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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	4MHz
Connectivity	-
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	13
Program Memory Size	1.75KB (1K x 14)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	4V ~ 5.5V
Data Converters	A/D 4x8b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	18-SOIC (0.295", 7.50mm Width)
Supplier Device Package	18-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16c712-04i-so

PIC16C712/716

2.2 Data Memory Organization

The data memory is partitioned into multiple banks which contain the General Purpose Registers and the Special Function Registers. Bits RP1 and RP0 are the bank select bits.

RP1 ⁽¹⁾	RP0	(STATUS<6:5>)
--------------------	-----	---------------

- = 00 → Bank 0
- = 01 → Bank 1
- = 10 → Bank 2 (not implemented)
- = 11 → Bank 3 (not implemented)

Note 1: Maintain this bit clear to ensure upward compatibility with future products.

Each bank extends up to 7Fh (128 bytes). The lower locations of each bank are reserved for the Special Function Registers. Above the Special Function Registers are General Purpose Registers, implemented as static RAM. All implemented banks contain Special Function Registers. Some “high use” Special Function Registers from one bank may be mirrored in another bank for code reduction and quicker access.

2.2.1 GENERAL PURPOSE REGISTER FILE

The register file can be accessed either directly, or indirectly through the File Select Register FSR (see **Section 2.5 “Indirect Addressing, INDF and FSR Registers”**).

FIGURE 2-3: REGISTER FILE MAP

File Address			File Address
00h	INDF ⁽¹⁾	INDF ⁽¹⁾	80h
01h	TMR0	OPTION_REG	81h
02h	PCL	PCL	82h
03h	STATUS	STATUS	83h
04h	FSR	FSR	84h
05h	PORTA	TRISA	85h
06h	PORTB	TRISB	86h
07h	DATAACP	TRISCCP	87h
08h			88h
09h			89h
0Ah	PCLATH	PCLATH	8Ah
0Bh	INTCON	INTCON	8Bh
0Ch	PIR1	PIE1	8Ch
0Dh			8Dh
0Eh	TMR1L	PCON	8Eh
0Fh	TMR1H		8Fh
10h	T1CON		90h
11h	TMR2		91h
12h	T2CON	PR2	92h
13h			93h
14h			94h
15h	CCPR1L		95h
16h	CCPR1H		96h
17h	CCP1CON		97h
18h			98h
19h			99h
1Ah			9Ah
1Bh			9Bh
1Ch			9Ch
1Dh			9Dh
1Eh	ADRES		9Eh
1Fh	ADCON0	ADCON1	9Fh
20h	General Purpose Registers 96 Bytes	General Purpose Registers 32 Bytes	A0h
			BFh
			C0h
			FFh
7Fh			
Bank 0		Bank 1	

☐ Unimplemented data memory locations, read as '0'.

Note 1: Not a physical register.

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TABLE 2-1: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets (4)
Bank 1											
80h	INDF ⁽¹⁾	Addressing this location uses contents of FSR to address data memory (not a physical register)								0000 0000	0000 0000
81h	OPTION_ REG	$\overline{\text{RBPU}}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
82h	PCL ⁽¹⁾	Program Counter's (PC) Least Significant Byte								0000 0000	0000 0000
83h	STATUS ⁽¹⁾	IRP ⁽⁴⁾	RP1 ⁽⁴⁾	RP0	$\overline{\text{TO}}$	$\overline{\text{PD}}$	Z	DC	C	rr01 1xxx	rr0q quuu
84h	FSR ⁽¹⁾	Indirect Data Memory Address Pointer								xxxx xxxx	uuuu uuuu
85h	TRISA	—	—	— ⁽⁷⁾	PORTA Data Direction Register					--x1 1111	--x1 1111
86h	TRISB	PORTB Data Direction Register								1111 1111	1111 1111
87h	TRISCCP	— ⁽⁷⁾	— ⁽⁷⁾	— ⁽⁷⁾	— ⁽⁷⁾	— ⁽⁷⁾	TCCP	— ⁽⁷⁾	TT1CK	xxxx x1x1	xxxx x1x1
88h-89h	—	Unimplemented								—	—
8Ah	PCLATH ^(1,2)	—	—	—	Write Buffer for the upper 5 bits of the Program Counter					---0 0000	---0 0000
8Bh	INTCON ⁽¹⁾	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
8Ch	PIE1	—	ADIE	—	—	—	CCP1IE	TMR2IE	TMR1IE	-0-- -000	-0-- -000
8Dh	—	Unimplemented								—	—
8Eh	PCON	—	—	—	—	—	—	$\overline{\text{POR}}$	$\overline{\text{BOR}}$	---- --gq	---- --uu
8Fh-91h	—	Unimplemented								—	—
92h	PR2	Timer2 Period Register								1111 1111	1111 1111
93h-9Eh	—	Unimplemented								—	—
9Fh	ADCON1	—	—	—	—	—	PCFG2	PCFG1	PCFG0	---- -000	---- -000

Legend: x = unknown, u = unchanged, q = value depends on condition, — = unimplemented, read as '0'.
Shaded locations are unimplemented, read as '0'.

Note 1: These registers can be addressed from either bank.

2: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for PC<12:8> whose contents are transferred to the upper byte of the program counter.

3: Other (non Power-up) Resets include: external Reset through $\overline{\text{MCLR}}$ and the Watchdog Timer Reset.

4: The IRP and RP1 bits are reserved. Always maintain these bits clear.

5: On any device Reset, these pins are configured as inputs.

6: This is the value that will be in the port output latch.

7: Reserved bits; Do Not Use.

2.2.2.3 INTCON Register

The INTCON Register is a readable and writable register which contains various enable and flag bits for the TMR0 register overflow, RB Port change and External RB0/INT pin interrupts.

Note: Interrupt flag bits get set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit, GIE (INTCON<7>). User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

FIGURE 2-6: INTCON REGISTER (ADDRESS 0Bh, 8Bh)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x
GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF
bit7							bit0
<p>bit 7: GIE: Global Interrupt Enable bit 1 = Enables all unmasked interrupts 0 = Disables all interrupts</p> <p>bit 6: PEIE: Peripheral Interrupt Enable bit 1 = Enables all unmasked peripheral interrupts 0 = Disables all peripheral interrupts</p> <p>bit 5: TOIE: TMR0 Overflow Interrupt Enable bit 1 = Enables the TMR0 interrupt 0 = Disables the TMR0 interrupt</p> <p>bit 4: IINTE: RB0/INT External Interrupt Enable bit 1 = Enables the RB0/INT external interrupt 0 = Disables the RB0/INT external interrupt</p> <p>bit 3: RBIE: RB Port Change Interrupt Enable bit 1 = Enables the RB port change interrupt 0 = Disables the RB port change interrupt</p> <p>bit 2: TOIF: TMR0 Overflow Interrupt Flag bit 1 = TMR0 register has overflowed (must be cleared in software) 0 = TMR0 register did not overflow</p> <p>bit 1: INTF: RB0/INT External Interrupt Flag bit 1 = The RB0/INT external interrupt occurred (must be cleared in software) 0 = The RB0/INT external interrupt did not occur</p> <p>bit 0: RBIF: RB Port Change Interrupt Flag bit 1 = At least one of the RB7:RB4 pins changed state (must be cleared in software) 0 = None of the RB7:RB4 pins have changed state</p>							
<p>R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR Reset</p>							

PORTB pins RB3:RB1 are multiplexed with several peripheral functions (Table 3-3). PORTB pins RB3:RB0 have Schmitt Trigger input buffers.

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTB pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. Since the TRIS bit override is in effect while the peripheral is enabled, read-modify-write instructions (BSF, BCF, XORWF) with TRISB as destination should be avoided. The user should refer to the corresponding peripheral section for the correct TRIS bit settings.

Four of PORTB's pins, RB7:RB4, have an interrupt-on-change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB7:RB4 pin configured as an output is excluded from the interrupt-on-change comparison). The input pins, RB7:RB4, are compared with the old value latched on the last read of

PORTB. The “mismatch” outputs of RB7:RB4 are OR'ed together to generate the RB Port Change Interrupt with flag bit RBIF (INTCON<0>).

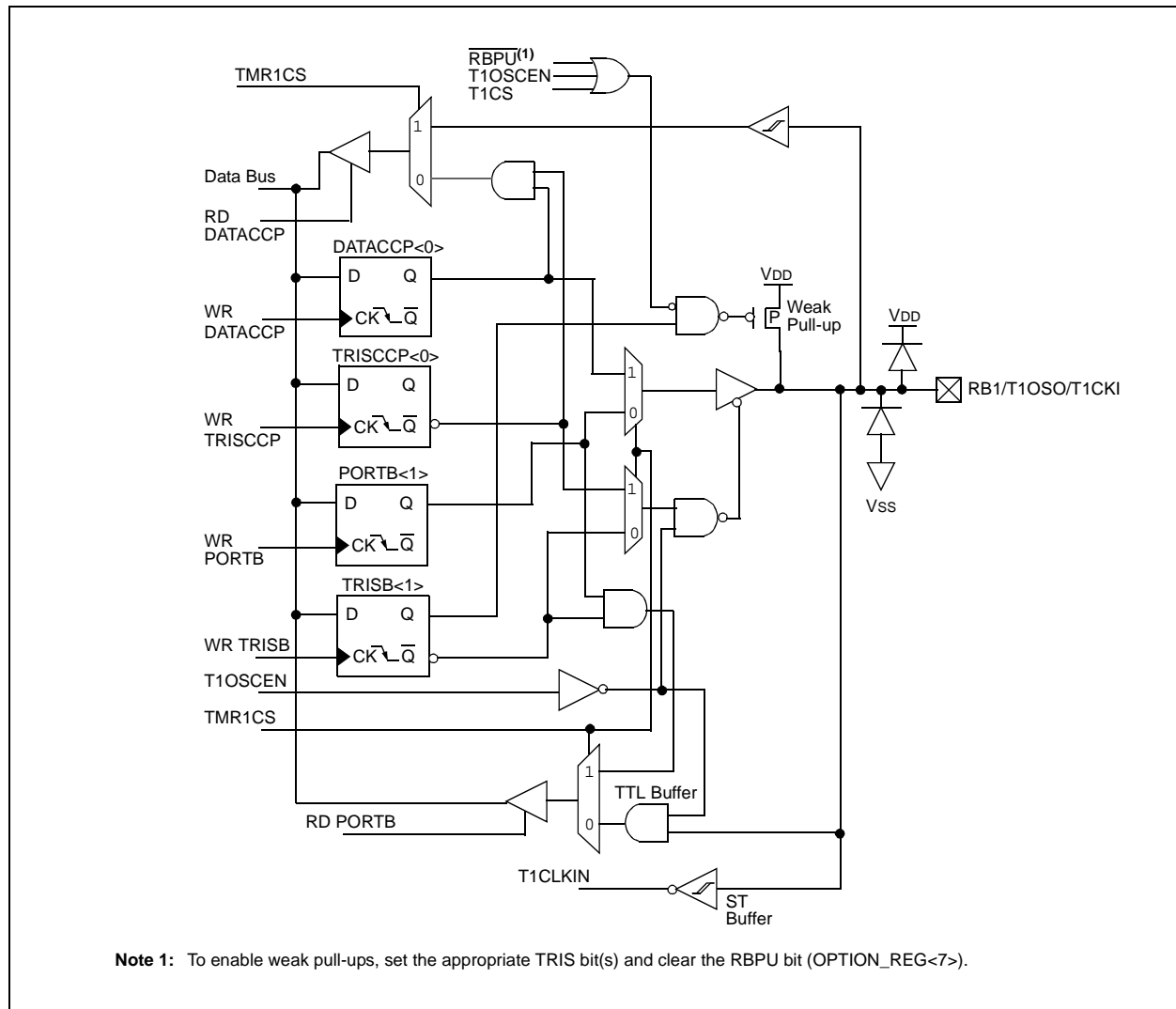
This interrupt can wake the device from Sleep. The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

- Any read or write of PORTB will end the mismatch condition.
- Clear flag bit RBIF.

A mismatch condition will continue to set flag bit RBIF. Reading PORTB will end the mismatch condition and allow flag bit RBIF to be cleared.

The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt-on-change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.

FIGURE 3-4: BLOCK DIAGRAM OF RB1/T1OSO/T1CKI PIN



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7.3 PWM Mode

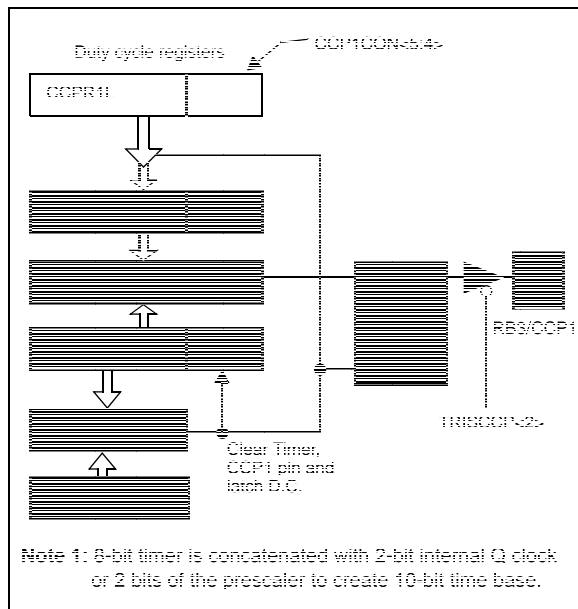
In Pulse Width Modulation (PWM) mode, the CCP1 pin produces up to a 10-bit resolution PWM output. Since the CCP1 pin is multiplexed with the PORTB data latch, the TRISCCP<2> bit must be cleared to make the CCP1 pin an output.

Note: Clearing the CCP1CON register will force the CCP1 PWM output latch to the default low level. This is neither the PORTB I/O data latch nor the DATAACP latch.

Figure 7-5 shows a simplified block diagram of the CCP module in PWM mode.

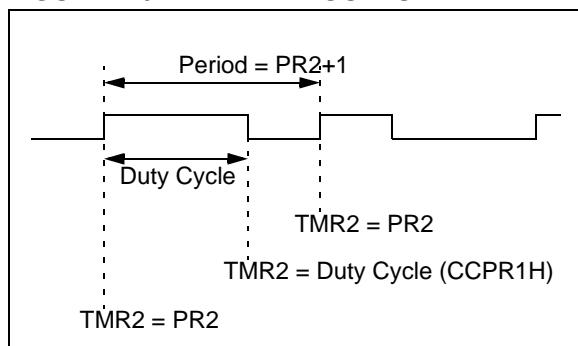
For a step by step procedure on how to set up the CCP module for PWM operation, see **Section 7.3.3 “Set-Up for PWM Operation”**.

FIGURE 7-5: SIMPLIFIED PWM BLOCK DIAGRAM



A PWM output (Figure 7-6) has a time base (period) and a time that the output stays high (duty cycle). The frequency of the PWM is the inverse of the period (1/period).

FIGURE 7-6: PWM OUTPUT



7.3.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula:

$$\text{PWM period} = [(PR2) + 1] \cdot 4 \cdot T_{osc} \cdot (\text{TMR2 prescale value})$$

PWM frequency is defined as $1 / [\text{PWM period}]$.

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCP1 pin is set (exception: if PWM duty cycle = 0%, the CCP1 pin will not be set)
- The PWM duty cycle is latched from CCPR1L into CCPR1H

Note: The Timer2 postscaler (see **Section 6.0 “Timer2 Module”**) is not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.

7.3.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPR1L register and to the CCP1CON<5:4> bits. Up to 10-bit resolution is available. The CCPR1L contains the eight MSBs and the CCP1CON<5:4> contains the two LSbs. This 10-bit value is represented by CCPR1L:CCP1CON<5:4>. The following equation is used to calculate the PWM duty cycle in time:

$$\text{PWM duty cycle} = (\text{CCPR1L:CCP1CON<5:4>}) \cdot T_{osc} \cdot (\text{TMR2 prescale value})$$

CCPR1L and CCP1CON<5:4> can be written to at any time, but the duty cycle value is not latched into CCPR1H until after a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPR1H is a read-only register.

The CCPR1H register and a 2-bit internal latch are used to double buffer the PWM duty cycle. This double buffering is essential for glitchless PWM operation.

When the CCPR1H and 2-bit latch match TMR2 concatenated with an internal 2-bit Q clock or 2 bits of the TMR2 prescaler, the CCP1 pin is cleared.

Maximum PWM resolution (bits) for a given PWM frequency:

$$= \frac{\log\left(\frac{F_{osc}}{F_{PWM}}\right)}{\log(2)} \text{ bits}$$

Note: If the PWM duty cycle value is longer than the PWM period the CCP1 pin will not be cleared.

For an example PWM period and duty cycle calculation, see the PIC® Mid-Range Reference Manual, (DS33023).

7.3.3 SET-UP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

1. Set the PWM period by writing to the PR2 register.
2. Set the PWM duty cycle by writing to the CCPR1L register and CCP1CON<5:4> bits.
3. Make the CCP1 pin an output by clearing the TRISCCP<2> bit.
4. Set the TMR2 prescale value and enable Timer2 by writing to T2CON.
5. Configure the CCP1 module for PWM operation.

TABLE 7-3: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 20 MHz

PWM Frequency	1.22 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescaler (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	5.5

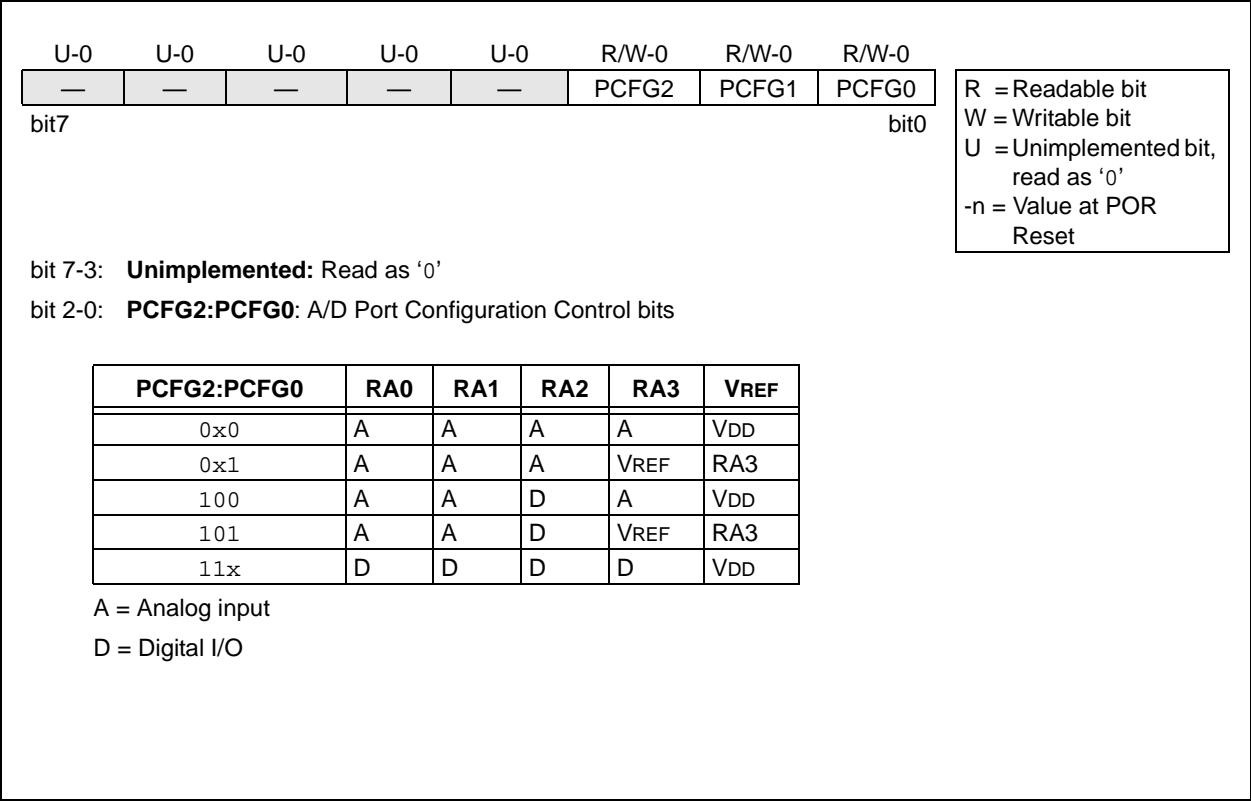
TABLE 7-4: REGISTERS ASSOCIATED WITH PWM AND TIMER2

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
07h	DATA CCP	—	—	—	—	—	DCCP	—	DT1CK	xxxx xxxx	xxxx xuxu
0Bh,8Bh	INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	—	ADIF	—	—	—	CCP1IF	TMR2IF	TMR1IF	-0-- -000	-0-- -000
11h	TMR2	Timer2 Module's Register								0000 0000	0000 0000
12h	T2CON	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	-000 0000
15h	CCPR1L	Capture/Compare/PWM Register 1 (LSB)								xxxx xxxx	uuuu uuuu
16h	CCPR1H	Capture/Compare/PWM Register 1 (MSB)								xxxx xxxx	uuuu uuuu
17h	CCP1CON	—	—	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	--00 0000	--00 0000
87h	TRISCCP	—	—	—	—	—	TCCP	—	TT1CK	xxxx x1x1	xxxx x1x1
8Ch	PIE1	—	ADIE	—	—	—	CCP1IE	TMR2IE	TMR1IE	-0-- -000	-0-- -000
92h	PR2	Timer2 Module's Period Register								1111 1111	1111 1111

Legend: x = unknown, u = unchanged, — = unimplemented read as '0'. Shaded cells are not used by PWM and Timer2.

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FIGURE 8-2: ADCON1 REGISTER (ADDRESS 9Fh)



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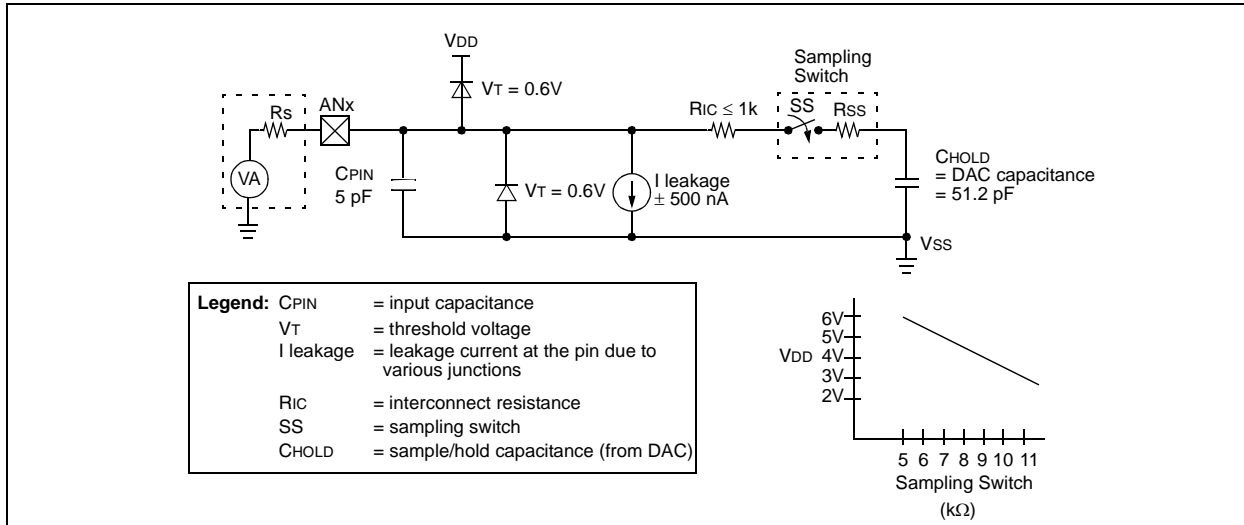
8.1 A/D Acquisition Requirements

For the A/D converter to meet its specified accuracy, the Charge Holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 8-4. The source impedance (R_s) and the internal sampling switch (R_{SS}) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (R_{SS}) impedance varies over the device voltage (V_{DD}). The source impedance affects the offset voltage at the analog input (due to pin leakage current). **The maximum recommended impedance for analog sources is 10 k Ω .** After the analog input channel is selected (changed) this acquisition must be done before the conversion can be started.

To calculate the minimum acquisition time, T_{ACQ} , see the PIC[®] Mid-Range Reference Manual, (DS33023). This equation calculates the acquisition time to within 1/2 LSb error (512 steps for the A/D). The 1/2 LSb error is the maximum error allowed for the A/D to meet its specified accuracy.

Note: When the conversion is started, the hold-ing capacitor is disconnected from the input pin.

FIGURE 8-4: ANALOG INPUT MODEL



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8.4 A/D Conversions

Note: The GO/DONE bit should **NOT** be set in the same instruction that turns on the A/D.

8.5 Use of the CCP Trigger

An A/D conversion can be started by the “Special Event Trigger” of the CCP1 module. This requires that the CCP1M3:CCP1M0 bits (CCP1CON<3:0>) be programmed as 1011 and that the A/D module is enabled (ADON bit is set). When the trigger occurs, the

GO/DONE bit will be set, starting the A/D conversion, and the Timer1 counter will be reset to zero. Timer1 is reset to automatically repeat the A/D acquisition period with minimal software overhead (moving the ADRES to the desired location). The appropriate analog input channel must be selected and the minimum acquisition done before the “Special Event Trigger” sets the GO/DONE bit (starts a conversion).

If the A/D module is not enabled (ADON is cleared), then the “Special Event Trigger” will be ignored by the A/D module, but will still reset the Timer1 counter.

TABLE 8-2: SUMMARY OF A/D REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets
05h	PORTA	—	—	— ⁽¹⁾	RA4	RA3	RA2	RA1	RA0	--xx xxxx	--xu uuuu
0Bh,8Bh	INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	—	ADIF	—	—	—	CCP1IF	TMR2IF	TMR1IF	-0-- -000	-0-- -000
1Eh	ADRES	A/D Result Register								xxxx xxxx	uuuu uuuu
1Fh	ADCON0	ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/DONE	—	ADON	0000 00-0	0000 00-0
85h	TRISA	—	—	— ⁽¹⁾	PORTA Data Direction Register					---1 1111	---1 1111
8Ch	PIE1	—	ADIE	—	—	—	CCP1IE	TMR2IE	TMR1IE	-0-- -000	-0-- 0000
9Fh	ADCON1	—	—	—	—	—	PCFG2	PCFG1	PCFG0	---- -000	---- -000

Legend: x = unknown, u = unchanged, — = unimplemented read as ‘0’. Shaded cells are not used for A/D conversion.

Note 1: Reserved bits; Do Not Use.

Note 1: This is a separate oscillator from the RC oscillator of the CLKIN pin.

The figure consists of three vertically stacked timing diagrams, each showing the relationship between VDD, VBOR, and Internal Reset signals.

- Top Diagram (Normal Reset):** VDD is initially high, then drops below VBOR (indicated by a dashed line) and returns high. Internal Reset is active (low) during the VDD drop and remains active for 72 ms after VDD returns to high.
- Middle Diagram (Brown-out Reset):** VDD is initially high, then drops below VBOR and returns high. Internal Reset is active (low) during the VDD drop and remains active for 72 ms after VDD returns to high. A second, shorter pulse of Internal Reset is shown during the VDD drop, labeled <72 ms.
- Bottom Diagram (Low-voltage Reset):** VDD is initially high, then drops below VBOR and returns high. Internal Reset is active (low) during the VDD drop and remains active for 72 ms after VDD returns to high.

9.10.1 INT INTERRUPT

External interrupt on RB0/INT pin is edge triggered, either rising if bit INTEDG (OPTION_REG<6>) is set, or falling if the INTEDG bit is clear. When a valid edge appears on the RB0/INT pin, flag bit INTF (INTCON<1>) is set. This interrupt can be disabled by clearing enable bit INTE (INTCON<4>). Flag bit INTF must be cleared in software in the Interrupt Service Routine before re-enabling this interrupt. The INT interrupt can wake-up the processor from Sleep, if bit INTE was set prior to going into Sleep. The status of global interrupt enable bit GIE decides whether or not the processor branches to the interrupt vector following wake-up. See **Section 9.13 “Power-down Mode (Sleep)”** for details on Sleep mode.

9.10.2 TMR0 INTERRUPT

An overflow (FFh → 00h) in the TMR0 register will set flag bit T0IF (INTCON<2>). The interrupt can be enabled/disabled by setting/clearing enable bit T0IE (INTCON<5>). (**Section 4.0 “Timer0 Module”**)

9.10.3 PORTB INTCON CHANGE

An input change on PORTB<7:4> sets flag bit RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit RBIE (INTCON<4>). (**Section 3.2 “PORTB and the TRISB Register”**)

9.11 Context Saving During Interrupts

During an interrupt, only the return PC value is saved on the stack. Typically, users may wish to save key registers during an interrupt, (i.e., W register and STATUS register). This will have to be implemented in software.

Example 9-1 stores and restores the W and STATUS registers. The register, W_TEMP, must be defined in each bank and must be defined at the same offset from the bank base address (i.e., if W_TEMP is defined at 0x20 in bank 0, it must also be defined at 0xA0 in bank 1).

The example:

- a) Stores the W register.
- b) Stores the STATUS register in bank 0.
- c) Stores the PCLATH register.
- d) Executes the Interrupt Service Routine code (User-generated).
- e) Restores the STATUS register (and bank select bit).
- f) Restores the W and PCLATH registers.

EXAMPLE 9-1: SAVING STATUS, W, AND PCLATH REGISTERS IN RAM

```

MOVWF    W_TEMP           ;Copy W to TEMP register, could be bank one or zero
SWAPF    STATUS,W         ;Swap status to be saved into W
CLRF     STATUS           ;bank 0, regardless of current bank, Clears IRP,RP1,RP0
MOVWF    STATUS_TEMP      ;Save status to bank zero STATUS_TEMP register
MOVF     PCLATH, W        ;Only required if using pages 1, 2 and/or 3
MOVWF    PCLATH_TEMP      ;Save PCLATH into W
CLRF     PCLATH           ;Page zero, regardless of current page
BCF      STATUS, IRP      ;Return to Bank 0
MOVF     FSR, W           ;Copy FSR to W
MOVWF    FSR_TEMP        ;Copy FSR from W to FSR_TEMP
:
:(ISR)
:
MOVF     PCLATH_TEMP, W   ;Restore PCLATH
MOVWF    PCLATH           ;Move W into PCLATH
SWAPF    STATUS_TEMP,W   ;Swap STATUS_TEMP register into W
                        ;(sets bank to original state)
MOVWF    STATUS           ;Move W into STATUS register
SWAPF    W_TEMP,F        ;Swap W_TEMP
SWAPF    W_TEMP,W        ;Swap W_TEMP into W

```

9.16 In-Circuit Serial Programming™

PIC16CXXX microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data, and three other lines for power, ground and the programming voltage. This allows customers to manufacture boards with unprogrammed devices, and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

For complete details on serial programming, please refer to the In-Circuit Serial Programming™ (ICSP™) Guide, (DS30277).

12.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings ^(†)

Ambient temperature under bias	-55°C to +125°C
Storage temperature	-65°C to +150°C
Voltage on any pin with respect to V _{SS} (except V _{DD} , $\overline{\text{MCLR}}$, and RA4)	-0.3V to (V _{DD} + 0.3V)
Voltage on V _{DD} with respect to V _{SS}	-0.3V to +7.5V
Voltage on $\overline{\text{MCLR}}$ with respect to V _{SS} (Note 2)	0V to +13.25V
Voltage on RA4 with respect to V _{SS}	0V to +8.5V
Total power dissipation (Note 1) (PDIP and SOIC)	1.0W
Total power dissipation (Note 1) (SSOP)	0.65W
Maximum current out of V _{SS} pin	300 mA
Maximum current into V _{DD} pin	250 mA
Input clamp current, I _{IK} (V _I < 0 or V _I > V _{DD})	±20 mA
Output clamp current, I _{OK} (V _O < 0 or V _O > V _{DD})	±20 mA
Maximum output current sunk by any I/O pin	25 mA
Maximum output current sourced by any I/O pin	25 mA
Maximum current sunk by PORTA and PORTB (combined)	200 mA
Maximum current sourced by PORTA and PORTB (combined)	200 mA

Note 1: Power dissipation is calculated as follows: $P_{dis} = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD} - V_{OH}) \times I_{OH}\} + \sum (V_{OL} \times I_{OL})$

- 2:** Voltage spikes below V_{SS} at the $\overline{\text{MCLR}}$ /V_{PP} pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100Ω should be used when applying a “low” level to the $\overline{\text{MCLR}}$ /V_{PP} pin rather than pulling this pin directly to V_{SS}.

† NOTICE: Stresses above those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

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12.2 DC Characteristics: PIC16LC712/716-04 (Commercial, Industrial)

DC CHARACTERISTICS		Standard Operating Conditions (unless otherwise stated)					
		Operating temperature					
		0°C ≤ TA ≤ +70°C for commercial					
		-40°C ≤ TA ≤ +85°C for industrial					
Param No.	Sym.	Characteristic	Min.	Typ†	Max.	Units	Conditions
D001	VDD	Supply Voltage	2.5	—	5.5	V	BOR enabled (Note 7)
			VBOR*	—	5.5	V	
D002*	VDR	RAM Data Retention Voltage⁽¹⁾	—	1.5	—	V	
D003	VPOR	VDD Start Voltage to ensure internal Power-on Reset signal	—	VSS	—	V	See section on Power-on Reset for details
D004*	SVDD	VDD Rise Rate to ensure internal Power-on Reset signal	0.05	—	—	V/ms	PWRT enabled (PWRT $\overline{\text{TE}}$ bit clear)
D004A*			TBD	—	—		PWRT disabled (PWRT $\overline{\text{TE}}$ bit set) See section on Power-on Reset for details
D005	VBOR	Brown-out Reset voltage trip point	3.65	—	4.35	V	BODEN bit set
D010	IDD	Supply Current^(2,5)	—	2.0	3.8	mA	XT, RC osc modes FOSC = 4 MHz, VDD = 3.0V (Note 4)
D010A			—	22.5	48	μA	LP osc mode FOSC = 32 kHz, VDD = 3.0V, WDT disabled
D020	IPD	Power-down Current^(3,5)	—	7.5	30	μA	VDD = 3.0V, WDT enabled, -40°C to +85°C
D021			—	0.9	5	μA	VDD = 3.0V, WDT disabled, 0°C to +70°C
D021A			—	0.9	5	μA	VDD = 3.0V, WDT disabled, -40°C to +85°C
D022*	ΔIWD $\overline{\text{T}}$	Module Differential Current⁽⁶⁾ Watchdog Timer	—	6.0	20	μA	WDTE bit set, VDD = 4.0V
D022A*	ΔIBOR	Brown-out Reset	—	TBD	200	μA	BODEN bit set, VDD = 5.0V
1A	FOSC	LP Oscillator Operating Frequency	0	—	200	KHz	All temperatures
		RC Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		XT Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		HS Oscillator Operating Frequency	0	—	20	MHz	All temperatures

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note1: This is the limit to which VDD can be lowered without losing RAM data.

- 2:** The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD,

MCLR = VDD; WDT enabled/disabled as specified.

- 3:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD and VSS.
- 4:** For RC Osc mode, current through REXT is not included. The current through the resistor can be estimated by the formula $I_r = V_{DD}/2R_{EXT}$ (mA) with REXT in kOhm.
- 5:** Timer1 oscillator (when enabled) adds approximately 20 μA to the specification. This value is from characterization and is for design guidance only. This is not tested.
- 6:** The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.
- 7:** This is the voltage where the device enters the Brown-out Reset. When BOR is enabled, the device will operate correctly to this trip point.

12.4 AC (Timing) Characteristics

12.4.1 TIMING PARAMETER SYMBOLOGY

The timing parameter symbols have been created using one of the following formats:

1. TppS2ppS
2. TppS

T			
F	Frequency	T	Time

Lowercase letters (pp) and their meanings:

pp			
cc	CCP1	osc	OSC1
ck	CLKOUT	rd	\overline{RD}
cs	\overline{CS}	rw	\overline{RD} or \overline{WR}
di	SDI	sc	SCK
do	SDO	ss	\overline{SS}
dt	Data in	t0	T0CKI
io	I/O port	t1	T1CKI
mc	\overline{MCLR}	wr	\overline{WR}

Uppercase letters and their meanings:

S			
F	Fall	P	Period
H	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance

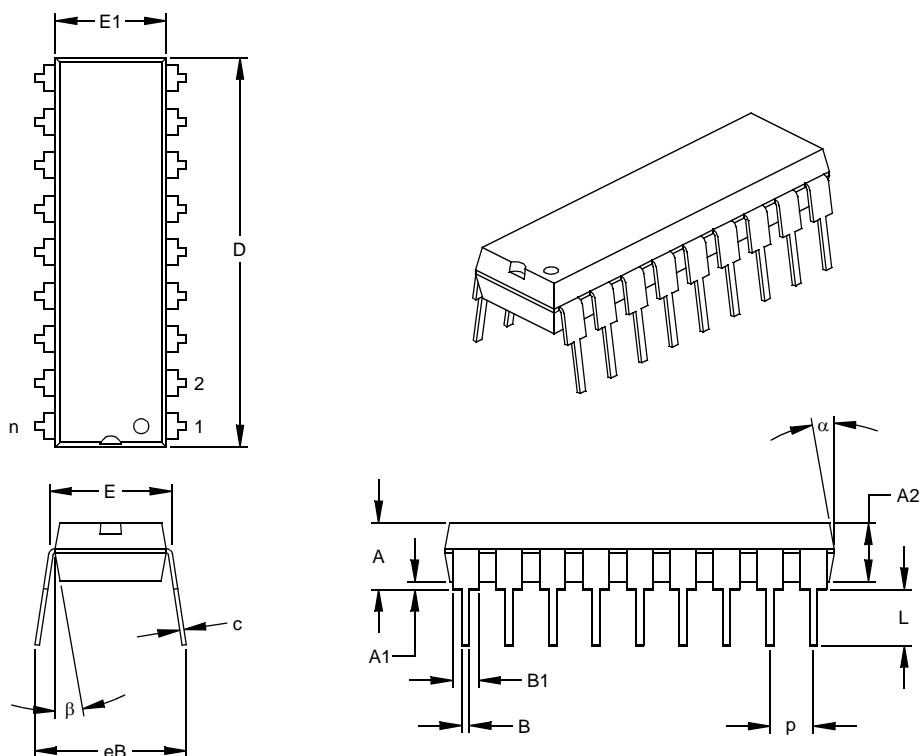
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13.2 Package Details

The following sections give the technical details of the packages.

18-Lead Plastic Dual In-line (P) – 300 mil (PDIP)

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Units		INCHES*			MILLIMETERS		
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		18			18	
Pitch	p		.100			2.54	
Top to Seating Plane	A	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015			0.38		
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.890	.898	.905	22.61	22.80	22.99
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	c	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	B	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing	eB	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

* Controlling Parameter

§ Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side.

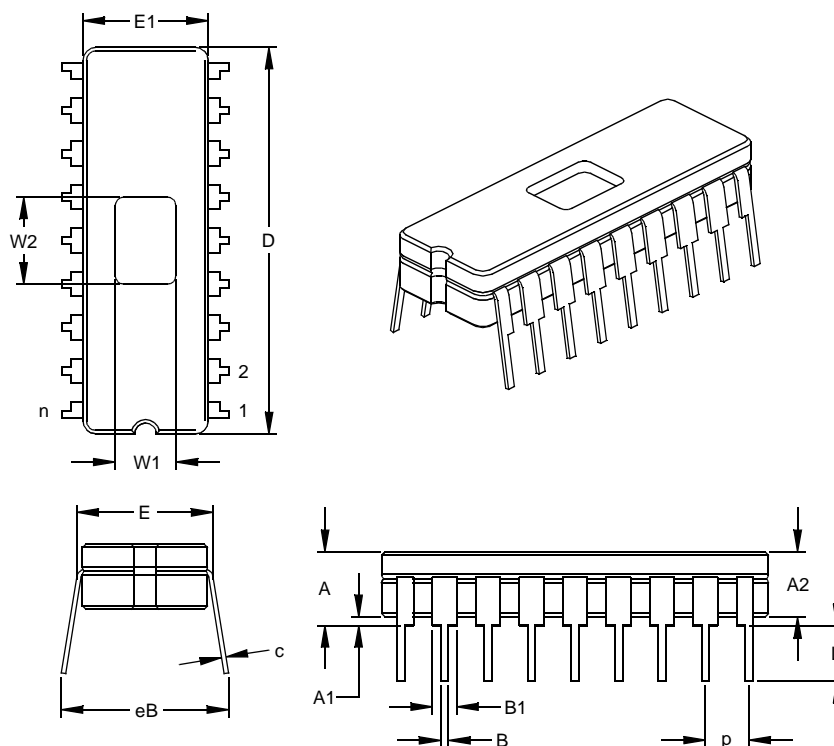
JEDEC Equivalent: MS-001

Drawing No. C04-007

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18-Lead Ceramic Dual In-line with Window (JW) – 300 mil (CERDIP)

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Units		INCHES*			MILLIMETERS		
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		18			18	
Pitch	p		.100			2.54	
Top to Seating Plane	A	.170	.183	.195	4.32	4.64	4.95
Ceramic Package Height	A2	.155	.160	.165	3.94	4.06	4.19
Standoff	A1	.015	.023	.030	0.38	0.57	0.76
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26
Ceramic Pkg. Width	E1	.285	.290	.295	7.24	7.37	7.49
Overall Length	D	.880	.900	.920	22.35	22.86	23.37
Tip to Seating Plane	L	.125	.138	.150	3.18	3.49	3.81
Lead Thickness	c	.008	.010	.012	0.20	0.25	0.30
Upper Lead Width	B1	.050	.055	.060	1.27	1.40	1.52
Lower Lead Width	B	.016	.019	.021	0.41	0.47	0.53
Overall Row Spacing	§ eB	.345	.385	.425	8.76	9.78	10.80
Window Width	W1	.130	.140	.150	3.30	3.56	3.81
Window Length	W2	.190	.200	.210	4.83	5.08	5.33

* Controlling Parameter
 § Significant Characteristic
 JEDEC Equivalent: MO-036
 Drawing No. C04-010

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