Intel - EP1S80F1508C6 Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	7904
Number of Logic Elements/Cells	79040
Total RAM Bits	7427520
Number of I/O	1203
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	1508-BBGA, FCBGA
Supplier Device Package	1508-FBGA, FC (40x40)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1s80f1508c6

asynchronous load, and clear signals. An asynchronous clear signal takes precedence if both signals are asserted simultaneously. Each LAB supports up to two clears and one preset signal.

In addition to the clear and preset ports, Stratix devices provide a chip-wide reset pin (`DEV_CLRn`) that resets all registers in the device. An option set before compilation in the Quartus II software controls this pin. This chip-wide reset overrides all other control signals.

MultiTrack Interconnect

In the Stratix architecture, connections between LEs, TriMatrix memory, DSP blocks, and device I/O pins are provided by the MultiTrack interconnect structure with DirectDrive™ technology. The MultiTrack interconnect consists of continuous, performance-optimized routing lines of different lengths and speeds used for inter- and intra-design block connectivity. The Quartus II Compiler automatically places critical design paths on faster interconnects to improve design performance.

DirectDrive technology is a deterministic routing technology that ensures identical routing resource usage for any function regardless of placement within the device. The MultiTrack interconnect and DirectDrive technology simplify the integration stage of block-based designing by eliminating the re-optimization cycles that typically follow design changes and additions.

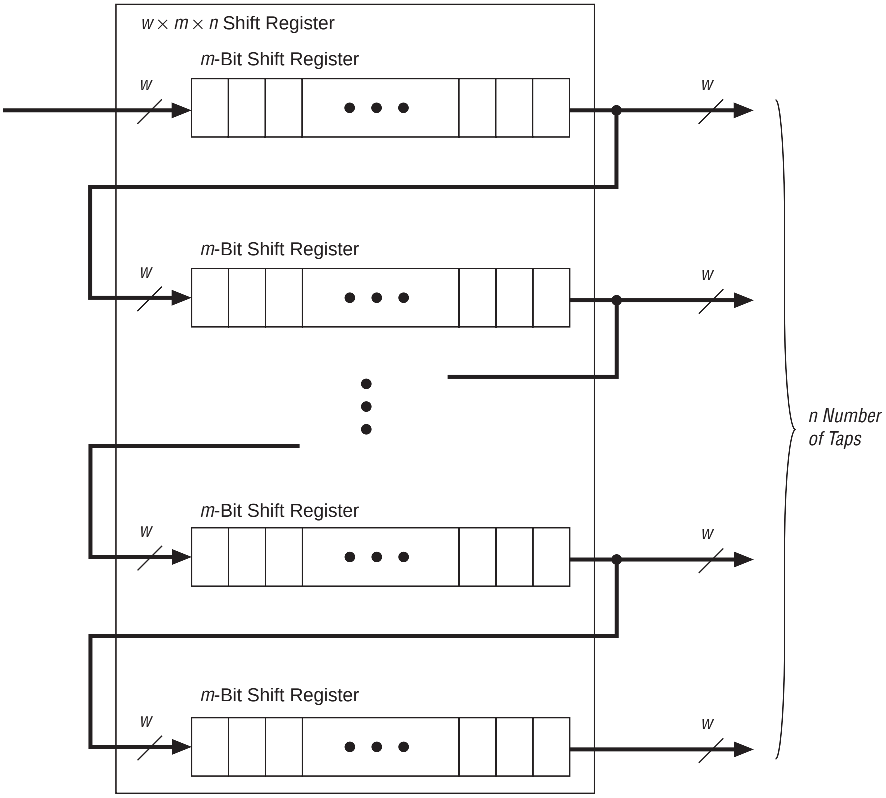
The MultiTrack interconnect consists of row and column interconnects that span fixed distances. A routing structure with fixed length resources for all devices allows predictable and repeatable performance when migrating through different device densities. Dedicated row interconnects route signals to and from LABs, DSP blocks, and TriMatrix memory within the same row. These row resources include:

- Direct link interconnects between LABs and adjacent blocks.
- R4 interconnects traversing four blocks to the right or left.
- R8 interconnects traversing eight blocks to the right or left.
- R24 row interconnects for high-speed access across the length of the device.

The direct link interconnect allows an LAB, DSP block, or TriMatrix memory block to drive into the local interconnect of its left and right neighbors and then back into itself. Only one side of a M-RAM block interfaces with direct link and row interconnects. This provides fast communication between adjacent LABs and/or blocks without using row interconnect resources.

The R4 interconnects span four LABs, three LABs and one M512 RAM block, two LABs and one M4K RAM block, or two LABs and one DSP block to the right or left of a source LAB. These resources are used for fast

Figure 2–14. Shift Register Memory Configuration



Memory Block Size

TriMatrix memory provides three different memory sizes for efficient application support. The large number of M512 blocks are ideal for designs with many shallow first-in first-out (FIFO) buffers. M4K blocks provide additional resources for channelized functions that do not require large amounts of storage. The M-RAM blocks provide a large single block of RAM ideal for data packet storage. The different-sized blocks allow Stratix devices to efficiently support variable-sized memory in designs.

The Quartus II software automatically partitions the user-defined memory into the embedded memory blocks using the most efficient size combinations. You can also manually assign the memory to a specific block size or a mixture of block sizes.

Figure 2–29. DSP Blocks Arranged in Columns

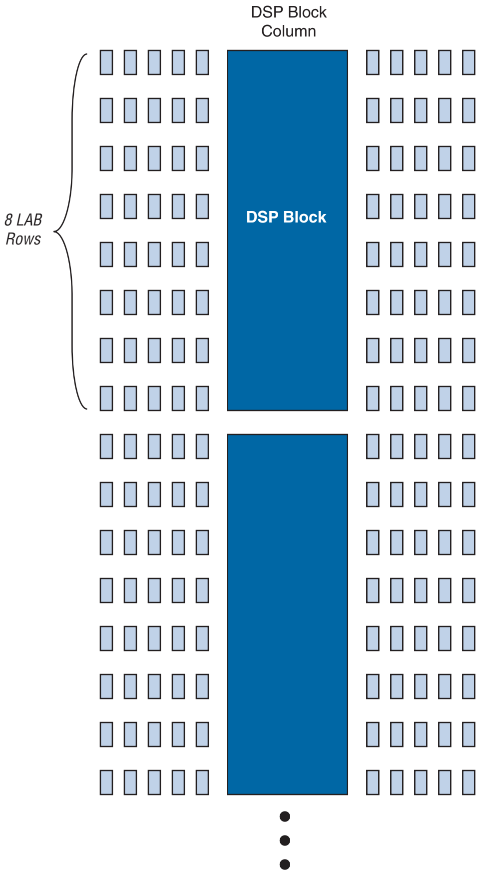


Table 2–13 shows the number of DSP blocks in each Stratix device.

Table 2–13. DSP Blocks in Stratix Devices Notes (1), (2)

Device	DSP Blocks	Total 9 × 9 Multipliers	Total 18 × 18 Multipliers	Total 36 × 36 Multipliers
EP1S10	6	48	24	6
EP1S20	10	80	40	10
EP1S25	10	80	40	10
EP1S30	12	96	48	12
EP1S40	14	112	56	14
EP1S60	18	144	72	18
EP1S80	22	176	88	22

Notes to Table 2–13:

(1) Each device has either the number of 9 × 9-, 18 × 18-, or 36 × 36-bit multipliers shown. The total number of multipliers for each device is not the sum of all the multipliers.

(2) The number of supported multiply functions shown is based on signed/signed or unsigned/unsigned implementations.

DSP block multipliers can optionally feed an adder/subtractor or accumulator within the block depending on the configuration. This makes routing to LEs easier, saves LE routing resources, and increases performance, because all connections and blocks are within the DSP block. Additionally, the DSP block input registers can efficiently implement shift registers for FIR filter applications.

Figure 2–30 shows the top-level diagram of the DSP block configured for 18 × 18-bit multiplier mode. Figure 2–31 shows the 9 × 9-bit multiplier configuration of the DSP block.

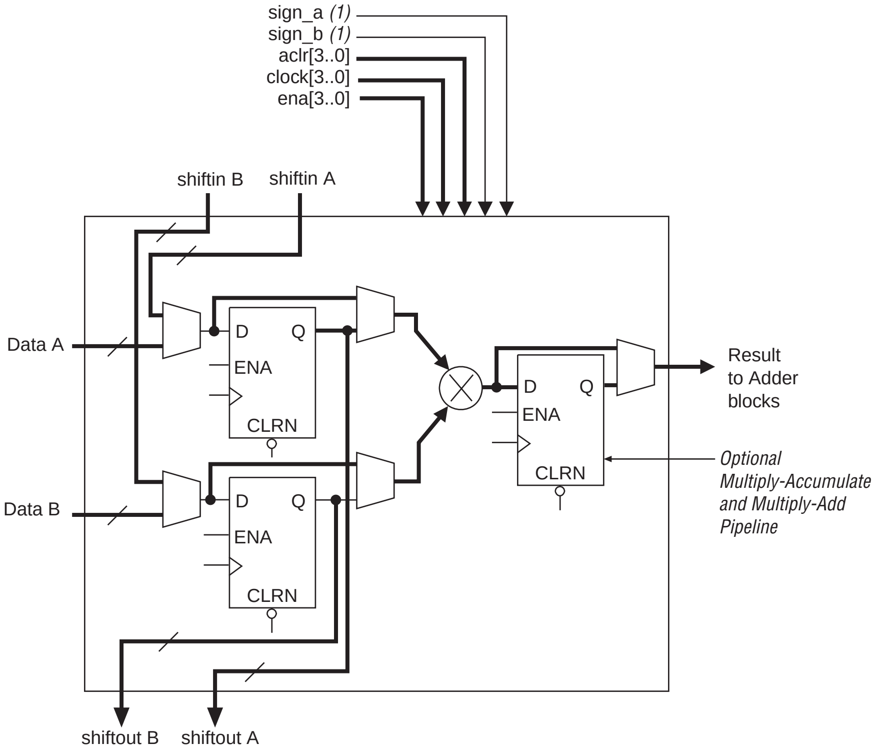
The DSP block consists of the following elements:

- Multiplier block
- Adder/output block

Multiplier Block

The DSP block multiplier block consists of the input registers, a multiplier, and pipeline register for pipelining multiply-accumulate and multiply-add/subtract functions as shown in Figure 2–32.

Figure 2–32. Multiplier Sub-Block within Stratix DSP Block



Note to Figure 2–32:

(1) These signals can be unregistered or registered once to match data path pipelines if required.

Table 2–14 shows the summary of input register modes for the DSP block.

Table 2–14. Input Register Modes

Register Input Mode	9 × 9	18 × 18	36 × 36
Parallel input	✓	✓	✓
Shift register input	✓	✓	

Multiplier

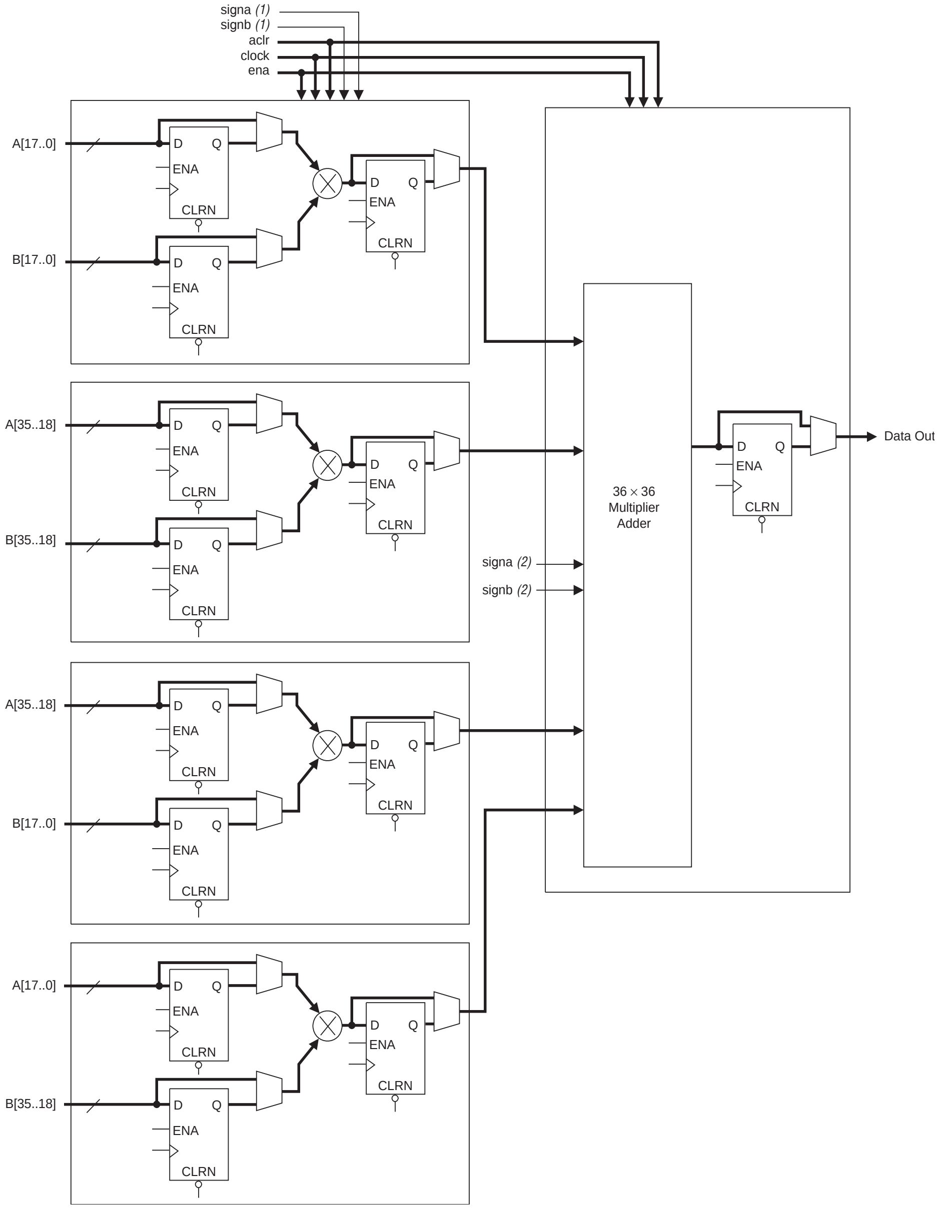
The multiplier supports 9 × 9-, 18 × 18-, or 36 × 36-bit multiplication. Each DSP block supports eight possible 9 × 9-bit or smaller multipliers. There are four multiplier blocks available for multipliers larger than 9 × 9 bits but smaller than 18 × 18 bits. There is one multiplier block available for multipliers larger than 18 × 18 bits but smaller than or equal to 36 × 36 bits. The ability to have several small multipliers is useful in applications such as video processing. Large multipliers greater than 18 × 18 bits are useful for applications such as the mantissa multiplication of a single-precision floating-point number.

The multiplier operands can be signed or unsigned numbers, where the result is signed if either input is signed as shown in Table 2–15. The `sign_a` and `sign_b` signals provide dynamic control of each operand's representation: a logic 1 indicates the operand is a signed number, a logic 0 indicates the operand is an unsigned number. These sign signals affect all multipliers and adders within a single DSP block and you can register them to match the data path pipeline. The multipliers are full precision (that is, 18 bits for the 18-bit multiply, 36-bits for the 36-bit multiply, and so on) regardless of whether `sign_a` or `sign_b` set the operands as signed or unsigned numbers.

Table 2–15. Multiplier Signed Representation

Data A	Data B	Result
Unsigned	Unsigned	Unsigned
Unsigned	Signed	Signed
Signed	Unsigned	Signed
Signed	Signed	Signed

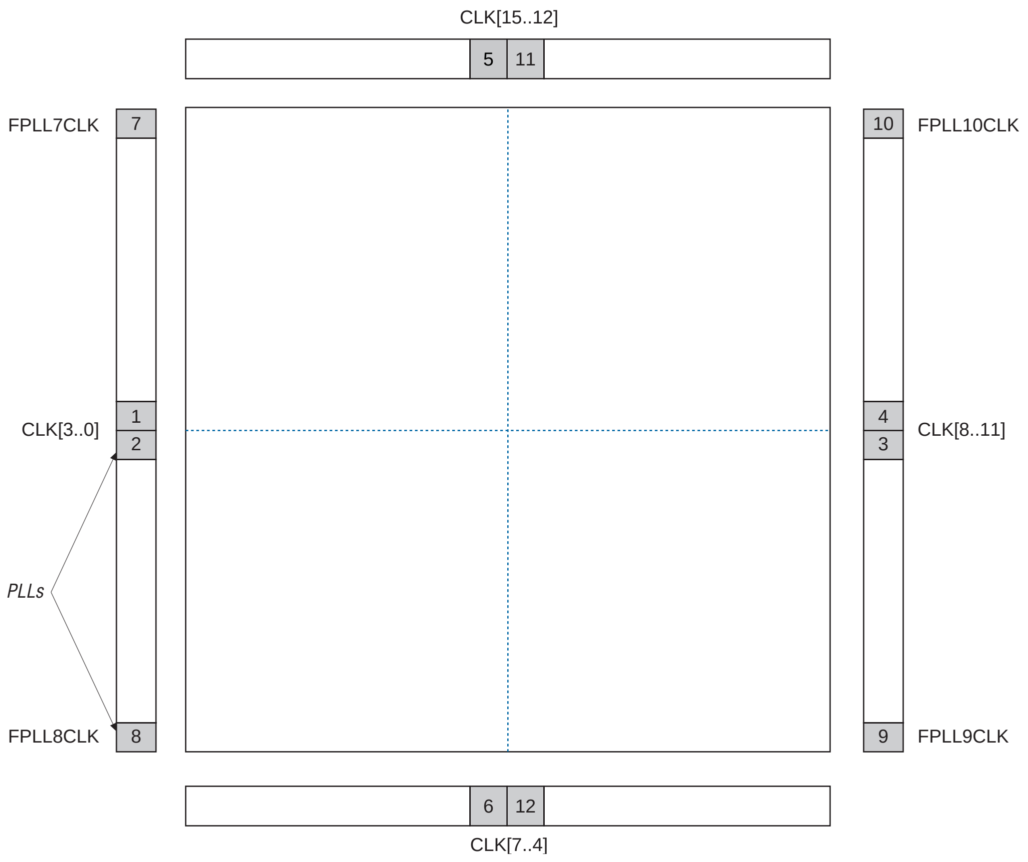
Figure 2–36. 36 × 36 Multiply Mode



Notes to Figure 2–36:
(1) These signals are not registered or registered once to match the pipeline.
(2) These signals are not registered, registered once, or registered twice for latency to match the pipeline.

Figure 2–49 shows a top-level diagram of the Stratix device and PLL floorplan.

Figure 2–49. PLL Locations



bandwidth is tuned by varying the charge pump current, loop filter resistor value, high frequency capacitor value, and *m* counter value. You can manually adjust these values if desired. Bandwidth is programmable from 200 kHz to 1.5 MHz.

External Clock Outputs

Enhanced PLLs 5 and 6 each support up to eight single-ended clock outputs (or four differential pairs). Differential SSTL and HSTL outputs are implemented using 2 single-ended output buffers which are programmed to have opposite polarity. In Quartus II software, simply assign the appropriate differential I/O standard and the software will implement the inversion. See Figure 2–55.

The only way you can use the `rx_data_align` is if one of the following is true:

- The receiver PLL is only clocking receive channels (no resources for the transmitter)
- If all channels can fit in one I/O bank

Table 2–38. EP1S30 Differential Channels *Note (1)*

Package	Transmitter /Receiver	Total Channels	Maximum Speed (Mbps)	Center Fast PLLs				Corner Fast PLLs *(2), (3)*			
				PLL1	PLL2	PLL3	PLL4	PLL7	PLL8	PLL9	PLL10
780-pin FineLine BGA	Transmitter *(4)*	70	840	18	17	17	18	*(6)*	*(6)*	*(6)*	*(6)*
			840 *(5)*	35	35	35	35	*(6)*	*(6)*	*(6)*	*(6)*
	Receiver	66	840	17	16	16	17	*(6)*	*(6)*	*(6)*	*(6)*
			840 *(5)*	33	33	33	33	*(6)*	*(6)*	*(6)*	*(6)*
956-pin BGA	Transmitter *(4)*	80	840	19	20	20	19	20	20	20	20
			840 *(5)*	39	39	39	39	20	20	20	20
	Receiver	80	840	20	20	20	20	19	20	20	19
			840 *(5)*	40	40	40	40	19	20	20	19
1,020-pin FineLine BGA	Transmitter *(4)*	80 (2) *(7)*	840	19 (1)	20	20	19 (1)	20	20	20	20
			840 *(5),(8)*	39 (1)	39 (1)	39 (1)	39 (1)	20	20	20	20
	Receiver	80 (2) *(7)*	840	20	20	20	20	19 (1)	20	20	19 (1)
			840 *(5),(8)*	40	40	40	40	19 (1)	20	20	19 (1)

Table 2–39. EP1S40 Differential Channels (Part 1 of 2) *Note (1)*

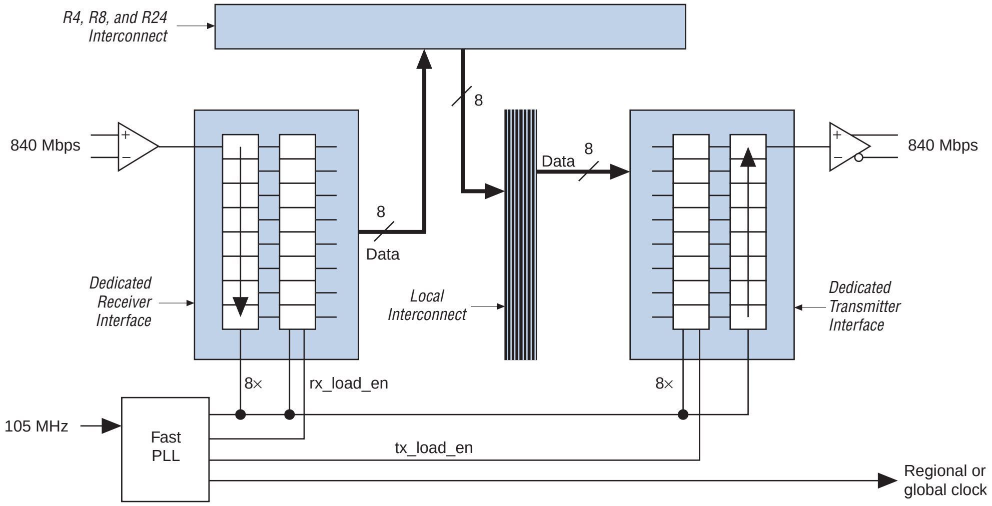
Package	Transmitter/ Receiver	Total Channels	Maximum Speed (Mbps)	Center Fast PLLs				Corner Fast PLLs *(2), (3)*			
				PLL1	PLL2	PLL3	PLL4	PLL7	PLL8	PLL9	PLL10
780-pin FineLine BGA	Transmitter *(4)*	68	840	18	16	16	18	*(6)*	*(6)*	*(6)*	*(6)*
			840 *(5)*	34	34	34	34	*(6)*	*(6)*	*(6)*	*(6)*
	Receiver	66	840	17	16	16	17	*(6)*	*(6)*	*(6)*	*(6)*
			840 *(5)*	33	33	33	33	*(6)*	*(6)*	*(6)*	*(6)*

- RapidIO
- HyperTransport

Dedicated Circuitry

Stratix devices support source-synchronous interfacing with LVDS, LVPECL, 3.3-V PCML, or HyperTransport signaling at up to 840 Mbps. Stratix devices can transmit or receive serial channels along with a low-speed or high-speed clock. The receiving device PLL multiplies the clock by a integer factor W (W = 1 through 32). For example, a HyperTransport application where the data rate is 800 Mbps and the clock rate is 400 MHz would require that W be set to 2. The SERDES factor J determines the parallel data width to deserialize from receivers or to serialize for transmitters. The SERDES factor J can be set to 4, 7, 8, or 10 and does not have to equal the PLL clock-multiplication W value. For a J factor of 1, the Stratix device bypasses the SERDES block. For a J factor of 2, the Stratix device bypasses the SERDES block, and the DDR input and output registers are used in the IOE. See Figure 2–73.

Figure 2–73. High-Speed Differential I/O Receiver / Transmitter Interface Example

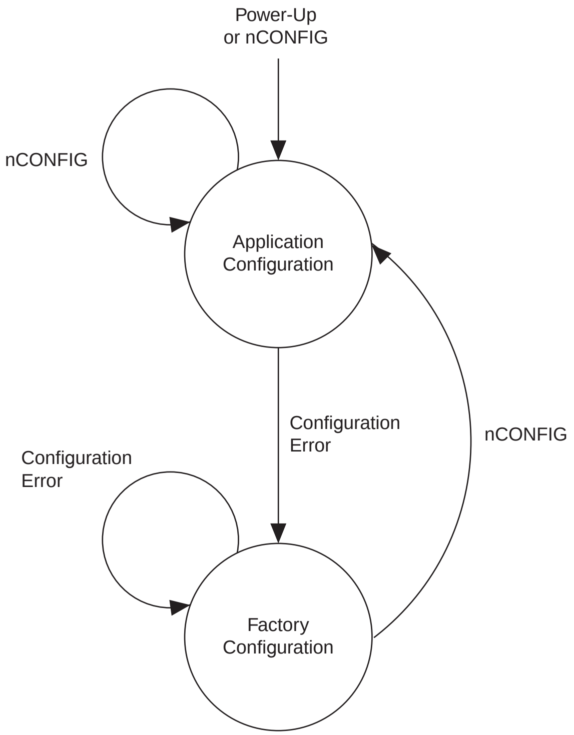


An external pin or global or regional clock can drive the fast PLLs, which can output up to three clocks: two multiplied high-speed differential I/O clocks to drive the SERDES block and/or external pin, and a low-speed clock to drive the logic array.

Local Update Mode

Local update mode is a simplified version of the remote update. This feature is intended for simple systems that need to load a single application configuration immediately upon power up without loading the factory configuration first. Local update designs have only one application configuration to load, so it does not require a factory configuration to determine which application configuration to use. Figure 3–4 shows the transition diagram for local update mode.

Figure 3–4. Local Update Transition Diagram



Stratix Automated Single Event Upset (SEU) Detection

Stratix devices offer on-chip circuitry for automated checking of single event upset (SEU) detection. FPGA devices that operate at high elevations or in close proximity to earth's North or South Pole require periodic checks to ensure continued data integrity. The error detection cyclic redundancy check (CRC) feature controlled by the **Device & Pin Options** dialog box in the Quartus II software uses a 32-bit CRC circuit to ensure data reliability and is one of the best options for mitigating SEU.

Table 4–7. 1.8-V I/O Specifications

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCIO}	Output supply voltage		1.65	1.95	V
V_{IH}	High-level input voltage		$0.65 \times V_{CCIO}$	2.25	V
V_{IL}	Low-level input voltage		–0.3	$0.35 \times V_{CCIO}$	V
V_{OH}	High-level output voltage	I_{OH} = –2 to –8 mA *(10)*	$V_{CCIO} - 0.45$		V
V_{OL}	Low-level output voltage	I_{OL} = 2 to 8 mA *(10)*		0.45	V

Table 4–8. 1.5-V I/O Specifications

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCIO}	Output supply voltage		1.4	1.6	V
V_{IH}	High-level input voltage		$0.65 \times V_{CCIO}$	$V_{CCIO} + 0.3$	V
V_{IL}	Low-level input voltage		–0.3	$0.35 \times V_{CCIO}$	V
V_{OH}	High-level output voltage	I_{OH} = –2 mA *(10)*	$0.75 \times V_{CCIO}$		V
V_{OL}	Low-level output voltage	I_{OL} = 2 mA *(10)*		$0.25 \times V_{CCIO}$	V

Notes to Tables 4–1 through 4–8:

(1) See the *Operating Requirements for Altera Devices Data Sheet*.

(2) Conditions beyond those listed in Table 4–1 may cause permanent damage to a device. Additionally, device operation at the absolute maximum ratings for extended periods of time may have adverse affects on the device.

(3) Minimum DC input is –0.5 V. During transitions, the inputs may undershoot to –2.0 V for input currents less than 100 mA and periods shorter than 20 ns, or overshoot to the voltage shown in Table 4–9, based on input duty cycle for input currents less than 100 mA. The overshoot is dependent upon duty cycle of the signal. The DC case is equivalent to 100% duty cycle.

(4) Maximum V_{CC} rise time is 100 ms, and V_{CC} must rise monotonically.

(5) V_{CCIO} maximum and minimum conditions for LVPECL, LVDS, and 3.3-V PCML are shown in parentheses.

(6) All pins, including dedicated inputs, clock, I/O, and JTAG pins, may be driven before V_{CCINT} and V_{CCIO} are powered.

(7) Typical values are for T_A = 25°C, V_{CCINT} = 1.5 V, and V_{CCIO} = 1.5 V, 1.8 V, 2.5 V, and 3.3 V.

(8) This value is specified for normal device operation. The value may vary during power-up. This applies for all V_{CCIO} settings (3.3, 2.5, 1.8, and 1.5 V).

(9) Pin pull-up resistance values will lower if an external source drives the pin higher than V_{CCIO}.

(10) Drive strength is programmable according to the values shown in the *Stratix Architecture* chapter of the *Stratix Device Handbook, Volume 1*.

Table 4–9. Overshoot Input Voltage with Respect to Duty Cycle (Part 1 of 2)

Vin (V)	Maximum Duty Cycle (%)
4.0	100
4.1	90
4.2	50

device. Decoupling capacitors were not used in this measurement. To factor in the current for decoupling capacitors, sum up the current for each capacitor using the following equation:

$I = C (dV/dt)$

If the regulator or power supply minimum output current is more than the Stratix device requires, then the device may consume more current than the maximum current listed in Table 4–34. However, the device does not require any more current to successfully power up than what is listed in Table 4–34.

Table 4–34. Stratix Power-Up Current (I_{CCINT}) Requirements Note (1)

Device	Power-Up Current Requirement		Unit
	Typical	Maximum	
EP1S10	250	700	mA
EP1S20	400	1,200	mA
EP1S25	500	1,500	mA
EP1S30	550	1,900	mA
EP1S40	650	2,300	mA
EP1S60	800	2,600	mA
EP1S80	1,000	3,000	mA

Note to Table 4–34:

(1) The maximum test conditions are for 0° C and typical test conditions are for 40° C.

The exact amount of current consumed varies according to the process, temperature, and power ramp rate. Stratix devices typically require less current during power up than shown in Table 4–34. The user-mode current during device operation is generally higher than the power-up current.

The duration of the I_{CCINT} power-up requirement depends on the V_{CCINT} voltage supply rise time. The power-up current consumption drops when the V_{CCINT} supply reaches approximately 0.75 V.

Table 4–59. EP1S10 External I/O Timing on Row Pins Using Regional Clock Networks *Note (1)*

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.161		2.336		2.685		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.434	4.889	2.434	5.226	2.434	5.643	NA	NA	ns
t_{XZ}	2.461	4.493	2.461	5.282	2.461	5.711	NA	NA	ns
t_{ZX}	2.461	4.493	2.461	5.282	2.461	5.711	NA	NA	ns
$t_{INSUPLL}$	1.057		1.172		1.315		NA		ns
t_{INHPLL}	0.000		0.000		0.000		NA		ns
$t_{OUTCOPLL}$	1.327	2.773	1.327	2.848	1.327	2.940	NA	NA	ns
t_{XZPLL}	1.354	2.827	1.354	2.904	1.354	3.008	NA	NA	ns
t_{ZXPLL}	1.354	2.827	1.354	2.904	1.354	3.008	NA	NA	ns

Table 4–60. EP1S10 External I/O Timing on Row Pins Using Global Clock Networks *Note (1)*

Parameter	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{INSU}	1.787		1.944		2.232		NA		ns
t_{INH}	0.000		0.000		0.000		NA		ns
t_{OUTCO}	2.647	5.263	2.647	5.618	2.647	6.069	NA	NA	ns
t_{XZ}	2.674	5.317	2.674	5.674	2.674	6.164	NA	NA	ns
t_{ZX}	2.674	5.317	2.674	5.674	2.674	6.164	NA	NA	ns
$t_{INSUPLL}$	1.371		1.1472		1.654		NA		ns
t_{INHPLL}	0.000		0.000		0.000		NA		ns
$t_{OUTCOPLL}$	1.144	2.459	1.144	2.548	1.144	2.601	NA	NA	ns
t_{XZPLL}	1.171	2.513	1.171	2.604	1.171	2.669	NA	NA	ns
t^{ZXPLL}	1.171	2.513	1.171	2.604	1.171	2.669	NA	NA	ns

Note to Tables 4–55 to 4–60:

(1) Only EP1S25, EP1S30, and EP1S40 have speed grade of -8.

Skew on Input Pins

Table 4–99 shows the package skews that were considered to get the worst case I/O skew value. You can use these values, for example, when calculating the timing budget on the input (read) side of a memory interface.

Table 4–99. Package Skew on Input Pins

Package Parameter	Worst-Case Skew (ps)
Pins in the same I/O bank	50
Pins in top/bottom (vertical I/O) banks	50
Pins in left/right side (horizontal I/O) banks	50
Pins across the entire device	100

PLL Counter & Clock Network Skews

Table 4–100 shows the clock skews between different clock outputs from the Stratix device PLL.

Table 4–100. PLL Counter & Clock Network Skews

Parameter	Worst-Case Skew (ps)
Clock skew between two external clock outputs driven by the same counter	100
Clock skew between two external clock outputs driven by the different counters with the same settings	150
Dual-purpose PLL dedicated clock output used as I/O pin vs. regular I/O pin	270 *(1)*
Clock skew between any two outputs of the PLL that drive global clock networks	150

Note to Table 4–100:

(1) The Quartus II software models 270 ps of delay on the PLL dedicated clock output (`PLL6_OUT[3..0]p/n` and `PLL5_OUT[3..0]p/n`) pins both when used as clocks and when used as I/O pins.

I/O Timing Measurement Methodology

Different I/O standards require different baseline loading techniques for reporting timing delays. Altera characterizes timing delays with the required termination and loading for each I/O standard. The timing information is specified from the input clock pin up to the output pin of

Table 4–102. Reporting Methodology For Minimum Timing For Single-Ended Output Pins (Part 2 of 2) *Notes (1), (2), (3)*

I/O Standard	Loading and Termination							Measurement Point
	R_{UP} Ω	R_{DN} Ω	R_S Ω	R_T Ω	V_{CCIO} (V)	VTT (V)	C_L (pF)	V_{MEAS}
3.3-V CTT	–	–	25	50	3.600	1.650	30	1.650

Notes to Table 4–102:

(1) Input measurement point at internal node is 0.5 × V_{CCINT}.

(2) Output measuring point for data is V_{MEAS}. When two values are given, the first is the measurement point on the rising edge and the other is for the falling edge.

(3) Input stimulus edge rate is 0 to V_{CCINT} in 0.5 ns (internal signal) from the driver preceding the I/O buffer.

(4) The first value is for the output rising edge and the second value is for the output falling edge. The hyphen (-) indicates infinite resistance or disconnection.

Figure 4–8 shows the measurement setup for output disable and output enable timing. The T_{CHZ} stands for clock to high Z time delay and is the same as T_{XZ}. The T_{CLZ} stands for clock to low Z (driving) time delay and is the same as T_{ZX}.

Figure 4–8. Measurement Setup for T_{XZ} and T_{ZX}

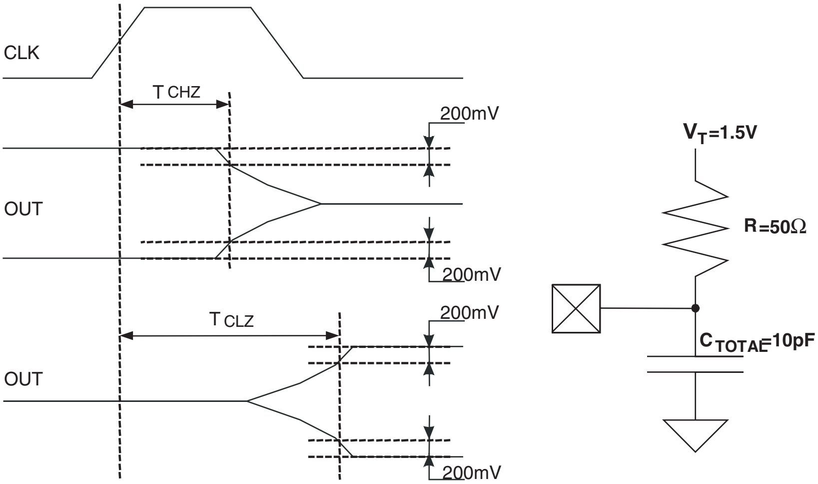


Table 4–114. Stratix Maximum Input Clock Rate for CLK[7..4] & CLK[15..12] Pins in Flip-Chip Packages (Part 2 of 2)

I/O Standard	-5 Speed Grade	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVDS *(1)*	645	645	622	622	MHz
HyperTransport technology *(1)*	500	500	450	450	MHz

Table 4–115. Stratix Maximum Input Clock Rate for CLK[0, 2, 9, 11] Pins & FPLL[10..7]CLK Pins in Flip-Chip Packages

I/O Standard	-5 Speed Grade	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	422	422	390	390	MHz
2.5 V	422	422	390	390	MHz
1.8 V	422	422	390	390	MHz
1.5 V	422	422	390	390	MHz
LVCMOS	422	422	390	390	MHz
GTL+	300	250	200	200	MHz
SSTL-3 Class I	400	350	300	300	MHz
SSTL-3 Class II	400	350	300	300	MHz
SSTL-2 Class I	400	350	300	300	MHz
SSTL-2 Class II	400	350	300	300	MHz
SSTL-18 Class I	400	350	300	300	MHz
SSTL-18 Class II	400	350	300	300	MHz
1.5-V HSTL Class I	400	350	300	300	MHz
1.8-V HSTL Class I	400	350	300	300	MHz
CTT	300	250	200	200	MHz
Differential 1.5-V HSTL C1	400	350	300	300	MHz
LVPECL *(1)*	717	717	640	640	MHz
PCML *(1)*	400	375	350	350	MHz
LVDS *(1)*	717	717	640	640	MHz
HyperTransport technology *(1)*	717	717	640	640	MHz

Table 4–120. Stratix Maximum Output Clock Rate for PLL[5, 6, 11, 12] Pins in Flip-Chip Packages (Part 2 of 2)

I/O Standard	-5 Speed Grade	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
SSTL-2 Class II *(3)*	200	200	167	167	MHz
SSTL-2 Class II *(4)*	200	200	167	167	MHz
SSTL-2 Class II *(5)*	150	134	134	134	MHz
SSTL-18 Class I	150	133	133	133	MHz
SSTL-18 Class II	150	133	133	133	MHz
1.5-V HSTL Class I	250	225	200	200	MHz
1.5-V HSTL Class II	225	200	200	200	MHz
1.8-V HSTL Class I	250	225	200	200	MHz
1.8-V HSTL Class II	225	200	200	200	MHz
3.3-V PCI	350	300	250	250	MHz
3.3-V PCI-X 1.0	350	300	250	250	MHz
Compact PCI	350	300	250	250	MHz
AGP 1×	350	300	250	250	MHz
AGP 2×	350	300	250	250	MHz
CTT	200	200	200	200	MHz
Differential 1.5-V HSTL C1	225	200	200	200	MHz
Differential 1.8-V HSTL Class I	250	225	200	200	MHz
Differential 1.8-V HSTL Class II	225	200	200	200	MHz
Differential SSTL-2 *(6)*	200	200	167	167	MHz
LVPECL *(2)*	500	500	500	500	MHz
PCML *(2)*	350	350	350	350	MHz
LVDS *(2)*	500	500	500	500	MHz
HyperTransport technology *(2)*	350	350	350	350	MHz

Table 4–121. Stratix Maximum Output Clock Rate (Using I/O Pins) for PLL[1, 2, 3, 4] Pins in Flip-Chip Packages

I/O Standard	-5 Speed Grade	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	400	350	300	300	MHz
2.5 V	400	350	300	300	MHz
1.8 V	400	350	300	300	MHz
1.5 V	350	300	300	300	MHz
LVCMOS	400	350	300	300	MHz
GTL	200	167	125	125	MHz
GTL+	200	167	125	125	MHz
SSTL-3 Class I	167	150	133	133	MHz
SSTL-3 Class II	167	150	133	133	MHz
SSTL-2 Class I	150	133	133	133	MHz
SSTL-2 Class II	150	133	133	133	MHz
SSTL-18 Class I	150	133	133	133	MHz
SSTL-18 Class II	150	133	133	133	MHz
1.5-V HSTL Class I	250	225	200	200	MHz
1.5-V HSTL Class II	225	225	200	200	MHz
1.8-V HSTL Class I	250	225	200	200	MHz
1.8-V HSTL Class II	225	225	200	200	MHz
3.3-V PCI	250	225	200	200	MHz
3.3-V PCI-X 1.0	225	225	200	200	MHz
Compact PCI	400	350	300	300	MHz
AGP 1×	400	350	300	300	MHz
AGP 2×	400	350	300	300	MHz
CTT	300	250	200	200	MHz
LVPECL *(2)*	717	717	500	500	MHz
PCML *(2)*	420	420	420	420	MHz
LVDS *(2)*	717	717	500	500	MHz
HyperTransport technology *(2)*	420	420	420	420	MHz