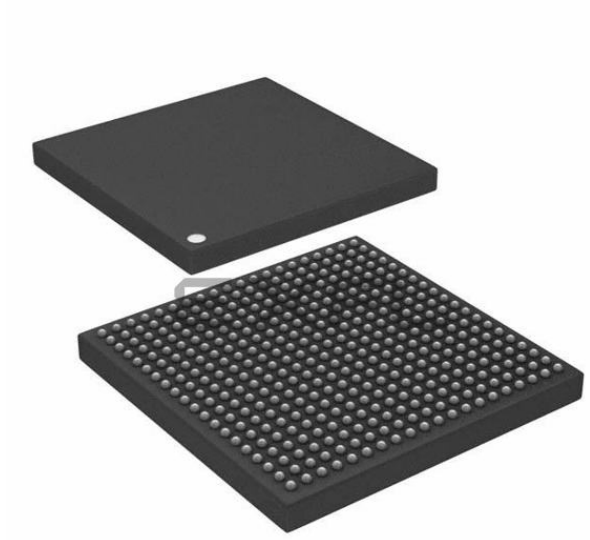

Welcome to E-XFL.COM

Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	Fixed Point
Interface	Host Interface, I²C, UART
Clock Rate	266MHz
Non-Volatile Memory	ROM (8kB)
On-Chip RAM	400kB
Voltage - I/O	3.30V
Voltage - Core	1.20V
Operating Temperature	-40°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	400-LFBGA
Supplier Device Package	400-LFBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/msc7116vf1000

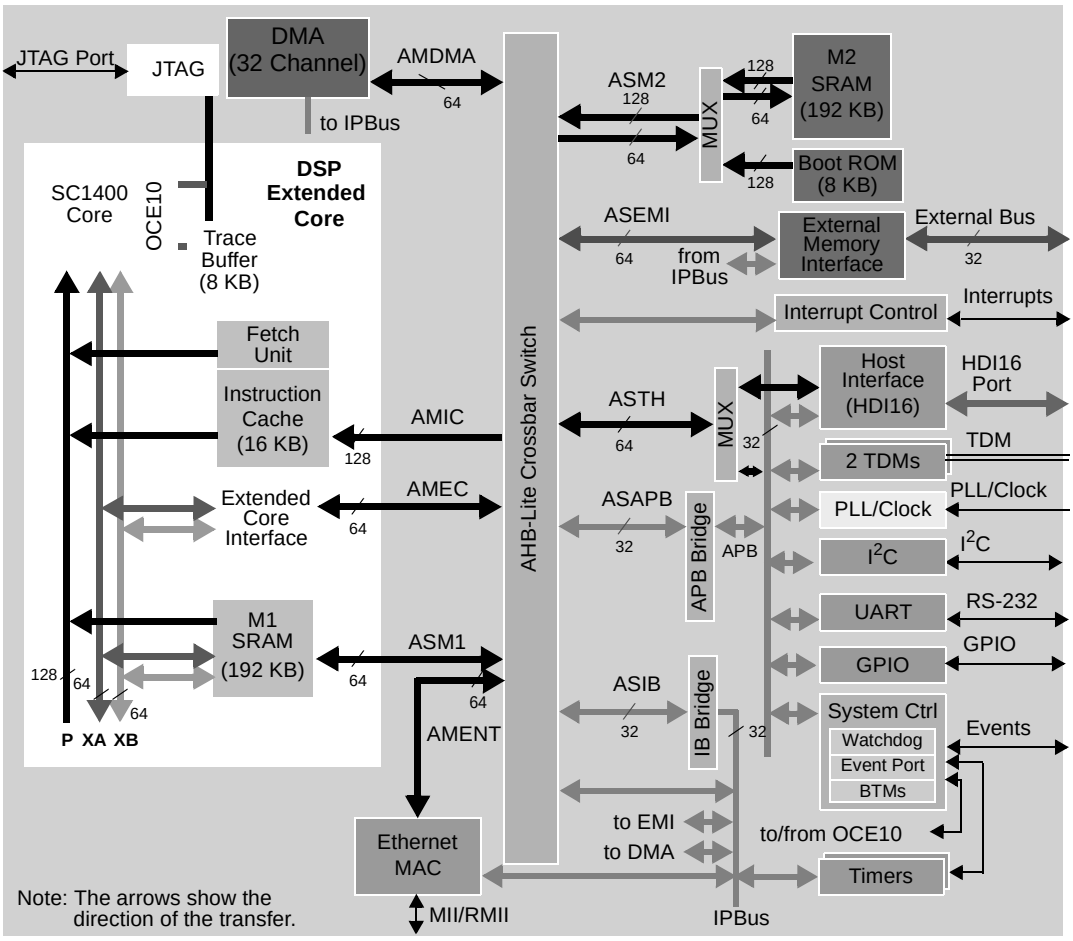

Figure 1. MSC7116 Block Diagram

1 Pin Assignments

This section includes diagrams of the MSC7116 package ball grid array layouts and pinout allocation tables.

1.1 MAP-BGA Ball Layout Diagrams

Top and bottom views of the MAP-BGA package are shown in **Figure 2** and **Figure 3** with their ball location index numbers.

Top View

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
A	GND	GND	DQM1	DQS2	CK	$\overline{CK}$	HD15	HD12	HD10	HD7	HD6	HD4	HD1	HD0	GND	BM3	NC	NC	NC	NC
B	V_{DDM}	NC	$\overline{CS0}$	DQM2	DQS3	DQS0	CKE	$\overline{WE}$	HD14	HD11	HD8	HD5	HD2	NC	BM2	NC	NC	NC	NC	NC
C	D24	D30	D25	$\overline{CS1}$	DQM3	DQM0	DQS1	$\overline{RAS}$	$\overline{CAS}$	HD13	HD9	HD3	NC	NC	NC	NC	NC	NC	NC	NC
D	V_{DDM}	D28	D27	GND	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDC}	NC	NC	NC
E	GND	D26	D31	V_{DDM}	V_{DDM}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDM}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDC}	V_{DDC}	NC	NC	NC
F	V_{DDM}	D15	D29	V_{DDC}	V_{DDC}	V_{DDC}	GND	GND	GND	V_{DDM}	V_{DDM}	GND	GND	GND	V_{DDIO}	V_{DDC}	V_{DDC}	NC	NC	NC
G	GND	D13	GND	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	NC	NC	NC
H	D14	D12	D11	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	NC	HA2	HA1
J	D10	V_{DDM}	D9	V_{DDM}	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDC}	HA3	$\overline{HACK}$	$\overline{HREQ}$
K	D0	GND	D8	V_{DDC}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	HA0	HDDS	$\overline{HDS}$
L	D1	GND	D3	V_{DDC}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDC}	$\overline{HCS2}$	$\overline{HCS1}$	HRW
M	D2	V_{DDM}	D5	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDC}	V_{DDC}	SDA	UTXD	URXD
N	D4	D6	V_{REF}	V_{DDM}	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDC}	V_{DDC}	CLKIN	SCL	V_{SSPLL}
P	D7	D17	D16	V_{DDM}	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	$\overline{PORESET}$	TPSEL	V_{DDPLL}
R	GND	D19	D18	V_{DDM}	V_{DDM}	V_{DDM}	GND	V_{DDM}	GND	V_{DDM}	GND	GND	V_{DDIO}	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	TDO	EE0	TEST0
T	V_{DDM}	D20	D22	V_{DDM}	V_{DDM}	V_{DDC}	V_{DDM}	V_{DDM}	V_{DDC}	V_{DDM}	V_{DDM}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDC}	V_{DDC}	MDIO	TMS	$\overline{HRESET}$
U	GND	D21	D23	V_{DDM}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	COL	TCK	$\overline{TRST}$
V	V_{DDM}	NC	A13	A11	A10	A5	A2	BA0	NC	EVNT0	EVNT4	T0TCK	T1RFS	T1TD	TX_ER	RXD2	RXD0	TX_EN	CRS	TDI
W	GND	V_{DDM}	A12	A8	A7	A6	A3	NC	EVNT1	EVNT2	T0RFS	T0TFS	T1RD	T1TFS	TXD2	RXD3	TXD1	TXCLK	RX_ER	MDC
Y	V_{DDM}	GND	A9	A1	A0	A4	BA1	$\overline{NMI}$	EVNT3	T0RCK	T0RD	TOTD	T1RCK	T1TCK	TXD3	RXCLK	TXD0	RXD1	GND	RX_DV

Figure 2. MSC7116 Molded Array Process-Ball Grid Array (MAP-BGA), Top View

Bottom View

	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
A	NC	NC	NC	NC	BM3	GND	HD0	HD1	HD4	HD6	HD7	HD10	HD12	HD15	$\overline{CK}$	CK	DQS2	DQM1	GND	GND
B	NC	NC	NC	NC	NC	BM2	NC	HD2	HD5	HD8	HD11	HD14	$\overline{WE}$	CKE	DQS0	DQS3	DQM2	$\overline{CS0}$	NC	V_{DDM}
C	NC	NC	NC	NC	NC	NC	NC	NC	HD3	HD9	HD13	$\overline{CAS}$	$\overline{RAS}$	DQS1	DQM0	DQM3	$\overline{CS1}$	D25	D30	D24
D	NC	NC	NC	V_{DD}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	GND	D27	D28	V_{DDM}
E	NC	NC	NC	V_{DD}	V_{DD}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDM}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DDM}	V_{DDM}	D31	D26	GND
F	NC	NC	NC	V_{DD}	V_{DD}	V_{DDIO}	GND	GND	GND	V_{DDM}	V_{DDM}	GND	GND	GND	V_{DD}	V_{DD}	V_{DD}	D29	D15	V_{DDM}
G	NC	NC	NC	V_{DD}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	GND	D13	GND
H	HA1	HA2	NC	V_{DD}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	D11	D12	D14
J	$\overline{HREQ}$	$\overline{HACK}$	HA3	V_{DD}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	V_{DDM}	D9	V_{DDM}	D10
K	$\overline{HDS}$	HDDS	HA0	V_{DD}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DD}	D8	GND	D0
L	HRW	$\overline{HCS1}$	$\overline{HCS2}$	V_{DD}	V_{DDIO}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DD}	D3	GND	D1
M	URXD	UTXD	SDA	V_{DD}	V_{DD}	GND	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	D5	V_{DDM}	D2
N	V_{SSPLL}	SCL	CLKIN	V_{DD}	V_{DD}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	V_{DDM}	V_{REF}	D6	D4
P	V_{DDPLL}	TPSEL	$\overline{PORESET}$	V_{DD}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	V_{DDM}	D16	D17	D7
R	TEST0	EE0	TDO	V_{DD}	V_{DDIO}	V_{DDIO}	GND	V_{DDIO}	GND	GND	V_{DDM}	GND	V_{DDM}	GND	V_{DDM}	V_{DDM}	V_{DDM}	D18	D19	GND
T	$\overline{HRESET}$	TMS	MDIO	V_{DD}	V_{DD}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDM}	V_{DDM}	V_{DD}	V_{DDM}	V_{DDM}	V_{DD}	V_{DDM}	V_{DDM}	D22	D20	V_{DDM}
U	$\overline{TRST}$	TCK	COL	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DDM}	D23	D21	GND
V	TDI	CRS	TX_EN	RXD0	RXD2	TX_ER	T1TD	T1RFS	T0TCK	EVNT4	EVNT0	NC	BA0	A2	A5	A10	A11	A13	NC	V_{DDM}
W	MDC	RX_ER	TXCLK	TXD1	RXD3	TXD2	T1TFS	T1RD	T0TFS	T0RFS	EVNT2	EVNT1	NC	A3	A6	A7	A8	A12	V_{DDM}	GND
Y	RX_DV	GND	RXD1	TXD0	RXCLK	TXD3	T1TCK	T1RCK	TOTD	T0RD	T0RCK	EVNT3	$\overline{NMI}$	BA1	A4	A0	A1	A9	GND	V_{DDM}

Figure 3. MSC7116 Molded Array Process-Ball Grid Array (MAP-BGA), Bottom View

1.2 Signal List By Ball Location

Table 1 lists the signals sorted by ball number and configuration.

Table 1. MSC7116 Signals by Ball Designator

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
A1	GND					
A2	GND					
A3	DQM1					
A4	DQS2					
A5	CK					
A6	$\overline{CK}$					
A7	GPIC7			GPOC7	HD15	
A8	GPIC4			GPOC4	HD12	
A9	GPIC2			GPOC2	HD10	
A10	reserved				HD7	
A11	reserved				HD6	
A12	reserved				HD4	
A13	reserved				HD1	
A14	reserved				HD0	
A15	GND					
A16	BM3	GPID8		GPOD8	reserved	
A17	NC					
A18	NC					
A19	NC					
A20	NC					
B1	V_{DDM}					
B2	NC					
B3	$\overline{CS0}$					
B4	DQM2					
B5	DQS3					
B6	DQS0					
B7	CKE					
B8	$\overline{WE}$					
B9	GPIC6			GPOC6	HD14	
B10	GPIC3			GPOC3	HD11	
B11	GPIC0			GPOC0	HD8	
B12	reserved				HD5	
B13	reserved				HD2	

Table 1. MSC7116 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
F6	V_{DDC}					
F7	GND					
F8	GND					
F9	GND					
F10	V_{DDM}					
F11	V_{DDM}					
F12	GND					
F13	GND					
F14	GND					
F15	V_{DDIO}					
F16	V_{DDC}					
F17	V_{DDC}					
F18	NC					
F19	NC					
F20	NC					
G1	GND					
G2	D13					
G3	GND					
G4	V_{DDM}					
G5	V_{DDM}					
G6	GND					
G7	GND					
G8	GND					
G9	GND					
G10	GND					
G11	GND					
G12	GND					
G13	GND					
G14	GND					
G15	V_{DDIO}					
G16	V_{DDIO}					
G17	V_{DDC}					
G18	NC					
G19	NC					
G20	NC					
H1	D14					

Table 1. MSC7116 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
H2	D12					
H3	D11					
H4	V_{DDM}					
H5	V_{DDM}					
H6	GND					
H7	GND					
H8	GND					
H9	GND					
H10	GND					
H11	GND					
H12	GND					
H13	GND					
H14	GND					
H15	V_{DDIO}					
H16	V_{DDIO}					
H17	V_{DDC}					
H18	NC					
H19	reserved				HA2	
H20	reserved				HA1	
J1	D10					
J2	V_{DDM}					
J3	D9					
J4	V_{DDM}					
J5	V_{DDM}					
J6	V_{DDM}					
J7	GND					
J8	GND					
J9	GND					
J10	GND					
J11	GND					
J12	GND					
J13	GND					
J14	GND					
J15	GND					
J16	V_{DDIO}					
J17	V_{DDC}					

Table 1. MSC7116 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
L14	V_{DDIO}					
L15	V_{DDIO}					
L16	V_{DDIO}					
L17	V_{DDC}					
L18	GPIB11			GPOB11	$\overline{HCS2}$/HCS2	
L19	reserved				$\overline{HCS1}$/HCS1	
L20	reserved				HRW or $\overline{HRD}$/HRD	
M1	D2					
M2	V_{DDM}					
M3	D5					
M4	V_{DDM}					
M5	V_{DDM}					
M6	GND					
M7	GND					
M8	GND					
M9	GND					
M10	GND					
M11	GND					
M12	GND					
M13	GND					
M14	GND					
M15	GND					
M16	V_{DDC}					
M17	V_{DDC}					
M18	GPIA14		$\overline{IRQ15}$	GPOA14	SDA	
M19	GPIA12		$\overline{IRQ3}$	GPOA12	UTXD	
M20	GPIA13		$\overline{IRQ2}$	GPOA13	URXD	
N1	D4					
N2	D6					
N3	V_{REF}					
N4	V_{DDM}					
N5	V_{DDM}					
N6	V_{DDM}					
N7	GND					
N8	GND					
N9	GND					

Table 1. MSC7116 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
N10	GND					
N11	GND					
N12	GND					
N13	GND					
N14	GND					
N15	V_{DDIO}					
N16	V_{DDC}					
N17	V_{DDC}					
N18	CLKIN					
N19	GPIA15		$\overline{IRQ14}$	GPOA15	SCL	
N20	V_{SSPLL}					
P1	D7					
P2	D17					
P3	D16					
P4	V_{DDM}					
P5	V_{DDM}					
P6	V_{DDM}					
P7	GND					
P8	GND					
P9	GND					
P10	GND					
P11	GND					
P12	GND					
P13	GND					
P14	GND					
P15	V_{DDIO}					
P16	V_{DDIO}					
P17	V_{DDC}					
P18	$\overline{PORESET}$					
P19	TPSEL					
P20	V_{DDPLL}					
R1	GND					
R2	D19					
R3	D18					
R4	V_{DDM}					
R5	V_{DDM}					

Table 18. DDR DRAM Output AC Timing (continued)

No.	Parameter	Symbol	Min	Max	Unit
209	Dn/DQMn output setup with respect to DQSn[3]	t_{DDKHDS}, t_{DDKLDS}	$0.25 \times t_{CK} - 750$	—	ps
210	Dn/DQMn output hold with respect to DQSn[3]	t_{DDKHDX}, t_{DDKLDX}	$0.25 \times t_{CK} - 750$	—	ps
211	DQSn preamble start[4]	t_{DDKHMP}	$-0.25 \times t_{CK}$	—	ps
212	DQSn epilogue end[5]	t_{DDKHME}	–600	600	ps

Notes:
1. All CK/$\overline{CK}$ referenced measurements are made from the crossing of the two signals ±0.1 V.
2. t_{DDKHMH} can be modified through the TCFG2[WRDD] DQSS override bits. The DRAM requires that the first write data strobe arrives 75–125% of a DRAM cycle after the write command is issued. Any skew between DQSn and CK must be considered when trying to achieve this 75%–125% goal. The TCFG2[WRDD] bits can be used to shift DQSn by 1/4 DRAM cycle increments. The skew in this case refers to an internal skew existing at the signal connections. By default, the CK/$\overline{CK}$ crossing occurs in the middle of the control signal (An/$\overline{RAS}$/$\overline{CAS}$/$\overline{WE}$/CKE) tenure. Setting TCFG2[ACSM] bit shifts the control signal assertion 1/2 DRAM cycle earlier than the default timing. This means that the signal is asserted no earlier than 600 ps before the CK/$\overline{CK}$ crossing and no later than 600 ps after the crossing time; the device uses 1200 ps of the skew budget (the interval from –600 to +600 ps). Timing is verified by referencing the falling edge of CK. See Chapter 10 of the *MSC711x Reference Manual* for details.
3. Determined by maximum possible skew between a data strobe (DQS) and any corresponding bit of data. The data strobe should be centered inside of the data eye.
4. Please note that this spec is in reference to the DQSn first rising edge. It could also be referenced from CK(r), but due to programmable delay of the write strobes (TCFG2[WRDD]), there pre-amble may be extended for a full DRAM cycle. For this reason, we reference from DQSn.
5. All outputs are referenced to the rising edge of CK. Note that this is essentially the CK/DQSn skew in spec 208. In addition there is no real “maximum” time for the epilogue end. JEDEC does not require this is as a device limitation, but simply for the chip to guarantee fast enough write-to-read turn-around times. This is already guaranteed by the memory controller operation.

Figure 6 shows the DDR DRAM output timing diagram.

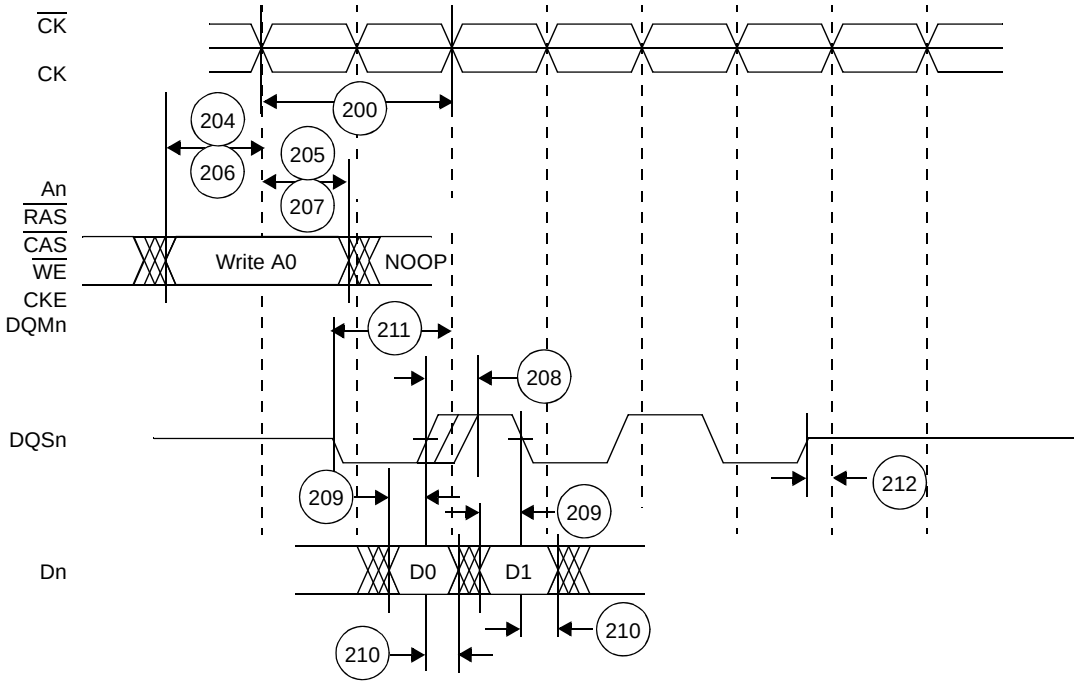

Figure 6. DDR DRAM Output Timing Diagram

2.5.6.2 Transmit Signal Timing

Table 22. Transmit Signal Timing

No.	Characteristics	Min	Max	Unit
800	Transmit clock period: • MII: TXCLK • RMII: REFCLK	 40 20	 — —	 ns ns
801	Transmit clock pulse width high—as a percent of clock period • MII: RXCLK • RMII: REFCLK	35 14 7	65 — —	% ns ns
802	Transmit clock pulse width low—as a percent of clock period: • MII: RXCLK • RMII: REFCLK	35 14 7	65 — —	% ns ns
805	Transmit clock to TXDn, TX_EN, TX_ER invalid	4	—	ns
806	Transmit clock to TXDn, TX_EN, TX_ER valid	—	14	ns

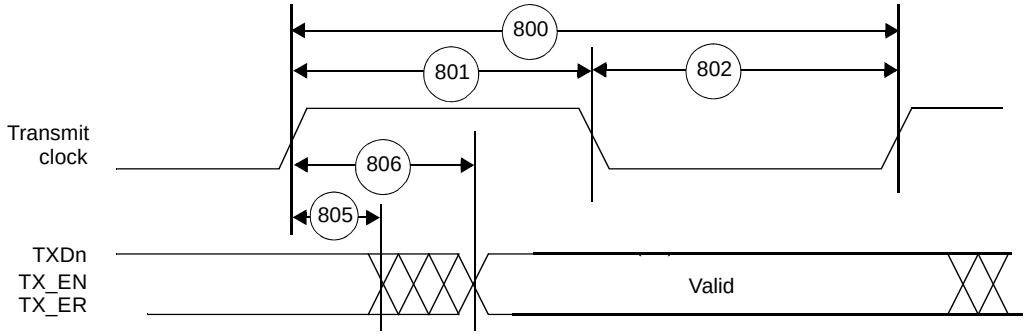

Figure 11. Ethernet Receive Signal Timing

2.5.6.3 Asynchronous Input Signal Timing

Table 23. Asynchronous Input Signal Timing

No.	Characteristics	Min	Max	Unit
807	• MII: CRS and COL minimum pulse width (1.5 × TXCLK period) • RMII: CRS_DV minimum pulse width (1.5 x REFCLK period)	60 30	— —	ns ns

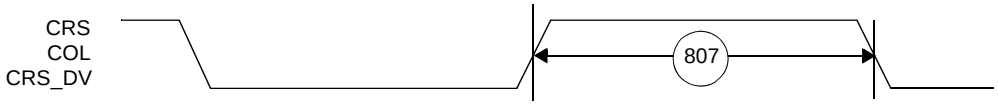

Figure 12. Asynchronous Input Signal Timing

2.5.6.4 Management Interface Timing

Table 24. Ethernet Controller Management Interface Timing

No.	Characteristics	Min	Max	Unit
808	MDC period	400	—	ns
809	MDC pulse width high	160	—	ns
810	MDC pulse width low	160	—	ns
811	MDS falling edge to MDIO output invalid (minimum propagation delay)	0	—	ns
812	MDS falling edge to MDIO output valid (maximum propagation delay)	—	15	ns
813	MDIO input to MDC rising edge setup time	10	—	ns
814	MDC rising edge to MDIO input hold time	10	—	ns

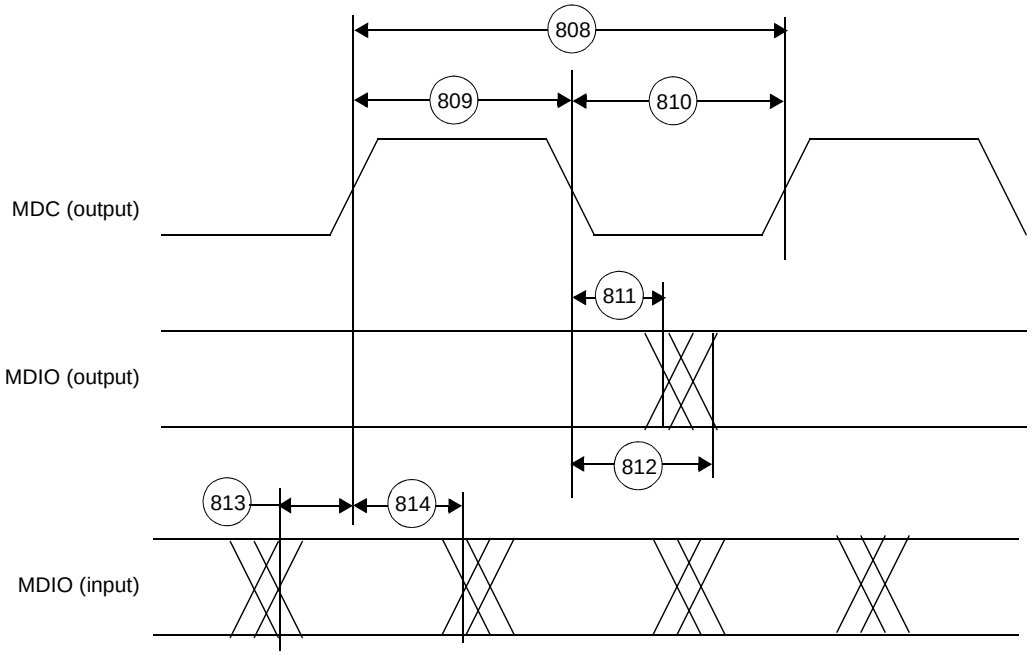

Figure 13. Serial Management Channel Timing

2.5.7 HDI16 Signals

Table 25. Host Interface (HDI16) Timing[1, 2]

No.	Characteristics[3]	Expression	Value	Unit
40	Host Interface Clock period	T_{CORE}	Note 1	ns
44a	Read data strobe minimum assertion width[4] HACK read minimum assertion width	$2.0 \times T_{CORE} + 9.0$	Note 11	ns
44b	Read data strobe minimum deassertion width[4] HACK read minimum deassertion width	$1.5 \times T_{CORE}$	Note 11	ns
44c	Read data strobe minimum deassertion width[4] after "Last Data Register" reads[5,6], or between two consecutive CVR, ICR, or ISR reads[7] HACK minimum deassertion width after "Last Data Register" reads[5,6]	$2.5 \times T_{CORE}$	Note 11	ns
45	Write data strobe minimum assertion width[8] HACK write minimum assertion width	$1.5 \times T_{CORE}$	Note 11	ns
46	Write data strobe minimum deassertion width[8] HACK write minimum deassertion width after ICR, CVR and Data Register writes[5]	$2.5 \times T_{CORE}$	Note 11	ns
47	Host data input minimum setup time before write data strobe deassertion[8] Host data input minimum setup time before HACK write deassertion	—	2.5	ns
48	Host data input minimum hold time after write data strobe deassertion[8] Host data input minimum hold time after HACK write deassertion	—	2.5	ns
49	Read data strobe minimum assertion to output data active from high impedance[4] HACK read minimum assertion to output data active from high impedance	—	1.0	ns
50	Read data strobe maximum assertion to output data valid[4] HACK read maximum assertion to output data valid	$(2.0 \times T_{CORE}) + 8.0$	Note 11	ns
51	Read data strobe maximum deassertion to output data high impedance[4] HACK read maximum deassertion to output data high impedance	—	9.0	ns
52	Output data minimum hold time after read data strobe deassertion[4] Output data minimum hold time after HACK read deassertion	—	1.0	ns
53	HCS[1–2] minimum assertion to read data strobe assertion[4]	—	0.5	ns
54	HCS[1–2] minimum assertion to write data strobe assertion[8]	—	0.0	ns
55	HCS[1–2] maximum assertion to output data valid	$(2.0 \times T_{CORE}) + 6.0$	Note 11	ns
56	HCS[1–2] minimum hold time after data strobe deassertion[9]	—	0.5	ns
57	HA[0–2], HRW minimum setup time before data strobe assertion[9]	—	5.0	ns
58	HA[0–2], HRW minimum hold time after data strobe deassertion[9]	—	5.0	ns
61	Maximum delay from read data strobe deassertion to host request deassertion for "Last Data Register" read[4, 5, 10]	$(3.0 \times T_{CORE}) + 6.0$	Note 11	ns
62	Maximum delay from write data strobe deassertion to host request deassertion for "Last Data Register" write[5,8,10]	$(3.0 \times T_{CORE}) + 6.0$	Note 11	ns
63	Minimum delay from DMA HACK (OAD=0) or Read/Write data strobe(OAD=1) deassertion to HREQ assertion.	$(2.0 \times T_{CORE}) + 1.0$	Note 11	ns
64	Maximum delay from DMA HACK (OAD=0) or Read/Write data strobe(OAD=1) assertion to HREQ deassertion	$(5.0 \times T_{CORE}) + 6.0$	Note 11	ns

Notes:
1. T_{CORE} = core clock period. At 300 MHz, T_{CORE} = 3.333 ns.
2. In the timing diagrams below, the controls pins are drawn as active low. The pin polarity is programmable.
3. V_{DD} = 3.3 V ± 0.15 V; T_J = –40°C to +105 °C, C_L = 30 pF for maximum delay timings and C_L = 0 pF for minimum delay timings.
4. The read data strobe is HRD/HRD in the dual data strobe mode and HDS/HDS in the single data strobe mode.
5. For 64-bit transfers, the "last data register" is the register at address 0x7, which is the last location to be read or written in data transfers. This is RX0/TX0 in the little endian mode (HBE = 0), or RX3/TX3 in the big endian mode (HBE = 1).
6. This timing is applicable only if a read from the "last data register" is followed by a read from the RX[0–3] registers without first polling RXDF or HREQ bits, or waiting for the assertion of the HREQ/HREQ signal.
7. This timing is applicable only if two consecutive reads from one of these registers are executed.
8. The write data strobe is HWR in the dual data strobe mode and HDS in the single data strobe mode.
9. The data strobe is host read (HRD/HRD) or host write (HWR/HWR) in the dual data strobe mode and host data strobe (HDS/HDS) in the single data strobe mode.
10. The host request is HREQ/HREQ in the single host request mode and HRRQ/HRRQ and HTRQ/HTRQ in the double host request mode. HRRQ/HRRQ is deasserted only when HOTX fifo is empty, HTRQ/HTRQ is deasserted only if HORX fifo is full
11. Compute the value using the expression.
12. The read and write data strobe minimum deassertion width for non-"last data register" accesses in single and dual data strobe modes is based on timings 57 and 58.

2.5.8 I²C Timing

Table 26. I²C Timing

No.	Characteristic	Fast		Unit
		Min	Max	
450	SCL clock frequency	0	400	kHz
451	Hold time START condition	(SCL clock period/2) – 0.3	—	µs
452	SCL low period	(SCL clock period/2) – 0.3	—	µs
453	SCL high period	(SCL clock period/2) – 0.1	—	µs
454	Repeated START set-up time (not shown in figure)	$2 \times 1/F_{BCK}$	—	µs
455	Data hold time	0	—	µs
456	Data set-up time	250	—	ns
457	SDA and SCL rise time	—	700	ns
458	SDA and SCL fall time	—	300	ns
459	Set-up time for STOP	(SCL clock period/2) – 0.7	—	µs
460	Bus free time between STOP and START	(SCL clock period/2) – 0.3	—	µs
Note:	SDA set-up time is referenced to the rising edge of SCL. SDA hold time is referenced to the falling edge of SCL. Load capacitance on SDA and SCL is 400 pF.			

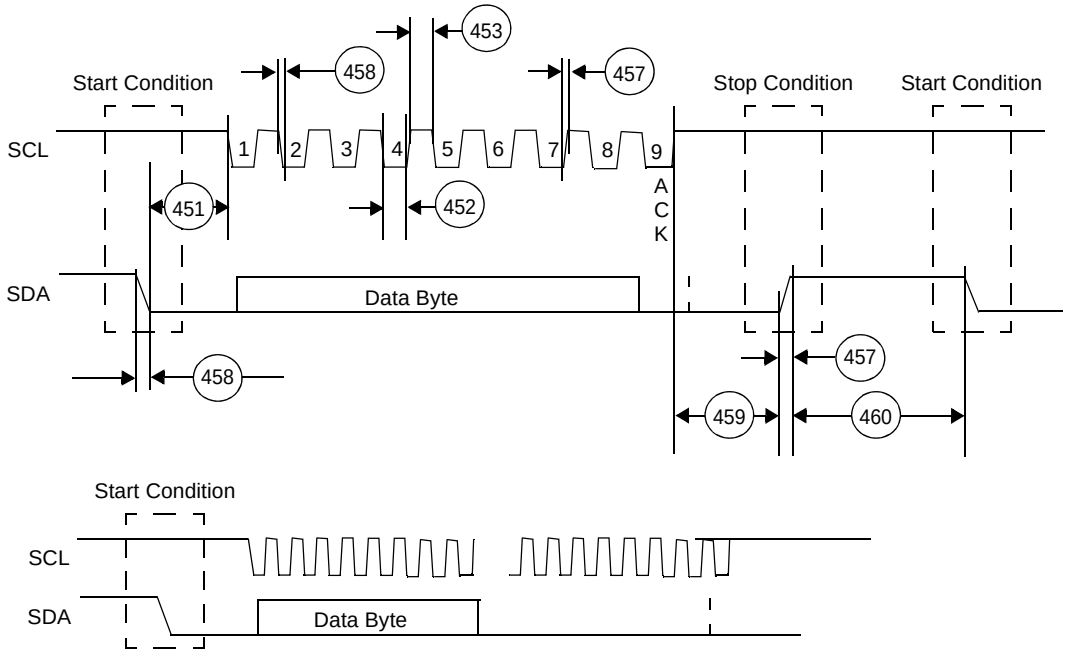

Figure 20. I²C Timing Diagram

3.2 Power Supply Design Considerations

This section outlines the MSC7116 power considerations: power supply, power sequencing, power planes, decoupling, power supply filtering, and power consumption. It also presents a recommended power supply design and options for low-power consumption. For information on AC/DC electrical specifications and thermal characteristics, refer to **Section 2**.

3.2.1 Power Supply

The MSC7116 requires four input voltages, as shown in **Table 32**.

Table 32. MSC7116 Voltages

Voltage	Symbol	Value
Core	V_{DDC}	1.2 V
Memory	V_{DDM}	2.5 V
Reference	V_{REF}	1.25 V
I/O	V_{DDIO}	3.3 V

You should supply the MSC7116 core voltage via a variable switching supply or regulator to allow for compatibility with possible core voltage changes on future silicon revisions. The core voltage is supplied with 1.2 V (+5% and –10%) across V_{DDC} and GND and the I/O section is supplied with 3.3 V (± 10%) across V_{DDIO} and GND. The memory and reference voltages supply the DDR memory controller block. The memory voltage is supplied with 2.5 V across V_{DDM} and GND. The reference voltage is supplied across V_{REF} and GND and must be between $0.49 \times V_{DDM}$ and $0.51 \times V_{DDM}$. Refer to the JEDEC standard JESD8 (*Stub Series Terminated Logic for 2.5 Volts* (STTL_2)) for memory voltage supply requirements.

3.2.2 Power Sequencing

One consequence of multiple power supplies is that the voltage rails ramp up at different rates when power is initially applied. The rates depend on the power supply, the type of load on each power supply, and the way different voltages are derived. It is extremely important to observe the power up and power down sequences at the board level to avoid latch-up, forward biasing of ESD devices, and excessive currents, which all lead to severe device damage.

Note: There are five possible power-up/power-down sequence cases. The first four cases listed in the following sections are recommended for new designs. The fifth case is not recommended for new designs and must be carefully evaluated for current spike risks based on actual information for the specific application.

3.2.2.1 Case 1

The power-up sequence is as follows:

1. Turn on the V_{DDIO} (3.3 V) supply first.
2. Turn on the V_{DDC} (1.2 V) supply second.
3. Turn on the V_{DDM} (2.5 V) supply third.
4. Turn on the V_{REF} (1.25 V) supply fourth (last).

The power-down sequence is as follows:

1. Turn off the V_{REF} (1.25 V) supply first.
2. Turn off the V_{DDM} (2.5 V) supply second.
3. Turn off the V_{DDC} (1.2 V) supply third.
4. Turn of the V_{DDIO} (3.3 V) supply fourth (last).

Use the following guidelines:

- Make sure that the time interval between the ramp-down of V_{DDIO} and V_{DDC} is less than 10 ms.
- Make sure that the time interval between the ramp-up or ramp-down for V_{DDC} and V_{DDM} is less than 10 ms for power-up and power-down.
- Refer to **Figure 30** for relative timing for power sequencing case 1.

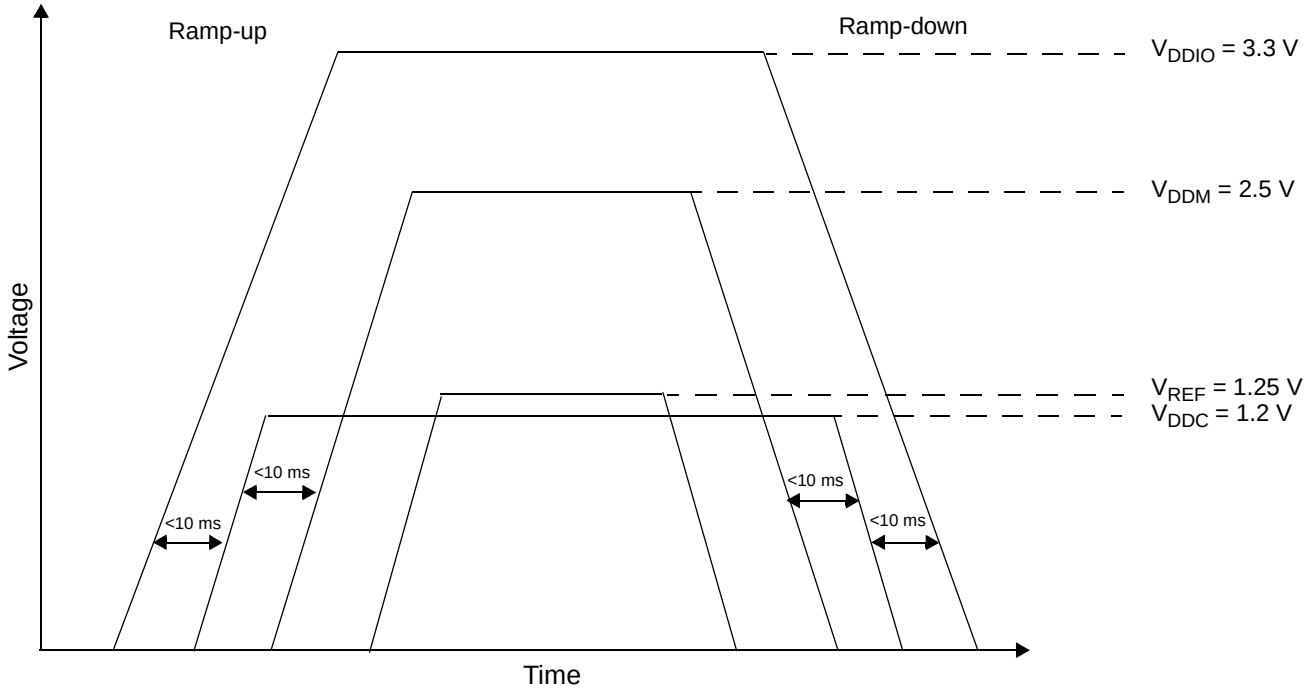

Figure 30. Voltage Sequencing Case 1

Table 35. Boot Mode Source Selection

BM[3–0]	Boot Port	Input Clock Frequency	Clock Divide	PLL	CKSEL	RNG Bit	Core Clock Frequency	Comments
HDI Boot Modes								
0000	HDI16	< F_{max}	N/A	N/A	00	0	< F_{max}	Not clocked by the PLL. Can boot as 8- or 16-bit HDI.
0101	HDI16	22.2-25 MHz	1	12	11	1	266–300 MHz	Can boot as 8- or 16-bit HDI.
0010	HDI16	25-33.3 MHz	2	32	01	1	200–266 MHz	
0111	HDI16	33-66 MHz	3	12	11	1	132–264 MHz	
0100	HDI16	44.3-50 MHz	2	12	11	1	266–300 MHz	
SPI Boot Modes - Using HA3, HCS2, BM3, BM2 Pins								
1000	SPI (SW)	< F_{max}	N/A	N/A	00	0	< F_{max}	The boot program automatically determines whether EEPROM or Flash memory.
1001	SPI (SW)	15.6-25 MHz	1	17	11	0	133–212.5 MHz	
1010	SPI (SW)	33-50 MHz	2	16	11	0	132–200 MHz	
1011	SPI (SW)	44.3-75 MHz	3	18	11	0	133–225 MHz	
SPI Boot Modes - Using URXD, UTXD, SCL, SDA Pins								
1100	SPI (SW)	< F_{max}	N/A	N/A	00	0	< F_{max}	Boots through different set of pins.
I^2C Boot Modes								
0001	I^2C	< 100 MHz	N/A	N/A	00	0	< 100 MHz	Not clocked by the PLL. I^2C is limited to a maximum bit rate of 400 Kbps. With a clock divider of 128, this limits the maximum input clock frequency to 100 MHz.
Reserved								
0011	Reserved	—	—	—	—	—	—	—
0110	Reserved	—	—	—	—	—	—	—
1101	Reserved	—	—	—	—	—	—	—
1110	Reserved	—	—	—	—	—	—	—
1111	Reserved	—	—	—	—	—	—	—

Notes: 1. The clock divider determines the value used in the clock module CLKCTRL[PLLDVF] field.
2. The clock multiplier determines the value used in the clock module CLKCTRL[PLLMLTF] field.
3. F_{max} is determined by the maximum frequency of the peripheral and of the SC1400 core as specified in the data sheet.

3.4.3 Boot

After a power-on reset, the PLL is bypassed and the device is directly clocked from the CLKIN pin. Thus, the device operates slowly during the boot process. After the boot program is loaded, it can enable the PLL and start the device operating at a higher speed. The MSC7116 can boot from an external host through the HDI16 or download a user program through the I^2C port. The boot operating mode is set by configuring the BM[0–3] signals sampled at the rising edge of $\overline{\text{PORESET}}$, as shown in **Table 35**. See the *MSC711x Reference Manual* for details of boot program operation.

3.4.3.1 HDI16 Boot

If the MSC7116 device boots from an external host through the HDI16, the port is configured as follows:

- Operate in Non-DMA mode.
- Operate in polled mode on the device side.
- Operate in polled mode on the external host side.
- External host must write four 16-bit values at a time with the first word as the most significant and the fourth word as the least significant.

When booting from a power-on reset, the HDI16 is additionally configurable as follows:

- 8- or 16-bit mode as specified by the H8BIT pin.
- Data strobe as specified by the HDSP and HDDS pins.

These pins are sampled only on the deassertion of power-on reset. During a boot from a hard reset, the configuration of these pins is unaffected.

Note: When the HDI16 is used for booting or other purposes, bit 0 is the least significant bit and not the most significant bit as for other DSP products.

3.4.3.2 I^2C Boot

When the MSC7116 device is configured to boot from the I^2C port, the boot program configures the GPIO pins for I^2C operation. Then the MSC7116 device initiates accesses to the I^2C module, downloading data to the MSC7116 device. The I^2C interface is configured as follows:

- PLL is disabled and bypassed so that the I^2C module is clocked with the IPBus clock.
- I^2C interface operates in master mode and polling is used.
- EPROM operates in slave mode.
- Clock divider is set to 128.
- Address of slave during boot is 0xA0.

The IPBus clock is internally divided to generate the bit clock, as follows:

- CLKIN must be a maximum of 100 MHz
- PLL is bypassed.
- IPBus clock = CLKIN/2 is a maximum of 50 MHz.
- I^2C bit clock must be less than or equal to:
  - IPBus clock/I^2C clock divider
  - 50 MHz (max)/128
  - 390.6 KHz

This satisfies the maximum clock rate requirement of 400 kbps for the I^2C interface. For details on the boot procedure, see the "Boot Program" chapter of the *MSC711x Reference Manual*.

3.4.3.3 SPI Boot

When the MSC7116 device is configured to boot from the SPI port, the boot program configures the GPIO pins for SPI operation. Then the MSC7116 device initiates accesses to the SPI module, downloading data to the MSC7116 device. When the SPI routines run in the boot ROM, the MSC7116 is always configured as the SPI master. Booting through the SPI is supported for serial EEPROM devices and serial Flash devices. When a READ_ID instruction is issued to the serial memory device and the device returns a value of 0x00 or 0xFF, the routines for accessing a serial EEPROM are used, at a maximum frequency of 4 Mbps. Otherwise, the routines for accessing a serial Flash are used, and they can run at faster speeds. Booting is performed through one of two sets of pins:

- Main set: BM[2–3], HA3, and HCS2, which allow use of the PLL.
- Alternate set: UTXD, URXD, SDA, and SCL, which cannot be used with the PLL.

In either configuration, an error during SPI boot is flagged on the EVNT3 pin. For details on the boot procedure, see the "Boot Program" chapter of the *MSC711x Reference Manual*.

3.5 DDR Memory System Guidelines

MSC7116 devices contain a memory controller that provides a glueless interface to external double data rate (DDR) SDRAM memory modules with Class 2 Series Stub Termination Logic 2.5 V (SSTL_2). There are two termination techniques, as shown in Figure 36. Technique B is the most popular termination technique.

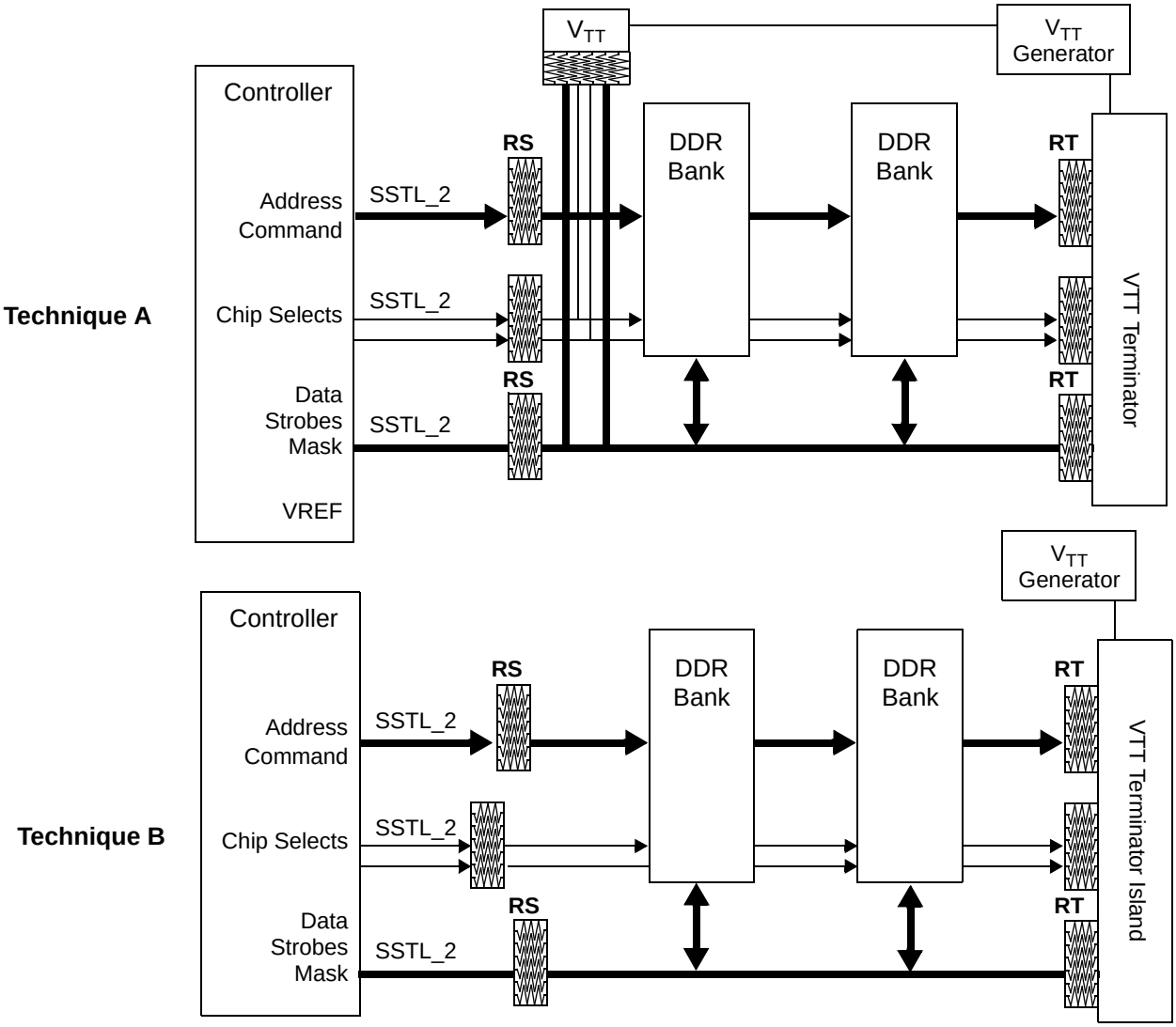

Figure 36. SSTL Termination Techniques

Figure 37 illustrates the power wattage for the resistors. Typical values for the resistors are as follows:

- RS = 22 Ω
- RT = 24 Ω

3.6 Connectivity Guidelines

This section summarizes the connections and special conditions, such as pull-up or pull-down resistors, for the MSC7116 device. Following are guidelines for signal groups and configuration settings:

- *Clock and reset signals*.
  - SWTE is used to configure the MSC7116 device and is sampled on the deassertion of $\overline{\text{PORESET}}$, so it should be tied to V_{DDC} or GND either directly or through pull-up or pull-down resistors until $\overline{\text{PORESET}}$ is deasserted. After $\overline{\text{PORESET}}$, this signal can be left floating.
  - BM[0–1] configure the MSC7116 device and are sampled until $\overline{\text{PORESET}}$ is deasserted, so they should be tied to V_{DDIO} or GND either directly or through pull-up or pull-down resistors.
  - $\overline{\text{HRESET}}$ should be pulled up.
- *Interrupt signals*. When used, $\overline{\text{IRQ}}$ pins must be pulled up.
- *HDI16 signals*.
  - When they are configured for open-drain, the $\overline{\text{HREQ}}$/HREQ or $\overline{\text{HTRQ}}$/HTRQ signals require a pull-up resistor. However, these pins are also sampled at power-on reset to determine the HDI16 boot mode and may need to be pulled down. When these pins must be pulled down on reset and pulled up otherwise, a buffer can be used with the $\overline{\text{HRESET}}$ signal as the enable.
  - When the device boots through the HDI16, the HDDS, HDSP and H8BIT pins should be pulled up or down, depending on the required boot mode settings.
- *Ethernet MAC/TDM2 signals*. The MDIO signal requires an external pull-up resistor.
- *I^2C signals*. The SCL and SDA signals, when programmed for I^2C, requires an external pull-up resistor.
- *General-purpose I/O (GPIO) signals*. An unused GPIO pin can be disconnected. After boot, program it as an output pin.
- Other signals.
  - The $\overline{\text{TEST0}}$ pin must be connected to ground.
  - The $\overline{\text{TPSEL}}$ pin should be pulled up to enable debug access via the EOnCE port and pulled down for boundary scan.
  - Pins labelled NO CONNECT (NC) must not be connected.
  - When a 16-pin double data rate (DDR) interface is used, the 16 unused data pins should be no connects (floating) if the used lines are terminated.
  - Do not connect DBREQ to DONE (as you would for the MSC8101 device). Connect DONE to one of the EVNT pins, and DBREQ to HRRQ.

4 Ordering Information

Consult a Freescale Semiconductor sales office or authorized distributor to determine product availability and place an order.

Part	Supply Voltage	Package Type	Pin Count	Core Frequency (MHz)	Solder Spheres	Order Number
MSC7116	1.2 V core 2.5 V memory 3.3 V I/O	Molded Array Process-Ball Grid Array (MAP-BGA)	400	266	Lead-free	MSC7116VM1000
					Lead-bearing	MSC7116VF1000

7 Revision History

Table 36 provides a revision history for this data sheet.

Table 36. Document Revision History

Revision	Date	Description
0	Apr 2004	• Initial public release.
1	May 2004	• Added ordering information and new package options.
2	Aug. 2004	• Updated clock parameter values. • Updated DDR timing specifications. • Updated I^2C timing specifications.
3	Sep. 2004	• Updated **Figures 1-2** and **1-2** to correct HDSP and DBREQ. • Corrected EE0 port reference. • Updated ball location for HDSP.
4	Jan. 2005	• Added signal HA3. • Updated absolute maximum ratings, DDR DRAM capacitance specifications, clock parameters, reset timing, and TDM timing. • Added note for timing reference for I^2C interface. • Expanded GPIO timing information. • Corrected pin T20 and K20 signal designation. • Corrected signal names to GPAO15 and $\overline{IRQ2}$. • Expanded design guidelines in Chapter 4.
5	Mar. 2005	• Updated features list. • Updated power specifications. • Changed CLKIN frequency range. • Added clock configuration information. • Updated JTAG timings.
6	Apr. 2005	• Added recommended power supply ratings and updated equations to estimate power consumption.
7	Oct. 2005	• Updated core and total power consumption examples.
8	Dec. 2005	• Added information about the new mask set 1M88B. Affected all sections.
9	Nov. 2006	• Updated arrows in Host DMA Writing Timing figure. • Updated boot overview in **Section 4.4.3**.
10	Apr. 2007	• Removed erroneous references to V_{CCSYN} and V_{CCSYN1}.
11	Jul. 2007	• Updated to new data sheet format. Reorganized and renumbered sections, figures, and tables. • Removed all references to obsolete mask set 1L44X and corresponding specification values. • Added a note to clarify the definition of TCK timing 700 in new **Table 31**. • Reworked reset and boot sections. • Expanded I^2C boot information and added SPI boot information. • Removed obsolete part numbers.
12	Aug 2007	• The power-up and power-down sequences described in **Section 3.2** starting on page 42 have been expanded to five possible design scenarios/cases. These cases replace the previously recommended power-up/power-down sequence recommendations. **Section 3.2** has been clarified by adding subsection headings.
13	Apr 2008	• Change the PLL filter resistor from 20 Ω to 2 Ω in **Section 3.2.5**.