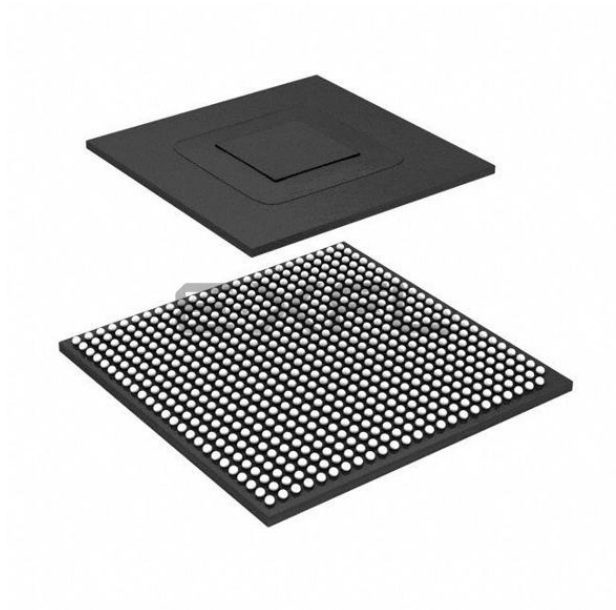


Welcome to E-XFL.COM

Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

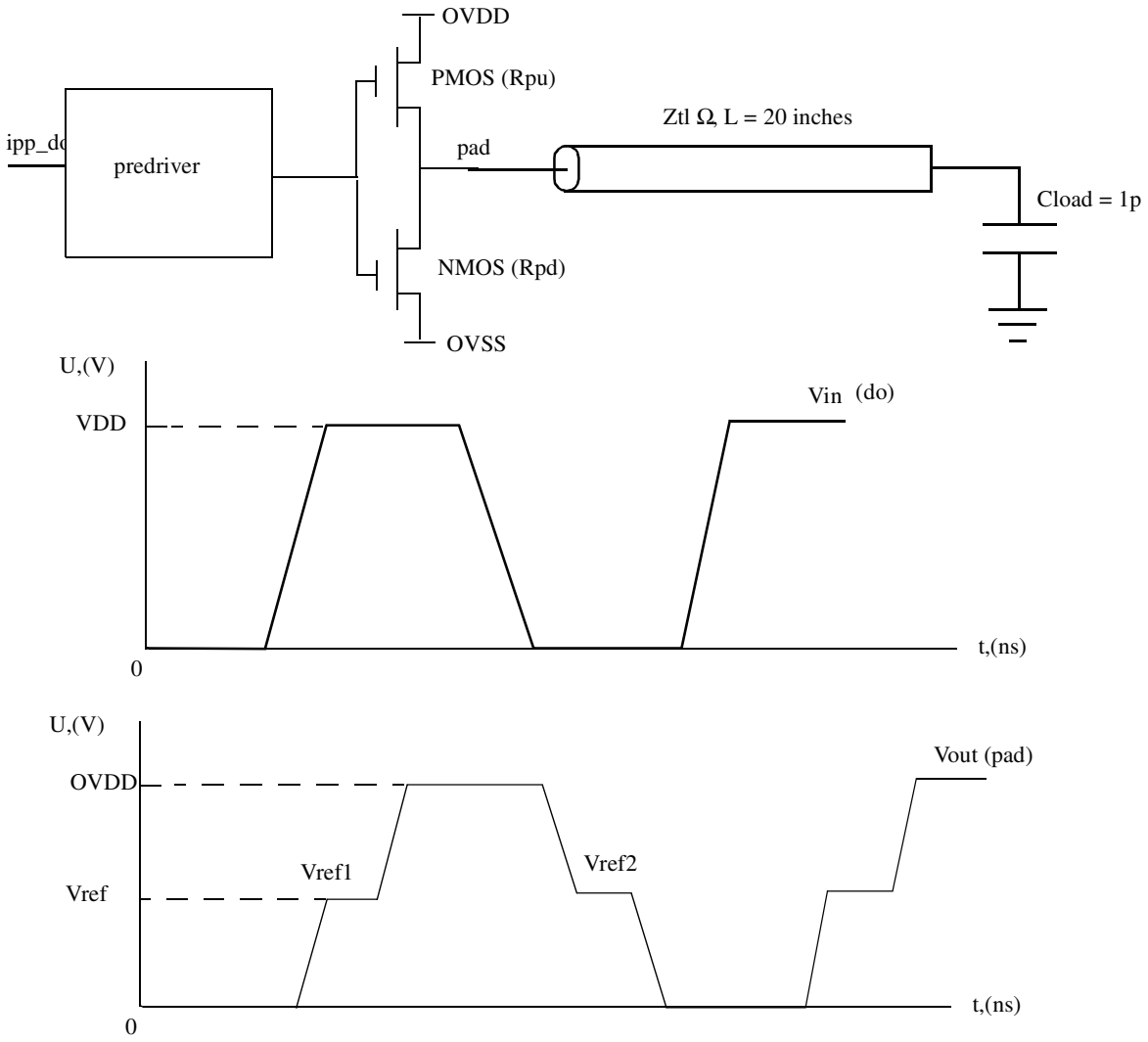
Details	
Product Status	Active
Core Processor	ARM® Cortex®-A9
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	800MHz
Co-Processors/DSP	Multimedia; NEON™ SIMD
RAM Controllers	LPDDR2, LVDDR3, DDR3
Graphics Acceleration	Yes
Display & Interface Controllers	Keypad, LCD
Ethernet	10/100/1000Mbps (1)
SATA	-
USB	USB 2.0 + PHY (4)
Voltage - I/O	1.8V, 2.5V, 2.8V, 3.3V
Operating Temperature	-40°C ~ 125°C (TJ)
Security Features	ARM TZ, Boot Security, Cryptography, RTIC, Secure Fusebox, Secure JTAG, Secure Memory, Secure RTC, Tamper Detection
Package / Case	624-LFBGA
Supplier Device Package	624-MAPBGA (21x21)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mcimx6s4avm08ac

Table 2. i.MX 6Solo/6DualLite Modules List (continued)

Block Mnemonic	Block Name	Subsystem	Brief Description
SSI-1 SSI-2 SSI-3	I2S/SSI/AC97 Interface	Connectivity Peripherals	The SSI is a full-duplex synchronous interface, which is used on the AP to provide connectivity with off-chip audio peripherals. The SSI supports a wide variety of protocols (SSI normal, SSI network, I2S, and AC-97), bit depths (up to 24 bits per word), and clock / frame sync options. The SSI has two pairs of 8x24 FIFOs and hardware support for an external DMA controller in order to minimize its impact on system performance. The second pair of FIFOs provides hardware interleaving of a second audio stream that reduces CPU overhead in use cases where two time slots are being used simultaneously.
TEMPMON	Temperature Monitor	System Control Peripherals	The Temperature sensor IP is used for detecting die temperature. The temperature read out does not reflect case or ambient temperature. It reflects the temperature in proximity of the sensor location on the die. Temperature distribution may not be uniformly distributed, therefore the read out value may not be the reflection of the temperature value of the entire die.
TZASC	Trust-Zone Address Space Controller	Security	The TZASC (TZC-380 by ARM) provides security address region control functions required for intended application. It is used on the path to the DRAM controller.
UART-1 UART-2 UART-3 UART-4 UART-5	UART Interface	Connectivity Peripherals	Each of the UARTv2 modules support the following serial data transmit/receive protocols and configurations: • 7- or 8-bit data words, 1 or 2 stop bits, programmable parity (even, odd or none) • Programmable baud rates up to 4 MHz. This is a higher max baud rate relative to the 1.875 MHz, which is stated by the TIA/EIA-232-F standard and the i.MX31 UART modules. • 32-byte FIFO on Tx and 32 half-word FIFO on Rx supporting auto-baud • IrDA 1.0 support (up to SIR speed of 115200 bps) • Option to operate as 8-pins full UART, DCE, or DTE
USBOH3	USB 2.0 High Speed OTG and 3x HS Hosts	Connectivity Peripherals	USBOH3 contains: • One high-speed OTG module with integrated HS USB PHY • One high-speed Host module with integrated HS USB PHY • Two identical high-speed Host modules connected to HSIC USB ports.
VDOA	VDOA	Multimedia Peripherals	Video Data Order Adapter (VDOA): used to re-order video data from the “tiled” order used by the VPU to the conventional raster-scan order needed by the IPU.

Table 3. Special Signal Considerations (continued)

Signal Name	Remarks
DRAM_VREF	When using DDR_VREF with DDR I/O, the nominal reference voltage must be half of the NVCC_DRAM supply. The user must tie DDR_VREF to a precision external resistor divider. Use a 1 kΩ 0.5% resistor to GND and a 1 kΩ 0.5% resistor to NVCC_DRAM. Shunt each resistor with a closely-mounted 0.1 µF capacitor. To reduce supply current, a pair of 1.5 kΩ 0.1% resistors can be used. Using resistors with recommended tolerances ensures the ± 2% DDR_VREF tolerance (per the DDR3 specification) is maintained when four DDR3 ICs plus the i.MX 6Solo/6DualLite are drawing current on the resistor divider. It is recommended to use regulated power supply for “big” memory configurations (more that eight devices)
ZQPAD	DRAM calibration resistor 240 Ω 1% used as reference during DRAM output buffer driver calibration should be connected between this pad and GND.
NVCC_LVDS2P5	The DDR pre-drivers share the NVCC_LVDS2P5 ball with the LVDS interface. This ball can be shorted to VDDHIGH_CAP on the circuit board.
VDD_FA FA_ANA	These signals are reserved for Freescale manufacturing use only. User must tie both connections to GND.
GPANAIO	This signal is reserved for Freescale manufacturing use only. User must leave this connection floating.
JTAG_*nnnn*	The JTAG interface is summarized in Table 4. Use of external resistors is unnecessary. However, if external resistors are used, the user must ensure that the on-chip pull-up/down configuration is followed. For example, do not use an external pull down on an input that has on-chip pull-up.
	JTAG_TDO is configured with a keeper circuit such that the floating condition is eliminated if an external pull resistor is not present. An external pull resistor on JTAG_TDO is detrimental and should be avoided.
	JTAG_MOD is referenced as SJC_MOD in the i.MX 6Solo/6DualLite reference manual. Both names refer to the same signal. JTAG_MOD must be externally connected to GND for normal operation. Termination to GND through an external pull-down resistor (such as 1 kΩ) is allowed. JTAG_MOD set to hi configures the JTAG interface to mode compliant with IEEE1149.1 standard. JTAG_MOD set to low configures the JTAG interface for common SW debug adding all the system TAPs to the chain.
NC	These signals are No Connect (NC) and should be floated by the user.
POR_B	This cold reset negative logic input resets all modules and logic in the IC. May be used in addition to internally generated power on reset signal (logical AND, both internal and external signals are considered active low).
ONOFF	In normal mode may be connected to ON/OFF button (De-bouncing provided at this input). Internally this pad is pulled up. Short connection to GND in OFF mode causes internal power management state machine to change state to ON. In ON mode short connection to GND generates interrupt (intended to SW controllable power down). Long above ~5s connection to GND causes “forced” OFF.
TEST_MODE	TEST_MODE is for Freescale factory use. This signal is internally connected to an on-chip pull-down device. The user must either float this signal or tie it to GND.
PCIE_REXT	The impedance calibration process requires connection of reference resistor 200 Ω 1% precision resistor on PCIE_REXT pad to ground.



$$Rpu = \frac{Vovdd - Vref1}{Vref1} \times Ztl$$

$$Rpd = \frac{Vref2}{Vovdd - Vref2} \times Ztl$$

Figure 9. Impedance Matching Load for Measurement

Table 43. EIM Bus Timing Parameters (continued)[1]

ID	Parameter	BCD = 0		BCD = 1		BCD = 2		BCD = 3	
		Min	Max	Min	Max	Min	Max	Min	Max
WE19	Input Data hold time from Clock rise	2	—	2	—	—	—	—	—
WE20	WAIT_B setup time to Clock rise	2	—	4	—	—	—	—	—
WE21	WAIT_B hold time from Clock rise	2	—	2	—	—	—	—	—

1 t is the maximal EIM logic (axi_clk) cycle time. The maximum allowed axi_clk frequency depends on the fixed/non-fixed latency configuration, whereas the maximum allowed BCLK frequency is:

—Fixed latency for both read and write is 132 MHz.

—Variable latency for read only is 132 MHz.

—Variable latency for write only is 52 MHz.

In variable latency configuration for write, if BCD = 0 & WBCDD = 1 or BCD = 1, axi_clk must be 104 MHz.Write BCD = 1 and 104 MHz axi_clk, will result in a BCLK of 52 MHz. When the clock branch to EIM is decreased to 104 MHz, other buses are impacted which are clocked from this source. See the CCM chapter of the *i.MX 6Solo/6DualLite Reference Manual (IMX6SDLRM)* for a detailed clock tree description.

2 BCLK parameters are being measured from the 50% point, that is, high is defined as 50% of signal value and low is defined as 50% as signal value.

3 For signal measurements, “High” is defined as 80% of signal value and “Low” is defined as 20% of signal value.

Figure 14 to Figure 17 provide few examples of basic EIM accesses to external memory devices with the timing parameters mentioned previously for specific control parameters settings.

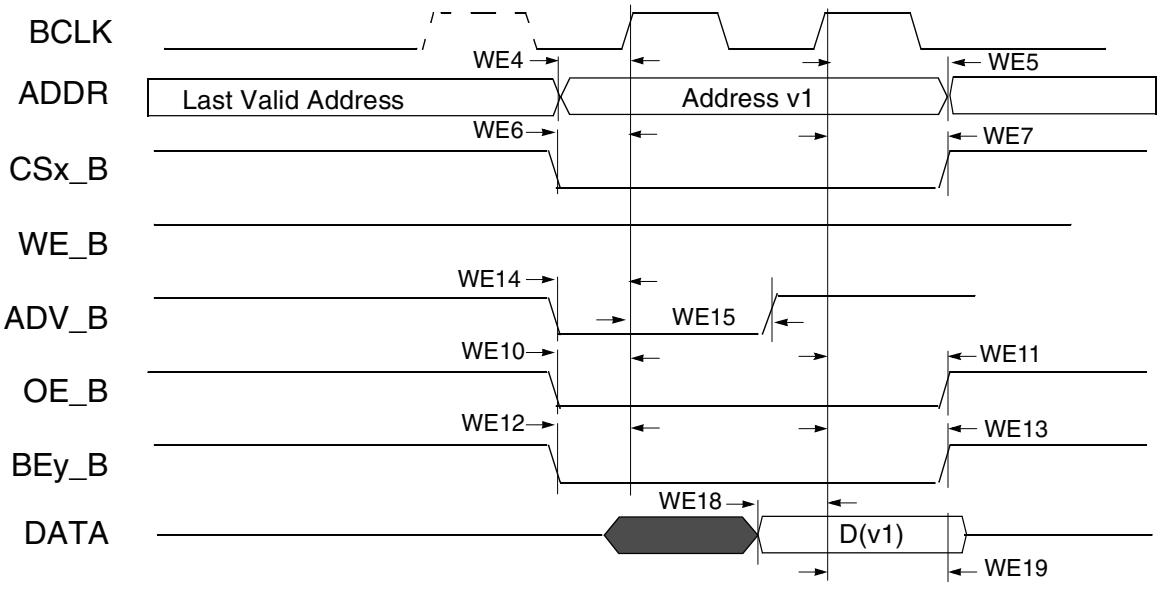


Figure 14. Synchronous Memory Read Access, WSC=1

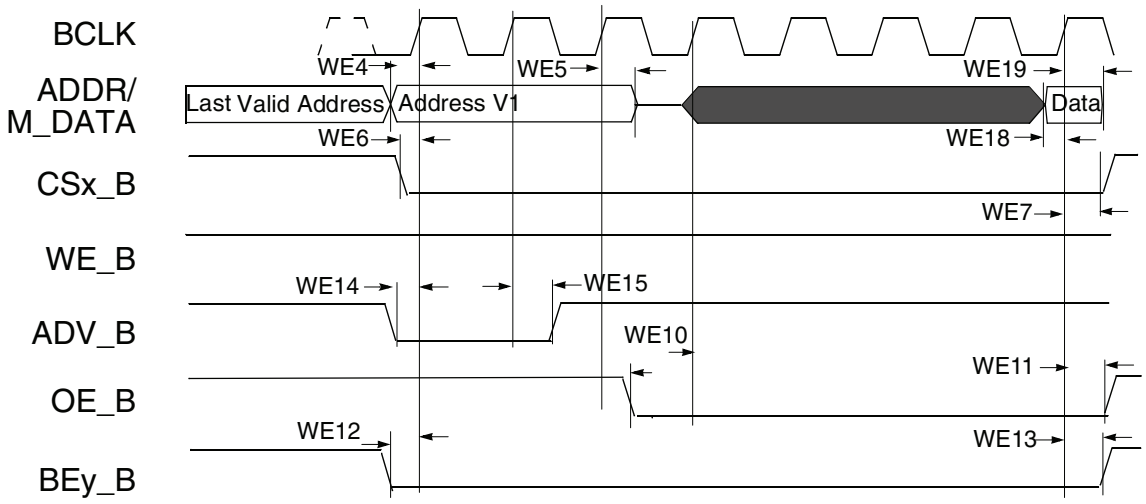


Figure 17. 16-Bit Muxed A/D Mode, Synchronous Read Access, WSC=7, RADVN=1, ADH=1, OEA=0

4.9.3.5 General EIM Timing-Asynchronous Mode

Figure 18 through Figure 22, and Table 44 help you determine timing parameters relative to the chip select (CS) state for asynchronous and DTACK EIM accesses with corresponding EIM bit fields and the timing parameters mentioned above.

Asynchronous read & write access length in cycles may vary from what is shown in Figure 18 through Figure 21 as RWSC, OEN & CSN is configured differently. See the *i.MX 6Solo/6DualLite Reference Manual (IMX6SDLRM)* for the EIM programming model.

4.10.3.2 Read and Write Timing

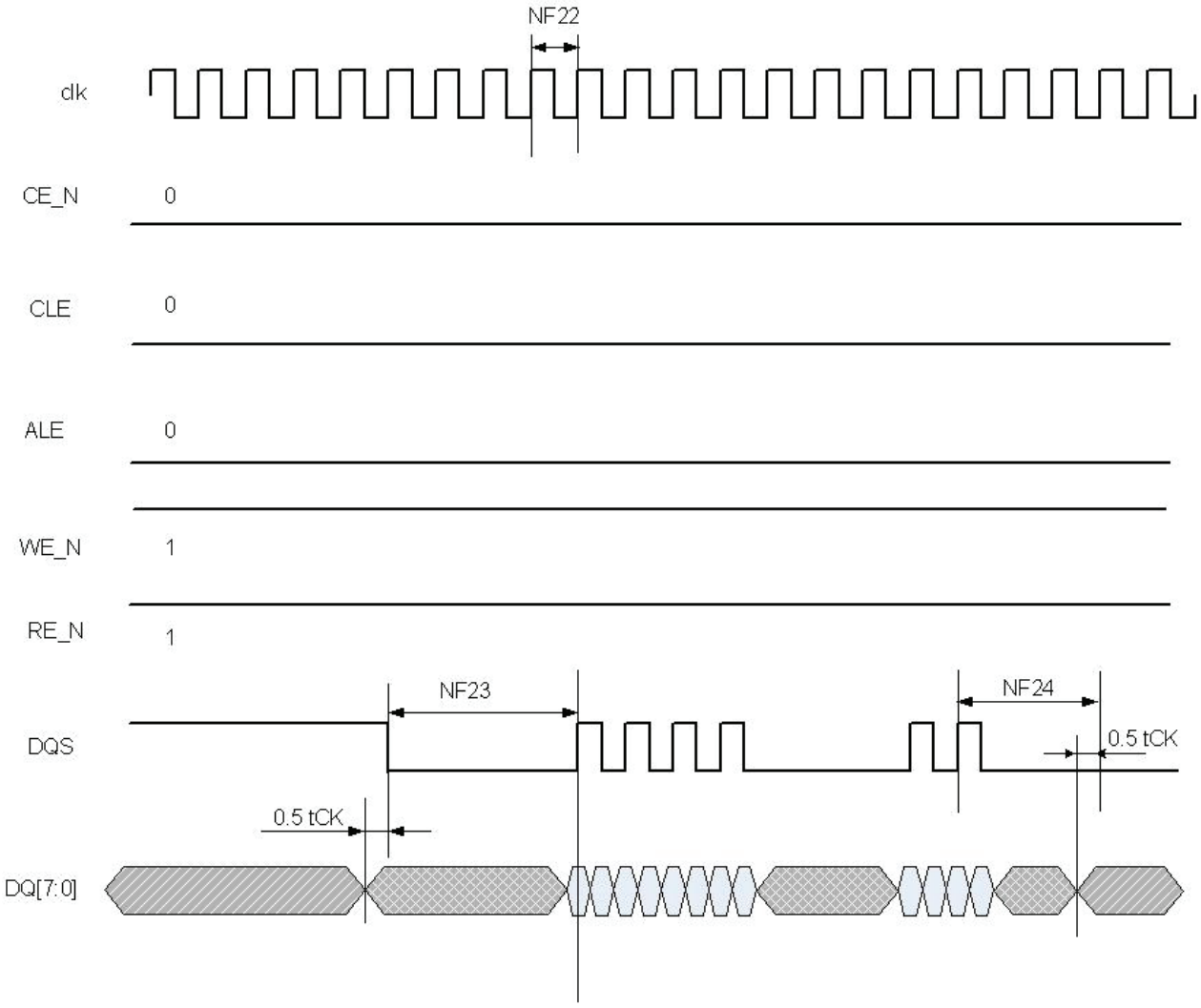


Figure 38. Samsung Toggle Mode Data Write Timing

4.11.2.1 ECSPI Master Mode Timing

Figure 40 depicts the timing of ECSPI in master mode. Table 54 lists the ECSPI master mode timing characteristics.

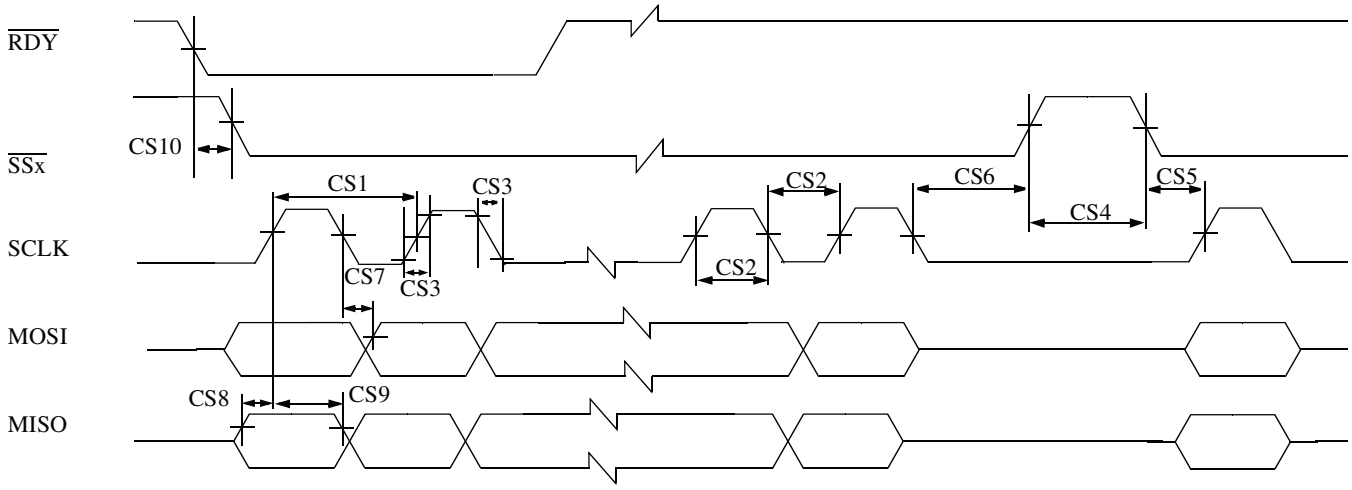


Figure 40. ECSPI Master Mode Timing Diagram

Table 54. ECSPI Master Mode Timing Parameters

ID	Parameter	Symbol	Min	Max	Unit
CS1	SCLK Cycle Time–Read SCLK Cycle Time–Write	t_{clk}	43 15	—	ns
CS2	SCLK High or Low Time–Read SCLK High or Low Time–Write	t_{SW}	21.5 7	—	ns
CS3	SCLK Rise or Fall[1]	$t_{RISE/FALL}$	—	—	ns
CS4	SSx pulse width	t_{CSLH}	Half SCLK period	—	ns
CS5	SSx Lead Time (CS setup time)	t_{SCS}	Half SCLK period - 4	—	ns
CS6	SSx Lag Time (CS hold time)	t_{HCS}	Half SCLK period - 2	—	ns
CS7	MOSI Propagation Delay (C_{LOAD} = 20 pF)	t_{PDmosi}	-1	1	ns
CS8	MISO Setup Time	t_{Smiso}	18	—	ns
CS9	MISO Hold Time	t_{Hmiso}	0	—	ns
CS10	RDY to SSx Time[2]	t_{SDRY}	5	—	ns

1 See specific I/O AC parameters Section 4.7, “I/O AC Parameters.”

2 SPI_RDY is sampled internally by ipg_clk and is asynchronous to all other CSPI signals.

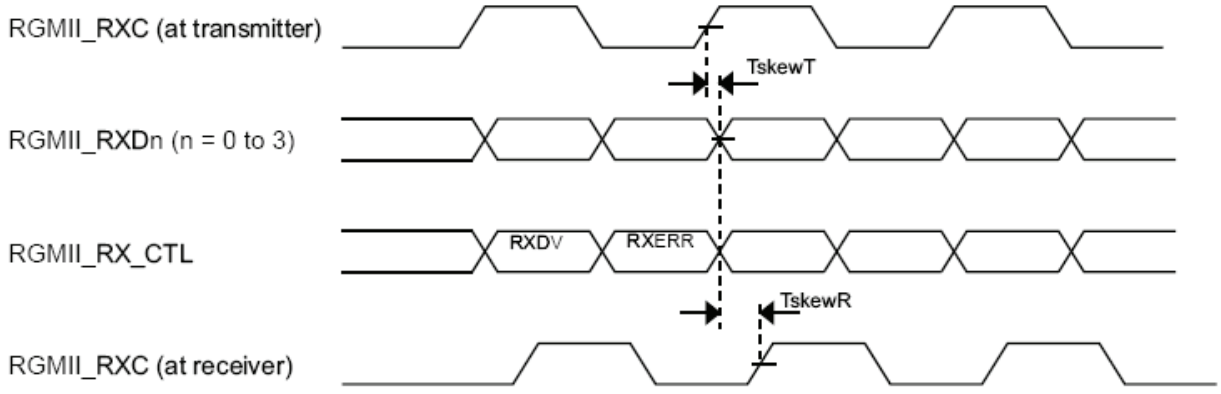


Figure 53. RGMII Receive Signal Timing Diagram Original

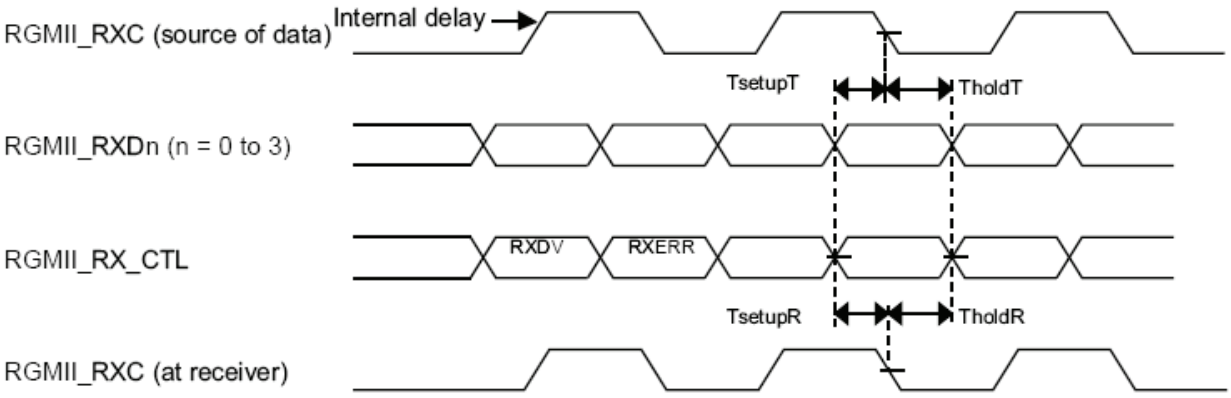


Figure 54. RGMII Receive Signal Timing Diagram with Internal Delay

4.11.6 Flexible Controller Area Network (FLEXCAN) AC Electrical Specifications

The Flexible Controller Area Network (FlexCAN) module is a communication controller implementing the CAN protocol according to the CAN 2.0B protocol specification. The processor has two CAN modules available for systems design. Tx and Rx ports for both modules are multiplexed with other I/O pins. See the IOMUXC chapter of the *i.MX 6Solo/6DualLite Reference Manual (IMX6SDLRM)* to see which pins expose Tx and Rx pins; these ports are named TXCAN and RXCAN, respectively.

4.11.7 HDMI Module Timing Parameters

4.11.7.1 Latencies and Timing Information

Power-up time (time between TX_PWRON assertion and TX_READY assertion) for the HDMI 3D Tx PHY while operating with the slowest input reference clock supported (13.5 MHz) is 3.35 ms.

Table 66. Electrical Characteristics

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
Operating conditions for HDMI						
avddtmds	Termination supply voltage	-	3.15	3.3	3.45	V
R_T	Termination resistance	-	45	50	55	Ω
TMDS drivers DC specifications						
V_{OFF}	Single-ended standby voltage	RT = 50 Ω For measurement conditions and definitions, see the first two figures above. Compliance point TP1 as defined in the HDMI specification, version 1.3a, section 4.2.4.	avddtmds ± 10 mV			mV
V_{SWING}	Single-ended output swing voltage		400	-	600	mV
V_H	Single-ended output high voltage For definition, see the second figure above	If attached sink supports TMDSCLK < or = 165 MHz	avddtmds ± 10 mV			mV
		If attached sink supports TMDSCLK > 165 MHz	avddtmds - 200 mV	-	avddtmds + 10 mV	mV
V_L	Single-ended output low voltage For definition, see the second figure above	If attached sink supports TMDSCLK < or = 165 MHz	avddtmds - 600 mV	-	avddtmds - 400mV	mV
		If attached sink supports TMDSCLK > 165 MHz	avddtmds - 700 mV	-	avddtmds - 400 mV	mV
R_{TERM}	Differential source termination load (inside HDMI 3D Tx PHY) Although the HDMI 3D Tx PHY includes differential source termination, the user-defined value is set for each single line (for illustration, see the third figure above). **Note:** R_{TERM} can also be configured to be open and not present on TMDS channels.	-	50	-	200	Ω
Hot plug detect specifications						
HPD^{VH}	Hot plug detect high range	-	2.0	-	5.3	V
$^{VHPD}_{VL}$	Hot plug detect low range	-	0	-	0.8	V
$^{HPD}_{Z}$	Hot plug detect input impedance	-	10	-	-	kΩ
$^{HPD}_{t}$	Hot plug detect time delay	-	-	-	100	µs

4.11.8 Switching Characteristics

Table 67 describes switching characteristics for the HDMI 3D Tx PHY. Figure 58 to Figure 62 illustrate various parameters specified in table.

NOTE

All dynamic parameters related to the TMDS line drivers' performance imply the use of assembly guidelines.

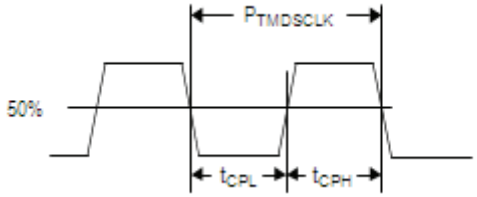


Figure 58. TMDS Clock Signal Definitions

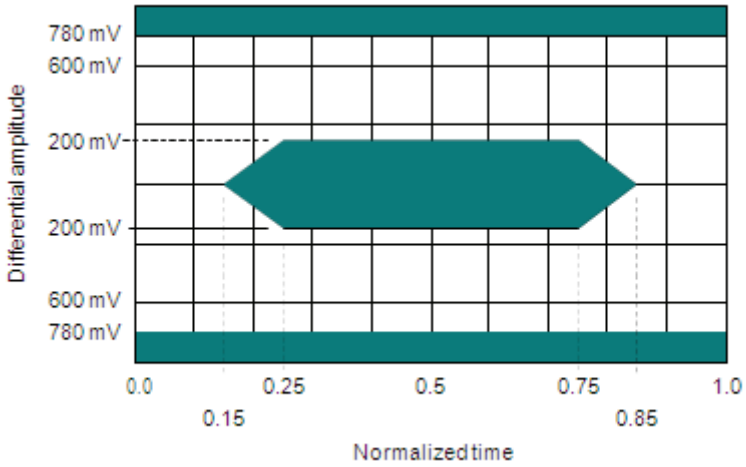


Figure 59. Eye Diagram Mask Definition for HDMI Driver Signal Specification at TP1

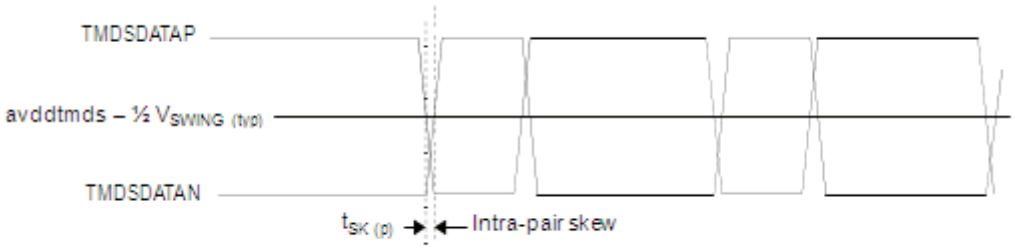


Figure 60. Intra-Pair Skew Definition

4.11.10.1 IPU Sensor Interface Signal Mapping

The IPU supports a number of sensor input formats. Table 69 defines the mapping of the Sensor Interface Pins used for various supported interface formats.

Table 69. Camera Input Signal Cross Reference, Format, and Bits Per Cycle

Signal Name[1]	RGB565 8 bits 2 cycles	RGB565[2] 8 bits 3 cycles	RGB666[3] 8 bits 3 cycles	RGB888 8 bits 3 cycles	YCbCr[4] 8 bits 2 cycles	RGB565[5] 16 bits 2 cycles	YCbCr[6] 16 bits 1 cycle	YCbCr[7] 16 bits 1 cycle	YCbCr[8] 20 bits 1 cycle
CSIx_DAT0	—	—	—	—	—	—	—	0	C[0]
CSIx_DAT1	—	—	—	—	—	—	—	0	C[1]
CSIx_DAT2	—	—	—	—	—	—	—	C[0]	C[2]
CSIx_DAT3	—	—	—	—	—	—	—	C[1]	C[3]
CSIx_DAT4	—	—	—	—	—	B[0]	C[0]	C[2]	C[4]
CSIx_DAT5	—	—	—	—	—	B[1]	C[1]	C[3]	C[5]
CSIx_DAT6	—	—	—	—	—	B[2]	C[2]	C[4]	C[6]
CSIx_DAT7	—	—	—	—	—	B[3]	C[3]	C[5]	C[7]
CSIx_DAT8	—	—	—	—	—	B[4]	C[4]	C[6]	C[8]
CSIx_DAT9	—	—	—	—	—	G[0]	C[5]	C[7]	C[9]
CSIx_DAT10	—	—	—	—	—	G[1]	C[6]	0	Y[0]
CSIx_DAT11	—	—	—	—	—	G[2]	C[7]	0	Y[1]
CSIx_DAT12	B[0], G[3]	R[2],G[4],B[2]	R/G/B[4]	R/G/B[0]	Y/C[0]	G[3]	Y[0]	Y[0]	Y[2]
CSIx_DAT13	B[1], G[4]	R[3],G[5],B[3]	R/G/B[5]	R/G/B[1]	Y/C[1]	G[4]	Y[1]	Y[1]	Y[3]
CSIx_DAT14	B[2], G[5]	R[4],G[0],B[4]	R/G/B[0]	R/G/B[2]	Y/C[2]	G[5]	Y[2]	Y[2]	Y[4]
CSIx_DAT15	B[3], R[0]	R[0],G[1],B[0]	R/G/B[1]	R/G/B[3]	Y/C[3]	R[0]	Y[3]	Y[3]	Y[5]
CSIx_DAT16	B[4], R[1]	R[1],G[2],B[1]	R/G/B[2]	R/G/B[4]	Y/C[4]	R[1]	Y[4]	Y[4]	Y[6]
CSIx_DAT17	G[0], R[2]	R[2],G[3],B[2]	R/G/B[3]	R/G/B[5]	Y/C[5]	R[2]	Y[5]	Y[5]	Y[7]
CSIx_DAT18	G[1], R[3]	R[3],G[4],B[3]	R/G/B[4]	R/G/B[6]	Y/C[6]	R[3]	Y[6]	Y[6]	Y[8]
CSIx_DAT19	G[2], R[4]	R[4],G[5],B[4]	R/G/B[5]	R/G/B[7]	Y/C[7]	R[4]	Y[7]	Y[7]	Y[9]

1 CSIx stands for CSI1 or CSI2

2 The MSB bits are duplicated on LSB bits implementing color extension

3 The two MSB bits are duplicated on LSB bits implementing color extension

4 YCbCr, 8 bits—Supported within the BT.656 protocol (sync embedded within the data stream).

5 RGB 16 bits— Supported in two ways: (1) As a “generic data” input, with no on-the-fly processing; (2) With on-the-fly processing, but only under some restrictions on the control protocol.

6 YCbCr 16 bits— Supported as a “generic-data” input, with no on-the-fly processing.

7 YCbCr 16 bits— Supported as a sub-case of the YCbCr, 20 bits, under the same conditions (BT.1120 protocol).

8 YCbCr, 20 bits, supported only within the BT.1120 protocol (syncs embedded within the data stream).

4.11.10.2 Sensor Interface Timings

There are three camera timing modes supported by the IPU.

4.11.10.2.1 BT.656 and BT.1120 Video Mode

Smart camera sensors, which include imaging processing, usually support video mode transfer. They use an embedded timing syntax to replace the SENSB_VSYNC and SENSB_HSYNC signals. The timing syntax is defined by the BT.656/BT.1120 standards.

This operation mode follows the recommendations of ITU BT.656/ ITU BT.1120 specifications. The only control signal used is SENSB_PIX_CLK. Start-of-frame and active-line signals are embedded in the data stream. An active line starts with a SAV code and ends with a EAV code. In some cases, digital blanking is inserted in between EAV and SAV code. The CSI decodes and filters out the timing-coding from the data stream, thus recovering SENSB_VSYNC and SENSB_HSYNC signals for internal use. On BT.656 one component per cycle is received over the SENSB_DATA bus. On BT.1120 two components per cycle are received over the SENSB_DATA bus.

4.11.10.2.2 Gated Clock Mode

The SENSB_VSYNC, SENSB_HSYNC, and SENSB_PIX_CLK signals are used in this mode. See Figure 64.

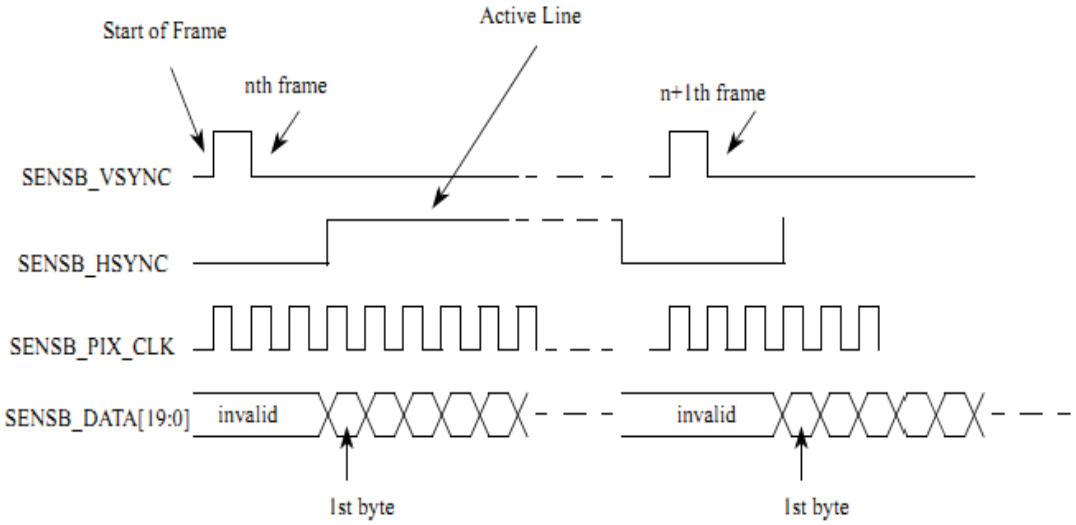


Figure 64. Gated Clock Mode Timing Diagram

A frame starts with a rising edge on SENSB_VSYNC (all the timings correspond to straight polarity of the corresponding signals). Then SENSB_HSYNC goes to high and hold for the entire line. Pixel clock is valid as long as SENSB_HSYNC is high. Data is latched at the rising edge of the valid pixel clocks. SENSB_HSYNC goes to low at the end of line. Pixel clocks then become invalid and the CSI stops receiving data from the stream. For next line the SENSB_HSYNC timing repeats. For next frame the SENSB_VSYNC timing repeats.

4.11.12.7 Forward High-Speed Data Transmission Timing

The timing relationship of the DDR Clock differential signal to the Data differential signal is shown in Figure 75:

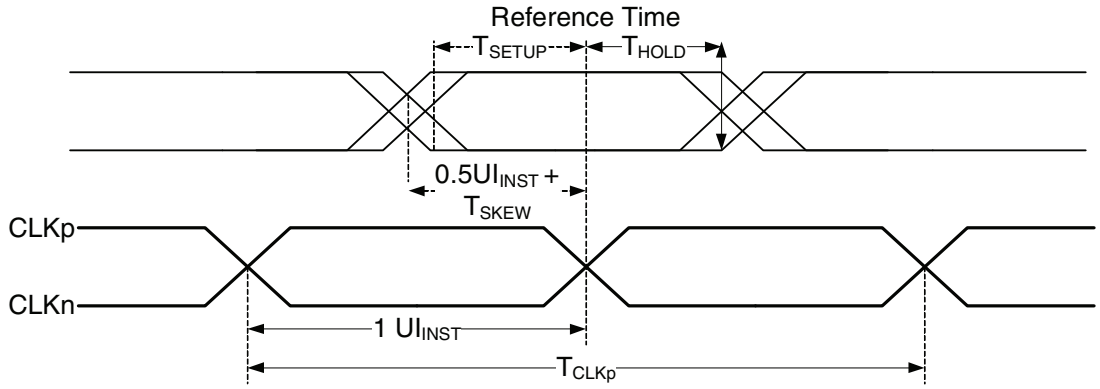


Figure 75. Data to Clock Timing Definitions

4.11.12.8 Reverse High-Speed Data Transmission Timing

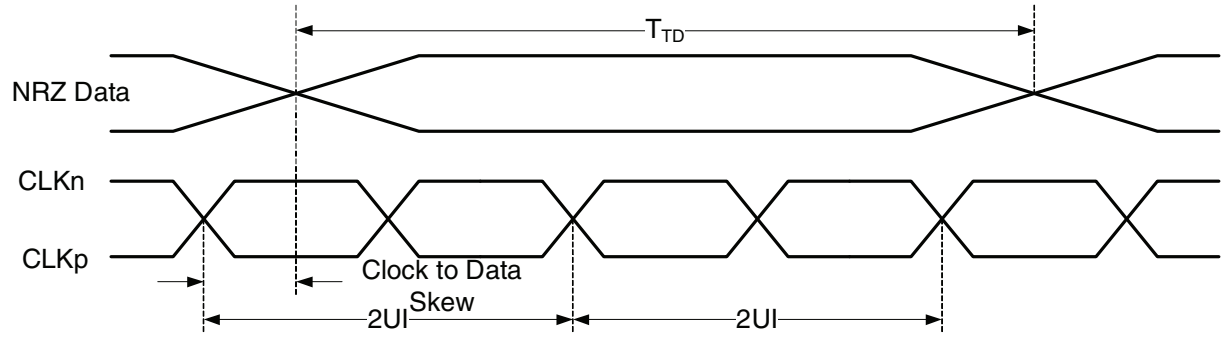


Figure 76. Reverse High-Speed Data Transmission Timing at Slave Side

4.11.12.9 Low-Power Receiver Timing

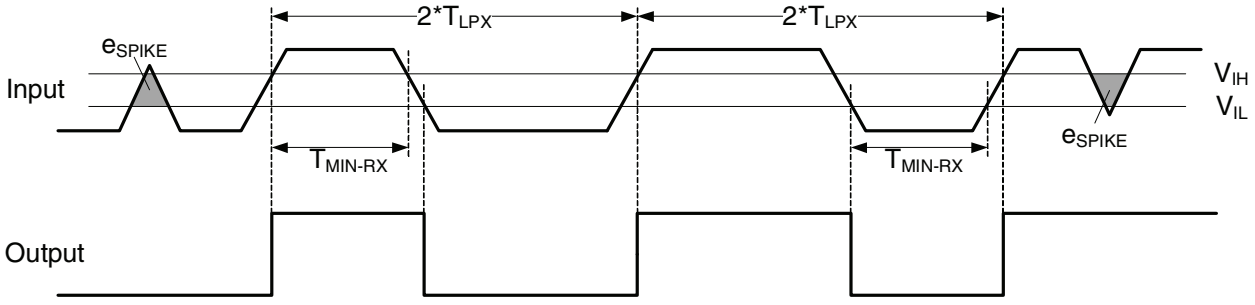


Figure 77. Input Glitch Rejection of Low-Power Receivers

Table 81. MLB 1024 Fs Timing Parameters (continued)

Parameter	Symbol	Min	Max	Unit	Comment
MLBSIG/MLBDAT output high impedance from MLBCLK low	t_{mcfdz}	0	t_{mckl}	ns	[3]
Bus Hold from MLBCLK low	t_{mdzh}	2	—	ns	—
MLBSIG/MLBDAT output valid from transition of MLBCLK (low to high)	t_{delay}	—	7	ns	—

[1] The controller can shut off MLBCLK to place MediaLB in a low-power state. Depending on the time the clock is shut off, a runt pulse can occur on MLBCLK.

[2] MLBCLK low/high time includes the pulse width variation.

[3] The MediaLB driver can release the MLBDAT/MLBSIG line as soon as MLBCLK is low; however, the logic state of the final driven bit on the line must remain on the bus for t_{mdzh}. Therefore, coupling must be minimized while meeting the maximum load capacitance listed.

Table 82 lists the MediaLB 6-pin interface timing characteristics, and Figure 87 shows the MLB 6-pin delay, setup, and hold times.

Table 82. MLB 6-Pin Interface Timing Parameters

Parameter	Symbol	Min	Max	Unit	Comment
Cycle-to-cycle system jitter	t_{jitter}	—	600	ps	—
Transmitter MLBSP/N (MLBDP/N) output valid from transition of MLBCP/N (low-to-high)[1]	t_{delay}	0.6	1.3	ns	
Disable turnaround time from transition of MLBCP/N (low-to-high)	t_{phz}	0.6	3.5	ns	
Enable turnaround time from transition of MLBCP/N (low-to-high)	t_{plz}	0.6	5.6	ns	
MLBSP/N (MLBDP/N) valid to transition of MLBCP/N (low-to-high)	t_{su}	0.05	—	ns	
MLBSP/N (MLBDP/N) hold from transition of MLBCP/N (low-to-high)[2]	t_{hd}	0.6			

[1] t_{delay}, t_{phz}, t_{plz}, t_{su}, and t_{hd} may also be referenced from a low-to-high transition of the recovered clock for 2:1 and 4:1 recovered-to-external clock ratios.

[2] The transmitting device must ensure valid data on MLBSP/N (MLBDP/N) for at least $t_{hd(min)}$ following the rising edge of MLBCP/N; receivers must latch MLBSP/N (MLBDP/N) data within $t_{hd(min)}$ of the rising edge of MLBCP/N.

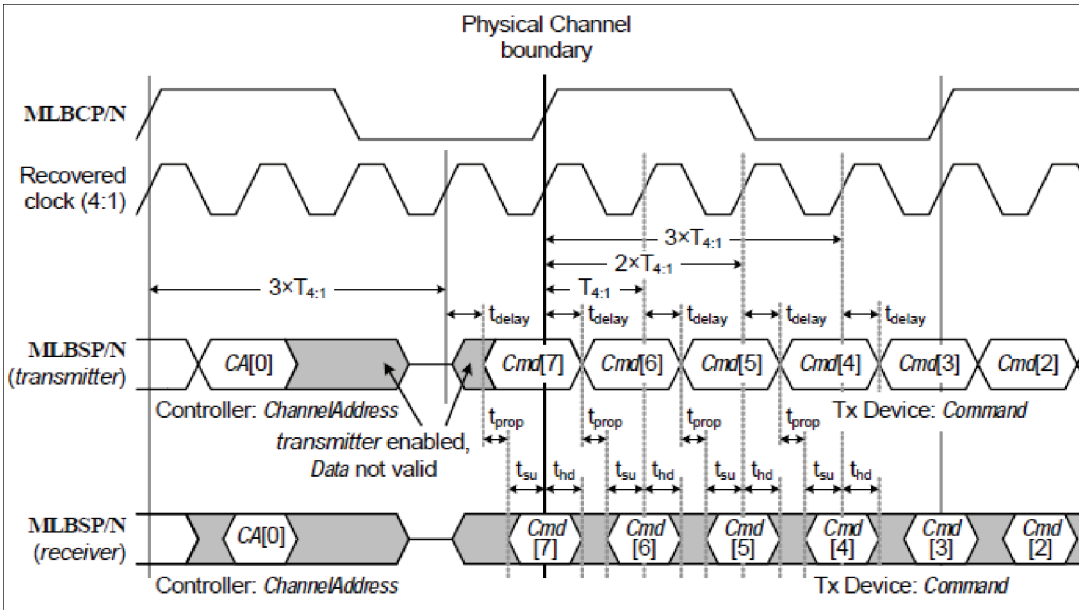


Figure 87. MLB 6-Pin Delay, Setup, and Hold Times

4.11.15 PCIe PHY Parameters

The PCIe interface complies with PCIe specification Gen2 x1 lane and supports the PCI Express 1.1/2.0 standard.

4.11.15.1 PCIE_REXT Reference Resistor Connection

The impedance calibration process requires connection of reference resistor 200 Ω. 1% precision resistor on PCIE_REXT pads to ground. It is used for termination impedance calibration.

4.11.16 Pulse Width Modulator (PWM) Timing Parameters

This section describes the electrical information of the PWM. The PWM can be programmed to select one of three clock signals as its source frequency. The selected clock signal is passed through a prescaler before being input to the counter. The output is available at the pulse-width modulator output (PWMO) external pin.

Figure 88 depicts the timing of the PWM, and Table 83 lists the PWM timing parameters.

NOTE

- All the timings for the SSI are given for a non-inverted serial clock polarity (TSCKP/RSCKP = 0) and a non-inverted frame sync (TFSI/RFSI = 0). If the polarity of the clock and/or the frame sync have been inverted, all the timing remains valid by inverting the clock signal STCK/SRCK and/or the frame sync STFS/SRFS shown in the tables and in the figures.
- All timings are on Audiomux Pads when SSI is being used for data transfer.
- "Tx" and "Rx" refer to the Transmit and Receive sections of the SSI.
- The terms, WL and BL, refer to Word Length (WL) and Bit Length (BL).
- For internal Frame Sync operation using external clock, the FS timing is same as that of Tx Data (for example, during AC97 mode of operation).

4.11.20 UART I/O Configuration and Timing Parameters

4.11.20.1 UART RS-232 I/O Configuration in Different Modes

The i.MX 6Solo/6DualLite UART interfaces can serve both as DTE or DCE device. This can be configured by the DCEDTE control bit (default 0 — DCE mode). Table 91 shows the UART I/O configuration based on the enabled mode.

Table 91. UART I/O Configuration vs. Mode

Port	DTE Mode		DCE Mode	
	Direction	Description	Direction	Description
RTS	Output	RTS from DTE to DCE	Input	RTS from DTE to DCE
CTS	Input	CTS from DCE to DTE	Output	CTS from DCE to DTE
DTR	Output	DTR from DTE to DCE	Input	DTR from DTE to DCE
DSR	Input	DSR from DCE to DTE	Output	DSR from DCE to DTE
DCD	Input	DCD from DCE to DTE	Output	DCD from DCE to DTE
RI	Input	RING from DCE to DTE	Output	RING from DCE to DTE
TXD_MUX	Input	Serial data from DCE to DTE	Output	Serial data from DCE to DTE
RXD_MUX	Output	Serial data from DTE to DCE	Input	Serial data from DTE to DCE

4.11.20.2 UART RS-232 Serial Mode Timing

The following sections describe the electrical information of the UART module in the RS-232 mode.

4.11.20.2.1 UART Transmitter

Figure 99 depicts the transmit timing of UART in the RS-232 serial mode, with 8 data bit/1 stop bit format. Table 92 lists the UART RS-232 serial mode transmit timing characteristics.

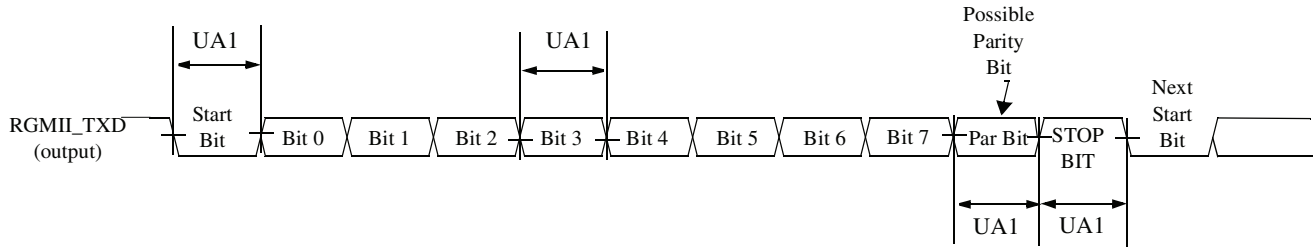


Figure 99. UART RS-232 Serial Mode Transmit Timing Diagram

Table 92. RS-232 Serial Mode Transmit Timing Parameters

ID	Parameter	Symbol	Min	Max	Unit
UA1	Transmit Bit Time	t_{Tbit}	$1/F_{baud\_rate}$[1] - T_{ref_clk}[2]	$1/F_{baud\_rate} + T_{ref\_clk}$	—

[1] F_{baud_rate}: Baud rate frequency. The maximum baud rate the UART can support is (*ipg_perclk* frequency)/16.

[2] T_{ref_clk}: The period of UART reference clock *ref_clk* (*ipg_perclk* after RFDIV divider).

4.11.20.2.2 UART Receiver

Figure 100 depicts the RS-232 serial mode receive timing with 8 data bit/1 stop bit format. Table 93 lists serial mode receive timing characteristics.

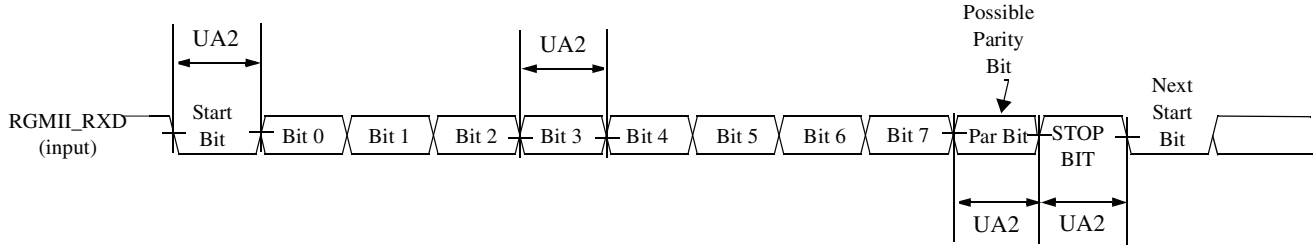


Figure 100. UART RS-232 Serial Mode Receive Timing Diagram

Table 93. RS-232 Serial Mode Receive Timing Parameters

ID	Parameter	Symbol	Min	Max	Unit
UA2	Receive Bit Time[1]	t_{Rbit}	$1/F_{baud\_rate}$[2] - $1/(16 \times F_{baud\_rate})$	$1/F_{baud\_rate} + 1/(16 \times F_{baud\_rate})$	—

[1] The UART receiver can tolerate $1/(16 \times F_{baud\_rate})$ tolerance in each bit. But accumulation tolerance in one frame must not exceed $3/(16 \times F_{baud\_rate})$.

[2] F_{baud_rate}: Baud rate frequency. The maximum baud rate the UART can support is (*ipg_perclk* frequency)/16.

Table 101. 21 x 21 mm Functional Contact Assignments[1] (continued)

Ball Name	Ball	Power Group	Ball Type	Out of Reset Condition[2]			
				Default Mode (Reset Mode)	Default Function	Input/ Output	Value
LVDS0_TX1_N	U4	NVCC_LVDS2P5					
LVDS0_TX1_P	U3	NVCC_LVDS2P5		ALT0	ldb.LVDS0_TX1	Input	Keeper
LVDS0_TX2_N	V2	NVCC_LVDS2P5					
LVDS0_TX2_P	V1	NVCC_LVDS2P5		ALT0	ldb.LVDS0_TX2	Input	Keeper
LVDS0_TX3_N	W2	NVCC_LVDS2P5					
LVDS0_TX3_P	W1	NVCC_LVDS2P5		ALT0	ldb.LVDS0_TX3	Input	Keeper
LVDS1_CLK_N	Y3	NVCC_LVDS2P5					
LVDS1_CLK_P	Y4	NVCC_LVDS2P5		ALT0	ldb.LVDS1_CLK	Input	Keeper
LVDS1_TX0_N	Y1	NVCC_LVDS2P5					
LVDS1_TX0_P	Y2	NVCC_LVDS2P5		ALT0	ldb.LVDS1_TX0	Input	Keeper
LVDS1_TX1_N	AA2	NVCC_LVDS2P5					
LVDS1_TX1_P	AA1	NVCC_LVDS2P5		ALT0	ldb.LVDS1_TX1	Input	Keeper
LVDS1_TX2_N	AB1	NVCC_LVDS2P5					
LVDS1_TX2_P	AB2	NVCC_LVDS2P5		ALT0	ldb.LVDS1_TX2	Input	Keeper
LVDS1_TX3_N	AA3	NVCC_LVDS2P5					
LVDS1_TX3_P	AA4	NVCC_LVDS2P5		ALT0	ldb.LVDS1_TX3	Input	Keeper
MLB_CN	A11	VDDHIGH_CAP					
MLB_CP	B11	VDDHIGH_CAP					
MLB_DN	B10	VDDHIGH_CAP					
MLB_DP	A10	VDDHIGH_CAP					
MLB_SN	A9	VDDHIGH_CAP					
MLB_SP	B9	VDDHIGH_CAP					
NANDF_ALE	A16	NVCC_NANDF	GPIO	ALT5	gpio6.GPIO[8]	Input	100 kΩ pull-up
NANDF_CLE	C15	NVCC_NANDF	GPIO	ALT5	gpio6.GPIO[7]	Input	100 kΩ pull-up
NANDF_CS0	F15	NVCC_NANDF	GPIO	ALT5	gpio6.GPIO[11]	Input	100 kΩ pull-up
NANDF_CS1	C16	NVCC_NANDF	GPIO	ALT5	gpio6.GPIO[14]	Input	100 kΩ pull-up
NANDF_CS2	A17	NVCC_NANDF	GPIO	ALT5	gpio6.GPIO[15]	Input	100 kΩ pull-up
NANDF_CS3	D16	NVCC_NANDF	GPIO	ALT5	gpio6.GPIO[16]	Input	100 kΩ pull-up
NANDF_D0	A18	NVCC_NANDF	GPIO	ALT5	gpio2.GPIO[0]	Input	100 kΩ pull-up
NANDF_D1	C17	NVCC_NANDF	GPIO	ALT5	gpio2.GPIO[1]	Input	100 kΩ pull-up

Table 101. 21 x 21 mm Functional Contact Assignments[1] (continued)

Ball Name	Ball	Power Group	Ball Type	Out of Reset Condition[2]			
				Default Mode (Reset Mode)	Default Function	Input/ Output	Value
SD4_DAT3	A20	NVCC_NANDF	GPIO	ALT5	gpio2.GPIO[11]	Input	100 kΩ pull-up
SD4_DAT4	E18	NVCC_NANDF	GPIO	ALT5	gpio2.GPIO[12]	Input	100 kΩ pull-up
SD4_DAT5	C19	NVCC_NANDF	GPIO	ALT5	gpio2.GPIO[13]	Input	100 kΩ pull-up
SD4_DAT6	B20	NVCC_NANDF	GPIO	ALT5	gpio2.GPIO[14]	Input	100 kΩ pull-up
SD4_DAT7	D19	NVCC_NANDF	GPIO	ALT5	gpio2.GPIO[15]	Input	100 kΩ pull-up
TAMPER	E11	VDD_SNVS_IN	GPIO	ALT0	snvs_lp_wrapper.SNVS_TD1	Input	100 kΩ pull-down
TEST_MODE	E12	VDD_SNVS_IN	GPIO	ALT0	tcu.TEST_MODE	Input	100 kΩ pull-down
USB_H1_DN	F10	VDDUSB_CAP					
USB_H1_DP	E10	VDDUSB_CAP					
USB_OTG_CHD_B	B8	VDDUSB_CAP					
USB_OTG_DN	B6	VDDUSB_CAP					
USB_OTG_DP	A6	VDDUSB_CAP					
XTALI	A7	NVCC_PLL_OUT					
XTALO	B7	NVCC_PLL_OUT					
NC	A1						
NC	A12						
NC	A14						
NC	B12						
NC	B14						
NC	E1						
NC	E2						
NC	F1						
NC	F2						

1 DRAM_D32 to DRAM_D63 are only available for i.MX 6DualLite chip; for i.MX 6Solo chip, these pins are NC.

2 The state immediately after reset and before ROM firmware or software has executed.

Table 103. 21 x 21 mm, 0.8 mm Pitch Ball Map (continued)

Y	W	V	U	T	R	P	N
LVDS1_TX0_N	LVDS0_TX3_P	LVDS0_TX2_P	LVDS0_TX0_P	GPIO_2	GPIO_17	CSI0_PIXCLK	CSI0_DAT4
LVDS1_TX0_P	LVDS0_TX3_N	LVDS0_TX2_N	LVDS0_TX0_N	GPIO_9	GPIO_16	CSI0_DAT5	CSI0_VSYNC
LVDS1_CLK_N	GND	LVDS0_CLK_P	LVDS0_TX1_P	GPIO_6	GPIO_7	CSI0_DATA_EN	CSI0_DAT7
LVDS1_CLK_P	KEY_ROW2	LVDS0_CLK_N	LVDS0_TX1_N	GPIO_1	GPIO_5	CSI0_MCLK	CSI0_DAT6
GND	KEY_COL0	KEY_ROW4	KEY_COL3	GPIO_0	GPIO_8	GPIO_19	CSI0_DAT9
DRAM_RESET	KEY_COL2	KEY_ROW0	KEY_ROW1	KEY_COL4	GPIO_4	GPIO_18	CSI0_DAT8
DRAM_D20	GND	NVCC_LVDS2P5	KEY_COL1	KEY_ROW3	GPIO_3	NVCC_GPIO	NVCC_CSI
DRAM_D21	GND	GND	GND	GND	GND	GND	GND
DRAM_D19	GND	NVCC_DRAM	VDDARM_IN	VDDARM_IN	VDDARM_IN	VDDARM_IN	VDDARM_IN
DRAM_D25	GND	NVCC_DRAM	VDDSOC_CAP	VDDSOC_CAP	VDDSOC_CAP	GND	GND
DRAM_SDCKE0	GND	NVCC_DRAM	GND	GND	VDDARM_CAP	VDDARM_CAP	VDDARM_CAP
DRAM_A15	GND	NVCC_DRAM	GND	GND	GND	GND	NC
DRAM_A7	GND	NVCC_DRAM	VDDSOC_CAP	VDDSOC_CAP	VDDARM_CAP	VDDARM_CAP	VDDARM_CAP
DRAM_A3	DRAM_A4	NVCC_DRAM	VDDSOC_CAP	VDDSOC_CAP	VDDARM_IN	VDDARM_IN	VDDARM_IN
DRAM_SDBA1	GND	NVCC_DRAM	GND	GND	GND	GND	GND
DRAM_CS0	GND	NVCC_DRAM	VDDSOC_IN	VDDSOC_IN	VDDSOC_IN	VDDSOC_IN	VDDSOC_IN
DRAM_D36	GND	NVCC_DRAM	GND	GND	GND	VDDPU_CAP	VDDPU_CAP
DRAM_D37	GND	NVCC_DRAM	NVCC_DRAM	NVCC_DRAM	NVCC_DRAM	GND	GND
DRAM_D40	GND	GND	GND	GND	NVCC_ENET	NVCC_LCD	DI0_DISP_CLK
DRAM_D44	ENET_TXD1	ENET_MDC	ENET_TXD0	DISP0_DAT21	DISP0_DAT13	DISP0_DAT4	DI0_PIN3
DRAM_DQM7	ENET_RXD0	ENET_TX_EN	ENET_CRS_DV	DISP0_DAT16	DISP0_DAT10	DISP0_DAT3	DI0_PIN15
DRAM_D59	ENET_RXD1	ENET_REF_CLK	DISP0_DAT20	DISP0_DAT15	DISP0_DAT8	DISP0_DAT1	EIM_BCLK
DRAM_D62	ENET_RX_ER	ENET_MDIO	DISP0_DAT19	DISP0_DAT11	DISP0_DAT6	DISP0_DAT2	EIM_DA14
GND	DISP0_DAT23	DISP0_DAT22	DISP0_DAT17	DISP0_DAT12	DISP0_DAT7	DISP0_DAT0	EIM_DA15
DRAM_D58	DRAM_D63	DISP0_DAT18	DISP0_DAT14	DISP0_DAT9	DISP0_DAT5	DI0_PIN4	DI0_PIN2
Y	W	V	U	T	R	P	N