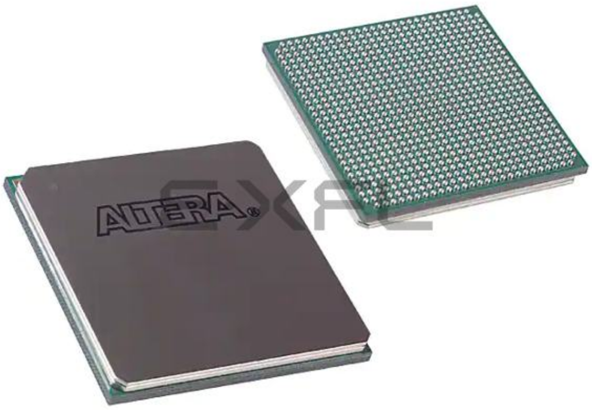

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	1057
Number of Logic Elements/Cells	10570
Total RAM Bits	920448
Number of I/O	426
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	780-BBGA
Supplier Device Package	780-FBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1s10f780c5n

Chapter	Date/Version	Changes Made
4	January 2005, 3.2	• Updated rise and fall input values.
	September 2004, v3.1	• Updated Note 3 in Table 4–8 on page 4–4. • Updated Table 4–10 on page 4–6. • Updated Table 4–20 on page 4–12 through Table 4–23 on page 4–13. Added rows $V_{IL(AC)}$ and $V_{IH(AC)}$ to each table. • Updated Table 4–26 on page 4–14 through Table 4–29 on page 4–15. • Updated Table 4–31 on page 4–16. • Updated description of "External Timing Parameters" on page 4–33. • Updated Table 4–36 on page 4–20. • Added signals t_{OUTCO}, T_{XZ}, and T_{ZX} to Figure 4–4 on page 4–33. • Added rows $t_{M512CLKENSU}$ and $t_{M512CLKENH}$ to Table 4–40 on page 4–24. • Added rows $t_{M4CLKENSU}$ and $t_{M4CLKENH}$ to Table 4–41 on page 4–24. • Updated Note 2 in Table 4–54 on page 4–35. • Added rows $t_{MRAMCLKENSU}$ and $t_{MRAMCLKENH}$ to Table 4–42 on page 4–25. • Updated Table 4–46 on page 4–29. • Updated Table 4–47 on page 4–29.

Figure 2–8 shows the carry-select circuitry in an LAB for a 10-bit full adder. One portion of the LUT generates the sum of two bits using the input signals and the appropriate carry-in bit; the sum is routed to the output of the LE. The register can be bypassed for simple adders or used for accumulator functions. Another portion of the LUT generates carry-out bits. An LAB-wide carry in bit selects which chain is used for the addition of given inputs. The carry-in signal for each chain, `carry-in0` or `carry-in1`, selects the carry-out to carry forward to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it is fed to local, row, or column interconnects.

The Quartus II Compiler automatically creates carry chain logic during design processing, or you can create it manually during design entry. Parameterized functions such as LPM functions automatically take advantage of carry chains for the appropriate functions.

The Quartus II Compiler creates carry chains longer than 10 LEs by linking LABs together automatically. For enhanced fitting, a long carry chain runs vertically allowing fast horizontal connections to TriMatrix™ memory and DSP blocks. A carry chain can continue as far as a full column.

can drive other R8 interconnects to extend their range as well as C8 interconnects for row-to-row connections. One R8 interconnect is faster than two R4 interconnects connected together.

R24 row interconnects span 24 LABs and provide the fastest resource for long row connections between LABs, TriMatrix memory, DSP blocks, and IOEs. The R24 row interconnects can cross M-RAM blocks. R24 row interconnects drive to other row or column interconnects at every fourth LAB and do not drive directly to LAB local interconnects. R24 row interconnects drive LAB local interconnects via R4 and C4 interconnects. R24 interconnects can drive R24, R4, C16, and C4 interconnects.

The column interconnect operates similarly to the row interconnect and vertically routes signals to and from LABs, TriMatrix memory, DSP blocks, and IOEs. Each column of LABs is served by a dedicated column interconnect, which vertically routes signals to and from LABs, TriMatrix memory and DSP blocks, and horizontal IOEs. These column resources include:

- LUT chain interconnects within an LAB
- Register chain interconnects within an LAB
- C4 interconnects traversing a distance of four blocks in up and down direction
- C8 interconnects traversing a distance of eight blocks in up and down direction
- C16 column interconnects for high-speed vertical routing through the device

Stratix devices include an enhanced interconnect structure within LABs for routing LE output to LE input connections faster using LUT chain connections and register chain connections. The LUT chain connection allows the combinatorial output of an LE to directly drive the fast input of the LE right below it, bypassing the local interconnect. These resources can be used as a high-speed connection for wide fan-in functions from LE 1 to LE 10 in the same LAB. The register chain connection allows the register output of one LE to connect directly to the register input of the next LE in the LAB for fast shift registers. The Quartus II Compiler automatically takes advantage of these resources to improve utilization and performance. Figure 2–10 shows the LUT chain and register chain interconnects.

TriMatrix memory architecture can implement pipelined RAM by registering both the input and output signals to the RAM block. All TriMatrix memory block inputs are registered providing synchronous write cycles. In synchronous operation, the memory block generates its own self-timed strobe write enable (WREN) signal derived from the global or regional clock. In contrast, a circuit using asynchronous RAM must generate the RAM WREN signal while ensuring its data and address signals meet setup and hold time specifications relative to the WREN signal. The output registers can be bypassed. Flow-through reading is possible in the simple dual-port mode of M512 and M4K RAM blocks by clocking the read enable and read address registers on the negative clock edge and bypassing the output registers.

Two single-port memory blocks can be implemented in a single M4K block as long as each of the two independent block sizes is equal to or less than half of the M4K block size.

The Quartus II software automatically implements larger memory by combining multiple TriMatrix memory blocks. For example, two 256 × 16-bit RAM blocks can be combined to form a 256 × 32-bit RAM block. Memory performance does not degrade for memory blocks using the maximum number of words available in one memory block. Logical memory blocks using less than the maximum number of words use physical blocks in parallel, eliminating any external control logic that would increase delays. To create a larger high-speed memory block, the Quartus II software automatically combines memory blocks with LE control logic.

Clear Signals

When applied to input registers, the asynchronous clear signal for the TriMatrix embedded memory immediately clears the input registers. However, the output of the memory block does not show the effects until the next clock edge. When applied to output registers, the asynchronous clear signal clears the output registers and the effects are seen immediately.

Parity Bit Support

The memory blocks support a parity bit for each byte. The parity bit, along with internal LE logic, can implement parity checking for error detection to ensure data integrity. You can also use parity-size data words to store user-specified control bits. In the M4K and M-RAM blocks, byte enables are also available for data input masking during write operations.

Shift Register Support

You can configure embedded memory blocks to implement shift registers for DSP applications such as pseudo-random number generators, multi-channel filtering, auto-correlation, and cross-correlation functions. These and other DSP applications require local data storage, traditionally implemented with standard flip-flops, which can quickly consume many logic cells and routing resources for large shift registers. A more efficient alternative is to use embedded memory as a shift register block, which saves logic cell and routing resources and provides a more efficient implementation with the dedicated circuitry.

The size of a $w \times m \times n$ shift register is determined by the input data width (w), the length of the taps (m), and the number of taps (n). The size of a $w \times m \times n$ shift register must be less than or equal to the maximum number of memory bits in the respective block: 576 bits for the M512 RAM block and 4,608 bits for the M4K RAM block. The total number of shift register outputs (number of taps $n \times$ width w) must be less than the maximum data width of the RAM block (18 for M512 blocks, 36 for M4K blocks). To create larger shift registers, the memory blocks are cascaded together.

Data is written into each address location at the falling edge of the clock and read from the address at the rising edge of the clock. The shift register mode logic automatically controls the positive and negative edge clocking to shift the data in one clock cycle. Figure 2–14 shows the TriMatrix memory block in the shift register mode.

Output Selection Multiplexer

The outputs from the various elements of the adder/output block are routed through an output selection multiplexer. Based on the DSP block operational mode and user settings, the multiplexer selects whether the output from the multiplier, the adder/subtractor/accumulator, or summation block feeds to the output.

Output Registers

Optional output registers for the DSP block outputs are controlled by four sets of control signals: `clock[3..0]`, `aclr[3..0]`, and `ena[3..0]`. Output registers can be used in any mode.

Modes of Operation

The adder, subtractor, and accumulate functions of a DSP block have four modes of operation:

- Simple multiplier
- Multiply-accumulator
- Two-multipliers adder
- Four-multipliers adder

☞ Each DSP block can only support one mode. Mixed modes in the same DSP block is not supported.

Simple Multiplier Mode

In simple multiplier mode, the DSP block drives the multiplier sub-block result directly to the output with or without an output register. Up to four 18×18-bit multipliers or eight 9×9-bit multipliers can drive their results directly out of one DSP block. See Figure 2–35.

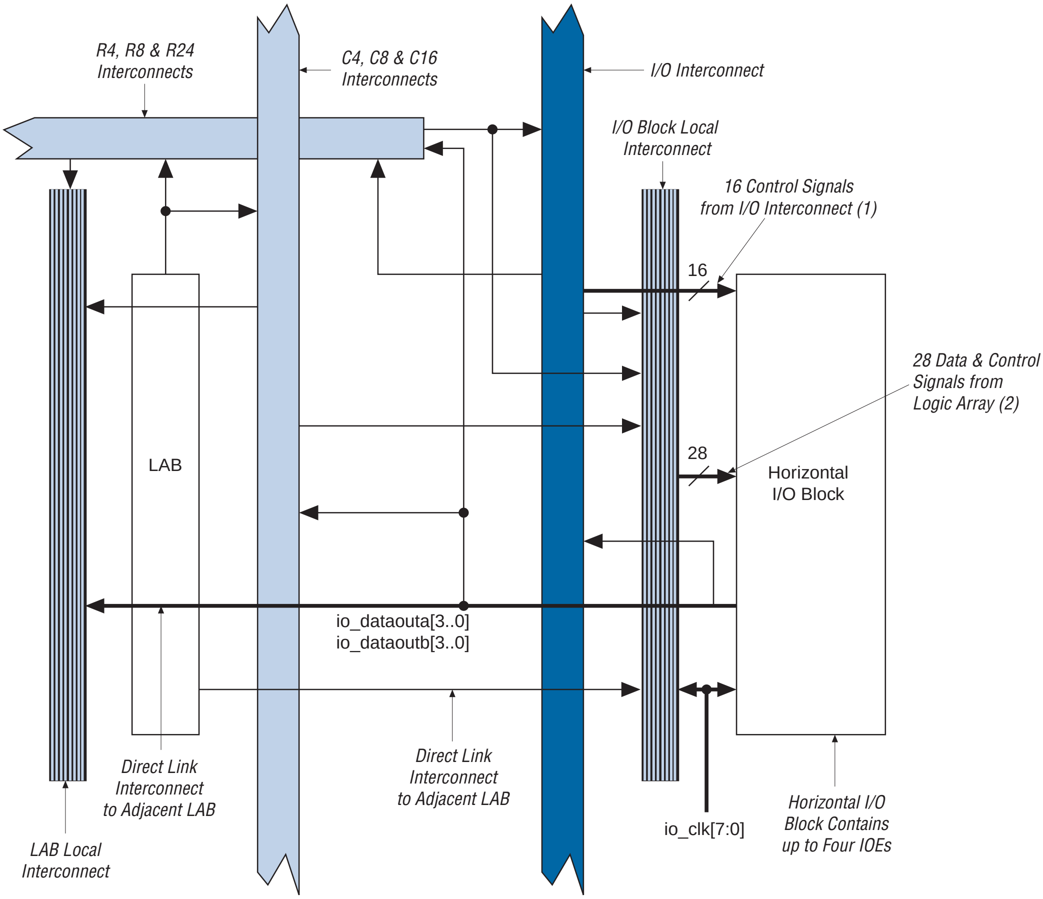
Clock Multiplication & Division

Each Stratix device enhanced PLL provides clock synthesis for PLL output ports using $m/(n \times \text{post-scale counter})$ scaling factors. The input clock is divided by a pre-scale divider, n, and is then multiplied by the m feedback factor. The control loop drives the VCO to match $f_{IN} \times (m/n)$. Each output port has a unique post-scale counter that divides down the high-frequency VCO. For multiple PLL outputs with different frequencies, the VCO is set to the least common multiple of the output frequencies that meets its frequency specifications. Then, the post-scale dividers scale down the output frequency for each output port. For example, if output frequencies required from one PLL are 33 and 66 MHz, set the VCO to 330 MHz (the least common multiple in the VCO's range). There is one pre-scale counter, n, and one multiply counter, m, per PLL, with a range of 1 to 512 on each. There are two post-scale counters (l) for regional clock output ports, four counters (g) for global clock output ports, and up to four counters (e) for external clock outputs, all ranging from 1 to 1024 with a 50% duty cycle setting. The post-scale counters range from 1 to 512 with any non-50% duty cycle setting. The Quartus II software automatically chooses the appropriate scaling factors according to the input frequency, multiplication, and division values entered.

Clock Switchover

To effectively develop high-reliability network systems, clocking schemes must support multiple clocks to provide redundancy. For this reason, Stratix device enhanced PLLs support a flexible clock switchover capability. Figure 2–53 shows a block diagram of the switchover circuit.The switchover circuit is configurable, so you can define how to implement it. Clock-sense circuitry automatically switches from the primary to secondary clock for PLL reference when the primary clock signal is not present.

Figure 2–60. Row I/O Block Connection to the Interconnect



Notes to Figure 2–60:

(1) The 16 control signals are composed of four output enables `io_boe[3..0]`, four clock enables `io_bce[3..0]`, four clocks `io_clk[3..0]`, and four clear signals `io_bclr[3..0]`.

(2) The 28 data and control signals consist of eight data out lines: four lines each for DDR applications `io_dataouta[3..0]` and `io_dataoutb[3..0]`, four output enables `io_coe[3..0]`, four input clock enables `io_cce_in[3..0]`, four output clock enables `io_cce_out[3..0]`, four clocks `io_cclk[3..0]`, and four clear signals `io_cclr[3..0]`.

Table 2–39. EP1S40 Differential Channels (Part 2 of 2) Note (1)											
Package	Transmitter/ Receiver	Total Channels	Maximum Speed (Mbps)	Center Fast PLLs				Corner Fast PLLs (2), (3)			
				PLL1	PLL2	PLL3	PLL4	PLL7	PLL8	PLL9	PLL10
956-pin BGA	Transmitter (4)	80	840	18	17	17	18	20	20	20	20
			840 (5)	35	35	35	35	20	20	20	20
	Receiver	80	840	20	20	20	20	18	17	17	18
			840 (5)	40	40	40	40	18	17	17	18
1,020-pin FineLine BGA	Transmitter (4)	80 (10) (7)	840	18 (2)	17 (3)	17 (3)	18 (2)	20	20	20	20
			840 (5), (8)	35 (5)	35 (5)	35 (5)	35 (5)	20	20	20	20
	Receiver	80 (10) (7)	840	20	20	20	20	18 (2)	17 (3)	17 (3)	18 (2)
			840 (5), (8)	40	40	40	40	18 (2)	17 (3)	17 (3)	18 (2)
1,508-pin FineLine BGA	Transmitter (4)	80 (10) (7)	840	18 (2)	17 (3)	17 (3)	18 (2)	20	20	20	20
			840 (5), (8)	35 (5)	35 (5)	35 (5)	35 (5)	20	20	20	20
	Receiver	80 (10) (7)	840	20	20	20	20	18 (2)	17 (3)	17 (3)	18 (2)
			840 (5), (8)	40	40	40	40	18 (2)	17 (3)	17 (3)	18 (2)

Table 2–40. EP1S60 Differential Channels (Part 1 of 2) Note (1)											
Package	Transmitter/ Receiver	Total Channels	Maximum Speed (Mbps)	Center Fast PLLs				Corner Fast PLLs (2), (3)			
				PLL1	PLL2	PLL3	PLL4	PLL7	PLL8	PLL9	PLL10
956-pin BGA	Transmitter (4)	80	840	12	10	10	12	20	20	20	20
			840 (5), (8)	22	22	22	22	20	20	20	20
	Receiver	80	840	20	20	20	20	12	10	10	12
			840 (5), (8)	40	40	40	40	12	10	10	12

Table 4–7. 1.8-V I/O Specifications					
Symbol	**Parameter**	**Conditions**	**Minimum**	**Maximum**	**Unit**
V_{CCIO}	Output supply voltage		1.65	1.95	V
V_{IH}	High-level input voltage		$0.65 \times V_{CCIO}$	2.25	V
V_{IL}	Low-level input voltage		–0.3	$0.35 \times V_{CCIO}$	V
V_{OH}	High-level output voltage	I_{OH} = –2 to –8 mA *(10)*	V_{CCIO} – 0.45		V
V_{OL}	Low-level output voltage	I_{OL} = 2 to 8 mA *(10)*		0.45	V

Table 4–8. 1.5-V I/O Specifications					
Symbol	**Parameter**	**Conditions**	**Minimum**	**Maximum**	**Unit**
V_{CCIO}	Output supply voltage		1.4	1.6	V
V_{IH}	High-level input voltage		$0.65 \times V_{CCIO}$	V_{CCIO} + 0.3	V
V_{IL}	Low-level input voltage		–0.3	$0.35 \times V_{CCIO}$	V
V_{OH}	High-level output voltage	I_{OH} = –2 mA *(10)*	$0.75 \times V_{CCIO}$		V
V_{OL}	Low-level output voltage	I_{OL} = 2 mA *(10)*		$0.25 \times V_{CCIO}$	V

Notes to Tables 4–1 through 4–8:

(1) See the *Operating Requirements for Altera Devices Data Sheet*.

(2) Conditions beyond those listed in Table 4–1 may cause permanent damage to a device. Additionally, device operation at the absolute maximum ratings for extended periods of time may have adverse affects on the device.

(3) Minimum DC input is –0.5 V. During transitions, the inputs may undershoot to –2.0 V for input currents less than 100 mA and periods shorter than 20 ns, or overshoot to the voltage shown in Table 4–9, based on input duty cycle for input currents less than 100 mA. The overshoot is dependent upon duty cycle of the signal. The DC case is equivalent to 100% duty cycle.

(4) Maximum V_{CC} rise time is 100 ms, and V_{CC} must rise monotonically.

(5) V_{CCIO} maximum and minimum conditions for LVPECL, LVDS, and 3.3-V PCML are shown in parentheses.

(6) All pins, including dedicated inputs, clock, I/O, and JTAG pins, may be driven before V_{CCINT} and V_{CCIO} are powered.

(7) Typical values are for T_A = 25°C, V_{CCINT} = 1.5 V, and V_{CCIO} = 1.5 V, 1.8 V, 2.5 V, and 3.3 V.

(8) This value is specified for normal device operation. The value may vary during power-up. This applies for all V_{CCIO} settings (3.3, 2.5, 1.8, and 1.5 V).

(9) Pin pull-up resistance values will lower if an external source drives the pin higher than V_{CCIO}.

(10) Drive strength is programmable according to the values shown in the *Stratix Architecture* chapter of the *Stratix Device Handbook, Volume 1*.

Table 4–9. Overshoot Input Voltage with Respect to Duty Cycle (Part 1 of 2)	
Vin (V)	**Maximum Duty Cycle (%)**
4.0	100
4.1	90
4.2	50

Table 4–36. Stratix Performance (Part 2 of 2) Notes (1), (2)

Applications		Resources Used			Performance				
		LEs	TriMatrix Memory Blocks	DSP Blocks	-5 Speed Grade	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Units
TriMatrix memory M-RAM block	True dual-port RAM 16K × 36 bit	0	1	0	269.83	237.69	206.82	175.74	MHz
	Single port RAM 32K × 18 bit	0	1	0	275.86	244.55	212.76	180.83	MHz
	Simple dual-port RAM 32K × 18 bit	0	1	0	275.86	244.55	212.76	180.83	MHz
	True dual-port RAM 32K × 18 bit	0	1	0	275.86	244.55	212.76	180.83	MHz
	Single port RAM 64K × 9 bit	0	1	0	287.85	253.29	220.36	187.26	MHz
	Simple dual-port RAM 64K × 9 bit	0	1	0	287.85	253.29	220.36	187.26	MHz
	True dual-port RAM 64K × 9 bit	0	1	0	287.85	253.29	220.36	187.26	MHz
DSP block	9 × 9-bit multiplier *(3)*	0	0	1	335.0	293.94	255.68	217.24	MHz
	18 × 18-bit multiplier *(4)*	0	0	1	278.78	237.41	206.52	175.50	MHz
	36 × 36-bit multiplier *(4)*	0	0	1	148.25	134.71	117.16	99.59	MHz
	36 × 36-bit multiplier *(5)*	0	0	1	278.78	237.41	206.52	175.5	MHz
	18-bit, 4-tap FIR filter	0	0	1	278.78	237.41	206.52	175.50	MHz
Larger Designs	8-bit, 16-tap parallel FIR filter	58	0	4	141.26	133.49	114.88	100.28	MHz
	8-bit, 1,024-point FFT function	870	5	1	261.09	235.51	205.21	175.22	MHz

Notes to Table 4–36:

(1) These design performance numbers were obtained using the Quartus II software.
(2) Numbers not listed will be included in a future version of the data sheet.
(3) This application uses registered inputs and outputs.
(4) This application uses registered multiplier input and output stages within the DSP block.
(5) This application uses registered multiplier input, pipeline, and output stages within the DSP block.

Table 4–39. DSP Block Internal Timing Microparameter Descriptions

Symbol	Parameter
t_{SU}	Input, pipeline, and output register setup time before clock
t_{H}	Input, pipeline, and output register hold time after clock
t_{CO}	Input, pipeline, and output register clock-to-output delay
$t_{INREG2PIPE9}$	Input Register to DSP Block pipeline register in 9 × 9-bit mode
$t_{INREG2PIPE18}$	Input Register to DSP Block pipeline register in 18 × 18-bit mode
$t_{PIPE2OUTREG2ADD}$	DSP Block Pipeline Register to output register delay in Two-Multipliers Adder mode
$t_{PIPE2OUTREG4ADD}$	DSP Block Pipeline Register to output register delay in Four-Multipliers Adder mode
t_{PD9}	Combinatorial input to output delay for 9 × 9
t_{PD18}	Combinatorial input to output delay for 18 × 18
t_{PD36}	Combinatorial input to output delay for 36 × 36
t_{CLR}	Minimum clear pulse width
t_{CLKHL}	Register minimum clock high or low time. This is a limit on the min time for the clock on the registers in these blocks. The actual performance is dependent upon the internal point-to-point delays in the blocks and may give slower performance as shown in Table 4–36 on page 4–20 and as reported by the timing analyzer in the Quartus II software.

Table 4–43. Routing Delay Internal Timing Microparameter Descriptions (Part 2 of 2)

Symbol	Parameter
t_{C4}	Delay for a C4 line with average loading; covers a distance of four LAB rows.
t_{C8}	Delay for a C8 line with average loading; covers a distance of eight LAB rows.
t_{C16}	Delay for a C16 line with average loading; covers a distance of 16 LAB rows.
t_{LOCAL}	Local interconnect delay, for connections within a LAB, and for the final routing hop of connections to LABs, DSP blocks, RAM blocks and I/Os.

Table 4–44. LE Internal Timing Microparameters

Parameter	-5		-6		-7		-8		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{SU}	10		10		11		13		ps
t_{H}	100		100		114		135		ps
t_{CO}		156		176		202		238	ps
t_{LUT}		366		459		527		621	ps
t_{CLR}	100		100		114		135		ps
t_{PRE}	100		100		114		135		ps
t_{CLKHL}	1000		1111		1190		1400		ps

Table 4–45. IOE Internal TSU Microparameter by Device Density (Part 1 of 2)

Device	Symbol	-5		-6		-7		-8		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
EP1S10	t_{SU_R}	76		80		80		80		ps
	t_{SU_C}	176		80		80		80		ps
EP1S20	t_{SU_R}	76		80		80		80		ps
	t_{SU_C}	76		80		80		80		ps
EP1S25	t_{SU_R}	276		280		280		280		ps
	t_{SU_C}	276		280		280		280		ps
EP1S30	t_{SU_R}	76		80		80		80		ps
	t_{SU_C}	176		180		180		180		ps

Table 4–45. IOE Internal TSU Microparameter by Device Density (Part 2 of 2)

Device	Symbol	-5		-6		-7		-8		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
EP1S40	t_{SU_R}	76		80		80		80		ps
	t_{SU_C}	376		380		380		380		ps
EP1S60	t_{SU_R}	276		280		280		280		ps
	t_{SU_C}	276		280		280		280		ps
EP1S80	t_{SU_R}	426		430		430		430		ps
	t_{SU_C}	76		80		80		80		ps

Table 4–46. IOE Internal Timing Microparameters

Symbol	-5		-6		-7		-8		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_H	68		71		82		96		ps
t_{CO_R}		171		179		206		242	ps
t_{CO_C}		171		179		206		242	ps
$t_{PIN2COMBOUT\_R}$		1,234		1,295		1,490		1,753	ps
$t_{PIN2COMBOUT\_C}$		1,087		1,141		1,312		1,544	ps
$t_{COMBIN2PIN\_R}$		3,894		4,089		4,089		4,089	ps
$t_{COMBIN2PIN\_C}$		4,299		4,494		4,494		4,494	ps
t_{CLR}	276		289		333		392		ps
t_{PRE}	260		273		313		369		ps
t_{CLKHL}	1,000		1,111		1,190		1,400		ps

Table 4–47. DSP Block Internal Timing Microparameters (Part 1 of 2)

Symbol	-5		-6		-7		-8		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{SU}	0		0		0		0		ps
t_H	67		75		86		101		ps
t_{CO}		142		158		181		214	ps
$t_{INREG2PIPE9}$		2,613		2,982		3,429		4,035	ps
$t_{INREG2PIPE18}$		3,390		3,993		4,591		5,402	ps

Table 4–81. EP1S40 External I/O Timing on Column Pins Using Global Clock Networks

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.126		2.268		2.558		2.930		ns
t_{INH}	0.000		0.000		0.000		0.000		ns
t_{OUTCO}	2.856	5.585	2.856	5.987	2.856	6.541	2.847	7.253	ns
t_{XZ}	2.796	5.459	2.796	5.855	2.796	6.417	2.787	7.138	ns
t_{ZX}	2.796	5.459	2.796	5.855	2.796	6.417	2.787	7.138	ns
$t_{INSUPLL}$	1.466		1.455		1.711		1.906		ns
t_{INHPLL}	0.000		0.000		0.000		0.000		ns
$t_{OUTCOPLL}$	1.092	2.345	1.092	2.510	1.092	2.455	1.089	2.473	ns
t_{XZPLL}	1.032	2.219	1.032	2.378	1.032	2.331	1.029	2.358	ns
t_{ZXPLL}	1.032	2.219	1.032	2.378	1.032	2.331	1.029	2.358	ns

Table 4–82. EP1S40 External I/O Timing on Row Pins Using Fast Regional Clock Networks

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.472		2.685		3.083		3.056		ns
t_{INH}	0.000		0.000		0.000		0.000		ns
t_{OUTCO}	2.631	5.258	2.631	5.625	2.631	6.105	2.745	7.324	ns
t_{XZ}	2.658	5.312	2.658	5.681	2.658	6.173	2.772	7.406	ns
t_{ZX}	2.658	5.312	2.658	5.681	2.658	6.173	2.772	7.406	ns

Tables 4–91 through 4–96 show the external timing parameters on column and row pins for EP1S80 devices.

Table 4–91. EP1S80 External I/O Timing on Column Pins Using Fast Regional Clock Networks Note (1)

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.328		2.528		2.900		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.422	4.830	2.422	5.169	2.422	5.633	NA	NA	ns
t_{XZ}	2.362	4.704	2.362	5.037	2.362	5.509	NA	NA	ns
t_{ZX}	2.362	4.704	2.362	5.037	2.362	5.509	NA	NA	ns

Table 4–92. EP1S80 External I/O Timing on Column Pins Using Regional Clock Networks Note (1)

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	1.760		1.912		2.194		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.761	5.398	2.761	5.785	2.761	6.339	NA	NA	ns
t_{XZ}	2.701	5.272	2.701	5.653	2.701	6.215	NA	NA	ns
t_{ZX}	2.701	5.272	2.701	5.653	2.701	6.215	NA	NA	ns
$t_{INSUPLL}$	0.462		0.606		0.785		NA		ns
t_{INHPLL}	0.000		0.000		0.000		NA		ns
$t_{OUTCOPLL}$	1.661	2.849	1.661	2.859	1.661	2.881	NA	NA	ns
t_{XZPLL}	1.601	2.723	1.601	2.727	1.601	2.757	NA	NA	ns
t_{ZXPLL}	1.601	2.723	1.601	2.727	1.601	2.757	NA	NA	ns

Table 4–104. Stratix I/O Standard Row Pin Input Delay Adders

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
LVCMOS		0		0		0		0	ps
3.3-V LVTTL		0		0		0		0	ps
2.5-V LVTTL		21		22		25		29	ps
1.8-V LVTTL		181		190		218		257	ps
1.5-V LVTTL		300		315		362		426	ps
GTL+		–152		–160		–184		–216	ps
CTT		–168		–177		–203		–239	ps
SSTL-3 Class I		–193		–203		–234		–275	ps
SSTL-3 Class II		–193		–203		–234		–275	ps
SSTL-2 Class I		–262		–276		–317		–373	ps
SSTL-2 Class II		–262		–276		–317		–373	ps
SSTL-18 Class I		–105		–111		–127		–150	ps
SSTL-18 Class II		0		0		0		0	ps
1.5-V HSTL Class I		–151		–159		–183		–215	ps
1.8-V HSTL Class I		–126		–133		–153		–179	ps
LVDS		–149		–157		–180		–212	ps
LVPECL		–149		–157		–180		–212	ps
3.3-V PCML		–65		–69		–79		–93	ps
HyperTransport		77		–81		–93		–110	ps

Table 4–117. Stratix Maximum Input Clock Rate for CLK[7..4] & CLK[15..12] Pins in Wire-Bond Packages (Part 2 of 2)

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
GTL+	250	200	200	MHz
SSTL-3 Class I	300	250	250	MHz
SSTL-3 Class II	300	250	250	MHz
SSTL-2 Class I	300	250	250	MHz
SSTL-2 Class II	300	250	250	MHz
SSTL-18 Class I	300	250	250	MHz
SSTL-18 Class II	300	250	250	MHz
1.5-V HSTL Class I	300	180	180	MHz
1.5-V HSTL Class II	300	180	180	MHz
1.8-V HSTL Class I	300	180	180	MHz
1.8-V HSTL Class II	300	180	180	MHz
3.3-V PCI	422	390	390	MHz
3.3-V PCI-X 1.0	422	390	390	MHz
Compact PCI	422	390	390	MHz
AGP 1×	422	390	390	MHz
AGP 2×	422	390	390	MHz
CTT	250	180	180	MHz
Differential 1.5-V HSTL C1	300	180	180	MHz
LVPECL *(1)*	422	400	400	MHz
PCML *(1)*	215	200	200	MHz
LVDS *(1)*	422	400	400	MHz
HyperTransport technology *(1)*	422	400	400	MHz

Table 4–118. Stratix Maximum Input Clock Rate for CLK[0, 2, 9, 11] Pins & FPLL[10..7]CLK Pins in Wire-Bond Packages (Part 1 of 2)

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	422	390	390	MHz
2.5 V	422	390	390	MHz
1.8 V	422	390	390	MHz
1.5 V	422	390	390	MHz

Table 4–123. Stratix Maximum Output Clock Rate (Using I/O Pins) for PLL[1, 2, 3, 4] Pins in Wire-Bond Packages (Part 2 of 2)				
I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVDS *(2)*	400	311	311	MHz
HyperTransport technology *(2)*	420	400	400	MHz

Notes to Tables 4–120 through 4–123:

(1) Differential SSTL-2 outputs are only available on column clock pins.
(2) These parameters are only available on row I/O pins.
(3) SSTL-2 in maximum drive strength condition. See Table 4–101 on page 4–62 for more information on exact loading conditions for each I/O standard.
(4) SSTL-2 in minimum drive strength with ≤10pF output load condition.
(5) SSTL-2 in minimum drive strength with > 10pF output load condition.
(6) Differential SSTL-2 outputs are only supported on column clock pins.