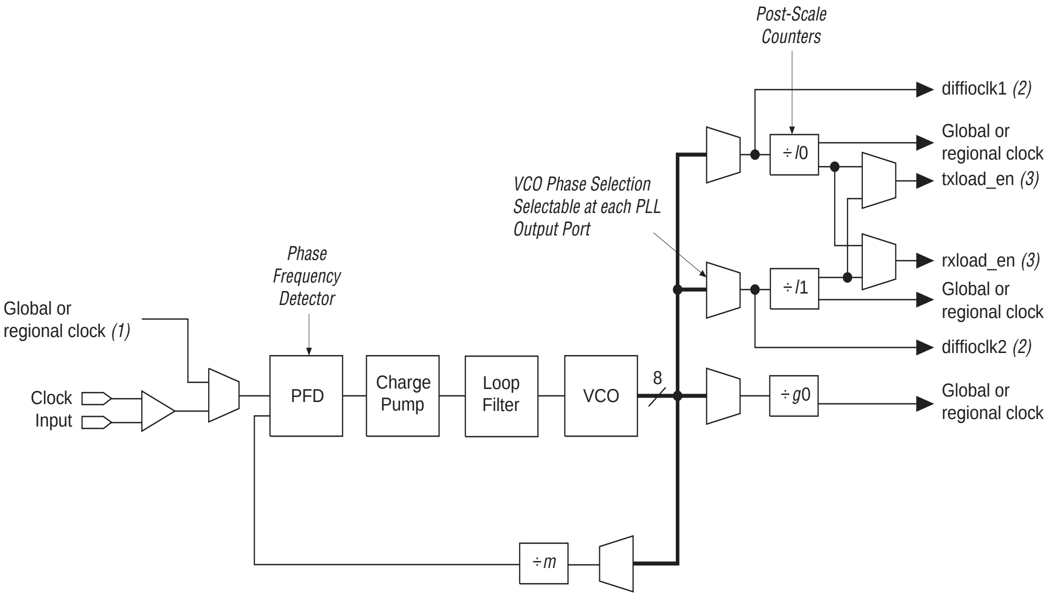
Programmable Duty Cycle

The programmable duty cycle allows enhanced PLLs to generate clock outputs with a variable duty cycle. This feature is supported on each enhanced PLL post-scale counter (*g*0..*g*3, *l*0..*l*3, *e*0..*e*3). The duty cycle setting is achieved by a low and high time count setting for the post-scale dividers. The Quartus II software uses the frequency input and the required multiply or divide rate to determine the duty cycle choices.

Advanced Clear & Enable Control

There are several control signals for clearing and enabling PLLs and their outputs. You can use these signals to control PLL resynchronization and gate PLL output clocks for low-power applications.

Figure 2–58. Stratix Device Fast PLL



Notes to Figure 2–58:

(1) The global or regional clock input can be driven by an output from another PLL or any dedicated CLK or FCLK pin. It cannot be driven by internally-generated global signals.

(2) In high-speed differential I/O support mode, this high-speed PLL clock feeds the SERDES. Stratix devices only support one rate of data transfer per fast PLL in high-speed differential I/O support mode.

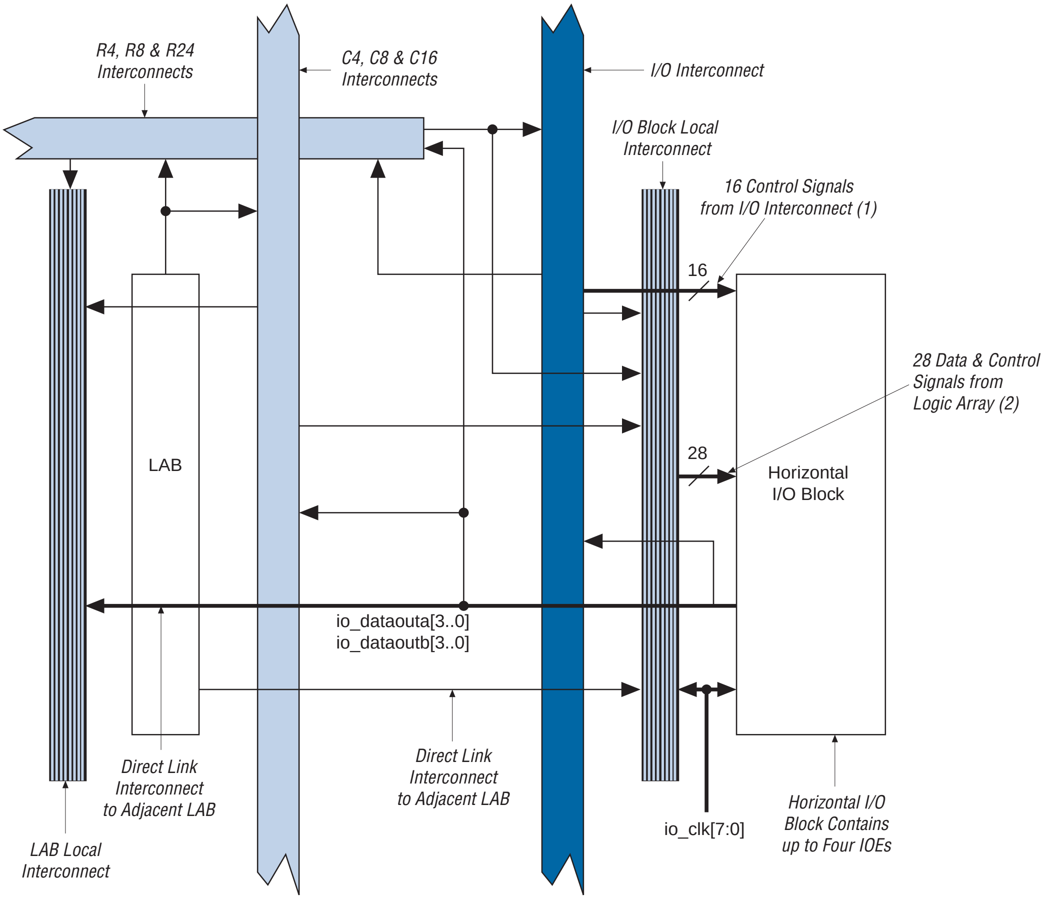
(3) This signal is a high-speed differential I/O support SERDES control signal.

Clock Multiplication & Division

Stratix device fast PLLs provide clock synthesis for PLL output ports using *m*/(post scaler) scaling factors. The input clock is multiplied by the *m* feedback factor. Each output port has a unique post scale counter to divide down the high-frequency VCO. There is one multiply divider, *m*, per fast PLL with a range of 1 to 32. There are two post scale L dividers for regional and/or LVDS interface clocks, and *g*0 counter for global clock output port; all range from 1 to 32.

In the case of a high-speed differential interface, set the output counter to 1 to allow the high-speed VCO frequency to drive the SERDES. When used for clocking the SERDES, the *m* counter can range from 1 to 30. The VCO frequency is equal to $f_{IN} \times m$, where VCO frequency must be between 300 and 1000 MHz.

Figure 2–60. Row I/O Block Connection to the Interconnect



Notes to Figure 2–60:

(1) The 16 control signals are composed of four output enables `io_boe[3..0]`, four clock enables `io_bce[3..0]`, four clocks `io_clk[3..0]`, and four clear signals `io_bclr[3..0]`.

(2) The 28 data and control signals consist of eight data out lines: four lines each for DDR applications `io_dataouta[3..0]` and `io_dataoutb[3..0]`, four output enables `io_coe[3..0]`, four input clock enables `io_cce_in[3..0]`, four output clock enables `io_cce_out[3..0]`, four clocks `io_cclk[3..0]`, and four clear signals `io_cclr[3..0]`.

Table 2–32. I/O Support by Bank (Part 2 of 2)

I/O Standard	Top & Bottom Banks (3, 4, 7 & 8)	Left & Right Banks (1, 2, 5 & 6)	Enhanced PLL External Clock Output Banks (9, 10, 11 & 12)
SSTL-3 Class II	✓	✓	✓
AGP (1× and 2×)	✓		✓
CTT	✓	✓	✓

Each I/O bank has its own `VCCIO` pins. A single device can support 1.5-, 1.8-, 2.5-, and 3.3-V interfaces; each bank can support a different standard independently. Each bank also has dedicated `VREF` pins to support any one of the voltage-referenced standards (such as SSTL-3) independently.

Each I/O bank can support multiple standards with the same V_{CCIO} for input and output pins. Each bank can support one voltage-referenced I/O standard. For example, when V_{CCIO} is 3.3 V, a bank can support LVTTL, LVCMOS, 3.3-V PCI, and SSTL-3 for inputs and outputs.

Differential On-Chip Termination

Stratix devices provide differential on-chip termination (LVDS I/O standard) to reduce reflections and maintain signal integrity. Differential on-chip termination simplifies board design by minimizing the number of external termination resistors required. Termination can be placed inside the package, eliminating small stubs that can still lead to reflections. The internal termination is designed using transistors in the linear region of operation.

Stratix devices support internal differential termination with a nominal resistance value of 137.5 Ω for LVDS input receiver buffers. LVPECL signals require an external termination resistor. Figure 2–71 shows the device with differential termination.

Figure 3–1 shows the timing requirements for the JTAG signals.

Figure 3–1. Stratix JTAG Waveforms

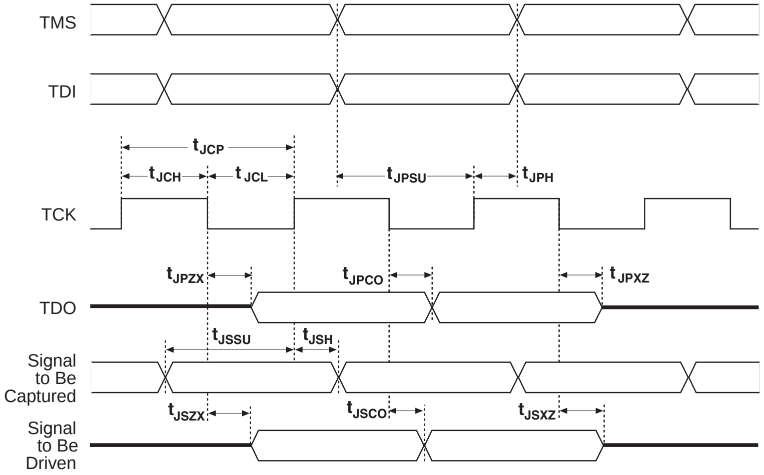


Table 3–4 shows the JTAG timing parameters and values for Stratix devices.

Table 3–4. Stratix JTAG Timing Parameters & Values

Symbol	Parameter	Min	Max	Unit
t_{JCP}	`TCK` clock period	100		ns
t_{JCH}	`TCK` clock high time	50		ns
t_{JCL}	`TCK` clock low time	50		ns
t_{JPSU}	JTAG port setup time	20		ns
t_{JPH}	JTAG port hold time	45		ns
t_{JPCO}	JTAG port clock to output		25	ns
t_{JPZX}	JTAG port high impedance to valid output		25	ns
t_{JPXZ}	JTAG port valid output to high impedance		25	ns
t_{JSSU}	Capture register setup time	20		ns
t_{JSH}	Capture register hold time	45		ns
t_{JSCO}	Update register clock to output		35	ns
t_{JSZX}	Update register high impedance to valid output		35	ns
t_{JSXZ}	Update register valid output to high impedance		35	ns

Table 4–13. HyperTransport Technology Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	I/O supply voltage		2.375	2.5	2.625	V
V_{ID} (peak-to-peak)	Input differential voltage swing (single-ended)		300		900	mV
V_{ICM}	Input common mode voltage		300		900	mV
V_{OD}	Output differential voltage (single-ended)	$R_L = 100\ \Omega$	380	485	820	mV
ΔV_{OD}	Change in V_{OD} between high and low	$R_L = 100\ \Omega$			50	mV
V_{OCM}	Output common mode voltage	$R_L = 100\ \Omega$	440	650	780	mV
ΔV_{OCM}	Change in V_{OCM} between high and low	$R_L = 100\ \Omega$			50	mV
R_L	Receiver differential input resistor		90	100	110	Ω

Table 4–14. 3.3-V PCI Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		3.0	3.3	3.6	V
V_{IH}	High-level input voltage		$0.5 \times V_{CCIO}$		$V_{CCIO} + 0.5$	V
V_{IL}	Low-level input voltage		–0.5		$0.3 \times V_{CCIO}$	V
V_{OH}	High-level output voltage	$I_{OUT} = -500\ \mu A$	$0.9 \times V_{CCIO}$			V
V_{OL}	Low-level output voltage	$I_{OUT} = 1{,}500\ \mu A$			$0.1 \times V_{CCIO}$	V

Table 4–18. SSTL-18 Class I Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		1.65	1.8	1.95	V
V_{REF}	Reference voltage		0.8	0.9	1.0	V
V_{TT}	Termination voltage		V_{REF} – 0.04	V_{REF}	V_{REF} + 0.04	V
$V_{IH(DC)}$	High-level DC input voltage		V_{REF} + 0.125			V
$V_{IL(DC)}$	Low-level DC input voltage				V_{REF} – 0.125	V
$V_{IH(AC)}$	High-level AC input voltage		V_{REF} + 0.275			V
$V_{IL(AC)}$	Low-level AC input voltage				V_{REF} – 0.275	V
V_{OH}	High-level output voltage	I_{OH} = –6.7 mA *(3)*	V_{TT} + 0.475			V
V_{OL}	Low-level output voltage	I_{OL} = 6.7 mA *(3)*			V_{TT} – 0.475	V

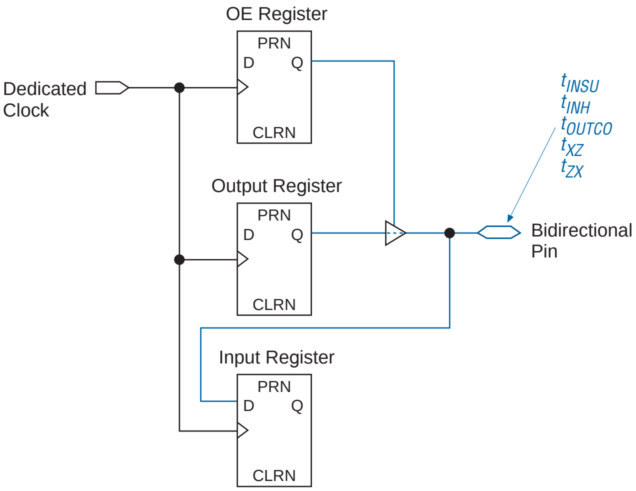
Table 4–19. SSTL-18 Class II Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		1.65	1.8	1.95	V
V_{REF}	Reference voltage		0.8	0.9	1.0	V
V_{TT}	Termination voltage		V_{REF} – 0.04	V_{REF}	V_{REF} + 0.04	V
$V_{IH(DC)}$	High-level DC input voltage		V_{REF} + 0.125			V
$V_{IL(DC)}$	Low-level DC input voltage				V_{REF} – 0.125	V
$V_{IH(AC)}$	High-level AC input voltage		V_{REF} + 0.275			V
$V_{IL(AC)}$	Low-level AC input voltage				V_{REF} – 0.275	V
V_{OH}	High-level output voltage	I_{OH} = –13.4 mA *(3)*	V_{TT} + 0.630			V
V_{OL}	Low-level output voltage	I_{OL} = 13.4 mA *(3)*			V_{TT} – 0.630	V

External Timing Parameters

External timing parameters are specified by device density and speed grade. Figure 4–4 shows the pin-to-pin timing model for bidirectional IOE pin timing. All registers are within the IOE.

Figure 4–4. External Timing in Stratix Devices



All external timing parameters reported in this section are defined with respect to the dedicated clock pin as the starting point. All external I/O timing parameters shown are for 3.3-V LVTTL I/O standard with the 24-mA current strength and fast slew rate. For external I/O timing using standards other than LVTTL or for different current strengths, use the I/O standard input and output delay adders in Tables 4–103 through 4–108.

Tables 4–85 through 4–90 show the external timing parameters on column and row pins for EP1S60 devices.

Table 4–85. EP1S60 External I/O Timing on Column Pins Using Fast Regional Clock Networks Note (1)

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	3.029		3.277		3.733		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.446	4.871	2.446	5.215	2.446	5.685	NA	NA	ns
t_{XZ}	2.386	4.745	2.386	5.083	2.386	5.561	NA	NA	ns
t_{ZX}	2.386	4.745	2.386	5.083	2.386	5.561	NA	NA	ns

Table 4–86. EP1S60 External I/O Timing on Column Pins Using Regional Clock Networks Note (1)

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.491		2.691		3.060		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.767	5.409	2.767	5.801	2.767	6.358	NA	NA	ns
t_{XZ}	2.707	5.283	2.707	5.669	2.707	6.234	NA	NA	ns
t_{ZX}	2.707	5.283	2.707	5.669	2.707	6.234	NA	NA	ns
$t_{INSUPLL}$	1.233		1.270		1.438		NA		ns
t_{INHPLL}	0.000		0.000		0.000		NA		ns
$t_{OUTCOPLL}$	1.078	2.278	1.078	2.395	1.078	2.428	NA	NA	ns
t_{XZPLL}	1.018	2.152	1.018	2.263	1.018	2.304	NA	NA	ns
t_{ZXPLL}	1.018	2.152	1.018	2.263	1.018	2.304	NA	NA	ns

the FPGA device. The Quartus II software calculates the I/O timing for each I/O standard with a default baseline loading as specified by the I/O standard.

Altera measures clock-to-output delays (t_{CO}) at worst-case process, minimum voltage, and maximum temperature (PVT) for the 3.3-V LVTTL I/O standard with 24 mA (default case) current drive strength setting and fast slew rate setting. I/O adder delays are measured to calculate the t_{CO} change at worst-case PVT across all I/O standards and current drive strength settings with the default loading shown in Table 4–101 on page 4–62. Timing derating data for additional loading is taken for t_{CO} across worst-case PVT for all I/O standards and drive strength settings. These three pieces of data are used to predict the timing at the output pin.

t_{CO} at pin = t_{OUTCO} max for 3.3-V 24 mA LVTTL + I/O Adder + Output Delay Adder for Loading

Simulation using IBIS models is required to determine the delays on the PCB traces in addition to the output pin delay timing reported by the Quartus II software and the timing model in the device handbook.

1. Simulate the output driver of choice into the generalized test setup using values from Table 4–101 on page 4–62.

2. Record the time to VMEAS.

3. Simulate the output driver of choice into the actual PCB trace and load, using the appropriate IBIS input buffer model or an equivalent capacitance value to represent the load.

4. Record the time to VMEAS.

5. Compare the results of steps 2 and 4. The increase or decrease in delay should be added to or subtracted from the I/O Standard Output Adder delays to yield the actual worst-case propagation delay (clock-to-input) of the PCB trace.

The Quartus II software reports maximum timing with the conditions shown in Table 4–101 on page 4–62 using the proceeding equation. Figure 4–7 on page 4–62 shows the model of the circuit that is represented by the Quartus II output timing.

Table 4–116. Stratix Maximum Input Clock Rate for CLK[1, 3, 8, 10] Pins in Flip-Chip Packages

I/O Standard	-5 Speed Grade	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	422	422	390	390	MHz
2.5 V	422	422	390	390	MHz
1.8 V	422	422	390	390	MHz
1.5 V	422	422	390	390	MHz
LVCMOS	422	422	390	390	MHz
GTL+	300	250	200	200	MHz
SSTL-3 Class I	400	350	300	300	MHz
SSTL-3 Class II	400	350	300	300	MHz
SSTL-2 Class I	400	350	300	300	MHz
SSTL-2 Class II	400	350	300	300	MHz
SSTL-18 Class I	400	350	300	300	MHz
SSTL-18 Class II	400	350	300	300	MHz
1.5-V HSTL Class I	400	350	300	300	MHz
1.8-V HSTL Class I	400	350	300	300	MHz
CTT	300	250	200	200	MHz
Differential 1.5-V HSTL C1	400	350	300	300	MHz
LVPECL *(1)*	645	645	640	640	MHz
PCML *(1)*	300	275	275	275	MHz
LVDS *(1)*	645	645	640	640	MHz
HyperTransport technology *(1)*	500	500	450	450	MHz

Table 4–117. Stratix Maximum Input Clock Rate for CLK[7..4] & CLK[15..12] Pins in Wire-Bond Packages (Part 1 of 2)

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	422	390	390	MHz
2.5 V	422	390	390	MHz
1.8 V	422	390	390	MHz
1.5 V	422	390	390	MHz
LVCMOS	422	390	390	MHz
GTL	250	200	200	MHz

Table 4–128. Enhanced PLL Specifications for -6 Speed Grades (Part 2 of 2)

Symbol	Parameter	Min	Typ	Max	Unit
$t_{SCANCLK}$	`scanclk` frequency *(5)*			22	MHz
t_{DLOCK}	Time required to lock dynamically (after switchover or reconfiguring any non-post-scale counters/delays) *(7)* *(11)*	*(9)*		100	µs
t_{LOCK}	Time required to lock from end of device configuration *(11)*	10		400	µs
f_{VCO}	PLL internal VCO operating range	300		800 *(8)*	MHz
t_{LSKEW}	Clock skew between two external clock outputs driven by the same counter		±50		ps
t_{SKEW}	Clock skew between two external clock outputs driven by the different counters with the same settings		±75		ps
f_{SS}	Spread spectrum modulation frequency	30		150	kHz
% spread	Percentage spread for spread spectrum frequency *(10)*	0.4	0.5	0.6	%
t_{ARESET}	Minimum pulse width on `areset` signal	10			ns

Table 4–129. Enhanced PLL Specifications for -7 Speed Grade (Part 1 of 2)

Symbol	Parameter	Min	Typ	Max	Unit
f_{IN}	Input clock frequency	3 *(1)*, *(2)*		565	MHz
f_{INPFD}	Input frequency to PFD	3		420	MHz
f_{INDUTY}	Input clock duty cycle	40		60	%
$f_{EINDUTY}$	External feedback clock input duty cycle	40		60	%
$t_{INJITTER}$	Input clock period jitter			±200 *(3)*	ps
$t_{EINJITTER}$	External feedback clock period jitter			±200 *(3)*	ps
t_{FCOMP}	External feedback clock compensation time *(4)*			6	ns
f_{OUT}	Output frequency for internal global or regional clock	0.3		420	MHz
f_{OUT_EXT}	Output frequency for external clock *(3)*	0.3		434	MHz

Table 4–132. Fast PLL Specifications for -7 Speed Grades (Part 2 of 2)

Symbol	Parameter	Min	Max	Unit
t_{JITTER}	Period jitter for DIFFIO clock out *(6)*		*(5)*	ps
t_{LOCK}	Time required for PLL to acquire lock	10	100	µs
m	Multiplication factors for *m* counter *(7)*	1	32	Integer
*l*0, *l*1, *g*0	Multiplication factors for *l*0, *l*1, and *g*0 counter *(7)*, *(8)*	1	32	Integer
t_{ARESET}	Minimum pulse width on `areset` signal	10		ns

Table 4–133. Fast PLL Specifications for -8 Speed Grades (Part 1 of 2)

Symbol	Parameter	Min	Max	Unit
f_{IN}	`CLKIN` frequency *(1)*, *(3)*	10	460	MHz
f_{INPFD}	Input frequency to PFD	10	500	MHz
f_{OUT}	Output frequency for internal global or regional clock *(4)*	9.375	420	MHz
f_{OUT_DIFFIO}	Output frequency for external clock driven out on a differential I/O data channel	*(5)*	*(5)*	MHz
f_{VCO}	VCO operating frequency	300	700	MHz
t_{INDUTY}	`CLKIN` duty cycle	40	60	%
$t_{INJITTER}$	Period jitter for `CLKIN` pin		±200	ps
t_{DUTY}	Duty cycle for `DFFIO` 1× `CLKOUT` pin *(6)*	45	55	%
t_{JITTER}	Period jitter for DIFFIO clock out *(6)*		*(5)*	ps
t_{LOCK}	Time required for PLL to acquire lock	10	100	µs
m	Multiplication factors for *m* counter *(7)*	1	32	Integer
*l*0, *l*1, *g*0	Multiplication factors for *l*0, *l*1, and *g*0 counter *(7)*, *(8)*	1	32	Integer