Welcome to E-XFL.COM

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	HCS12Z
Core Size	16-Bit
Speed	32MHz
Connectivity	CANbus, I²C, LINbus, SCI, SPI
Peripherals	DMA, LCD, POR, PWM, WDT
Number of I/O	100
Program Memory Size	64KB (64K x 8)
Program Memory Type	FLASH
EEPROM Size	4K x 8
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	5.5V ~ 18V
Data Converters	A/D 8x10b SAR
Oscillator Type	External, Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	144-LQFP
Supplier Device Package	144-LQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/s912zvhy64f1vlq

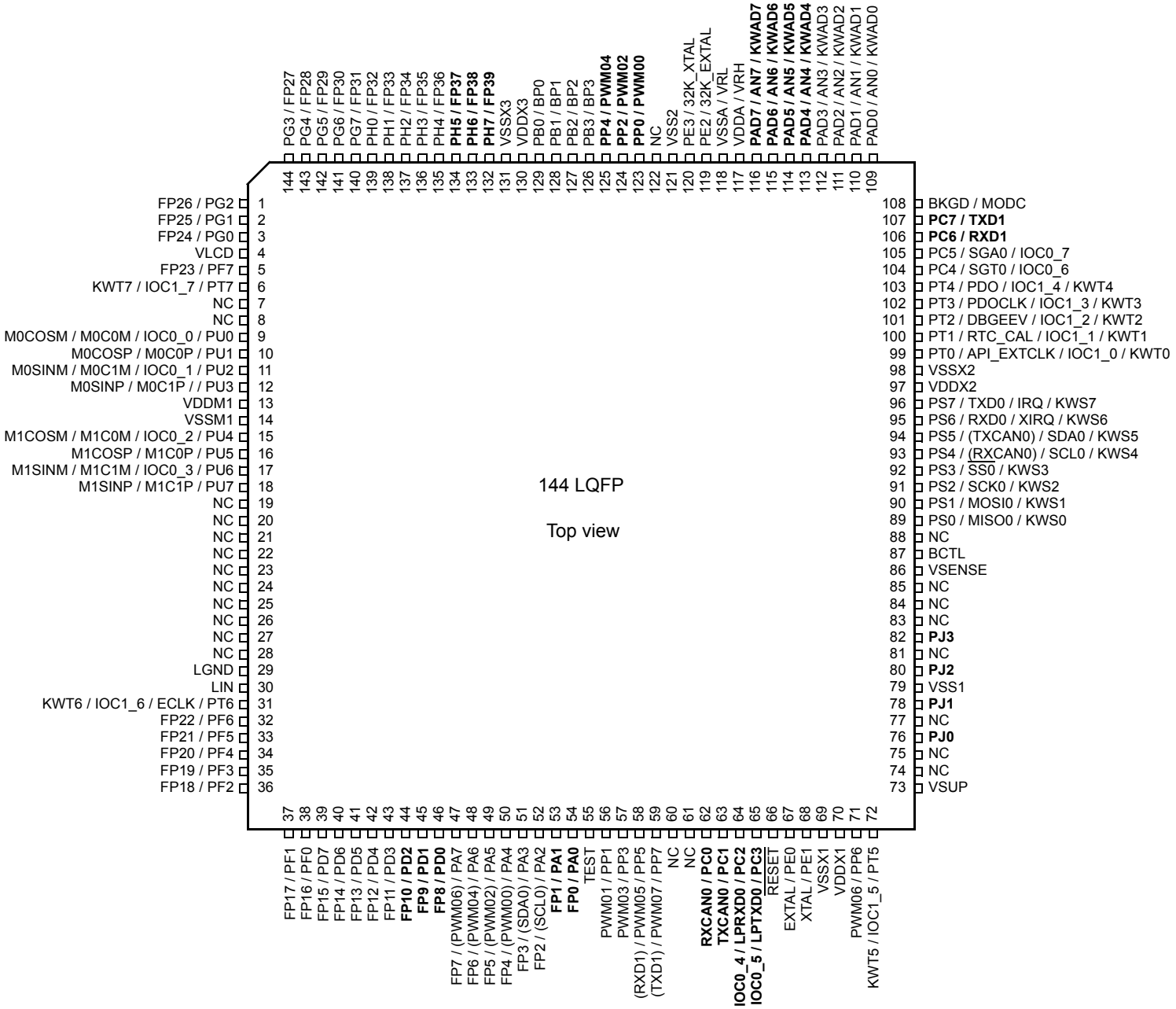


Figure 1-4. MC9S12ZVHY/MC9S12ZVHL Families 144-pin LQFP pin out

Table 1-7. Pin Summary

LQFP Option			Function					Power Supply	Internal Pull Resistor	
144	100	Pin	1st Func.	2nd Func.	3rd Func.	4th Func.	5th Func.		CTRL	Reset State
44	—	PD2	FP10	—	—	—	—	VDDX	PERD/ PPSD	Pull Down
45	—	PD1	FP9	—	—	—	—	VDDX	PERD/ PPSD	Pull Down
46	—	PD0	FP8	—	—	—	—	VDDX	PERD/ PPSD	Pull Down
47	33	PA7	(PWM6)	FP7	—	—	—	VDDX	PERA/ PPSA	Pull Down
48	34	PA6	(PWM4)	FP6	—	—	—	VDDX	PERA/ PPSA	Pull Down
49	35	PA5	(PWM2)	FP5	—	—	—	VDDX	PERA/ PPSA	Pull Down
50	36	PA4	(PWM0)	FP4	—	—	—	VDDX	PERA/ PPSA	Pull Down
51	37	PA3	(SDA0)	FP3	—	—	—	VDDX	PERA/ PPSA	Pull Down
52	38	PA2	(SCL0)	FP2	—	—	—	VDDX	PERA/ PPSA	Pull Down
53	—	PA1	FP1	—	—	—	—	VDDX	PERA/ PPSA	Pull Down
54	—	PA0	FP0	—	—	—	—	VDDX	PERA/ PPSA	Pull Down
55	39	TEST	—	—	—	—	—	VDDX	—	—
56	40	PP1	PWM1	—	—	—	—	V_{DDX}	PERP/ PPSP	Disabled
57	41	PP3	PWM3	—	—	—	—	V_{DDX}	PERP/ PPSP	Disabled
58	42	PP5	PWM5	(RXD1)	—	—	—	V_{DDX}	PERP/ PPSP	Disabled
59	43	PP7	PWM7	(TXD1)	—	—	—	V_{DDX}	PERP/ PPSP	Disabled
60	—	—	—	—	—	—	—	—	—	—
61	—	—	—	—	—	—	—	—	—	—
62	—	PC0	RXCAN0	—	—	—	—	V_{DDX}	PERC/ PPSC	Disabled

The BDC serial interface uses a clocking scheme in which the external host generates a falling edge on the BKGD pin to indicate the start of each bit time. This falling edge is sent for every bit whether data is transmitted or received. Data is transferred most significant bit (MSB) first at 16 target clock cycles per bit. The interface times out if during a command 512 clock cycles occur between falling edges from the host. The timeout forces the current command to be discarded.

The BKGD pin is a pseudo open-drain pin and has a weak on-chip active pull-up that is enabled at all times. It is assumed that there is an external pull-up and that drivers connected to BKGD do not typically drive the high level. Since R-C rise time could be unacceptably long, the target system and host provide brief drive-high (speedup) pulses to drive BKGD to a logic 1. The source of this speedup pulse is the host for transmit cases and the target for receive cases.

The timing for host-to-target is shown in Figure 5-6 and that of target-to-host in Figure 5-7 and Figure 5-8. All cases begin when the host drives the BKGD pin low to generate a falling edge. Since the host and target operate from separate clocks, it can take the target up to one full clock cycle to recognize this edge; this synchronization uncertainty is illustrated in Figure 5-6. The target measures delays from this perceived start of the bit time while the host measures delays from the point it actually drove BKGD low to start the bit up to one target clock cycle earlier. Synchronization between the host and target is established in this manner at the start of every bit time.

Figure 5-6 shows an external host transmitting a logic 1 and transmitting a logic 0 to the BKGD pin of a target system. The host is asynchronous to the target, so there is up to a one clock-cycle delay from the host-generated falling edge to where the target recognizes this edge as the beginning of the bit time. Ten target clock cycles later, the target senses the bit level on the BKGD pin. Internal glitch detect logic requires the pin be driven high no later than eight target clock cycles after the falling edge for a logic 1 transmission.

Since the host drives the high speedup pulses in these two cases, the rising edges look like digitally driven signals.

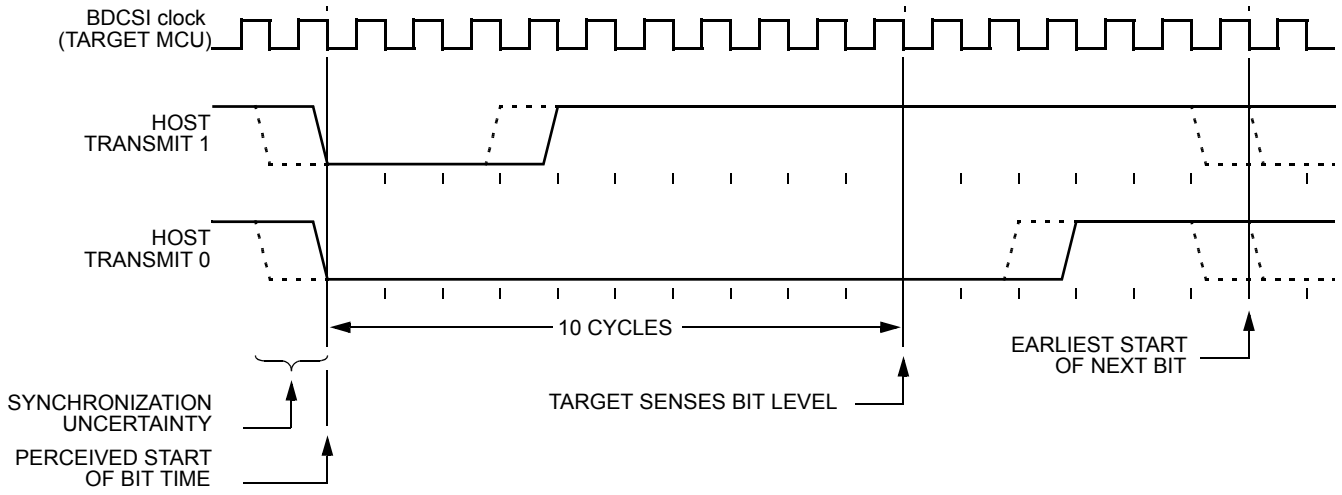


Figure 5-6. BDC Host-to-Target Serial Bit Timing

Figure 5-7 shows the host receiving a logic 1 from the target system. The host holds the BKGD pin low long enough for the target to recognize it (at least two target clock cycles). The host must release the low

6.3.2.3 Debug Trace Control Register High (DBGTCRH)

Address: 0x0102

	7	6	5	4	3	2	1	0
R W	reserved	TSOURCE	TRANGE		TRCMOD		TALIGN	
Reset	0	0	0	0	0	0	0	0

Figure 6-5. Debug Trace Control Register (DBGTCRH)

Read: Anytime.

Write: Anytime the module is disarmed and PTACT is clear.

WARNING

DBGTCR[7] is reserved. Setting this bit maps the tracing to an unimplemented bus, thus preventing proper operation.

This register configures the trace buffer for tracing and profiling.

Table 6-8. DBGTCRH Field Descriptions

Field	Description
6 TSOURCE	**Trace Control Bits** — The TSOURCE enables the tracing session. 0 No CPU tracing/profiling selected 1 CPU tracing/profiling selected
5–4 TRANGE	**Trace Range Bits** — The TRANGE bits allow filtering of trace information from a selected address range when tracing from the CPU in Detail mode. These bits have no effect in other tracing modes. To use a comparator for range filtering, the corresponding COMPE bit must remain cleared. If the COMPE bit is set then the comparator is used to generate events and the TRANGE bits have no effect. See Table 6-9 for range boundary definition.
3–2 TRCMOD	**Trace Mode Bits** — See Section 6.4.5.2 for detailed Trace Mode descriptions. In Normal Mode, change of flow information is stored. In Loop1 Mode, change of flow information is stored but redundant entries into trace memory are inhibited. In Detail Mode, address and data for all memory and register accesses is stored. See Table 6-10.
1–0 TALIGN	**Trigger Align Bits** — These bits control whether the trigger is aligned to the beginning, end or the middle of a tracing or profiling session. See Table 6-11.

Table 6-9. TRANGE Trace Range Encoding

TRANGE	Tracing Range
00	Trace from all addresses (No filter)
01	Trace only in address range from $00000 to Comparator D
10	Trace only in address range from Comparator C to $FFFFFF
11	Trace only in range from Comparator C to Comparator D

Table 6-10. TRCMOD Trace Mode Bit Encoding

TRCMOD	Description
00	Normal
01	Loop1

Write: If DBG not armed and PTACT is clear.

This register can be accessed with a byte resolution, whereby DBGADM0, DBGADM1, DBGADM2, DBGADM3 map to DBGADM[31:0] respectively.

Table 6-29. DBGADM Field Descriptions

Field	Description
31–16 Bits[31:16] (DBGADM0, DBGADM1)	**Comparator Data Mask Bits** — These bits control whether the comparator compares the data bus bits to the corresponding comparator data compare bits. 0 Do not compare corresponding data bit 1 Compare corresponding data bit
15-0 Bits[15:0] (DBGADM2, DBGADM3)	**Comparator Data Mask Bits** — These bits control whether the comparator compares the data bus bits to the corresponding comparator data compare bits. 0 Do not compare corresponding data bit 1 Compare corresponding data bit

6.3.2.16 Debug Comparator B Control Register (DBGBCTL)

Address: 0x0120

	7	6	5	4	3	2	1	0
R	0	0	INST	0	RW	RWE	reserved	COMPE
W								
Reset	0	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 6-18. Debug Comparator B Control Register

Read: Anytime.

Write: If DBG not armed and PTACT is clear.

Table 6-30. DBGBCTL Field Descriptions

Field[1]	Description
5 INST	**Instruction Select** — This bit configures the comparator to compare PC or data access addresses. 0 Comparator compares addresses of data accesses 1 Comparator compares PC address
3 RW	**Read/Write Comparator Value Bit** — The RW bit controls whether read or write is used in compare for the associated comparator. The RW bit is ignored if RWE is clear or INST is set. 0 Write cycle is matched 1 Read cycle is matched
2 RWE	**Read/Write Enable Bit** — The RWE bit controls whether read or write comparison is enabled for the associated comparator. This bit is ignored when INST is set. 0 Read/Write is not used in comparison 1 Read/Write is used in comparison
0 COMPE	**Enable Bit** — Determines if comparator is enabled 0 The comparator is not enabled 1 The comparator is enabled

1. If the ABCM field selects range mode comparisons, then DBGACTL bits configure the comparison, DBGBCTL is ignored.

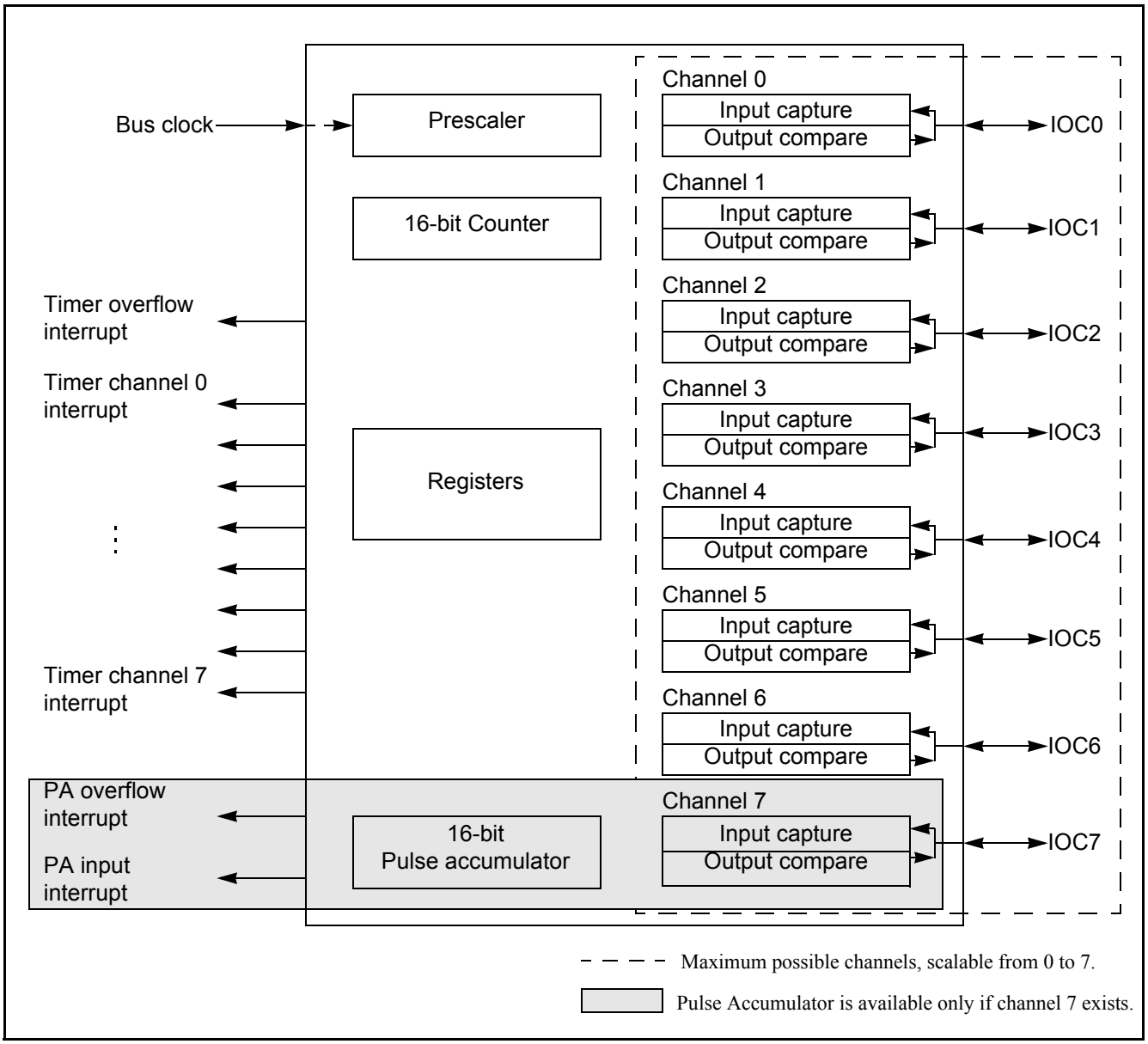


Figure 8-1. TIM16B8CV3 Block Diagram

8.3.2 Register Descriptions

This section consists of register descriptions in address order. Each description includes a standard register diagram with an associated figure number. Details of register bit and field function follow the register diagrams, in bit order.

Only bits related to implemented channels are valid.

Register Name		Bit 7	6	5	4	3	2	1	Bit 0
0x0000 TIOS	R	IOS7	IOS6	IOS5	IOS4	IOS3	IOS2	IOS1	IOS0
	W								
0x0001 CFORC	R	0	0	0	0	0	0	0	0
	W	FOC7	FOC6	FOC5	FOC4	FOC3	FOC2	FOC1	FOC0
0x0002 OC7M	R	OC7M7	OC7M6	OC7M5	OC7M4	OC7M3	OC7M2	OC7M1	OC7M0
	W								
0x0003 OC7D	R	OC7D7	OC7D6	OC7D5	OC7D4	OC7D3	OC7D2	OC7D1	OC7D0
	W								
0x0004 TCNTH	R	TCNT15	TCNT14	TCNT13	TCNT12	TCNT11	TCNT10	TCNT9	TCNT8
	W								
0x0005 TCNTL	R	TCNT7	TCNT6	TCNT5	TCNT4	TCNT3	TCNT2	TCNT1	TCNT0
	W								
0x0006 TSCR1	R	TEN	TSWAI	TSFRZ	TFFCA	PRNT	0	0	0
	W								
0x0007 TTOV	R	TOV7	TOV6	TOV5	TOV4	TOV3	TOV2	TOV1	TOV0
	W								
0x0008 TCTL1	R	OM7	OL7	OM6	OL6	OM5	OL5	OM4	OL4
	W								
0x0009 TCTL2	R	OM3	OL3	OM2	OL2	OM1	OL1	OM0	OL0
	W								
0x000A TCTL3	R	EDG7B	EDG7A	EDG6B	EDG6A	EDG5B	EDG5A	EDG4B	EDG4A
	W								
0x000B TCTL4	R	EDG3B	EDG3A	EDG2B	EDG2A	EDG1B	EDG1A	EDG0B	EDG0A
	W								
0x000C TIE	R	C7I	C6I	C5I	C4I	C3I	C2I	C1I	C0I
	W								
0x000D TSCR2	R	TOI	0	0	0	TCRE	PR2	PR1	PR0
	W								
0x000E TFLG1	R	C7F	C6F	C5F	C4F	C3F	C2F	C1F	C0F
	W								
0x000F TFLG2	R	TOF	0	0	0	0	0	0	0
	W								
0x0010–0x001F TCxH–TCxL[1]	R	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8
	W								
	R	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	W								
0x0020 PACTL	R	0	PAEN	PAMOD	PEDGE	CLK1	CLK0	PAOVI	PAI
	W								

Figure 8-5. TIM16B8CV3 Register Summary (Sheet 1 of 2)

NOTE

The newly selected prescale factor will not take effect until the next synchronized edge where all prescale counter stages equal zero.

8.3.2.12 Main Timer Interrupt Flag 1 (TFLG1)

Module Base + 0x000E

	7	6	5	4	3	2	1	0
R W	C7F	C6F	C5F	C4F	C3F	C2F	C1F	C0F
Reset	0	0	0	0	0	0	0	0

Figure 8-20. Main Timer Interrupt Flag 1 (TFLG1)

Read: Anytime

Write: Used in the clearing mechanism (set bits cause corresponding bits to be cleared). Writing a zero will not affect current status of the bit.

Table 8-16. TRLG1 Field Descriptions

Note: Writing to unavailable bits has no effect. Reading from unavailable bits return a zero.

Field	Description
7:0 C[7:0]F	**Input Capture/Output Compare Channel "x" Flag** — These flags are set when an input capture or output compare event occurs. Clearing requires writing a one to the corresponding flag bit while TEN or PAEN is set to one. **Note:** When TFFCA bit in TSCR register is set, a read from an input capture or a write into an output compare channel (0x0010–0x001F) will cause the corresponding channel flag CxF to be cleared.

8.3.2.13 Main Timer Interrupt Flag 2 (TFLG2)

Module Base + 0x000F

	7	6	5	4	3	2	1	0
R	TOF	0	0	0	0	0	0	0
W								
Reset	0	0	0	0	0	0	0	0

Unimplemented or Reserved

Figure 8-21. Main Timer Interrupt Flag 2 (TFLG2)

TFLG2 indicates when interrupt conditions have occurred. To clear a bit in the flag register, write the bit to one while TEN bit of TSCR1 or PAEN bit of PACTL is set to one.

Read: Anytime

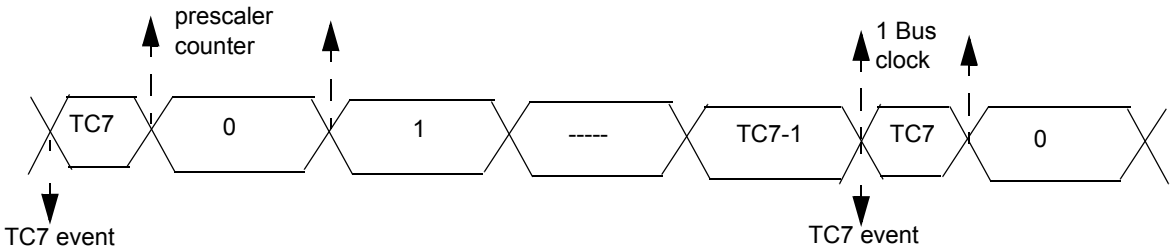
Write: Used in clearing mechanism (set bits cause corresponding bits to be cleared).

Writing to the timer port bit of an output compare pin does not affect the pin state. The value written is stored in an internal latch. When the pin becomes available for general-purpose output, the last value written to the bit appears at the pin.

When TCRE is set and TC7 is not equal to 0, then TCNT will cycle from 0 to TC7. When TCNT reaches TC7 value, it will last only one Bus cycle then reset to 0.

Note: in Figure 8-31,if PR[2:0] is equal to 0, one prescaler counter equal to one Bus clock

Figure 8-31. The TCNT cycle diagram under TCRE=1 condition



8.4.3.1 OC Channel Initialization

The internal register whose output drives OCx can be programmed before the timer drives OCx. The desired state can be programmed to this internal register by writing a one to CFORCx bit with TIOSx, OCPDx and TEN bits set to one.

Set OCx: Write a 1 to FOCx while TEN=1, IOSx=1, OMx=1, OLx=1 and OCPDx=1

Clear OCx: Write a 1 to FOCx while TEN=1, IOSx=1, OMx=1, OLx=0 and OCPDx=1

Setting OCPDx to zero allows the internal register to drive the programmed state to OCx. This allows a glitch free switch over of port from general purpose I/O to timer output once the OCPDx bit is set to zero.

8.4.4 Pulse Accumulator

The pulse accumulator (PACNT) is a 16-bit counter that can operate in two modes:

Event counter mode — Counting edges of selected polarity on the pulse accumulator input pin, PAI.

Gated time accumulation mode — Counting pulses from a divide-by-64 clock. The PAMOD bit selects the mode of operation.

The minimum pulse width for the PAI input is greater than two Bus clocks.

Module Base + 0x000E — Access: User read/write(1)

	7	6	5	4	3	2	1	0
R	RXERR7	RXERR6	RXERR5	RXERR4	RXERR3	RXERR2	RXERR1	RXERR0
W								
Reset:	0	0	0	0	0	0	0	0

= Unimplemented

Figure 11-18. MSCAN Receive Error Counter (CANRXERR)

1. Read: Only when in sleep mode (SLPRQ = 1 and SLPAK = 1) or initialization mode (INITRQ = 1 and INITAK = 1)
   Write: Unimplemented

NOTE

Reading this register when in any other mode other than sleep or initialization mode may return an incorrect value. For MCUs with dual CPUs, this may result in a CPU fault condition.

11.3.2.16 MSCAN Transmit Error Counter (CANTXERR)

This register reflects the status of the MSCAN transmit error counter.

Module Base + 0x000F — Access: User read/write(1)

	7	6	5	4	3	2	1	0
R	TXERR7	TXERR6	TXERR5	TXERR4	TXERR3	TXERR2	TXERR1	TXERR0
W								
Reset:	0	0	0	0	0	0	0	0

= Unimplemented

Figure 11-19. MSCAN Transmit Error Counter (CANTXERR)

1. Read: Only when in sleep mode (SLPRQ = 1 and SLPAK = 1) or initialization mode (INITRQ = 1 and INITAK = 1)
   Write: Unimplemented

NOTE

Reading this register when in any other mode other than sleep or initialization mode, may return an incorrect value. For MCUs with dual CPUs, this may result in a CPU fault condition.

The MSCAN facilitates a sophisticated message storage system which addresses the requirements of a broad range of network applications.

11.4.2.1 Message Transmit Background

Modern application layer software is built upon two fundamental assumptions:

- Any CAN node is able to send out a stream of scheduled messages without releasing the CAN bus between the two messages. Such nodes arbitrate for the CAN bus immediately after sending the previous message and only release the CAN bus in case of lost arbitration.
- The internal message queue within any CAN node is organized such that the highest priority message is sent out first, if more than one message is ready to be sent.

The behavior described in the bullets above cannot be achieved with a single transmit buffer. That buffer must be reloaded immediately after the previous message is sent. This loading process lasts a finite amount of time and must be completed within the inter-frame sequence (IFS) to be able to send an uninterrupted stream of messages. Even if this is feasible for limited CAN bus speeds, it requires that the CPU reacts with short latencies to the transmit interrupt.

A double buffer scheme de-couples the reloading of the transmit buffer from the actual message sending and, therefore, reduces the reactiveness requirements of the CPU. Problems can arise if the sending of a message is finished while the CPU re-loads the second buffer. No buffer would then be ready for transmission, and the CAN bus would be released.

At least three transmit buffers are required to meet the first of the above requirements under all circumstances. The MSCAN has three transmit buffers.

The second requirement calls for some sort of internal prioritization which the MSCAN implements with the "local priority" concept described in Section 11.4.2.2, "Transmit Structures."

11.4.2.2 Transmit Structures

The MSCAN triple transmit buffer scheme optimizes real-time performance by allowing multiple messages to be set up in advance. The three buffers are arranged as shown in Figure 11-39.

All three buffers have a 13-byte data structure similar to the outline of the receive buffers (see Section 11.3.3, "Programmer's Model of Message Storage"). An additional Transmit Buffer Priority Register (TBPR) contains an 8-bit local priority field (PRIO) (see Section 11.3.3.4, "Transmit Buffer Priority Register (TBPR)"). The remaining two bytes are used for time stamping of a message, if required (see Section 11.3.3.5, "Time Stamp Register (TSRH–TSRL)").

To transmit a message, the CPU must identify an available transmit buffer, which is indicated by a set transmitter buffer empty (TXEx) flag (see Section 11.3.2.7, "MSCAN Transmitter Flag Register (CANTFLG)"). If a transmit buffer is available, the CPU must set a pointer to this buffer by writing to the CANTBSEL register (see Section 11.3.2.11, "MSCAN Transmit Buffer Selection Register (CANTBSEL)"). This makes the respective buffer accessible within the CANTXFG address space (see Section 11.3.3, "Programmer's Model of Message Storage"). The algorithmic feature associated with the CANTBSEL register simplifies the transmit buffer selection. In addition, this scheme makes the handler

Chapter 14
Inter-Integrated Circuit (IICV3) Block Description

Table 14-1. Revision History

Revision Number	Revision Date	Sections Affected	Description of Changes
V01.03	28 Jul 2006	14.7.1.7/14-565	- Update flow-chart of interrupt routine for 10-bit address
V01.04	17 Nov 2006	14.3.1.2/14-545	- Revise Table1-5
V01.05	14 Aug 2007	14.3.1.1/14-545	- Backward compatible for IBAD bit name

14.1 Introduction

The inter-IC bus (IIC) is a two-wire, bidirectional serial bus that provides a simple, efficient method of data exchange between devices. Being a two-wire device, the IIC bus minimizes the need for large numbers of connections between devices, and eliminates the need for an address decoder.

This bus is suitable for applications requiring occasional communications over a short distance between a number of devices. It also provides flexibility, allowing additional devices to be connected to the bus for further expansion and system development.

The interface is designed to operate up to 100 kbps with maximum bus loading and timing. The device is capable of operating at higher baud rates, up to a maximum of clock/20, with reduced bus loading. The maximum communication length and the number of devices that can be connected are limited by a maximum bus capacitance of 400 pF.

14.1.1 Features

The IIC module has the following key features:

- Compatible with I2C bus standard
- Multi-master operation
- Software programmable for one of 256 different serial clock frequencies
- Software selectable acknowledge bit
- Interrupt driven byte-by-byte data transfer
- Arbitration lost interrupt with automatic mode switching from master to slave
- Calling address identification interrupt
- Start and stop signal generation/detection
- Repeated start signal generation

14.3.1.1 IIC Address Register (IBAD)

Module Base +0x0000

	7	6	5	4	3	2	1	0
R	ADR7	ADR6	ADR5	ADR4	ADR3	ADR2	ADR1	0
W								
Reset	0	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 14-3. IIC Bus Address Register (IBAD)

Read and write anytime

This register contains the address the IIC bus will respond to when addressed as a slave; note that it is not the address sent on the bus during the address transfer.

Table 14-2. IBAD Field Descriptions

Field	Description
7:1 ADR[7:1]	**Slave Address** — Bit 1 to bit 7 contain the specific slave address to be used by the IIC bus module.The default mode of IIC bus is slave mode for an address match on the bus.
0 Reserved	Reserved — Bit 0 of the IBAD is reserved for future compatibility. This bit will always read 0.

14.3.1.2 IIC Frequency Divider Register (IBFD)

Module Base + 0x0001

	7	6	5	4	3	2	1	0
R	IBC7	IBC6	IBC5	IBC4	IBC3	IBC2	IBC1	IBC0
W								
Reset	0	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 14-4. IIC Bus Frequency Divider Register (IBFD)

Read and write anytime

Table 14-3. IBFD Field Descriptions

Field	Description
7:0 IBC[7:0]	**I Bus Clock Rate 7:0** — This field is used to prescale the clock for bit rate selection. The bit clock generator is implemented as a prescale divider — IBC7:6, prescaled shift register — IBC5:3 select the prescaler divider and IBC2-0 select the shift register tap point. The IBC bits are decoded to give the tap and prescale values as shown in Table 14-4.

from where was during the previous transmission. It is not possible for the IIC to wake up the CPU when its internal clocks are stopped.

If it were the case that the IBSWAI bit was cleared when the WAI instruction was executed, the IIC internal clocks and interface would remain alive, continuing the operation which was currently underway. It is also possible to configure the IIC such that it will wake up the CPU via an interrupt at the conclusion of the current operation. See the discussion on the IBIF and IBIE bits in the IBSR and IBCR, respectively.

14.3.1.4 IIC Status Register (IBSR)

Module Base + 0x0003

	7	6	5	4	3	2	1	0
R	TCF	IAAS	IBB	IBAL	0	SRW	IBIF	RXAK
W								
Reset	1	0	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 14-7. IIC Bus Status Register (IBSR)

This status register is read-only with exception of bit 1 (IBIF) and bit 4 (IBAL), which are software clearable.

Table 14-9. IBSR Field Descriptions

Field	Description
7 TCF	**Data Transferring Bit** — While one byte of data is being transferred, this bit is cleared. It is set by the falling edge of the 9th clock of a byte transfer. Note that this bit is only valid during or immediately following a transfer to the IIC module or from the IIC module. 0 Transfer in progress 1 Transfer complete
6 IAAS	**Addressed as a Slave Bit** — When its own specific address (I-bus address register) is matched with the calling address or it receives the general call address with GCEN== 1,this bit is set.The CPU is interrupted provided the IBIE is set.Then the CPU needs to check the SRW bit and set its Tx/$\overline{\text{Rx}}$ mode accordingly.Writing to the I-bus control register clears this bit. 0 Not addressed 1 Addressed as a slave
5 IBB	Bus Busy Bit 0 This bit indicates the status of the bus. When a START signal is detected, the IBB is set. If a STOP signal is detected, IBB is cleared and the bus enters idle state. 1 Bus is busy
4 IBAL	**Arbitration Lost** — The arbitration lost bit (IBAL) is set by hardware when the arbitration procedure is lost. Arbitration is lost in the following circumstances: 1. SDA sampled low when the master drives a high during an address or data transmit cycle. 2. SDA sampled low when the master drives a high during the acknowledge bit of a data receive cycle. 3. A start cycle is attempted when the bus is busy. 4. A repeated start cycle is requested in slave mode. 5. A stop condition is detected when the master did not request it. This bit must be cleared by software, by writing a one to it. A write of 0 has no effect on this bit.

16.4.2 PWM Duty Cycle

The PWM duty cycle for the motor controller channel x can be determined by dividing the decimal representation of bits D[10:0] in MCDCx by the decimal representation of the bits P[10:0] in MCPER and multiplying the result by 100% as shown in the equation below:

$$\text{Effective PWM Channel X \% Duty Cycle} = \frac{\text{DUTY}}{\text{MCPER}} \cdot 100\%$$

NOTE

x = PWM Channel Number = 0, 1, 2, 3 ... 8. This equation is only valid if DUTY <= MCPER and MCPER is not equal to 0.

Whenever D[10:0] >= P[10:0], a constant low level (RECIRC = 0) or high level (RECIRC = 1) will be output.

16.4.3 Motor Controller Counter Clock Source

Figure 16-22 shows how the PWM motor controller timer counter clock source is selected.

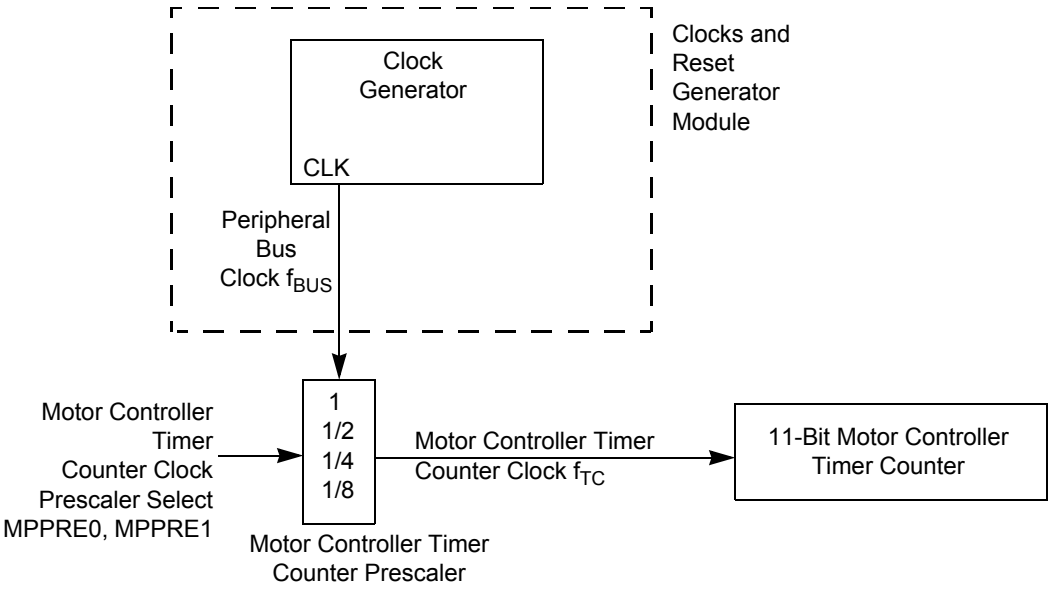


Figure 16-22. Motor Controller Counter Clock Selection

The peripheral bus clock is the source for the motor controller counter prescaler. The motor controller counter clock rate, f_{TC}, is set by selecting the appropriate prescaler value. The prescaler is selected with the MCPRE[1:0] bits in motor controller control register 0 (MCCTL0). The motor controller channel frequency of operation can be calculated using the following formula if DITH = 0:

$$\text{Motor Channel Frequency (Hz)} = \frac{f_{TC}}{\text{MCPER} \cdot \text{M}}$$

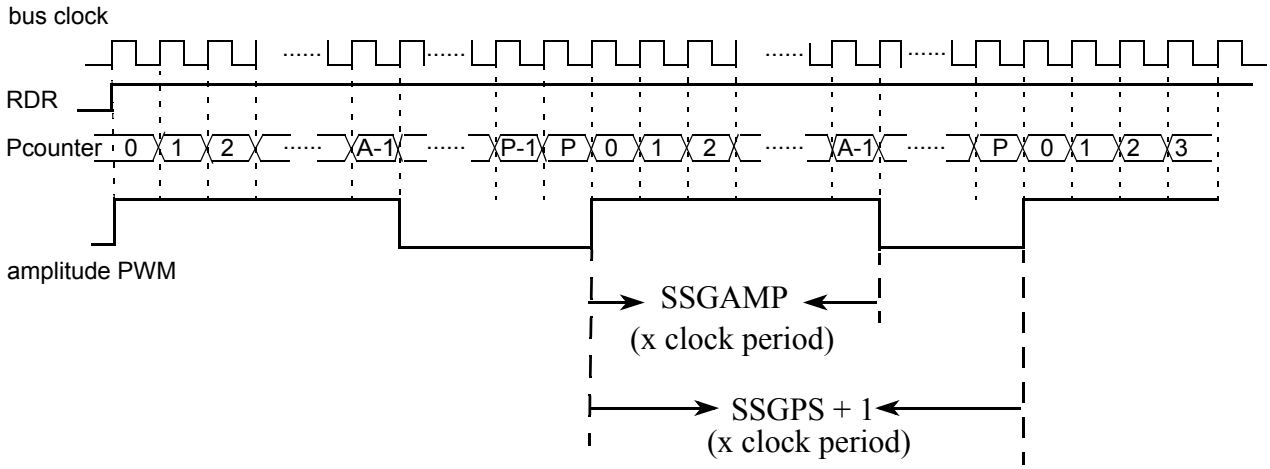
Table 19-15. SSGDCNT Field Descriptions

Field	Description
7–0 DCNT[7:0]	**SSG Tone Duration Counter bits —** The counter register is read only, It contains the tone cycle number. The counter will be reset to 0 when it reaches the SSGDUR value. In any SSG stop case, the counter will also be reset to 0.

19.4 Functional Description

19.4.1 SSG Amplitude Generation

The SSG sound amplitude is presented as the duty cycle of PWM signal. The prescaler signal period is the period of the PWM and the amplitude register SSGAMP contains the duty, the amplitude duty cycle is SSGAMP/(SSGPS+1), when SSGAMP>=SSGPS+1, it will be 100%. SSGAMP has a buffer register, so changing the SSGAMP will not disturb the amplitude PWM waveform before the synchronous reload takes place.



where : RDR is the RDR bit of SSGCR

Pcounter is the prescaler counter

A is value of SSGAMP buffer

P is the value of SSGPS buffer

Figure 19-21. SSG Amplitude Generation

23.5 Application Information

23.5.1 Module Initialization

The following steps should be used to configure the module before starting the transmission:

1. Set the slew rate in the LPSLRM register to the desired transmission baud rate.
2. When using the LIN Physical Layer for other purposes than LIN transmission, de-activate the dominant timeout feature in the LPSLRM register if needed.
3. In most cases, the internal pullup should be enabled in the LPCR register.
4. Route the desired source in the PIM module to the LIN Physical Layer.
5. Select the transmit mode (Receive only mode or Normal mode) in the LPCR register.
6. If the SCI is selected as source, activate the wake-up feature in the LPCR register if needed for the application (SCI active edge interrupt must also be enabled).
7. Enable the LIN Physical Layer in the LPCR register.
8. Wait for a minimum of a transmit bit.
9. Begin transmission if needed.

NOTE

It is not allowed to try to clear LPOCIF or LPDTIF if they are already cleared. Before trying to clear an error flag, always make sure that it is already set.

23.5.2 Interrupt handling in Interrupt Service Routine (ISR)

Both interrupts (TxD-dominant timeout and overcurrent) represent a failure in transmission. To avoid more disturbances on the transmission line, the transmitter is de-activated in both cases. The interrupt subroutine must take care of clearing the error condition and starting the routine that re-enables the transmission. For that purpose, the following steps are recommended:

1. First, the cause of the interrupt must be cleared:
   - The overcurrent will be gone after the transmitter has been disabled.
   - The TxD-dominant timeout condition will be gone once the selected source for LPTxD has turned recessive.
2. Clear the corresponding enable bit (LPDTIE or LPOCIE) to avoid entering the ISR again until the flags are cleared.
3. Notify the application of the error condition (LIN Error handler) and leave the ISR.

In the LIN Error handler, the following sequence is recommended:

1. Disable the LIN Physical Layer (LPCR) while re-configuring the transmission.
   - If the receiver must remain enabled, set the LIN Physical Layer into receive only mode instead.
2. Do all required configurations (SCI, etc.) to re-enable the transmission.
3. Wait for a transmit bit (this is needed to successfully re-enable the transmitter).

Appendix O
Package Information

0x0480–0x04A7 Pulse-Width-Modulator (PWM)

0x0498	PWMPER4	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x0499	PWMPER5	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x049A	PWMPER6	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x049B	PWMPER7	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x049C	PWMDTY0	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x049D	PWMDTY1	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x049E	PWMDTY2	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x049F	PWMDTY3	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x04A0	PWMDTY4	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x04A1	PWMDTY5	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x04A2	PWMDTY6	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x04A3	PWMDTY7	R W	Bit 7	6	5	4	3	2	1	Bit 0
0x04A4- 0x04A7	Reserved	R W	0	0	0	0	0	0	0	0

0x04A8–0x05BF Reserved

Address	Name		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x04A8- 0x05BF	Reserved	R W	0	0	0	0	0	0	0	0

0x05C0–0x05EF Timer Module (TIM0)

Address	Name		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0x05C0	TIM0TIOS	R W	IOS7	IOS6	IOS5	IOS4	IOS3	IOS2	IOS1	IOS0