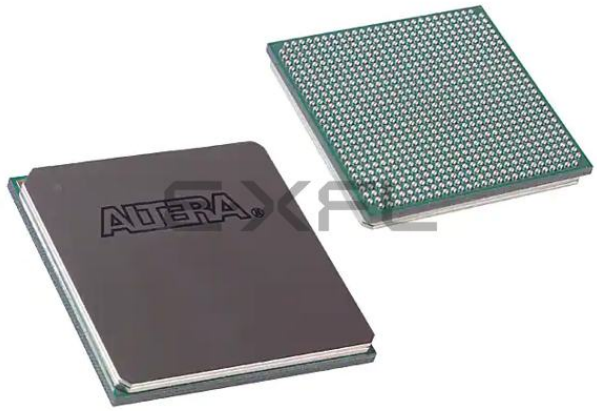

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	4125
Number of Logic Elements/Cells	41250
Total RAM Bits	3423744
Number of I/O	615
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	780-BBGA, FCBGA
Supplier Device Package	780-FBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1s40f780i6n

Chapter Revision Dates

The chapters in this book, *Stratix Device Handbook, Volume 1*, were revised on the following dates. Where chapters or groups of chapters are available separately, part numbers are listed.

Chapter 1. Introduction
Revised: *July 2005*
Part number: *S51001-3.2*

Chapter 2. Stratix Architecture
Revised: *July 2005*
Part number: *S51002-3.2*

Chapter 3. Configuration & Testing
Revised: *July 2005*
Part number: *S51003-1.3*

Chapter 4. DC & Switching Characteristics
Revised: *January 2006*
Part number: *S51004-3.4*

Chapter 5. Reference & Ordering Information
Revised: *September 2004*
Part number: *S51005-2.1*

Chapter	Date/Version	Changes Made
2	July 2003, v2.0	• Added reference on page 2-73 to Figures 2-50 and 2-51 for RCLK connections. • Updated ranges for EPLL post-scale and pre-scale dividers on page 2-85. • Updated PLL Reconfiguration frequency from 25 to 22 MHz on page 2-87. • New requirement to assert are set signal each PLL when it has to re-acquire lock on either a new clock after loss of lock (page 2-96). • Updated max input frequency for CLK[1,3,8,10] from 462 to 500, Table 2-24. • Renamed impedance matching to series termination throughout. • Updated naming convention for DQS pins on page 2-112 to match pin tables. • Added DDR SDRAM Performance Specification on page 2-117. • Added external reference resistor values for terminator technology (page 2-136). • Added Terminator Technology Specification on pages 2-137 and 2-138. • Updated Tables 2-45 to 2-49 to reflect PLL cross-bank support for high speed differential channels at full speed. • Wire bond package performance specification for "high" speed channels was increased to 624 Mbps from 462 Mbps throughout chapter.
3	July 2005, v1.3	• Updated "Operating Modes" section. • Updated "Temperature Sensing Diode" section. • Updated "IEEE Std. 1149.1 (JTAG) Boundary-Scan Support" section. • Updated "Configuration" section.
	January 2005, v1.2	• Updated limits for JTAG chain of devices.
	September 2004, v1.1	• Added new section, "Stratix Automated Single Event Upset (SEU) Detection" on page 3–12. • Updated description of "Custom-Built Circuitry" on page 3–13.
	April 2003, v1.0	• No new changes in *Stratix Device Handbook* v2.0.
4	January 2006, v3.4	• Added Table 4–135.
	July 2005, v3.3	• Updated Tables 4–6 and 4–30. • Updated Tables 4–103 through 4–108. • Updated Tables 4–114 through 4–124. • Updated Table 4–129. • Added Table 4–130.

functions. Another special packing mode allows the register output to feed back into the LUT of the same LE so that the register is packed with its own fan-out LUT. This provides another mechanism for improved fitting. The LE can also drive out registered and unregistered versions of the LUT output.

LUT Chain & Register Chain

In addition to the three general routing outputs, the LEs within an LAB have LUT chain and register chain outputs. LUT chain connections allow LUTs within the same LAB to cascade together for wide input functions. Register chain outputs allow registers within the same LAB to cascade together. The register chain output allows an LAB to use LUTs for a single combinatorial function and the registers to be used for an unrelated shift register implementation. These resources speed up connections between LABs while saving local interconnect resources. See "MultiTrack Interconnect" on page 2–14 for more information on LUT chain and register chain connections.

addnsub Signal

The LE's dynamic adder/subtractor feature saves logic resources by using one set of LEs to implement both an adder and a subtractor. This feature is controlled by the LAB-wide control signal `addnsub`. The `addnsub` signal sets the LAB to perform either A + B or A – B. The LUT computes addition, and subtraction is computed by adding the two's complement of the intended subtractor. The LAB-wide signal converts to two's complement by inverting the B bits within the LAB and setting carry-in = 1 to add one to the least significant bit (LSB). The LSB of an adder/subtractor must be placed in the first LE of the LAB, where the LAB-wide `addnsub` signal automatically sets the carry-in to 1. The Quartus II Compiler automatically places and uses the adder/subtractor feature when using adder/subtractor parameterized functions.

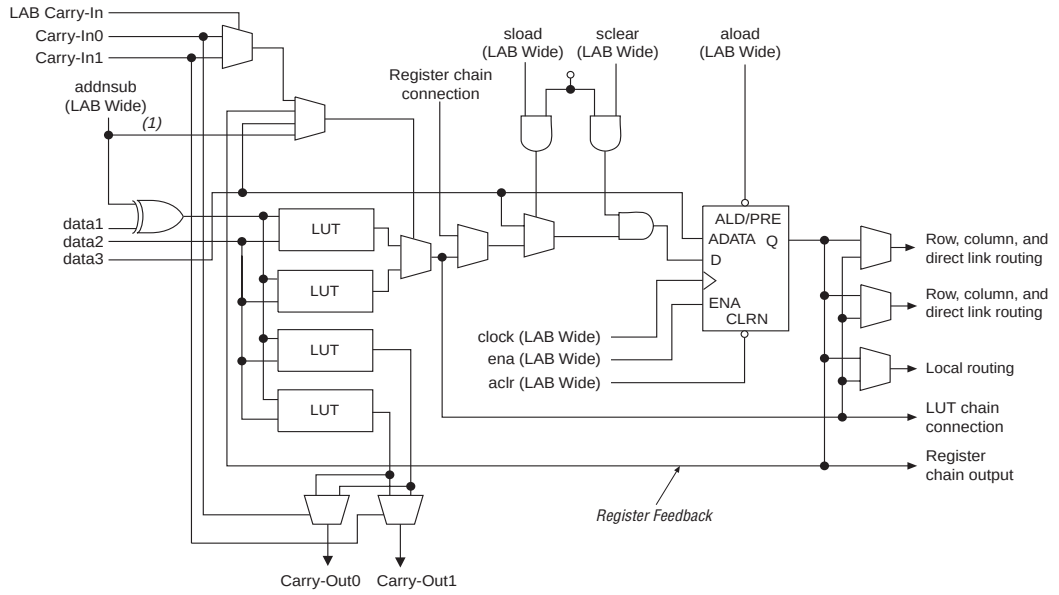
LE Operating Modes

The Stratix LE can operate in one of the following modes:

- Normal mode
- Dynamic arithmetic mode

Each mode uses LE resources differently. In each mode, eight available inputs to the LE—the four data inputs from the LAB local interconnect; `carry-in0` and `carry-in1` from the previous LE; the LAB carry-in from the previous carry-chain LAB; and the register chain connection—are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear,

Figure 2–7. LE in Dynamic Arithmetic Mode



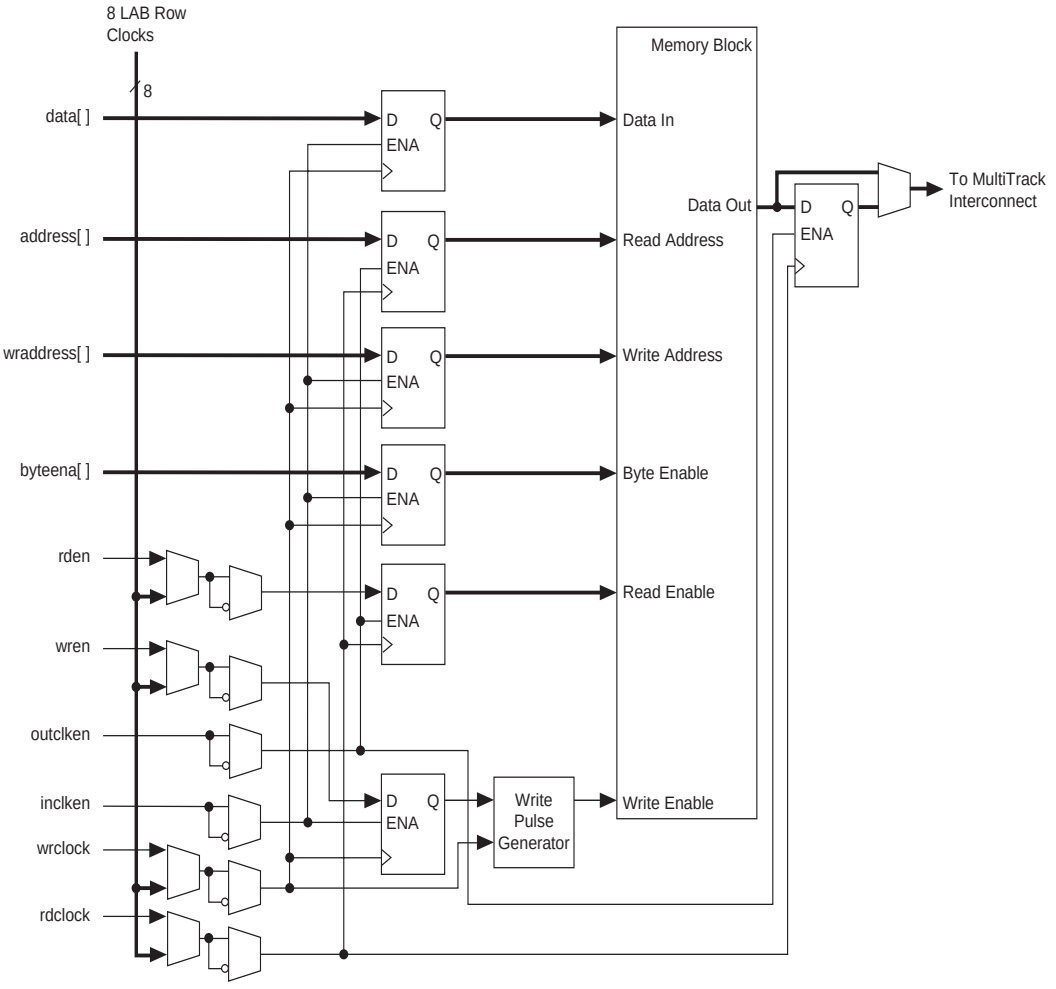
Note to Figure 2–7:
(1) The `addnsub` signal is tied to the carry input for the first LE of a carry chain only.

Carry-Select Chain

The carry-select chain provides a very fast carry-select function between LEs in arithmetic mode. The carry-select chain uses the redundant carry calculation to increase the speed of carry functions. The LE is configured to calculate outputs for a possible carry-in of 1 and carry-in of 0 in parallel. The `carry-in0` and `carry-in1` signals from a lower-order bit feed forward into the higher-order bit via the parallel carry chain and feed into both the LUT and the next portion of the carry chain. Carry-select chains can begin in any LE within an LAB.

The speed advantage of the carry-select chain is in the parallel pre-computation of carry chains. Since the LAB carry-in selects the precomputed carry chain, not every LE is in the critical path. Only the propagation delay between LAB carry-in generation (LE 5 and LE 10) are now part of the critical path. This feature allows the Stratix architecture to implement high-speed counters, adders, multipliers, parity functions, and comparators of arbitrary width.

Figure 2–27. Read/Write Clock Mode in Simple Dual-Port Mode Notes (1), (2)



Notes to Figure 2–27:

(1) All registers shown except the `rden` register have asynchronous clear ports.

(2) Violating the setup or hold time on the address registers could corrupt the memory contents. This applies to both read and write operations.

Clock Multiplication & Division

Each Stratix device enhanced PLL provides clock synthesis for PLL output ports using $m/(n \times$ post-scale counter) scaling factors. The input clock is divided by a pre-scale divider, n, and is then multiplied by the m feedback factor. The control loop drives the VCO to match $f_{IN} \times (m/n)$. Each output port has a unique post-scale counter that divides down the high-frequency VCO. For multiple PLL outputs with different frequencies, the VCO is set to the least common multiple of the output frequencies that meets its frequency specifications. Then, the post-scale dividers scale down the output frequency for each output port. For example, if output frequencies required from one PLL are 33 and 66 MHz, set the VCO to 330 MHz (the least common multiple in the VCO's range). There is one pre-scale counter, n, and one multiply counter, m, per PLL, with a range of 1 to 512 on each. There are two post-scale counters (l) for regional clock output ports, four counters (g) for global clock output ports, and up to four counters (e) for external clock outputs, all ranging from 1 to 1024 with a 50% duty cycle setting. The post-scale counters range from 1 to 512 with any non-50% duty cycle setting. The Quartus II software automatically chooses the appropriate scaling factors according to the input frequency, multiplication, and division values entered.

Clock Switchover

To effectively develop high-reliability network systems, clocking schemes must support multiple clocks to provide redundancy. For this reason, Stratix device enhanced PLLs support a flexible clock switchover capability. Figure 2–53 shows a block diagram of the switchover circuit.The switchover circuit is configurable, so you can define how to implement it. Clock-sense circuitry automatically switches from the primary to secondary clock for PLL reference when the primary clock signal is not present.

During switchover, the PLL VCO continues to run and will either slow down or speed up, generating frequency drift on the PLL outputs. The clock switchover transitions without any glitches. After the switch, there is a finite resynchronization period to lock onto new clock as the VCO ramps up. The exact amount of time it takes for the PLL to relock relates to the PLL configuration and may be adjusted by using the programmable bandwidth feature of the PLL. The specification for the maximum time to relock is 100 µs.

For more information on clock switchover, see *AN 313, Implementing Clock Switchover in Stratix & Stratix GX Devices*.

PLL Reconfiguration

The PLL reconfiguration feature enables system logic to change Stratix device enhanced PLL counters and delay elements without reloading a Programmer Object File (**.pof**). This provides considerable flexibility for frequency synthesis, allowing real-time PLL frequency and output clock delay variation. You can sweep the PLL output frequencies and clock delay in prototype environments. The PLL reconfiguration feature can also dynamically or intelligently control system clock speeds or t_{CO} delays in end systems.

Clock delay elements at each PLL output port implement variable delay. Figure 2–54 shows a diagram of the overall dynamic PLL control feature for the counters and the clock delay elements. The configuration time is less than 20 µs for the enhanced PLL using a input shift clock rate of 22 MHz. The charge pump, loop filter components, and phase shifting using VCO phase taps cannot be dynamically adjusted.

Stratix devices have an I/O interconnect similar to the R4 and C4 interconnect to drive high-fanout signals to and from the I/O blocks. There are 16 signals that drive into the I/O blocks composed of four output enables `io_boe[3..0]`, four clock enables `io_bce[3..0]`, four clocks `io_bclk[3..0]`, and four clear signals `io_bclr[3..0]`. The pin's `datain` signals can drive the IO interconnect, which in turn drives the logic array or other I/O blocks. In addition, the control and data signals can be driven from the logic array, providing a slower but more flexible routing resource. The row or column IOE clocks, `io_clk[7..0]`, provide a dedicated routing resource for low-skew, high-speed clocks. I/O clocks are generated from regional, global, or fast regional clocks (see "PLLs & Clock Networks" on page 2–73). Figure 2–62 illustrates the signal paths through the I/O block.

Figure 2–62. Signal Path through the I/O Block

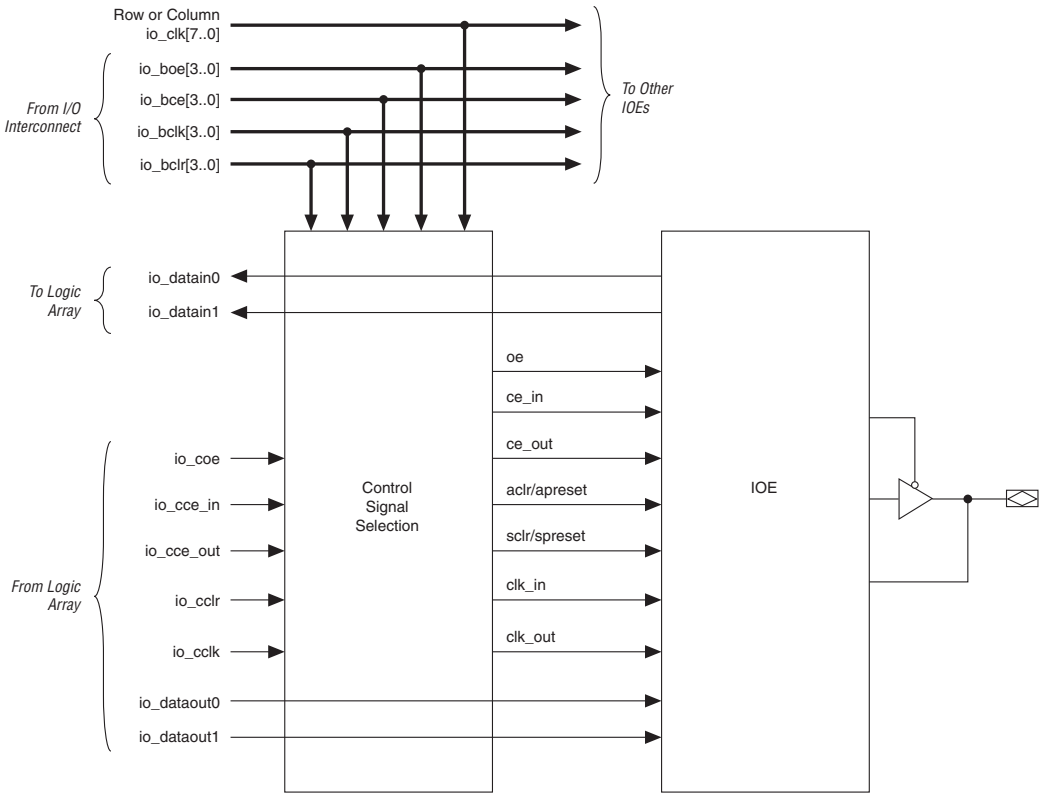


Table 2–26. External RAM Support in EP1S60 & EP1S80 Devices

DDR Memory Type	I/O Standard	Maximum Clock Rate (MHz)		
		-5 Speed Grade	-6 Speed Grade	-7 Speed Grade
DDR SDRAM *(1)*, *(2)*	SSTL-2	167	167	133
DDR SDRAM - side banks *(2)*, *(3)*	SSTL-2	150	133	133
QDR SRAM *(4)*	1.5-V HSTL	133	133	133
QDRII SRAM *(4)*	1.5-V HSTL	167	167	133
ZBT SRAM *(5)*	LVTTL	200	200	167

Notes to Table 2–26:

(1) These maximum clock rates apply if the Stratix device uses DQS phase-shift circuitry to interface with DDR SDRAM. DQS phase-shift circuitry is only available in the top and bottom I/O banks (I/O banks 3, 4, 7, and 8).

(2) For more information on DDR SDRAM, see *AN 342: Interfacing DDR SDRAM with Stratix & Stratix GX Devices*.

(3) DDR SDRAM is supported on the Stratix device side I/O banks (I/O banks 1, 2, 5, and 6) without dedicated DQS phase-shift circuitry. The read DQS signal is ignored in this mode. Numbers are preliminary.

(4) For more information on QDR or QDRII SRAM, see *AN 349: QDR SRAM Controller Reference Design for Stratix & Stratix GX Devices*.

(5) For more information on ZBT SRAM, see *AN 329: ZBT SRAM Controller Reference Design for Stratix & Stratix GX Devices*.

In addition to six I/O registers and one input latch in the IOE for interfacing to these high-speed memory interfaces, Stratix devices also have dedicated circuitry for interfacing with DDR SDRAM. In every Stratix device, the I/O banks at the top (I/O banks 3 and 4) and bottom (I/O banks 7 and 8) of the device support DDR SDRAM up to 200 MHz. These pins support DQS signals with DQ bus modes of ×8, ×16, or ×32.

Table 2–27 shows the number of DQ and DQS buses that are supported per device.

Table 2–27. DQS & DQ Bus Mode Support (Part 1 of 2) Note (1)

Device	Package	Number of ×8 Groups	Number of ×16 Groups	Number of ×32 Groups
EP1S10	672-pin BGA 672-pin FineLine BGA	12 *(2)*	0	0
	484-pin FineLine BGA 780-pin FineLine BGA	16 *(3)*	0	4
EP1S20	484-pin FineLine BGA	18*(4)*	7 *(5)*	4
	672-pin BGA 672-pin FineLine BGA	16*(3)*	7 *(5)*	4
	780-pin FineLine BGA	20	7 *(5)*	4

The Stratix device instruction register length is 10 bits and the USERCODE register length is 32 bits. Tables 3–2 and 3–3 show the boundary-scan register length and device IDCODE information for Stratix devices.

Table 3–2. Stratix Boundary-Scan Register Length

Device	Boundary-Scan Register Length
EP1S10	1,317
EP1S20	1,797
EP1S25	2,157
EP1S30	2,253
EP1S40	2,529
EP1S60	3,129
EP1S80	3,777

Table 3–3. 32-Bit Stratix Device IDCODE

Device	IDCODE (32 Bits) *(1)*			
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer Identity (11 Bits)	LSB (1 Bit) *(2)*
EP1S10	0000	0010 0000 0000 0001	000 0110 1110	1
EP1S20	0000	0010 0000 0000 0010	000 0110 1110	1
EP1S25	0000	0010 0000 0000 0011	000 0110 1110	1
EP1S30	0000	0010 0000 0000 0100	000 0110 1110	1
EP1S40	0000	0010 0000 0000 0101	000 0110 1110	1
EP1S60	0000	0010 0000 0000 0110	000 0110 1110	1
EP1S80	0000	0010 0000 0000 0111	000 0110 1110	1

Notes to Tables 3–2 and 3–3:
(1) The most significant bit (MSB) is on the left.
(2) The IDCODE's least significant bit (LSB) is always 1.

While in the factory configuration, the factory-configuration logic performs the following operations:

- Loads a remote update-control register to determine the page address of the new application configuration
- Determines whether to enable a user watchdog timer for the application configuration
- Determines what the watchdog timer setting should be if it is enabled

The user watchdog timer is a counter that must be continually reset within a specific amount of time in the user mode of an application configuration to ensure that valid configuration occurred during a remote update. Only valid application configurations designed for remote update can reset the user watchdog timer in user mode. If a valid application configuration does not reset the user watchdog timer in a specific amount of time, the timer updates a status register and loads the factory configuration. The user watchdog timer is automatically disabled for factory configurations.

If an error occurs in loading the application configuration, the configuration logic writes a status register to specify the cause of the error. Once this occurs, the Stratix device automatically loads the factory configuration, which reads the status register and determines the reason for reconfiguration. Based on the reason, the factory configuration will take appropriate steps and will write the remote update control register to specify the next application configuration page to be loaded.

When the Stratix device successfully loads the application configuration, it enters into user mode. The Stratix device then executes the main application of the user. Intellectual property (IP), such as a Nios® (16-bit ISA) and Nios® II (32-bit ISA) embedded processors, can help the Stratix device determine when remote update is coming. The Nios embedded processor or user logic receives incoming data, writes it to the configuration device, and loads the factory configuration. The factory configuration will read the remote update status register and determine the valid application configuration to load. Figure 3–2 shows the Stratix remote update. Figure 3–3 shows the transition diagram for remote update mode.

Table 4–22. SSTL-3 Class I Specifications (Part 2 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
$V_{IL(AC)}$	Low-level AC input voltage				V_{REF} – 0.4	V
V_{OH}	High-level output voltage	I_{OH} = –8 mA *(3)*	V_{TT} + 0.6			V
V_{OL}	Low-level output voltage	I_{OL} = 8 mA *(3)*			V_{TT} – 0.6	V

Table 4–23. SSTL-3 Class II Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		3.0	3.3	3.6	V
V_{TT}	Termination voltage		V_{REF} – 0.05	V_{REF}	V_{REF} + 0.05	V
V_{REF}	Reference voltage		1.3	1.5	1.7	V
$V_{IH(DC)}$	High-level DC input voltage		V_{REF} + 0.2		V_{CCIO} + 0.3	V
$V_{IL(DC)}$	Low-level DC input voltage		–0.3		V_{REF} – 0.2	V
$V_{IH(AC)}$	High-level AC input voltage		V_{REF} + 0.4			V
$V_{IL(AC)}$	Low-level AC input voltage				V_{REF} – 0.4	V
V_{OH}	High-level output voltage	I_{OH} = –16 mA *(3)*	V_{TT} + 0.8			V
V_{OL}	Low-level output voltage	I_{OL} = 16 mA *(3)*			V_{TT} – 0.8	V

Table 4–24. 3.3-V AGP 2× Specifications

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		3.15	3.3	3.45	V
V_{REF}	Reference voltage		0.39 × V_{CCIO}		0.41 × V_{CCIO}	V
V_{IH}	High-level input voltage *(4)*		0.5 × V_{CCIO}		V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage *(4)*				0.3 × V_{CCIO}	V
V_{OH}	High-level output voltage	I_{OUT} = –0.5 mA	0.9 × V_{CCIO}		3.6	V
V_{OL}	Low-level output voltage	I_{OUT} = 1.5 mA			0.1 × V_{CCIO}	V

Table 4–25. 3.3-V AGP 1× Specifications (Part 1 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage		3.15	3.3	3.45	V
V_{IH}	High-level input voltage *(4)*		0.5 × V_{CCIO}		V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage *(4)*				0.3 × V_{CCIO}	V

device. Decoupling capacitors were not used in this measurement. To factor in the current for decoupling capacitors, sum up the current for each capacitor using the following equation:

$I = C (dV/dt)$

If the regulator or power supply minimum output current is more than the Stratix device requires, then the device may consume more current than the maximum current listed in Table 4–34. However, the device does not require any more current to successfully power up than what is listed in Table 4–34.

Table 4–34. Stratix Power-Up Current (I_{CCINT}) Requirements Note (1)

Device	Power-Up Current Requirement		Unit
	Typical	Maximum	
EP1S10	250	700	mA
EP1S20	400	1,200	mA
EP1S25	500	1,500	mA
EP1S30	550	1,900	mA
EP1S40	650	2,300	mA
EP1S60	800	2,600	mA
EP1S80	1,000	3,000	mA

Note to Table 4–34:

(1) The maximum test conditions are for 0° C and typical test conditions are for 40° C.

The exact amount of current consumed varies according to the process, temperature, and power ramp rate. Stratix devices typically require less current during power up than shown in Table 4–34. The user-mode current during device operation is generally higher than the power-up current.

The duration of the I_{CCINT} power-up requirement depends on the V_{CCINT} voltage supply rise time. The power-up current consumption drops when the V_{CCINT} supply reaches approximately 0.75 V.

Internal Timing Parameters

Internal timing parameters are specified on a speed grade basis independent of device density. Tables 4–37 through 4–42 describe the Stratix device internal timing microparameters for LEs, IOEs, TriMatrix™ memory structures, DSP blocks, and MultiTrack interconnects.

Table 4–37. LE Internal Timing Microparameter Descriptions

Symbol	Parameter
t_{SU}	LE register setup time before clock
t_{H}	LE register hold time after clock
t_{CO}	LE register clock-to-output delay
t_{LUT}	LE combinatorial LUT delay for data-in to data-out
t_{CLR}	Minimum clear pulse width
t_{PRE}	Minimum preset pulse width
t_{CLKHL}	Register minimum clock high or low time. The maximum core clock frequency can be calculated by $1/(2 \times t_{CLKHL})$.

Table 4–38. IOE Internal Timing Microparameter Descriptions

Symbol	Parameter
t_{SU_R}	Row IOE input register setup time
t_{SU_C}	Column IOE input register setup time
t_{H}	IOE input and output register hold time after clock
t_{CO_R}	Row IOE input and output register clock-to-output delay
t_{CO_C}	Column IOE input and output register clock-to-output delay
$t_{PIN2COMBOUT\_R}$	Row input pin to IOE combinatorial output
$t_{PIN2COMBOUT\_C}$	Column input pin to IOE combinatorial output
$t_{COMBIN2PIN\_R}$	Row IOE data input to combinatorial output pin
$t_{COMBIN2PIN\_C}$	Column IOE data input to combinatorial output pin
t_{CLR}	Minimum clear pulse width
t_{PRE}	Minimum preset pulse width
t_{CLKHL}	Register minimum clock high or low time. The maximum I/O clock frequency can be calculated by $1/(2 \times t_{CLKHL})$. Performance may also be affected by I/O timing, use of PLL, and I/O programmable settings.

Table 4–41. M4K Block Internal Timing Microparameter Descriptions (Part 2 of 2)

Symbol	Parameter
$t_{M4KDATAAH}$	A port data hold time after clock
$t_{M4KADDRASU}$	A port address setup time before clock
$t_{M4KADDRAH}$	A port address hold time after clock
$t_{M4KDATABSU}$	B port data setup time before clock
$t_{M4KDATABH}$	B port data hold time after clock
$t_{M4KADDRBSU}$	B port address setup time before clock
$t_{M4KADDRBH}$	B port address hold time after clock
$t_{M4KDATACO1}$	Clock-to-output delay when using output registers
$t_{M4KDATACO2}$	Clock-to-output delay without output registers
$t_{M4KCLKHL}$	Register minimum clock high or low time. This is a limit on the min time for the clock on the registers in these blocks. The actual performance is dependent upon the internal point-to-point delays in the blocks and may give slower performance as shown inTable 4–36 on page 4–20 and as reported by the timing analyzer in the Quartus II software.
t_{M4KCLR}	Minimum clear pulse width

Table 4–42. M-RAM Block Internal Timing Microparameter Descriptions (Part 1 of 2)

Symbol	Parameter
t_{MRAMRC}	Synchronous read cycle time
t_{MRAMWC}	Synchronous write cycle time
$t_{MRAMWERESU}$	Write or read enable setup time before clock
$t_{MRAMWEREH}$	Write or read enable hold time after clock
$t_{MRAMCLKENSU}$	Clock enable setup time before clock
$t_{MRAMCLKENH}$	Clock enable hold time after clock
$t_{MRAMBESU}$	Byte enable setup time before clock
$t_{MRAMBEH}$	Byte enable hold time after clock
$t_{MRAMDATAASU}$	A port data setup time before clock
$t_{MRAMDATAAH}$	A port data hold time after clock
$t_{MRAMADDRASU}$	A port address setup time before clock
$t_{MRAMADDRAH}$	A port address hold time after clock
$t_{MRAMDATABSU}$	B port setup time before clock

Table 4–63. EP1S20 External I/O Timing on Column Pins Using Global Clock Networks *Note (1)*

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	1.351		1.479		1.699		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.732	5.380	2.732	5.728	2.732	6.240	NA	NA	ns
t_{XZ}	2.672	5.254	2.672	5.596	2.672	6.116	NA	NA	ns
t_{ZX}	2.672	5.254	2.672	5.596	2.672	6.116	NA	NA	ns
$t_{INSUPLL}$	0.923		0.971		1.098		NA		ns
t_{INHPLL}	0.000		0.000		0.000		NA		ns
$t_{OUTCOPLL}$	1.210	2.544	1.210	2.648	1.210	2.715	NA	NA	ns
t_{XZPLL}	1.150	2.418	1.150	2.516	1.150	2.591	NA	NA	ns
t_{ZXPLL}	1.150	2.418	1.150	2.516	1.150	2.591	NA	NA	ns

Table 4–64. EP1S20 External I/O Timing on Row Pins Using Fast Regional Clock Networks *Note (1)*

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.032		2.207		2.535		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.492	5.018	2.492	5.355	2.492	5.793	NA	NA	ns
t_{XZ}	2.519	5.072	2.519	5.411	2.519	5.861	NA	NA	ns
t_{ZX}	2.519	5.072	2.519	5.411	2.519	5.861	NA	NA	ns

Table 4–102. Reporting Methodology For Minimum Timing For Single-Ended Output Pins (Part 2 of 2)
Notes (1), (2), (3)

I/O Standard	Loading and Termination							Measurement Point
	R_{UP} Ω	R_{DN} Ω	R_S Ω	R_T Ω	V_{CCIO} (V)	VTT (V)	C_L (pF)	V_{MEAS}
3.3-V CTT	–	–	25	50	3.600	1.650	30	1.650

Notes to Table 4–102:
(1) Input measurement point at internal node is 0.5 × V_{CCINT}.
(2) Output measuring point for data is V_{MEAS}. When two values are given, the first is the measurement point on the rising edge and the other is for the falling edge.
(3) Input stimulus edge rate is 0 to V_{CCINT} in 0.5 ns (internal signal) from the driver preceding the I/O buffer.
(4) The first value is for the output rising edge and the second value is for the output falling edge. The hyphen (-) indicates infinite resistance or disconnection.

Figure 4–8 shows the measurement setup for output disable and output enable timing. The T_{CHZ} stands for clock to high Z time delay and is the same as T_{XZ}. The T_{CLZ} stands for clock to low Z (driving) time delay and is the same as T_{ZX}.

Figure 4–8. Measurement Setup for T_{XZ} and T_{ZX}

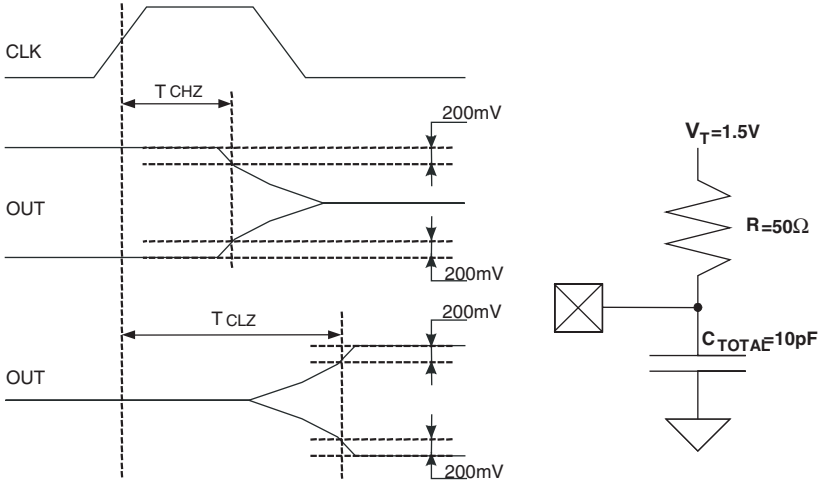


Table 4–118. Stratix Maximum Input Clock Rate for CLK[0, 2, 9, 11] Pins & FPLL[10..7]CLK Pins in Wire-Bond Packages (Part 2 of 2)

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVCMOS	422	390	390	MHz
GTL+	250	200	200	MHz
SSTL-3 Class I	350	300	300	MHz
SSTL-3 Class II	350	300	300	MHz
SSTL-2 Class I	350	300	300	MHz
SSTL-2 Class II	350	300	300	MHz
SSTL-18 Class I	350	300	300	MHz
SSTL-18 Class II	350	300	300	MHz
1.5-V HSTL Class I	350	300	300	MHz
1.8-V HSTL Class I	350	300	300	MHz
CTT	250	200	200	MHz
Differential 1.5-V HSTL C1	350	300	300	MHz
LVPECL *(1)*	717	640	640	MHz
PCML *(1)*	375	350	350	MHz
LVDS *(1)*	717	640	640	MHz
HyperTransport technology *(1)*	717	640	640	MHz

Table 4–119. Stratix Maximum Input Clock Rate for CLK[1, 3, 8, 10] Pins in Wire-Bond Packages (Part 1 of 2)

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	422	390	390	MHz
2.5 V	422	390	390	MHz
1.8 V	422	390	390	MHz
1.5 V	422	390	390	MHz
LVCMOS	422	390	390	MHz
GTL+	250	200	200	MHz
SSTL-3 Class I	350	300	300	MHz
SSTL-3 Class II	350	300	300	MHz
SSTL-2 Class I	350	300	300	MHz
SSTL-2 Class II	350	300	300	MHz

Table 4–125. High-Speed I/O Specifications for Flip-Chip Packages (Part 3 of 4) *Notes (1), (2)*

Symbol	Conditions	-5 Speed Grade			-6 Speed Grade			-7 Speed Grade			-8 Speed Grade			Unit
		Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	
SW	PCML (*J* = 4, 7, 8, 10)	750			750			800			800			ps
	PCML (*J* = 2)	900			900			1,200			1,200			ps
	PCML (*J* = 1)	1,500			1,500			1,700			1,700			ps
	LVDS and LVPECL (*J* = 1)	500			500			550			550			ps
	LVDS, LVPECL, HyperTransport technology (*J* = 2 through 10)	440			440			500			500			ps
Input jitter tolerance (peak-to-peak)	All			250			250			250			250	ps
Output jitter (peak-to-peak)	All			160			160			200			200	ps
Output t_{RISE}	LVDS	80	110	120	80	110	120	80	110	120	80	110	120	ps
	HyperTransport technology	110	170	200	110	170	200	120	170	200	120	170	200	ps
	LVPECL	90	130	150	90	130	150	100	135	150	100	135	150	ps
	PCML	80	110	135	80	110	135	80	110	135	80	110	135	ps
Output t_{FALL}	LVDS	80	110	120	80	110	120	80	110	120	80	110	120	ps
	HyperTransport technology	110	170	200	110	170	200	110	170	200	110	170	200	ps
	LVPECL	90	130	160	90	130	160	100	135	160	100	135	160	ps
	PCML	105	140	175	105	140	175	110	145	175	110	145	175	ps