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Understanding [Embedded - Microprocessors](#)

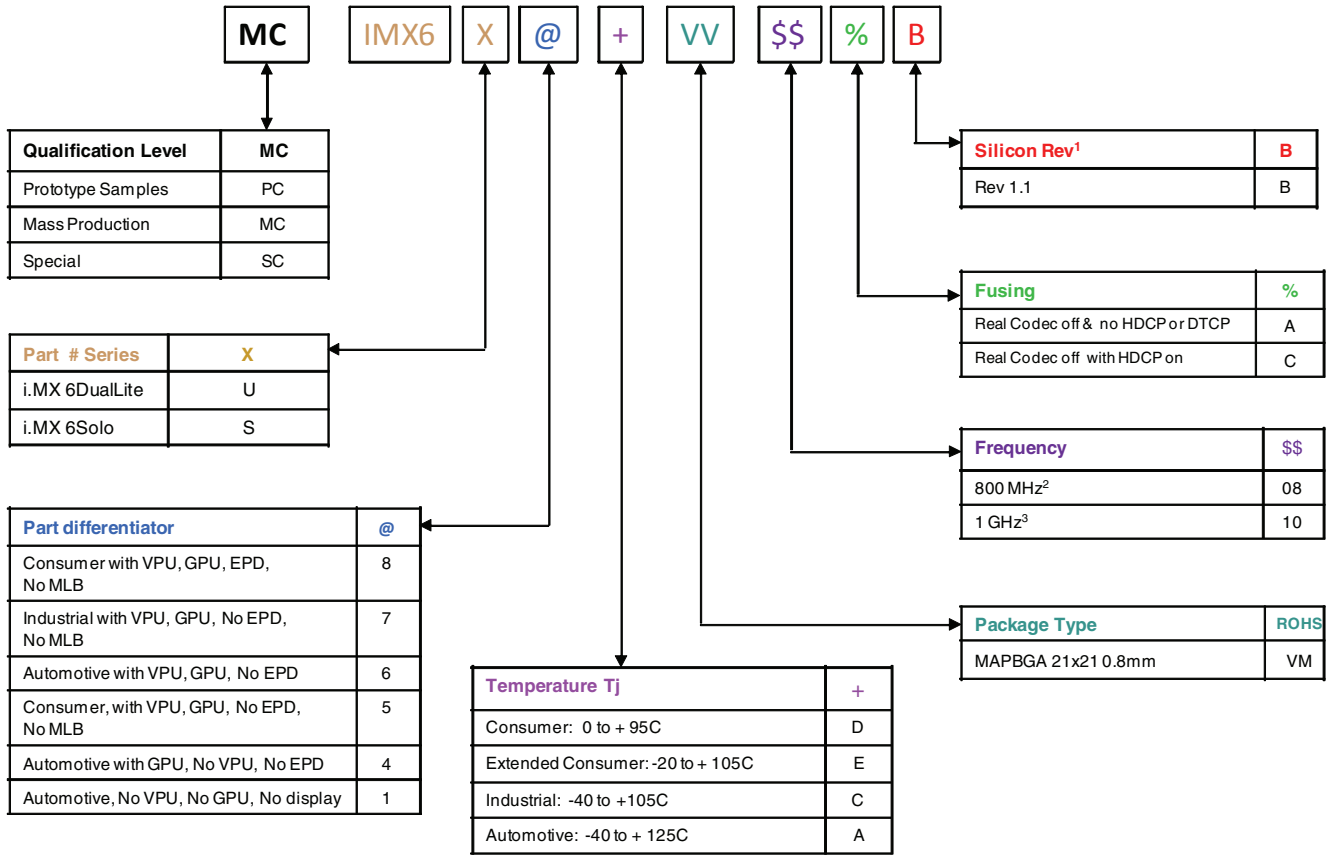
Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Active
Core Processor	ARM® Cortex®-A9
Number of Cores/Bus Width	2 Core, 32-Bit
Speed	800MHz
Co-Processors/DSP	Multimedia; NEON™ SIMD
RAM Controllers	LPDDR2, LVDDR3, DDR3
Graphics Acceleration	Yes
Display & Interface Controllers	Keypad, LCD
Ethernet	10/100/1000Mbps (1)
SATA	-
USB	USB 2.0 + PHY (4)
Voltage - I/O	1.8V, 2.5V, 2.8V, 3.3V
Operating Temperature	-40°C ~ 125°C (TJ)
Security Features	ARM TZ, Boot Security, Cryptography, RTIC, Secure Fusebox, Secure JTAG, Secure Memory, Secure RTC, Tamper Detection
Package / Case	624-LFBGA
Supplier Device Package	624-MAPBGA (21x21)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mcimx6u4avm08acr



1. See the [freescale.com\imx6series](http://freescale.com/imx6series) Web page for latest information on the available silicon revision.
2. If a 24 MHz input clock is used (required for USB), the maximum SoC speed is limited to 792 MHz.
3. If a 24 MHz input clock is used (required for USB), the maximum SoC speed is limited to 996 MHz.

Figure 1. Part Number Nomenclature—i.MX 6DualLite and 6Solo

1.2 Features

The i.MX 6Solo/6DualLite processors are based on ARM Cortex-A9 MPCore™ Platform, which has the following features:

- The i.MX 6Solo supports single ARM Cortex-A9 MPCore (with TrustZone)
- The i.MX 6DualLite supports dual ARM Cortex-A9 MPCore (with TrustZone)
- The core configuration is symmetric, where each core includes:
 - 32 KByte L1 Instruction Cache
 - 32 KByte L1 Data Cache
 - Private Timer and Watchdog
 - Cortex-A9 NEON MPE (Media Processing Engine) Co-processor

The ARM Cortex-A9 MPCore complex includes:

- General Interrupt Controller (GIC) with 128 interrupt support
- Global Timer

NOTE

The numbers in brackets indicate number of module instances. For example, PWM (4) indicates four separate PWM peripherals.

3 Modules List

The i.MX 6Solo/6DualLite processors contain a variety of digital and analog modules. [Table 2](#) describes these modules in alphabetical order.

Table 2. i.MX 6Solo/6DualLite Modules List

Block Mnemonic	Block Name	Subsystem	Brief Description
ARM	ARM Platform	ARM	The ARM Core Platform includes 1x (Solo) Cortex-A9 core for i.MX 6Solo and 2x (Dual) Cortex-A9 cores for i.MX 6DualLite. It also includes associated sub-blocks, such as the Level 2 Cache Controller, SCU (Snoop Control Unit), GIC (General Interrupt Controller), private timers, watchdog, and CoreSight debug modules.
APBH-DMA	NAND Flash and BCH ECC DMA controller	System Control Peripherals	DMA controller used for GPMMI2 operation
ASRC	Asynchronous Sample Rate Converter	Multimedia Peripherals	The Asynchronous Sample Rate Converter (ASRC) converts the sampling rate of a signal associated to an input clock into a signal associated to a different output clock. The ASRC supports concurrent sample rate conversion of up to 10 channels of about -120dB THD+N. The sample rate conversion of each channel is associated to a pair of incoming and outgoing sampling rates. The ASRC supports up to three sampling rate pairs.
AUDMUX	Digital Audio Mux	Multimedia Peripherals	The AUDMUX is a programmable interconnect for voice, audio, and synchronous data routing between host serial interfaces (for example, SSI1, SSI2, and SSI3) and peripheral serial interfaces (audio and voice codecs). The AUDMUX has seven ports with identical functionality and programming models. A desired connectivity is achieved by configuring two or more AUDMUX ports.
BCH40	Binary-BCH ECC Processor	System Control Peripherals	The BCH40 module provides up to 40-bit ECC encryption/decryption for NAND Flash controller (GPMMI)
CAAM	Cryptographic accelerator and assurance module	Security	CAAM is a cryptographic accelerator and assurance module. CAAM implements several encryption and hashing functions, a run-time integrity checker, and a Pseudo Random Number Generator (PRNG). The pseudo random number generator is certified by Cryptographic Algorithm Validation Program (CAVP) of National Institute of Standards and Technology (NIST). Its DRBG validation number is 94 and its SHS validation number is 1455. CAAM also implements a Secure Memory mechanism. In i.MX 6Solo/6DualLite processors, the security memory provided is 16 KB.

Table 2. i.MX 6Solo/6DualLite Modules List (continued)

Block Mnemonic	Block Name	Subsystem	Brief Description
IOMUXC	IOMUX Control	System Control Peripherals	This module enables flexible IO multiplexing. Each IO pad has default and several alternate functions. The alternate functions are software configurable.
IPUV3H	Image Processing Unit, ver.3H	Multimedia Peripherals	IPUV3H enables connectivity to displays and video sources, relevant processing and synchronization and control capabilities, allowing autonomous operation. The IPUv3H supports concurrent output to two display ports and concurrent input from two camera ports, through the following interfaces: <ul style="list-style-type: none"> • Parallel Interfaces for both display and camera • Single/dual channel LVDS display interface • HDMI transmitter • MIPI/DSI transmitter • MIPI/CSI-2 receiver The processing includes: <ul style="list-style-type: none"> • Image conversions: resizing, rotation, inversion, and color space conversion • A high-quality de-interlacing filter • Video/graphics combining • Image enhancement: color adjustment and gamut mapping, gamma correction, and contrast enhancement • Support for display backlight reduction
KPP	Key Pad Port	Connectivity Peripherals	KPP Supports 8x8 external key pad matrix. KPP features are: <ul style="list-style-type: none"> • Open drain design • Glitch suppression circuit design • Multiple keys detection • Standby key press detection
LDB	LVDS Display Bridge	Connectivity Peripherals	LVDS Display Bridge is used to connect the IPU (Image Processing Unit) to External LVDS Display Interface. LDB supports two channels; each channel has following signals: <ul style="list-style-type: none"> • One clock pair • Four data pairs Each signal pair contains LVDS special differential pad (PadP, PadM).
MLB150	MediaLB	Connectivity / Multimedia Peripherals	The MLB interface module provides a link to a MOST [®] data network, using the standardized MediaLB protocol (up to 6144 fs). The module is backward compatible to MLB-50.
MMDC	Multi-Mode DDR Controller	Connectivity Peripherals	DDR Controller has the following features: <ul style="list-style-type: none"> • Supports 16/32-bit DDR3-800 (LV) or LPDDR2-800 in i.MX 6Solo • Supports 16/32/64-bit DDR3-800 (LV) or LPDDR2-800 in i.MX 6DualLite • Supports 2x32 LPDDR2-800 in i.MX 6DualLite • Supports up to 4 GByte DDR memory space

their input supply ripple rejection and their on-die trimming. This translates into more stable voltage for the on-chip logics.

These regulators have three basic modes:

- Bypass. The regulation FET is switched fully on passing the external voltage, to the load unaltered. The analog part of the regulator is powered down in this state, removing any loss other than the IR drop through the power grid and FET.
- Power Gate. The regulation FET is switched fully off limiting the current draw from the supply. The analog part of the regulator is powered down here limiting the power consumption.
- Analog regulation mode. The regulation FET is controlled such that the output voltage of the regulator equals the programmed target voltage. The target voltage is fully programmable in 25 mV steps.

For additional information, see the i.MX 6Solo/6DualLite reference manual.

4.3.2 Analog Regulators

4.3.2.1 LDO_1P1

The LDO_1P1 regulator implements a programmable linear-regulator function from VDDHIGH_IN (see [Table 9](#) for minimum and maximum input requirements). Typical Programming Operating Range is 1.0 V to 1.2 V with the nominal default setting as 1.1 V. Since the accuracy or the % regulation is not tested, and only tested with the LDO set to either 1.0V or 1.2V, this is the only range that is guaranteed. The regulator has been designed to be stable with a minimum external low ESR decoupling capacitor of 1 μ F (2.2 μ F should be considered the recommended minimum value for component selection), though the actual capacitance required should be determined by the application. A programmable brown-out detector is included in the regulator that can be used by the system to determine when the load capability of the regulator is being exceeded to take the necessary steps. Current-limiting can be enabled to allow for in-rush current requirements during start-up, if needed. Active-pull-down can also be enabled for systems requiring this feature.

For additional information, see the i.MX 6Solo/6DualLite reference manual.

4.3.2.2 LDO_2P5

The LDO_2P5 module implements a programmable linear-regulator function from VDDHIGH_IN (see [Table 9](#) for minimum and maximum input requirements). Typical Programming Operating Range is 2.25 V to 2.75 V with the nominal default setting as 2.5 V. Since the accuracy or the % regulation is not tested, and only tested with the LDO set to either 2.25V or 2.75V, this is the only range that is guaranteed. The regulator has been designed to be stable with a minimum external low ESR decoupling capacitor of 1 μ F (2.2 μ F should be considered the recommended minimum value for component selection), though the actual capacitance required should be determined by the application. A programmable brown-out detector is included in the regulator that can be used by the system to determine when the load capability of the regulator is being exceeded, to take the necessary steps. Current-limiting can be enabled to allow for in-rush current requirements during start-up, if needed. Active-pull-down can also be enabled for systems requiring this feature. An alternate self-biased low-precision weak-regulator is included that can be

- ² To maintain a valid level, the transition edge of the input must sustain a constant slew rate (monotonic) from the current DC level through to the target DC level, V_{il} or V_{ih} . Monotonic input transition time is from 0.1 ns to 1 s.
- ³ Hysteresis of 250 mV is guaranteed over all operating conditions when hysteresis is enabled.

4.6.2 DDR I/O DC Parameters

The DDR I/O pads support LPDDR2 and DDR3/DDR3L operational modes.

4.6.2.1 LPDDR2 Mode I/O DC Parameters

The LPDDR2 interface mode fully complies with JESD209-2B LPDDR2 JEDEC standard release June, 2009.

Table 25. LPDDR2 I/O DC Electrical Parameters¹

Parameters	Symbol	Test Conditions	Min	Max	Unit
High-level output voltage	V_{OH}	$I_{oh} = -0.1\text{mA}$	$0.9 \cdot OVDD$		V
Low-level output voltage	V_{OL}	$I_{ol} = 0.1\text{mA}$		$0.1 \cdot OVDD$	V
Input Reference Voltage	V_{ref}		$0.49 \cdot OVDD$	$0.51 \cdot OVDD$	V
DC High-Level input voltage	V_{ih_DC}		$V_{ref} + 0.13$	$OVDD$	V
DC Low-Level input voltage	V_{il_DC}		$OVSS$	$V_{ref} - 0.13$	V
Differential Input Logic High	V_{ih_diff}		0.26	Note ²	
Differential Input Logic Low	V_{il_diff}		Note ³	-0.26	
Pull-up/Pull-down Impedance Mismatch	$Mmpupd$		-15	15	%
240 Ω unit calibration resolution	R_{res}			10	Ω
Keeper Circuit Resistance	R_{keep}		110	175	$k\Omega$
Input current (no pull-up/down)	I_{in}	$V_I = 0, V_I = OVDD$	-2.5	2.5	μA

¹ Note that the JEDEC LPDDR2 specification (JESD209_2B) supersedes any specification in this document.

² The single-ended signals need to be within the respective limits ($V_{ih}(dc)$ max, $V_{il}(dc)$ min) for single-ended signals as well as the limitations for overshoot and undershoot.

4.9.3.4 Examples of EIM Synchronous Accesses

Table 43. EIM Bus Timing Parameters ¹

ID	Parameter	BCD = 0		BCD = 1		BCD = 2		BCD = 3	
		Min	Max	Min	Max	Min	Max	Min	Max
WE1	BCLK Cycle time ²	t	—	2 x t		3 x t	—	4 x t	—
WE2	BCLK Low Level Width	0.4 x t	—	0.8 x t		1.2 x t	—	1.6 x t	—
WE3	BCLK High Level Width	0.4 x t	—	0.8 x t		1.2 x t	—	1.6 x t	—
WE4	Clock rise to address valid ³	-0.5 x t - 1.25	-0.5 x t + 1.75	-t - 1.25	-t + 1.75	-1.5 x t - 1.25	-1.5 x t + 1.75	-2 x t - 1.25	-2 x t + 1.75
WE5	Clock rise to address invalid	0.5 x t - 1.25	0.5 x t + 1.75	t - 1.25	t + 1.75	1.5 x t - 1.25	1.5 x t + 1.75	2 x t - 1.25	2 x t + 1.75
WE6	Clock rise to CSx_B valid	-0.5 x t - 1.25	-0.5 x t + 1.75	-t - 1.25	-t + 1.75	-1.5 x t - 1.25	-1.5 x t + 1.75	-2 x t - 1.25	-2 x t + 1.75
WE7	Clock rise to CSx_B invalid	0.5 x t - 1.25	0.5 x t + 1.75	t - 1.25	t + 1.75	1.5 x t - 1.25	1.5 x t + 1.75	2 x t - 1.25	2 x t + 1.75
WE8	Clock rise to WE_B Valid	-0.5 x t - 1.25	-0.5 x t + 1.75	-t - 1.25	-t + 1.75	-1.5 x t - 1.25	-1.5 x t + 1.75	-2 x t - 1.25	-2 x t + 1.75
WE9	Clock rise to WE_B Invalid	0.5 x t - 1.25	0.5 x t + 1.75	t - 1.25	t + 1.75	1.5 x t - 1.25	1.5 x t + 1.75	2 x t - 1.25	2 x t + 1.75
WE10	Clock rise to OE_B Valid	-0.5 x t - 1.25	-0.5 x t + 1.75	-t - 1.25	-t + 1.75	-1.5 x t - 1.25	-1.5 x t + 1.75	-2 x t - 1.25	-2 x t + 1.75
WE11	Clock rise to OE_B Invalid	0.5 x t - 1.25	0.5 x t + 1.75	t - 1.25	t + 1.75	1.5 x t - 1.25	1.5 x t + 1.75	2 x t - 1.25	2 x t + 1.75
WE12	Clock rise to BEy_B Valid	-0.5 x t - 1.25	-0.5 x t + 1.75	-t - 1.25	-t + 1.75	-1.5 x t - 1.25	-1.5 x t + 1.75	-2 x t - 1.25	-2 x t + 1.75
WE13	Clock rise to BEy_B Invalid	0.5 x t - 1.25	0.5 x t + 1.75	t - 1.25	t + 1.75	1.5 x t - 1.25	1.5 x t + 1.75	2 x t - 1.25	2 x t + 1.75
WE14	Clock rise to ADV_B Valid	-0.5 x t - 1.25	-0.5 x t + 1.75	-t - 1.25	-t + 1.75	-1.5 x t - 1.25	-1.5 x t + 1.75	-2 x t - 1.25	-2 x t + 1.75
WE15	Clock rise to ADV_B Invalid	0.5 x t - 1.25	0.5 x t + 1.75	t - 1.25	t + 1.75	1.5 x t - 1.25	1.5 x t + 1.75	2 x t - 1.25	2 x t + 1.75
WE16	Clock rise to Output Data Valid	-0.5 x t - 1.25	-0.5 x t + 1.75	-t - 1.25	-t + 1.75	-1.5 x t - 1.25	-1.5 x t + 1.75	-2 x t - 1.25	-2 x t + 1.75
WE17	Clock rise to Output Data Invalid	0.5 x t - 1.25	0.5 x t + 1.75	t - 1.25	t + 1.75	1.5 x t - 1.25	1.5 x t + 1.75	2 x t - 1.25	2 x t + 1.75
WE18	Input Data setup time to Clock rise	2	—	4	—	—	—	—	—

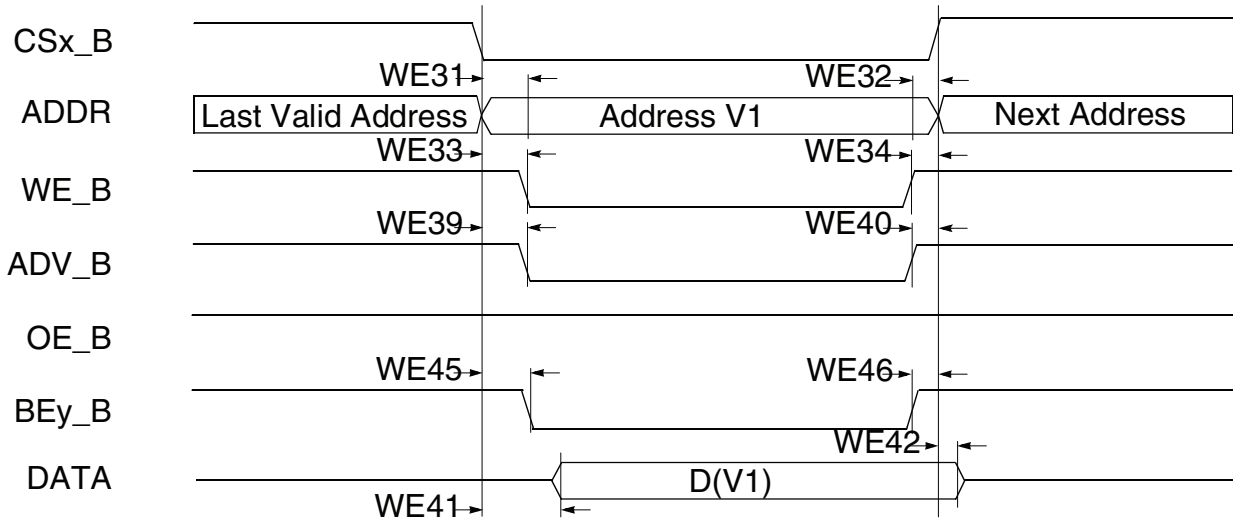


Figure 20. Asynchronous Memory Write Access

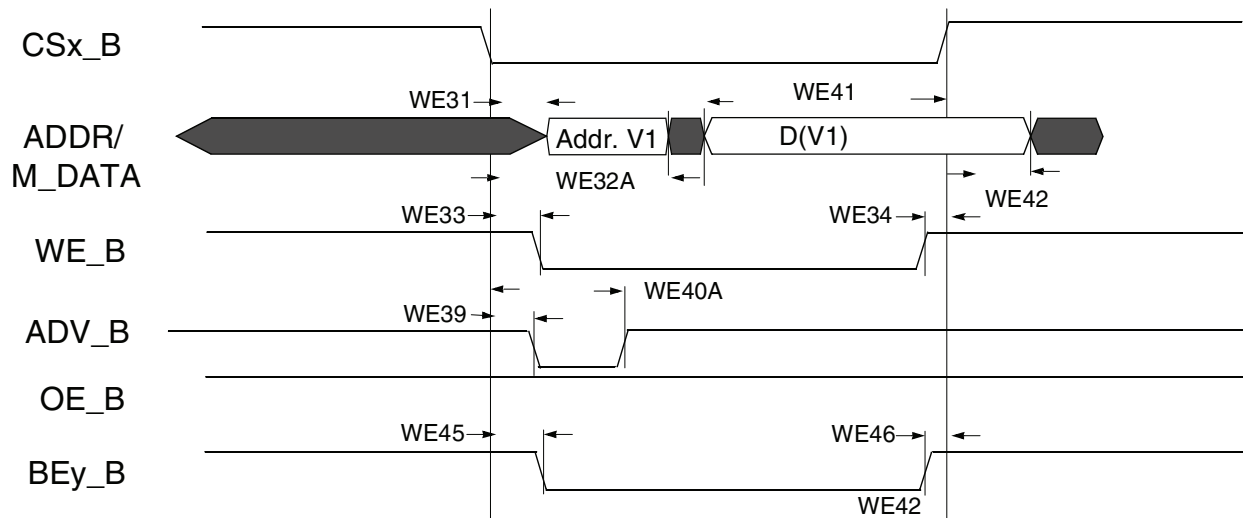


Figure 21. Asynchronous A/D Muxed Write Access

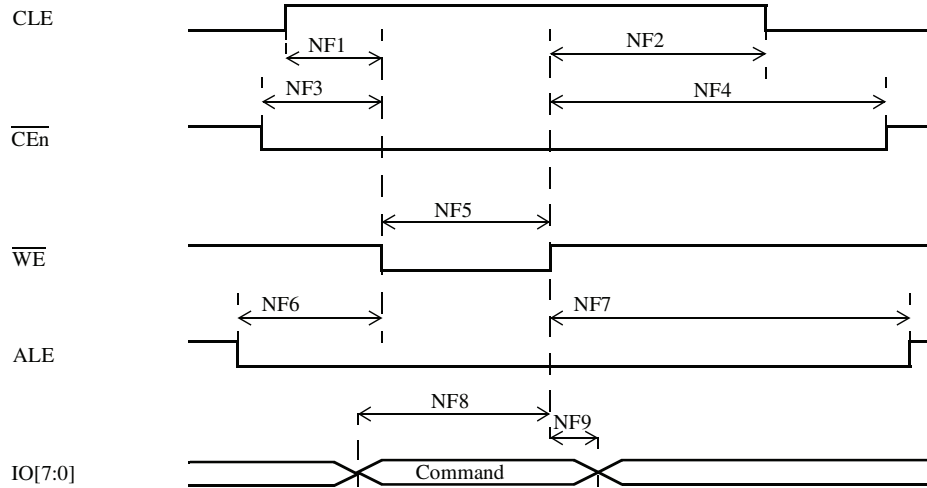


Figure 30. Command Latch Cycle Timing Diagram

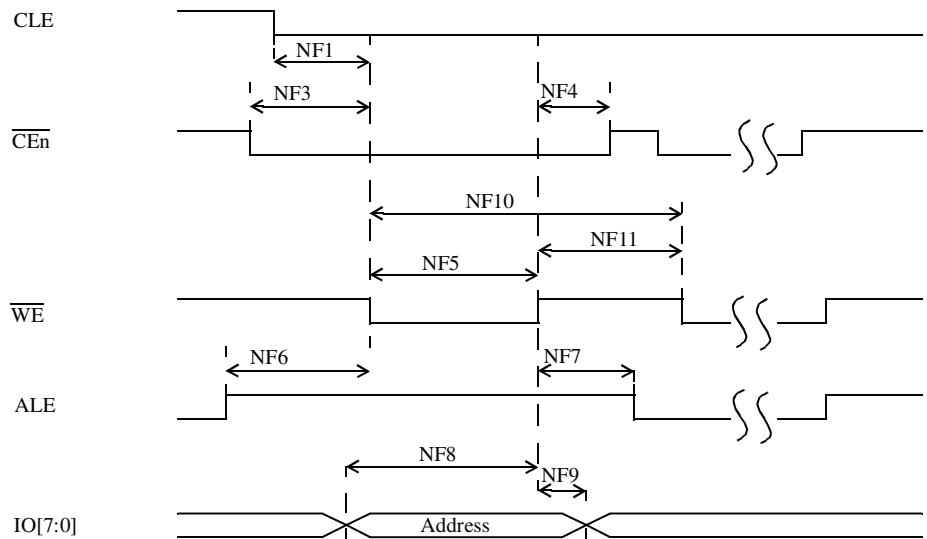


Figure 31. Address Latch Cycle Timing Diagram

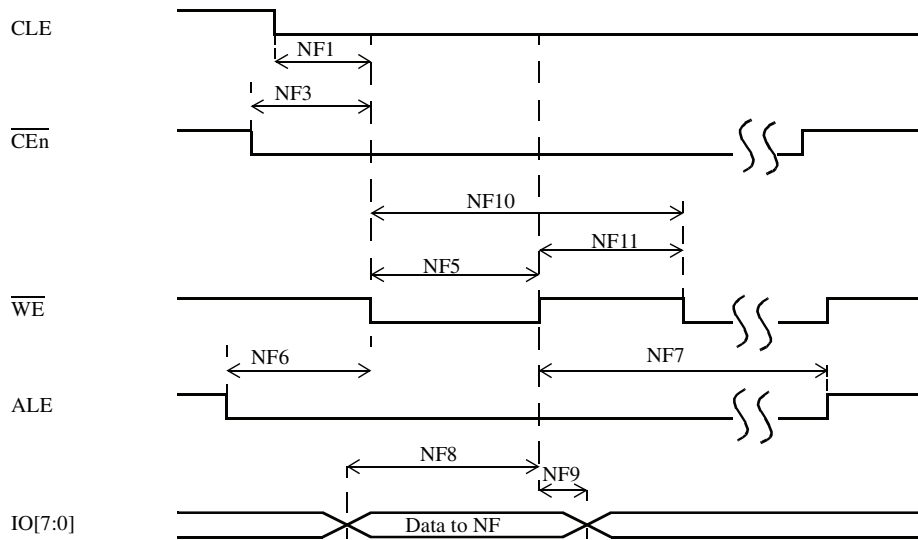


Figure 32. Write Data Latch Cycle Timing Diagram

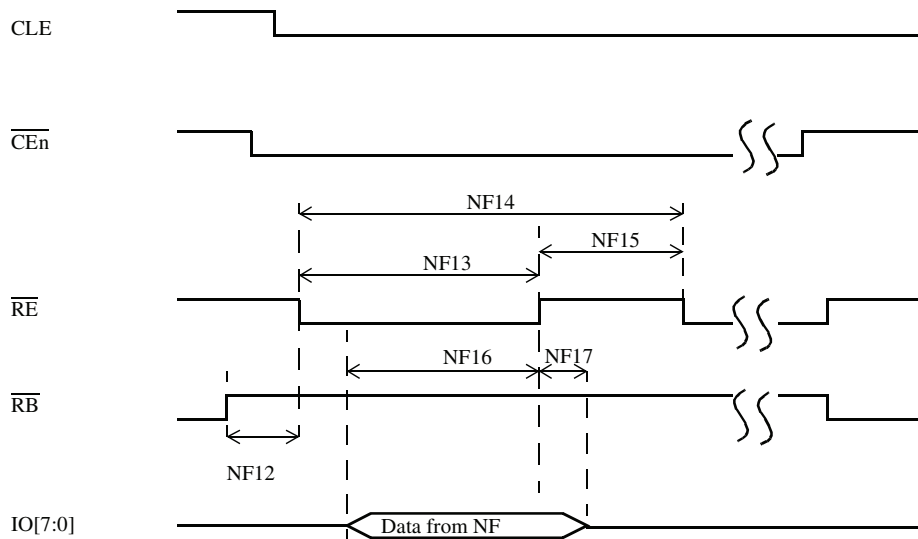


Figure 33. Read Data Latch Cycle Timing Diagram

Table 51. Asynchronous Mode Timing Parameters¹

ID	Parameter	Symbol	Timing T = GPMI Clock Cycle		Example Timing for GPMI Clock ≈ 100 MHz T = 10 ns		Unit
			Min.	Max.	Min.	Max.	
NF1	CLE setup time	tCLS	(AS+1) x T	—	10	—	ns
NF2	CLE hold time	tCLH	(DH+1) x T	—	20	—	ns
NF3	$\overline{\text{CEn}}$ setup time	tCS	(AS+1) x T	—	10	—	ns
NF4	$\overline{\text{CE}}$ hold time	tCH	(DH+1) x T	—	20	—	ns

4.11.3 Enhanced Serial Audio Interface (ESAI) Timing Parameters

The ESAI consists of independent transmitter and receiver sections, each section with its own clock generator. Table 56 shows the interface timing values. The number field in the table refers to timing signals found in Figure 42 and Figure 43.

Table 56. Enhanced Serial Audio Interface (ESAI) Timing

No.	Characteristics ^{1,2}	Symbol	Expression ²	Min	Max	Condition ³	Unit
62	Clock cycle ⁴	t_{SSICC}	$4 \times T_C$ $4 \times T_C$	30.0 30.0	— —	i ck i ck	ns
63	Clock high period: • For internal clock • For external clock	— —	$2 \times T_C - 9.0$ $2 \times T_C$	6 15	— —	— —	ns
64	Clock low period: • For internal clock • For external clock	— —	$2 \times T_C - 9.0$ $2 \times T_C$	6 15	— —	— —	ns
65	SCKR rising edge to FSR out (bl) high	— —	— —	— —	17.0 7.0	x ck i ck a	ns
66	SCKR rising edge to FSR out (bl) low	— —	— —	— —	17.0 7.0	x ck i ck a	ns
67	SCKR rising edge to FSR out (wr) high ⁵	— —	— —	— —	19.0 9.0	x ck i ck a	ns
68	SCKR rising edge to FSR out (wr) low ⁵	— —	— —	— —	19.0 9.0	x ck i ck a	ns
69	SCKR rising edge to FSR out (wl) high	— —	— —	— —	16.0 6.0	x ck i ck a	ns
70	SCKR rising edge to FSR out (wl) low	— —	— —	— —	17.0 7.0	x ck i ck a	ns
71	Data in setup time before SCKR (SCK in synchronous mode) falling edge	— —	— —	12.0 19.0	— —	x ck i ck	ns
72	Data in hold time after SCKR falling edge	— —	— —	3.5 9.0	— —	x ck i ck	ns
73	FSR input (bl, wr) high before SCKR falling edge ⁵	— —	— —	2.0 12.0	— —	x ck i ck a	ns
74	FSR input (wl) high before SCKR falling edge	— —	— —	2.0 12.0	— —	x ck i ck a	ns
75	FSR input hold time after SCKR falling edge	— —	— —	2.5 8.5	— —	x ck i ck a	ns
78	SCKT rising edge to FST out (bl) high	— —	— —	— —	18.0 8.0	x ck i ck	ns
79	SCKT rising edge to FST out (bl) low	— —	— —	— —	20.0 10.0	x ck i ck	ns
80	SCKT rising edge to FST out (wr) high ⁵	— —	— —	— —	20.0 10.0	x ck i ck	ns

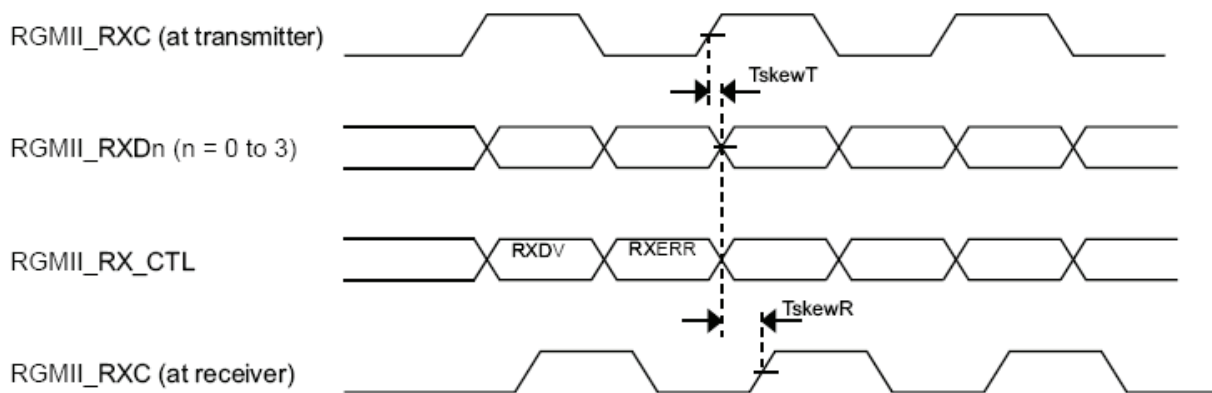


Figure 53. RGMII Receive Signal Timing Diagram Original

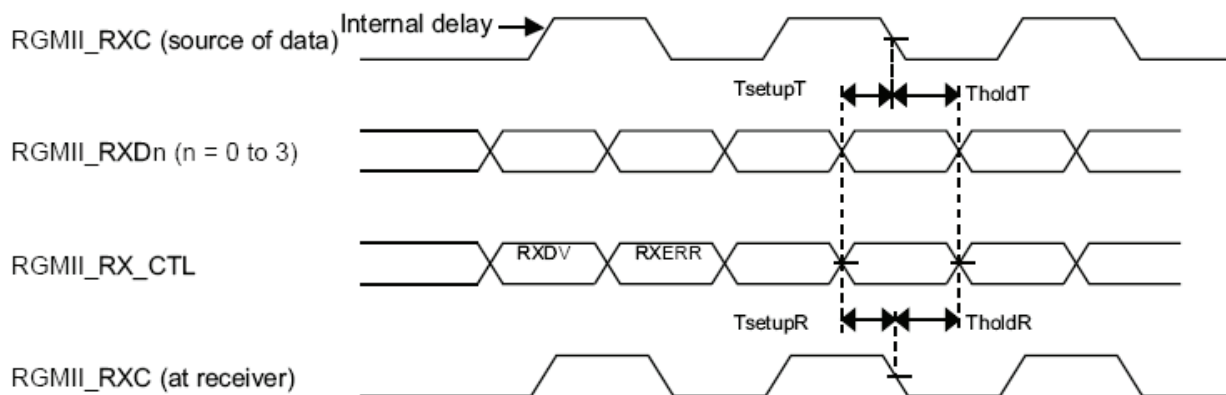


Figure 54. RGMII Receive Signal Timing Diagram with Internal Delay

4.11.6 Flexible Controller Area Network (FLEXCAN) AC Electrical Specifications

The Flexible Controller Area Network (FlexCAN) module is a communication controller implementing the CAN protocol according to the CAN 2.0B protocol specification. The processor has two CAN modules available for systems design. Tx and Rx ports for both modules are multiplexed with other I/O pins. See the IOMUXC chapter of the *i.MX 6Solo/6DualLite Reference Manual (IMX6SDLRM)* to see which pins expose Tx and Rx pins; these ports are named TXCAN and RXCAN, respectively.

4.11.7 HDMI Module Timing Parameters

4.11.7.1 Latencies and Timing Information

Power-up time (time between TX_PWRON assertion and TX_READY assertion) for the HDMI 3D Tx PHY while operating with the slowest input reference clock supported (13.5 MHz) is 3.35 ms.

Table 66. Electrical Characteristics

Symbol	Parameter	Condition	Min.	Typ.	Max.	Unit
Operating conditions for HDMI						
avddtmds	Termination supply voltage	-	3.15	3.3	3.45	V
R_T	Termination resistance	-	45	50	55	Ω
TMDS drivers DC specifications						
V_{OFF}	Single-ended standby voltage	$R_T = 50 \Omega$	$avddtmds \pm 10 \text{ mV}$			mV
V_{SWING}	Single-ended output swing voltage	For measurement conditions and definitions, see the first two figures above. Compliance point TP1 as defined in the HDMI specification, version 1.3a, section 4.2.4.	400	-	600	mV
V_H	Single-ended output high voltage For definition, see the second figure above	If attached sink supports TMDSCCLK < or = 165 MHz	$avddtmds \pm 10 \text{ mV}$			mV
		If attached sink supports TMDSCCLK > 165 MHz	$avddtmds - 200 \text{ mV}$	-	$avddtmds + 10 \text{ mV}$	mV
V_L	Single-ended output low voltage For definition, see the second figure above	If attached sink supports TMDSCCLK < or = 165 MHz	$avddtmds - 600 \text{ mV}$	-	$avddtmds - 400 \text{ mV}$	mV
		If attached sink supports TMDSCCLK > 165 MHz	$avddtmds - 700 \text{ mV}$	-	$avddtmds - 400 \text{ mV}$	mV
R_{TERM}	Differential source termination load (inside HDMI 3D Tx PHY) Although the HDMI 3D Tx PHY includes differential source termination, the user-defined value is set for each single line (for illustration, see the third figure above). Note: R_{TERM} can also be configured to be open and not present on TMDS channels.	-	50	-	200	Ω
Hot plug detect specifications						
HPD^{VH}	Hot plug detect high range	-	2.0	-	5.3	V
$V_{HPD_{VL}}$	Hot plug detect low range	-	0	-	0.8	V
HPD_Z	Hot plug detect input impedance	-	10	-	-	k Ω
HPD_t	Hot plug detect time delay	-	-	-	100	μs

4.11.8 Switching Characteristics

Table 67 describes switching characteristics for the HDMI 3D Tx PHY. Figure 58 to Figure 62 illustrate various parameters specified in table.

Table 71. Video Signal Cross-Reference

i.MX 6Solo/6DualLite	LCD							Comment ¹
Port Name (x=0, 1)	RGB, Signal Name (General)	RGB/TV Signal Allocation (Example)						
		16-bit RGB	18-bit RGB	24 Bit RGB	8-bit Y/Cb ²	16-bit Y/Cb	20-bit Y/Cb	
DISPx_DAT0	DAT[0]	B[0]	B[0]	B[0]	Y/C[0]	C[0]	C[0]	The restrictions are as follows: <ul style="list-style-type: none"> • There are maximal three continuous groups of bits that could be independently mapped to the external bus. Groups should not be overlapped. • The bit order is expressed in each of the bit groups, for example, B[0] = least significant blue pixel bit
DISPx_DAT1	DAT[1]	B[1]	B[1]	B[1]	Y/C[1]	C[1]	C[1]	
DISPx_DAT2	DAT[2]	B[2]	B[2]	B[2]	Y/C[2]	C[2]	C[2]	
DISPx_DAT3	DAT[3]	B[3]	B[3]	B[3]	Y/C[3]	C[3]	C[3]	
DISPx_DAT4	DAT[4]	B[4]	B[4]	B[4]	Y/C[4]	C[4]	C[4]	
DISPx_DAT5	DAT[5]	G[0]	B[5]	B[5]	Y/C[5]	C[5]	C[5]	
DISPx_DAT6	DAT[6]	G[1]	G[0]	B[6]	Y/C[6]	C[6]	C[6]	
DISPx_DAT7	DAT[7]	G[2]	G[1]	B[7]	Y/C[7]	C[7]	C[7]	
DISPx_DAT8	DAT[8]	G[3]	G[2]	G[0]	—	Y[0]	C[8]	
DISPx_DAT9	DAT[9]	G[4]	G[3]	G[1]	—	Y[1]	C[9]	
DISPx_DAT10	DAT[10]	G[5]	G[4]	G[2]	—	Y[2]	Y[0]	
DISPx_DAT11	DAT[11]	R[0]	G[5]	G[3]	—	Y[3]	Y[1]	
DISPx_DAT12	DAT[12]	R[1]	R[0]	G[4]	—	Y[4]	Y[2]	
DISPx_DAT13	DAT[13]	R[2]	R[1]	G[5]	—	Y[5]	Y[3]	
DISPx_DAT14	DAT[14]	R[3]	R[2]	G[6]	—	Y[6]	Y[4]	
DISPx_DAT15	DAT[15]	R[4]	R[3]	G[7]	—	Y[7]	Y[5]	
DISPx_DAT16	DAT[16]	—	R[4]	R[0]	—	—	Y[6]	
DISPx_DAT17	DAT[17]	—	R[5]	R[1]	—	—	Y[7]	
DISPx_DAT18	DAT[18]	—	—	R[2]	—	—	Y[8]	
DISPx_DAT19	DAT[19]	—	—	R[3]	—	—	Y[9]	
DISPx_DAT20	DAT[20]	—	—	R[4]	—	—	—	
DISPx_DAT21	DAT[21]	—	—	R[5]	—	—	—	

4.11.10.6 Synchronous Interfaces to Standard Active Matrix TFT LCD Panels

4.11.10.6.1 IPU Display Operating Signals

The IPU uses four control signals and data to operate a standard synchronous interface:

- IPP_DISP_CLK—Clock to display
- HSYNC—Horizontal synchronization
- VSYNC—Vertical synchronization
- DRDY—Active data

All synchronous display controls are generated on the base of an internally generated “local start point”. The synchronous display controls can be placed on time axis with DI’s offset, up and down parameters. The display access can be whole number of DI clock (Tdiclk) only. The IPP_DATA can not be moved relative to the local start point. The data bus of the synchronous interface is output direction only.

4.11.10.6.2 LCD Interface Functional Description

Figure 67 depicts the LCD interface timing for a generic active matrix color TFT panel. In this figure, signals are shown with negative polarity. The sequence of events for active matrix interface timing is:

- DI_CLK internal DI clock is used for calculation of other controls.
- IPP_DISP_CLK latches data into the panel on its negative edge (when positive polarity is selected). In active mode, IPP_DISP_CLK runs continuously.
- HSYNC causes the panel to start a new line. (Usually IPP_PIN_2 is used as HSYNC.)
- VSYNC causes the panel to start a new frame. It always encompasses at least one HSYNC pulse. (Usually IPP_PIN_3 is used as VSYNC.)
- DRDY acts like an output enable signal to the CRT display. This output enables the data to be shifted onto the display. When disabled, the data is invalid and the trace is off. (DRDY can be used either synchronous or asynchronous generic purpose pin as well.)

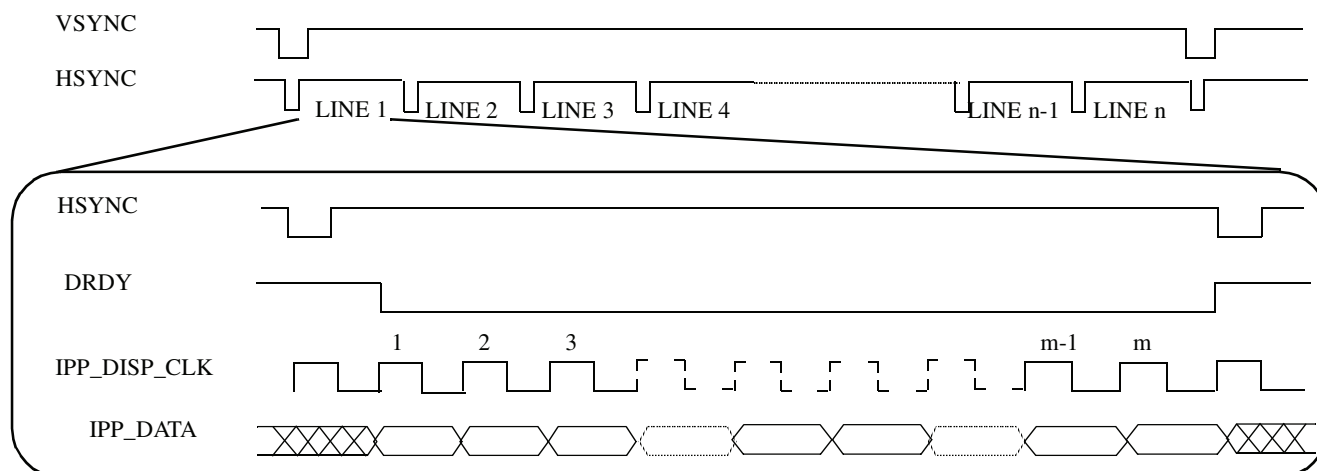


Figure 67. Interface Timing Diagram for TFT (Active Matrix) Panels

Table 81. MLB 1024 Fs Timing Parameters (continued)

Parameter	Symbol	Min	Max	Unit	Comment
MLBSIG/MLBDAT output high impedance from MLBCLK low	t_{mcfdz}	0	t_{mckl}	ns	3
Bus Hold from MLBCLK low	t_{mdzh}	2	—	ns	—
MLBSIG/MLBDAT output valid from transition of MLBCLK (low to high)	t_{delay}	—	7	ns	—

¹ The controller can shut off MLBCLK to place MediaLB in a low-power state. Depending on the time the clock is shut off, a runt pulse can occur on MLBCLK.

² MLBCLK low/high time includes the pulse width variation.

³ The MediaLB driver can release the MLBDAT/MLBSIG line as soon as MLBCLK is low; however, the logic state of the final driven bit on the line must remain on the bus for t_{mdzh} . Therefore, coupling must be minimized while meeting the maximum load capacitance listed.

Table 82 lists the MediaLB 6-pin interface timing characteristics, and Figure 87 shows the MLB 6-pin delay, setup, and hold times.

Table 82. MLB 6-Pin Interface Timing Parameters

Parameter	Symbol	Min	Max	Unit	Comment
Cycle-to-cycle system jitter	t_{jitter}	—	600	ps	—
Transmitter MLBSP/N (MLBDP/N) output valid from transition of MLBCP/N (low-to-high) ¹	t_{delay}	0.6	1.3	ns	
Disable turnaround time from transition of MLBCP/N (low-to-high)	t_{phz}	0.6	3.5	ns	
Enable turnaround time from transition of MLBCP/N (low-to-high)	t_{plz}	0.6	5.6	ns	
MLBSP/N (MLBDP/N) valid to transition of MLBCP/N (low-to-high)	t_{su}	0.05	—	ns	
MLBSP/N (MLBDP/N) hold from transition of MLBCP/N (low-to-high) ²	t_{hd}	0.6			

¹ t_{delay} , t_{phz} , t_{plz} , t_{su} , and t_{hd} may also be referenced from a low-to-high transition of the recovered clock for 2:1 and 4:1 recovered-to-external clock ratios.

² The transmitting device must ensure valid data on MLBSP/N (MLBDP/N) for at least $t_{hd(min)}$ following the rising edge of MLBCP/N; receivers must latch MLBSP/N (MLBDP/N) data within $t_{hd(min)}$ of the rising edge of MLBCP/N.

4.11.19 SSI Timing Parameters

This section describes the timing parameters of the SSI module. The connectivity of the serial synchronous interfaces are summarized in [Table 86](#).

Table 86. AUDMUX Port Allocation

Port	Signal Nomenclature	Type and Access
AUDMUX port 1	SSI 1	Internal
AUDMUX port 2	SSI 2	Internal
AUDMUX port 3	AUD3	External— AUD3 I/O
AUDMUX port 4	AUD4	External— EIM or CSPI1 I/O through IOMUXC
AUDMUX port 5	AUD5	External— EIM or SD1 I/O through IOMUXC
AUDMUX port 6	AUD6	External— EIM or DISP2 through IOMUXC
AUDMUX port 7	SSI 3	Internal

NOTE

- The terms WL and BL used in the timing diagrams and tables see Word Length (WL) and Bit Length (BL).
- The SSI timing diagrams use generic signal names wherein the names used in the *i.MX 6Solo/6DualLite Reference Manual (IMX6SDLRM)* are channel specific signal names. For example, a channel clock referenced in the IOMUXC chapter as AUD3_TXC appears in the timing diagram as RGMII_TXC.

4.11.19.2 SSI Receiver Timing with Internal Clock

Figure 96 depicts the SSI receiver internal clock timing and Table 88 lists the timing parameters for the receiver timing with the internal clock.

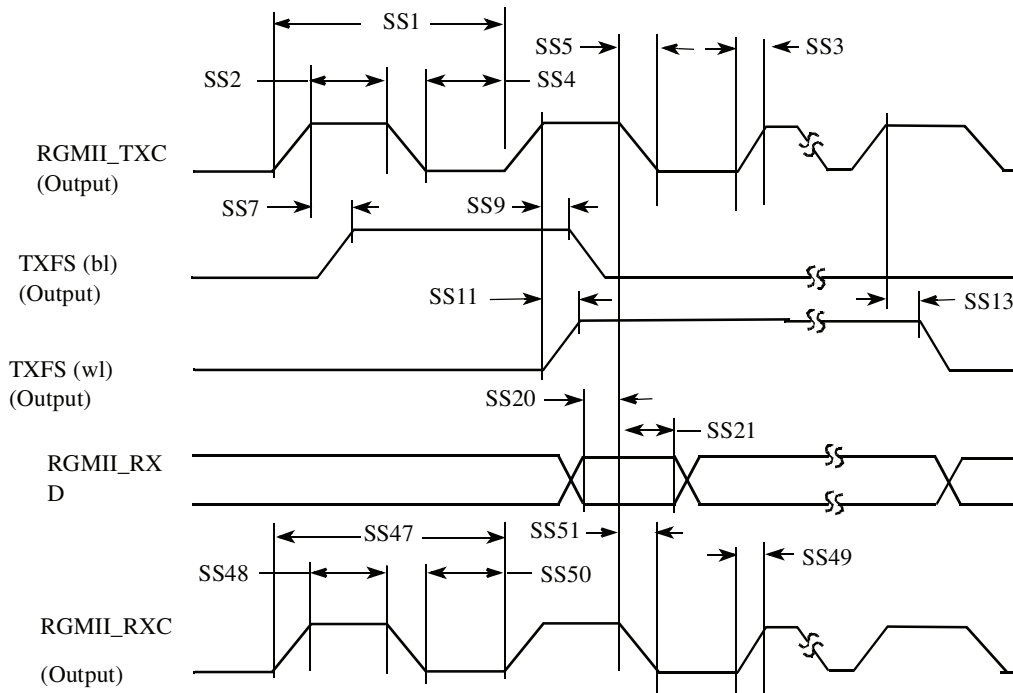


Figure 96. SSI Receiver Internal Clock Timing Diagram

Table 88. SSI Receiver Timing with Internal Clock

ID	Parameter	Min	Max	Unit
Internal Clock Operation				
SS1	(Tx/Rx) CK clock period	81.4	—	ns
SS2	(Tx/Rx) CK clock high period	36.0	—	ns
SS3	(Tx/Rx) CK clock rise time	—	6.0	ns
SS4	(Tx/Rx) CK clock low period	36.0	—	ns
SS5	(Tx/Rx) CK clock fall time	—	6.0	ns
SS7	(Rx) CK high to FS (bl) high	—	15.0	ns
SS9	(Rx) CK high to FS (bl) low	—	15.0	ns
SS11	(Rx) CK high to FS (wl) high	—	15.0	ns
SS13	(Rx) CK high to FS (wl) low	—	15.0	ns
SS20	SRXD setup time before (Rx) CK low	10.0	—	ns
SS21	SRXD hold time after (Rx) CK low	0.0	—	ns
Oversampling Clock Operation				

4.11.20.2.1 UART Transmitter

Figure 99 depicts the transmit timing of UART in the RS-232 serial mode, with 8 data bit/1 stop bit format. Table 92 lists the UART RS-232 serial mode transmit timing characteristics.

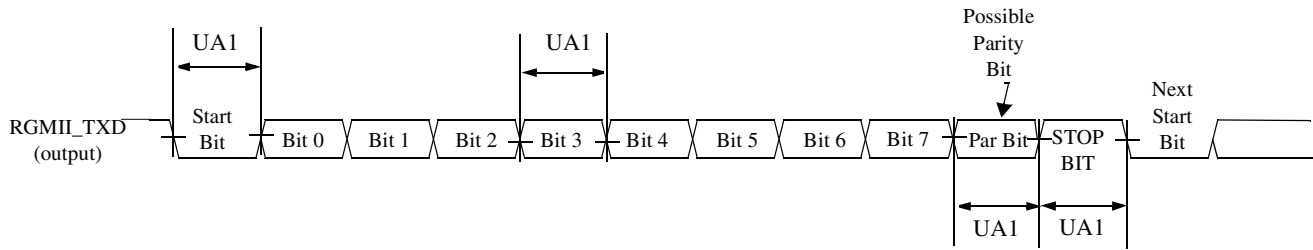


Figure 99. UART RS-232 Serial Mode Transmit Timing Diagram

Table 92. RS-232 Serial Mode Transmit Timing Parameters

ID	Parameter	Symbol	Min	Max	Unit
UA1	Transmit Bit Time	t_{Tbit}	$1/F_{baud_rate}^1 - T_{ref_clk}^2$	$1/F_{baud_rate} + T_{ref_clk}$	—

¹ F_{baud_rate} : Baud rate frequency. The maximum baud rate the UART can support is $(ipg_perclk \text{ frequency})/16$.

² T_{ref_clk} : The period of UART reference clock ref_clk (ipg_perclk after RFDIV divider).

4.11.20.2.2 UART Receiver

Figure 100 depicts the RS-232 serial mode receive timing with 8 data bit/1 stop bit format. Table 93 lists serial mode receive timing characteristics.

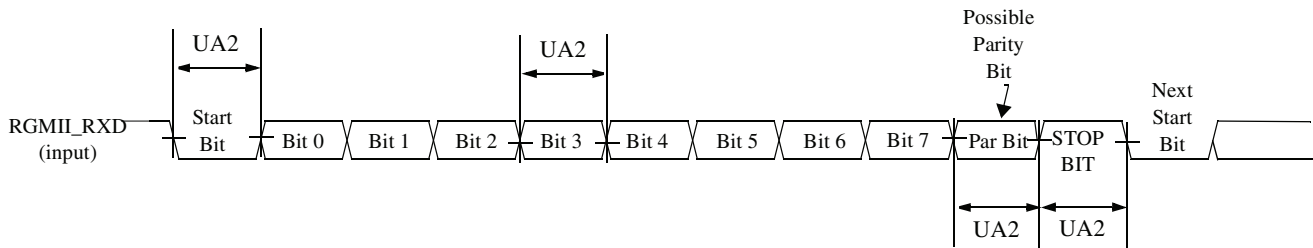


Figure 100. UART RS-232 Serial Mode Receive Timing Diagram

Table 93. RS-232 Serial Mode Receive Timing Parameters

ID	Parameter	Symbol	Min	Max	Unit
UA2	Receive Bit Time ¹	t_{Rbit}	$1/F_{baud_rate}^2 - 1/(16 \times F_{baud_rate})$	$1/F_{baud_rate} + 1/(16 \times F_{baud_rate})$	—

¹ The UART receiver can tolerate $1/(16 \times F_{baud_rate})$ tolerance in each bit. But accumulation tolerance in one frame must not exceed $3/(16 \times F_{baud_rate})$.

² F_{baud_rate} : Baud rate frequency. The maximum baud rate the UART can support is $(ipg_perclk \text{ frequency})/16$.

5.2 Boot Device Interface Allocation

Table 99 lists the interfaces that can be used by the boot process in accordance with the specific boot mode configuration. The table also describes the interface's specific modes and IOMUXC allocation, which are configured during boot when appropriate.

Table 99. Interface Allocation During Boot

Interface	IP Instance	Allocated Pads During Boot	Comment
SPI	ECSPI-1	EIM_D17, EIM_D18, EIM_D16, EIM_EB2, EIM_D19, EIM_D24, EIM_D25	
SPI	ECSPI-2	CSI0_DAT10, CSI0_DAT9, CSI0_DAT8, CSI0_DAT11, EIM_LBA, EIM_D24, EIM_D25	
SPI	ECSPI-3	DISP0_DAT2, DISP0_DAT1, DISP0_DAT0, DISP0_DAT3, DISP0_DAT4, DISP0_DAT5, DISP0_DAT6	
SPI	ECSPI-4	EIM_D22, EIM_D28, EIM_D21, EIM_D20, EIM_A25, EIM_D24, EIM_D25	
EIM	EIM	EIM_DA[15:0], EIM_D[31:16], CSI0_DAT[19:4], CSI0_DATA_EN, CSI0_VSYNC	Used for NOR, OneNAND boot Only CS0 is supported
NAND Flash	GPMI	NANDF_CLE, NANDF_ALE, NANDF_WP_B, SD4_CMD, SD4_CLK, NANDF_RB0, SD4_DAT0, NANDF_CS0, NANDF_CS1, NANDF_CS2, NANDF_CS3, NANDF_D[7:0]	8 bit Only CS0 is supported
SD/MMC	USDHC-1	SD1_CLK, SD1_CMD, SD1_DAT0, SD1_DAT1, SD1_DAT2, SD1_DAT3, GPIO_1, NANDF_D0, NANDF_D1, NANDF_D2, NANDF_D3, KEY_COL1	1, 4, or 8 bit
SD/MMC	USDHC-2	SD2_CLK, SD2_CMD, SD2_DAT0, SD2_DAT1, SD2_DAT2, SD2_DAT3, GPIO_4, NANDF_D5, NANDF_D6, NANDF_D7, NANDF_D8, KEY_ROW1	1, 4, or 8 bit
SD/MMC	USDHC-3	SD3_CLK, SD3_CMD, SD3_DAT0, SD3_DAT1, SD3_DAT2, SD3_DAT3, SD3_DAT4, SD3_DAT5, SD3_DAT6, SD3_DAT7, SD3_RST, GPIO_18	1, 4, or 8 bit
SD/MMC	USDHC-4	SD4_CLK, SD4_CMD, SD4_DAT0, SD4_DAT1, SD4_DAT2, SD4_DAT3, SD4_DAT4, SD4_DAT5, SD4_DAT6, SD4_DAT7, NANDF_ALE, NANDF_CS1	1, 4, or 8 bit
I ² C	I ² C-1	EIM_D28, EIM_D21	—
I ² C	I ² C-2	EIM_D16, EIM_EB2	—
I ² C	I ² C-3	EIM_D18, EIM_D17	—
USB	USB-OTG PHY	USB_OTG_DP USB_OTG_DN USB_OTG_VBUS	—

Table 101. 21 x 21 mm Functional Contact Assignments¹ (continued)

Ball Name	Ball	Power Group	Ball Type	Out of Reset Condition ²			
				Default Mode (Reset Mode)	Default Function	Input/Output	Value
DRAM_D4	AC1	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[4]	Input	100 kΩ pull-up
DRAM_D40	Y19	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[40]	Input	100 kΩ pull-up
DRAM_D41	AB20	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[41]	Input	100 kΩ pull-up
DRAM_D42	AB21	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[42]	Input	100 kΩ pull-up
DRAM_D43	AD21	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[43]	Input	100 kΩ pull-up
DRAM_D44	Y20	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[44]	Input	100 kΩ pull-up
DRAM_D45	AA20	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[45]	Input	100 kΩ pull-up
DRAM_D46	AE21	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[46]	Input	100 kΩ pull-up
DRAM_D47	AC21	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[47]	Input	100 kΩ pull-up
DRAM_D48	AC22	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[48]	Input	100 kΩ pull-up
DRAM_D49	AE22	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[49]	Input	100 kΩ pull-up
DRAM_D5	AD1	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[5]	Input	100 kΩ pull-up
DRAM_D50	AE24	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[50]	Input	100 kΩ pull-up
DRAM_D51	AC24	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[51]	Input	100 kΩ pull-up
DRAM_D52	AB22	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[52]	Input	100 kΩ pull-up
DRAM_D53	AC23	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[53]	Input	100 kΩ pull-up
DRAM_D54	AD25	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[54]	Input	100 kΩ pull-up
DRAM_D55	AC25	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[55]	Input	100 kΩ pull-up
DRAM_D56	AB25	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[56]	Input	100 kΩ pull-up
DRAM_D57	AA21	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[57]	Input	100 kΩ pull-up
DRAM_D58	Y25	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[58]	Input	100 kΩ pull-up
DRAM_D59	Y22	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[59]	Input	100 kΩ pull-up
DRAM_D6	AB4	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[6]	Input	100 kΩ pull-up
DRAM_D60	AB23	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[60]	Input	100 kΩ pull-up
DRAM_D61	AA23	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[61]	Input	100 kΩ pull-up
DRAM_D62	Y23	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[62]	Input	100 kΩ pull-up
DRAM_D63	W25	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[63]	Input	100 kΩ pull-up
DRAM_D7	AE4	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[7]	Input	100 kΩ pull-up
DRAM_D8	AD5	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[8]	Input	100 kΩ pull-up
DRAM_D9	AE5	NVCC_DRAM	DDR	ALT0	mmdc.DRAM_D[9]	Input	100 kΩ pull-up