AMD Xilinx - XCS05XL-5VQ100C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	100
Number of Logic Elements/Cells	238
Total RAM Bits	3200
Number of I/O	77
Number of Gates	5000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	100-TQFP
Supplier Device Package	100-VQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcs05xl-5vq100c

General Overview

Spartan series FPGAs are implemented with a regular, flexible, programmable architecture of Configurable Logic Blocks (CLBs), interconnected by a powerful hierarchy of versatile routing resources (routing channels), and surrounded by a perimeter of programmable Input/Output Blocks (IOBs), as seen in Figure 1. They have generous routing resources to accommodate the most complex interconnect patterns.

The devices are customized by loading configuration data into internal static memory cells. Re-programming is possible an unlimited number of times. The values stored in these memory cells determine the logic functions and interconnections implemented in the FPGA. The FPGA can either actively read its configuration data from an external serial PROM (Master Serial mode), or the configuration data can be written into the FPGA from an external device (Slave Serial mode).

Spartan series FPGAs can be used where hardware must be adapted to different user applications. FPGAs are ideal for shortening design and development cycles, and also offer a cost-effective solution for production rates well beyond 50,000 systems per month.

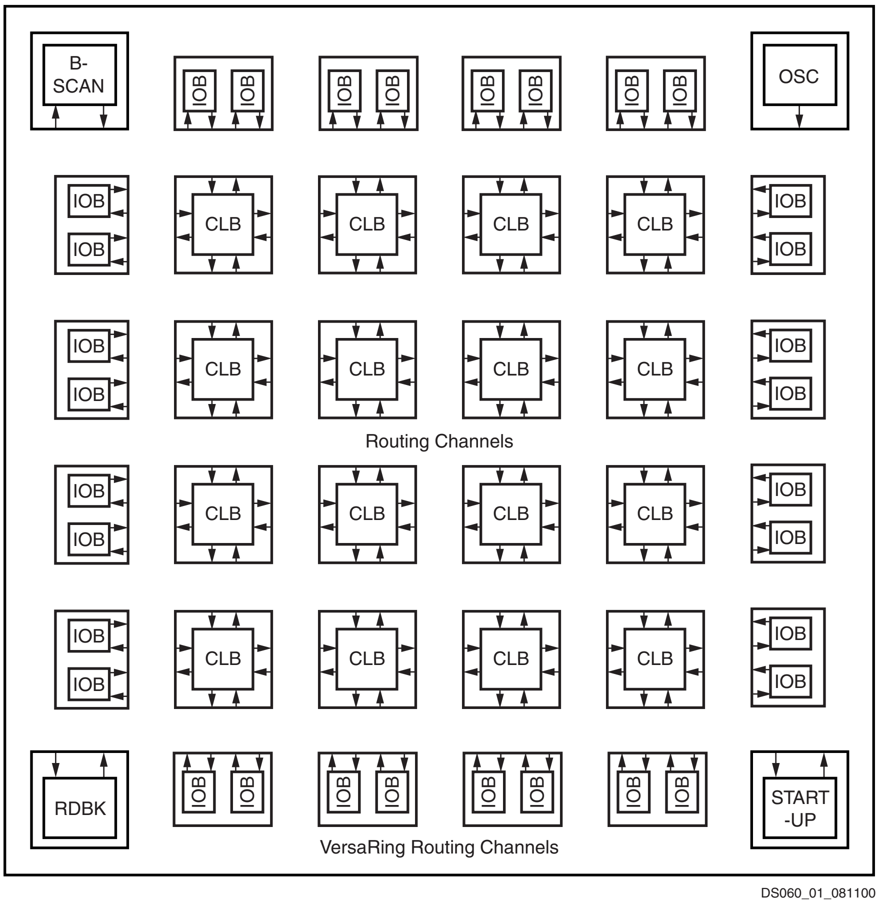


Figure 1: **Basic FPGA Block Diagram**

Spartan and Spartan-XL devices provide system clock rates exceeding 80 MHz and internal performance in excess of 150 MHz. In addition to the conventional benefit of high volume programmable logic solutions, Spartan series FPGAs also offer on-chip edge-triggered single-port and dual-port RAM, clock enables on all flip-flops, fast carry logic, and many other features.

The Spartan/XL families leverage the highly successful XC4000 architecture with many of that family's features and benefits. Technology advancements have been derived from the XC4000XLA process developments.

Logic Functional Description

The Spartan series uses a standard FPGA structure as shown in Figure 1, page 2. The FPGA consists of an array of configurable logic blocks (CLBs) placed in a matrix of routing channels. The input and output of signals is achieved through a set of input/output blocks (IOBs) forming a ring around the CLBs and routing channels.

- CLBs provide the functional elements for implementing the user's logic.
- IOBs provide the interface between the package pins and internal signal lines.
- Routing channels provide paths to interconnect the inputs and outputs of the CLBs and IOBs.

The functionality of each circuit block is customized during configuration by programming internal static memory cells. The values stored in these memory cells determine the logic functions and interconnections implemented in the FPGA.

Configurable Logic Blocks (CLBs)

The CLBs are used to implement most of the logic in an FPGA. The principal CLB elements are shown in the simplified block diagram in Figure 2. There are three look-up tables (LUT) which are used as logic function generators, two flip-flops and two groups of signal steering multiplexers. There are also some more advanced features provided by the CLB which will be covered in the **Advanced Features Description**, page 13.

Function Generators

Two 16 x 1 memory look-up tables (F-LUT and G-LUT) are used to implement 4-input function generators, each offering unrestricted logic implementation of any Boolean function of up to four independent input signals (F1 to F4 or G1 to G4). Using memory look-up tables the propagation delay is independent of the function implemented.

A third 3-input function generator (H-LUT) can implement any Boolean function of its three inputs. Two of these inputs are controlled by programmable multiplexers (see box "A" of Figure 2). These inputs can come from the F-LUT or G-LUT outputs or from CLB inputs. The third input always comes from a CLB input. The CLB can, therefore, implement certain functions of up to nine inputs, like parity checking. The three LUTs in the CLB can also be combined to do any arbitrarily defined Boolean function of five inputs.

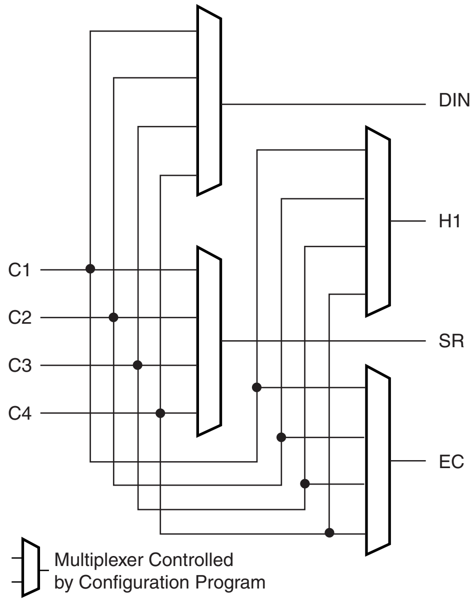


Figure 4: **CLB Control Signal Interface**

The four internal control signals are:

- EC: Enable Clock
- SR: Asynchronous Set/Reset or H function generator Input 0
- DIN: Direct In or H function generator Input 2
- H1: H function generator Input 1.

Input/Output Blocks (IOBs)

User-configurable input/output blocks (IOBs) provide the interface between external package pins and the internal logic. Each IOB controls one package pin and can be configured for input, output, or bidirectional signals. Figure 6 shows a simplified functional block diagram of the Spartan/XL FPGA IOB.

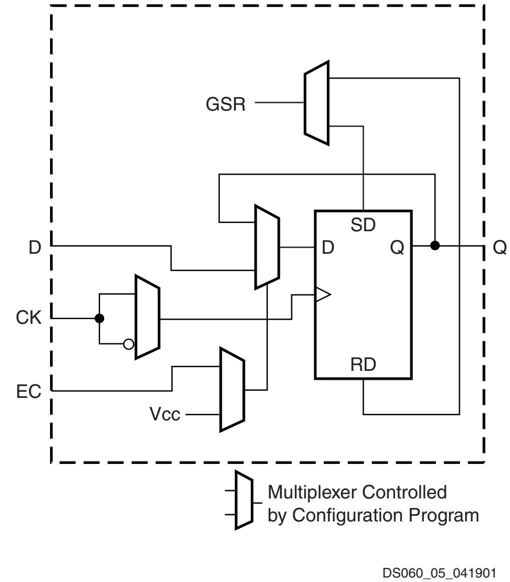


Figure 5: **IOB Flip-Flop/Latch Functional Block Diagram**

IOB Input Signal Path

The input signal to the IOB can be configured to either go directly to the routing channels (via I1 and I2 in Figure 6) or to the input register. The input register can be programmed as either an edge-triggered flip-flop or a level-sensitive latch. The functionality of this register is shown in Table 3, and a simplified block diagram of the register can be seen in Figure 5.

Table 3: **Input Register Functionality**

Mode	CK	EC	D	Q
Power-Up or GSR	X	X	X	SR
Flip-Flop	_/	1*	D	D
	0	X	X	Q
Latch	1	1*	X	Q
	0	1*	D	D
Both	X	0	X	Q

Legend:

X	Don't care.
_/	Rising edge (clock not inverted).
SR	Set or Reset value. Reset is default.
0*	Input is Low or unconnected (default value)
1*	Input is High or unconnected (default value)

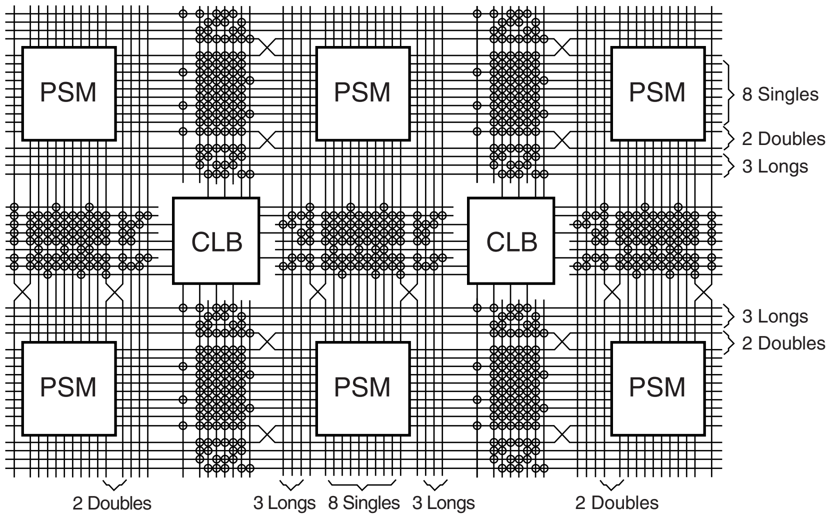


Figure 8: **Spartan/XL CLB Routing Channels and Interface Block Diagram**

CLB Interface

A block diagram of the CLB interface signals is shown in Figure 9. The input signals to the CLB are distributed evenly on all four sides providing maximum routing flexibility. In general, the entire architecture is symmetrical and regular. It is well suited to established placement and routing algorithms. Inputs, outputs, and function generators can freely swap positions within a CLB to avoid routing congestion during the placement and routing operation. The exceptions are the clock (K) input and CIN/COUT signals. The K input is routed to dedicated global vertical lines as well as four single-length lines and is on the left side of the CLB. The CIN/COUT signals are routed through dedicated interconnects which do not interfere with the general routing structure. The output signals from the CLB are available to drive both vertical and horizontal channels.

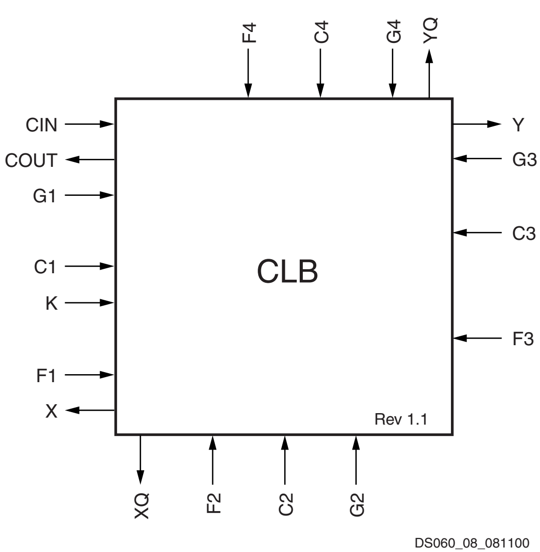


Figure 9: **CLB Interconnect Signals**

Programmable Switch Matrices

The horizontal and vertical single- and double-length lines intersect at a box called a programmable switch matrix (PSM). Each PSM consists of programmable pass transistors used to establish connections between the lines (see Figure 10).

For example, a single-length signal entering on the right side of the switch matrix can be routed to a single-length line on the top, left, or bottom sides, or any combination thereof, if multiple branches are required. Similarly, a double-length signal can be routed to a double-length line on any or all of the other three edges of the programmable switch matrix.

Single-Length Lines

Single-length lines provide the greatest interconnect flexibility and offer fast routing between adjacent blocks. There are eight vertical and eight horizontal single-length lines associated with each CLB. These lines connect the switching matrices that are located in every row and column of CLBs. Single-length lines are connected by way of the programmable switch matrices, as shown in Figure 10. Routing connectivity is shown in Figure 8.

Single-length lines incur a delay whenever they go through a PSM. Therefore, they are not suitable for routing signals for long distances. They are normally used to conduct signals within a localized area and to provide the branching for nets with fanout greater than one.

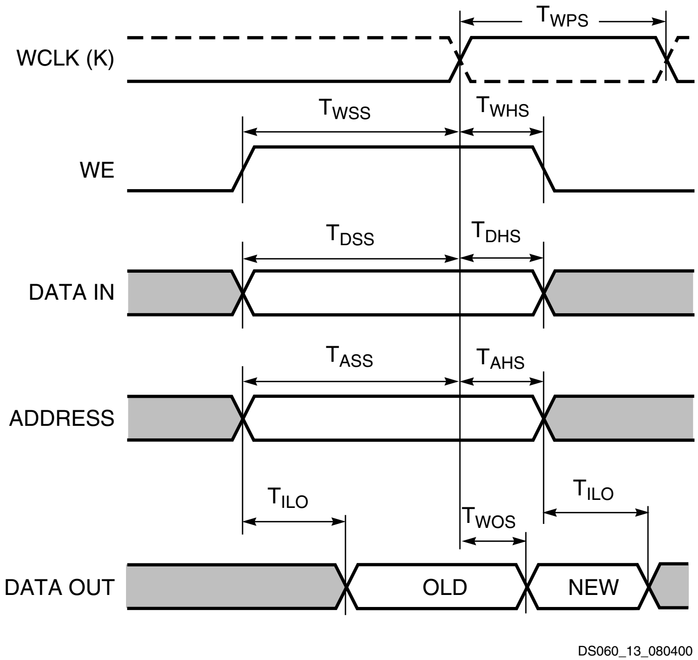


Figure 13: **Data Write and Access Timing for RAM**

WCLK can be configured as active on either the rising edge (default) or the falling edge. While the WCLK input to the RAM accepts the same signal as the clock input to the associated CLB's flip-flops, the sense of this WCLK input can be inverted with respect to the sense of the flip-flop clock inputs. Consequently, within the same CLB, data at the RAM SPO line can be stored in a flip-flop with either the same or the inverse clock polarity used to write data to the RAM.

The WE input is active High and cannot be inverted within the CLB.

Allowing for settling time, the data on the SPO output reflects the contents of the RAM location currently addressed. When the address changes, following the asynchronous delay T_{ILO}, the data stored at the new address location will appear on SPO. If the data at a particular RAM address is overwritten, after the delay T_{WOS}, the new data will appear on SPO.

Dual-Port Mode

In dual-port mode, the function generators (F-LUT and G-LUT) are used to create a 16 x 1 dual-port memory. Of the two data ports available, one permits read and write operations at the address specified by A[3:0] while the second provides only for read operations at the address specified independently by DPRA[3:0]. As a result, simultaneous read/write operations at different addresses (or even at the same address) are supported.

The functional organization of the 16 x 1 dual-port RAM is shown in Figure 14. The dual-port RAM signals and the

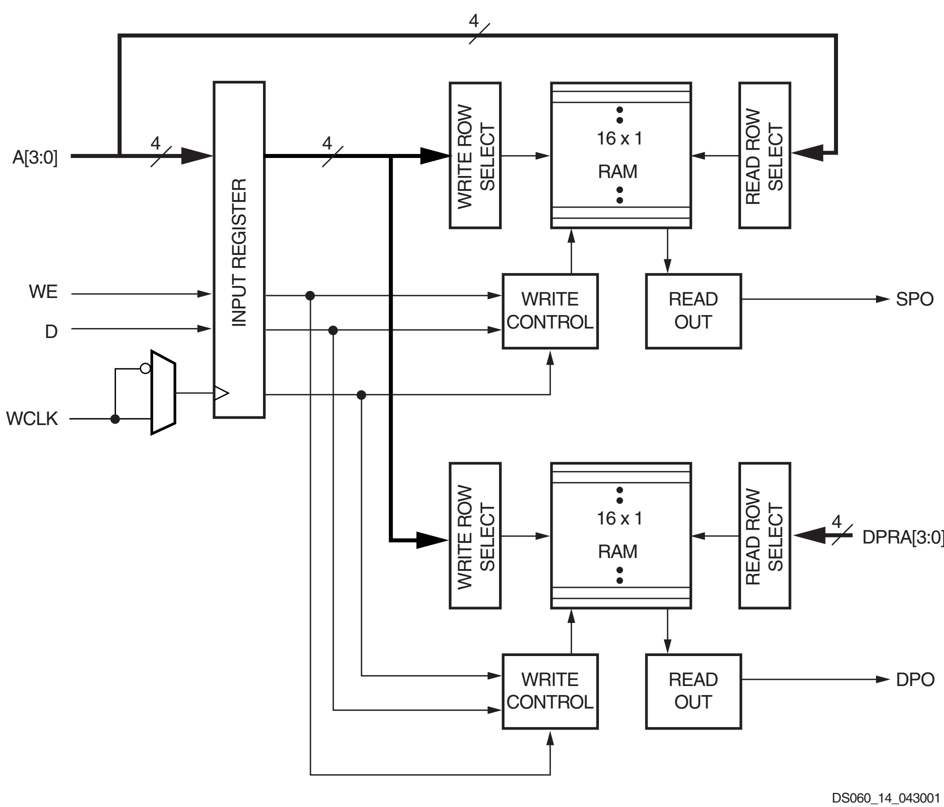


Figure 14: **Logic Diagram for the Dual-Port RAM**

CLB signals from which they are originally derived are shown in Table 10.

Table 10: **Dual-Port RAM Signals**

RAM Signal	Function	CLB Signal
D	Data In	DIN
A[3:0]	Read Address for Single-Port. Write Address for Single-Port and Dual-Port.	F[4:1]
DPRA[3:0]	Read Address for Dual-Port	G[4:1]
WE	Write Enable	SR
WCLK	Clock	K
SPO	Single Port Out (addressed by A[3:0])	F_{OUT}
DPO	Dual Port Out (addressed by DPRA[3:0])	G_{OUT}

The RAM16X1D primitive used to instantiate the dual-port RAM consists of an upper and a lower 16 x 1 memory array. The address port labeled A[3:0] supplies both the read and write addresses for the lower memory array, which behaves the same as the 16 x 1 single-port RAM array described previously. Single Port Out (SPO) serves as the data output for the lower memory. Therefore, SPO reflects the data at address A[3:0].

The other address port, labeled DPRA[3:0] for Dual Port Read Address, supplies the read address for the upper memory. The write address for this memory, however, comes from the address A[3:0]. Dual Port Out (DPO) serves as the data output for the upper memory. Therefore, DPO reflects the data at address DPRA[3:0].

By using A[3:0] for the write address and DPRA[3:0] for the read address, and reading only the DPO output, a FIFO that can read and write simultaneously is easily generated. The simultaneous read/write capability possible with the dual-port RAM can provide twice the effective data throughput of a single-port RAM alternating read and write operations.

The timing relationships for the dual-port RAM mode are shown in Figure 13.

Note that write operations to RAM are synchronous (edge-triggered); however, data access is asynchronous.

Initializing RAM at FPGA Configuration

Both RAM and ROM implementations in the Spartan/XL families are initialized during device configuration. The initial contents are defined via an INIT attribute or property attached to the RAM or ROM symbol, as described in the library guide. If not defined, all RAM contents are initialized to zeros, by default.

RAM initialization occurs only during device configuration. The RAM content is not affected by GSR.

More Information on Using RAM Inside CLBs

Three application notes are available from Xilinx that discuss synchronous (edge-triggered) RAM: *"Xilinx Edge-Triggered and Dual-Port RAM Capability," "Implementing FIFOs in Xilinx RAM,"* and *"Synchronous and Asynchronous FIFO Designs."* All three application notes apply to both the Spartan and the Spartan-XL families.

Fast Carry Logic

Each CLB F-LUT and G-LUT contains dedicated arithmetic logic for the fast generation of carry and borrow signals. This extra output is passed on to the function generator in the adjacent CLB. The carry chain is independent of normal routing resources. (See Figure 15.)

Dedicated fast carry logic greatly increases the efficiency and performance of adders, subtractors, accumulators, comparators and counters. It also opens the door to many new applications involving arithmetic operation, where the previous generations of FPGAs were not fast enough or too inefficient. High-speed address offset calculations in microprocessor or graphics systems, and high-speed addition in digital signal processing are two typical applications.

The two 4-input function generators can be configured as a 2-bit adder with built-in hidden carry that can be expanded to any length. This dedicated carry circuitry is so fast and efficient that conventional speed-up methods like carry generate/propagate are meaningless even at the 16-bit level, and of marginal benefit at the 32-bit level. This fast carry logic is one of the more significant features of the Spartan

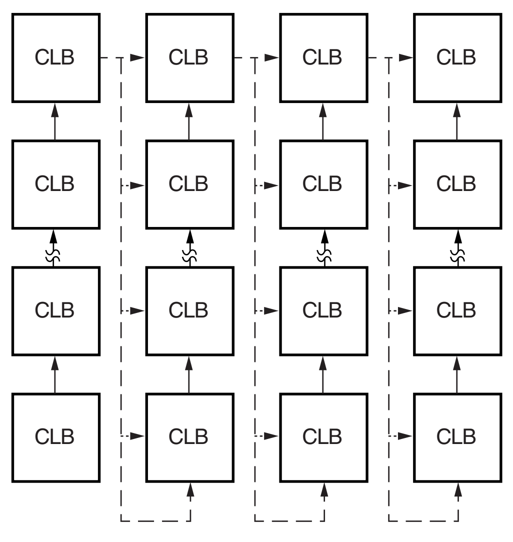


Figure 15: **Available Spartan/XL Carry Propagation Paths**

Figure 20 is a diagram of the Spartan/XL FPGA boundary scan logic. It includes three bits of Data Register per IOB, the IEEE 1149.1 Test Access Port controller, and the Instruction Register with decodes.

Spartan/XL devices can also be configured through the boundary scan logic. See **Configuration Through the Boundary Scan Pins**, page 37.

Data Registers

The primary data register is the boundary scan register. For each IOB pin in the FPGA, bonded or not, it includes three bits for In, Out and 3-state Control. Non-IOB pins have appropriate partial bit population for In or Out only. PROGRAM, CCLK and DONE are not included in the boundary scan register. Each EXTEST CAPTURE-DR state captures all In, Out, and 3-state pins.

The data register also includes the following non-pin bits: TDO.T, and TDO.O, which are always bits 0 and 1 of the data register, respectively, and BSCANT.UPD, which is always the last bit of the data register. These three boundary scan bits are special-purpose Xilinx test signals.

The other standard data register is the single flip-flop BYPASS register. It synchronizes data being passed through the FPGA to the next downstream boundary scan device.

The FPGA provides two additional data registers that can be specified using the BSCAN macro. The FPGA provides two user pins (BSCAN.SEL1 and BSCAN.SEL2) which are the decodes of two user instructions. For these instructions, two corresponding pins (BSCAN.TDO1 and BSCAN.TDO2) allow user scan data to be shifted out on TDO. The data register clock (BSCAN.DRCK) is available for control of test logic which the user may wish to implement with CLBs. The NAND of TCK and RUN-TEST-IDLE is also provided (BSCAN.IDLE).

Instruction Set

The Spartan/XL FPGA boundary scan instruction set also includes instructions to configure the device and read back the configuration data. The instruction set is coded as shown in Table 12.

Table 12: **Boundary Scan Instructions**

Instruction			Test Selected	TDO Source	I/O Data Source
I2	I1	I0			
0	0	0	EXTEST	DR	DR
0	0	1	SAMPLE/ PRELOAD	DR	Pin/Logic
0	1	0	USER 1	BSCAN. TDO1	User Logic
0	1	1	USER 2	BSCAN. TDO2	User Logic
1	0	0	READBACK	Readback Data	Pin/Logic
1	0	1	CONFIGURE	DOUT	Disabled
1	1	0	IDCODE (Spartan-XL only)	IDCODE Register	-
1	1	1	BYPASS	Bypass Register	-

Bit Sequence

The bit sequence within each IOB is: In, Out, 3-state. The input-only pins contribute only the In bit to the boundary scan I/O data register, while the output-only pins contributes all three bits.

The first two bits in the I/O data register are TDO.T and TDO.O, which can be used for the capture of internal signals. The final bit is BSCANT.UPD, which can be used to drive an internal net. These locations are primarily used by Xilinx for internal testing.

From a cavity-up view of the chip (as shown in the FPGA Editor), starting in the upper right chip corner, the boundary scan data-register bits are ordered as shown in Figure 21. The device-specific pinout tables for the Spartan/XL devices include the boundary scan locations for each IOB pin.

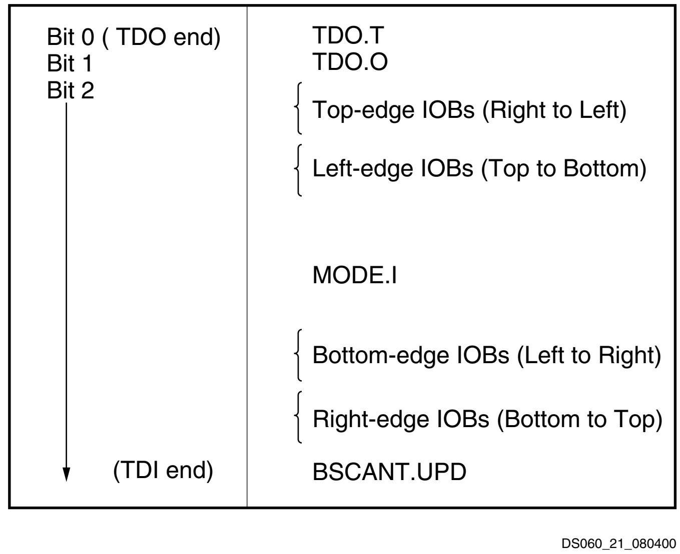


Figure 21: **Boundary Scan Bit Sequence**

BSDL (Boundary Scan Description Language) files for Spartan/XL devices are available on the Xilinx website in the File Download area. Note that the 5V Spartan devices and 3V Spartan-XL devices have different BSDL files.

Including Boundary Scan in a Design

If boundary scan is only to be used during configuration, no special elements need be included in the schematic or HDL code. In this case, the special boundary scan pins TDI, TMS, TCK and TDO can be used for user functions after configuration.

To indicate that boundary scan remain enabled after configuration, place the BSCAN library symbol and connect the TDI, TMS, TCK and TDO pad symbols to the appropriate pins, as shown in Figure 22.

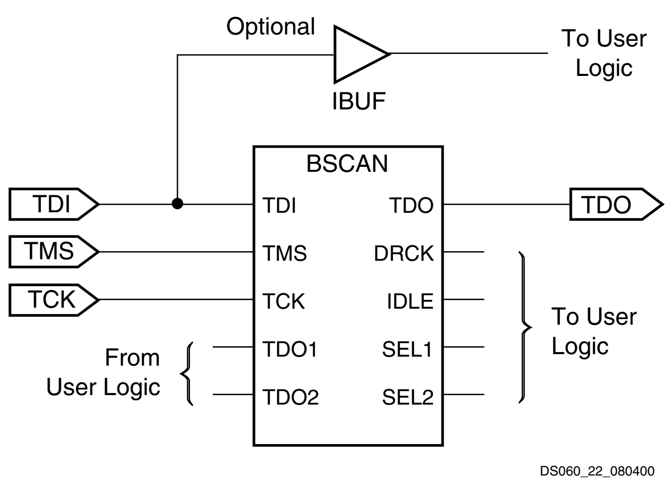


Figure 22: **Boundary Scan Example**

Even if the boundary scan symbol is used in a design, the input pins TMS, TCK, and TDI can still be used as inputs to be routed to internal logic. Care must be taken not to force the chip into an undesired boundary scan state by inadvertently applying boundary scan input patterns to these pins. The simplest way to prevent this is to keep TMS High, and then apply whatever signal is desired to TDI and TCK.

Avoiding Inadvertent Boundary Scan

If TMS or TCK is used as user I/O, care must be taken to ensure that at least one of these pins is held constant during configuration. In some applications, a situation may occur where TMS or TCK is driven during configuration. This may cause the device to go into boundary scan mode and disrupt the configuration process.

To prevent activation of boundary scan during configuration, do either of the following:

- TMS: Tie High to put the Test Access Port controller in a benign RESET state.
- TCK: Tie High or Low—do not toggle this clock input.

For more information regarding boundary scan, refer to the Xilinx Application Note, "*Boundary Scan in FPGA Devices*."

Boundary Scan Enhancements (Spartan-XL Family Only)

Spartan-XL devices have improved boundary scan functionality and performance in the following areas:

IDCODE: The IDCODE register is supported. By using the IDCODE, the device connected to the JTAG port can be determined. The use of the IDCODE enables selective configuration dependent on the FPGA found.

The IDCODE register has the following binary format:

```
vvvv:ffff:fffa:aaaa:aaaa:cccc:cccc:ccc1
```

where

c = the company code (49h for Xilinx)

a = the array dimension in CLBs (ranges from 0Ah for XCS05XL to 1Ch for XCS40XL)

f = the family code (02h for Spartan-XL family)

v = the die version number

Table 13: **IDCODEs Assigned to Spartan-XL FPGAs**

FPGA	IDCODE
XCS05XL	0040A093h
XCS10XL	0040E093h
XCS20XL	00414093h
XCS30XL	00418093h
XCS40XL	0041C093h

Configuration State: The configuration state is available to JTAG controllers.

Configuration Disable: The JTAG port can be prevented from configuring the FPGA.

TCK Startup: TCK can now be used to clock the start-up block in addition to other user clocks.

CCLK Holdoff: Changed the requirement for Boundary Scan Configure or EXTEST to be issued prior to the release of INIT pin and CCLK cycling.

Reissue Configure: The Boundary Scan Configure can be reissued to recover from an unfinished attempt to configure the device.

Bypass FF: Bypass FF and IOB is modified to provide DRCLOCK only during BYPASS for the bypass flip-flop, and during EXTEST or SAMPLE/PRELOAD for the IOB register.

Power-Down (Spartan-XL Family Only)

All Spartan/XL devices use a combination of efficient segmented routing and advanced process technology to provide low power consumption under all conditions. The 3.3V Spartan-XL family adds a dedicated active Low power-down pin ($\overline{PWRDWN}$) to reduce supply current to 100 μA typical. The $\overline{PWRDWN}$ pin takes advantage of one of the unused No Connect locations on the 5V Spartan device. The user must de-select the "5V Tolerant I/Os" option in the Configuration Options to achieve the specified Power Down current. The $\overline{PWRDWN}$ pin has a default internal pull-up resistor, allowing it to be left unconnected if unused.

V_{CC} must continue to be supplied during Power-down, and configuration data is maintained. When the $\overline{PWRDWN}$ pin is pulled Low, the input and output buffers are disabled. The inputs are internally forced to a logic Low level, including the MODE pins, DONE, CCLK, and TDO, and all internal pull-up resistors are turned off. The $\overline{PROGRAM}$ pin is not affected by Power Down. The GSR net is asserted during Power Down, initializing all the flip-flops to their start-up state.

$\overline{PWRDWN}$ has a minimum pulse width of 50 ns (Figure 23). On entering the Power-down state, the inputs will be disabled and the flip-flops set/reset, and then the outputs are disabled about 10 ns later. The user may prefer to assert the GTS or GSR signals before $\overline{PWRDWN}$ to affect the order of events. When the $\overline{PWRDWN}$ signal is returned High, the inputs will be enabled first, followed immediately by the release of the GSR signal initializing the flip-flops. About 10 ns later, the outputs will be enabled. Allow 50 ns after the release of $\overline{PWRDWN}$ before using the device.

Master Serial Mode

The Master serial mode uses an internal oscillator to generate a Configuration Clock (CCLK) for driving potential slave devices and the Xilinx serial-configuration PROM (SPROM). The CCLK speed is selectable as either 1 MHz (default) or 8 MHz. Configuration always starts at the default slow frequency, then can switch to the higher frequency during the first frame. Frequency tolerance is –50% to +25%.

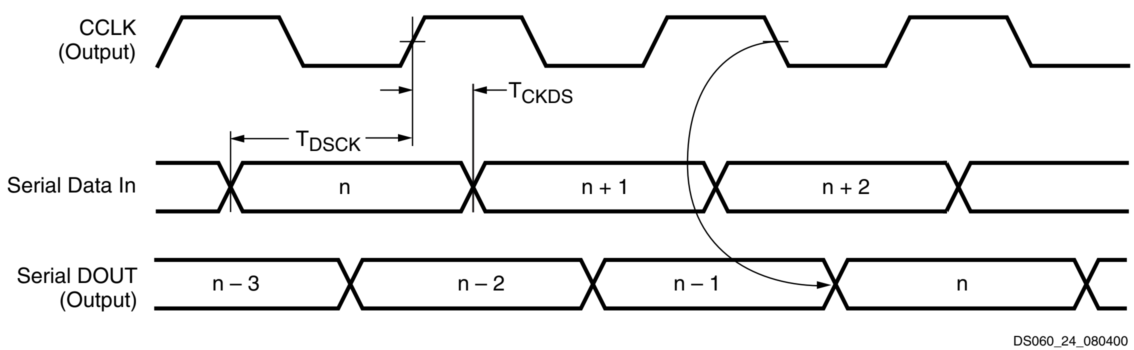
In Master Serial mode, the CCLK output of the device drives a Xilinx SPROM that feeds the FPGA DIN input. Each rising edge of the CCLK output increments the Serial PROM internal address counter. The next data bit is put on the SPROM data output, connected to the FPGA DIN pin. The FPGA accepts this data on the subsequent rising CCLK edge.

When used in a daisy-chain configuration the Master Serial FPGA is placed as the first device in the chain and is referred to as the lead FPGA. The lead FPGA presents the preamble data, and all data that overflows the lead device, on its DOUT pin. There is an internal pipeline delay of 1.5 CCLK periods, which means that DOUT changes on the falling CCLK edge, and the next FPGA in the daisy chain accepts data on the subsequent rising CCLK edge. See the timing diagram in Figure 24.

In the bitstream generation software, the user can specify Fast Configuration Rate, which, starting several bits into the first frame, increases the CCLK frequency by a factor of eight. For actual timing values please refer to the specification section. Be sure that the serial PROM and slaves are fast enough to support this data rate. Earlier families such as the XC3000 series do not support the Fast Configuration Rate option.

The SPROM CE input can be driven from either $\overline{\text{LDC}}$ or DONE. Using $\overline{\text{LDC}}$ avoids potential contention on the DIN pin, if this pin is configured as user I/O, but $\overline{\text{LDC}}$ is then restricted to be a permanently High user output after configuration. Using DONE can also avoid contention on DIN, provided the Early DONE option is invoked.

Figure 25 shows a full master/slave system. The leftmost device is in Master Serial mode, all other devices in the chain are in Slave Serial mode.



	Symbol	Description	Min	Units
CCLK	T_{DSCK}	DIN setup	20	ns
	T_{CKDS}	DIN hold	0	ns

Notes:

1. At power-up, V_{CC} must rise from 2.0V to V_{CC} min in less than 25 ms, otherwise delay configuration by pulling $\overline{\text{PROGRAM}}$ Low until V_{CC} is valid.
2. Master Serial mode timing is based on testing in slave mode.

Figure 24: **Master Serial Mode Programming Switching Characteristics**

Slave Serial Mode

In Slave Serial mode, the FPGA receives serial configuration data on the rising edge of CCLK and, after loading its configuration, passes additional data out, resynchronized on the next falling edge of CCLK.

In this mode, an external signal drives the CCLK input of the FPGA (most often from a Master Serial device). The serial configuration bitstream must be available at the DIN input of the lead FPGA a short setup time before each rising CCLK edge.

The lead FPGA then presents the preamble data—and all data that overflows the lead device—on its DOUT pin. There is an internal delay of 0.5 CCLK periods, which means that DOUT changes on the falling CCLK edge, and the next FPGA in the daisy chain accepts data on the subsequent rising CCLK edge.

Figure 25 shows a full master/slave system. A Spartan/XL device in Slave Serial mode should be connected as shown in the third device from the left.

to the DONE pin. User I/Os for each device become active after the DONE pin for that device goes High. (The exact timing is determined by development system options.) Since the DONE pin is open-drain and does not drive a High value, tying the DONE pins of all devices together prevents all devices in the chain from going High until the last device in the chain has completed its configuration cycle. If the DONE pin of a device is left unconnected, the device becomes active as soon as that device has been configured. Only devices supporting Express mode can be used to form an Express mode daisy chain.

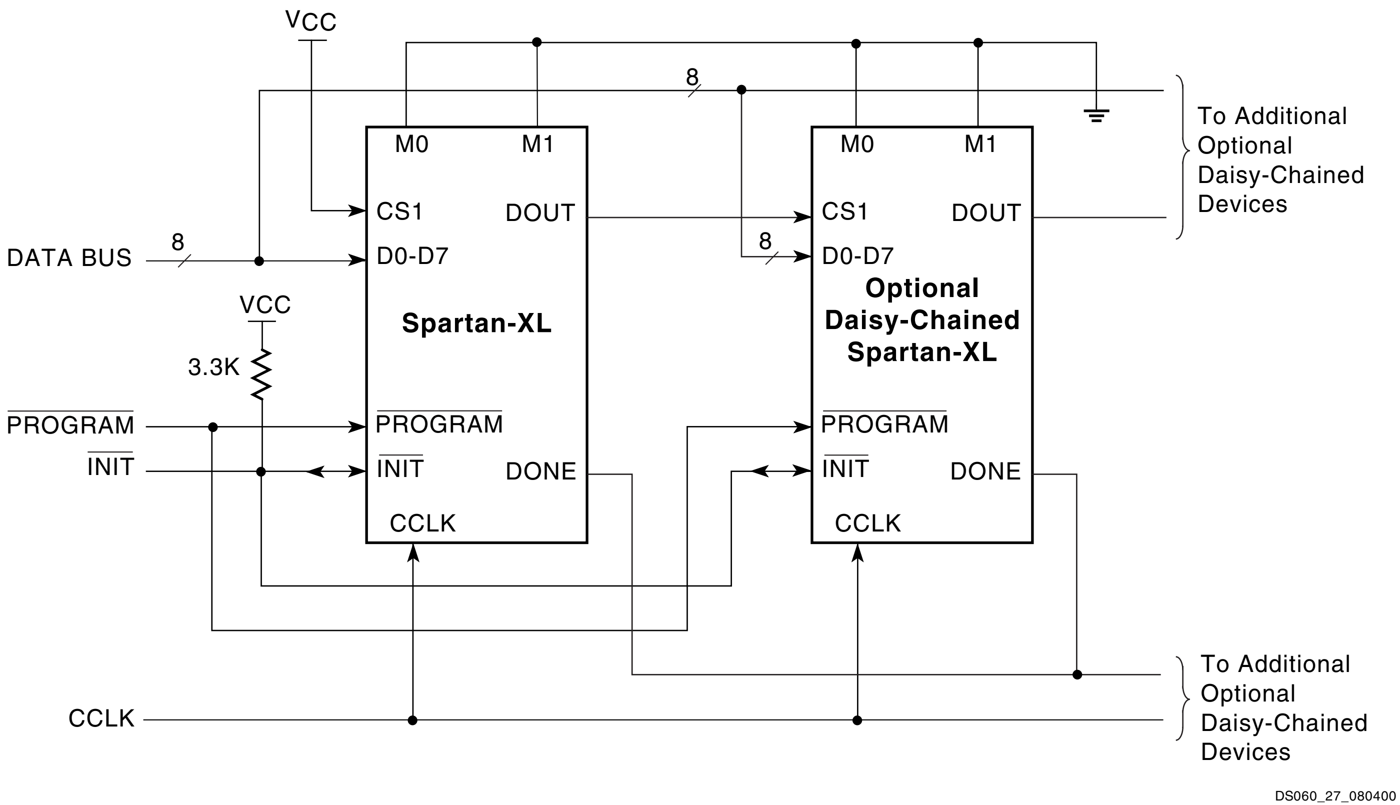


Figure 27: **Express Mode Circuit Diagram**

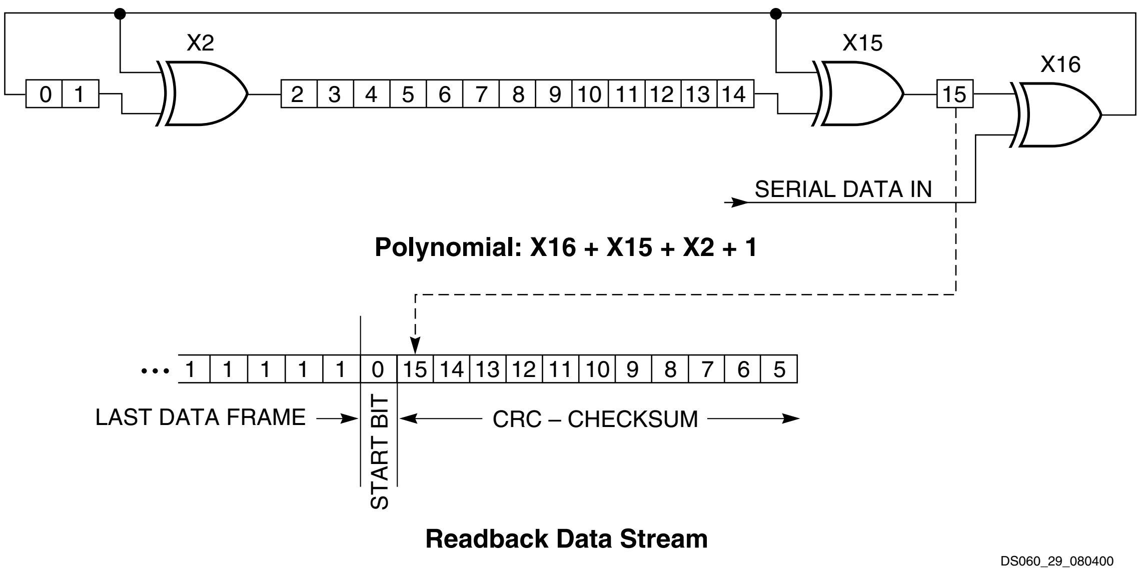


Figure 29: **Circuit for Generating CRC-16**

Configuration Sequence

There are four major steps in the Spartan/XL FPGA power-up configuration sequence.

- Configuration Memory Clear
- Initialization
- Configuration
- Start-up

The full process is illustrated in Figure 30.

Configuration Memory Clear

When power is first applied or is reapplied to an FPGA, an internal circuit forces initialization of the configuration logic. When V_{CC} reaches an operational level, and the circuit passes the write and read test of a sample pair of configuration bits, a time delay is started. This time delay is nominally 16 ms. The delay is four times as long when in Master Serial Mode to allow ample time for all slaves to reach a stable V_{CC}. When all $\overline{\text{INIT}}$ pins are tied together, as recommended, the longest delay takes precedence. Therefore, devices with different time delays can easily be mixed and matched in a daisy chain.

This delay is applied only on power-up. It is not applied when reconfiguring an FPGA by pulsing the $\overline{\text{PROGRAM}}$ pin Low. During this time delay, or as long as the $\overline{\text{PROGRAM}}$ input is asserted, the configuration logic is held in a Configuration Memory Clear state. The configuration-memory frames are consecutively initialized, using the internal oscillator.

At the end of each complete pass through the frame addressing, the power-on time-out delay circuitry and the level of the $\overline{\text{PROGRAM}}$ pin are tested. If neither is asserted, the logic initiates one additional clearing of the configuration frames and then tests the $\overline{\text{INIT}}$ input.

Initialization

During initialization and configuration, user pins HDC, $\overline{\text{LDC}}$, $\overline{\text{INIT}}$ and DONE provide status outputs for the system interface. The outputs $\overline{\text{LDC}}$, $\overline{\text{INIT}}$ and DONE are held Low and HDC is held High starting at the initial application of power.

The open drain $\overline{\text{INIT}}$ pin is released after the final initialization pass through the frame addresses. There is a deliberate delay before a Master-mode device recognizes an inactive $\overline{\text{INIT}}$. Two internal clocks after the $\overline{\text{INIT}}$ pin is recognized as High, the device samples the MODE pin to determine the configuration mode. The appropriate interface lines become active and the configuration preamble and data can be loaded.

to wait after completing the configuration memory clear operation. When INIT is no longer held Low externally, the device determines its configuration mode by capturing the state of the Mode pins, and is ready to start the configuration process. A master device waits up to an additional 300 μs to make sure that any slaves in the optional daisy chain have seen that INIT is High.

For more details on Configuration, refer to the Xilinx Application Note "FPGA Configuration Guidelines" (XAPP090).

Start-Up

Start-up is the transition from the configuration process to the intended user operation. This transition involves a change from one clock source to another, and a change from interfacing parallel or serial configuration data where most outputs are 3-stated, to normal operation with I/O pins active in the user system. Start-up must make sure that the user logic 'wakes up' gracefully, that the outputs become active without causing contention with the configuration signals, and that the internal flip-flops are released from the Global Set/Reset (GSR) at the right time.

Start-Up Initiation

Two conditions have to be met in order for the start-up sequence to begin:

- The chip's internal memory must be full, and
- The configuration length count must be met, exactly.

In all configuration modes except Express mode, Spartan/XL devices read the expected length count from the bitstream and store it in an internal register. The length count varies according to the number of devices and the composition of the daisy chain. Each device also counts the number of CCLKs during configuration.

In Express mode, there is no length count. The start-up sequence for each device begins when the device has received its quota of configuration data. Wiring the DONE pins of several devices together delays start-up of all devices until all are fully configured.

Start-Up Events

The device can be programmed to control three start-up events.

- The release of the open-drain DONE output
- The termination of the Global Three-State and the change of configuration-related pins to the user function, activating all IOBs.
- The termination of the Global Set/Reset initialization of all CLB and IOB storage elements.

Figure 31 describes start-up timing in detail. The three events — DONE going High, the internal GSR being de-activated, and the user I/O going active — can all occur in any arbitrary sequence. This relative timing is selected by options in the bitstream generation software. Heavy lines in Figure 31 show the default timing. The thin lines indicate all other possible timing options. The start-up logic must be clocked until the "F" (Finished) state is reached.

The default option, and the most practical one, is for DONE to go High first, disconnecting the configuration data source and avoiding any contention when the I/Os become active one clock later. GSR is then released another clock period later to make sure that user operation starts from stable internal conditions. This is the most common sequence, shown with heavy lines in Figure 31, but the designer can modify it to meet particular requirements.

Start-Up Clock

Normally, the start-up sequence is controlled by the internal device oscillator (CCLK), which is asynchronous to the system clock. As a configuration option, they can be triggered by an on-chip user net called UCLK. This user net can be accessed by placing the STARTUP library symbol, and the start-up modes are known as UCLK_NOSYNC or UCLK_SYNC. This allows the device to wake up in synchronism with the user system.

DONE Pin

Note that DONE is an open-drain output and does not go High unless an internal pull-up is activated or an external pull-up is attached. The internal pull-up is activated as the default by the bitstream generation software.

The DONE pin can also be wire-ANDed with DONE pins of other FPGAs or with other external signals, and can then be used as input to the start-up control logic. This is called "Start-up Timing Synchronous to Done In" and is selected by either CCLK_SYNC or UCLK_SYNC. When DONE is not used as an input, the operation is called "Start-up Timing Not Synchronous to DONE In," and is selected by either CCLK_NOSYNC or UCLK_NOSYNC. Express mode configuration always uses either CCLK_SYNC or UCLK_SYNC timing, while the other configuration modes can use any of the four timing sequences.

When the UCLK_SYNC option is enabled, the user can externally hold the open-drain DONE output Low, and thus stall all further progress in the start-up sequence until DONE is released and has gone High. This option can be used to force synchronization of several FPGAs to a common user clock, or to guarantee that all devices are successfully configured before any I/Os go active.

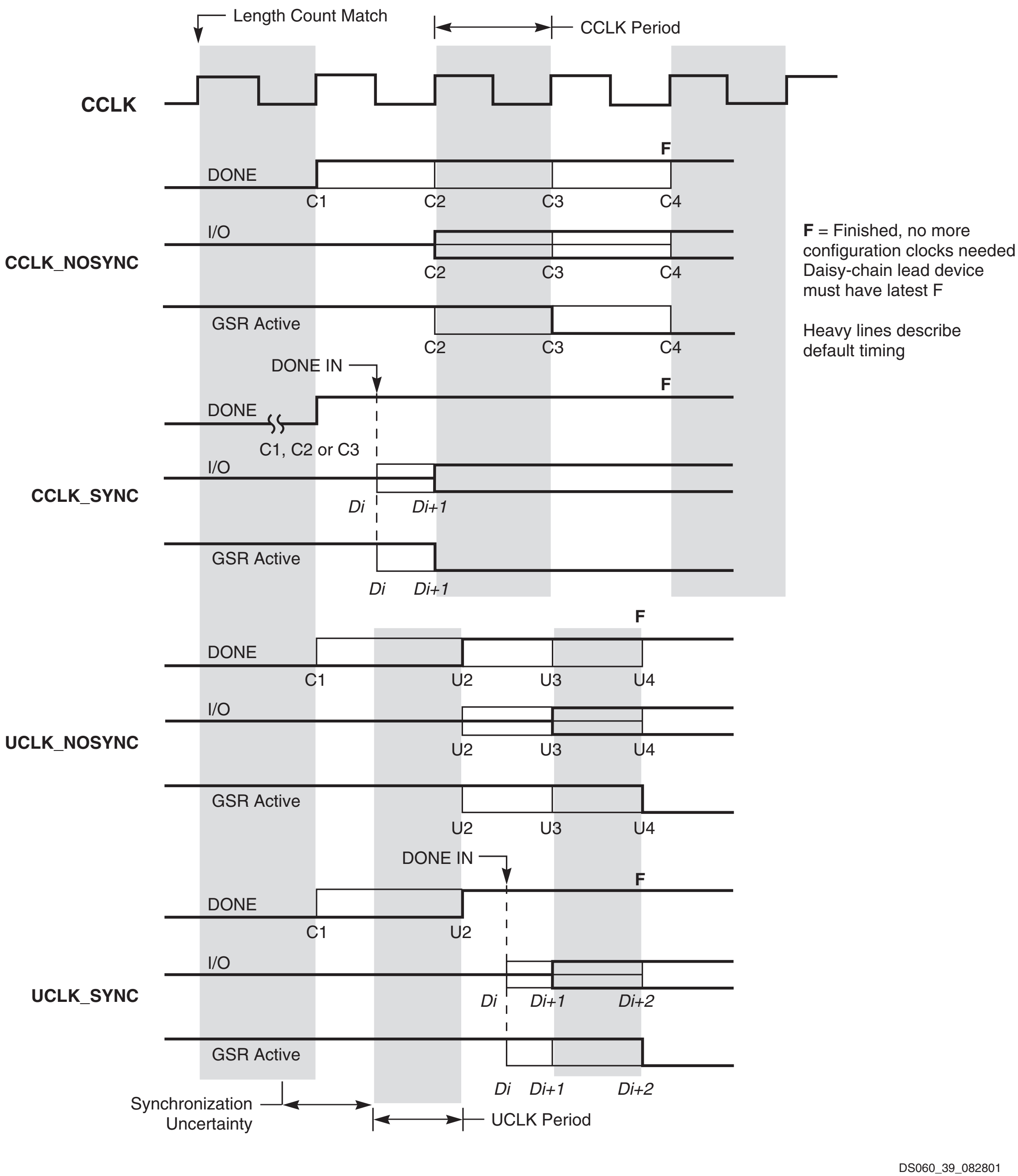


Figure 31: **Start-up Timing**

Configuration Through the Boundary Scan Pins

Spartan/XL devices can be configured through the boundary scan pins. The basic procedure is as follows:

- Power up the FPGA with INIT held Low (or drive the PROGRAM pin Low for more than 300 ns followed by a High while holding INIT Low). Holding INIT Low allows enough time to issue the CONFIG command to the FPGA. The pin can be used as I/O after configuration if a resistor is used to hold INIT Low.
- Issue the CONFIG command to the TMS input.
- Wait for INIT to go High.
- Sequence the boundary scan Test Access Port to the SHIFT-DR state.
- Toggle TCK to clock data into TDI pin.

The user must account for all TCK clock cycles after INIT goes High, as all of these cycles affect the Length Count compare.

For more detailed information, refer to the Xilinx application note, "*Boundary Scan in FPGA Devices*." This application note applies to Spartan and Spartan-XL devices.

Spartan Family IOB Output Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. These path delays, provided as a guideline, have been extracted from the static timing analyzer report. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values are expressed in nanoseconds unless otherwise noted.

Symbol	Description	Device	Speed Grade				Units
			-4		-3		
			Min	Max	Min	Max	
Clocks							
T_{CH}	Clock High	All devices	3.0	-	4.0	-	ns
T_{CL}	Clock Low	All devices	3.0	-	4.0	-	ns
Propagation Delays - TTL Outputs(1,2)							
T_{OKPOF}	Clock (OK) to Pad, fast	All devices	-	3.3	-	4.5	ns
T_{OKPOS}	Clock (OK to Pad, slew-rate limited	All devices	-	6.9	-	7.0	ns
T_{OPF}	Output (O) to Pad, fast	All devices	-	3.6	-	4.8	ns
T_{OPS}	Output (O) to Pad, slew-rate limited	All devices	-	7.2	-	7.3	ns
T_{TSHZ}	3-state to Pad High-Z (slew-rate independent)	All devices	-	3.0	-	3.8	ns
T_{TSONF}	3-state to Pad active and valid, fast	All devices	-	6.0	-	7.3	ns
T_{TSONS}	3-state to Pad active and valid, slew-rate limited	All devices	-	9.6	-	9.8	ns
Setup and Hold Times							
T_{OOK}	Output (O) to clock (OK) setup time	All devices	2.5	-	3.8	-	ns
T_{OKO}	Output (O) to clock (OK) hold time	All devices	0.0	-	0.0	-	ns
T_{ECOK}	Clock Enable (EC) to clock (OK) setup time	All devices	2.0	-	2.7	-	ns
T_{OKEC}	Clock Enable (EC) to clock (OK) hold time	All devices	0.0	-	0.5	-	ns
Global Set/Reset							
T_{MRW}	Minimum GSR pulse width	All devices	11.5		13.5		ns
T_{RPO}	Delay from GSR input to any Pad	XCS05	-	12.0	-	15.0	ns
		XCS10	-	12.5	-	15.7	ns
		XCS20	-	13.0	-	16.2	ns
		XCS30	-	13.5	-	16.9	ns
		XCS40	-	14.0	-	17.5	ns

Notes:
1. Delay adder for CMOS Outputs option (with fast slew rate option): for -3 speed grade, add 1.0 ns; for -4 speed grade, add 0.8 ns.
2. Delay adder for CMOS Outputs option (with slow slew rate option): for -3 speed grade, add 2.0 ns; for -4 speed grade, add 1.5 ns.
3. Output timing is measured at ~50% V_{CC} threshold, with 50 pF external capacitive loads including test fixture. Slew-rate limited output rise/fall times are approximately two times longer than fast output rise/fall times.
4. Voltage levels of unused pads, bonded or unbonded, must be valid logic levels. Each can be configured with the internal pull-up (default) or pull-down resistor, or configured as a driven output, or can be driven from an external source.

Spartan-XL Family Global Buffer Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.

When fewer vertical clock lines are connected, the clock distribution is faster; when multiple clock lines per column are driven from the same global clock, the delay is longer. For more specific, more precise, and worst-case guaranteed data, reflecting the actual routing structure, use the values provided by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. These path delays, provided as a guideline, have been extracted from the static timing analyzer report. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature).

Symbol	Description	Device	Speed Grade		Units
			-5	-4	
			Max	Max	
T_{GLS}	From pad through buffer, to any clock K	XCS05XL	1.4	1.5	ns
		XCS10XL	1.7	1.8	ns
		XCS20XL	2.0	2.1	ns
		XCS30XL	2.3	2.5	ns
		XCS40XL	2.6	2.8	ns

Table 18: **Pin Descriptions *(Continued)***

Pin Name	I/O During Config.	I/O After Config.	Pin Description
PWRDWN	I	I	PWRDWN is an active Low input that forces the FPGA into the Power Down state and reduces power consumption. When PWRDWN is Low, the FPGA disables all I/O and initializes all flip-flops. All inputs are interpreted as Low independent of their actual level. VCC must be maintained, and the configuration data is maintained. PWRDWN halts configuration if asserted before or during configuration, and re-starts configuration when removed. When PWRDWN returns High, the FPGA becomes operational by first enabling the inputs and flip-flops and then enabling the outputs. PWRDWN has a default internal pull-up resistor.
User I/O Pins That Can Have Special Functions			
TDO	O	O	If boundary scan is used, this pin is the Test Data Output. If boundary scan is not used, this pin is a 3-state output without a register, after configuration is completed. To use this pin, place the library component TDO instead of the usual pad symbol. An output buffer must still be used.
TDI, TCK, TMS	I	I/O or I (JTAG)	If boundary scan is used, these pins are Test Data In, Test Clock, and Test Mode Select inputs respectively. They come directly from the pads, bypassing the IOBs. These pins can also be used as inputs to the CLB logic after configuration is completed. If the BSCAN symbol is not placed in the design, all boundary scan functions are inhibited once configuration is completed, and these pins become user-programmable I/O. In this case, they must be called out by special library elements. To use these pins, place the library components TDI, TCK, and TMS instead of the usual pad symbols. Input or output buffers must still be used.
HDC	O	I/O	High During Configuration (HDC) is driven High until the I/O go active. It is available as a control output indicating that configuration is not yet completed. After configuration, HDC is a user-programmable I/O pin.
LDC	O	I/O	Low During Configuration (LDC) is driven Low until the I/O go active. It is available as a control output indicating that configuration is not yet completed. After configuration, LDC is a user-programmable I/O pin.
INIT	I/O	I/O	Before and during configuration, INIT is a bidirectional signal. A 1 kΩ to 10 kΩ external pull-up resistor is recommended. As an active Low open-drain output, INIT is held Low during the power stabilization and internal clearing of the configuration memory. As an active Low input, it can be used to hold the FPGA in the internal WAIT state before the start of configuration. Master mode devices stay in a WAIT state an additional 30 to 300 μs after INIT has gone High. During configuration, a Low on this output indicates that a configuration data error has occurred. After the I/O go active, INIT is a user-programmable I/O pin.
PGCK1 - PGCK4 (Spartan)	Weak Pull-up	I or I/O	Four Primary Global inputs each drive a dedicated internal global net with short delay and minimal skew. If not used to drive a global buffer, any of these pins is a user-programmable I/O. The PGCK1-PGCK4 pins drive the four Primary Global Buffers. Any input pad symbol connected directly to the input of a BUFGP symbol is automatically placed on one of these pins.

XCS10 and XCS10XL Device Pinouts

XCS10/XL Pad Name	PC84[4]	VQ100	CS144[2,4]	TQ144	Bndry Scan
VCC	P33	P25	N1	P37	-
Not Connect-ed[1] $\overline{\text{PWRDWN}}$[2]	P34	P26	N2	P38	174 [1]
I/O, PGCK2[1] GCK3[2]	P35	P27	M3	P39	175[3]
I/O (HDC)	P36	P28	N3	P40	178 [3]
I/O	-	-	K4	P41	181 [3]
I/O	-	-	L4	P42	184 [3]
I/O	-	P29	M4	P43	187 [3]
I/O ($\overline{\text{LDC}}$)	P37	P30	N4	P44	190 [3]
GND	-	-	K5	P45	-
I/O	-	-	L5	P46	193 [3]
I/O	-	-	M5	P47	196 [3]
I/O	P38	P31	N5	P48	199 [3]
I/O	P39	P32	K6	P49	202 [3]
I/O	-	P33	L6	P50	205 [3]
I/O	-	P34	M6	P51	208 [3]
I/O	P40	P35	N6	P52	211 [3]
I/O ($\overline{\text{INIT}}$)	P41	P36	M7	P53	214 [3]
VCC	P42	P37	N7	P54	-
GND	P43	P38	L7	P55	-
I/O	P44	P39	K7	P56	217 [3]
I/O	P45	P40	N8	P57	220 [3]
I/O	-	P41	M8	P58	223 [3]
I/O	-	P42	L8	P59	226 [3]
I/O	P46	P43	K8	P60	229 [3]
I/O	P47	P44	N9	P61	232 [3]
I/O	-	-	M9	P62	235 [3]
I/O	-	-	L9	P63	238 [3]
GND	-	-	K9	P64	-
I/O	P48	P45	N10	P65	241 [3]
I/O	P49	P46	M10	P66	244 [3]
I/O	-	-	L10	P67	247 [3]
I/O	-	-	N11	P68	250 [3]
I/O	P50	P47	M11	P69	253 [3]
I/O, SGCK3[1] GCK4[2]	P51	P48	L11	P70	256 [3]
GND	P52	P49	N12	P71	-
DONE	P53	P50	M12	P72	-
VCC	P54	P51	N13	P73	-
$\overline{\text{PROGRAM}}$	P55	P52	M13	P74	-
I/O (D7[2])	P56	P53	L12	P75	259 [3]

XCS10 and XCS10XL Device Pinouts

XCS10/XL Pad Name	PC84[4]	VQ100	CS144[2,4]	TQ144	Bndry Scan
I/O, PGCK3[1] GCK5[2]	P57	P54	L13	P76	262 [3]
I/O	-	-	K10	P77	265 [3]
I/O	-	-	K11	P78	268 [3]
I/O (D6[2])	P58	P55	K12	P79	271 [3]
I/O	-	P56	K13	P80	274 [3]
GND	-	-	J10	P81	-
I/O	-	-	J11	P82	277 [3]
I/O	-	-	J12	P83	280 [3]
I/O (D5[2])	P59	P57	J13	P84	283 [3]
I/O	P60	P58	H10	P85	286 [3]
I/O	-	P59	H11	P86	289 [3]
I/O	-	P60	H12	P87	292 [3]
I/O (D4[2])	P61	P61	H13	P88	295 [3]
I/O	P62	P62	G12	P89	298 [3]
VCC	P63	P63	G13	P90	-
GND	P64	P64	G11	P91	-
I/O (D3[2])	P65	P65	G10	P92	301 [3]
I/O	P66	P66	F13	P93	304 [3]
I/O	-	P67	F12	P94	307 [3]
I/O	-	-	F11	P95	310 [3]
I/O (D2[2])	P67	P68	F10	P96	313 [3]
I/O	P68	P69	E13	P97	316 [3]
I/O	-	-	E12	P98	319 [3]
I/O	-	-	E11	P99	322 [3]
GND	-	-	E10	P100	-
I/O (D1[2])	P69	P70	D13	P101	325 [3]
I/O	P70	P71	D12	P102	328 [3]
I/O	-	-	D11	P103	331 [3]
I/O	-	-	C13	P104	334 [3]
I/O (D0[2], DIN)	P71	P72	C12	P105	337 [3]
I/O, SGCK4[1] GCK6[2] (DOUT)	P72	P73	C11	P106	340 [3]
CCLK	P73	P74	B13	P107	-
VCC	P74	P75	B12	P108	-
O, TDO	P75	P76	A13	P109	0
GND	P76	P77	A12	P110	-
I/O	P77	P78	B11	P111	2
I/O, PGCK4[1] GCK7[2]	P78	P79	A11	P112	5
I/O	-	-	D10	P113	8
I/O	-	-	C10	P114	11
I/O (CS1[2])	P79	P80	B10	P115	14

XCS30 and XCS30XL Device Pinouts *(Continued)*

XCS30/XL Pad Name	VQ100(5)	TQ144	PQ208	PQ240	BG256(5)	CS280(2,5)	Bndry Scan
I/O	-	-	P85	P97	U12	T11	382 (3)
I/O	-	-	-	P99	V13	U12	385 (3)
I/O	-	-	-	P100	Y14	T12	388 (3)
VCC	-	-	P86	P101	VCC(4)	W13	-
I/O	P43	P60	P87	P102	Y15	V13	391 (3)
I/O	P44	P61	P88	P103	V14	U13	394 (3)
I/O	-	P62	P89	P104	W15	T13	397 (3)
I/O	-	P63	P90	P105	Y16	W14	400 (3)
GND	-	P64	P91	P106	GND(4)	GND(4)	-
I/O	-	-	-	P107	V15	V14	403 (3)
I/O	-	-	P92	P108	W16	U14	406 (3)
I/O	-	-	P93	P109	Y17	T14	409 (3)
I/O	-	-	P94	P110	V16	R14	412 (3)
I/O	-	-	P95	P111	W17	W15	415 (3)
I/O	-	-	P96	P112	Y18	U15	418 (3)
I/O	P45	P65	P97	P113	U16	V16	421 (3)
I/O	P46	P66	P98	P114	V17	U16	424 (3)
I/O	-	P67	P99	P115	W18	W17	427 (3)
I/O	-	P68	P100	P116	Y19	W18	430 (3)
I/O	P47	P69	P101	P117	V18	V17	433 (3)
I/O, SGCK3(1), GCK4(2)	P48	P70	P102	P118	W19	V18	436 (3)
GND	P49	P71	P103	P119	GND(4)	GND(4)	-
DONE	P50	P72	P104	P120	Y20	W19	-
VCC	P51	P73	P105	P121	VCC(4)	U17	-
$\overline{\text{PROGRAM}}$	P52	P74	P106	P122	V19	U18	-
I/O (D7(2))	P53	P75	P107	P123	U19	V19	439 (3)
I/O, PGCK3(1), GCK5(2)	P54	P76	P108	P124	U18	U19	442 (3)
I/O	-	P77	P109	P125	T17	T16	445 (3)
I/O	-	P78	P110	P126	V20	T17	448 (3)
I/O	-	-	-	P127	U20	T18	451 (3)
I/O	-	-	P111	P128	T18	T19	454 (3)
I/O (D6(2))	P55	P79	P112	P129	T19	R16	457 (3)
I/O	P56	P80	P113	P130	T20	R19	460 (3)
I/O	-	-	P114	P131	R18	P15	463 (3)
I/O	-	-	P115	P132	R19	P17	466 (3)
I/O	-	-	P116	P133	R20	P18	469 (3)
I/O	-	-	P117	P134	P18	P16	472 (3)
GND	-	P81	P118	P135	GND(4)	GND(4)	-
I/O	-	-	-	P136	P20	P19	475 (3)
I/O	-	-	-	P137	N18	N17	478 (3)
I/O	-	P82	P119	P138	N19	N18	481 (3)
I/O	-	P83	P120	P139	N20	N19	484 (3)
VCC	-	-	P121	P140	VCC(4)	N16	-
I/O (D5(2))	P57	P84	P122	P141	M17	M19	487 (3)
I/O	P58	P85	P123	P142	M18	M17	490 (3)

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
I/O	P90	P105	Y16	W14	466 (3)
GND	P91	P106	GND(4)	GND(4)	-
I/O	-	P107	V15	V14	469 (3)
I/O	P92	P108	W16	U14	472 (3)
I/O	P93	P109	Y17	T14	475 (3)
I/O	P94	P110	V16	R14	478 (3)
I/O	P95	P111	W17	W15	481 (3)
I/O	P96	P112	Y18	U15	484 (3)
I/O	-	-	-	T15	487 (3)
I/O	-	-	-	W16	490 (3)
I/O	P97	P113	U16	V16	493 (3)
I/O	P98	P114	V17	U16	496 (3)
I/O	P99	P115	W18	W17	499 (3)
I/O	P100	P116	Y19	W18	502 (3)
I/O	P101	P117	V18	V17	505 (3)
I/O, SGCK3(1), GCK4(2)	P102	P118	W19	V18	508 (3)
GND	P103	P119	GND(4)	GND(4)	-
DONE	P104	P120	Y20	W19	-
VCC	P105	P121	VCC(4)	VCC(4)	-
$\overline{\text{PROGRAM}}$	P106	P122	V19	U18	-
I/O (D7(2))	P107	P123	U19	V19	511 (3)
I/O, PGCK3(1), GCK5(2)	P108	P124	U18	U19	514 (3)
I/O	P109	P125	T17	T16	517 (3)
I/O	P110	P126	V20	T17	520 (3)
I/O	-	P127	U20	T18	523 (3)
I/O	P111	P128	T18	T19	526 (3)
I/O	-	-	-	R15	529 (3)
I/O	-	-	-	R17	523 (3)
I/O (D6(2))	P112	P129	T19	R16	535 (3)
I/O	P113	P130	T20	R19	538 (3)
I/O	P114	P131	R18	P15	541 (3)
I/O	P115	P132	R19	P17	544 (3)
I/O	P116	P133	R20	P18	547 (3)
I/O	P117	P134	P18	P16	550 (3)
GND	P118	P135	GND(4)	GND(4)	-
I/O	-	P136	P20	P19	553 (3)
I/O	-	P137	N18	N17	556 (3)
I/O	P119	P138	N19	N18	559 (3)
I/O	P120	P139	N20	N19	562 (3)
VCC	P121	P140	VCC(4)	VCC(4)	-
I/O (D5(2))	P122	P141	M17	M19	565 (3)
I/O	P123	P142	M18	M17	568 (3)

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
I/O	-	-	-	M18	571 (3)
I/O	-	-	M19	M16	574 (3)
I/O	P124	P144	M20	L19	577 (3)
I/O	P125	P145	L19	L18	580 (3)
I/O	P126	P146	L18	L17	583 (3)
I/O	P127	P147	L20	L16	586 (3)
I/O (D4(2))	P128	P148	K20	K19	589 (3)
I/O	P129	P149	K19	K18	592 (3)
VCC	P130	P150	VCC(4)	VCC(4)	-
GND	P131	P151	GND(4)	GND(4)	-
I/O (D3(2))	P132	P152	K18	K16	595 (3)
I/O	P133	P153	K17	K15	598 (3)
I/O	P134	P154	J20	J19	601 (3)
I/O	P135	P155	J19	J18	604 (3)
I/O	P136	P156	J18	J17	607 (3)
I/O	P137	P157	J17	J16	610 (3)
I/O	-	-	H20	H19	613 (3)
I/O	-	-	-	H18	616 (3)
I/O (D2(2))	P138	P159	H19	H17	619 (3)
I/O	P139	P160	H18	H16	622 (3)
VCC	P140	P161	VCC(4)	VCC(4)	-
I/O	P141	P162	G19	G18	625 (3)
I/O	P142	P163	F20	G17	628 (3)
I/O	-	P164	G18	G16	631 (3)
I/O	-	P165	F19	F19	634 (3)
GND	P143	P166	GND(4)	GND(4)	-
I/O	-	P167	F18	F18	637 (3)
I/O	P144	P168	E19	F17	640 (3)
I/O	P145	P169	D20	F16	643 (3)
I/O	P146	P170	E18	F15	646 (3)
I/O	P147	P171	D19	E19	649 (3)
I/O	P148	P172	C20	E17	652 (3)
I/O (D1(2))	P149	P173	E17	E16	655 (3)
I/O	P150	P174	D18	D19	658 (3)
I/O	-	-	-	D18	661 (3)
I/O	-	-	-	D17	664 (3)
I/O	P151	P175	C19	C19	667 (3)
I/O	P152	P176	B20	B19	670 (3)
I/O (D0(2), DIN)	P153	P177	C18	C18	673 (3)
I/O, SGCK4(1), GCK6(2) (DOUT)	P154	P178	B19	B18	676 (3)
CCLK	P155	P179	A20	A19	-
VCC	P156	P180	VCC(4)	VCC(4)	-