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# Chapter 6

## S12Z Debug (S12ZDBGV2) Module

**Table 6-1. Revision History Table**

Revision Number	Revision Date	Sections Affected	Description Of Changes
2.04	19.APR.2012	<a href="#">Section 6.4.5.2.1</a>	Documented DBGTB read dependency on PROFILE bit
2.05	23.MAY.2012	General	Formatting changes to support DBGV3 from single source
2.06	10.SEP.2012	<a href="#">Section 6.4.5.3</a>	Added NOTE about PC trace buffer entries for Comp D timestamps
2.07	18.OCT.2012	General	Formatting corrections
2.08	16.NOV.2012	<a href="#">Section 6.5.1</a>	Modified step over breakpoint information
2.09	19.DEC.2012	General	Formatting corrections
2.10	28.JUN.2013	General <a href="#">Section 6.3.2.21</a> <a href="#">Section 6.3.2.1</a> <a href="#">Section 6.3.2.5</a>	Emphasized need to set TSOURCE for tracing or profiling Corrected DBGCDM write access dependency Corrected ARM versus PTACT dependency Modified DBGTBH read access dependencies
2.11	15.JUL.2013	<a href="#">Section 6.3.2</a>	Added explicit names to state control register bit fields

### 6.1 Introduction

The DBG module provides on-chip breakpoints and trace buffer with flexible triggering capability to allow non-intrusive debug of application software. The DBG module is optimized for the S12Z architecture and allows debugging of CPU module operations.

Typically the DBG module is used in conjunction with the BDC module, whereby the user configures the DBG module for a debugging session over the BDC interface. Once configured the DBG module is armed and the device leaves active BDM returning control to the user program, which is then monitored by the DBG module. Alternatively the DBG module can be configured over a serial interface using SWI routines.

#### 6.1.1 Glossary

**Table 6-2. Glossary Of Terms**

Term	Definition
COF	Change Of Flow. Change in the program flow due to a conditional branch, indexed jump or interrupt
PC	Program Counter

**Table 7-4. Reference Clock Frequency Selection if OSC\_LCP is enabled**

REFCLK Frequency Ranges (OSCE=1)	REFFRQ[1:0]
$1\text{MHz} \leq f_{\text{REF}} \leq 2\text{MHz}$	00
$2\text{MHz} < f_{\text{REF}} \leq 6\text{MHz}$	01
$6\text{MHz} < f_{\text{REF}} \leq 12\text{MHz}$	10
$f_{\text{REF}} > 12\text{MHz}$	11

### 7.3.2.10 S12CPMU\_UHV\_V5 COP Control Register (CPMUCOP)

This register controls the COP (Computer Operating Properly) watchdog.

The clock source for the COP is either ACLK, IRCCLK or OSCCLK depending on the setting of the COPOSCSEL0 and COPOSCSEL1 bit (see also [Table 7-8](#)).

In Stop Mode with PSTP=1 (Pseudo Stop Mode), COPOSCSEL0=1 and COPOSCSEL1=0 and PCE=1 the COP continues to run, else the COP counter halts in Stop Mode with COPOSCSEL1 =0.

In Full Stop Mode and Pseudo Stop Mode with COPOSCSEL1=1 the COP continues to run.

Module Base + 0x000C

	7	6	5	4	3	2	1	0
R	WCOP	RSBCK	0	0	0	CR2	CR1	CR0
W			WRTMASK					
Reset	F	0	0	0	0	F	F	F

After de-assert of System Reset the values are automatically loaded from the Flash memory. See Device specification for details.

 = Unimplemented or Reserved

**Figure 7-13. S12CPMU\_UHV\_V5 COP Control Register (CPMUCOP)**

Read: Anytime

Write:

1. RSBCK: Anytime in Special Mode; write to “1” but not to “0” in Normal Mode
2. WCOP, CR2, CR1, CR0:
  - Anytime in Special Mode, when WRTMASK is 0, otherwise it has no effect
  - Write once in Normal Mode, when WRTMASK is 0, otherwise it has no effect.
    - Writing CR[2:0] to “000” has no effect, but counts for the “write once” condition.
    - Writing WCOP to “0” has no effect, but counts for the “write once” condition.

When a non-zero value is loaded from Flash to CR[2:0] the COP time-out period is started.

A change of the COPOSCSEL0 or COPOSCSEL1 bit (writing a different value) or loosing UPOSC status while COPOSCSEL1 is clear and COPOSCSEL0 is set, re-starts the COP time-out period.

In Normal Mode the COP time-out period is restarted if either of these conditions is true:

1. Writing a non-zero value to CR[2:0] (anytime in special mode, once in normal mode) with WRTMASK = 0.
2. Writing WCOP bit (anytime in Special Mode, once in Normal Mode) with WRTMASK = 0.
3. Changing RSBCK bit from “0” to “1”.

In Special Mode, any write access to CPMUCOP register restarts the COP time-out period.

### 8.3.2.8 Timer Control Register 1/Timer Control Register 2 (TCTL1/TCTL2)

Module Base + 0x0008

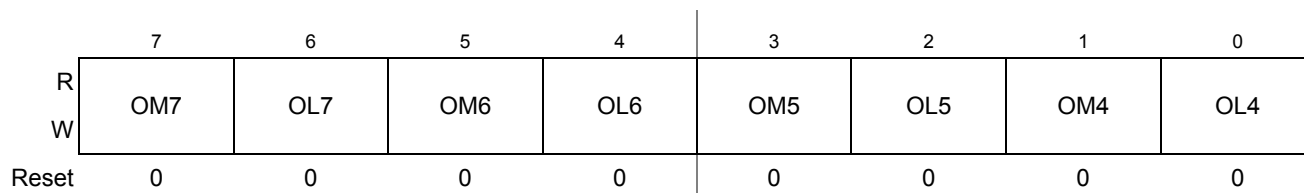


Figure 8-14. Timer Control Register 1 (TCTL1)

Module Base + 0x0009

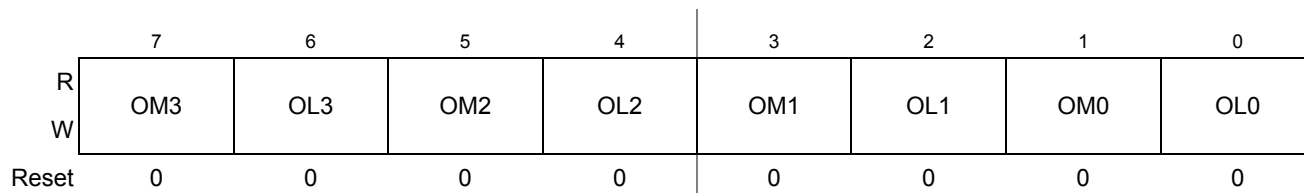


Figure 8-15. Timer Control Register 2 (TCTL2)

Read: Anytime

Write: Anytime

Table 8-8. TCTL1/TCTL2 Field Descriptions

**Note:** Writing to unavailable bits has no effect. Reading from unavailable bits return a zero

Field	Description
7:0 OMx	<p><b>Output Mode</b> — These eight pairs of control bits are encoded to specify the output action to be taken as a result of a successful OCx compare. When either OMx or OLx is 1, the pin associated with OCx becomes an output tied to OCx.</p> <p><b>Note:</b> To enable output action by OMx bits on timer port, the corresponding bit in OC7M should be cleared. For an output line to be driven by an OCx the OCPDx must be cleared.</p>
7:0 OLx	<p><b>Output Level</b> — These eightpairs of control bits are encoded to specify the output action to be taken as a result of a successful OCx compare. When either OMx or OLx is 1, the pin associated with OCx becomes an output tied to OCx.</p> <p><b>Note:</b> To enable output action by OLx bits on timer port, the corresponding bit in OC7M should be cleared. For an output line to be driven by an OCx the OCPDx must be cleared.</p>

Table 8-9. Compare Result Output Action

OMx	OLx	Action
0	0	No output compare action on the timer output signal
0	1	Toggle OCx output line
1	0	Clear OCx output line to zero
1	1	Set OCx output line to one

## 10.3 Signal Description

This section lists all inputs to the ADC12B\_LBA block.

### 10.3.1 Detailed Signal Descriptions

#### 10.3.1.1 AN $x$ ( $x = n, \dots, 2, 1, 0$ )

This pin serves as the analog input Channel  $x$ . The maximum input channel number is  $n$ . Please refer to the device reference manual for the maximum number of input channels.

#### 10.3.1.2 VRH\_0, VRH\_1, VRL\_0, VRL\_1

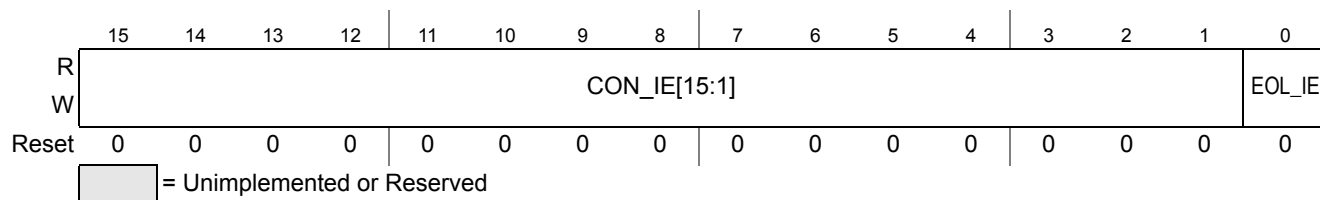
VRH\_0/1 are the high reference voltages, VRL0/1 are the low reference voltages for a ADC conversion selectable on a conversion command basis. Please refer to the device reference manual for availability and connectivity of these pins.

#### 10.3.1.3 VDDA, VSSA

These pins are the power supplies for the analog circuitry of the ADC12B\_LBA block.

### 10.4.2.11 ADC Conversion Interrupt Enable Register (ADCCONIE)

Module Base + 0x000A



**Figure 10-14. ADC Conversion Interrupt Enable Register (ADCCONIE)**

Read: Anytime

Write: Anytime

**Table 10-15. ADCCONIE Field Descriptions**

Field	Description
15-1 CON_IE[15:1]	<b>Conversion Interrupt Enable Bits</b> — These bits enable the individual interrupts which can be triggered via interrupt flags CON_IF[15:1]. 0 ADC conversion interrupt disabled. 1 ADC conversion interrupt enabled.
0 EOL_IE	<b>End Of List Interrupt Enable Bit</b> — This bit enables the end of conversion sequence list interrupt. 0 End of list interrupt disabled. 1 End of list interrupt enabled.



### 10.5.3.2.2 Introduction of the two Command Sequence Lists (CSLs)

The two Command Sequence Lists (CSLs) can be referred to via the Command Base Pointer Register plus the Command and Result Offset Registers plus the Command Index Register (ADCCBP, ADCCROFF\_0/1, ADCCIDX).

The final address for conversion command loading is calculated by the sum of these registers (e.g.: ADCCBP+ADCCROFF\_0+ADCCIDX or ADCCBP+ADCCROFF\_1+ADCCIDX).

Bit CSL\_BMOD selects if the CSL is used in double buffer or single buffer mode. In double buffer mode, the CSL can be swapped by flow control bits LDOK and RSTA. For detailed information about when and how the CSL is swapped, please refer to [Section 10.5.3.2.5, “The four ADC conversion flow control bits - description of Restart Event + CSL Swap](#), [Section 10.8.7.1, “Initial Start of a Command Sequence List](#) and [Section 10.8.7.3, “Restart CSL execution with new/other CSL \(alternative CSL becomes active CSL\) — CSL swapping](#)

Which list is actively used for ADC command loading is indicated by bit CSL\_SEL. The register to define the CSL start addresses (ADCCBP) can be set to any even location of the system RAM or NVM area. It is the user’s responsibility to make sure that the different ADC lists do not overlap or exceed the system RAM or the NVM area, respectively. The error flag IA\_EIF will be set for accesses to ranges outside system RAM area and cause an error interrupt if enabled.

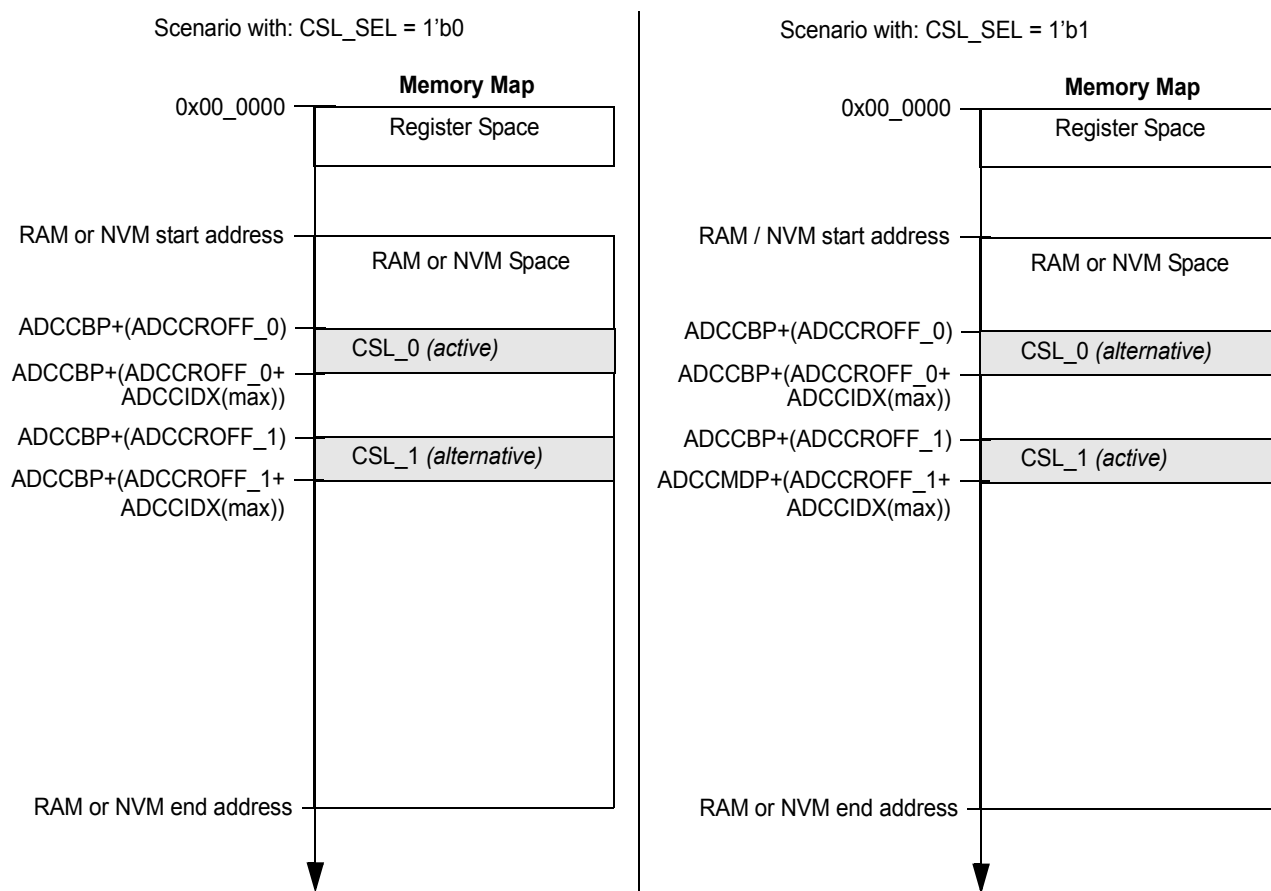


Figure 10-31. Command Sequence List Schema in Double Buffer Mode

### 11.4.5.5 MSCAN Sleep Mode

The CPU can request the MSCAN to enter this low power mode by asserting the SLPRQ bit in the CANCTL0 register. The time when the MSCAN enters sleep mode depends on a fixed synchronization delay and its current activity:

- If there are one or more message buffers scheduled for transmission (TXEx = 0), the MSCAN will continue to transmit until all transmit message buffers are empty (TXEx = 1, transmitted successfully or aborted) and then goes into sleep mode.
- If the MSCAN is receiving, it continues to receive and goes into sleep mode as soon as the CAN bus next becomes idle.
- If the MSCAN is neither transmitting nor receiving, it immediately goes into sleep mode.

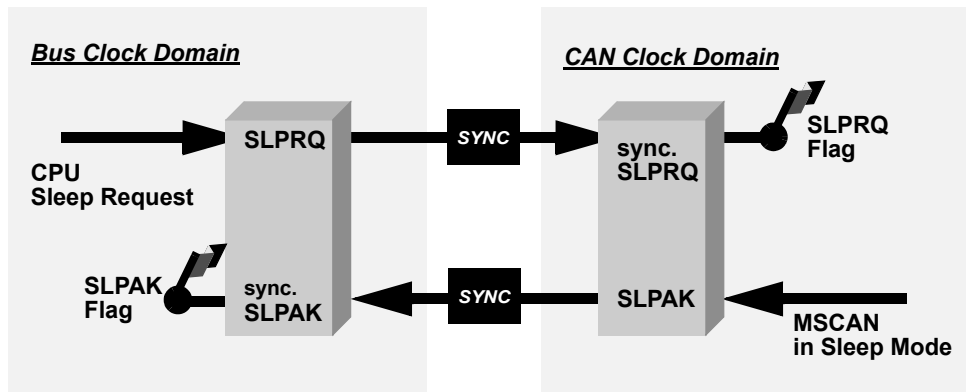


Figure 11-46. Sleep Request / Acknowledge Cycle

**NOTE**

The application software must avoid setting up a transmission (by clearing one or more TXEx flag(s)) and immediately request sleep mode (by setting SLPRQ). Whether the MSCAN starts transmitting or goes into sleep mode directly depends on the exact sequence of operations.

If sleep mode is active, the SLPRQ and SLPK bits are set (Figure 11-46). The application software must use SLPK as a handshake indication for the request (SLPRQ) to go into sleep mode.

When in sleep mode (SLPRQ = 1 and SLPK = 1), the MSCAN stops its internal clocks. However, clocks that allow register accesses from the CPU side continue to run.

If the MSCAN is in bus-off state, it stops counting the 128 occurrences of 11 consecutive recessive bits due to the stopped clocks. TXCAN remains in a recessive state. If RXF = 1, the message can be read and RXF can be cleared. Shifting a new message into the foreground buffer of the receiver FIFO (RxFG) does not take place while in sleep mode.

It is possible to access the transmit buffers and to clear the associated TXE flags. No message abort takes place while in sleep mode.

In Figure 12-24, a large burst of noise is perceived as the beginning of a start bit, although the test sample at RT5 is high. The RT5 sample sets the noise flag. Although this is a worst-case misalignment of perceived bit time, the data samples RT8, RT9, and RT10 are within the bit time and data recovery is successful.

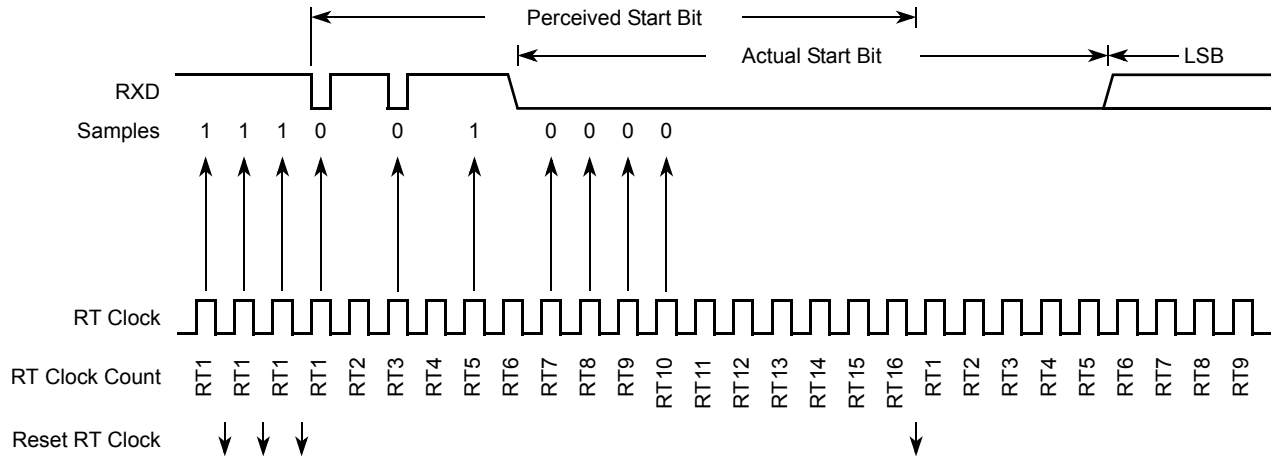


Figure 12-24. Start Bit Search Example 3

Figure 12-25 shows the effect of noise early in the start bit time. Although this noise does not affect proper synchronization with the start bit time, it does set the noise flag.

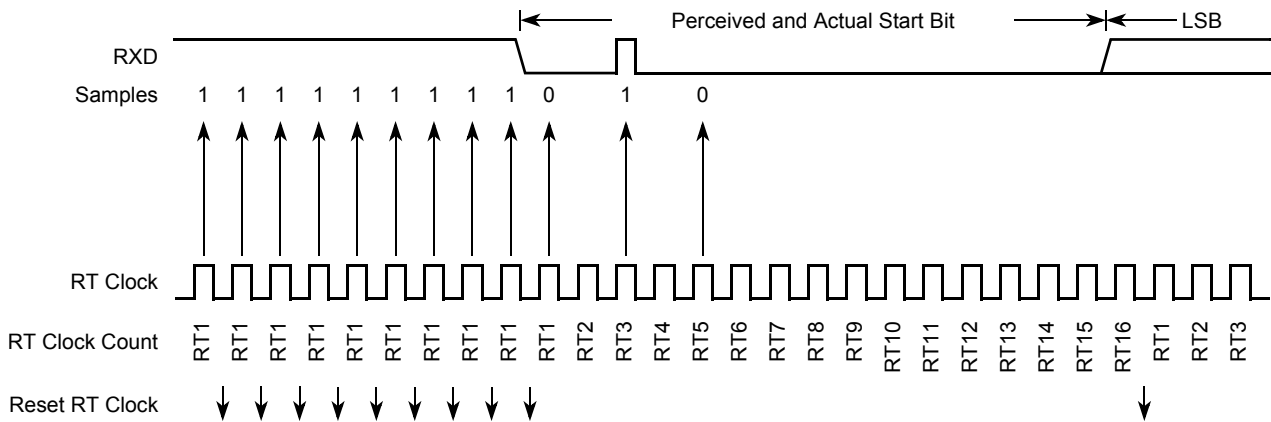


Figure 12-25. Start Bit Search Example 4

**Table 12-20. SCI Interrupt Sources**

RXEDGIF	SCIASR1[7]	RXEDGIE	Active high level. Indicates that an active edge (falling for RXPOL = 0, rising for RXPOL = 1) was detected.
BERRIF	SCIASR1[1]	BERRIE	Active high level. Indicates that a mismatch between transmitted and received data in a single wire application has happened.
BKDIF	SCIASR1[0]	BRKDIE	Active high level. Indicates that a break character has been received.

### 12.5.3.1 Description of Interrupt Operation

The SCI only originates interrupt requests. The following is a description of how the SCI makes a request and how the MCU should acknowledge that request. The interrupt vector offset and interrupt number are chip dependent. The SCI only has a single interrupt line (SCI Interrupt Signal, active high operation) and all the following interrupts, when generated, are ORed together and issued through that port.

#### 12.5.3.1.1 TDRE Description

The TDRE interrupt is set high by the SCI when the transmit shift register receives a byte from the SCI data register. A TDRE interrupt indicates that the transmit data register (SCIDRH/L) is empty and that a new byte can be written to the SCIDRH/L for transmission. Clear TDRE by reading SCI status register 1 with TDRE set and then writing to SCI data register low (SCIDRL).

#### 12.5.3.1.2 TC Description

The TC interrupt is set by the SCI when a transmission has been completed. Transmission is completed when all bits including the stop bit (if transmitted) have been shifted out and no data is queued to be transmitted. No stop bit is transmitted when sending a break character and the TC flag is set (providing there is no more data queued for transmission) when the break character has been shifted out. A TC interrupt indicates that there is no transmission in progress. TC is set high when the TDRE flag is set and no data, preamble, or break character is being transmitted. When TC is set, the TXD pin becomes idle (logic 1). Clear TC by reading SCI status register 1 (SCISR1) with TC set and then writing to SCI data register low (SCIDRL). TC is cleared automatically when data, preamble, or break is queued and ready to be sent.

#### 12.5.3.1.3 RDRF Description

The RDRF interrupt is set when the data in the receive shift register transfers to the SCI data register. A RDRF interrupt indicates that the received data has been transferred to the SCI data register and that the byte can now be read by the MCU. The RDRF interrupt is cleared by reading the SCI status register one (SCISR1) and then reading SCI data register low (SCIDRL).

#### 12.5.3.1.4 OR Description

The OR interrupt is set when software fails to read the SCI data register before the receive shift register receives the next frame. The newly acquired data in the shift register will be lost in this case, but the data already in the SCI data registers is not affected. The OR interrupt is cleared by reading the SCI status register one (SCISR1) and then reading SCI data register low (SCIDRL).

#### 12.5.3.1.5 IDLE Description

The IDLE interrupt is set when 10 consecutive logic 1s (if M = 0) or 11 consecutive logic 1s (if M = 1) appear on the receiver input. Once the IDLE is cleared, a valid frame must again set the RDRF flag before an idle condition can set the IDLE flag. Clear IDLE by reading SCI status register 1 (SCISR1) with IDLE set and then reading SCI data register low (SCIDRL).

The following sequence should be used to update the current magnitude and direction for coil 0 and coil 1 of the motor to achieve consistent PWM output:

1. Write to duty cycle register x
2. Write to duty cycle register x + 1.

At the next timer counter overflow, the duty cycle registers will be copied to the working duty cycle registers. Sequential writes to the duty cycle register x will result in the previous data being overwritten.

### 16.4.1.1.2 Full H-Bridge Mode (MCOM = 10)

In full H-bridge mode, the PWM channels x and x + 1 operate independently. The duty cycle working registers are updated whenever a timer counter overflow occurs.

### 16.4.1.1.3 Half H-Bridge Mode (MCOM = 00 or 01)

In half H-bridge mode, the PWM channels x and x + 1 operate independently. In this mode, each PWM channel can be configured such that one pin is released and the other pin is a PWM output. [Figure 16-11](#) shows a typical configuration in half H-bridge mode.

The two pins associated with each channel are switchable between released mode and PWM output dependent upon the state of the MCOM[1:0] bits in the MCCCx (channel control) register. See register description in [Section 16.3.2.4, “Motor Controller Channel Control Registers”](#). In half H-bridge mode, the state of the S bit has no effect.

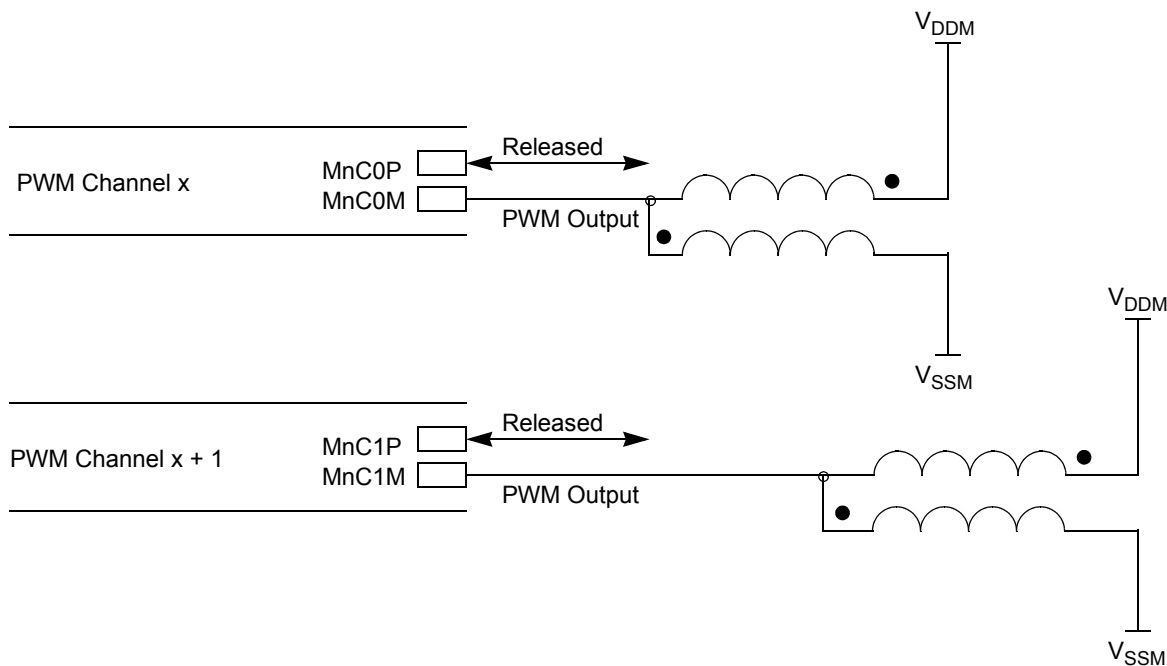


Figure 16-11. Typical Quad Half H-Bridge Mode Configuration

## 16.4.2 PWM Duty Cycle

The PWM duty cycle for the motor controller channel  $x$  can be determined by dividing the decimal representation of bits  $D[10:0]$  in  $MCDCx$  by the decimal representation of the bits  $P[10:0]$  in  $MCPER$  and multiplying the result by 100% as shown in the equation below:

$$\text{Effective PWM Channel X \% Duty Cycle} = \frac{\text{DUTY}}{\text{MCPER}} \cdot 100\%$$

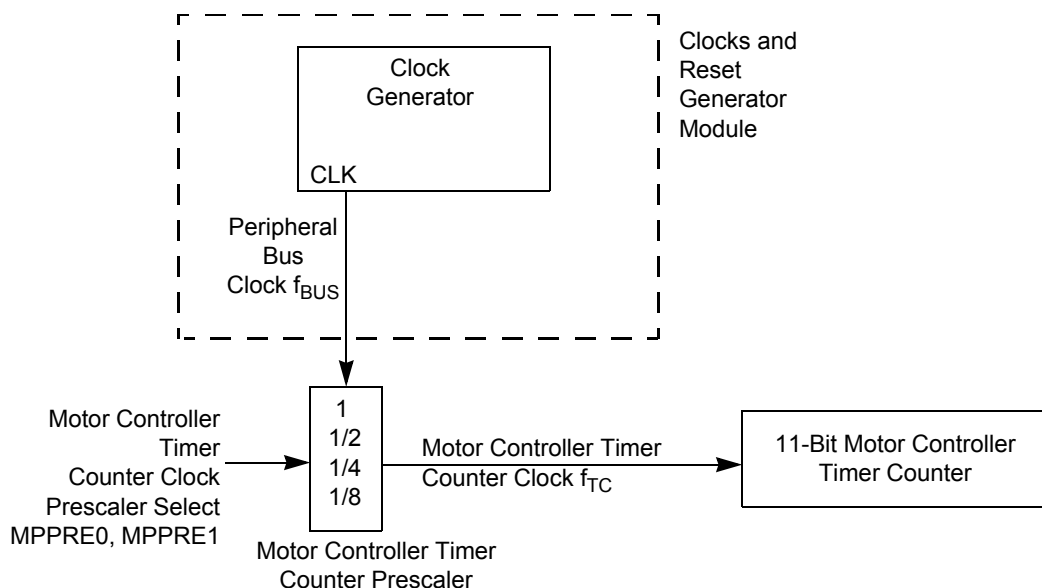
### NOTE

$x$  = PWM Channel Number = 0, 1, 2, 3 ... 8. This equation is only valid if  $\text{DUTY} \leq \text{MCPER}$  and  $\text{MCPER}$  is not equal to 0.

Whenever  $D[10:0] \geq P[10:0]$ , a constant low level ( $\text{RECIRC} = 0$ ) or high level ( $\text{RECIRC} = 1$ ) will be output.

## 16.4.3 Motor Controller Counter Clock Source

Figure 16-22 shows how the PWM motor controller timer counter clock source is selected.



**Figure 16-22. Motor Controller Counter Clock Selection**

The peripheral bus clock is the source for the motor controller counter prescaler. The motor controller counter clock rate,  $f_{TC}$ , is set by selecting the appropriate prescaler value. The prescaler is selected with the  $\text{MCPRE}[1:0]$  bits in motor controller control register 0 ( $\text{MCCTL0}$ ). The motor controller channel frequency of operation can be calculated using the following formula if  $\text{DITH} = 0$ :

$$\text{Motor Channel Frequency (Hz)} = \frac{f_{TC}}{\text{MCPER} \cdot M}$$

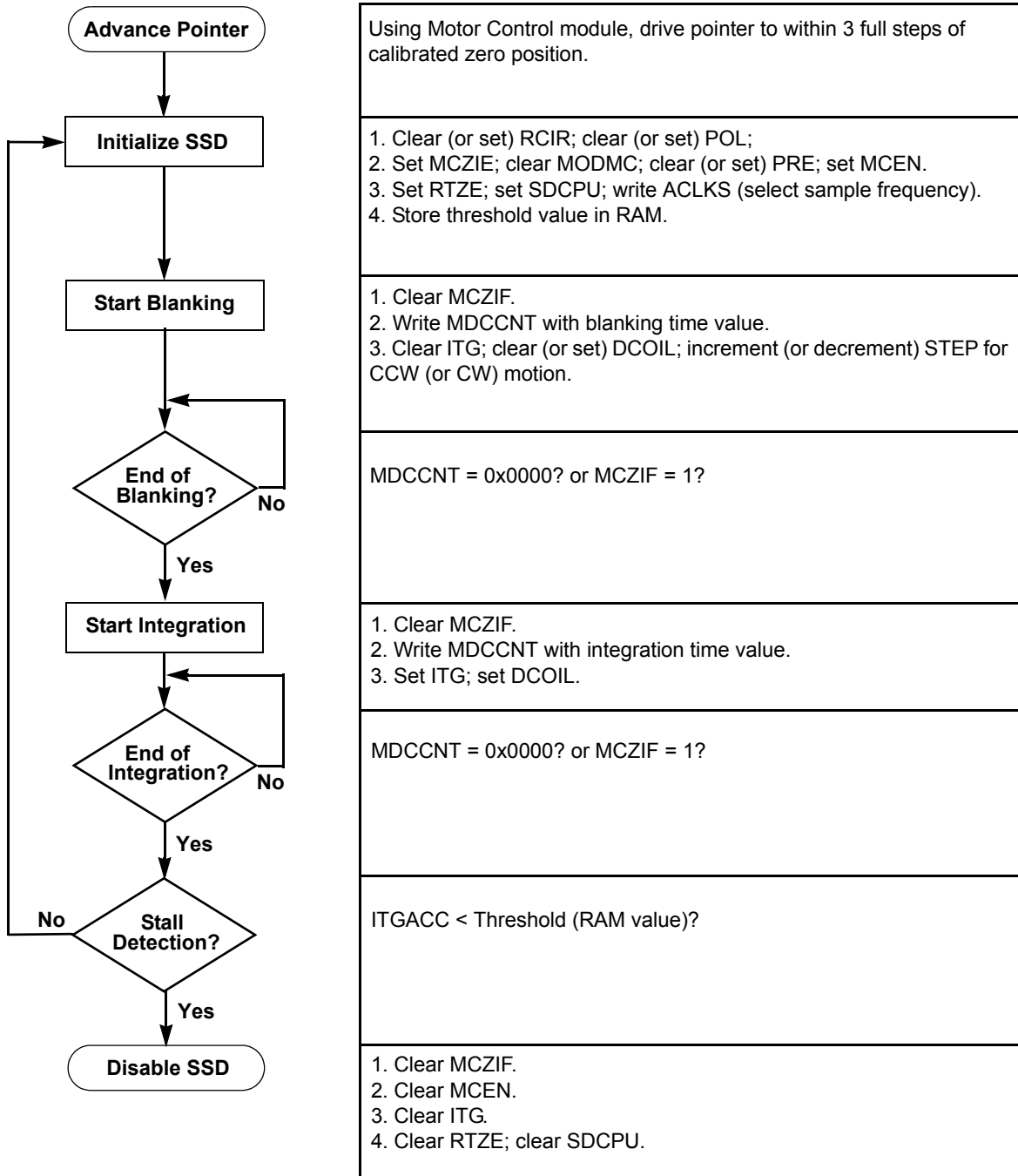


Figure 17-15. Return-to-Zero Flowchart



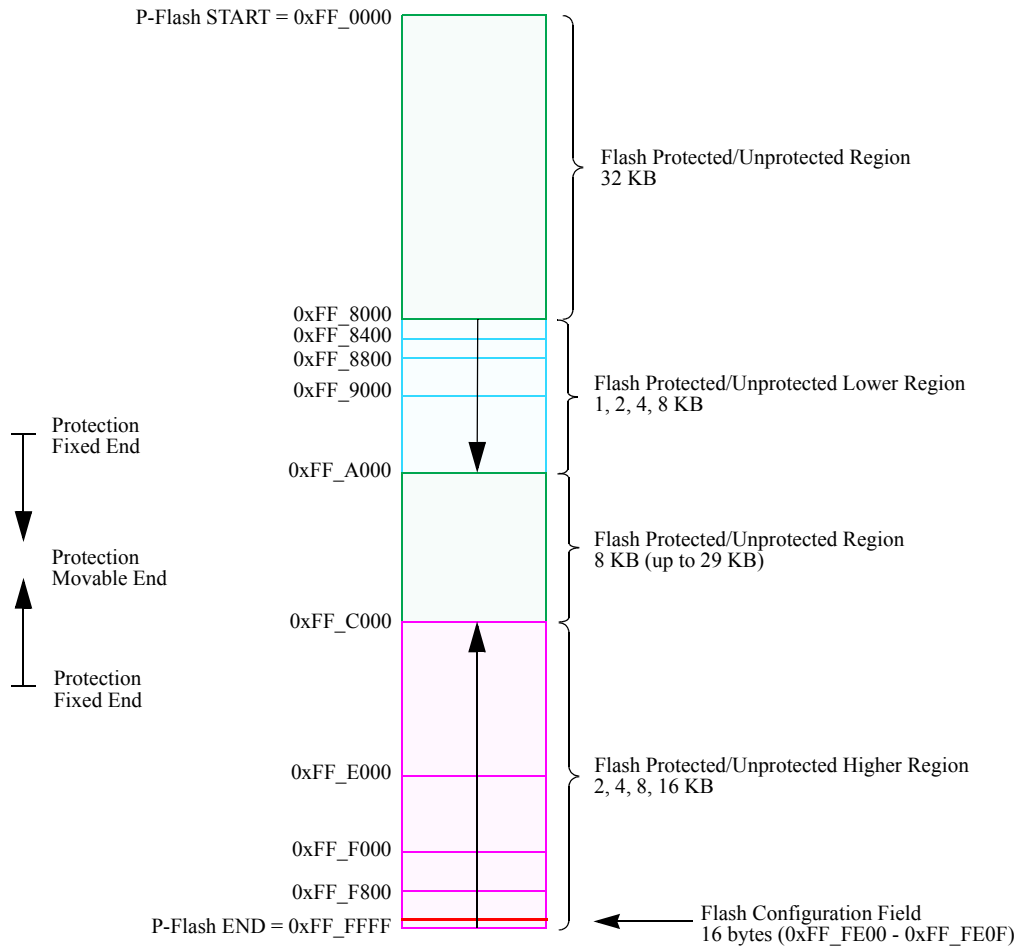


Figure 21-2. P-Flash Memory Map

**CAUTION**

The FCLKDIV register should never be written while a Flash command is executing (CCIF=0).

**Table 21-7. FCLKDIV Field Descriptions**

Field	Description
7 FDIVLD	<b>Clock Divider Loaded</b> 0 FCLKDIV register has not been written since the last reset 1 FCLKDIV register has been written since the last reset
6 FDIVLCK	<b>Clock Divider Locked</b> 0 FDIV field is open for writing 1 FDIV value is locked and cannot be changed. Once the lock bit is set high, only reset can clear this bit and restore writability to the FDIV field in normal mode.
5–0 FDIV[5:0]	<b>Clock Divider Bits</b> — FDIV[5:0] must be set to effectively divide BUSCLK down to 1 MHz to control timed events during Flash program and erase algorithms. <a href="#">Table 21-8</a> shows recommended values for FDIV[5:0] based on the BUSCLK frequency. Please refer to <a href="#">Section 21.4.5, “Flash Command Operations,”</a> for more information.

**Table 21-8. FDIV values for various BUSCLK Frequencies**

BUSCLK Frequency (MHz)		FDIV[5:0]	BUSCLK Frequency (MHz)		FDIV[5:0]
MIN <sup>(1)</sup>	MAX <sup>(2)</sup>		MIN <sup>1</sup>	MAX <sup>2</sup>	
1.0	1.6	0x00	26.6	27.6	0x1A
1.6	2.6	0x01	27.6	28.6	0x1B
2.6	3.6	0x02	28.6	29.6	0x1C
3.6	4.6	0x03	29.6	30.6	0x1D
4.6	5.6	0x04	30.6	31.6	0x1E
5.6	6.6	0x05	31.6	32.6	0x1F
6.6	7.6	0x06	32.6	33.6	0x20
7.6	8.6	0x07	33.6	34.6	0x21
8.6	9.6	0x08	34.6	35.6	0x22
9.6	10.6	0x09	35.6	36.6	0x23
10.6	11.6	0x0A	36.6	37.6	0x24
11.6	12.6	0x0B	37.6	38.6	0x25
12.6	13.6	0x0C	38.6	39.6	0x26
13.6	14.6	0x0D	39.6	40.6	0x27
14.6	15.6	0x0E	40.6	41.6	0x28
15.6	16.6	0x0F	41.6	42.6	0x29
16.6	17.6	0x10	42.6	43.6	0x2A
17.6	18.6	0x11	43.6	44.6	0x2B

### 21.3.2.9.1 P-Flash Protection Restrictions

The general guideline is that P-Flash protection can only be added and not removed. [Table 21-23](#) specifies all valid transitions between P-Flash protection scenarios. Any attempt to write an invalid scenario to the FPROT register will be ignored. The contents of the FPROT register reflect the active protection scenario. See the FPHS and FPLS bit descriptions for additional restrictions.

**Table 21-23. P-Flash Protection Scenario Transitions**

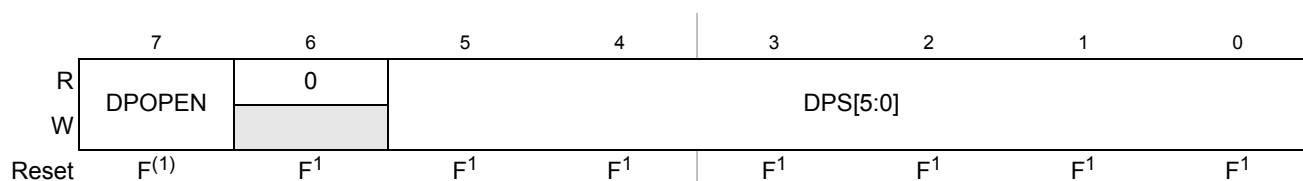
From Protection Scenario	To Protection Scenario <sup>(1)</sup>							
	0	1	2	3	4	5	6	7
0	X	X	X	X				
1		X		X				
2			X	X				
3				X				
4				X	X			
5			X	X	X	X		
6		X		X	X		X	
7	X	X	X	X	X	X	X	X

1. Allowed transitions marked with X, see [Figure 21-14](#) for a definition of the scenarios.

### 21.3.2.10 EEPROM Protection Register (DFPROT)

The DFPROT register defines which EEPROM sectors are protected against program and erase operations.

Offset Module Base + 0x0009



**Figure 21-15. EEPROM Protection Register (DFPROT)**

1. Loaded from Flash configuration field, during reset sequence.

The (unreserved) bits of the DFPROT register are writable with the restriction that protection can be added but not removed. Writes must increase the DPS value and the DPOPEN bit can only be written from 1 (protection disabled) to 0 (protection enabled). If the DPOPEN bit is set, the state of the DPS bits is irrelevant.

During the reset sequence, fields DPOPEN and DPS of the DFPROT register are loaded with the contents of the EEPROM protection byte in the Flash configuration field at global address 0xFF\_FE0D located in P-Flash memory (see [Table 21-4](#)) as indicated by reset condition F in [Table 21-25](#). To change the

## 23.2 External Signal Description

This section lists and describes the signals that connect off chip as well as internal supply nodes and special signals.

### 23.2.1 LIN — LIN Bus Pin

This pad is connected to the single-wire LIN data bus.

### 23.2.2 LGND — LIN Ground Pin

This pin is the device LIN ground connection. It is used to sink currents related to the LIN Bus pin. A decoupling capacitor external to the device (typically 220 pF, X7R ceramic) between LIN and LGND can further improve the quality of this ground and filter noise.

### 23.2.3 VLINSUP — Positive Power Supply

External power supply to the chip. The VLINSUP supply mapping is described in device level documentation.

### 23.2.4 LPTxD — LIN Transmit Pin

This pin can be routed to the SCI, LPDR1 register bit, an external pin, or other options. Please refer to the PIM chapter of the device specification for the available routing options.

This input is only used in normal mode; in other modes the value of this pin is ignored.

### 23.2.5 LPRxD — LIN Receive Pin

This pin can be routed to the SCI, an external pin, or other options. Please refer to the PIM chapter of the device specification for the available routing options.

In standby mode this output is disabled, and sends only a short pulse in case the wake-up functionality is enabled and a valid wake-up pulse was received in the LIN Bus.

