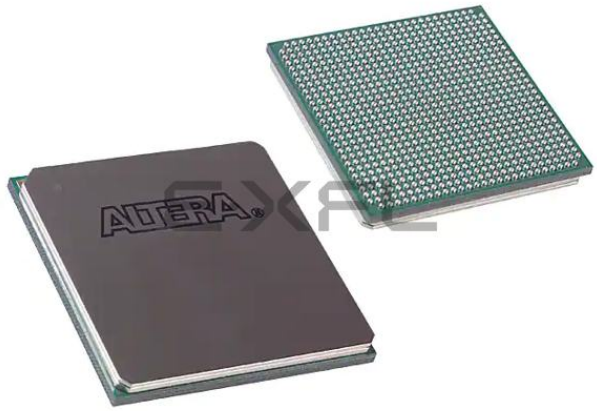

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

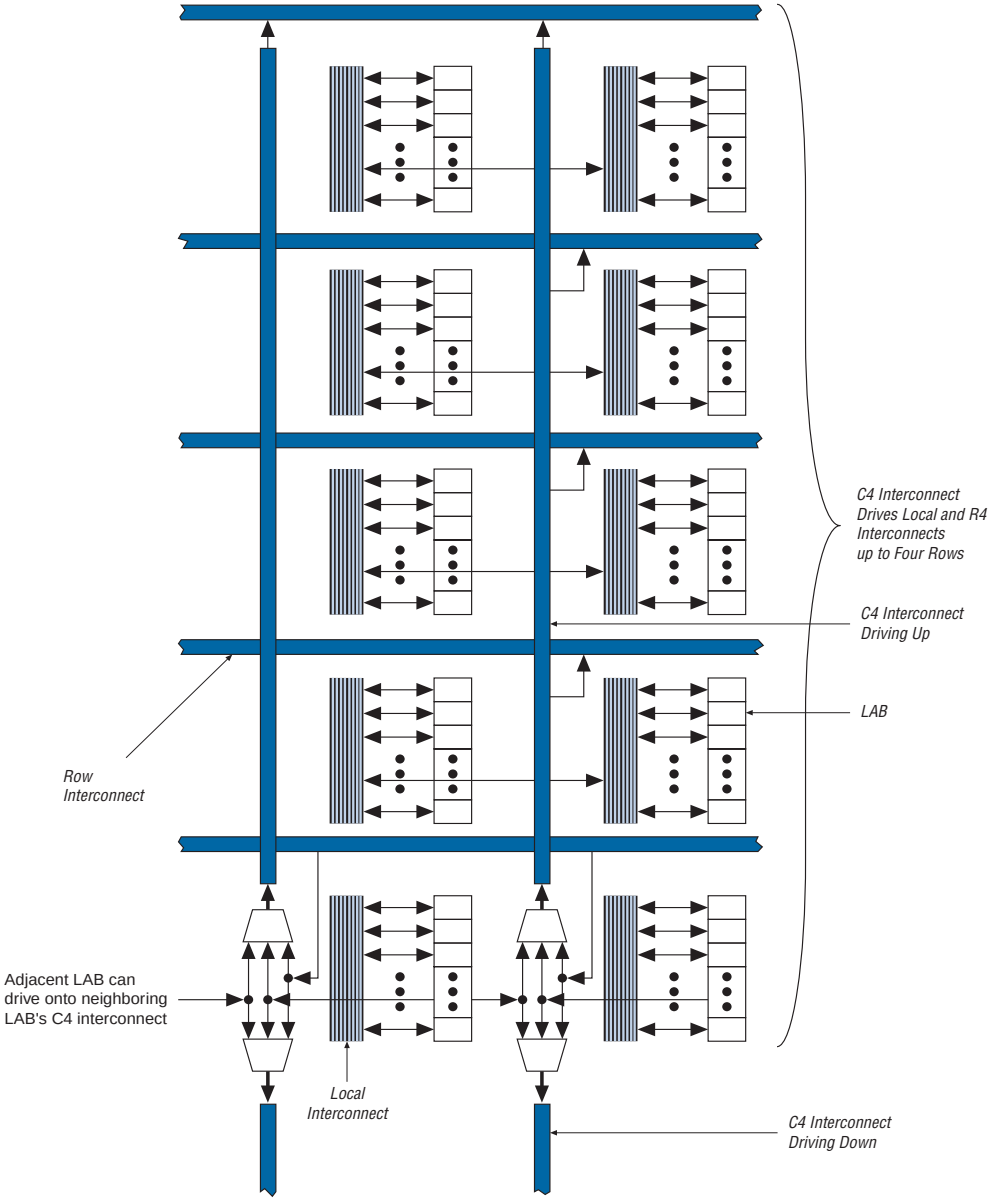
Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	2566
Number of Logic Elements/Cells	25660
Total RAM Bits	1944576
Number of I/O	597
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	780-BBGA
Supplier Device Package	780-FBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1s25f780c5n

Chapter	Date/Version	Changes Made
4	October 2003, v2.1	• Added -8 speed grade information. • Updated performance information in Table 4–36. • Updated timing information in Tables 4–55 through 4–96. • Updated delay information in Tables 4–103 through 4–108. • Updated programmable delay information in Tables 4–100 and 4–103.
	July 2003, v2.0	• Updated clock rates in Tables 4–114 through 4–123. • Updated speed grade information in the introduction on page 4-1. • Corrected figures 4-1 & 4-2 and Table 4-9 to reflect how VID and VOD are specified. • Added note 6 to Table 4-32. • Updated Stratix Performance Table 4-35. • Updated EP1S60 and EP1S80 timing parameters in Tables 4-82 to 4-93. The Stratix timing models are final for all devices. • Updated Stratix IOE programmable delay chains in Tables 4-100 to 4-101. • Added single-ended I/O standard output pin delay adders for loading in Table 4-102. • Added spec for FPLL[10..7]CLK pins in Tables 4-104 and 4-107. • Updated high-speed I/O specification for J=2 in Tables 4-114 and 4-115. • Updated EPLL specification and fast PLL specification in Tables 4-116 to 4-120.
5	September 2004, v2.1	• Updated reference to device pin-outs on page 5–1 to indicate that device pin-outs are no longer included in this manual and are now available on the Altera web site.
	April 2003, v1.0	• No new changes in Stratix Device Handbook v2.0.

Figure 2–11. C4 Interconnect Connections Note (1)



Note to Figure 2–11:
(1) Each C4 interconnect can drive either up or down four rows.

C8 interconnects span eight LABs, M512, or M4K blocks up or down from a source LAB. Every LAB has its own set of C8 interconnects to drive either up or down. C8 interconnect connections between the LABs in a column are similar to the C4 connections shown in Figure 2–11 with the exception that they connect to eight LABs above and below. The C8 interconnects can drive and be driven by all types of architecture blocks similar to C4 interconnects. C8 interconnects can drive each other to extend their range as well as R8 interconnects for column-to-column connections. C8 interconnects are faster than two C4 interconnects.

C16 column interconnects span a length of 16 LABs and provide the fastest resource for long column connections between LABs, TriMatrix memory blocks, DSP blocks, and IOEs. C16 interconnects can cross M-RAM blocks and also drive to row and column interconnects at every fourth LAB. C16 interconnects drive LAB local interconnects via C4 and R4 interconnects and do not drive LAB local interconnects directly.

All embedded blocks communicate with the logic array similar to LAB-to-LAB interfaces. Each block (i.e., TriMatrix memory and DSP blocks) connects to row and column interconnects and has local interconnect regions driven by row and column interconnects. These blocks also have direct link interconnects for fast connections to and from a neighboring LAB. All blocks are fed by the row LAB clocks, `labclk[7..0]`.

Table 2–9. M-RAM Block Configurations (True Dual-Port)				
Port A	Port B			
	64K × 9	32K × 18	16K × 36	8K × 72
64K × 9	✓	✓	✓	✓
32K × 18	✓	✓	✓	✓
16K × 36	✓	✓	✓	✓
8K × 72	✓	✓	✓	✓

The read and write operation of the memory is controlled by the WREN signal, which sets the ports into either read or write modes. There is no separate read enable (RE) signal.

Writing into RAM is controlled by both the WREN and byte enable (byteena) signals for each port. The default value for the byteena signal is high, in which case writing is controlled only by the WREN signal. The byte enables are available for the ×18, ×36, and ×72 modes. In the ×144 simple dual-port mode, the two sets of byteena signals (byteena_a and byteena_b) are combined to form the necessary 16 byte enables. Tables 2–10 and 2–11 summarize the byte selection.

Table 2–10. Byte Enable for M-RAM Blocks *Notes (1), (2)*			
byteena[3..0]	datain ×18	datain ×36	datain ×72
[0] = 1	[8..0]	[8..0]	[8..0]
[1] = 1	[17..9]	[17..9]	[17..9]
[2] = 1	–	[26..18]	[26..18]
[3] = 1	–	[35..27]	[35..27]
[4] = 1	–	–	[44..36]
[5] = 1	–	–	[53..45]
[6] = 1	–	–	[62..54]
[7] = 1	–	–	[71..63]

Input Registers

A bank of optional input registers is located at the input of each multiplier and multiplicand inputs to the multiplier. When these registers are configured for parallel data inputs, they are driven by regular routing resources. You can use a clock signal, asynchronous clear signal, and a clock enable signal to independently control each set of A and B inputs for each multiplier in the DSP block. You select these control signals from a set of four different `clock[3..0]`, `aclr[3..0]`, and `ena[3..0]` signals that drive the entire DSP block.

You can also configure the input registers for a shift register application. In this case, the input registers feed the multiplier and drive two dedicated shift output lines: `shiftoutA` and `shiftoutB`. The shift outputs of one multiplier block directly feed the adjacent multiplier block in the same DSP block (or the next DSP block) as shown in Figure 2–33, to form a shift register chain. This chain can terminate in any block, that is, you can create any length of shift register chain up to 224 registers. You can use the input shift registers for FIR filter applications. One set of shift inputs can provide data for a filter, and the other are coefficients that are optionally loaded in serial or parallel. When implementing 9 × 9- and 18 × 18-bit multipliers, you do not need to implement external shift registers in LAB LEs. You implement all the filter circuitry within the DSP block and its routing resources, saving LE and general routing resources for general logic. External registers are needed for shift register inputs when using 36 × 36-bit multipliers.

Pipeline/Post Multiply Register

The output of 9 × 9- or 18 × 18-bit multipliers can optionally feed a register to pipeline multiply-accumulate and multiply-add/subtract functions. For 36 × 36-bit multipliers, this register will pipeline the multiplier function.

Adder/Output Blocks

The result of the multiplier sub-blocks are sent to the adder/output block which consist of an adder/subtractor/accumulator unit, summation unit, output select multiplexer, and output registers. The results are used to configure the adder/output block as a pure output, accumulator, a simple two-multiplier adder, four-multiplier adder, or final stage of the 36-bit multiplier. You can configure the adder/output block to use output registers in any mode, and must use output registers for the accumulator. The system cannot use adder/output blocks independently of the multiplier. Figure 2–34 shows the adder and output stages.

Output Selection Multiplexer

The outputs from the various elements of the adder/output block are routed through an output selection multiplexer. Based on the DSP block operational mode and user settings, the multiplexer selects whether the output from the multiplier, the adder/subtractor/accumulator, or summation block feeds to the output.

Output Registers

Optional output registers for the DSP block outputs are controlled by four sets of control signals: `clock[3..0]`, `aclr[3..0]`, and `ena[3..0]`. Output registers can be used in any mode.

Modes of Operation

The adder, subtractor, and accumulate functions of a DSP block have four modes of operation:

- Simple multiplier
- Multiply-accumulator
- Two-multipliers adder
- Four-multipliers adder

Each DSP block can only support one mode. Mixed modes in the same DSP block is not supported.

Simple Multiplier Mode

In simple multiplier mode, the DSP block drives the multiplier sub-block result directly to the output with or without an output register. Up to four 18 × 18-bit multipliers or eight 9 × 9-bit multipliers can drive their results directly out of one DSP block. See Figure 2–35.

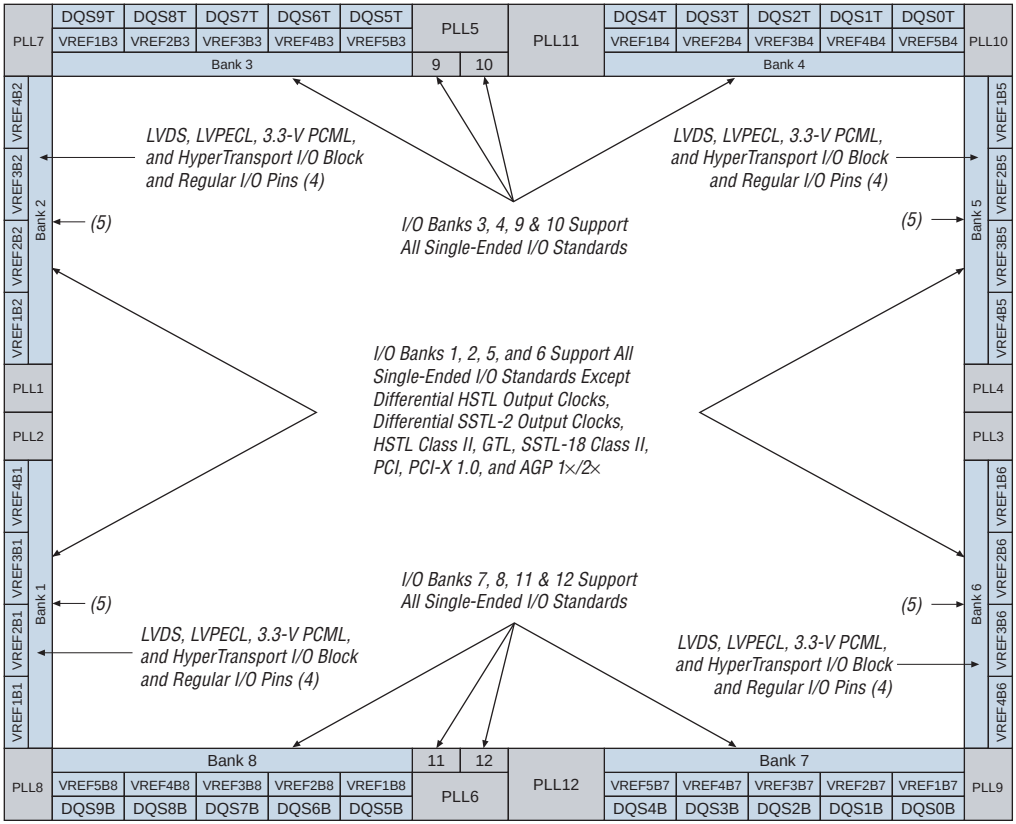
Clock Multiplication & Division

Each Stratix device enhanced PLL provides clock synthesis for PLL output ports using $m/(n \times$ post-scale counter) scaling factors. The input clock is divided by a pre-scale divider, n, and is then multiplied by the m feedback factor. The control loop drives the VCO to match $f_{IN} \times (m/n)$. Each output port has a unique post-scale counter that divides down the high-frequency VCO. For multiple PLL outputs with different frequencies, the VCO is set to the least common multiple of the output frequencies that meets its frequency specifications. Then, the post-scale dividers scale down the output frequency for each output port. For example, if output frequencies required from one PLL are 33 and 66 MHz, set the VCO to 330 MHz (the least common multiple in the VCO's range). There is one pre-scale counter, n, and one multiply counter, m, per PLL, with a range of 1 to 512 on each. There are two post-scale counters (l) for regional clock output ports, four counters (g) for global clock output ports, and up to four counters (e) for external clock outputs, all ranging from 1 to 1024 with a 50% duty cycle setting. The post-scale counters range from 1 to 512 with any non-50% duty cycle setting. The Quartus II software automatically chooses the appropriate scaling factors according to the input frequency, multiplication, and division values entered.

Clock Switchover

To effectively develop high-reliability network systems, clocking schemes must support multiple clocks to provide redundancy. For this reason, Stratix device enhanced PLLs support a flexible clock switchover capability. Figure 2–53 shows a block diagram of the switchover circuit.The switchover circuit is configurable, so you can define how to implement it. Clock-sense circuitry automatically switches from the primary to secondary clock for PLL reference when the primary clock signal is not present.

Figure 2–70. Stratix I/O Banks *Notes (1), (2), (3)*



Notes to Figure 2–70:

(1) Figure 2–70 is a top view of the silicon die. This will correspond to a top-down view for non-flip-chip packages, but will be a reverse view for flip-chip packages.

(2) Figure 2–70 is a graphic representation only. See the device pin-outs on the web (**www.altera.com**) and the Quartus II software for exact locations.

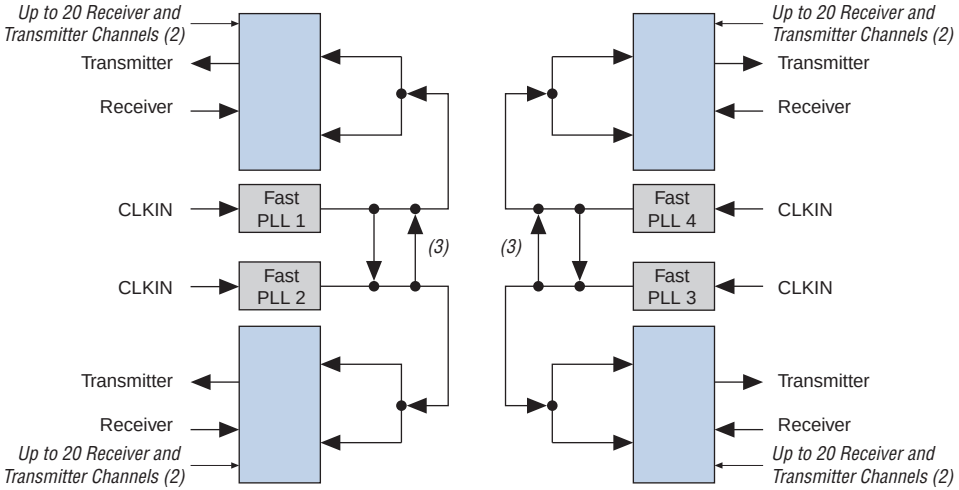
(3) Banks 9 through 12 are enhanced PLL external clock output banks.

(4) If the high-speed differential I/O pins are not used for high-speed differential signaling, they can support all of the I/O standards except HSTL Class I and II, GTL, SSTL-18 Class II, PCI, PCI-X 1.0, and AGP 1× /2×.

(5) For guidelines for placing single-ended I/O pads next to differential I/O pads, see the *Selectable I/O Standards in Stratix and Stratix GX Devices* chapter in the *Stratix Device Handbook, Volume 2*.

The Quartus II MegaWizard® Plug-In Manager only allows the implementation of up to 20 receiver or 20 transmitter channels for each fast PLL. These channels operate at up to 840 Mbps. The receiver and transmitter channels are interleaved such that each I/O bank on the left and right side of the device has one receiver channel and one transmitter channel per LAB row. Figure 2–74 shows the fast PLL and channel layout in EP1S10, EP1S20, and EP1S25 devices. Figure 2–75 shows the fast PLL and channel layout in the EP1S30 to EP1S80 devices.

Figure 2–74. Fast PLL & Channel Layout in the EP1S10, EP1S20 or EP1S25 Devices *Note (1)*



Notes to Figure 2–74:

(1) Wire-bond packages support up to 624 Mbps.

(2) See Table 2–41 for the number of channels each device supports.

(3) There is a multiplexer here to select the PLL clock source. If a PLL uses this multiplexer to clock channels outside of its bank quadrant, those clocked channels support up to 840 Mbps for "high" speed channels and 462 Mbps for "low" speed channels, as labeled in the device pin-outs at **www.altera.com**.

Table 3–1. Stratix JTAG Instructions

JTAG Instruction	Instruction Code	Description
SAMPLE/PRELOAD	00 0000 0101	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins. Also used by the SignalTap II embedded logic analyzer.
EXTEST *(1)*	00 0000 0000	Allows the external circuitry and board-level interconnects to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	11 1111 1111	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through selected devices to adjacent devices during normal device operation.
USERCODE	00 0000 0111	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO.
IDCODE	00 0000 0110	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.
HIGHZ *(1)*	00 0000 1011	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through selected devices to adjacent devices during normal device operation, while tri-stating all of the I/O pins.
CLAMP *(1)*	00 0000 1010	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through selected devices to adjacent devices during normal device operation while holding I/O pins to a state defined by the data in the boundary-scan register.
ICR instructions		Used when configuring an Stratix device via the JTAG port with a MasterBlaster™, ByteBlasterMV™, or ByteBlaster™ II download cable, or when using a Jam File or Jam Byte-Code File via an embedded processor or JRunner.
PULSE_NCONFIG	00 0000 0001	Emulates pulsing the nCONFIG pin low to trigger reconfiguration even though the physical pin is unaffected.
CONFIG_IO	00 0000 1101	Allows configuration of I/O standards through the JTAG chain for JTAG testing. Can be executed before, after, or during configuration. Stops configuration if executed during configuration. Once issued, the CONFIG_IO instruction will hold nSTATUS low to reset the configuration device. nSTATUS is held low until the device is reconfigured.
SignalTap II instructions		Monitors internal device operation with the SignalTap II embedded logic analyzer.

Note to Table 3–1:

(1) Bus hold and weak pull-up resistor features override the high-impedance state of HIGHZ, CLAMP, and EXTEST.

Figure 3–1 shows the timing requirements for the JTAG signals.

Figure 3–1. Stratix JTAG Waveforms

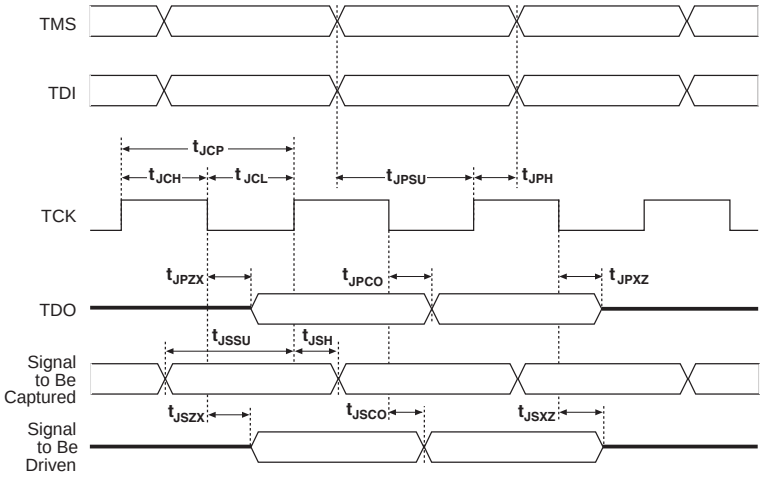
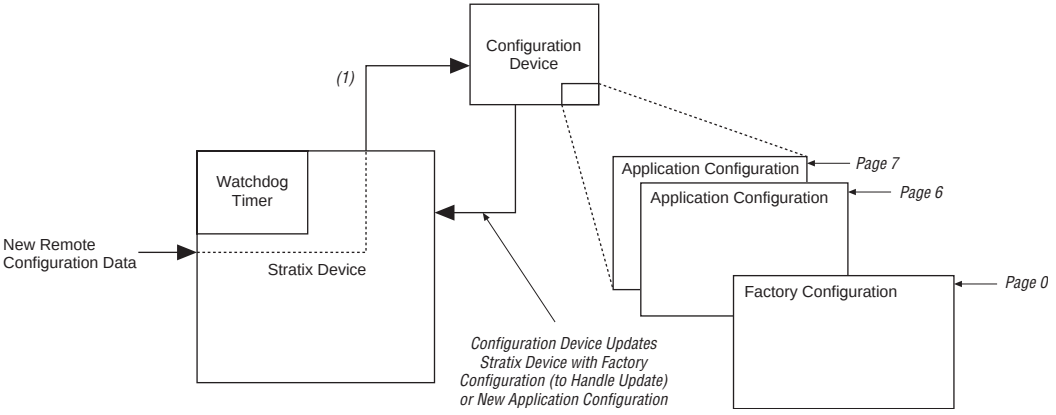


Table 3–4 shows the JTAG timing parameters and values for Stratix devices.

Table 3–4. Stratix JTAG Timing Parameters & Values

Symbol	Parameter	Min	Max	Unit
t_{JCP}	TCK clock period	100		ns
t_{JCH}	TCK clock high time	50		ns
t_{JCL}	TCK clock low time	50		ns
t_{JPSU}	JTAG port setup time	20		ns
t_{JPH}	JTAG port hold time	45		ns
t_{JPCO}	JTAG port clock to output		25	ns
t_{JPZX}	JTAG port high impedance to valid output		25	ns
t_{JPXZ}	JTAG port valid output to high impedance		25	ns
t_{JSSU}	Capture register setup time	20		ns
t_{JSH}	Capture register hold time	45		ns
t_{JSCO}	Update register clock to output		35	ns
t_{JSZX}	Update register high impedance to valid output		35	ns
t_{JSXZ}	Update register valid output to high impedance		35	ns

Figure 3–2. Stratix Device Remote Update



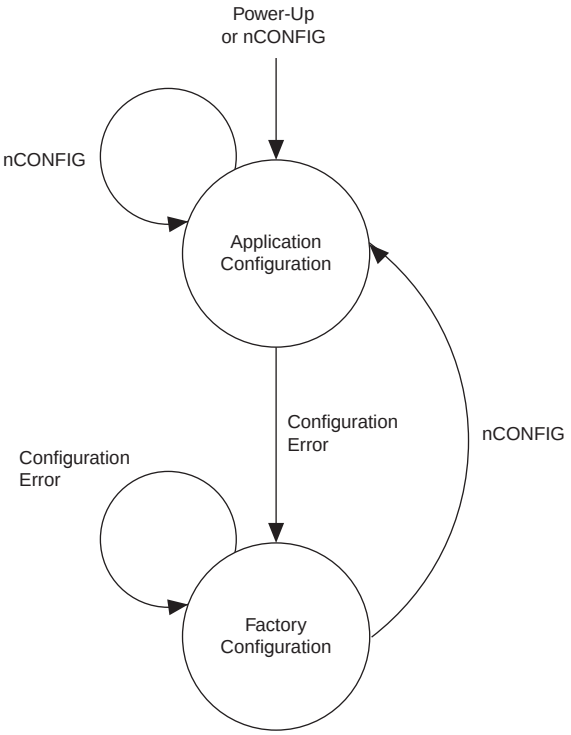
Note to Figure 3–2:

(1) When the Stratix device is configured with the factory configuration, it can handle update data from EPC16, EPC8, or EPC4 configuration device pages and point to the next page in the configuration device.

Local Update Mode

Local update mode is a simplified version of the remote update. This feature is intended for simple systems that need to load a single application configuration immediately upon power up without loading the factory configuration first. Local update designs have only one application configuration to load, so it does not require a factory configuration to determine which application configuration to use. Figure 3–4 shows the transition diagram for local update mode.

Figure 3–4. Local Update Transition Diagram



Stratix Automated Single Event Upset (SEU) Detection

Stratix devices offer on-chip circuitry for automated checking of single event upset (SEU) detection. FPGA devices that operate at high elevations or in close proximity to earth's North or South Pole require periodic checks to ensure continued data integrity. The error detection cyclic redundancy check (CRC) feature controlled by the **Device & Pin Options** dialog box in the Quartus II software uses a 32-bit CRC circuit to ensure data reliability and is one of the best options for mitigating SEU.

Performance

Table 4–36 shows Stratix performance for some common designs. All performance values were obtained with Quartus II software compilation of LPM, or MegaCore® functions for the FIR and FFT designs.

Table 4–36. Stratix Performance (Part 1 of 2) Notes (1), (2)

Applications		Resources Used			Performance				
		LEs	TriMatrix Memory Blocks	DSP Blocks	-5 Speed Grade	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Units
LE	16-to-1 multiplexer *(1)*	22	0	0	407.83	324.56	288.68	228.67	MHz
	32-to-1 multiplexer *(3)*	46	0	0	318.26	255.29	242.89	185.18	MHz
	16-bit counter	16	0	0	422.11	422.11	390.01	348.67	MHz
	64-bit counter	64	0	0	321.85	290.52	261.23	220.5	MHz
TriMatrix memory M512 block	Simple dual-port RAM 32 × 18 bit	0	1	0	317.76	277.62	241.48	205.21	MHz
	FIFO 32 × 18 bit	30	1	0	319.18	278.86	242.54	206.14	MHz
TriMatrix memory M4K block	Simple dual-port RAM 128 × 36 bit	0	1	0	290.86	255.55	222.27	188.89	MHz
	True dual-port RAM 128 × 18 bit	0	1	0	290.86	255.55	222.27	188.89	MHz
	FIFO 128 × 36 bit	34	1	0	290.86	255.55	222.27	188.89	MHz
TriMatrix memory M-RAM block	Single port RAM 4K × 144 bit	1	1	0	255.95	223.06	194.06	164.93	MHz
	Simple dual-port RAM 4K × 144 bit	0	1	0	255.95	233.06	194.06	164.93	MHz
	True dual-port RAM 4K × 144 bit	0	1	0	255.95	233.06	194.06	164.93	MHz
	Single port RAM 8K × 72 bit	0	1	0	278.94	243.19	211.59	179.82	MHz
	Simple dual-port RAM 8K × 72 bit	0	1	0	255.95	223.06	194.06	164.93	MHz
	True dual-port RAM 8K × 72 bit	0	1	0	255.95	223.06	194.06	164.93	MHz
	Single port RAM 16K × 36 bit	0	1	0	280.66	254.32	221.28	188.00	MHz
	Simple dual-port RAM 16K × 36 bit	0	1	0	269.83	237.69	206.82	175.74	MHz

Table 4–47. DSP Block Internal Timing Microparameters (Part 2 of 2)

Symbol	-5		-6		-7		-8		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
$t_{PIPE2OUTREG2ADD}$		2,002		2,203		2,533		2,980	ps
$t_{PIPE2OUTREG4ADD}$		2,899		3,189		3,667		4,314	ps
t_{PD9}		3,709		4,081		4,692		5,520	ps
t_{PD18}		4,795		5,275		6,065		7,135	ps
t_{PD36}		7,495		8,245		9,481		11,154	ps
t_{CLR}	450		500		575		676		ps
t_{CLKHL}	1,350		1,500		1,724		2,029		ps

Table 4–48. M512 Block Internal Timing Microparameters

Symbol	-5		-6		-7		-8		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{M512RC}		3,340		3,816		4,387		5,162	ps
t_{M512WC}		3,138		3,590		4,128		4,860	ps
$t_{M512WERESU}$	110		123		141		166		ps
$t_{M512WEREH}$	34		38		43		51		ps
$t_{M512CLKENSU}$	215		215		247		290		ps
$t_{M512CLKENH}$	–70		–70		–81		–95		ps
$t_{M512DATASU}$	110		123		141		166		ps
$t_{M512DATAH}$	34		38		43		51		ps
$t_{M512WADDRSU}$	110		123		141		166		ps
$t_{M512WADDRH}$	34		38		43		51		ps
$t_{M512RADDRSU}$	110		123		141		166		ps
$t_{M512RADDRH}$	34		38		43		51		ps
$t_{M512DATACO1}$		424		472		541		637	ps
$t_{M512DATACO2}$		3,366		3,846		4,421		5,203	ps
$t_{M512CLKHL}$	1,000		1,111		1,190		1,400		ps
$t_{M512CLR}$	170		189		217		255		ps

Table 4–57. EP1S10 External I/O Timing on Column Pins Using Global Clock Networks *Note (1)*

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	1.647		1.692		1.940		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.619	5.184	2.619	5.515	2.619	5.999	NA	NA	ns
t_{XZ}	2.559	5.058	2.559	5.383	2.559	5.875	NA	NA	ns
t_{ZX}	2.559	5.058	2.559	5.383	2.559	5.875	NA	NA	ns
$t_{INSUPLL}$	1.239		1.229		1.374		NA		ns
t_{INHPLL}	0.000		0.000		0.000		NA		ns
$t_{OUTCOPLL}$	1.109	2.372	1.109	2.436	1.109	2.492	NA	NA	ns
t_{XZPLL}	1.049	2.246	1.049	2.304	1.049	2.368	NA	NA	ns
t_{ZXPLL}	1.049	2.246	1.049	2.304	1.049	2.368	NA	NA	ns

Table 4–58. EP1S10 External I/O Timing on Row Pin Using Fast Regional Clock Network *Note (1)*

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.212		2.403		2.759		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.391	4.838	2.391	5.159	2.391	5.569	NA	NA	ns
t_{XZ}	2.418	4.892	2.418	5.215	2.418	5.637	NA	NA	ns
t_{ZX}	2.418	4.892	2.418	5.215	2.418	5.637	NA	NA	ns

Table 4–101. Reporting Methodology For Maximum Timing For Single-Ended Output Pins (Part 2 of 2)
Notes (1), (2), (3)

I/O Standard	Loading and Termination							Measurement Point
	R_{UP} Ω	R_{DN} Ω	R_S Ω	R_T Ω	V_{CCIO} (V)	VTT (V)	C_L (pF)	V_{MEAS}
3.3-V SSTL-3 Class I	–	–	25	50	2.950	1.250	30	1.250
2.5-V SSTL-2 Class II	–	–	25	25	2.370	1.110	30	1.110
2.5-V SSTL-2 Class I	–	–	25	50	2.370	1.110	30	1.110
1.8-V SSTL-18 Class II	–	–	25	25	1.650	0.760	30	0.760
1.8-V SSTL-18 Class I	–	–	25	50	1.650	0.760	30	0.760
1.5-V HSTL Class II	–	–	0	25	1.400	0.700	20	0.680
1.5-V HSTL Class I	–	–	0	50	1.400	0.700	20	0.680
1.8-V HSTL Class II	–	–	0	25	1.650	0.700	20	0.880
1.8-V HSTL Class I	–	–	0	50	1.650	0.700	20	0.880
3.3-V PCI *(4)*	–/25	25/–	0	–	2.950	2.950	10	0.841/1.814
3.3-V PCI-X 1.0 *(4)*	–/25	25/–	0	–	2.950	2.950	10	0.841/1.814
3.3-V Compact PCI *(4)*	–/25	25/–	0	–	2.950	2.950	10	0.841/1.814
3.3-V AGP 1X *(4)*	–/25	25/–	0	–	2.950	2.950	10	0.841/1.814
3.3-V CTT	–	–	25	50	2.050	1.350	30	1.350

Notes to Table 4–101:

(1) Input measurement point at internal node is $0.5 \times V_{CCINT}$.
(2) Output measuring point for data is V_{MEAS}.
(3) Input stimulus edge rate is 0 to V_{CCINT} in 0.5 ns (internal signal) from the driver preceding the IO buffer.
(4) The first value is for output rising edge and the second value is for output falling edge. The hyphen (-) indicates infinite resistance or disconnection.

Table 4–122. Stratix Maximum Output Clock Rate for PLL[5, 6, 11, 12] Pins in Wire-Bond Packages (Part 1 of 2)

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	175	150	150	MHz
2.5 V	175	150	150	MHz
1.8 V	175	150	150	MHz
1.5 V	175	150	150	MHz
LVCMOS	175	150	150	MHz
GTL	125	100	100	MHz
GTL+	125	100	100	MHz
SSTL-3 Class I	110	90	90	MHz
SSTL-3 Class II	133	125	125	MHz
SSTL-2 Class I	166	133	133	MHz
SSTL-2 Class II	133	100	100	MHz
SSTL-18 Class I	110	100	100	MHz
SSTL-18 Class II	110	100	100	MHz
1.5-V HSTL Class I	167	167	167	MHz
1.5-V HSTL Class II	167	133	133	MHz
1.8-V HSTL Class I	167	167	167	MHz
1.8-V HSTL Class II	167	133	133	MHz
3.3-V PCI	167	167	167	MHz
3.3-V PCI-X 1.0	167	133	133	MHz
Compact PCI	175	150	150	MHz
AGP 1×	175	150	150	MHz
AGP 2×	175	150	150	MHz
CTT	125	100	100	MHz
Differential 1.5-V HSTL C1	167	133	133	MHz
Differential 1.8-V HSTL Class I	167	167	167	MHz
Differential 1.8-V HSTL Class II	167	133	133	MHz
Differential SSTL-2 *(1)*	110	100	100	MHz
LVPECL *(2)*	311	275	275	MHz
PCML *(2)*	250	200	200	MHz

Table 4–130. Enhanced PLL Specifications for -8 Speed Grade (Part 3 of 3)

Symbol	Parameter	Min	Typ	Max	Unit
t_{LSKEW}	Clock skew between two external clock outputs driven by the same counter		±50		ps
t_{SKEW}	Clock skew between two external clock outputs driven by the different counters with the same settings		±75		ps
f_{SS}	Spread spectrum modulation frequency	30		150	kHz
% spread	Percentage spread for spread spectrum frequency *(10)*	0.5		0.6	%
t_{ARESET}	Minimum pulse width on `areset` signal	10			ns

Notes to Tables 4–127 through 4–130:

(1) The minimum input clock frequency to the PFD (f_{IN}/N) must be at least 3 MHz for Stratix device enhanced PLLs.
(2) Use this equation ($f_{OUT} = f_{IN} * ml(n \times$ post-scale counter)) in conjunction with the specified f_{INPFD} and f_{VCO} ranges to determine the allowed PLL settings.
(3) See "Maximum Input & Output Clock Rates" on page 4–76.
(4) t_{FCOMP} can also equal 50% of the input clock period multiplied by the pre-scale divider *n* (whichever is less).
(5) This parameter is timing analyzed by the Quartus II software because the `scanclk` and `scandata` ports can be driven by the logic array.
(6) Actual jitter performance may vary based on the system configuration.
(7) Total required time to reconfigure and lock is equal to $t_{DLOCK} + t_{CONFIG}$. If only post-scale counters and delays are changed, then t_{DLOCK} is equal to 0.
(8) When using the spread-spectrum feature, the minimum VCO frequency is 500 MHz. The maximum VCO frequency is determined by the speed grade selected.
(9) Lock time is a function of PLL configuration and may be significantly faster depending on bandwidth settings or feedback counter change increment.
(10) Exact, user-controllable value depends on the PLL settings.
(11) The LOCK circuit on Stratix PLLs does not work for industrial devices below -20C unless the PFD frequency > 200 MHz. See the *Stratix FPGA Errata Sheet* for more information on the PLL.