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Details

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Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	4MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	33
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	128 x 8
RAM Size	192 x 8
Voltage - Supply (Vcc/Vdd)	4V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16f874-04i-pt

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2.2 Data Memory Organization

The data memory is partitioned into multiple banks which contain the General Purpose Registers and the Special Function Registers. Bits RP1 (STATUS<6>) and RP0 (STATUS<5>) are the bank select bits.

RP1:RP0	Bank
00	0
01	1
10	2
11	3

Each bank extends up to 7Fh (128 bytes). The lower locations of each bank are reserved for the Special Function Registers. Above the Special Function Registers are General Purpose Registers, implemented as static RAM. All implemented banks contain Special Function Registers. Some frequently used Special Function Registers from one bank may be mirrored in another bank for code reduction and quicker access.

Note:	EEPROM Data Memory description can be found in Section 4.0 of this data sheet.	
0.0.4		

2.2.1 GENERAL PURPOSE REGISTER FILE

The register file can be accessed either directly, or indirectly through the File Select Register (FSR).

2.5 Indirect Addressing, INDF and FSR Registers

The INDF register is not a physical register. Addressing the INDF register will cause indirect addressing.

Indirect addressing is possible by using the INDF register. Any instruction using the INDF register actually accesses the register pointed to by the File Select Register, FSR. Reading the INDF register itself, indirectly (FSR = '0') will read 00h. Writing to the INDF register indirectly results in a no operation (although status bits may be affected). An effective 9-bit address is obtained by concatenating the 8-bit FSR register and the IRP bit (STATUS<7>), as shown in Figure 2-6. A simple program to clear RAM locations 20h-2Fh using indirect addressing is shown in Example 2-2.

EXAMPLE 2-2: INDIRECT ADDRESSING

	MOVLW	0x20	;initialize pointer
	MOVWF	FSR	;to RAM
NEXT	CLRF	INDF	clear INDF register;
	INCF	FSR,F	;inc pointer
	BTFSS	FSR,4	;all done?
	GOTO	NEXT	;no clear next
CONTINUE			
	:		;yes continue





3.6 Parallel Slave Port

The Parallel Slave Port (PSP) is not implemented on the PIC16F873 or PIC16F876.

PORTD operates as an 8-bit wide Parallel Slave Port or microprocessor port, when control bit PSPMODE (TRISE<4>) is set. In Slave mode, it is asynchronously readable and writable by the external world through RD control input pin RE0/RD and WR control input pin RE1/WR.

The PSP can directly interface to an 8-bit microprocessor data bus. The external microprocessor can read or write the PORTD latch as an 8-bit latch. Setting bit PSPMODE enables port pin RE0/RD to be the RD input, RE1/WR to be the WR input and RE2/CS to be the CS (chip select) input. For this functionality, the corresponding data direction bits of the TRISE register (TRISE<2:0>) must be configured as inputs (set). The A/D port configuration bits PCFG3:PCFG0 (ADCON1<3:0>) must be set to configure pins RE2:RE0 as digital I/O.

There are actually two 8-bit latches: one for data output, and one for data input. The user writes 8-bit data to the PORTD data latch and reads data from the port pin latch (note that they have the same address). In this mode, the TRISD register is ignored, since the external device is controlling the direction of data flow.

A write to the PSP occurs when both the \overline{CS} and \overline{WR} lines are first detected low. When either the \overline{CS} or \overline{WR} lines become high (level triggered), the Input Buffer Full (IBF) status flag bit (TRISE<7>) is set on the Q4 clock cycle, following the next Q2 cycle, to signal the write is complete (Figure 3-10). The interrupt flag bit PSPIF (PIR1<7>) is also set on the same Q4 clock cycle. IBF can only be cleared by reading the PORTD input latch. The Input Buffer Overflow (IBOV) status flag bit (TRISE<5>) is set if a second write to the PSP is attempted when the previous byte has not been read out of the buffer.

A read from the PSP occurs when both the \overline{CS} and \overline{RD} lines are first detected low. The Output Buffer Full (OBF) status flag bit (TRISE<6>) is cleared immediately (Figure 3-11), indicating that the PORTD latch is waiting to be read by the external bus. When either the \overline{CS} or \overline{RD} pin becomes high (level triggered), the interrupt flag bit PSPIF is set on the Q4 clock cycle, following the next Q2 cycle, indicating that the read is complete. OBF remains low until data is written to PORTD by the user firmware.

When not in PSP mode, the IBF and OBF bits are held clear. However, if flag bit IBOV was previously set, it must be cleared in firmware.

An interrupt is generated and latched into flag bit PSPIF when a read or write operation is completed. PSPIF must be cleared by the user in firmware and the interrupt can be disabled by clearing the interrupt enable bit PSPIE (PIE1<7>).



PORTD AND PORTE BLOCK DIAGRAM (PARALLEL SLAVE



SSPSTAT: SYNC SERIAL PORT STATUS REGISTER (ADDRESS: 94h) REGISTER 9-1: R/W-0 R/W-0 R-0 R-0 R-0 R-0 R-0 R-0 SMP D/A Р R/W BF CKE S UA bit 7 bit 0 bit 7 SMP: Sample bit SPI Master mode: 1 = Input data sampled at end of data output time 0 = Input data sampled at middle of data output time SPI Slave mode: SMP must be cleared when SPI is used in slave mode In I²C Master or Slave mode: 1 = Slew rate control disabled for standard speed mode (100 kHz and 1 MHz) 0 = Slew rate control enabled for high speed mode (400 kHz) bit 6 CKE: SPI Clock Edge Select (Figure 9-2, Figure 9-3 and Figure 9-4) SPI mode: For CKP = 0 1 = Data transmitted on rising edge of SCK 0 = Data transmitted on falling edge of SCK For CKP = 1 1 = Data transmitted on falling edge of SCK 0 = Data transmitted on rising edge of SCK In I²C Master or Slave mode: 1 = Input levels conform to SMBus spec 0 = Input levels conform to I²C specs **D/A**: Data/Address bit (I²C mode only) bit 5 1 = Indicates that the last byte received or transmitted was data 0 = Indicates that the last byte received or transmitted was address bit 4 P: STOP bit (I²C mode only. This bit is cleared when the MSSP module is disabled, SSPEN is cleared.) 1 = Indicates that a STOP bit has been detected last (this bit is '0' on RESET) 0 = STOP bit was not detected last bit 3 S: START bit (I²C mode only. This bit is cleared when the MSSP module is disabled, SSPEN is cleared.) 1 = Indicates that a START bit has been detected last (this bit is '0' on RESET) 0 = START bit was not detected last bit 2 **R/W**: Read/Write bit Information (I²C mode only) This bit holds the R/W bit information following the last address match. This bit is only valid from the address match to the next START bit, STOP bit or not ACK bit. In I²C Slave mode: 1 = Read0 = WriteIn I²C Master mode: 1 = Transmit is in progress 0 = Transmit is not in progress Logical OR of this bit with SEN, RSEN, PEN, RCEN, or ACKEN will indicate if the MSSP is in IDLE mode. bit 1 **UA**: Update Address (10-bit I²C mode only) 1 = Indicates that the user needs to update the address in the SSPADD register 0 = Address does not need to be updated bit BF: Buffer Full Status bit Receive (SPI and I²C modes): 1 = Receive complete, SSPBUF is full 0 = Receive not complete, SSPBUF is empty Transmit (I²C mode only): 1 = Data transmit in progress (does not include the ACK and STOP bits), SSPBUF is full 0 = Data transmit complete (does not include the ACK and STOP bits), SSPBUF is empty Legend: R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'

- n = Value at POR

'1' = Bit is set

x = Bit is unknown

'0' = Bit is cleared



FIGURE 9-7: I²C WAVEFORMS FOR TRANSMISSION (7-BIT ADDRESS)

9.2.2 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the I^2C bus is such that the first byte after the START condition usually determines which device will be the slave addressed by the master. The exception is the general call address, which can address all devices. When this address is used, all devices should, in theory, respond with an acknowledge.

The general call address is one of eight addresses reserved for specific purposes by the I²C protocol. It consists of all 0's with R/W = 0.

The general call address is recognized when the General Call Enable bit (GCEN) is enabled (SSPCON2<7> is set). Following a START bit detect, 8 bits are shifted into SSPSR and the address is compared against SSPADD. It is also compared to the general call address and fixed in hardware.

If the general call address matches, the SSPSR is transferred to the SSPBUF, the BF flag is set (eighth bit), and on the falling edge of the ninth bit (ACK bit), the SSPIF flag is set.

When the interrupt is serviced, the source for the interrupt can be checked by reading the contents of the SSPBUF to determine if the address was device specific, or a general call address.

In 10-bit mode, the SSPADD is required to be updated for the second half of the address to match, and the UA bit is set (SSPSTAT<1>). If the general call address is sampled when GCEN is set, while the slave is configured in 10-bit address mode, then the second half of the address is not necessary, the UA bit will not be set, and the slave will begin receiving data after the Acknowledge (Figure 9-8).



FIGURE 9-8: SLAVE MODE GENERAL CALL ADDRESS SEQUENCE (7 OR 10-BIT MODE)

9.2.7 I²C MASTER MODE SUPPORT

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPCON and by setting the SSPEN bit. Once Master mode is enabled, the user has six options:

- Assert a START condition on SDA and SCL.
- Assert a Repeated START condition on SDA and SCL.
- Write to the SSPBUF register initiating transmission of data/address.
- Generate a STOP condition on SDA and SCL.
- Configure the I²C port to receive data.
- Generate an Acknowledge condition at the end of a received byte of data.
- Note: The MSSP Module, when configured in I²C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a START condition and immediately write the SSPBUF register to initiate transmission before the START condition is complete. In this case, the SSPBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPBUF did not occur.

9.2.7.1 I²C Master Mode Operation

The master device generates all of the serial clock pulses and the START and STOP conditions. A transfer is ended with a STOP condition or with a Repeated START condition. Since the Repeated START condition is also the beginning of the next serial transfer, the l^2C bus will not be released.

In Master Transmitter mode, serial data is output through SDA, while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic '0'. Serial data is transmitted 8 bits at a time. After each byte is transmitted, an Acknowledge bit is received. START and STOP conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/W bit. In this case, the R/W bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address followed by a '1' to indicate receive bit. Serial data is received via SDA, while SCL outputs the serial clock. Serial data is received 8 bits at a time. After each byte is received, an Acknowledge bit is transmitted. START and STOP conditions indicate the beginning and end of transmission.

The baud rate generator used for SPI mode operation is now used to set the SCL clock frequency for either 100 kHz, 400 kHz, or 1 MHz I²C operation. The baud rate generator reload value is contained in the lower 7 bits of the SSPADD register. The baud rate generator will automatically begin counting on a write to the SSPBUF. Once the given operation is complete (i.e., transmission of the last data bit is followed by ACK), the internal clock will automatically stop counting and the SCL pin will remain in its last state.

A typical transmit sequence would go as follows:

- a) User generates a START condition by setting the START enable bit (SEN) in SSPCON2.
- b) SSPIF is set. The module will wait the required start time before any other operation takes place.
- c) User loads SSPBUF with address to transmit.
- d) Address is shifted out the SDA pin until all 8 bits are transmitted.
- e) MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPCON2 register (SSPCON2<6>).
- f) MSSP module generates an interrupt at the end of the ninth clock cycle by setting SSPIF.
- g) User loads SSPBUF with eight bits of data.
- h) DATA is shifted out the SDA pin until all 8 bits are transmitted.
- MSSP module shifts in the ACK bit from the slave device, and writes its value into the SSPCON2 register (SSPCON2<6>).
- MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
- k) User generates a STOP condition by setting the STOP enable bit, PEN, in SSPCON2.
- I) Interrupt is generated once the STOP condition is complete.

9.2.8 BAUD RATE GENERATOR

In I^2C Master mode, the reload value for the BRG is located in the lower 7 bits of the SSPADD register (Figure 9-10). When the BRG is loaded with this value, the BRG counts down to 0 and stops until another reload has taken place. The BRG count is decremented twice per instruction cycle (Tcr), on the Q2 and Q4 clock.

In I²C Master mode, the BRG is reloaded automatically. If clock arbitration is taking place, the BRG will be reloaded when the SCL pin is sampled high (Figure 9-11).

Note: Baud Rate = Fosc / (4 * (SSPADD + 1))

FIGURE 9-10:

BAUD RATE GENERATOR BLOCK DIAGRAM



PIC16F87X

9.2.18.1 Bus Collision During a START Condition

During a START condition, a bus collision occurs if:

- a) SDA or SCL are sampled low at the beginning of the START condition (Figure 9-20).
- b) SCL is sampled low before SDA is asserted low (Figure 9-21).

During a START condition, both the SDA and the SCL pins are monitored. If either the SDA pin <u>or</u> the SCL pin is already low, then these events all occur:

- the START condition is aborted,
- and the BCLIF flag is set,
- <u>and</u> the SSP module is reset to its IDLE state (Figure 9-20).

The START condition begins with the SDA and SCL pins de-asserted. When the SDA pin is sampled high, the baud rate generator is loaded from SSPADD<6:0> and counts down to 0. If the SCL pin is sampled low while SDA is high, a bus collision occurs, because it is assumed that another master is attempting to drive a data '1' during the START condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 9-22). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The baud rate generator is then reloaded and counts down to 0. During this time, if the SCL pins are sampled as '0', a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a START condition is that no two bus masters can assert a START condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision, because the two masters must be allowed to arbitrate the first address following the START condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated START, or STOP conditions.





PIC16F87X

	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x	
	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	
	bit 7							bit 0	
bit 7	SPEN: Serial Port Enable bit								
	1 = Serial p	oort enabled	(configures	RC7/RX/D	I and RC6/I	X/CK pins a	is serial port	t pins)	
hit 6		Pacaiva Enc	u ahla hit						
DIL U	1 = Selects	9-bit recept	tion						
	0 = Selects	s 8-bit recept	tion						
bit 5	SREN: Sin	gle Receive	Enable bit						
	Asynchron	<u>ous mode:</u>							
	Don't care								
	Synchrono	<u>us mode - m</u> a aingla raa	<u>naster:</u>						
	1 = Enable 0 = Disable	s single rece	eive						
	This bit is c	cleared after	reception is	s complete.					
	<u>Synchrono</u>	us mode - sl	lave:						
	Don't care								
bit 4	CREN: Co	ntinuous Re	ceive Enabl	e bit					
	Asynchron	<u>ous mode:</u>							
	1 = Enable 0 = Disable	s continuou: es continuou	s receive						
	Synchronous mode:								
	1 = Enables continuous receive until enable bit CREN is cleared (CREN overrides SREN)								
	0 = Disables continuous receive								
bit 3	ADDEN: A	ddress Dete	ect Enable b	it					
	Asynchron	<u>ous mode 9-</u>	<u>-bit (RX9 = ´</u>	<u>1):</u> ablasinterru	امحما اممر	of the reaction	a huffaruuh		
	$\perp = \text{Enable}$ RSR<8	s address de l> is set	etection, ena	ables Interru	ipt and load	of the receiv	e butter wh	en	
	0 = Disable	es address d	letection, all	bytes are re	eceived, and	ninth bit ca	n be used a	s parity bit	
bit 2	FERR: Fra	ming Error b	oit						
	1 = Framin	g error (can	be updated	by reading	RCREG reg	ister and rec	ceive next va	alid byte)	
	0 = No frar	ning error							
bit 1		errun Error	bit	hu olooring l					
	1 = Overru	rrun error	be cleared	by cleaning i					
bit 0	RX9D: 9th bit of Received Data (can be parity bit, but must be calculated by user firmware)								
20						. se calculu			
	Legend:								
	R = Reada	ble bit	W = W	/ritable bit	U = Unim	plemented	bit. read as	'0'	

'1' = Bit is set

'0' = Bit is cleared

REGISTER 10-2: RCSTA: RECEIVE STATUS AND CONTROL REGISTER (ADDRESS 18h)

- n = Value at POR

x = Bit is unknown

TABLE 10-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other RESETS
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	R0IF	0000 000x	0000 000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
18h	RCSTA	SPEN	RX9	SREN	CREN	_	FERR	OERR	RX9D	0000 -00x	0000 -00x
19h	TXREG	USART Tr	USART Transmit Register							0000 0000	0000 0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
98h	TXSTA	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	Baud Rate	Baud Rate Generator Register								0000 0000

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission. **Note 1:** Bits PSPIE and PSPIF are reserved on PIC16F873/876 devices; always maintain these bits clear.

FIGURE 10-9: SYNCHRONOUS TRANSMISSION



FIGURE 10-10: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)



11.2 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires a minimum 12TAD per 10-bit conversion. The source of the A/D conversion clock is software selected. The four possible options for TAD are:

- 2Tosc
- 8Tosc
- 32Tosc
- Internal A/D module RC oscillator (2-6 μs)

For correct A/D conversions, the A/D conversion clock (TAD) must be selected to ensure a minimum TAD time of 1.6 $\mu s.$

Table 11-1 shows the resultant TAD times derived from the device operating frequencies and the A/D clock source selected.

TABLE 11-1: TAD VS. MAXIMUM DEVICE OPERATING FREQUENCIES (STANDARD DEVICES (C))

AD Clock	Maximum Device Frequency	
Operation	ADCS1:ADCS0	Max.
2Tosc	0.0	1.25 MHz
8Tosc	01	5 MHz
32Tosc	10	20 MHz
RC ^(1, 2, 3)	11	(Note 1)

Note 1: The RC source has a typical TAD time of 4 μ s, but can vary between 2-6 μ s.

2: When the device frequencies are greater than 1 MHz, the RC A/D conversion clock source is only recommended for SLEEP operation.

3: For extended voltage devices (LC), please refer to the Electrical Characteristics (Sections 15.1 and 15.2).

11.3 Configuring Analog Port Pins

The ADCON1 and TRIS registers control the operation of the A/D port pins. The port pins that are desired as analog inputs must have their corresponding TRIS bits set (input). If the TRIS bit is cleared (output), the digital output level (VOH or VOL) will be converted.

The A/D operation is independent of the state of the CHS2:CHS0 bits and the TRIS bits.

Note	1:	When reading the port register, any pin
		configured as an analog input channel will
		read as cleared (a low level). Pins config-
		ured as digital inputs will convert an ana-
		log input. Analog levels on a digitally
		configured input will not affect the conver-
		sion accuracy.

2: Analog levels on any pin that is defined as a digital input (including the AN7:AN0 pins), may cause the input buffer to consume current that is out of the device specifications.

11.4 A/D Conversions

Clearing the GO/DONE bit during a conversion will abort the current conversion. The A/D result register pair will NOT be updated with the partially completed A/D conversion sample. That is, the ADRESH:ADRESL registers will continue to contain the value of the last completed conversion (or the last value written to the ADRESH:ADRESL registers). After the A/D conversion is aborted, a 2TAD wait is required before the next

FIGURE 11-3: A/D CONVERSION TAD CYCLES

acquisition is started. After this 2TAD wait, acquisition on the selected channel is automatically started. The GO/DONE bit can then be set to start the conversion.

In Figure 11-3, after the GO bit is set, the first time segment has a minimum of TCY and a maximum of TAD.

Note: The GO/DONE bit should NOT be set in the same instruction that turns on the A/D.



11.4.1 A/D RESULT REGISTERS

The ADRESH:ADRESL register pair is the location where the 10-bit A/D result is loaded at the completion of the A/D conversion. This register pair is 16-bits wide. The A/D module gives the flexibility to left or right justify the 10-bit result in the 16-bit result register. The A/D Format Select bit (ADFM) controls this justification. Figure 11-4 shows the operation of the A/D result justification. The extra bits are loaded with '0's'. When an A/D result will not overwrite these locations (A/D disable), these registers may be used as two general purpose 8-bit registers.

FIGURE 11-4: A/D RESULT JUSTIFICATION



12.10.1 INT INTERRUPT

External interrupt on the RB0/INT pin is edge triggered, either rising, if bit INTEDG (OPTION_REG<6>) is set, or falling, if the INTEDG bit is clear. When a valid edge appears on the RB0/INT pin, flag bit INTF (INTCON<1>) is set. This interrupt can be disabled by clearing enable bit INTE (INTCON<4>). Flag bit INTF must be cleared in software in the Interrupt Service Routine before re-enabling this interrupt. The INT interrupt can wake-up the processor from SLEEP, if bit INTE was set prior to going into SLEEP. The status of global interrupt enable bit, GIE, decides whether or not the processor branches to the interrupt vector following wake-up. See Section 12.13 for details on SLEEP mode.

12.10.2 TMR0 INTERRUPT

An overflow (FFh \rightarrow 00h) in the TMR0 register will set flag bit T0IF (INTCON<2>). The interrupt can be enabled/disabled by setting/clearing enable bit T0IE (INTCON<5>) (Section 5.0).

12.10.3 PORTB INTCON CHANGE

An input change on PORTB<7:4> sets flag bit RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit RBIE (INTCON<4>) (Section 3.2).

12.11 Context Saving During Interrupts

During an interrupt, only the return PC value is saved on the stack. Typically, users may wish to save key registers during an interrupt, (i.e., W register and STATUS register). This will have to be implemented in software.

For the PIC16F873/874 devices, the register W_TEMP must be defined in both banks 0 and 1 and must be defined at the same offset from the bank base address (i.e., If W_TEMP is defined at 0x20 in bank 0, it must also be defined at 0xA0 in bank 1). The registers, PCLATH_TEMP and STATUS_TEMP, are only defined in bank 0.

Since the upper 16 bytes of each bank are common in the PIC16F876/877 devices, temporary holding registers W_TEMP, STATUS_TEMP, and PCLATH_TEMP should be placed in here. These 16 locations don't require banking and therefore, make it easier for context save and restore. The same code shown in Example 12-1 can be used.

MOVWF	W_TEMP	;Copy W to TEMP register
SWAPF	STATUS,W	;Swap status to be saved into W
CLRF	STATUS	;bank 0, regardless of current bank, Clears IRP,RP1,RP0
MOVWF	STATUS_TEMP	;Save status to bank zero STATUS_TEMP register
MOVF	PCLATH, W	;Only required if using pages 1, 2 and/or 3
MOVWF	PCLATH_TEMP	;Save PCLATH into W
CLRF	PCLATH	;Page zero, regardless of current page
:		
:(ISR)		;(Insert user code here)
:		
MOVF	PCLATH_TEMP, W	;Restore PCLATH
MOVWF	PCLATH	;Move W into PCLATH
SWAPF	STATUS_TEMP,W	;Swap STATUS_TEMP register into W
		;(sets bank to original state)
MOVWF	STATUS	;Move W into STATUS register
SWAPF	W_TEMP,F	;Swap W_TEMP
SWAPF	W_TEMP,W	;Swap W_TEMP into W

EXAMPLE 12-1: SAVING STATUS, W, AND PCLATH REGISTERS IN RAM

	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1		Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4
OSC1	·/~~~~/			- ANNON		<u>;</u> ~~~~~~		
CLKOUT ⁽⁴⁾	\/\		<u> </u>	Tost(2)	/	\	¦/ \	
INT pin	· · ·	1		· ·		1 1	1 I 1 I	
INTF Flag (INTCON<1>)		י י י י	4	/		Interrupt Latency	(2)	
GIE bit (INTCON<7>)	1 I 1 I 1 I 1 I 1 I	i 	Processor i	n		·	1 1 1 1 1 1	ו ו ו ו
INSTRUCTIO	N FLOW			· ·		1	1 1 1 1	1
PC	Х РС Х	PC+1	ХР	C+2	PC+2	X PC + 2	<u>X 0004h X</u>	0005h
Instruction { Fetched {	Inst(PC) = SLEEP	Inst(PC + 1)		1	Inst(PC + 2)	1 1 1	Inst(0004h)	Inst(0005h)
Instruction	Inst(PC - 1)	SLEEP		, , ,	Inst(PC + 1)	Dummy cycle	Dummy cycle	Inst(0004h)
Note 1: XT, 1 2: Tos	HS or LP oscillator n r = 1024Tosc (drawi	node assumed. ing not to scale)	This delay v	vill not be th	here for RC osc n	node.		

F

If GIE = '0', execution will continue in-line.

4: CLKOUT is not available in these osc modes, but shown here for timing reference.

12.14 In-Circuit Debugger

When the DEBUG bit in the configuration word is programmed to a '0', the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB® ICD. When the microcontroller has this feature enabled, some of the resources are not available for general use. Table 12-8 shows which features are consumed by the background debugger.

I/O pins	RB6, RB7
Stack	1 level
Program Memory	Address 0000h must be NOP
	Last 100h words
Data Memory	0x070 (0x0F0, 0x170, 0x1F0) 0x1EB - 0x1EF

To use the In-Circuit Debugger function of the microcontroller, the design must implement In-Circuit Serial Programming connections to MCLR/VPP, VDD, GND, RB7 and RB6. This will interface to the In-Circuit Debugger module available from Microchip, or one of the third party development tool companies.

12.15 Program Verification/Code Protection

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out for verification purposes.

12.16 ID Locations

Four memory locations (2000h - 2003h) are designated as ID locations, where the user can store checksum or other code identification numbers. These locations are not accessible during normal execution, but are readable and writable during program/verify. It is recommended that only the 4 Least Significant bits of the ID location are used.



TABLE 15-2:	CLKOUT AND I/O TIMING REQUIREMENTS
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Param No.	Symbol	Charac	teristic	Min	Тур†	Мах	Units	Conditions
10*	TosH2ckL	OSC1↑ to CLKOUT↓	—	75	200	ns	(Note 1)	
11*	TosH2ck H	OSC1↑ to CLKOUT↑	—	75	200	ns	(Note 1)	
12*	TckR	CLKOUT rise time		—	35	100	ns	(Note 1)
13*	TckF	CLKOUT fall time		—	35	100	ns	(Note 1)
14*	TckL2ioV	CLKOUT ↓ to Port out vali	—	_	0.5TCY + 20	ns	(Note 1)	
15*	TioV2ckH	Port in valid before CLKO	Tosc + 200	_	—	ns	(Note 1)	
16*	TckH2iol	Port in hold after CLKOUT	0	_	—	ns	(Note 1)	
17*	TosH2ioV	OSC1↑ (Q1 cycle) to Port out valid	—	100	255	ns		
18*	TosH2iol	OSC1↑ (Q2 cycle) to	Standard (F)	100	_	—	ns	
		Port input invalid (I/O in hold time)	Extended (LF)	200	—	—	ns	
19*	TioV2osH	Port input valid to OSC1 [↑]	0	_	—	ns		
20*	TioR	Port output rise time	Standard (F)	—	10	40	ns	
			Extended (LF)	—	_	145	ns	
21*	TioF	Port output fall time	Standard (F)	—	10	40	ns	
			Extended (LF)			145	ns	
22††*	Tinp	INT pin high or low time	Тсү		—	ns		
23††*	Trbp	RB7:RB4 change INT high	n or low time	Тсү	_	—	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

t These parameters are asynchronous events not related to any internal clock edges.

Note 1: Measurements are taken in RC mode where CLKOUT output is 4 x Tosc.

TABLE 15-12:PIC16F87X-04 (COMMERCIAL, INDUSTRIAL, EXTENDED)PIC16F87X-10 (EXTENDED)PIC16F87X-20 (COMMERCIAL, INDUSTRIAL)PIC16LF87X-04 (COMMERCIAL, INDUSTRIAL)

Param No.	Sym	Characterist	ic	Min	Тур†	Max	Units	Conditions
A01	NR	Resolution		—	—	10-bits	bit	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$
A03	EIL	Integral linearity error		—	—	< ± 1	LSb	$\begin{array}{l} VREF=VDD=5.12V,\\ VSS\leqVAIN\leqVREF \end{array}$
A04	Edl	Differential linearity err	or	—	_	< ± 1	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$
A06	EOFF	Offset error		—	—	< ± 2	LSb	$\begin{array}{l} VREF=VDD=5.12V,\\ VSS\leqVAIN\leqVREF \end{array}$
A07	Egn	Gain error		_	—	< ± 1	LSb	$\begin{array}{l} VREF=VDD=5.12V,\\ VSS\leqVAIN\leqVREF \end{array}$
A10	_	Monotonicity ⁽³⁾		—	guaranteed	_	_	$V\text{SS} \leq V\text{AIN} \leq V\text{REF}$
A20	Vref	Reference voltage (VREF+ - VREF-)		2.0	—	Vdd + 0.3	V	Absolute minimum electrical spec. To ensure 10-bit accuracy.
A21	VREF+	Reference voltage High		AVDD - 2.5V		AVDD + 0.3V	V	
A22	Vref-	Reference voltage low		AVss - 0.3V		VREF+ - 2.0V	V	
A25	VAIN	Analog input voltage		Vss - 0.3 V	—	VREF + 0.3 V	V	
A30	Zain	Recommended impedance of analog voltage source		_	—	10.0	kΩ	
A40	IAD	A/D conversion	Standard	—	220		μΑ	Average current consumption
		current (VDD)	Extended	—	90		μΑ	when A/D is on (Note 1)
A50 IREF VREF input current (Note 2)		10	_	1000	μA	During VAIN acquisition. Based on differential of VHOLD to VAIN to charge CHOLD, see Section 11.1.		
		—	—	10	μA	During A/D Conversion cycle		

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: When A/D is off, it will not consume any current other than minor leakage current.

The power-down current spec includes any such leakage from the A/D module.

2: VREF current is from RA3 pin or VDD pin, whichever is selected as reference input.

3: The A/D conversion result never decreases with an increase in the input voltage, and has no missing codes.

PIC16F87X





FIGURE 16-12: TYPICAL AND MAXIMUM △ITMR1 vs. VDD OVER TEMPERATURE (-10°C TO 70°C, TIMER1 WITH OSCILLATOR, XTAL=32 kHZ, C1 AND C2=50 pF)





FIGURE 16-21: MINIMUM AND MAXIMUM VIN vs. VDD (ST INPUT, -40°C TO 125°C)





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28-Lead Plastic Small Outline (SO) - Wide, 300 mil (SOIC)

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units		INCHES*		MILLIMETERS			
Dimension	MIN	NOM	MAX	MIN	NOM	MAX		
Number of Pins	n		28			28		
Pitch	р		.050			1.27		
Overall Height	Α	.093	.099	.104	2.36	2.50	2.64	
Molded Package Thickness	A2	.088	.091	.094	2.24	2.31	2.39	
Standoff §	A1	.004	.008	.012	0.10	0.20	0.30	
Overall Width	Е	.394	.407	.420	10.01	10.34	10.67	
Molded Package Width	E1	.288	.295	.299	7.32	7.49	7.59	
Overall Length	D	.695	.704	.712	17.65	17.87	18.08	
Chamfer Distance	h	.010	.020	.029	0.25	0.50	0.74	
Foot Length	L	.016	.033	.050	0.41	0.84	1.27	
Foot Angle Top	¢	0	4	8	0	4	8	
Lead Thickness	С	.009	.011	.013	0.23	0.28	0.33	
Lead Width	В	.014	.017	.020	0.36	0.42	0.51	
Mold Draft Angle Top	α	0	12	15	0	12	15	
Mold Draft Angle Bottom	β	0	12	15	0	12	15	

* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-013 Drawing No. C04-052

44-Lead Plastic Metric Quad Flatpack (PQ) 10x10x2 mm Body, 1.6/0.15 mm Lead Form (MQFP)

For the most current package drawings, please see the Microchip Packaging Specification located Note: at http://www.microchip.com/packaging



		INCHES		MILLIMETERS*			
Dimension	MIN	NOM	MAX	MIN	NOM	MAX	
Number of Pins	n		44			44	
Pitch	р		.031			0.80	
Pins per Side	n1		11			11	
Overall Height	А	.079	.086	.093	2.00	2.18	2.35
Molded Package Thickness	A2	.077	.080	.083	1.95	2.03	2.10
Standoff §	A1	.002	.006	.010	0.05	0.15	0.25
Foot Length	L	.029	.035	.041	0.73	0.88	1.03
Footprint (Reference)	(F)		.063			1.60	
Foot Angle	ф	0	3.5	7	0	3.5	7
Overall Width	E	.510	.520	.530	12.95	13.20	13.45
Overall Length	D	.510	.520	.530	12.95	13.20	13.45
Molded Package Width	E1	.390	.394	.398	9.90	10.00	10.10
Molded Package Length	D1	.390	.394	.398	9.90	10.00	10.10
Lead Thickness	С	.005	.007	.009	0.13	0.18	0.23
Lead Width	В	.012	.015	.018	0.30	0.38	0.45
Pin 1 Corner Chamfer	СН	.025	.035	.045	0.64	0.89	1.14
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side.

JEDEC Equivalent: MS-022 Drawing No. C04-071

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