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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	2566
Number of Logic Elements/Cells	25660
Total RAM Bits	1944576
Number of I/O	473
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	672-BBGA
Supplier Device Package	672-BGA (35x35)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1s25b672c6



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Section I. Stratix Device Family Data Sheet

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Table 1–1. Stratix Device Features — EP1S10, EP1S20, EP1S25, EP1S30

Feature	EP1S10	EP1S20	EP1S25	EP1S30
LEs	10,570	18,460	25,660	32,470
M512 RAM blocks (32×18 bits)	94	194	224	295
M4K RAM blocks (128×36 bits)	60	82	138	171
M-RAM blocks ($4K \times 144$ bits)	1	2	2	4
Total RAM bits	920,448	1,669,248	1,944,576	3,317,184
DSP blocks	6	10	10	12
Embedded multipliers (1)	48	80	80	96
PLLs	6	6	6	10
Maximum user I/O pins	426	586	706	726

Table 1–2. Stratix Device Features — EP1S40, EP1S60, EP1S80

Feature	EP1S40	EP1S60	EP1S80
LEs	41,250	57,120	79,040
M512 RAM blocks (32×18 bits)	384	574	767
M4K RAM blocks (128×36 bits)	183	292	364
M-RAM blocks ($4K \times 144$ bits)	4	6	9
Total RAM bits	3,423,744	5,215,104	7,427,520
DSP blocks	14	18	22
Embedded multipliers (1)	112	144	176
PLLs	12	12	12
Maximum user I/O pins	822	1,022	1,238

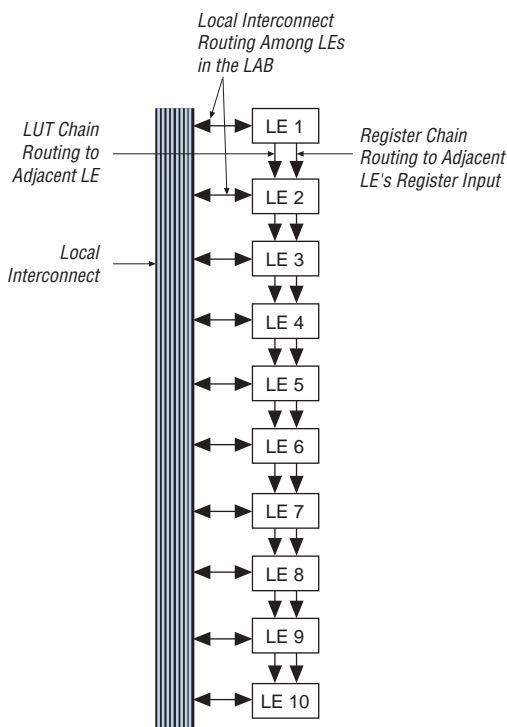
Note to Tables 1–1 and 1–2:

- (1) This parameter lists the total number of 9×9 -bit multipliers for each device. For the total number of 18×18 -bit multipliers per device, divide the total number of 9×9 -bit multipliers by 2. For the total number of 36×36 -bit multipliers per device, divide the total number of 9×9 -bit multipliers by 8.

Dynamic Arithmetic Mode

The dynamic arithmetic mode is ideal for implementing adders, counters, accumulators, wide parity functions, and comparators. An LE in dynamic arithmetic mode uses four 2-input LUTs configurable as a dynamic adder/subtractor. The first two 2-input LUTs compute two summations based on a possible carry-in of 1 or 0; the other two LUTs generate carry outputs for the two chains of the carry select circuitry. As shown in [Figure 2-7](#), the LAB carry-in signal selects either the carry-in0 or carry-in1 chain. The selected chain's logic level in turn determines which parallel sum is generated as a combinatorial or registered output. For example, when implementing an adder, the sum output is the selection of two possible calculated sums: $\text{data1} + \text{data2} + \text{carry-in0}$ or $\text{data1} + \text{data2} + \text{carry-in1}$. The other two LUTs use the data1 and data2 signals to generate two possible carry-out signals—one for a carry of 1 and the other for a carry of 0. The carry-in0 signal acts as the carry select for the carry-out0 output and carry-in1 acts as the carry select for the carry-out1 output. LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output.

The dynamic arithmetic mode also offers clock enable, counter enable, synchronous up/down control, synchronous clear, synchronous load, and dynamic adder/subtractor options. The LAB local interconnect data inputs generate the counter enable and synchronous up/down control signals. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. The Quartus II software automatically places any registers that are not used by the counter into other LABs. The addnsub LAB-wide signal controls whether the LE acts as an adder or subtractor.

Figure 2–10. LUT Chain & Register Chain Interconnects

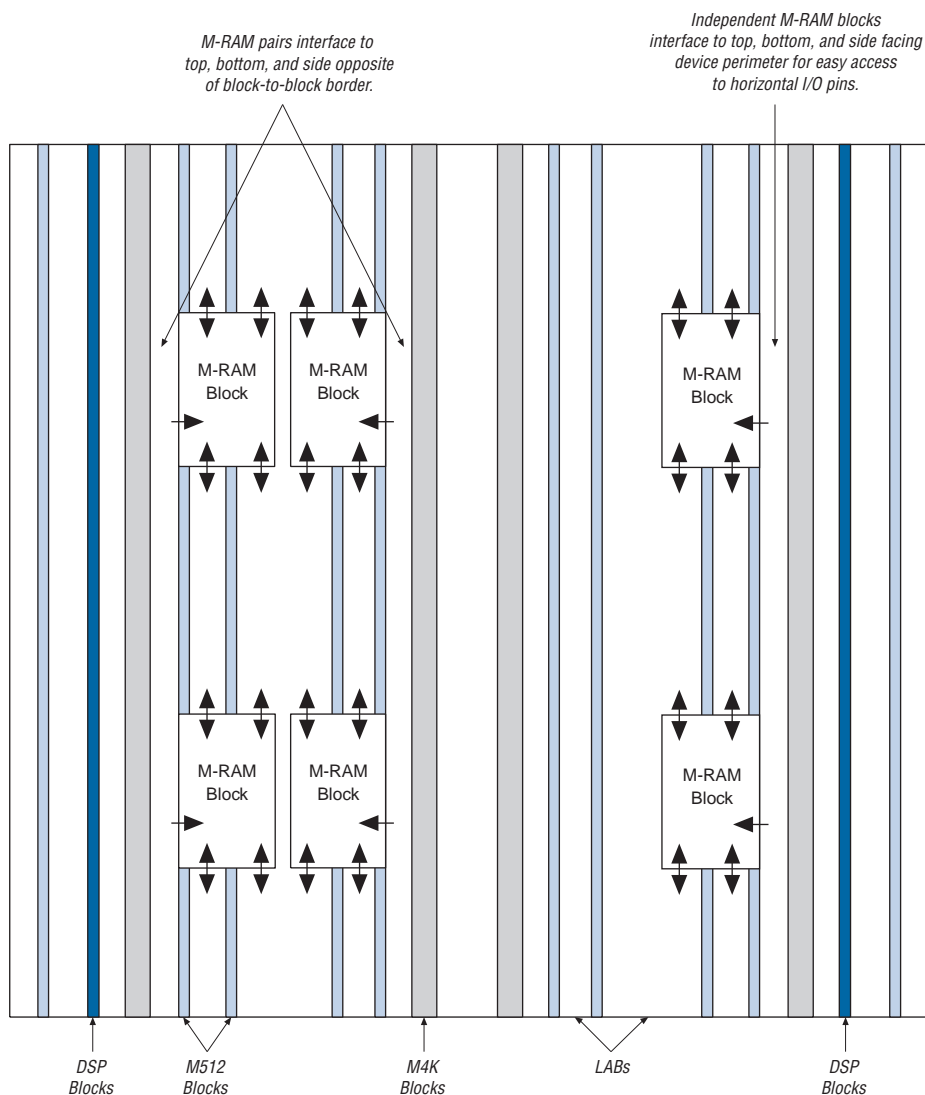
The C4 interconnects span four LABs, M512, or M4K blocks up or down from a source LAB. Every LAB has its own set of C4 interconnects to drive either up or down. [Figure 2–11](#) shows the C4 interconnect connections from an LAB in a column. The C4 interconnects can drive and be driven by all types of architecture blocks, including DSP blocks, TriMatrix memory blocks, and vertical IOEs. For LAB interconnection, a primary LAB or its LAB neighbor can drive a given C4 interconnect. C4 interconnects can drive each other to extend their range as well as drive row interconnects for column-to-column connections.

Shift Register Support

You can configure embedded memory blocks to implement shift registers for DSP applications such as pseudo-random number generators, multi-channel filtering, auto-correlation, and cross-correlation functions. These and other DSP applications require local data storage, traditionally implemented with standard flip-flops, which can quickly consume many logic cells and routing resources for large shift registers. A more efficient alternative is to use embedded memory as a shift register block, which saves logic cell and routing resources and provides a more efficient implementation with the dedicated circuitry.

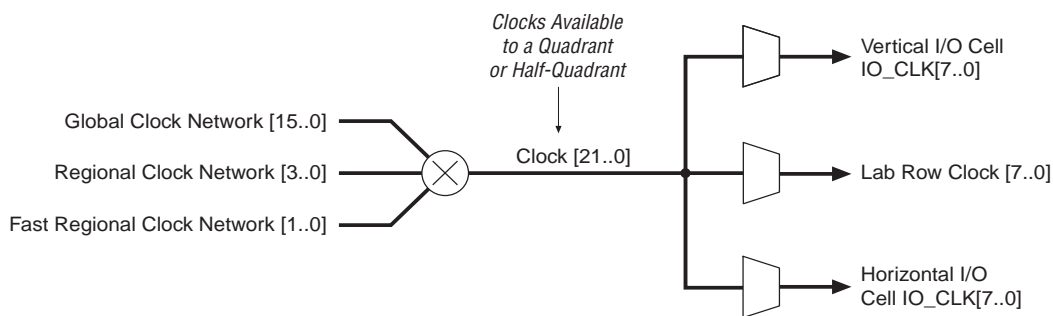
The size of a $w \times m \times n$ shift register is determined by the input data width (w), the length of the taps (m), and the number of taps (n). The size of a $w \times m \times n$ shift register must be less than or equal to the maximum number of memory bits in the respective block: 576 bits for the M512 RAM block and 4,608 bits for the M4K RAM block. The total number of shift register outputs (number of taps $n \times$ width w) must be less than the maximum data width of the RAM block (18 for M512 blocks, 36 for M4K blocks). To create larger shift registers, the memory blocks are cascaded together.

Data is written into each address location at the falling edge of the clock and read from the address at the rising edge of the clock. The shift register mode logic automatically controls the positive and negative edge clocking to shift the data in one clock cycle. [Figure 2-14](#) shows the TriMatrix memory block in the shift register mode.

Figure 2–20. EP1S60 Device with M-RAM Interface Locations *Note (1)***Note to Figure 2–20:**

(1) Device shown is an EP1S60 device. The number and position of M-RAM blocks varies in other devices.

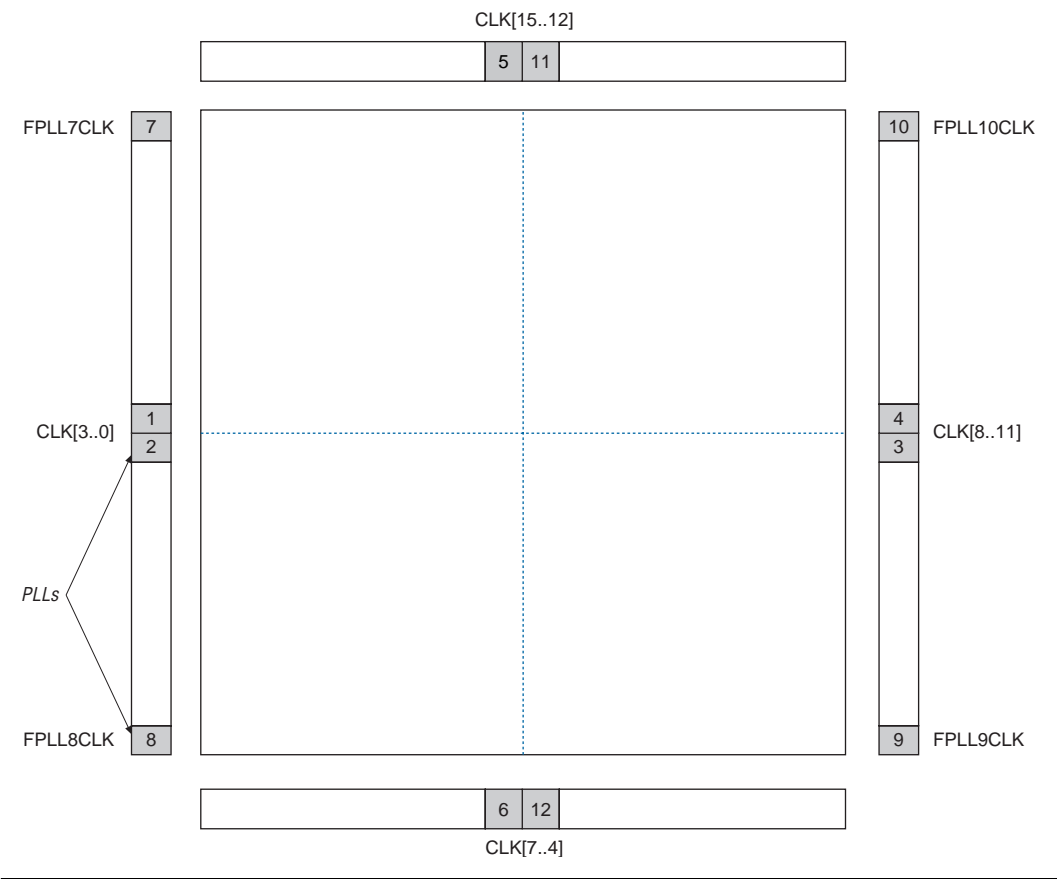
The M-RAM block local interconnect is driven by the R4, R8, C4, C8, and direct link interconnects from adjacent LABs. For independent M-RAM blocks, up to 10 direct link address and control signal input connections to the M-RAM block are possible from the left adjacent LABs for M-RAM

Figure 2–46. Regional Clock Bus

IOE clocks have horizontal and vertical block regions that are clocked by eight I/O clock signals chosen from the 22 quadrant or half-quadrant clock resources. [Figures 2–47](#) and [2–48](#) show the quadrant and half-quadrant relationship to the I/O clock regions, respectively. The vertical regions (column pins) have less clock delay than the horizontal regions (row pins).

Figure 2–49 shows a top-level diagram of the Stratix device and PLL floorplan.

Figure 2–49. PLL Locations



VCO period from up to eight taps for individual fine step selection. Also, each clock output counter can use a unique initial count setting to achieve individual coarse shift selection in steps of one VCO period. The combination of coarse and fine shifts allows phase shifting for the entire input clock period.

The equation to determine the precision of the phase shifting in degrees is: $45^\circ \div \text{post-scale counter value}$. Therefore, the maximum step size is 45° , and smaller steps are possible depending on the multiplication and division ratio necessary on the output counter port.

This type of phase shift provides the highest precision since it is the least sensitive to process, supply, and temperature variation.

Clock Delay

In addition to the phase shift feature, the ability to fine tune the Δt clock delay provides advanced time delay shift control on each of the four PLL outputs. There are time delays for each post-scale counter (e , g , or l) from the PLL, the n counter, and m counter. Each of these can shift in 250-ps increments for a range of 3.0 ns. The m delay shifts all outputs earlier in time, while n delay shifts all outputs later in time. Individual delays on post-scale counters (e , g , and l) provide positive delay for each output. Table 2-21 shows the combined delay for each output for normal or zero delay buffer mode where Δt_e , Δt_g , or Δt_l is unique for each PLL output.

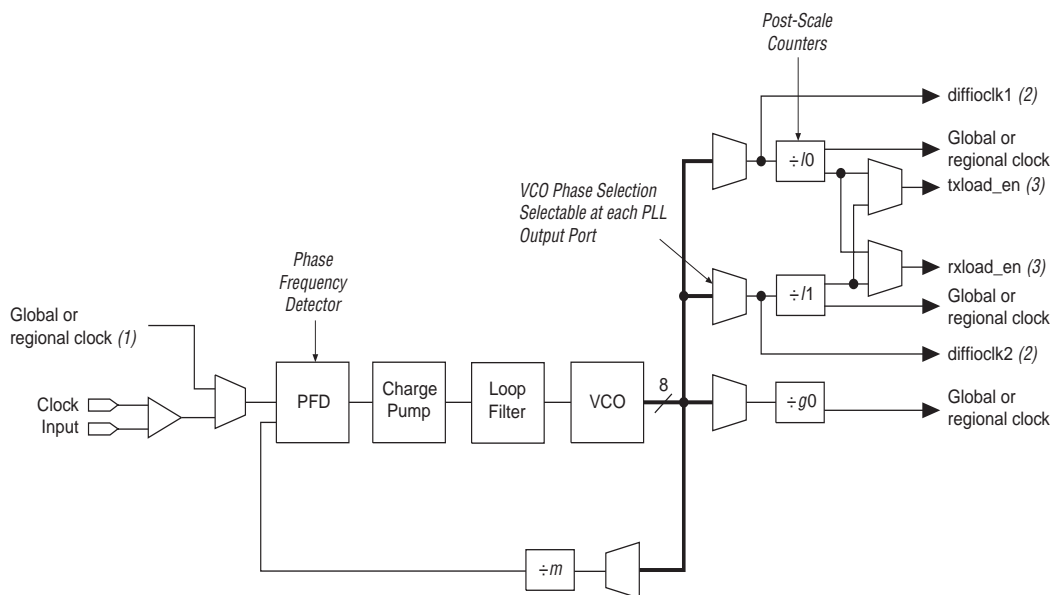
The t_{OUTPUT} for a single output can range from -3 ns to $+6$ ns. The total delay shift difference between any two PLL outputs, however, must be less than ± 3 ns. For example, shifts on two outputs of -1 and $+2$ ns is allowed, but not -1 and $+2.5$ ns because these shifts would result in a difference of 3.5 ns. If the design uses external feedback, the Δt_e delay will remove delay from outputs, represented by a negative sign (see Table 2-21). This effect occurs because the Δt_e delay is then part of the feedback loop.

Table 2-21. Output Clock Delay for Enhanced PLLs

Normal or Zero Delay Buffer Mode	External Feedback Mode
$\Delta t_{e\text{OUTPUT}} = \Delta t_n - \Delta t_m + \Delta t_e$	$\Delta t_{e\text{OUTPUT}} = \Delta t_n - \Delta t_m - \Delta t_e$ (1)
$\Delta t_{g\text{OUTPUT}} = \Delta t_n - \Delta t_m + \Delta t_g$	$\Delta t_{g\text{OUTPUT}} = \Delta t_n - \Delta t_m + \Delta t_g$
$\Delta t_{l\text{OUTPUT}} = \Delta t_n - \Delta t_m + \Delta t_l$	$\Delta t_{l\text{OUTPUT}} = \Delta t_n - \Delta t_m + \Delta t_l$

Note to Table 2-21:

(1) Δt_e removes delay from outputs in external feedback mode.

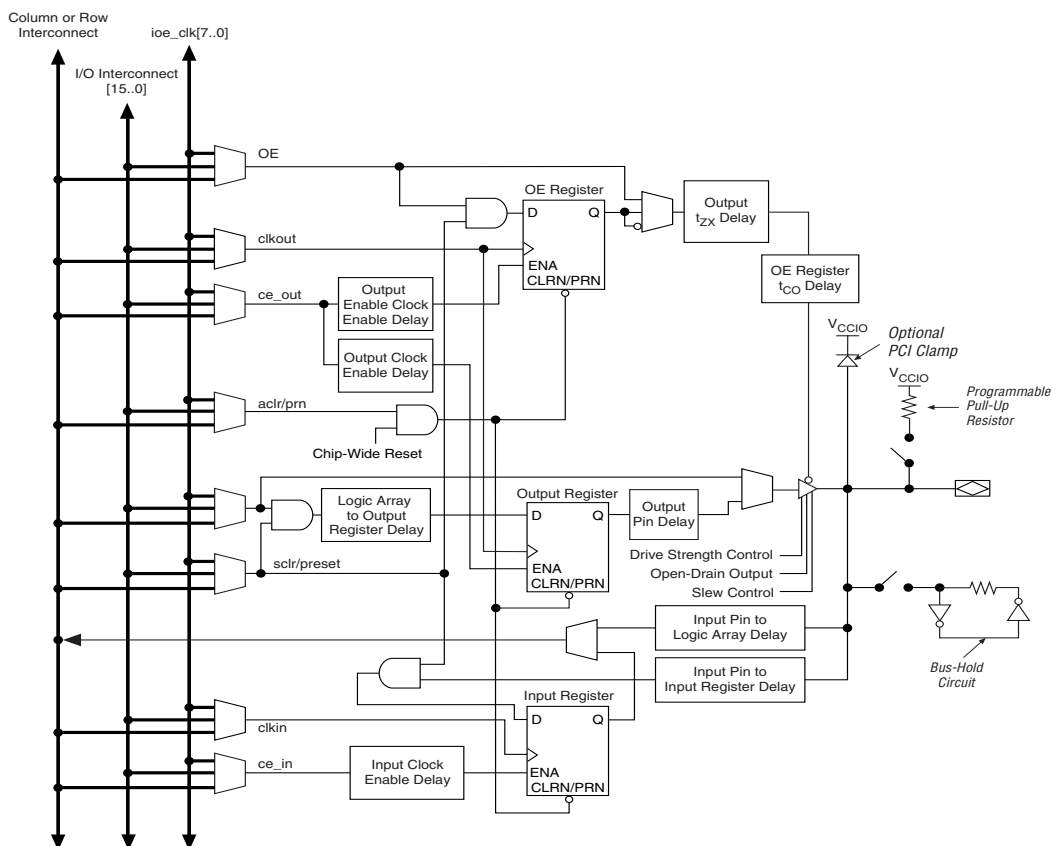
Figure 2–58. Stratix Device Fast PLL**Notes to Figure 2–58:**

- (1) The global or regional clock input can be driven by an output from another PLL or any dedicated CLK or FCLK pin. It cannot be driven by internally-generated global signals.
- (2) In high-speed differential I/O support mode, this high-speed PLL clock feeds the SERDES. Stratix devices only support one rate of data transfer per fast PLL in high-speed differential I/O support mode.
- (3) This signal is a high-speed differential I/O support SERDES control signal.

Clock Multiplication & Division

Stratix device fast PLLs provide clock synthesis for PLL output ports using m /(post scaler) scaling factors. The input clock is multiplied by the m feedback factor. Each output port has a unique post scale counter to divide down the high-frequency VCO. There is one multiply divider, m , per fast PLL with a range of 1 to 32. There are two post scale L dividers for regional and/or LVDS interface clocks, and $g0$ counter for global clock output port; all range from 1 to 32.

In the case of a high-speed differential interface, set the output counter to 1 to allow the high-speed VCO frequency to drive the SERDES. When used for clocking the SERDES, the m counter can range from 1 to 30. The VCO frequency is equal to $f_{IN} \times m$, where VCO frequency must be between 300 and 1000 MHz.

Figure 2–64. Stratix IOE in Bidirectional I/O Configuration Note (1)**Note to Figure 2–64:**

(1) All input signals to the IOE can be inverted at the IOE.

The Stratix device IOE includes programmable delays that can be activated to ensure zero hold times, input IOE register-to-logic array register transfers, or logic array-to-output IOE register transfers.

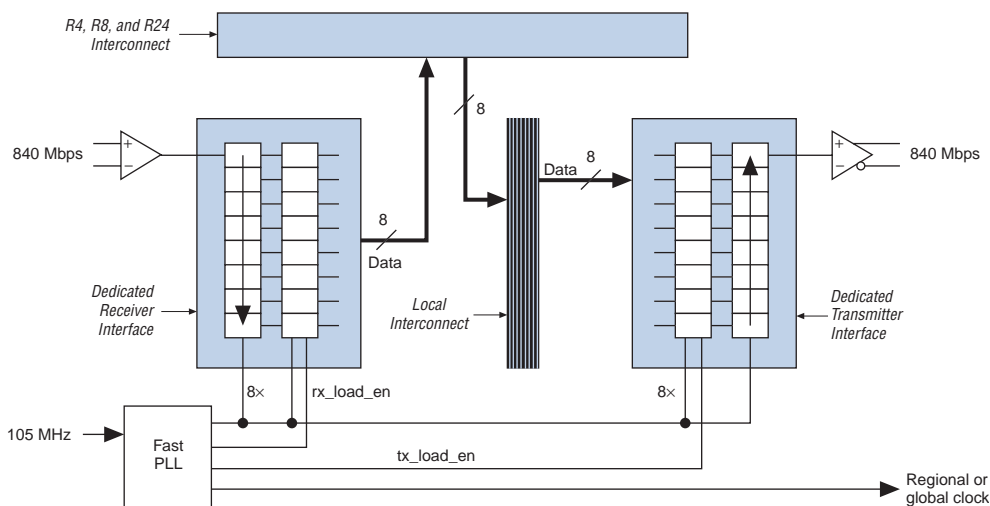
A path in which a pin directly drives a register may require the delay to ensure zero hold time, whereas a path in which a pin drives a register through combinatorial logic may not require the delay. Programmable delays exist for decreasing input-pin-to-logic-array and IOE input register delays. The Quartus II Compiler can program these delays to automatically minimize setup time while providing a zero hold time. Programmable delays can increase the register-to-pin delays for output

- RapidIO
- HyperTransport

Dedicated Circuitry

Stratix devices support source-synchronous interfacing with LVDS, LVPECL, 3.3-V PCML, or HyperTransport signaling at up to 840 Mbps. Stratix devices can transmit or receive serial channels along with a low-speed or high-speed clock. The receiving device PLL multiplies the clock by a integer factor W ($W = 1$ through 32). For example, a HyperTransport application where the data rate is 800 Mbps and the clock rate is 400 MHz would require that W be set to 2. The SERDES factor J determines the parallel data width to deserialize from receivers or to serialize for transmitters. The SERDES factor J can be set to 4, 7, 8, or 10 and does not have to equal the PLL clock-multiplication W value. For a J factor of 1, the Stratix device bypasses the SERDES block. For a J factor of 2, the Stratix device bypasses the SERDES block, and the DDR input and output registers are used in the IOE. See [Figure 2-73](#).

Figure 2-73. High-Speed Differential I/O Receiver / Transmitter Interface Example



An external pin or global or regional clock can drive the fast PLLs, which can output up to three clocks: two multiplied high-speed differential I/O clocks to drive the SERDES block and/or external pin, and a low-speed clock to drive the logic array.

Operating Conditions

Stratix® devices are offered in both commercial and industrial grades. Industrial devices are offered in -6 and -7 speed grades and commercial devices are offered in -5 (fastest), -6, -7, and -8 speed grades. This section specifies the operation conditions for operating junction temperature, V_{CCINT} and V_{CCIO} voltage levels, and input voltage requirements. The voltage specifications in this section are specified at the pins of the device (and not the power supply). If the device operates outside these ranges, then all DC and AC specifications are not guaranteed. Furthermore, the reliability of the device may be affected. The timing parameters in this chapter apply to both commercial and industrial temperature ranges unless otherwise stated.

Tables 4–1 through 4–8 provide information on absolute maximum ratings.

Table 4–1. Stratix Device Absolute Maximum Ratings Notes (1), (2)

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCINT}	Supply voltage	With respect to ground	–0.5	2.4	V
V_{CCIO}			–0.5	4.6	V
V_I	DC input voltage (3)		–0.5	4.6	V
I_{OUT}	DC output current, per pin		–25	40	mA
T_{STG}	Storage temperature	No bias	–65	150	°C
T_J	Junction temperature	BGA packages under bias		135	°C

Table 4–2. Stratix Device Recommended Operating Conditions (Part 1 of 2)

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCINT}	Supply voltage for internal logic and input buffers	(4)	1.425	1.575	V

Table 4–2. Stratix Device Recommended Operating Conditions (Part 2 of 2)

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCIO}	Supply voltage for output buffers, 3.3-V operation	(4), (5)	3.00 (3.135)	3.60 (3.465)	V
	Supply voltage for output buffers, 2.5-V operation	(4)	2.375	2.625	V
	Supply voltage for output buffers, 1.8-V operation	(4)	1.71	1.89	V
	Supply voltage for output buffers, 1.5-V operation	(4)	1.4	1.6	V
V_I	Input voltage	(3), (6)	–0.5	4.0	V
V_O	Output voltage		0	V_{CCIO}	V
T_J	Operating junction temperature	For commercial use	0	85	°C
		For industrial use	–40	100	°C

Table 4–3. Stratix Device DC Operating Conditions Note (7) (Part 1 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
I_I	Input pin leakage current	$V_I = V_{CCIOmax}$ to 0 V (8)	–10		10	μA
I_{OZ}	Tri-stated I/O pin leakage current	$V_O = V_{CCIOmax}$ to 0 V (8)	–10		10	μA
I_{CC0}	V_{CC} supply current (standby) (All memory blocks in power-down mode)	V_I = ground, no load, no toggling inputs				mA
		EP1S10. V_I = ground, no load, no toggling inputs		37		mA
		EP1S20. V_I = ground, no load, no toggling inputs		65		mA
		EP1S25. V_I = ground, no load, no toggling inputs		90		mA
		EP1S30. V_I = ground, no load, no toggling inputs		114		mA
		EP1S40. V_I = ground, no load, no toggling inputs		145		mA
		EP1S60. V_I = ground, no load, no toggling inputs		200		mA
		EP1S80. V_I = ground, no load, no toggling inputs		277		mA

Table 4–36. Stratix Performance (Part 2 of 2) Notes (1), (2)

Applications		Resources Used			Performance				
		LEs	TriMatrix Memory Blocks	DSP Blocks	-5 Speed Grade	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Units
TriMatrix memory M-RAM block	True dual-port RAM 16K × 36 bit	0	1	0	269.83	237.69	206.82	175.74	MHz
	Single port RAM 32K × 18 bit	0	1	0	275.86	244.55	212.76	180.83	MHz
	Simple dual-port RAM 32K × 18 bit	0	1	0	275.86	244.55	212.76	180.83	MHz
	True dual-port RAM 32K × 18 bit	0	1	0	275.86	244.55	212.76	180.83	MHz
	Single port RAM 64K × 9 bit	0	1	0	287.85	253.29	220.36	187.26	MHz
	Simple dual-port RAM 64K × 9 bit	0	1	0	287.85	253.29	220.36	187.26	MHz
	True dual-port RAM 64K × 9 bit	0	1	0	287.85	253.29	220.36	187.26	MHz
DSP block	9 × 9-bit multiplier (3)	0	0	1	335.0	293.94	255.68	217.24	MHz
	18 × 18-bit multiplier (4)	0	0	1	278.78	237.41	206.52	175.50	MHz
	36 × 36-bit multiplier (4)	0	0	1	148.25	134.71	117.16	99.59	MHz
	36 × 36-bit multiplier (5)	0	0	1	278.78	237.41	206.52	175.5	MHz
	18-bit, 4-tap FIR filter	0	0	1	278.78	237.41	206.52	175.50	MHz
Larger Designs	8-bit, 16-tap parallel FIR filter	58	0	4	141.26	133.49	114.88	100.28	MHz
	8-bit, 1,024-point FFT function	870	5	1	261.09	235.51	205.21	175.22	MHz

Notes to Table 4–36:

- (1) These design performance numbers were obtained using the Quartus II software.
- (2) Numbers not listed will be included in a future version of the data sheet.
- (3) This application uses registered inputs and outputs.
- (4) This application uses registered multiplier input and output stages within the DSP block.
- (5) This application uses registered multiplier input, pipeline, and output stages within the DSP block.

Stratix External I/O Timing

These timing parameters are for both column IOE and row IOE pins. In EP1S30 devices and above, you can decrease the t_{SU} time by using the FPLLCLK, but may get positive hold time in EP1S60 and EP1S80 devices. You should use the Quartus II software to verify the external devices for any pin.

Tables 4–55 through 4–60 show the external timing parameters on column and row pins for EP1S10 devices.

Table 4–55. EP1S10 External I/O Timing on Column Pins Using Fast Regional Clock Networks *Note (1)*

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.238		2.325		2.668		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.240	4.549	2.240	4.836	2.240	5.218	NA	NA	ns
t_{XZ}	2.180	4.423	2.180	4.704	2.180	5.094	NA	NA	ns
t_{ZX}	2.180	4.423	2.180	4.704	2.180	5.094	NA	NA	ns

Table 4–56. EP1S10 External I/O Timing on Column Pins Using Regional Clock Networks *Note (1)*

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max			
t_{INSU}	1.992		2.054		2.359		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.395	4.795	2.395	5.107	2.395	5.527	NA	NA	ns
t_{XZ}	2.335	4.669	2.335	4.975	2.335	5.403	NA	NA	ns
t_{ZX}	2.335	4.669	2.335	4.975	2.335	5.403	NA	NA	ns
$t_{INSUPLL}$	0.975		0.985		1.097		NA		ns
t_{INHPLL}	0.000		0.000		0.000		NA	NA	ns
$t_{OUTCOPLL}$	1.262	2.636	1.262	2.680	1.262	2.769	NA	NA	ns
t_{XZPLL}	1.202	2.510	1.202	2.548	1.202	2.645	NA	NA	ns
t_{ZXPLL}	1.202	2.510	1.202	2.548	1.202	2.645	NA	NA	ns

Table 4–69. EP1S25 External I/O Timing on Column Pins Using Global Clock Networks

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	1.371		1.471		1.657		1.916		ns
t_{INH}	0.000		0.000		0.000		0.000		ns
t_{OUTCO}	2.809	5.516	2.809	5.890	2.809	6.429	2.809	7.155	ns
t_{xZ}	2.749	5.390	2.749	5.758	2.749	6.305	2.749	7.040	ns
t_{ZX}	2.749	5.390	2.749	5.758	2.749	6.305	2.749	7.040	ns
t_{INSUPLL}	1.271		1.327		1.491		1.677		ns
t_{INHPLL}	0.000		0.000		0.000		0.000		ns
t_{OUTCOPLL}	1.124	2.396	1.124	2.492	1.124	2.522	1.124	2.602	ns
t_{xZPLL}	1.064	2.270	1.064	2.360	1.064	2.398	1.064	2.487	ns
t_{ZXPLL}	1.064	2.270	1.064	2.360	1.064	2.398	1.064	2.487	ns

Table 4–70. EP1S25 External I/O Timing on Row Pins Using Fast Regional Clock Networks

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.429		2.631		2.990		3.503		ns
t_{INH}	0.000		0.000		0.000		0.000		ns
t_{OUTCO}	2.376	4.821	2.376	5.131	2.376	5.538	2.376	6.063	ns
t_{xZ}	2.403	4.875	2.403	5.187	2.403	5.606	2.403	6.145	ns
t_{ZX}	2.403	4.875	2.403	5.187	2.403	5.606	2.403	6.145	ns

External I/O Delay Parameters

External I/O delay timing parameters for I/O standard input and output adders and programmable input and output delays are specified by speed grade independent of device density. All of the timing parameters in this section apply to both flip-chip and wire-bond packages.

Tables 4–103 and 4–104 show the input adder delays associated with column and row I/O pins. If an I/O standard is selected other than 3.3-V LVTTTL or LVCMOS, add the selected delay to the external t_{INSU} and t_{INSUPLL} I/O parameters shown in Tables 4–54 through 4–96.

Table 4–103. Stratix I/O Standard Column Pin Input Delay Adders

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
LVCMOS		0		0		0		0	ps
3.3-V LVTTTL		0		0		0		0	ps
2.5-V LVTTTL		19		19		22		26	ps
1.8-V LVTTTL		221		232		266		313	ps
1.5-V LVTTTL		352		369		425		500	ps
GTL		–45		–48		–55		–64	ps
GTL+		–75		–79		–91		–107	ps
3.3-V PCI		0		0		0		0	ps
3.3-V PCI-X 1.0		0		0		0		0	ps
Compact PCI		0		0		0		0	ps
AGP 1×		0		0		0		0	ps
AGP 2×		0		0		0		0	ps
CTT		120		126		144		170	ps
SSTL-3 Class I		–162		–171		–196		–231	ps
SSTL-3 Class II		–162		–171		–196		–231	ps
SSTL-2 Class I		–202		–213		–244		–287	ps
SSTL-2 Class II		–202		–213		–244		–287	ps
SSTL-18 Class I		78		81		94		110	ps
SSTL-18 Class II		78		81		94		110	ps
1.5-V HSTL Class I		–76		–80		–92		–108	ps
1.5-V HSTL Class II		–76		–80		–92		–108	ps
1.8-V HSTL Class I		–52		–55		–63		–74	ps
1.8-V HSTL Class II		–52		–55		–63		–74	ps

PLL Specifications

Tables 4–127 through 4–129 describe the Stratix device enhanced PLL specifications.

Table 4–127. Enhanced PLL Specifications for -5 Speed Grades (Part 1 of 2)

Symbol	Parameter	Min	Typ	Max	Unit
f_{IN}	Input clock frequency	3 (1), (2)		684	MHz
f_{INPFD}	Input frequency to PFD	3		420	MHz
f_{INDUTY}	Input clock duty cycle	40		60	%
$f_{EINDUTY}$	External feedback clock input duty cycle	40		60	%
$t_{INJITTER}$	Input clock period jitter			±200 (3)	ps
$t_{EINJITTER}$	External feedback clock period jitter			±200 (3)	ps
t_{FCOMP}	External feedback clock compensation time (4)			6	ns
f_{OUT}	Output frequency for internal global or regional clock	0.3		500	MHz
f_{OUT_EXT}	Output frequency for external clock (3)	0.3		526	MHz
$t_{OUTDUTY}$	Duty cycle for external clock output (when set to 50%)	45		55	%
t_{JITTER}	Period jitter for external clock output (6)			±100 ps for >200-MHz outclk ±20 mUI for <200-MHz outclk	ps or mUI
$t_{CONFIG5,6}$	Time required to reconfigure the scan chains for PLLs 5 and 6			$289/f_{SCANCLK}$	
$t_{CONFIG11,12}$	Time required to reconfigure the scan chains for PLLs 11 and 12			$193/f_{SCANCLK}$	
$t_{SCANCLK}$	scanclk frequency (5)			22	MHz
t_{DLOCK}	Time required to lock dynamically (after switchover or reconfiguring any non-post-scale counters/delays) (7)			100	μs
t_{LOCK}	Time required to lock from end of device configuration	10		400	μs
f_{VCO}	PLL internal VCO operating range	300		800 (8)	MHz
t_{LSKEW}	Clock skew between two external clock outputs driven by the same counter		±50		ps

Table 4–128. Enhanced PLL Specifications for -6 Speed Grades (Part 2 of 2)

Symbol	Parameter	Min	Typ	Max	Unit
t_{SCANCLK}	scanclk frequency (5)			22	MHz
t_{DLOCK}	Time required to lock dynamically (after switchover or reconfiguring any non-post-scale counters/delays) (7) (11)	(9)		100	μs
t_{LOCK}	Time required to lock from end of device configuration (11)	10		400	μs
f_{VCO}	PLL internal VCO operating range	300		800 (8)	MHz
t_{LSKEW}	Clock skew between two external clock outputs driven by the same counter		± 50		ps
t_{SKEW}	Clock skew between two external clock outputs driven by the different counters with the same settings		± 75		ps
f_{SS}	Spread spectrum modulation frequency	30		150	kHz
% spread	Percentage spread for spread spectrum frequency (10)	0.4	0.5	0.6	%
t_{ARESET}	Minimum pulse width on areset signal	10			ns

Table 4–129. Enhanced PLL Specifications for -7 Speed Grade (Part 1 of 2)

Symbol	Parameter	Min	Typ	Max	Unit
f_{IN}	Input clock frequency	3 (1), (2)		565	MHz
f_{INPFD}	Input frequency to PFD	3		420	MHz
f_{INDUTY}	Input clock duty cycle	40		60	%
f_{EINDUTY}	External feedback clock input duty cycle	40		60	%
t_{INJITTER}	Input clock period jitter			± 200 (3)	ps
$t_{\text{EINJITTER}}$	External feedback clock period jitter			± 200 (3)	ps
t_{FCOMP}	External feedback clock compensation time (4)			6	ns
f_{OUT}	Output frequency for internal global or regional clock	0.3		420	MHz
$f_{\text{OUT_EXT}}$	Output frequency for external clock (3)	0.3		434	MHz