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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	4125
Number of Logic Elements/Cells	41250
Total RAM Bits	3423744
Number of I/O	615
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	780-BBGA, FCBGA
Supplier Device Package	780-FBGA (29x29)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=ep1s40f780c7n

2. Stratix Architecture

S51002-3.2

Functional Description

Stratix® devices contain a two-dimensional row- and column-based architecture to implement custom logic. A series of column and row interconnects of varying length and speed provide signal interconnects between logic array blocks (LABs), memory block structures, and DSP blocks.

The logic array consists of LABs, with 10 logic elements (LEs) in each LAB. An LE is a small unit of logic providing efficient implementation of user logic functions. LABs are grouped into rows and columns across the device.

M512 RAM blocks are simple dual-port memory blocks with 512 bits plus parity (576 bits). These blocks provide dedicated simple dual-port or single-port memory up to 18-bits wide at up to 318 MHz. M512 blocks are grouped into columns across the device in between certain LABs.

M4K RAM blocks are true dual-port memory blocks with 4K bits plus parity (4,608 bits). These blocks provide dedicated true dual-port, simple dual-port, or single-port memory up to 36-bits wide at up to 291 MHz. These blocks are grouped into columns across the device in between certain LABs.

M-RAM blocks are true dual-port memory blocks with 512K bits plus parity (589,824 bits). These blocks provide dedicated true dual-port, simple dual-port, or single-port memory up to 144-bits wide at up to 269 MHz. Several M-RAM blocks are located individually or in pairs within the device's logic array.

Digital signal processing (DSP) blocks can implement up to either eight full-precision 9 × 9-bit multipliers, four full-precision 18 × 18-bit multipliers, or one full-precision 36 × 36-bit multiplier with add or subtract features. These blocks also contain 18-bit input shift registers for digital signal processing applications, including FIR and infinite impulse response (IIR) filters. DSP blocks are grouped into two columns in each device.

Each Stratix device I/O pin is fed by an I/O element (IOE) located at the end of LAB rows and columns around the periphery of the device. I/O pins support numerous single-ended and differential I/O standards. Each IOE contains a bidirectional I/O buffer and six registers for registering input, output, and output-enable signals. When used with

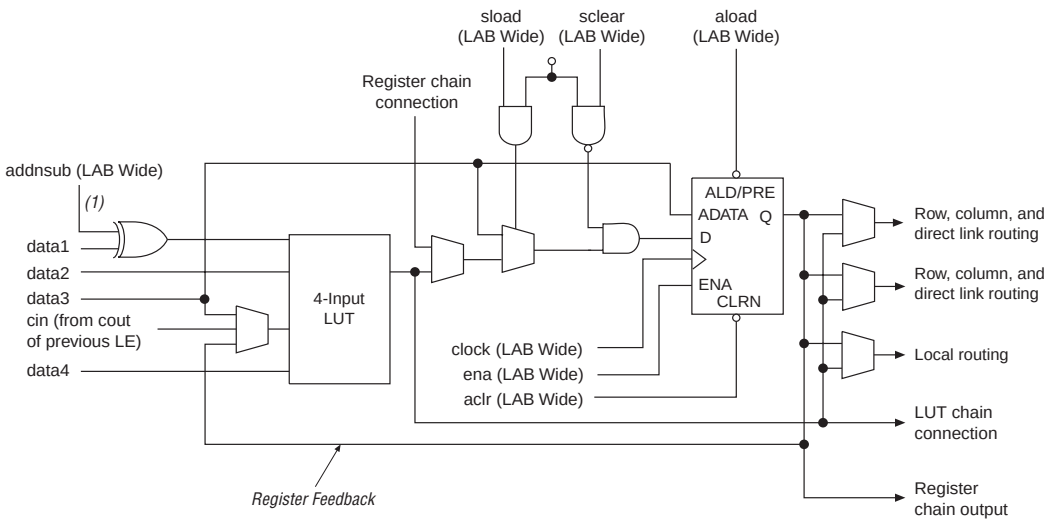
asynchronous preset load, synchronous clear, synchronous load, and clock enable control for the register. These LAB-wide signals are available in all LE modes. The addnsub control signal is allowed in arithmetic mode.

The Quartus II software, in conjunction with parameterized functions such as library of parameterized modules (LPM) functions, automatically chooses the appropriate mode for common functions such as counters, adders, subtractors, and arithmetic functions. If required, you can also create special-purpose functions that specify which LE operating mode to use for optimal performance.

Normal Mode

The normal mode is suitable for general logic applications and combinatorial functions. In normal mode, four data inputs from the LAB local interconnect are inputs to a four-input LUT (see Figure 2–6). The Quartus II Compiler automatically selects the carry-in or the data3 signal as one of the inputs to the LUT. Each LE can use LUT chain connections to drive its combinatorial output directly to the next LE in the LAB. Asynchronous load data for the register comes from the data3 input of the LE. LEs in normal mode support packed registers.

Figure 2–6. LE in Normal Mode



Note to Figure 2–6:
(1) This signal is only allowed in normal mode if the LE is at the end of an adder/subtractor chain.

asynchronous load, and clear signals. An asynchronous clear signal takes precedence if both signals are asserted simultaneously. Each LAB supports up to two clears and one preset signal.

In addition to the clear and preset ports, Stratix devices provide a chip-wide reset pin (DEV_CLRn) that resets all registers in the device. An option set before compilation in the Quartus II software controls this pin. This chip-wide reset overrides all other control signals.

MultiTrack Interconnect

In the Stratix architecture, connections between LEs, TriMatrix memory, DSP blocks, and device I/O pins are provided by the MultiTrack interconnect structure with DirectDrive™ technology. The MultiTrack interconnect consists of continuous, performance-optimized routing lines of different lengths and speeds used for inter- and intra-design block connectivity. The Quartus II Compiler automatically places critical design paths on faster interconnects to improve design performance.

DirectDrive technology is a deterministic routing technology that ensures identical routing resource usage for any function regardless of placement within the device. The MultiTrack interconnect and DirectDrive technology simplify the integration stage of block-based designing by eliminating the re-optimization cycles that typically follow design changes and additions.

The MultiTrack interconnect consists of row and column interconnects that span fixed distances. A routing structure with fixed length resources for all devices allows predictable and repeatable performance when migrating through different device densities. Dedicated row interconnects route signals to and from LABs, DSP blocks, and TriMatrix memory within the same row. These row resources include:

- Direct link interconnects between LABs and adjacent blocks.
- R4 interconnects traversing four blocks to the right or left.
- R8 interconnects traversing eight blocks to the right or left.
- R24 row interconnects for high-speed access across the length of the device.

The direct link interconnect allows an LAB, DSP block, or TriMatrix memory block to drive into the local interconnect of its left and right neighbors and then back into itself. Only one side of a M-RAM block interfaces with direct link and row interconnects. This provides fast communication between adjacent LABs and/or blocks without using row interconnect resources.

The R4 interconnects span four LABs, three LABs and one M512 RAM block, two LABs and one M4K RAM block, or two LABs and one DSP block to the right or left of a source LAB. These resources are used for fast

can drive other R8 interconnects to extend their range as well as C8 interconnects for row-to-row connections. One R8 interconnect is faster than two R4 interconnects connected together.

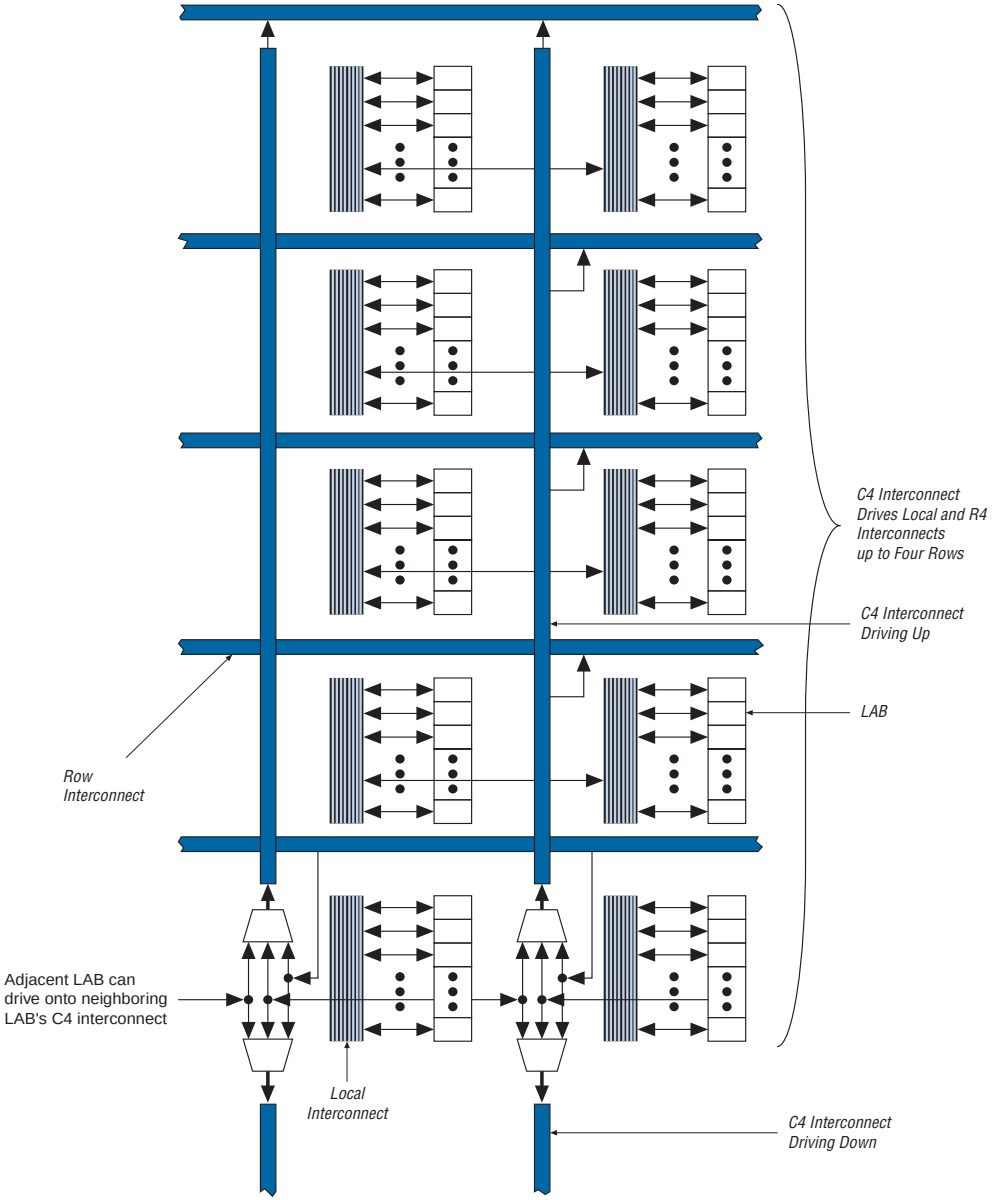
R24 row interconnects span 24 LABs and provide the fastest resource for long row connections between LABs, TriMatrix memory, DSP blocks, and IOEs. The R24 row interconnects can cross M-RAM blocks. R24 row interconnects drive to other row or column interconnects at every fourth LAB and do not drive directly to LAB local interconnects. R24 row interconnects drive LAB local interconnects via R4 and C4 interconnects. R24 interconnects can drive R24, R4, C16, and C4 interconnects.

The column interconnect operates similarly to the row interconnect and vertically routes signals to and from LABs, TriMatrix memory, DSP blocks, and IOEs. Each column of LABs is served by a dedicated column interconnect, which vertically routes signals to and from LABs, TriMatrix memory and DSP blocks, and horizontal IOEs. These column resources include:

- LUT chain interconnects within an LAB
- Register chain interconnects within an LAB
- C4 interconnects traversing a distance of four blocks in up and down direction
- C8 interconnects traversing a distance of eight blocks in up and down direction
- C16 column interconnects for high-speed vertical routing through the device

Stratix devices include an enhanced interconnect structure within LABs for routing LE output to LE input connections faster using LUT chain connections and register chain connections. The LUT chain connection allows the combinatorial output of an LE to directly drive the fast input of the LE right below it, bypassing the local interconnect. These resources can be used as a high-speed connection for wide fan-in functions from LE 1 to LE 10 in the same LAB. The register chain connection allows the register output of one LE to connect directly to the register input of the next LE in the LAB for fast shift registers. The Quartus II Compiler automatically takes advantage of these resources to improve utilization and performance. Figure 2–10 shows the LUT chain and register chain interconnects.

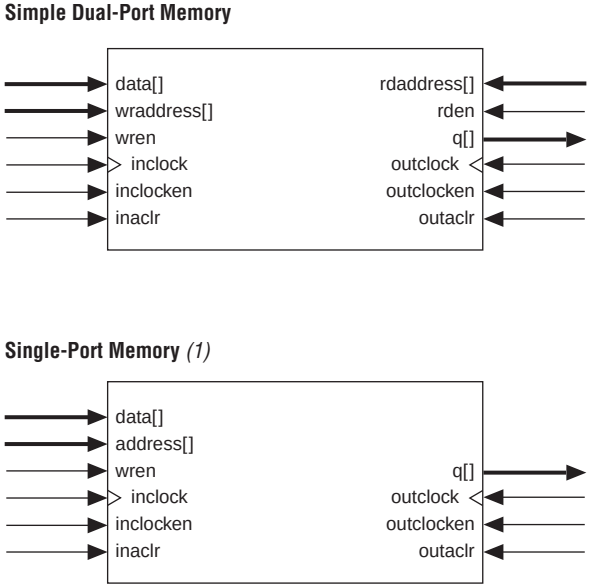
Figure 2–11. C4 Interconnect Connections Note (1)



Note to Figure 2–11:
(1) Each C4 interconnect can drive either up or down four rows.

In addition to true dual-port memory, the memory blocks support simple dual-port and single-port RAM. Simple dual-port memory supports a simultaneous read and write and can either read old data before the write occurs or just read the don't care bits. Single-port memory supports non-simultaneous reads and writes, but the q[] port will output the data once it has been written to the memory (if the outputs are not registered) or after the next rising edge of the clock (if the outputs are registered). For more information, see Chapter 2, TriMatrix Embedded Memory Blocks in Stratix & Stratix GX Devices of the *Stratix Device Handbook, Volume 2*. Figure 2–13 shows these different RAM memory port configurations for TriMatrix memory.

Figure 2–13. Simple Dual-Port & Single-Port Memory Configurations



Note to Figure 2–13:

(1) Two single-port memory blocks can be implemented in a single M4K block as long as each of the two independent block sizes is equal to or less than half of the M4K block size.

The memory blocks also enable mixed-width data ports for reading and writing to the RAM ports in dual-port RAM configuration. For example, the memory block can be written in ×1 mode at port A and read out in ×16 mode from port B.

Table 2–11. M-RAM Combined Byte Selection for ×144 Mode Notes (1), (2)	
byteena[15..0]	**datain ×144**
[0] = 1	[8..0]
[1] = 1	[17..9]
[2] = 1	[26..18]
[3] = 1	[35..27]
[4] = 1	[44..36]
[5] = 1	[53..45]
[6] = 1	[62..54]
[7] = 1	[71..63]
[8] = 1	[80..72]
[9] = 1	[89..81]
[10] = 1	[98..90]
[11] = 1	[107..99]
[12] = 1	[116..108]
[13] = 1	[125..117]
[14] = 1	[134..126]
[15] = 1	[143..135]

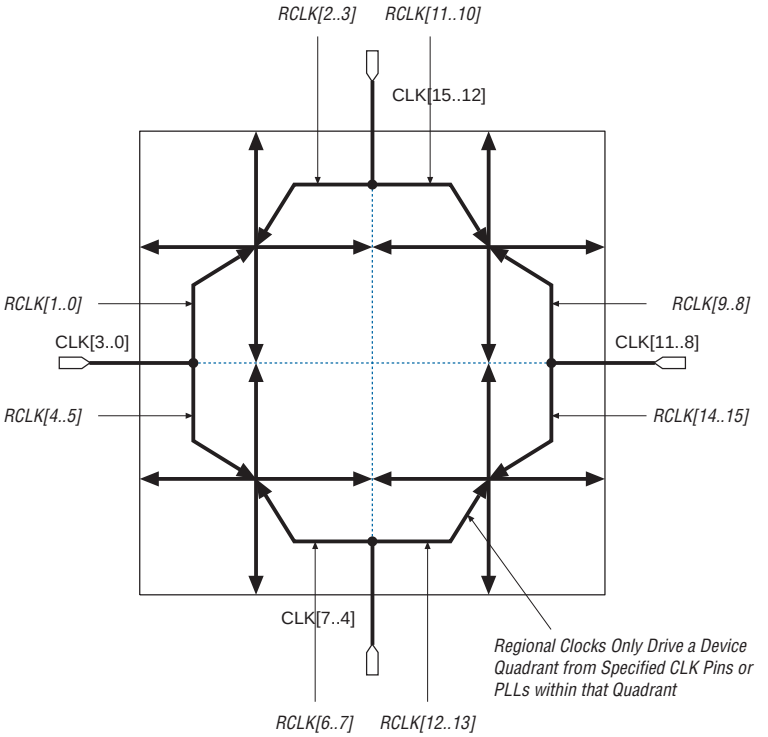
Notes to Tables 2–10 and 2–11:

(1) Any combination of byte enables is possible.

(2) Byte enables can be used in the same manner with 8-bit words, i.e., in ×16, ×32, ×64, and ×128 modes.

Similar to all RAM blocks, M-RAM blocks can have different clocks on their inputs and outputs. All input registers—`renwe`, `datain`, address, and byte enable registers—are clocked together from either of the two clocks feeding the block. The output register can be bypassed. The eight `labclk` signals or local interconnect can drive the control signals for the A and B ports of the M-RAM block. LEs can also control the `clock_a`, `clock_b`, `renwe_a`, `renwe_b`, `clr_a`, clr_b, `clocken_a`, and `clocken_b` signals as shown in Figure 2–19.

Figure 2–43. Regional Clocks

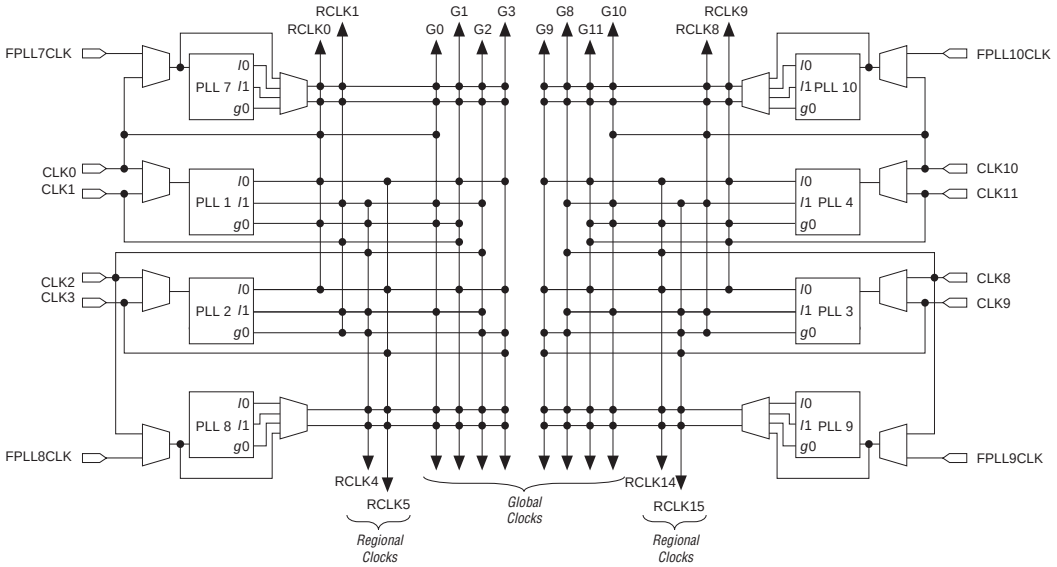


Fast Regional Clock Network

In EP1S25, EP1S20, and EP1S10 devices, there are two fast regional clock networks, FCLK[1..0], within each quadrant, fed by input pins that can connect to fast regional clock networks (see Figure 2–44). In EP1S30 and larger devices, there are two fast regional clock networks within each half-quadrant (see Figure 2–45). Dual-purpose FCLK pins drive the fast clock networks. All devices have eight FCLK pins to drive fast regional clock networks. Any I/O pin can drive a clock or control signal onto any fast regional clock network with the addition of a delay. This signal is driven via the I/O interconnect. The fast regional clock networks can also be driven from internal logic elements.

Figure 2–50 shows the global and regional clocking from the PLL outputs and the CLK pins.

Figure 2–50. Global & Regional Clock Connections from Side Pins & Fast PLL Outputs *Note (1), (2)*



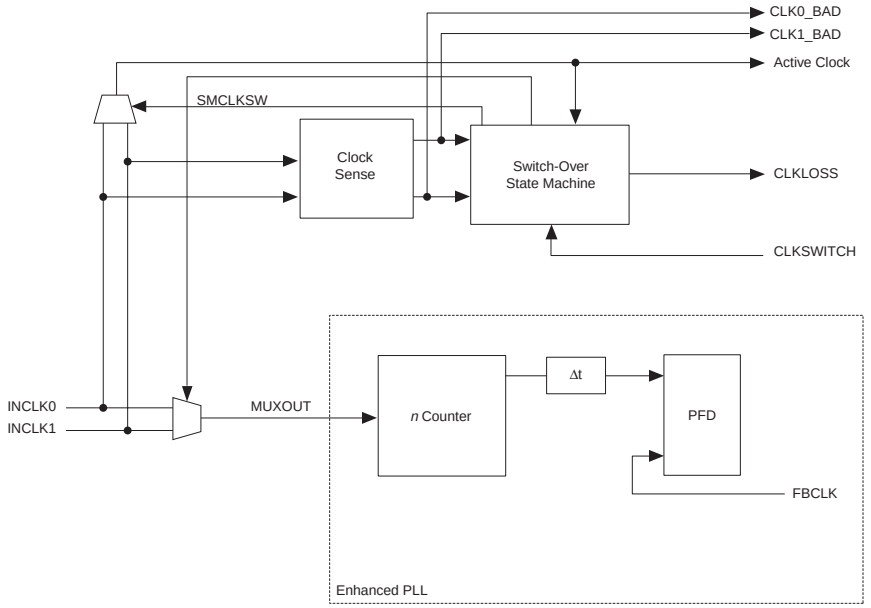
Notes to Figure 2–50:

(1) PLLs 1 to 4 and 7 to 10 are fast PLLs. PLLs 5, 6, 11, and 12 are enhanced PLLs.

(2) The global or regional clocks in a fast PLL's quadrant can drive the fast PLL input. A pin or other PLL must drive the global or regional source. The source cannot be driven by internally generated logic before driving the fast PLL.

Figure 2–51 shows the global and regional clocking from enhanced PLL outputs and top `CLK` pins.

Figure 2–53. Clock Switchover Circuitry



There are two possible ways to use the clock switchover feature.

- Use automatic switchover circuitry for switching between inputs of the same frequency. For example, in applications that require a redundant clock with the same frequency as the primary clock, the switchover state machine generates a signal that controls the multiplexer select input on the bottom of Figure 2–53. In this case, the secondary clock becomes the reference clock for the PLL.
- Use the `clkswitch` input for user- or system-controlled switch conditions. This is possible for same-frequency switchover or to switch between inputs of different frequencies. For example, if `inclk0` is 66 MHz and `inclk1` is 100 MHz, you must control the switchover because the automatic clock-sense circuitry cannot monitor primary and secondary clock frequencies with a frequency difference of more than ±20%. This feature is useful when clock sources can originate from multiple cards on the backplane, requiring a system-controlled switchover between frequencies of operation. You can use `clkswitch` together with the lock signal to trigger the switch from a clock that is running but becomes unstable and cannot be locked onto.

bandwidth is tuned by varying the charge pump current, loop filter resistor value, high frequency capacitor value, and *m* counter value. You can manually adjust these values if desired. Bandwidth is programmable from 200 kHz to 1.5 MHz.

External Clock Outputs

Enhanced PLLs 5 and 6 each support up to eight single-ended clock outputs (or four differential pairs). Differential SSTL and HSTL outputs are implemented using 2 single-ended output buffers which are programmed to have opposite polarity. In Quartus II software, simply assign the appropriate differential I/O standard and the software will implement the inversion. See Figure 2–55.

Table 2–37. EP1S10, EP1S20 & EP1S25 Device Differential Channels (Part 2 of 2) Note (1)

Device	Package	Transmitter/ Receiver	Total Channels	Maximum Speed (Mbps)	Center Fast PLLs			
					PLL 1	PLL 2	PLL 3	PLL 4
EP1S25	672-pin FineLine BGA 672-pin BGA	Transmitter (2)	56	624 (4)	14	14	14	14
				624 (3)	28	28	28	28
		Receiver	58	624 (4)	14	15	15	14
				624 (3)	29	29	29	29
	780-pin FineLine BGA	Transmitter (2)	70	840 (4)	18	17	17	18
				840 (3)	35	35	35	35
		Receiver	66	840 (4)	17	16	16	17
				840 (3)	33	33	33	33
	1,020-pin FineLine BGA	Transmitter (2)	78	840 (4)	19	20	20	19
				840 (3)	39	39	39	39
		Receiver	78	840 (4)	19	20	20	19
				840 (3)	39	39	39	39

Notes to Table 2–37:

(1) The first row for each transmitter or receiver reports the number of channels driven directly by the PLL. The second row below it shows the maximum channels a PLL can drive if cross bank channels are used from the adjacent center PLL. For example, in the 484-pin FineLine BGA EP1S10 device, PLL 1 can drive a maximum of five channels at 840 Mbps or a maximum of 10 channels at 840 Mbps. The Quartus II software may also merge receiver and transmitter PLLs when a receiver is driving a transmitter. In this case, one fast PLL can drive both the maximum numbers of receiver and transmitter channels.

(2) The number of channels listed includes the transmitter clock output (`tx_outclock`) channel. If the design requires a DDR clock, it can use an extra data channel.

(3) These channels span across two I/O banks per side of the device. When a center PLL clocks channels in the opposite bank on the same side of the device it is called cross-bank PLL support. Both center PLLs can clock cross-bank channels simultaneously if, for example, PLL_1 is clocking all receiver channels and PLL_2 is clocking all transmitter channels. You cannot have two adjacent PLLs simultaneously clocking cross-bank receiver channels or two adjacent PLLs simultaneously clocking transmitter channels. Cross-bank allows for all receiver channels on one side of the device to be clocked on one clock while all transmitter channels on the device are clocked on the other center PLL. Crossbank PLLs are supported at full-speed, 840 Mbps. For wire-bond devices, the full-speed is 624 Mbps.

(4) These values show the channels available for each PLL without crossing another bank.

When you span two I/O banks using cross-bank support, you can route only two load enable signals total between the PLLs. When you enable `rx_data_align`, you use both `rxloadena` and `txloadena` of a PLL. That leaves no `loadena` for the second PLL.

3. Configuration & Testing

S51003-1.3

IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All Stratix® devices provide JTAG BST circuitry that complies with the IEEE Std. 1149.1a-1990 specification. JTAG boundary-scan testing can be performed either before or after, but not during configuration. Stratix devices can also use the JTAG port for configuration together with either the Quartus® II software or hardware using either Jam Files (**.jam**) or Jam Byte-Code Files (**.jbc**).

Stratix devices support IOE I/O standard setting reconfiguration through the JTAG BST chain. The JTAG chain can update the I/O standard for all input and output pins any time before or during user mode through the `CONFIG_IO` instruction. You can use this ability for JTAG testing before configuration when some of the Stratix pins drive or receive from other devices on the board using voltage-referenced standards. Since the Stratix device may not be configured before JTAG testing, the I/O pins may not be configured for appropriate electrical standards for chip-to-chip communication. Programming those I/O standards via JTAG allows you to fully test the I/O connection to other devices.

The enhanced PLL reconfiguration bits are part of the JTAG chain before configuration and after power-up. After device configuration, the PLL reconfiguration bits are not part of the JTAG chain.

The JTAG pins support 1.5-V/1.8-V or 2.5-V/3.3-V I/O standards. The `TDO` pin voltage is determined by the V_{CCIO} of the bank where it resides. The `VCCSEL` pin selects whether the JTAG inputs are 1.5-V, 1.8-V, 2.5-V, or 3.3-V compatible.

Stratix devices also use the JTAG port to monitor the logic operation of the device with the SignalTap® II embedded logic analyzer. Stratix devices support the JTAG instructions shown in Table 3–1.

The Quartus II software has an Auto Usercode feature where you can choose to use the checksum value of a programming file as the JTAG user code. If selected, the checksum is automatically loaded to the `USERCODE` register. In the Settings dialog box in the Assignments menu, click **Device & Pin Options**, then **General,** and then turn on the **Auto Usercode** option.

Table 4–15. PCI-X 1.0 Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		3.0		3.6	V
V_{IH}	High-level input voltage		0.5 × V_{CCIO}		V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage		–0.5		0.35 × V_{CCIO}	V
V_{IPU}	Input pull-up voltage		0.7 × V_{CCIO}			V
V_{OH}	High-level output voltage	I_{OUT} = –500 µA	0.9 × V_{CCIO}			V
V_{OL}	Low-level output voltage	I_{OUT} = 1,500 µA			0.1 × V_{CCIO}	V

Table 4–16. GTL+ I/O Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{TT}	Termination voltage		1.35	1.5	1.65	V
V_{REF}	Reference voltage		0.88	1.0	1.12	V
V_{IH}	High-level input voltage		V_{REF} + 0.1			V
V_{IL}	Low-level input voltage				V_{REF} – 0.1	V
V_{OL}	Low-level output voltage	I_{OL} = 34 mA *(3)*			0.65	V

Table 4–17. GTL I/O Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{TT}	Termination voltage		1.14	1.2	1.26	V
V_{REF}	Reference voltage		0.74	0.8	0.86	V
V_{IH}	High-level input voltage		V_{REF} + 0.05			V
V_{IL}	Low-level input voltage				V_{REF} – 0.05	V
V_{OL}	Low-level output voltage	I_{OL} = 40 mA *(3)*			0.4	V

Table 4–22. SSTL-3 Class I Specifications (Part 2 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
$V_{IL(AC)}$	Low-level AC input voltage				V_{REF} – 0.4	V
V_{OH}	High-level output voltage	I_{OH} = –8 mA *(3)*	V_{TT} + 0.6			V
V_{OL}	Low-level output voltage	I_{OL} = 8 mA *(3)*			V_{TT} – 0.6	V

Table 4–23. SSTL-3 Class II Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		3.0	3.3	3.6	V
V_{TT}	Termination voltage		V_{REF} – 0.05	V_{REF}	V_{REF} + 0.05	V
V_{REF}	Reference voltage		1.3	1.5	1.7	V
$V_{IH(DC)}$	High-level DC input voltage		V_{REF} + 0.2		V_{CCIO} + 0.3	V
$V_{IL(DC)}$	Low-level DC input voltage		–0.3		V_{REF} – 0.2	V
$V_{IH(AC)}$	High-level AC input voltage		V_{REF} + 0.4			V
$V_{IL(AC)}$	Low-level AC input voltage				V_{REF} – 0.4	V
V_{OH}	High-level output voltage	I_{OH} = –16 mA *(3)*	V_{TT} + 0.8			V
V_{OL}	Low-level output voltage	I_{OL} = 16 mA *(3)*			V_{TT} – 0.8	V

Table 4–24. 3.3-V AGP 2× Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		3.15	3.3	3.45	V
V_{REF}	Reference voltage		0.39 × V_{CCIO}		0.41 × V_{CCIO}	V
V_{IH}	High-level input voltage *(4)*		0.5 × V_{CCIO}		V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage *(4)*				0.3 × V_{CCIO}	V
V_{OH}	High-level output voltage	I_{OUT} = –0.5 mA	0.9 × V_{CCIO}		3.6	V
V_{OL}	Low-level output voltage	I_{OUT} = 1.5 mA			0.1 × V_{CCIO}	V

Table 4–25. 3.3-V AGP 1× Specifications (Part 1 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		3.15	3.3	3.45	V
V_{IH}	High-level input voltage *(4)*		0.5 × V_{CCIO}		V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage *(4)*				0.3 × V_{CCIO}	V

device. Decoupling capacitors were not used in this measurement. To factor in the current for decoupling capacitors, sum up the current for each capacitor using the following equation:

$I = C (dV/dt)$

If the regulator or power supply minimum output current is more than the Stratix device requires, then the device may consume more current than the maximum current listed in Table 4–34. However, the device does not require any more current to successfully power up than what is listed in Table 4–34.

Table 4–34. Stratix Power-Up Current (I_{CCINT}) Requirements Note (1)

Device	Power-Up Current Requirement		Unit
	Typical	Maximum	
EP1S10	250	700	mA
EP1S20	400	1,200	mA
EP1S25	500	1,500	mA
EP1S30	550	1,900	mA
EP1S40	650	2,300	mA
EP1S60	800	2,600	mA
EP1S80	1,000	3,000	mA

Note to Table 4–34:

(1) The maximum test conditions are for 0° C and typical test conditions are for 40° C.

The exact amount of current consumed varies according to the process, temperature, and power ramp rate. Stratix devices typically require less current during power up than shown in Table 4–34. The user-mode current during device operation is generally higher than the power-up current.

The duration of the I_{CCINT} power-up requirement depends on the V_{CCINT} voltage supply rise time. The power-up current consumption drops when the V_{CCINT} supply reaches approximately 0.75 V.

Table 4–97. Output Pin Timing Skew Definitions (Part 2 of 2)	
Symbol	**Definition**
t_{LR_HIO}	Across all HIO banks (1, 2, 5, 6); across four similar type I/O banks
t_{TB_VIO}	Across all VIO banks (3, 4, 7, 8); across four similar type I/O banks
$t_{OVERALL}$	Output timing skew for all I/O pins on the device.

Notes to Table 4–97:
(1) See Figure 4–5 on page 4–57.
(2) See Figure 4–6 on page 4–58.

Table 4–98 shows the I/O skews when using the same global or regional clock to feed IOE registers in I/O banks around each device. These values can be used for calculating the timing budget on the output (write) side of a memory interface. These values already factor in the package skew.

Table 4–98. Output Skew for Stratix by Device Density			
Symbol	**Skew (ps)** *(1)*		
	EP1S10 to EP1S30	**EP1S40**	**EP1S60 & EP1S80**
t_{SB_HIO}	90	290	500
t_{SB_VIO}	160	290	500
t_{SS_HIO}	90	460	600
t_{SS_VIO}	180	520	630
t_{LR_HIO}	150	490	600
t_{TB_VIO}	190	580	670
$t_{OVERALL}$	430	630	880

Note to Table 4–98:
(1) The skew numbers in Table 4–98 account for worst case package skews.

Tables 4–109 and 4–110 show the adder delays for the column and row IOE programmable delays. These delays are controlled with the Quartus II software logic options listed in the Parameter column.

Table 4–109. Stratix IOE Programmable Delays on Column Pins *Note (1)*

Parameter	Setting	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
Decrease input delay to internal cells	Off		3,970		4,367		5,022		5,908	ps
	Small		3,390		3,729		4,288		5,045	ps
	Medium		2,810		3,091		3,554		4,181	ps
	Large		224		235		270		318	ps
	On		224		235		270		318	ps
Decrease input delay to input register	Off		3,900		4,290		4,933		5,804	ps
	On		0		0		0		0	ps
Decrease input delay to output register	Off		1,240		1,364		1,568		1,845	ps
	On		0		0		0		0	ps
Increase delay to output pin	Off		0		0		0		0	ps
	On		397		417		417		417	ps
Increase delay to output enable pin	Off		0		0		0		0	ps
	On		338		372		427		503	ps
Increase output clock enable delay	Off		0		0		0		0	ps
	Small		540		594		683		804	ps
	Large		1,016		1,118		1,285		1,512	ps
	On		1,016		1,118		1,285		1,512	ps
Increase input clock enable delay	Off		0		0		0		0	ps
	Small		540		594		683		804	ps
	Large		1,016		1,118		1,285		1,512	ps
	On		1,016		1,118		1,285		1,512	ps
Increase output enable clock enable delay	Off		0		0		0		0	ps
	Small		540		594		683		804	ps
	Large		1,016		1,118		1,285		1,512	ps
	On		1,016		1,118		1,285		1,512	ps
Increase t_{ZX} delay to output pin	Off		0		0		0		0	ps
	On		2,199		2,309		2,309		2,309	ps

Table 4–122. Stratix Maximum Output Clock Rate for PLL[5, 6, 11, 12] Pins in Wire-Bond Packages (Part 2 of 2)

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVDS *(2)*	311	275	275	MHz
HyperTransport technology *(2)*	311	275	275	MHz

Table 4–123. Stratix Maximum Output Clock Rate (Using I/O Pins) for PLL[1, 2, 3, 4] Pins in Wire-Bond Packages (Part 1 of 2)

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	200	175	175	MHz
2.5 V	200	175	175	MHz
1.8 V	200	175	175	MHz
1.5 V	200	175	175	MHz
LVCMOS	200	175	175	MHz
GTL	125	100	100	MHz
GTL+	125	100	100	MHz
SSTL-3 Class I	110	90	90	MHz
SSTL-3 Class II	150	133	133	MHz
SSTL-2 Class I	90	80	80	MHz
SSTL-2 Class II	110	100	100	MHz
SSTL-18 Class I	110	100	100	MHz
SSTL-18 Class II	110	100	100	MHz
1.5-V HSTL Class I	225	200	200	MHz
1.5-V HSTL Class II	200	167	167	MHz
1.8-V HSTL Class I	225	200	200	MHz
1.8-V HSTL Class II	200	167	167	MHz
3.3-V PCI	200	175	175	MHz
3.3-V PCI-X 1.0	200	175	175	MHz
Compact PCI	200	175	175	MHz
AGP 1×	200	175	175	MHz
AGP 2×	200	175	175	MHz
CTT	125	100	100	MHz
LVPECL *(2)*	311	270	270	MHz
PCML *(2)*	400	311	311	MHz

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