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Details

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Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	33
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	128 x 8
RAM Size	192 x 8
Voltage - Supply (Vcc/Vdd)	4V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16f874-20i-pt

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FI	GU	IRF	2-3	-
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PIC16F877/876 REGISTER FILE MAP

Indirect addr.(*)	00h	Indirect addr.(*)	80h	Indirect addr.(*)	100h	Indirect addr.(*)	18
TMR0	01h	OPTION_REG	81h	TMR0	101h	OPTION_REG	18
PCL	02h	PCL	82h	PCL	102h	PCL	182
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183
FSR	04h	FSR	84h	FSR	104h	FSR	184
PORTA	05h	TRISA	85h		105h		18
PORTB	06h	TRISB	86h	PORTB	106h	TRISB	186
PORTC	07h	TRISC	87h		107h		187
PORTD ⁽¹⁾	08h	TRISD ⁽¹⁾	88h		108h		188
PORTE ⁽¹⁾	09h	TRISE ⁽¹⁾	89h		109h		189
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18/
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18
PIR1	0Ch	PIE1	8Ch	EEDATA	10Ch	EECON1	180
PIR2	0Dh	PIE2	8Dh	EEADR	10Dh	EECON2	181
TMR1L	0Eh	PCON	8Eh	EEDATH	10Eh	Reserved ⁽²⁾	18
TMR1H	0Fh		8Fh	EEADRH	10Fh	Reserved ⁽²⁾	18
T1CON	10h		90h		110h		190
TMR2	11h	SSPCON2	91h		111h		19 [.]
T2CON	12h	PR2	92h		112h		192
SSPBUF	13h	SSPADD	93h		113h		193
SSPCON	14h	SSPSTAT	94h		114h		194
CCPR1L	15h		95h		115h		19
CCPR1H	16h		96h		116h		196
CCP1CON	17h		97h	General	117h	General	197
RCSTA	18h	TXSTA	98h	Purpose Register	118h	Purpose Register	198
TXREG	19h	SPBRG	99h	16 Bytes	119h	16 Bytes	199
RCREG	1Ah		9Ah	-	11Ah		19/
CCPR2L	1Bh		9Bh		11Bh		198
CCPR2H	1Ch		9Ch		11Ch		190
CCP2CON	1Dh		9Dh		11Dh		19[
ADRESH	1Eh	ADRESL	9Eh		11Eh		19
ADCON0	1Fh	ADCON1	9Fh		11Fh		191
	20h		A0h		120h		1A
General Purpose Register		General Purpose Register 80 Bytes		General Purpose Register 80 Bytes		General Purpose Register 80 Bytes	
96 Bytes		UU Dyico	EFh	00 Dytos	16Fh	00 Dyi00	1EI
	7Fh	accesses 70h-7Fh	F0h FFh	accesses 70h-7Fh	170h 17Fh	accesses 70h - 7Fh	1F(1Fl
Bank 0		Bank 1	FFII	Bank 2	/. !!	Bank 3	11.1
1 1 m 1 m m 1	بارار منعم	a memory location	a				

2: These registers are reserved, maintain these registers clear.

FIGURE 2-4: PIC16F874/873 REGISTER FILE MAP

,	File Address	A	File ddress	/	File Address		File Addres
Indirect addr. ^(*)	00h	Indirect addr.(*)	80h	Indirect addr.(*)	100h	Indirect addr.(*)	180h
TMR0	01h	OPTION_REG	81h	TMR0	101h	OPTION_REG	181h
PCL	02h	PCL	82h	PCL	102h	PCL	182h
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183h
FSR	04h	FSR	84h	FSR	104h	FSR	184h
PORTA	05h	TRISA	85h		105h		185h
PORTB	06h	TRISB	86h	PORTB	106h	TRISB	186h
PORTC	07h	TRISC	87h		107h		187h
PORTD ⁽¹⁾	08h	TRISD ⁽¹⁾	88h		108h		188h
PORTE ⁽¹⁾	09h	TRISE ⁽¹⁾	89h		109h		189h
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18Ah
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18Bh
PIR1	0Ch	PIE1	8Ch	EEDATA	10Ch	EECON1	18Cł
PIR2	0Dh	PIE2	8Dh	EEADR	10Dh	EECON2	18Dł
TMR1L	0Eh	PCON	8Eh	EEDATH	10Eh	Reserved ⁽²⁾	18Eh
TMR1H	0Fh		8Fh	EEADRH	10Fh	Reserved ⁽²⁾	18Fh
T1CON	10h		90h		110h		190h
TMR2	11h	SSPCON2	91h				
T2CON	12h	PR2	92h				
SSPBUF	13h	SSPADD	93h				
SSPCON	14h	SSPSTAT	94h				
CCPR1L	15h		95h				
CCPR1H	16h		96h				
CCP1CON	17h		97h				
RCSTA	18h	TXSTA	98h				
TXREG	19h	SPBRG	99h				
RCREG	1Ah		9Ah				
CCPR2L	1Bh		9Bh				
CCPR2H	1Ch		9Ch				
CCP2CON	1Dh		9Dh				
ADRESH	1Eh	ADRESL	9Eh				
ADCON0	1Fh	ADCON1	9Fh		1206		1A0h
	20h		A0h		120h		
General Purpose Register		General Purpose Register		accesses 20h-7Fh		accesses A0h - FFh	
96 Bytes		96 Bytes		2011 11 11	16Fh 170h		1EFt 1F0h
	754				1756		4
Bank 0	J 7Fh	Bank 1	FFh	Bank 2	17Fh	Bank 3	1FFł
* Not a Note 1: These	physical reg e registers	data memory locat gister. are not implemen are reserved, mai	ted on the	e PIC16F873.			

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Details on page:
Bank 1											
80h ⁽³⁾	INDF	Addressing	g this location	n uses conte	ents of FSR to	address dat	a memory (no	a physical r	egister)	0000 0000	27
81h	OPTION_REG	RBPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	19
82h ⁽³⁾	PCL	Program C	Counter (PC)	Least Signif	icant Byte					0000 0000	26
83h ⁽³⁾	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	18
84h ⁽³⁾	FSR	Indirect Da	rect Data Memory Address Pointer								27
85h	TRISA	_		PORTA Da	ta Direction R	egister				11 1111	29
86h	TRISB	PORTB Da	ata Direction	Register						1111 1111	31
87h	TRISC	PORTC D	ata Direction	Register						1111 1111	33
88h ⁽⁴⁾	TRISD	PORTD D	ata Direction	Register						1111 1111	35
89h ⁽⁴⁾	TRISE	IBF	OBF	IBOV	PSPMODE		PORTE Data	Direction Bi	its	0000 -111	37
8Ah ^(1,3)	PCLATH	_	_	_	Write Buffer	for the uppe	r 5 bits of the F	Program Cou	unter	0 0000	26
8Bh ⁽³⁾	INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	20
8Ch	PIE1	PSPIE ⁽²⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	21
8Dh	PIE2	_	(5)		EEIE	BCLIE	_		CCP2IE	-r-0 00	23
8Eh	PCON	_	— — — — — — POR BOR								25
8Fh	—	Unimplem	ented							_	—
90h	—	Unimplem	ented							_	_
91h	SSPCON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0000 0000	68
92h	PR2	Timer2 Pe	riod Register	r						1111 1111	55
93h	SSPADD	Synchrono	ous Serial Po	ort (I ² C mode) Address Re	gister				0000 0000	73, 74
94h	SSPSTAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	0000 0000	66
95h	_	Unimplem	ented							_	_
96h	—	Unimplem	ented							_	_
97h	—	Unimplem	ented							_	_
98h	TXSTA	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	0000 -010	95
99h	SPBRG	Baud Rate	e Generator F	Register						0000 0000	97
9Ah	—	Unimplem	ented							_	_
9Bh	—	Unimplem	ented							_	_
9Ch	—	Unimplem	ented							_	_
9Dh	—	Unimplem	ented							_	_
9Eh	ADRESL	A/D Result	t Register Lo	w Byte						xxxx xxxx	116
9Fh	ADCON1	ADFM	_	_	—	PCFG3	PCFG2	PCFG1	PCFG0	0 0000	112

TABLE 2-1: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.
Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8> whose contents are transferred to the upper byte of the program counter.
2: Bits PSPIE and PSPIF are reserved on PIC16F873/876 devices; always maintain these bits clear.
3: These registers can be addressed from any bank.
4: PORTD, PORTE, TRISD, and TRISE are not physically implemented on PIC16F873/876 devices; read as '0'.
5: PIR2<6> and PIE2<6> are reserved on these devices; always maintain these bits clear.

3.3 PORTC and the TRISC Register

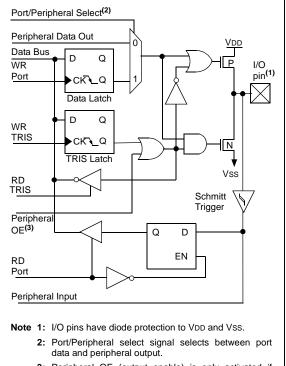
PORTC is an 8-bit wide, bi-directional port. The corresponding data direction register is TRISC. Setting a TRISC bit (= 1) will make the corresponding PORTC pin an input (i.e., put the corresponding output driver in a Hi-Impedance mode). Clearing a TRISC bit (= 0) will make the corresponding PORTC pin an output (i.e., put the contents of the output latch on the selected pin).

PORTC is multiplexed with several peripheral functions (Table 3-5). PORTC pins have Schmitt Trigger input buffers.

When the I^2C module is enabled, the PORTC<4:3> pins can be configured with normal I^2C levels, or with SMBus levels by using the CKE bit (SSPSTAT<6>).

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTC pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. Since the TRIS bit override is in effect while the peripheral is enabled, read-modify-write instructions (BSF, BCF, XORWF) with TRISC as destination, should be avoided. The user should refer to the corresponding peripheral section for the correct TRIS bit settings.

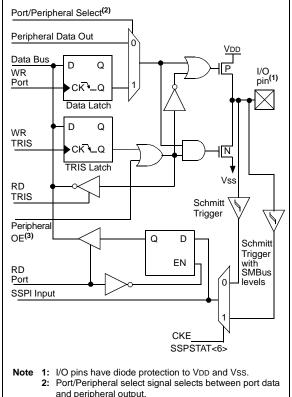
FIGURE 3-5: PORTC BLOCK DIAGRAM (PERIPHERAL OUTPUT OVERRIDE) RC<2:0>, RC<7:5>



3: Peripheral OE (output enable) is only activated if peripheral select is active.

FIGURE 3-6:

PORTC BLOCK DIAGRAM (PERIPHERAL OUTPUT OVERRIDE) RC<4:3>



 Peripheral OE (output enable) is only activated if peripheral select is active.

3.5 PORTE and TRISE Register

PORTE and TRISE are not implemented on the PIC16F873 or PIC16F876.

PORTE has three pins (RE0/RD/AN5, RE1/WR/AN6, and RE2/CS/AN7) which are individually configureable as inputs or outputs. These pins have Schmitt Trigger input buffers.

The PORTE pins become the I/O control inputs for the microprocessor port when bit PSPMODE (TRISE<4>) is set. In this mode, the user must make certain that the TRISE<2:0> bits are set, and that the pins are configured as digital inputs. Also ensure that ADCON1 is configured for digital I/O. In this mode, the input buffers are TTL.

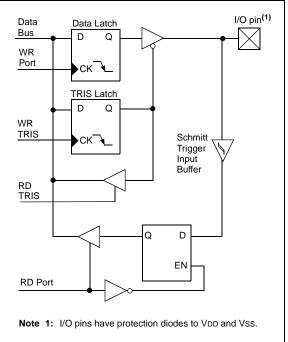
Register 3-1 shows the TRISE register, which also controls the parallel slave port operation.

PORTE pins are multiplexed with analog inputs. When selected for analog input, these pins will read as '0's.

TRISE controls the direction of the RE pins, even when they are being used as analog inputs. The user must make sure to keep the pins configured as inputs when using them as analog inputs.

Note: On a Power-on Reset, these pins are configured as analog inputs, and read as '0'.

FIGURE 3-8: PORTE BLOCK DIAGRAM (IN I/O PORT MODE)



Name	Bit#	Buffer Type	Function
RE0/RD/AN5	bit0	ST/TTL ⁽¹⁾	$\frac{I/O}{RD}$ port pin or read control input in Parallel Slave Port mode or analog input: 1 = Idle
REU/RD/ANS	bito	51/112 /	 a line a Read operation. Contents of PORTD register are output to PORTD I/O pins (if chip selected)
RE1/WR/AN6	bit1	ST/TTL ⁽¹⁾	 I/O port pin or write control input in Parallel Slave Port mode or analog input: WR 1 = Idle 0 = Write operation. Value of PORTD I/O pins is latched into PORTD register (if chip selected)
RE2/CS/AN7	bit2	ST/TTL ⁽¹⁾	I/O port pin or chip select control input in Parallel Slave Port mode or analog input: CS 1 = Device is not selected 0 = Device is selected

TABLE 3-9:PORTE FUNCTIONS

Legend: ST = Schmitt Trigger input, TTL = TTL input

Note 1: Input buffers are Schmitt Triggers when in I/O mode and TTL buffers when in Parallel Slave Port mode.

TABLE 3-10: SUMMARY OF REGISTERS ASSOCIATED WITH PORTE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other RESETS
09h	PORTE	—	—	_	—	_	RE2	RE1	RE0	xxx	uuu
89h	TRISE	IBF	OBF	IBOV	PSPMODE	_	PORTE [PORTE Data Direction Bits		0000 -111	0000 -111
9Fh	ADCON1	ADFM	—		_	PCFG3	PCFG2	PCFG1	PCFG0	0- 0000	0- 0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by PORTE.

NOTES:

4.0 DATA EEPROM AND FLASH PROGRAM MEMORY

The Data EEPROM and FLASH Program Memory are readable and writable during normal operation over the entire VDD range. These operations take place on a single byte for Data EEPROM memory and a single word for Program memory. A write operation causes an erase-then-write operation to take place on the specified byte or word. A bulk erase operation may not be issued from user code (which includes removing code protection).

Access to program memory allows for checksum calculation. The values written to program memory do not need to be valid instructions. Therefore, up to 14-bit numbers can be stored in memory for use as calibration parameters, serial numbers, packed 7-bit ASCII, etc. Executing a program memory location containing data that form an invalid instruction, results in the execution of a NOP instruction.

The EEPROM Data memory is rated for high erase/ write cycles (specification D120). The FLASH program memory is rated much lower (specification D130), because EEPROM data memory can be used to store frequently updated values. An on-chip timer controls the write time and it will vary with voltage and temperature, as well as from chip to chip. Please refer to the specifications for exact limits (specifications D122 and D133).

A byte or word write automatically erases the location and writes the new value (erase before write). Writing to EEPROM data memory does not impact the operation of the device. Writing to program memory will cease the execution of instructions until the write is complete. The program memory cannot be accessed during the write. During the write operation, the oscillator continues to run, the peripherals continue to function and interrupt events will be detected and essentially "queued" until the write is complete. When the write completes, the next instruction in the pipeline is executed and the branch to the interrupt vector will take place, if the interrupt is enabled and occurred during the write.

Read and write access to both memories take place indirectly through a set of Special Function Registers (SFR). The six SFRs used are:

- EEDATA
- EEDATH
- EEADR
- EEADRH
- EECON1
- EECON2

The EEPROM data memory allows byte read and write operations without interfering with the normal operation of the microcontroller. When interfacing to EEPROM data memory, the EEADR register holds the address to be accessed. Depending on the operation, the EEDATA register holds the data to be written, or the data read, at the address in EEADR. The PIC16F873/874 devices have 128 bytes of EEPROM data memory and therefore, require that the MSb of EEADR remain clear. The EEPROM data memory on these devices do not wrap around to 0, i.e., 0x80 in the EEADR does not map to 0x00. The PIC16F876/877 devices have 256 bytes of EEPROM data memory and therefore, uses all 8-bits of the EEADR.

The FLASH program memory allows non-intrusive read access, but write operations cause the device to stop executing instructions, until the write completes. When interfacing to the program memory, the EEADRH:EEADR registers form a two-byte word, which holds the 13-bit address of the memory location being accessed. The register combination of EEDATH:EEDATA holds the 14-bit data for writes, or reflects the value of program memory after a read operation. Just as in EEPROM data memory accesses, the value of the EEADRH:EEADR registers must be within the valid range of program memory, depending on the device: 0000h to 1FFFh for the PIC16F873/874. or 0000h to 3FFFh for the PIC16F876/877. Addresses outside of this range do not wrap around to 0000h (i.e., 4000h does not map to 0000h on the PIC16F877).

4.1 EECON1 and EECON2 Registers

The EECON1 register is the control register for configuring and initiating the access. The EECON2 register is not a physically implemented register, but is used exclusively in the memory write sequence to prevent inadvertent writes.

There are many bits used to control the read and write operations to EEPROM data and FLASH program memory. The EEPGD bit determines if the access will be a program or data memory access. When clear, any subsequent operations will work on the EEPROM data memory. When set, all subsequent operations will operate in the program memory.

Read operations only use one additional bit, RD, which initiates the read operation from the desired memory location. Once this bit is set, the value of the desired memory location will be available in the data registers. This bit cannot be cleared by firmware. It is automatically cleared at the end of the read operation. For EEPROM data memory reads, the data will be available in the EEDATA register in the very next instruction cycle after the RD bit is set. For program memory reads, the data will be loaded into the EEDATH:EEDATA registers, following the second instruction after the RD bit is set.

4.4 Reading the FLASH Program Memory

Reading FLASH program memory is much like that of EEPROM data memory, only two NOP instructions must be inserted after the RD bit is set. These two instruction cycles that the NOP instructions execute, will be used by the microcontroller to read the data out of program the memory and insert value into the EEDATH:EEDATA registers. Data will be available following the second NOP instruction. EEDATH and EEDATA will hold their value until another read operation is initiated, or until they are written by firmware.

The steps to reading the FLASH program memory are:

- 1. Write the address to EEADRH:EEADR. Make sure that the address is not larger than the memory size of the PIC16F87X device.
- 2. Set the EEPGD bit to point to FLASH program memory.
- 3. Set the RD bit to start the read operation.
- 4. Execute two NOP instructions to allow the microcontroller to read out of program memory.
- 5. Read the data from the EEDATH:EEDATA registers.

EXAMPLE 4-3: FLASH PROGRAM READ

BSF	STATUS, RP1	;						
BCF	STATUS, RPO	;Bank 2						
MOVF	ADDRL, W	;Write the						
MOVWF	EEADR	;address bytes						
MOVF	ADDRH,W	;for the desired						
MOVWF	EEADRH	;address to read						
BSF	STATUS, RPO	;Bank 3						
BSF	EECON1, EEPGD	;Point to Program memory						
BSF	EECON1, RD	;Start read operation						
NOP		;Required two NOPs						
NOP		;						
BCF	STATUS, RPO	;Bank 2						
MOVF	EEDATA, W	;DATAL = EEDATA						
MOVWF	DATAL	;						
MOVF	EEDATH,W	;DATAH = EEDATH						
MOVWF	DATAH	;						

4.5 Writing to the FLASH Program Memory

Writing to FLASH program memory is unique, in that the microcontroller does not execute instructions while programming is taking place. The oscillator continues to run and all peripherals continue to operate and queue interrupts, if enabled. Once the write operation completes (specification D133), the processor begins executing code from where it left off. The other important difference when writing to FLASH program memory, is that the WRT configuration bit, when clear, prevents any writes to program memory (see Table 4-1).

Just like EEPROM data memory, there are many steps in writing to the FLASH program memory. Both address and data values must be written to the SFRs. The EEPGD bit must be set, and the WREN bit must be set to enable writes. The WREN bit should be kept clear at all times, except when writing to the FLASH Program memory. The WR bit can only be set if the WREN bit was set in a previous operation, i.e., they both cannot be set in the same operation. The WREN bit should then be cleared by firmware after the write. Clearing the WREN bit before the write actually completes will not terminate the write in progress.

Writes to program memory must also be prefaced with a special sequence of instructions that prevent inadvertent write operations. This is a sequence of five instructions that must be executed without interruption for each byte written. These instructions must then be followed by two NOP instructions to allow the microcontroller to setup for the write operation. Once the write is complete, the execution of instructions starts with the instruction after the second NOP.

The steps to write to program memory are:

- 1. Write the address to EEADRH:EEADR. Make sure that the address is not larger than the memory size of the PIC16F87X device.
- 2. Write the 14-bit data value to be programmed in the EEDATH:EEDATA registers.
- 3. Set the EEPGD bit to point to FLASH program memory.
- 4. Set the WREN bit to enable program operations.
- 5. Disable interrupts (if enabled).
- 6. Execute the special five instruction sequence:
 - Write 55h to EECON2 in two steps (first to W, then to EECON2)
 - Write AAh to EECON2 in two steps (first to W, then to EECON2)
 - Set the WR bit
- 7. Execute two NOP instructions to allow the microcontroller to setup for write operation.
- 8. Enable interrupts (if using interrupts).
- 9. Clear the WREN bit to disable program operations.

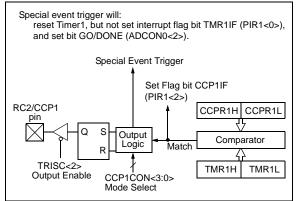
8.2 Compare Mode

In Compare mode, the 16-bit CCPR1 register value is constantly compared against the TMR1 register pair value. When a match occurs, the RC2/CCP1 pin is:

- Driven high
- Driven low
- Remains unchanged

The action on the pin is based on the value of control bits CCP1M3:CCP1M0 (CCP1CON<3:0>). At the same time, interrupt flag bit CCP1IF is set.

FIGURE 8-2: COMPARE MODE OPERATION BLOCK DIAGRAM



8.2.1 CCP PIN CONFIGURATION

The user must configure the RC2/CCP1 pin as an output by clearing the TRISC<2> bit.

Note: Clearing the CCP1CON register will force the RC2/CCP1 compare output latch to the default low level. This is not the PORTC I/O data latch.

8.2.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode, or Synchronized Counter mode, if the CCP module is using the compare feature. In Asynchronous Counter mode, the compare operation may not work.

8.2.3 SOFTWARE INTERRUPT MODE

When Generate Software Interrupt mode is chosen, the CCP1 pin is not affected. The CCPIF bit is set, causing a CCP interrupt (if enabled).

8.2.4 SPECIAL EVENT TRIGGER

In this mode, an internal hardware trigger is generated, which may be used to initiate an action.

The special event trigger output of CCP1 resets the TMR1 register pair. This allows the CCPR1 register to effectively be a 16-bit programmable period register for Timer1.

The special event trigger output of CCP2 resets the TMR1 register pair and starts an A/D conversion (if the A/D module is enabled).

Note: The special event trigger from the CCP1and CCP2 modules will not set interrupt flag bit TMR1IF (PIR1<0>).

PIC16F87X

8.3.3 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

- 1. Set the PWM period by writing to the PR2 register.
- 2. Set the PWM duty cycle by writing to the CCPR1L register and CCP1CON<5:4> bits.
- 3. Make the CCP1 pin an output by clearing the TRISC<2> bit.
- 4. Set the TMR2 prescale value and enable Timer2 by writing to T2CON.
- 5. Configure the CCP1 module for PWM operation.

TABLE 8-3: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 20 MHz

PWM Frequency	1.22 kHz	4.88 kHz	19.53 kHz	78.12kHz	156.3 kHz	208.3 kHz
Timer Prescaler (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0xFFh	0xFFh	0xFFh	0x3Fh	0x1Fh	0x17h
Maximum Resolution (bits)	10	10	10	8	7	5.5

TABLE 8-4: REGISTERS ASSOCIATED WITH CAPTURE, COMPARE, AND TIMER1

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other RESETS
0Bh,8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	0000 000x	0000 000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
0Dh	PIR2	—	_	_	_	—	—	_	CCP2IF	0	0
8Ch	PIE1	PSPIE ⁽¹⁾	PSPIE ⁽¹⁾ ADIE RCIE TXIE SSPIE CCP1IE TMR2IE TMR1IE						0000 0000	0000 0000	
8Dh	PIE2	—	_	CCP2IE	0	0					
87h	TRISC	PORTC D	ata Direct		1111 1111	1111 1111					
0Eh	TMR1L	Holding R	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register								uuuu uuuu
0Fh	TMR1H	Holding R	egister for	r the Most S	Significant B	yte of the 16	6-bit TMR1	Register		xxxx xxxx	uuuu uuuu
10h	T1CON	—	—	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	00 0000	uu uuuu
15h	CCPR1L	Capture/C	ompare/F	WM Regist	ter1 (LSB)					XXXX XXXX	uuuu uuuu
16h	CCPR1H	Capture/C	ompare/F	WM Regist	ter1 (MSB)					xxxx xxxx	uuuu uuuu
17h	CCP1CON	—	—	CCP1X	CCP1Y	CCP1M3	CCP1M2	CCP1M1	CCP1M0	00 0000	00 0000
1Bh	CCPR2L	Capture/C	ompare/F	WM Regist	ter2 (LSB)	-			-	xxxx xxxx	uuuu uuuu
1Ch	CCPR2H	Capture/C	ompare/F	WM Regist	ter2 (MSB)					xxxx xxxx	uuuu uuuu
1Dh	CCP2CON	—	—	CCP2X	CCP2Y	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 0000	00 0000

 $\label{eq:legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by Capture and Timer1.$

Note 1: The PSP is not implemented on the PIC16F873/876; always maintain these bits clear.

9.1.1 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave (Processor 2, Figure 9-5) is to broad-cast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI module is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "line activity monitor".

The clock polarity is selected by appropriately programming bit CKP (SSPCON<4>). This then, would give waveforms for SPI communication as shown in Figure 9-6, Figure 9-8 and Figure 9-9, where the MSb is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 Tcy)
- Fosc/64 (or 16 Tcy)
- Timer2 output/2

This allows a maximum bit clock frequency (at 20 MHz) of 5.0 MHz.

Figure 9-6 shows the waveforms for Master mode. When CKE = 1, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPBUF is loaded with the received data is shown.

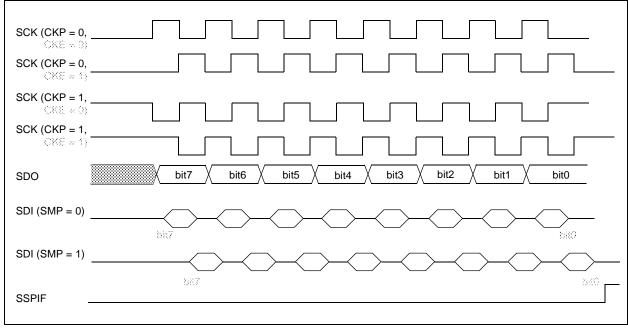
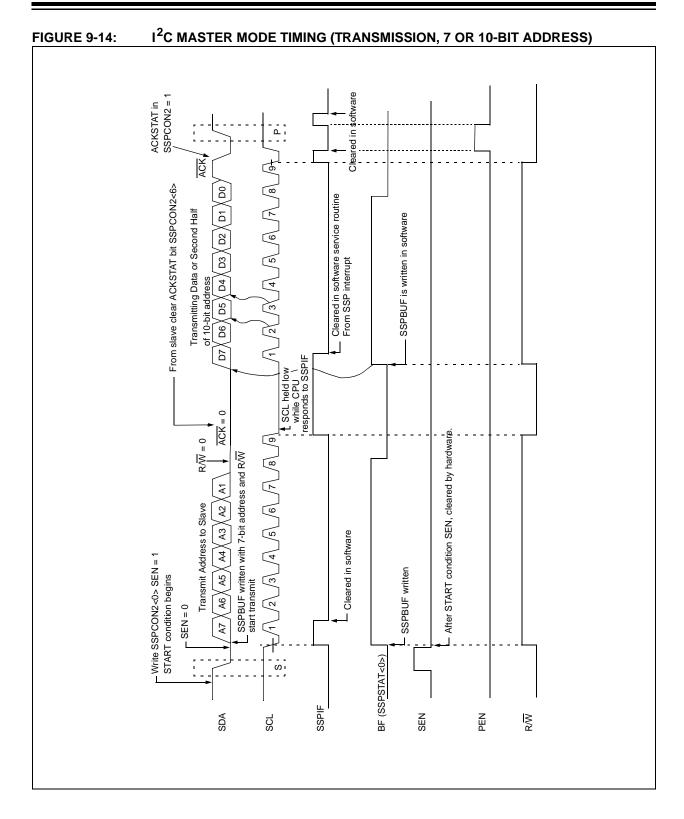


FIGURE 9-2: SPI MODE TIMING, MASTER MODE



9.2.18 MULTI -MASTER COMMUNICATION, BUS COLLISION, AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDA pin, arbitration takes place when the master outputs a '1' on SDA, by letting SDA float high and another master asserts a '0'. When the SCL pin floats high, data should be stable. If the expected data on SDA is a '1' and the data sampled on the SDA pin = '0', a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCLIF and reset the I^2C port to its IDLE state (Figure 9-19).

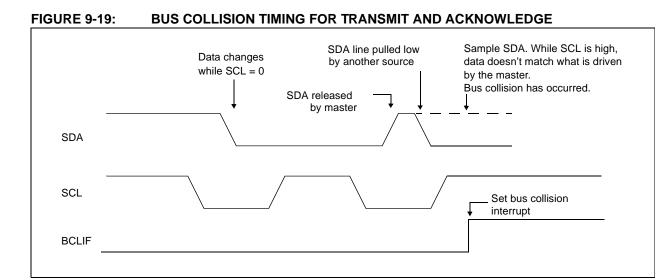
If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDA and SCL lines are de-asserted, and the SSPBUF can be written to. When the user services the bus collision Interrupt Service Routine, and if the I^2C bus is free, the user can resume communication by asserting a START condition.

If a START, Repeated START, STOP, or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDA and SCL lines are de-asserted, and the respective control bits in the SSPCON2 register are cleared. When the user services the bus collision Interrupt Service Routine, and if the l^2C bus is free, the user can resume communication by asserting a START condition.

The master will continue to monitor the SDA and SCL pins and if a STOP condition occurs, the SSPIF bit will be set.

A write to the SSPBUF will start the transmission of data at the first data bit, regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of START and STOP conditions allows the determination of when the bus is free. Control of the I^2C bus can be taken when the P bit is set in the SSPSTAT register, or the bus is idle and the S and P bits are cleared.



9.2.18.2 Bus Collision During a Repeated START Condition

During a Repeated START condition, a bus collision occurs if:

- a) A low level is sampled on SDA when SCL goes from low level to high level.
- SCL goes low before SDA is asserted low, indicating that another master is attempting to transmit a data '1'.

When the user de-asserts SDA and the pin is allowed to float high, the BRG is loaded with SSPADD<6:0> and counts down to 0. The SCL pin is then de-asserted, and when sampled high, the SDA pin is sampled. If SDA is low, a bus collision has occurred (i.e., another master is attempting to transmit a data'0'). If, however, SDA is sampled high, the BRG is reloaded and begins counting. If SDA goes from high to low before the BRG times out, no bus collision occurs, because no two masters can assert SDA at exactly the same time.

If, however, SCL goes from high to low before the BRG times out and SDA has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data'1' during the Repeated START condition.

If at the end of the BRG time-out, both SCL and SDA are still high, the SDA pin is driven low, the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCL pin, the SCL pin is driven low and the Repeated START condition is complete (Figure 9-23).

FIGURE 9-23: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)

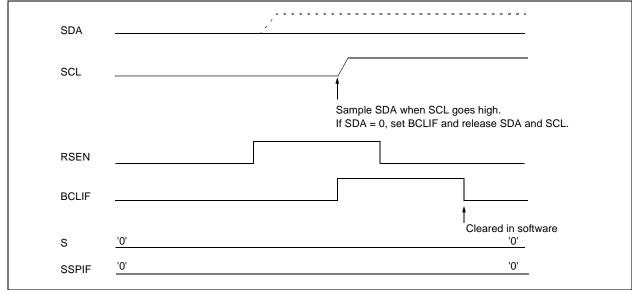
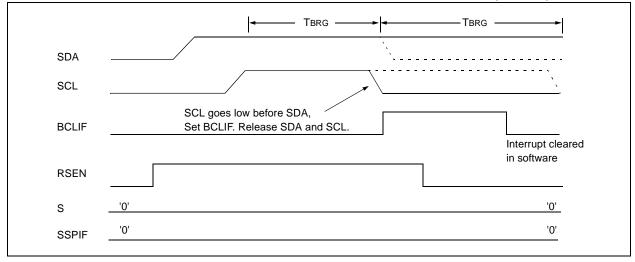


FIGURE 9-24: BUS COLLISION DURING REPEATED START CONDITION (CASE 2)



10.1 USART Baud Rate Generator (BRG)

The BRG supports both the Asynchronous and Synchronous modes of the USART. It is a dedicated 8-bit baud rate generator. The SPBRG register controls the period of a free running 8-bit timer. In Asynchronous mode, bit BRGH (TXSTA<2>) also controls the baud rate. In Synchronous mode, bit BRGH is ignored. Table 10-1 shows the formula for computation of the baud rate for different USART modes which only apply in Master mode (internal clock).

Given the desired baud rate and FOSC, the nearest integer value for the SPBRG register can be calculated using the formula in Table 10-1. From this, the error in baud rate can be determined.

It may be advantageous to use the high baud rate (BRGH = 1), even for slower baud clocks. This is because the FOSC/(16(X + 1)) equation can reduce the baud rate error in some cases.

Writing a new value to the SPBRG register causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate.

10.1.1 SAMPLING

The data on the RC7/RX/DT pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RX pin.

TABLE 10-1: BAUD RATE FORMULA

SYNC	BRGH = 0 (Low Speed)	BRGH = 1 (High Speed)
0	(Asynchronous) Baud Rate = Fosc/(64(X+1))	Baud Rate = Fosc/(16(X+1))
1	(Synchronous) Baud Rate = FOSC/(4(X+1))	N/A

X = value in SPBRG (0 to 255)

TABLE 10-2: REGISTERS ASSOCIATED WITH BAUD RATE GENERATOR

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other RESETS
98h	TXSTA	CSRC	TX9	TXEN	SYNC		BRGH	TRMT	TX9D	0000 -010	0000 -010
18h	RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	0000 000x
99h	SPBRG	Baud Rat	te Genera		0000 0000	0000 0000					

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used by the BRG.

TABLE 10-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other RESETS
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	R0IF	0000 000x	0000 000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
18h	RCSTA	SPEN	RX9	SREN	CREN	_	FERR	OERR	RX9D	0000 -00x	0000 -00x
19h	TXREG	USART Tr	USART Transmit Register							0000 0000	0000 0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
98h	TXSTA	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	Baud Rate Generator Register							0000 0000	0000 0000	

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission. **Note 1:** Bits PSPIE and PSPIF are reserved on PIC16F873/876 devices; always maintain these bits clear.

FIGURE 10-9: SYNCHRONOUS TRANSMISSION

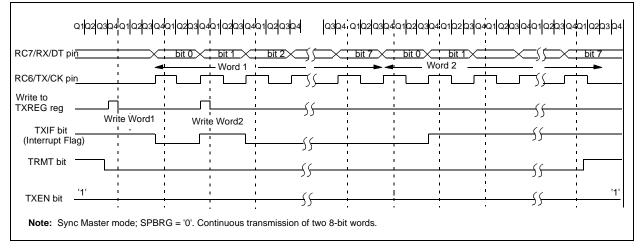
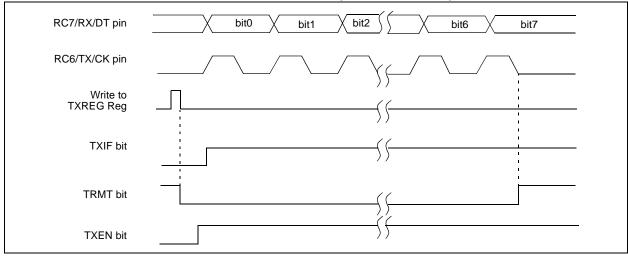


FIGURE 10-10: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)



	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1		Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4
OSC1	·/~~~~/			-		<u>;</u> ~~~~~~		
CLKOUT ⁽⁴⁾	\/\		<u> </u>	Tost(2)	/	\	¦/ \	
INT pin	· · ·	1		· ·		1 1	1 I 1 I	
INTF Flag (INTCON<1>)		י י י י	4	/		Interrupt Latency	(2)	
GIE bit (INTCON<7>)	1 I 1 I 1 I 1 I 1 I	i 	Processor i	n		·	1 1 1 1 1 1	ו ו ו ו
INSTRUCTIO	N FLOW			· ·		1	1 1 1 1	1
PC	Х РС Х	PC+1	ХР	C+2	PC+2	X PC + 2	<u>X 0004h X</u>	0005h
Instruction { Fetched {	Inst(PC) = SLEEP	Inst(PC + 1)		1	Inst(PC + 2)	1 1 1	Inst(0004h)	Inst(0005h)
Instruction	Inst(PC - 1)	SLEEP		, , ,	Inst(PC + 1)	Dummy cycle	Dummy cycle	Inst(0004h)
2: Tost	HS or LP oscillator n τ = 1024Tosc (drawi = '1' assumed. In th	ing not to scale)						

F

If GIE = '0', execution will continue in-line.

4: CLKOUT is not available in these osc modes, but shown here for timing reference.

12.14 In-Circuit Debugger

When the DEBUG bit in the configuration word is programmed to a '0', the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB® ICD. When the microcontroller has this feature enabled, some of the resources are not available for general use. Table 12-8 shows which features are consumed by the background debugger.

TABLE 12-8: DE	BUGGER RESOURCES
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I/O pins	RB6, RB7
Stack	1 level
Program Memory	Address 0000h must be NOP
	Last 100h words
Data Memory	0x070 (0x0F0, 0x170, 0x1F0) 0x1EB - 0x1EF

To use the In-Circuit Debugger function of the microcontroller, the design must implement In-Circuit Serial Programming connections to MCLR/VPP, VDD, GND, RB7 and RB6. This will interface to the In-Circuit Debugger module available from Microchip, or one of the third party development tool companies.

12.15 Program Verification/Code Protection

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out for verification purposes.

12.16 ID Locations

Four memory locations (2000h - 2003h) are designated as ID locations, where the user can store checksum or other code identification numbers. These locations are not accessible during normal execution, but are readable and writable during program/verify. It is recommended that only the 4 Least Significant bits of the ID location are used.

15.3

DC Characteristics: PIC16F873/874/876/877-04 (Extended) PIC16F873/874/876/877-10 (Extended) (Continued)

PIC16F873/874/876/877-04 PIC16F873/874/876/877-20 (Extended)			Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +125^{\circ}C$						
Param No.	Symbol Characteristic/ Device			Тур†	Max	Units	Conditions		
	IDD	Supply Current ^(2,5)							
D010			—	1.6	4	mA	RC osc configurations Fosc = 4 MHz, VDD = 5.5V		
D013			-	7	15	mA	HS osc configuration, Fosc = 10 MHz, VDD = 5.5V		
D015	ΔIBOR	Brown-out Reset Current ⁽⁶⁾	—	85	200	μΑ	BOR enabled, VDD = 5.0V		
	IPD	Power-down Current ^(3,5)							
D020A				10.5	60	μΑ	VDD = 4.0V, WDT enabled		
D021B				1.5	30	μA	VDD = 4.0V, WDT disabled		
D023	ΔIBOR	Brown-out Reset Current ⁽⁶⁾	—	85	200	μΑ	BOR enabled, VDD = 5.0V		

† Data is "Typ" column is at 5V, 25°C, unless otherwise stated. These parameters are for design guidance only, and are not tested.

Note 1: This is the limit to which VDD can be lowered without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading, switching rate, oscillator type, internal code execution pattern and temperature also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail to rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

- **3:** The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD and Vss.
- 4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in kOhm.
- **5:** Timer1 oscillator (when enabled) adds approximately 20 μA to the specification. This value is from characterization and is for design guidance only. This is not tested.
- 6: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.
- 7: When BOR is enabled, the device will operate correctly until the VBOR voltage trip point is reached.

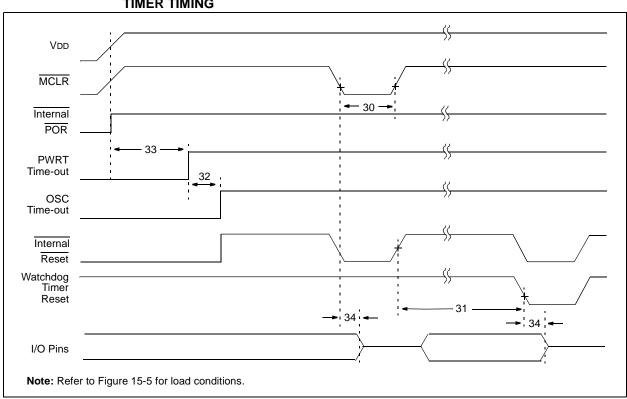


FIGURE 15-8: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

FIGURE 15-9: BROWN-OUT RESET TIMING

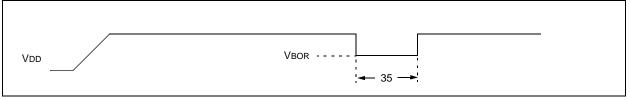


TABLE 15-3:RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER,
AND BROWN-OUT RESET REQUIREMENTS

Parameter No.	Symbol	Characteristic	Min	Тур†	Max	Units	Conditions
30	TmcL	MCLR Pulse Width (low)	2	-		μs	VDD = 5V, -40°C to +85°C
31*	Twdt	Watchdog Timer Time-out Period (No Prescaler)	7	18	33	ms	VDD = 5V, -40°C to +85°C
32	Tost	Oscillation Start-up Timer Period	_	1024 Tosc	_	—	Tosc = OSC1 period
33*	Tpwrt	Power-up Timer Period	28	72	132	ms	VDD = 5V, -40°C to +85°C
34	Tıoz	I/O Hi-impedance from MCLR Low or Watchdog Timer Reset	—	—	2.1	μS	
35	TBOR	Brown-out Reset pulse width	100	_		μs	$VDD \le VBOR (D005)$

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Parameter No.	Symbol	Characteristic			Тур	Max	Units	Conditions
90	Tsu:sta	START condition	100 kHz mode	4700	—	_	ns	Only relevant for Repeated
		Setup time	400 kHz mode	600	—	—		START condition
91	Thd:sta	START condition	100 kHz mode	4000	_	_	ns	After this period, the first clock
		Hold time	400 kHz mode	600	_	_		pulse is generated
92	Tsu:sto	STOP condition	100 kHz mode	4700	_	_	ns	
		Setup time	400 kHz mode	600	-	_		
93	Thd:sto	STOP condition	100 kHz mode	4000	-	_	ns	
		Hold time	400 kHz mode	600	_	_		

TABLE 15-8:	I ² C BUS START/STOP BITS REQUIREMENTS
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FIGURE 15-18: I²C BUS DATA TIMING

