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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	400
Number of Logic Elements/Cells	950
Total RAM Bits	12800
Number of I/O	160
Number of Gates	20000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcs20xl-4pq208i

General Overview

Spartan series FPGAs are implemented with a regular, flexible, programmable architecture of Configurable Logic Blocks (CLBs), interconnected by a powerful hierarchy of versatile routing resources (routing channels), and surrounded by a perimeter of programmable Input/Output Blocks (IOBs), as seen in Figure 1. They have generous routing resources to accommodate the most complex interconnect patterns.

The devices are customized by loading configuration data into internal static memory cells. Re-programming is possible an unlimited number of times. The values stored in these memory cells determine the logic functions and interconnections implemented in the FPGA. The FPGA can either actively read its configuration data from an external serial PROM (Master Serial mode), or the configuration data can be written into the FPGA from an external device (Slave Serial mode).

Spartan series FPGAs can be used where hardware must be adapted to different user applications. FPGAs are ideal for shortening design and development cycles, and also offer a cost-effective solution for production rates well beyond 50,000 systems per month.

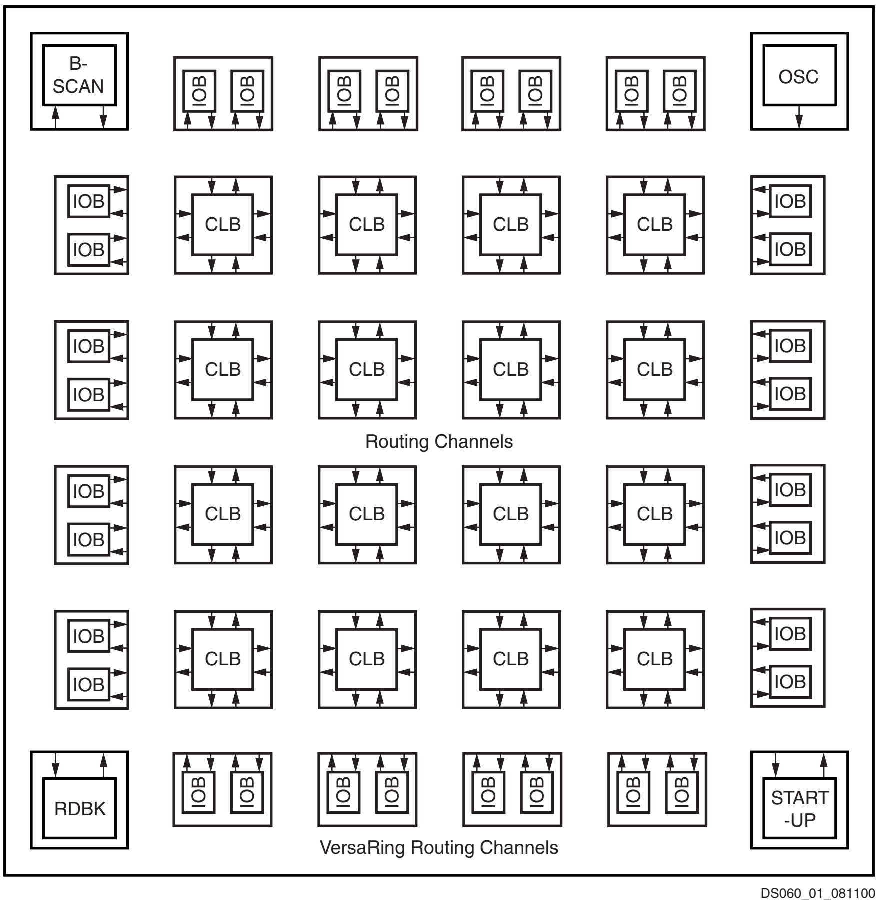


Figure 1: **Basic FPGA Block Diagram**

This high value makes them unsuitable as wired-AND pull-up resistors.

Table 7: **Supported Destinations for Spartan/XL Outputs**

Destination	Spartan-XL Outputs	Spartan Outputs	
	3.3V, CMOS	5V, TTL	5V, CMOS
Any device, V_{CC} = 3.3V, CMOS-threshold inputs	√	√	Some(1)
Any device, V_{CC} = 5V, TTL-threshold inputs	√	√	√
Any device, V_{CC} = 5V, CMOS-threshold inputs	Unreliable Data		√

Notes:
1. Only if destination device has 5V tolerant inputs.

After configuration, voltage levels of unused pads, bonded or unbonded, must be valid logic levels, to reduce noise sensitivity and avoid excess current. Therefore, by default, unused pads are configured with the internal pull-up resistor active. Alternatively, they can be individually configured with the pull-down resistor, or as a driven output, or to be driven by an external source. To activate the internal pull-up, attach the PULLUP library component to the net attached to the pad. To activate the internal pull-down, attach the PULL-DOWN library component to the net attached to the pad.

Set/Reset

As with the CLB registers, the GSR signal can be used to set or clear the input and output registers, depending on the value of the INIT attribute or property. The two flip-flops can be individually configured to set or clear on reset and after configuration. Other than the global GSR net, no user-controlled set/reset signal is available to the I/O flip-flops (Figure 5). The choice of set or reset applies to both the initial state of the flip-flop and the response to the GSR pulse.

Independent Clocks

Separate clock signals are provided for the input (IK) and output (OK) flip-flops. The clock can be independently inverted for each flip-flop within the IOB, generating either falling-edge or rising-edge triggered flip-flops. The clock inputs for each IOB are independent.

Common Clock Enables

The input and output flip-flops in each IOB have a common clock enable input (see EC signal in Figure 5), which through configuration, can be activated individually for the input or output flip-flop, or both. This clock enable operates exactly like the EC signal on the Spartan/XL FPGA CLB. It cannot be inverted within the IOB.

Routing Channel Description

All internal routing channels are composed of metal segments with programmable switching points and switching matrices to implement the desired routing. A structured, hierarchical matrix of routing channels is provided to achieve efficient automated routing.

This section describes the routing channels available in Spartan/XL devices. Figure 8 shows a general block diagram of the CLB routing channels. The implementation software automatically assigns the appropriate resources based on the density and timing requirements of the design. The following description of the routing channels is for information only and is simplified with some minor details omitted. For an exact interconnect description the designer should open a design in the FPGA Editor and review the actual connections in this tool.

The routing channels will be discussed as follows;

- CLB routing channels which run along each row and column of the CLB array.
- IOB routing channels which form a ring (called a VersaRing) around the outside of the CLB array. It connects the I/O with the CLB routing channels.
- Global routing consists of dedicated networks primarily designed to distribute clocks throughout the device with minimum delay and skew. Global routing can also be used for other high-fanout signals.

CLB Routing Channels

The routing channels around the CLB are derived from three types of interconnects; single-length, double-length, and longlines. At the intersection of each vertical and horizontal routing channel is a signal steering matrix called a Programmable Switch Matrix (PSM). Figure 8 shows the basic routing channel configuration showing single-length lines, double-length lines and longlines as well as the CLBs and PSMs. The CLB to routing channel interface is shown as well as how the PSMs interface at the channel intersections.

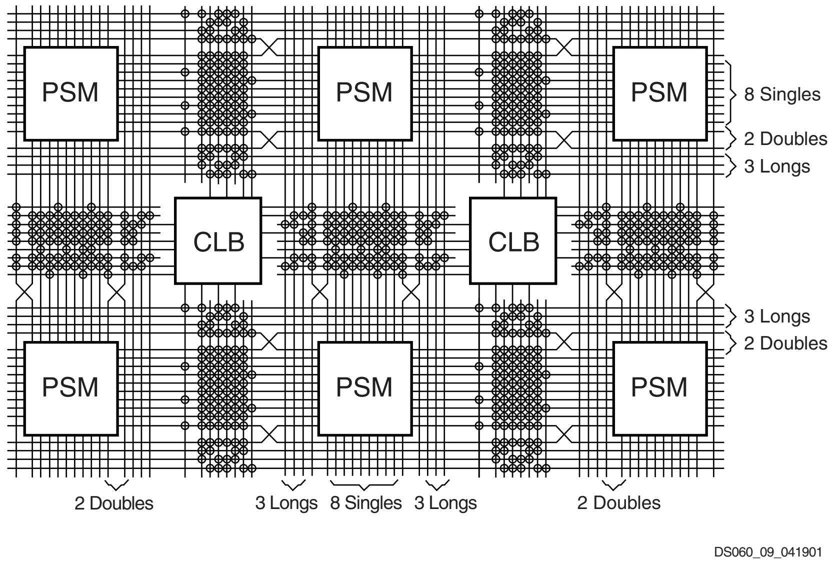


Figure 8: **Spartan/XL CLB Routing Channels and Interface Block Diagram**

CLB Interface

A block diagram of the CLB interface signals is shown in Figure 9. The input signals to the CLB are distributed evenly on all four sides providing maximum routing flexibility. In general, the entire architecture is symmetrical and regular. It is well suited to established placement and routing algorithms. Inputs, outputs, and function generators can freely swap positions within a CLB to avoid routing congestion during the placement and routing operation. The exceptions are the clock (K) input and CIN/COUT signals. The K input is routed to dedicated global vertical lines as well as four single-length lines and is on the left side of the CLB. The CIN/COUT signals are routed through dedicated interconnects which do not interfere with the general routing structure. The output signals from the CLB are available to drive both vertical and horizontal channels.

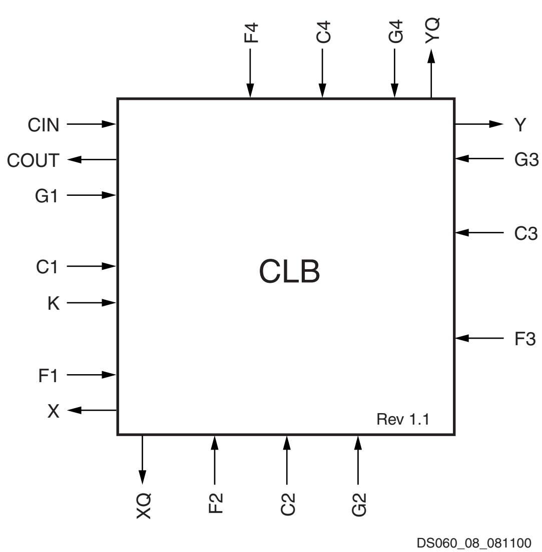


Figure 9: **CLB Interconnect Signals**

Programmable Switch Matrices

The horizontal and vertical single- and double-length lines intersect at a box called a programmable switch matrix (PSM). Each PSM consists of programmable pass transistors used to establish connections between the lines (see Figure 10).

For example, a single-length signal entering on the right side of the switch matrix can be routed to a single-length line on the top, left, or bottom sides, or any combination thereof, if multiple branches are required. Similarly, a double-length signal can be routed to a double-length line on any or all of the other three edges of the programmable switch matrix.

Single-Length Lines

Single-length lines provide the greatest interconnect flexibility and offer fast routing between adjacent blocks. There are eight vertical and eight horizontal single-length lines associated with each CLB. These lines connect the switching matrices that are located in every row and column of CLBs. Single-length lines are connected by way of the programmable switch matrices, as shown in Figure 10. Routing connectivity is shown in Figure 8.

Single-length lines incur a delay whenever they go through a PSM. Therefore, they are not suitable for routing signals for long distances. They are normally used to conduct signals within a localized area and to provide the branching for nets with fanout greater than one.

CLB signals from which they are originally derived are shown in Table 10.

Table 10: **Dual-Port RAM Signals**

RAM Signal	Function	CLB Signal
D	Data In	DIN
A[3:0]	Read Address for Single-Port. Write Address for Single-Port and Dual-Port.	F[4:1]
DPRA[3:0]	Read Address for Dual-Port	G[4:1]
WE	Write Enable	SR
WCLK	Clock	K
SPO	Single Port Out (addressed by A[3:0])	F_{OUT}
DPO	Dual Port Out (addressed by DPRA[3:0])	G_{OUT}

The RAM16X1D primitive used to instantiate the dual-port RAM consists of an upper and a lower 16 x 1 memory array. The address port labeled A[3:0] supplies both the read and write addresses for the lower memory array, which behaves the same as the 16 x 1 single-port RAM array described previously. Single Port Out (SPO) serves as the data output for the lower memory. Therefore, SPO reflects the data at address A[3:0].

The other address port, labeled DPRA[3:0] for Dual Port Read Address, supplies the read address for the upper memory. The write address for this memory, however, comes from the address A[3:0]. Dual Port Out (DPO) serves as the data output for the upper memory. Therefore, DPO reflects the data at address DPRA[3:0].

By using A[3:0] for the write address and DPRA[3:0] for the read address, and reading only the DPO output, a FIFO that can read and write simultaneously is easily generated. The simultaneous read/write capability possible with the dual-port RAM can provide twice the effective data throughput of a single-port RAM alternating read and write operations.

The timing relationships for the dual-port RAM mode are shown in Figure 13.

Note that write operations to RAM are synchronous (edge-triggered); however, data access is asynchronous.

Initializing RAM at FPGA Configuration

Both RAM and ROM implementations in the Spartan/XL families are initialized during device configuration. The initial contents are defined via an INIT attribute or property attached to the RAM or ROM symbol, as described in the library guide. If not defined, all RAM contents are initialized to zeros, by default.

RAM initialization occurs only during device configuration. The RAM content is not affected by GSR.

More Information on Using RAM Inside CLBs

Three application notes are available from Xilinx that discuss synchronous (edge-triggered) RAM: *"Xilinx Edge-Triggered and Dual-Port RAM Capability," "Implementing FIFOs in Xilinx RAM,"* and *"Synchronous and Asynchronous FIFO Designs."* All three application notes apply to both the Spartan and the Spartan-XL families.

Fast Carry Logic

Each CLB F-LUT and G-LUT contains dedicated arithmetic logic for the fast generation of carry and borrow signals. This extra output is passed on to the function generator in the adjacent CLB. The carry chain is independent of normal routing resources. (See Figure 15.)

Dedicated fast carry logic greatly increases the efficiency and performance of adders, subtractors, accumulators, comparators and counters. It also opens the door to many new applications involving arithmetic operation, where the previous generations of FPGAs were not fast enough or too inefficient. High-speed address offset calculations in microprocessor or graphics systems, and high-speed addition in digital signal processing are two typical applications.

The two 4-input function generators can be configured as a 2-bit adder with built-in hidden carry that can be expanded to any length. This dedicated carry circuitry is so fast and efficient that conventional speed-up methods like carry generate/propagate are meaningless even at the 16-bit level, and of marginal benefit at the 32-bit level. This fast carry logic is one of the more significant features of the Spartan

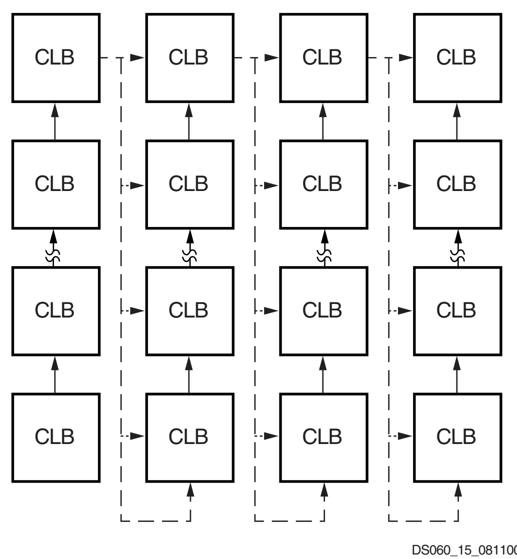


Figure 15: **Available Spartan/XL Carry Propagation Paths**

On-Chip Oscillator

Spartan/XL devices include an internal oscillator. This oscillator is used to clock the power-on time-out, for configuration memory clearing, and as the source of CCLK in Master configuration mode. The oscillator runs at a nominal 8 MHz frequency that varies with process, V_{CC}, and temperature. The output frequency falls between 4 MHz and 10 MHz.

The oscillator output is optionally available after configuration. Any two of four resynchronized taps of a built-in divider are also available. These taps are at the fourth, ninth, fourteenth and nineteenth bits of the divider. Therefore, if the primary oscillator output is running at the nominal 8 MHz, the user has access to an 8-MHz clock, plus any two of 500 kHz, 16 kHz, 490 Hz and 15 Hz. These frequencies can vary by as much as -50% or +25%.

These signals can be accessed by placing the OSC4 library element in a schematic or in HDL code. The oscillator is automatically disabled after configuration if the OSC4 symbol is not used in the design.

Global Signals: GSR and GTS

Global Set/Reset

A separate Global Set/Reset line, as shown in Figure 3, page 5 for the CLB and Figure 5, page 6 for the IOB, sets or clears each flip-flop during power-up, reconfiguration, or when a dedicated Reset net is driven active. This global net (GSR) does not compete with other routing resources; it uses a dedicated distribution network.

Each flip-flop is configured as either globally set or reset in the same way that the local set/reset (SR) is specified. Therefore, if a flip-flop is set by SR, it is also set by GSR. Similarly, if in reset mode, it is reset by both SR and GSR.

GSR can be driven from any user-programmable pin as a global reset input. To use this global net, place an input pad and input buffer in the schematic or HDL code, driving the GSR pin of the STARTUP symbol. (See Figure 19.) A specific pin location can be assigned to this input using a LOC attribute or property, just as with any other user-programmable pad. An inverter can optionally be inserted after the input buffer to invert the sense of the GSR signal. Alternatively, GSR can be driven from any internal node.

Global 3-State

A separate Global 3-state line (GTS) as shown in Figure 6, page 7 forces all FPGA outputs to the high-impedance state, unless boundary scan is enabled and is executing an EXTEST instruction. GTS does not compete with other routing resources; it uses a dedicated distribution network.

GTS can be driven from any user-programmable pin as a global 3-state input. To use this global net, place an input pad and input buffer in the schematic or HDL code, driving the GTS pin of the STARTUP symbol. This is similar to what is shown in Figure 19 for GSR except the IBUF would be connected to GTS. A specific pin location can be assigned to this input using a LOC attribute or property, just as with any other user-programmable pad. An inverter can optionally be inserted after the input buffer to invert the sense of the Global 3-state signal. Alternatively, GTS can be driven from any internal node.

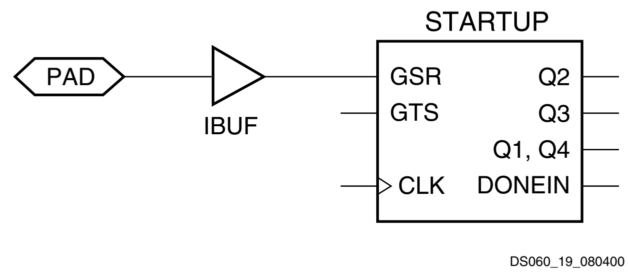


Figure 19: **Symbols for Global Set/Reset**

Boundary Scan

The "bed of nails" has been the traditional method of testing electronic assemblies. This approach has become less appropriate, due to closer pin spacing and more sophisticated assembly methods like surface-mount technology and multi-layer boards. The IEEE Boundary Scan Standard 1149.1 was developed to facilitate board-level testing of electronic assemblies. Design and test engineers can embed a standard test logic structure in their device to achieve high fault coverage for I/O and internal logic. This structure is easily implemented with a four-pin interface on any boundary scan compatible device. IEEE 1149.1-compatible devices may be serial daisy-chained together, connected in parallel, or a combination of the two.

The Spartan and Spartan-XL families implement IEEE 1149.1-compatible BYPASS, PRELOAD/SAMPLE and EXTEST boundary scan instructions. When the boundary scan configuration option is selected, three normal user I/O pins become dedicated inputs for these functions. Another user output pin becomes the dedicated boundary scan output. The details of how to enable this circuitry are covered later in this section.

By exercising these input signals, the user can serially load commands and data into these devices to control the driving of their outputs and to examine their inputs. This method is an improvement over bed-of-nails testing. It avoids the need to over-drive device outputs, and it reduces the user interface to four pins. An optional fifth pin, a reset for the control logic, is described in the standard but is not implemented in the Spartan/XL devices.

The dedicated on-chip logic implementing the IEEE 1149.1 functions includes a 16-state machine, an instruction register and a number of data registers. The functional details can be found in the IEEE 1149.1 specification and are also discussed in the Xilinx application note: "*Boundary Scan in FPGA Devices*."

Master Serial Mode

The Master serial mode uses an internal oscillator to generate a Configuration Clock (CCLK) for driving potential slave devices and the Xilinx serial-configuration PROM (SPROM). The CCLK speed is selectable as either 1 MHz (default) or 8 MHz. Configuration always starts at the default slow frequency, then can switch to the higher frequency during the first frame. Frequency tolerance is –50% to +25%.

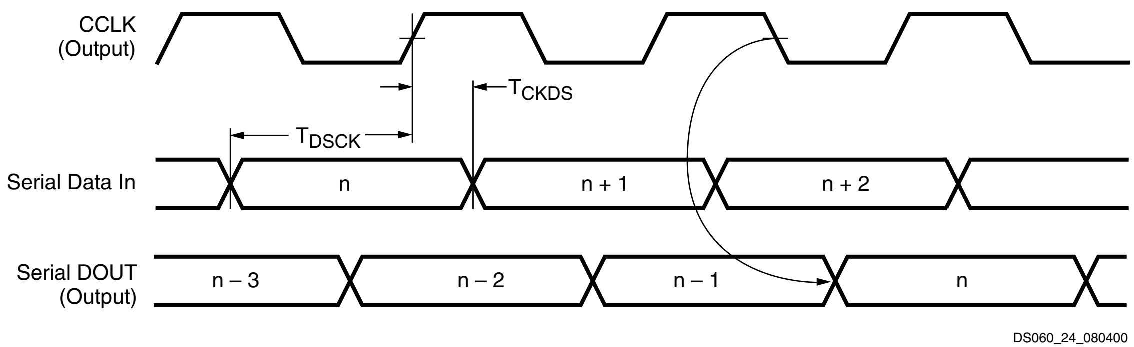
In Master Serial mode, the CCLK output of the device drives a Xilinx SPROM that feeds the FPGA DIN input. Each rising edge of the CCLK output increments the Serial PROM internal address counter. The next data bit is put on the SPROM data output, connected to the FPGA DIN pin. The FPGA accepts this data on the subsequent rising CCLK edge.

When used in a daisy-chain configuration the Master Serial FPGA is placed as the first device in the chain and is referred to as the lead FPGA. The lead FPGA presents the preamble data, and all data that overflows the lead device, on its DOUT pin. There is an internal pipeline delay of 1.5 CCLK periods, which means that DOUT changes on the falling CCLK edge, and the next FPGA in the daisy chain accepts data on the subsequent rising CCLK edge. See the timing diagram in Figure 24.

In the bitstream generation software, the user can specify Fast Configuration Rate, which, starting several bits into the first frame, increases the CCLK frequency by a factor of eight. For actual timing values please refer to the specification section. Be sure that the serial PROM and slaves are fast enough to support this data rate. Earlier families such as the XC3000 series do not support the Fast Configuration Rate option.

The SPROM CE input can be driven from either $\overline{\text{LDC}}$ or DONE. Using $\overline{\text{LDC}}$ avoids potential contention on the DIN pin, if this pin is configured as user I/O, but $\overline{\text{LDC}}$ is then restricted to be a permanently High user output after configuration. Using DONE can also avoid contention on DIN, provided the Early DONE option is invoked.

Figure 25 shows a full master/slave system. The leftmost device is in Master Serial mode, all other devices in the chain are in Slave Serial mode.



	Symbol	Description	Min	Units
CCLK	T_{DSCK}	DIN setup	20	ns
	T_{CKDS}	DIN hold	0	ns

Notes:
1. At power-up, V_{CC} must rise from 2.0V to V_{CC} min in less than 25 ms, otherwise delay configuration by pulling $\overline{\text{PROGRAM}}$ Low until V_{CC} is valid.
2. Master Serial mode timing is based on testing in slave mode.

Figure 24: **Master Serial Mode Programming Switching Characteristics**

Slave Serial Mode

In Slave Serial mode, the FPGA receives serial configuration data on the rising edge of CCLK and, after loading its configuration, passes additional data out, resynchronized on the next falling edge of CCLK.

In this mode, an external signal drives the CCLK input of the FPGA (most often from a Master Serial device). The serial configuration bitstream must be available at the DIN input of the lead FPGA a short setup time before each rising CCLK edge.

The lead FPGA then presents the preamble data—and all data that overflows the lead device—on its DOUT pin. There is an internal delay of 0.5 CCLK periods, which means that DOUT changes on the falling CCLK edge, and the next FPGA in the daisy chain accepts data on the subsequent rising CCLK edge.

Figure 25 shows a full master/slave system. A Spartan/XL device in Slave Serial mode should be connected as shown in the third device from the left.

Slave Serial is the default mode if the Mode pins are left unconnected, as they have weak pull-up resistors during configuration.

Multiple slave devices with identical configurations can be wired with parallel DIN inputs. In this way, multiple devices can be configured simultaneously.

Serial Daisy Chain

Multiple devices with different configurations can be connected together in a "daisy chain," and a single combined bitstream used to configure the chain of slave devices.

To configure a daisy chain of devices, wire the CCLK pins of all devices in parallel, as shown in Figure 25. Connect the DOUT of each device to the DIN of the next. The lead or master FPGA and following slaves each passes resynchronized configuration data coming from a single source. The header data, including the length count, is passed through and is captured by each FPGA when it recognizes the 0010 preamble. Following the length-count data, each FPGA outputs a High on DOUT until it has received its required number of data frames.

After an FPGA has received its configuration data, it passes on any additional frame start bits and configuration data on DOUT. When the total number of configuration clocks applied after memory initialization equals the value of the 24-bit length count, the FPGAs begin the start-up sequence and become operational together. FPGA I/O are normally released two CCLK cycles after the last configuration bit is received.

The daisy-chained bitstream is not simply a concatenation of the individual bitstreams. The PROM File Formatter must be used to combine the bitstreams for a daisy-chained configuration.

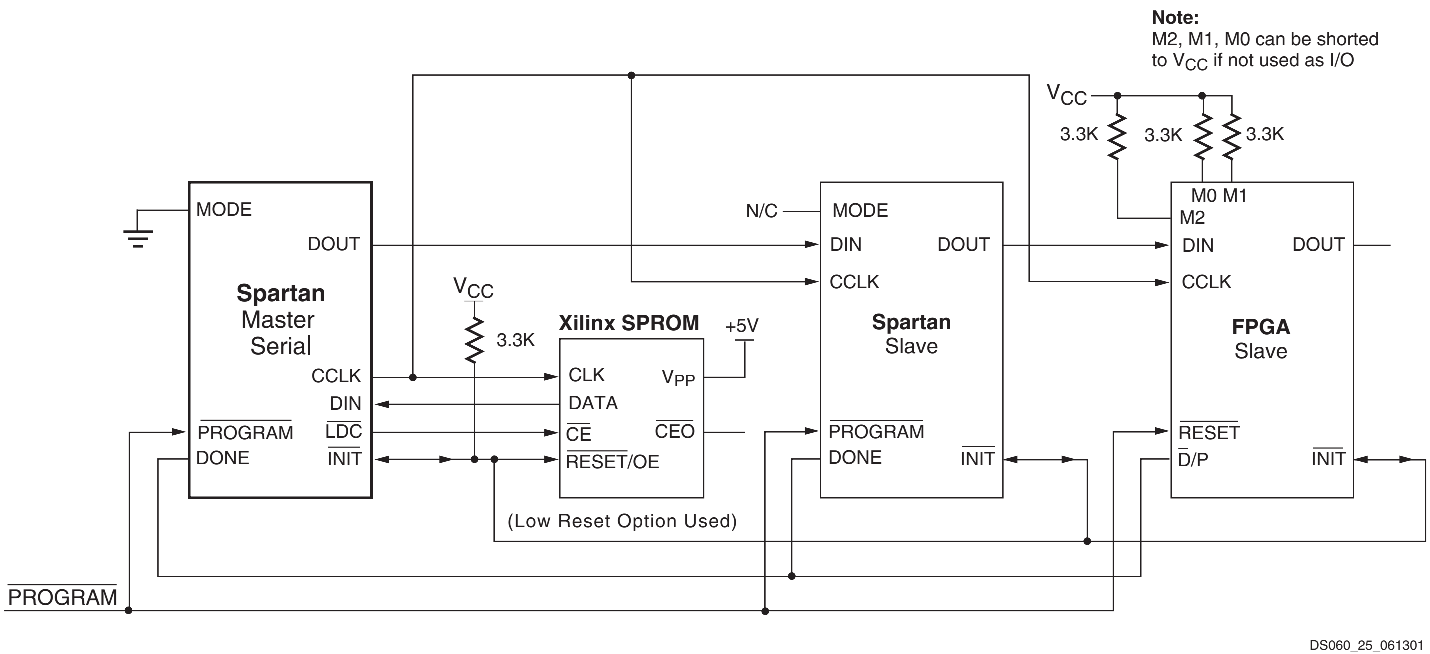


Figure 25: **Master/Slave Serial Mode Circuit Diagram**

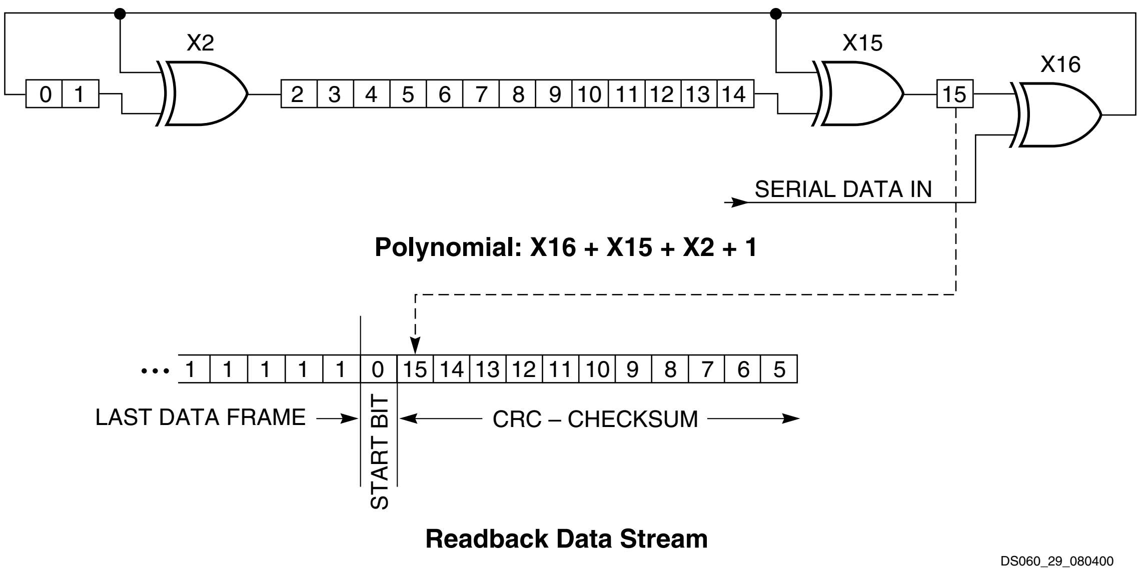


Figure 29: **Circuit for Generating CRC-16**

Configuration Sequence

There are four major steps in the Spartan/XL FPGA power-up configuration sequence.

- Configuration Memory Clear
- Initialization
- Configuration
- Start-up

The full process is illustrated in Figure 30.

Configuration Memory Clear

When power is first applied or is reapplied to an FPGA, an internal circuit forces initialization of the configuration logic. When V_{CC} reaches an operational level, and the circuit passes the write and read test of a sample pair of configuration bits, a time delay is started. This time delay is nominally 16 ms. The delay is four times as long when in Master Serial Mode to allow ample time for all slaves to reach a stable V_{CC}. When all $\overline{\text{INIT}}$ pins are tied together, as recommended, the longest delay takes precedence. Therefore, devices with different time delays can easily be mixed and matched in a daisy chain.

This delay is applied only on power-up. It is not applied when reconfiguring an FPGA by pulsing the $\overline{\text{PROGRAM}}$ pin Low. During this time delay, or as long as the $\overline{\text{PROGRAM}}$ input is asserted, the configuration logic is held in a Configuration Memory Clear state. The configuration-memory frames are consecutively initialized, using the internal oscillator.

At the end of each complete pass through the frame addressing, the power-on time-out delay circuitry and the level of the $\overline{\text{PROGRAM}}$ pin are tested. If neither is asserted, the logic initiates one additional clearing of the configuration frames and then tests the $\overline{\text{INIT}}$ input.

Initialization

During initialization and configuration, user pins HDC, $\overline{\text{LDC}}$, $\overline{\text{INIT}}$ and DONE provide status outputs for the system interface. The outputs $\overline{\text{LDC}}$, $\overline{\text{INIT}}$ and DONE are held Low and HDC is held High starting at the initial application of power.

The open drain $\overline{\text{INIT}}$ pin is released after the final initialization pass through the frame addresses. There is a deliberate delay before a Master-mode device recognizes an inactive $\overline{\text{INIT}}$. Two internal clocks after the $\overline{\text{INIT}}$ pin is recognized as High, the device samples the MODE pin to determine the configuration mode. The appropriate interface lines become active and the configuration preamble and data can be loaded.

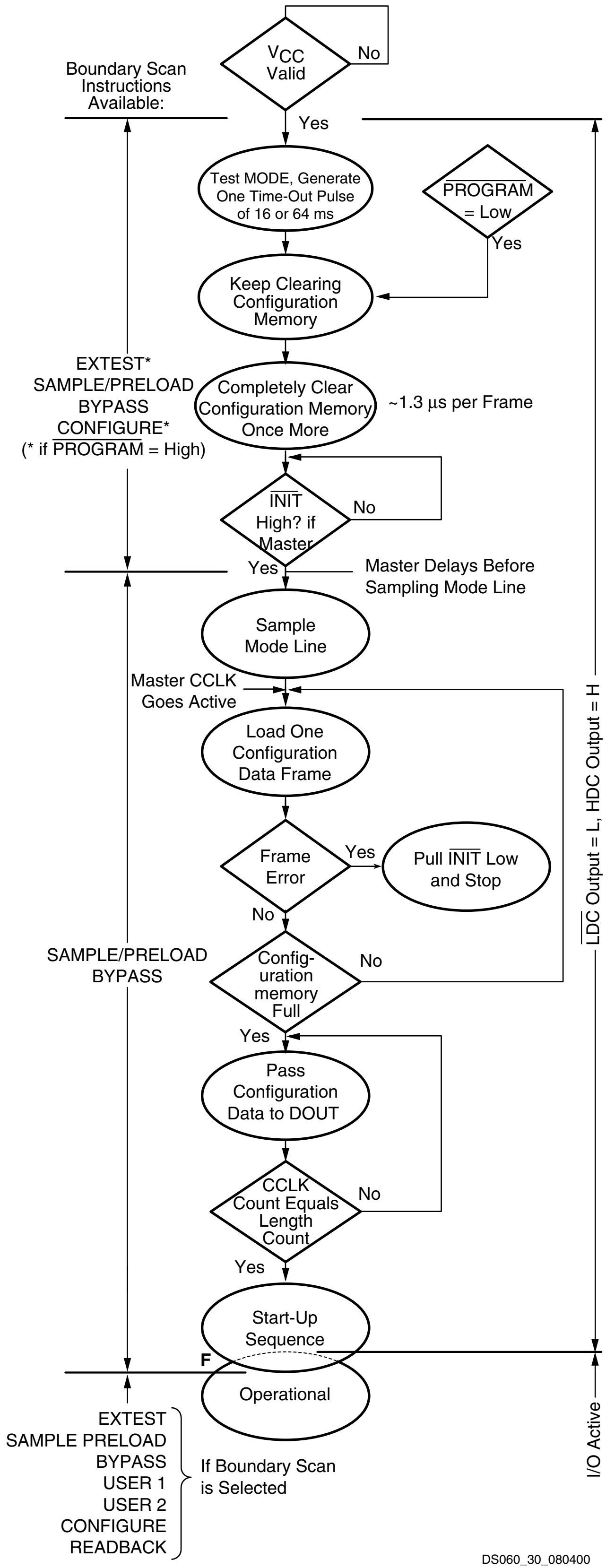


Figure 30: **Power-up Configuration Sequence**

Configuration

The 0010 preamble code indicates that the following 24 bits represent the length count for serial modes. The length count is the total number of configuration clocks needed to load the complete configuration data. (Four additional configuration clocks are required to complete the configuration process, as discussed below.) After the preamble and the length count have been passed through to any device in the daisy chain, its DOUT is held High to prevent frame start bits from reaching any daisy-chained devices. In Spartan-XL family Express mode, the length count bits are ignored, and DOUT is held Low, to disable the next device in the pseudo daisy chain.

A specific configuration bit, early in the first frame of a master device, controls the configuration-clock rate and can increase it by a factor of eight. Therefore, if a fast configuration clock is selected by the bitstream, the slower clock rate is used until this configuration bit is detected.

Each frame has a start field followed by the frame-configuration data bits and a frame error field. If a frame data error is detected, the FPGA halts loading, and signals the error by pulling the open-drain INIT pin Low. After all configuration frames have been loaded into an FPGA using a serial mode, DOUT again follows the input data so that the remaining data is passed on to the next device. In Spartan-XL family Express mode, when the first device is fully programmed, DOUT goes High to enable the next device in the chain.

Delaying Configuration After Power-Up

There are two methods of delaying configuration after power-up: put a logic Low on the PROGRAM input, or pull the bidirectional INIT pin Low, using an open-collector (open-drain) driver. (See Figure 30.)

A Low on the PROGRAM input is the more radical approach, and is recommended when the power-supply rise time is excessive or poorly defined. As long as PROGRAM is Low, the FPGA keeps clearing its configuration memory. When PROGRAM goes High, the configuration memory is cleared one more time, followed by the beginning of configuration, provided the INIT input is not externally held Low. Note that a Low on the PROGRAM input automatically forces a Low on the INIT output. The Spartan/XL FPGA PROGRAM pin has a permanent weak pull-up.

Avoid holding PROGRAM Low for more than 500 µs. The 500 µs maximum limit is only a recommendation, not a requirement. The only effect of holding PROGRAM Low for more than 500 µs is an increase in current, measured at about 40 mA in the XCS40XL. This increased current cannot damage the device. This applies only during reconfiguration, not during power-up. The INIT pin can also be held Low to delay reconfiguration, and the same characteristics apply as for the PROGRAM pin.

Using an open-collector or open-drain driver to hold INIT Low before the beginning of configuration causes the FPGA

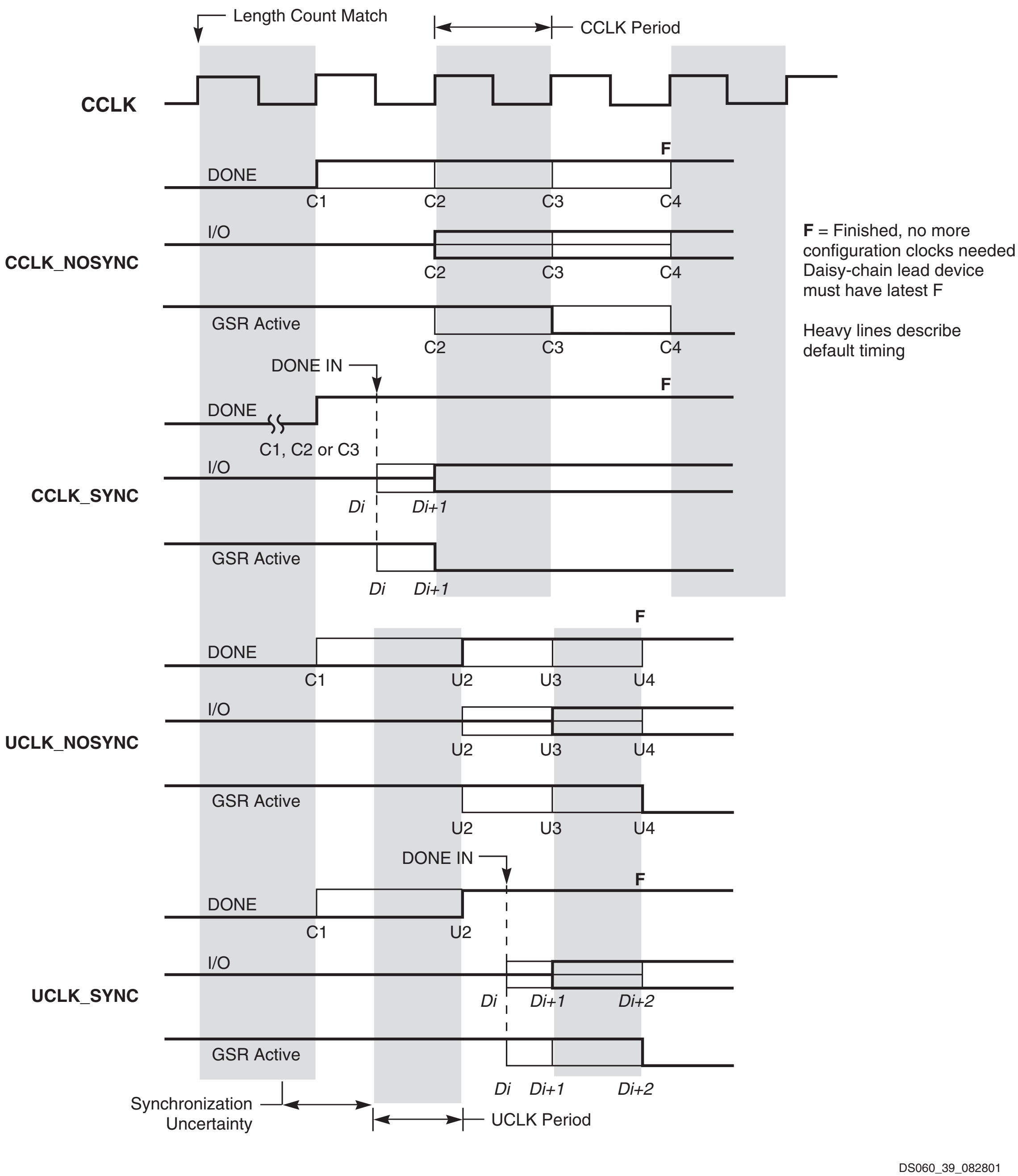


Figure 31: **Start-up Timing**

Configuration Through the Boundary Scan Pins

Spartan/XL devices can be configured through the boundary scan pins. The basic procedure is as follows:

- Power up the FPGA with $\overline{\text{INIT}}$ held Low (or drive the $\overline{\text{PROGRAM}}$ pin Low for more than 300 ns followed by a High while holding $\overline{\text{INIT}}$ Low). Holding $\overline{\text{INIT}}$ Low allows enough time to issue the CONFIG command to the FPGA. The pin can be used as I/O after configuration if a resistor is used to hold $\overline{\text{INIT}}$ Low.
- Issue the CONFIG command to the TMS input.
- Wait for $\overline{\text{INIT}}$ to go High.
- Sequence the boundary scan Test Access Port to the SHIFT-DR state.
- Toggle TCK to clock data into TDI pin.

The user must account for all TCK clock cycles after $\overline{\text{INIT}}$ goes High, as all of these cycles affect the Length Count compare.

For more detailed information, refer to the Xilinx application note, "*Boundary Scan in FPGA Devices*." This application note applies to Spartan and Spartan-XL devices.

Readback

The user can read back the content of configuration memory and the level of certain internal nodes without interfering with the normal operation of the device.

Readback not only reports the downloaded configuration bits, but can also include the present state of the device, represented by the content of all flip-flops and latches in CLBs and IOBs, as well as the content of function generators used as RAMs.

Although readback can be performed while the device is operating, for best results and to freeze a known capture state, it is recommended that the clock inputs be stopped until readback is complete.

Readback of Spartan-XL family Express mode bitstreams results in data that does not resemble the original bitstream, because the bitstream format differs from other modes.

Spartan/XL FPGA Readback does not use any dedicated pins, but uses four internal nets (RDBK.TRIG, RDBK.DATA, RDBK.RIP and RDBK.CLK) that can be routed to any IOB. To access the internal Readback signals, instantiate the READBACK library symbol and attach the appropriate pad symbols, as shown in Figure 32.

After Readback has been initiated by a Low-to-High transition on RDBK.TRIG, the RDBK.RIP (Read In Progress) output goes High on the next rising edge of RDBK.CLK. Subsequent rising edges of this clock shift out Readback data on the RDBK.DATA net.

Readback data does not include the preamble, but starts with five dummy bits (all High) followed by the Start bit (Low) of the first frame. The first two data bits of the first frame are always High.

Each frame ends with four error check bits. They are read back as High. The last seven bits of the last frame are also read back as High. An additional Start bit (Low) and an 11-bit Cyclic Redundancy Check (CRC) signature follow, before RDBK.RIP returns Low.

Readback Options

Readback options are: Readback Capture, Readback Abort, and Clock Select. They are set with the bitstream generation software.

Readback Capture

When the Readback Capture option is selected, the data stream includes sampled values of CLB and IOB signals. The rising edge of RDBK.TRIG latches the inverted values of the four CLB outputs, the IOB output flip-flops and the input signals I1 and I2. Note that while the bits describing configuration (interconnect, function generators, and RAM content) are *not* inverted, the CLB and IOB output signals *are* inverted. RDBK.TRIG is located in the lower-left corner of the device.

When the Readback Capture option is not selected, the values of the capture bits reflect the configuration data originally written to those memory locations. If the RAM capability of the CLBs is used, RAM data are available in Readback, since they directly overwrite the F and G function-table configuration of the CLB.

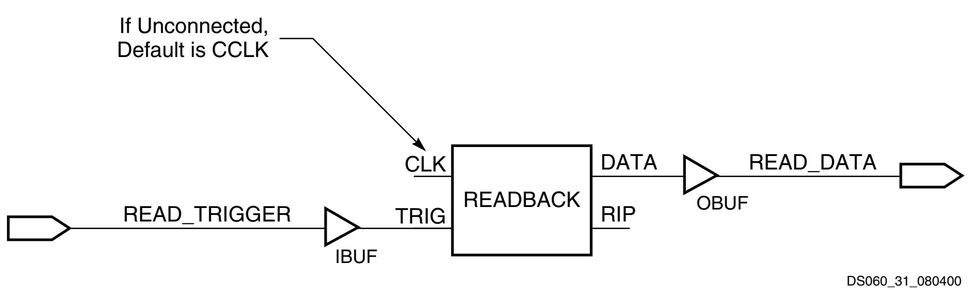


Figure 32: **Readback Example**

Spartan Family Detailed Specifications

Definition of Terms

In the following tables, some specifications may be designated as Advance or Preliminary. These terms are defined as follows:

Advance: Initial estimates based on simulation and/or extrapolation from other speed grades, devices, or families. Values are subject to change. Use as estimates, not for production.

Preliminary: Based on preliminary characterization. Further changes are not expected.

Unmarked: Specifications not identified as either Advance or Preliminary are to be considered Final.

Notwithstanding the definition of the above terms, all specifications are subject to change without notice.

Except for pin-to-pin input and output parameters, the AC parameter delay specifications included in this document are derived from measuring internal test patterns. All specifications are representative of worst-case supply voltage and junction temperature conditions. The parameters included are common to popular designs and typical applications.

Spartan Family Absolute Maximum Ratings(1)

Symbol	Description		Value	Units
V_{CC}	Supply voltage relative to GND		–0.5 to +7.0	V
V_{IN}	Input voltage relative to GND(2,3)		–0.5 to V_{CC} +0.5	V
V_{TS}	Voltage applied to 3-state output(2,3)		–0.5 to V_{CC} +0.5	V
T_{STG}	Storage temperature (ambient)		–65 to +150	°C
T_J	Junction temperature	Plastic packages	+125	°C

Notes:
1. Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those listed under Operating Conditions is not implied. Exposure to Absolute Maximum Ratings conditions for extended periods of time may affect device reliability.
2. Maximum DC overshoot (above V_{CC}) or undershoot (below GND) must be limited to either 0.5V or 10 mA, whichever is easier to achieve.
3. Maximum AC (during transitions) conditions are as follows; the device pins may undershoot to –2.0V or overshoot to +7.0V, provided this overshoot or undershoot lasts no more than 11 ns with a forcing current no greater than 100 mA.
4. For soldering guidelines, see the Package Information on the Xilinx website.

Spartan Family Recommended Operating Conditions

Symbol	Description		Min	Max	Units
V_{CC}	Supply voltage relative to GND, T_J = 0°C to +85°C	Commercial	4.75	5.25	V
	Supply voltage relative to GND, T_J = –40°C to +100°C(1)	Industrial	4.5	5.5	V
V_{IH}	High-level input voltage(2)	TTL inputs	2.0	V_{CC}	V
		CMOS inputs	70%	100%	V_{CC}
V_{IL}	Low-level input voltage(2)	TTL inputs	0	0.8	V
		CMOS inputs	0	20%	V_{CC}
T_{IN}	Input signal transition time		-	250	ns

Notes:
1. At junction temperatures above those listed as Recommended Operating Conditions, all delay parameters increase by 0.35% per °C.
2. Input and output measurement thresholds are: 1.5V for TTL and 2.5V for CMOS.

Spartan-XL Family Global Buffer Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.

When fewer vertical clock lines are connected, the clock distribution is faster; when multiple clock lines per column are driven from the same global clock, the delay is longer. For more specific, more precise, and worst-case guaranteed data, reflecting the actual routing structure, use the values provided by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. These path delays, provided as a guideline, have been extracted from the static timing analyzer report. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature).

Symbol	Description	Device	Speed Grade -5 Max	Speed Grade -4 Max	Units
T_{GLS}	From pad through buffer, to any clock K	XCS05XL	1.4	1.5	ns
		XCS10XL	1.7	1.8	ns
		XCS20XL	2.0	2.1	ns
		XCS30XL	2.3	2.5	ns
		XCS40XL	2.6	2.8	ns

Spartan-XL Family CLB RAM Synchronous (Edge-Triggered) Write Operation Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values apply to all Spartan-XL devices and are expressed in nanoseconds unless otherwise noted.

Symbol	Single Port RAM	Size(1)	Speed Grade -5 Min	-5 Max	-4 Min	-4 Max	Units
Write Operation							
T_{WCS}	Address write cycle time (clock K period)	16x2	7.7	-	8.4	-	ns
T_{WCTS}		32x1	7.7	-	8.4	-	ns
T_{WPS}	Clock K pulse width (active edge)	16x2	3.1	-	3.6	-	ns
T_{WPTS}		32x1	3.1	-	3.6	-	ns
T_{ASS}	Address setup time before clock K	16x2	1.3	-	1.5	-	ns
T_{ASTS}		32x1	1.5	-	1.7	-	ns
T_{DSS}	DIN setup time before clock K	16x2	1.5	-	1.7	-	ns
T_{DSTS}		32x1	1.8	-	2.1	-	ns
T_{WSS}	WE setup time before clock K	16x2	1.4	-	1.6	-	ns
T_{WSTS}		32x1	1.3	-	1.5	-	ns
	All hold times after clock K	16x2	0.0	-	0.0	-	ns
T_{WOS}	Data valid after clock K	32x1	-	4.5	-	5.3	ns
T_{WOTS}		16x2	-	5.4	-	6.3	ns
Read Operation							
T_{RC}	Address read cycle time	16x2	2.6	-	3.1	-	ns
T_{RCT}		32x1	3.8	-	5.5	-	ns
T_{ILO}	Data Valid after address change (no Write Enable)	16x2	-	1.0	-	1.1	ns
T_{IHO}		32x1	-	1.7	-	2.0	ns
T_{ICK}	Address setup time before clock K	16x2	0.6	-	0.7	-	ns
T_{IHCK}		32x1	1.3	-	1.6	-	ns

Notes:
1. Timing for 16 x 1 RAM option is identical to 16 x 2 RAM timing.

Spartan-XL Family Pin-to-Pin Input Parameter Guidelines

All devices are 100% functionally tested. Pin-to-pin timing parameters are derived from measuring external and internal test patterns and are guaranteed over worst-case operating conditions (supply voltage and junction temperature). Listed below are representative values for typical pin locations and normal clock loading.

Spartan-XL Family Setup and Hold

Symbol	Description	Device	Speed Grade -5 Max	Speed Grade -4 Max	Units
Input Setup/Hold Times Using Global Clock and IFF					
T_{SUF}/T_{HF}	No Delay	XCS05XL	1.1/2.0	1.6/2.6	ns
		XCS10XL	1.0/2.2	1.5/2.8	ns
		XCS20XL	0.9/2.4	1.4/3.0	ns
		XCS30XL	0.8/2.6	1.3/3.2	ns
		XCS40XL	0.7/2.8	1.2/3.4	ns
T_{SU}/T_{H}	Full Delay	XCS05XL	3.9/0.0	5.1/0.0	ns
		XCS10XL	4.1/0.0	5.3/0.0	ns
		XCS20XL	4.3/0.0	5.5/0.0	ns
		XCS30XL	4.5/0.0	5.7/0.0	ns
		XCS40XL	4.7/0.0	5.9/0.0	ns

Notes:
1. IFF = Input Flip-Flop or Latch
2. Setup time is measured with the fastest route and the lightest load. Hold time is measured using the furthest distance and a reference load of one clock pin per IOB/CLB.

Capacitive Load Factor

Figure 35 shows the relationship between I/O output delay and load capacitance. It allows a user to adjust the specified output delay if the load capacitance is different than 50 pF. For example, if the actual load capacitance is 120 pF, add 2.5 ns to the specified delay. If the load capacitance is 20 pF, subtract 0.8 ns from the specified output delay. Figure 35 is usable over the specified operating conditions of voltage and temperature and is independent of the output slew rate control.

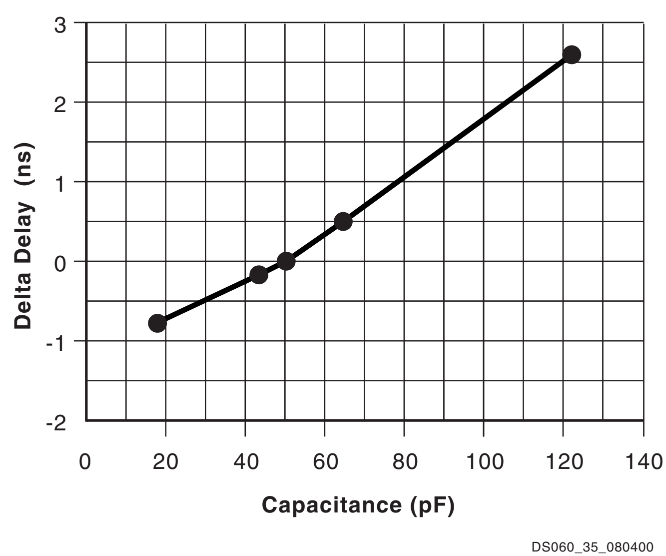


Figure 35: **Delay Factor at Various Capacitive Loads**

Device-Specific Pinout Tables

Device-specific tables include all packages for each Spartan and Spartan-XL device. They follow the pad locations around the die, and include boundary scan register locations.

Some Spartan-XL devices are available in Pb-free package options. The Pb-free package options have the same pinouts as the standard package options.

XCS05 and XCS05XL Device Pinouts

XCS05/XL Pad Name	PC84(4)	VQ100	Bndry Scan
VCC	P2	P89	-
I/O	P3	P90	32
I/O	P4	P91	35
I/O	-	P92	38
I/O	-	P93	41
I/O	P5	P94	44
I/O	P6	P95	47
I/O	P7	P96	50
I/O	P8	P97	53
I/O	P9	P98	56
I/O, SGCK1(1), GCK8(2)	P10	P99	59
VCC	P11	P100	-
GND	P12	P1	-
I/O, PGCK1(1), GCK1(2)	P13	P2	62
I/O	P14	P3	65
I/O, TDI	P15	P4	68
I/O, TCK	P16	P5	71
I/O, TMS	P17	P6	74
I/O	P18	P7	77
I/O	-	P8	83
I/O	P19	P9	86
I/O	P20	P10	89
GND	P21	P11	-
VCC	P22	P12	-
I/O	P23	P13	92
I/O	P24	P14	95
I/O	-	P15	98
I/O	P25	P16	104
I/O	P26	P17	107
I/O	P27	P18	110
I/O	-	P19	113
I/O	P28	P20	116
I/O, SGCK2(1), GCK2(2)	P29	P21	119
Not Connected(1), M1(2)	P30	P22	122
GND	P31	P23	-
MODE(1), M0 (2)	P32	P24	125
VCC	P33	P25	-

XCS05 and XCS05XL Device Pinouts

XCS05/XL Pad Name	PC84(4)	VQ100	Bndry Scan
Not Connected(1), PWRDWN(2)	P34	P26	126(1)
I/O, PGCK2(1), GCK3 (2)	P35	P27	127(3)
I/O (HDC)	P36	P28	130(3)
I/O	-	P29	133(3)
I/O (LDC)	P37	P30	136(3)
I/O	P38	P31	139(3)
I/O	P39	P32	142(3)
I/O	-	P33	145(3)
I/O	-	P34	148(3)
I/O	P40	P35	151(3)
I/O (INIT)	P41	P36	154(3)
VCC	P42	P37	-
GND	P43	P38	-
I/O	P44	P39	157(3)
I/O	P45	P40	160(3)
I/O	-	P41	163(3)
I/O	-	P42	166(3)
I/O	P46	P43	169(3)
I/O	P47	P44	172(3)
I/O	P48	P45	175(3)
I/O	P49	P46	178(3)
I/O	P50	P47	181(3)
I/O, SGCK3(1), GCK4(2)	P51	P48	184(3)
GND	P52	P49	-
DONE	P53	P50	-
VCC	P54	P51	-
PROGRAM	P55	P52	-
I/O (D7(2))	P56	P53	187(3)
I/O, PGCK3(1), GCK5(2)	P57	P54	190(3)
I/O (D6(2))	P58	P55	193(3)
I/O	-	P56	196(3)
I/O (D5(2))	P59	P57	199(3)
I/O	P60	P58	202(3)
I/O	-	P59	205(3)
I/O	-	P60	208(3)
I/O (D4(2))	P61	P61	211(3)
I/O	P62	P62	214(3)
VCC	P63	P63	-
GND	P64	P64	-
I/O (D3(2))	P65	P65	217(3)
I/O	P66	P66	220(3)
I/O	-	P67	223(3)
I/O (D2(2))	P67	P68	229(3)
I/O	P68	P69	232(3)
I/O (D1(2))	P69	P70	235(3)

XCS10 and XCS10XL Device Pinouts

XCS10/XL Pad Name	PC84(4)	VQ100	CS144(2,4)	TQ144	Bndry Scan
VCC	P33	P25	N1	P37	-
Not Connected(1) $\overline{\text{PWRDWN}}$(2)	P34	P26	N2	P38	174 (1)
I/O, PGCK2(1) GCK3(2)	P35	P27	M3	P39	175(3)
I/O (HDC)	P36	P28	N3	P40	178 (3)
I/O	-	-	K4	P41	181 (3)
I/O	-	-	L4	P42	184 (3)
I/O	-	P29	M4	P43	187 (3)
I/O ($\overline{\text{LDC}}$)	P37	P30	N4	P44	190 (3)
GND	-	-	K5	P45	-
I/O	-	-	L5	P46	193 (3)
I/O	-	-	M5	P47	196 (3)
I/O	P38	P31	N5	P48	199 (3)
I/O	P39	P32	K6	P49	202 (3)
I/O	-	P33	L6	P50	205 (3)
I/O	-	P34	M6	P51	208 (3)
I/O	P40	P35	N6	P52	211 (3)
I/O ($\overline{\text{INIT}}$)	P41	P36	M7	P53	214 (3)
VCC	P42	P37	N7	P54	-
GND	P43	P38	L7	P55	-
I/O	P44	P39	K7	P56	217 (3)
I/O	P45	P40	N8	P57	220 (3)
I/O	-	P41	M8	P58	223 (3)
I/O	-	P42	L8	P59	226 (3)
I/O	P46	P43	K8	P60	229 (3)
I/O	P47	P44	N9	P61	232 (3)
I/O	-	-	M9	P62	235 (3)
I/O	-	-	L9	P63	238 (3)
GND	-	-	K9	P64	-
I/O	P48	P45	N10	P65	241 (3)
I/O	P49	P46	M10	P66	244 (3)
I/O	-	-	L10	P67	247 (3)
I/O	-	-	N11	P68	250 (3)
I/O	P50	P47	M11	P69	253 (3)
I/O, SGCK3(1) GCK4(2)	P51	P48	L11	P70	256 (3)
GND	P52	P49	N12	P71	-
DONE	P53	P50	M12	P72	-
VCC	P54	P51	N13	P73	-
$\overline{\text{PROGRAM}}$	P55	P52	M13	P74	-
I/O (D7(2))	P56	P53	L12	P75	259 (3)

XCS10 and XCS10XL Device Pinouts

XCS10/XL Pad Name	PC84(4)	VQ100	CS144(2,4)	TQ144	Bndry Scan
I/O, PGCK3(1) GCK5(2)	P57	P54	L13	P76	262 (3)
I/O	-	-	K10	P77	265 (3)
I/O	-	-	K11	P78	268 (3)
I/O (D6(2))	P58	P55	K12	P79	271 (3)
I/O	-	P56	K13	P80	274 (3)
GND	-	-	J10	P81	-
I/O	-	-	J11	P82	277 (3)
I/O	-	-	J12	P83	280 (3)
I/O (D5(2))	P59	P57	J13	P84	283 (3)
I/O	P60	P58	H10	P85	286 (3)
I/O	-	P59	H11	P86	289 (3)
I/O	-	P60	H12	P87	292 (3)
I/O (D4(2))	P61	P61	H13	P88	295 (3)
I/O	P62	P62	G12	P89	298 (3)
VCC	P63	P63	G13	P90	-
GND	P64	P64	G11	P91	-
I/O (D3(2))	P65	P65	G10	P92	301 (3)
I/O	P66	P66	F13	P93	304 (3)
I/O	-	P67	F12	P94	307 (3)
I/O	-	-	F11	P95	310 (3)
I/O (D2(2))	P67	P68	F10	P96	313 (3)
I/O	P68	P69	E13	P97	316 (3)
I/O	-	-	E12	P98	319 (3)
I/O	-	-	E11	P99	322 (3)
GND	-	-	E10	P100	-
I/O (D1(2))	P69	P70	D13	P101	325 (3)
I/O	P70	P71	D12	P102	328 (3)
I/O	-	-	D11	P103	331 (3)
I/O	-	-	C13	P104	334 (3)
I/O (D0(2), DIN)	P71	P72	C12	P105	337 (3)
I/O, SGCK4(1) GCK6(2) (DOUT)	P72	P73	C11	P106	340 (3)
CCLK	P73	P74	B13	P107	-
VCC	P74	P75	B12	P108	-
O, TDO	P75	P76	A13	P109	0
GND	P76	P77	A12	P110	-
I/O	P77	P78	B11	P111	2
I/O, PGCK4(1) GCK7(2)	P78	P79	A11	P112	5
I/O	-	-	D10	P113	8
I/O	-	-	C10	P114	11
I/O (CS1(2))	P79	P80	B10	P115	14

Additional XCS20/XL Package Pins

PQ208					
Not Connected Pins					
P12	P18 (1)	P33 (1)	P39	P65	P71 (1)
P86 (1)	P92	P111	P121(1)	P140(1)	P144
P165	P173(1)	P192(1)	P202	P203	-
9/16/98					

Notes:
1. 5V Spartan family only
2. 3V Spartan-XL family only
3. The "PWRDWN" on the XCS20XL is not part of the Boundary Scan chain. For the XCS20XL, subtract 1 from all Boundary Scan numbers from GCK3 on (247 and higher).
4. CS144 package discontinued by **PDN2004-01**

XCS30 and XCS30XL Device Pinouts

XCS30/XL Pad Name	VQ100(5)	TQ144	PQ208	PQ240	BG256(5)	CS280(2,5)	Bndry Scan
VCC	P89	P128	P183	P212	VCC(4)	C10	-
I/O	P90	P129	P184	P213	C10	D10	74
I/O	P91	P130	P185	P214	D10	E10	77
I/O	P92	P131	P186	P215	A9	A9	80
I/O	P93	P132	P187	P216	B9	B9	83
I/O	-	-	P188	P217	C9	C9	86
I/O	-	-	P189	P218	D9	D9	89
I/O	P94	P133	P190	P220	A8	A8	92
I/O	P95	P134	P191	P221	B8	B8	95
VCC	-	-	P192	P222	VCC(4)	A7	-
I/O	-	-	-	P223	A6	B7	98
I/O	-	-	-	P224	C7	C7	101
I/O	-	P135	P193	P225	B6	D7	104
I/O	-	P136	P194	P226	A5	A6	107
GND	-	P137	P195	P227	GND(4)	GND(4)	-
I/O	-	-	P196	P228	C6	B6	110
I/O	-	-	P197	P229	B5	C6	113
I/O	-	-	P198	P230	A4	D6	116
I/O	-	-	P199	P231	C5	E6	119
I/O	P96	P138	P200	P232	B4	A5	122
I/O	P97	P139	P201	P233	A3	C5	125
I/O	-	-	P202	P234	D5	B4	128
I/O	-	-	P203	P235	C4	C4	131
I/O	-	P140	P204	P236	B3	A3	134
I/O	-	P141	P205	P237	B2	A2	137
I/O	P98	P142	P206	P238	A2	B3	140
I/O, SGCK1(1), GCK8(2)	P99	P143	P207	P239	C3	B2	143
VCC	P100	P144	P208	P240	VCC(4)	A1	-
GND	P1	P1	P1	P1	GND(4)	GND(4)	-
I/O, PGCK1(1), GCK1(2)	P2	P2	P2	P2	B1	C3	146
I/O	P3	P3	P3	P3	C2	C2	149
I/O	-	P4	P4	P4	D2	B1	152

XCS30 and XCS30XL Device Pinouts *(Continued)*

XCS30/XL Pad Name	VQ100(5)	TQ144	PQ208	PQ240	BG256(5)	CS280(2,5)	Bndry Scan
I/O	P18	P28	P44	P52	V1	T1	272
I/O	P19	P29	P45	P53	T4	T2	275
I/O	-	P30	P46	P54	U3	T3	278
I/O	-	P31	P47	P55	V2	U1	281
I/O	P20	P32	P48	P56	W1	V1	284
I/O, SGCK2(1), GCK2(2)	P21	P33	P49	P57	V3	U2	287
Not Connected(1), M1(2)	P22	P34	P50	P58	W2	V2	290
GND	P23	P35	P51	P59	GND(4)	GND(4)	-
MODE(1), M0(2)	P24	P36	P52	P60	Y1	W1	293
VCC	P25	P37	P53	P61	VCC(4)	U3	-
Not Connected (1), $\overline{\text{PWRDWN}}$(2)	P26	P38	P54	P62	W3	V3	294 (1)
I/O, PGCK2(1), GCK3(2)	P27	P39	P55	P63	Y2	W2	295 (3)
I/O (HDC)	P28	P40	P56	P64	W4	W3	298 (3)
I/O	-	P41	P57	P65	V4	T4	301 (3)
I/O	-	P42	P58	P66	U5	U4	304 (3)
I/O	P29	P43	P59	P67	Y3	V4	307 (3)
I/O ($\overline{\text{LDC}}$)	P30	P44	P60	P68	Y4	W4	310 (3)
I/O	-	-	P61	P69	V5	T5	313 (3)
I/O	-	-	P62	P70	W5	W5	316 (3)
I/O	-	-	P63	P71	Y5	R6	319 (3)
I/O	-	-	P64	P72	V6	U6	322 (3)
I/O	-	-	P65	P73	W6	V6	325 (3)
I/O	-	-	-	P74	Y6	T6	328 (3)
GND	-	P45	P66	P75	GND(4)	GND(4)	-
I/O	-	P46	P67	P76	W7	W6	331 (3)
I/O	-	P47	P68	P77	Y7	U7	334 (3)
I/O	P31	P48	P69	P78	V8	V7	337 (3)
I/O	P32	P49	P70	P79	W8	W7	340 (3)
VCC	-	-	P71	P80	VCC(4)	T7	-
I/O	-	-	P72	P81	Y8	W8	343 (3)
I/O	-	-	P73	P82	U9	U8	346 (3)
I/O	-	-	-	P84	Y9	W9	349 (3)
I/O	-	-	-	P85	W10	V9	352 (3)
I/O	P33	P50	P74	P86	V10	U9	355 (3)
I/O	P34	P51	P75	P87	Y10	T9	358 (3)
I/O	P35	P52	P76	P88	Y11	W10	361 (3)
I/O ($\overline{\text{INIT}}$)	P36	P53	P77	P89	W11	V10	364 (3)
VCC	P37	P54	P78	P90	VCC(4)	U10	-
GND	P38	P55	P79	P91	GND(4)	GND(4)	-
I/O	P39	P56	P80	P92	V11	T10	367 (3)
I/O	P40	P57	P81	P93	U11	R10	370 (3)
I/O	P41	P58	P82	P94	Y12	W11	373 (3)
I/O	P42	P59	P83	P95	W12	V11	376 (3)
I/O	-	-	P84	P96	V12	U11	379 (3)

CS280

VCC Pins					
E5	E7	E8	E9	E11	E12
E13	G5	G15	H5	H15	J5
J15	L5	L15	M5	M15	N5
N15	R7	R8	R9	R11	R12
R13	-	-	-	-	-
Not Connected Pins					
A4	A12	C8	C12	C15	D1
D2	D5	D8	D17	D18	E15
H2	H3	H18	H19	L4	M1
M16	M18	R2	R4	R5	R15
R17	T8	T15	U5	V8	V12
W12	W16	-	-	-	-
Not Connected Pins (VCC in XCS40XL)					
B5	B15	E3	E18	R3	R18
V5	V15	-	-	-	-

5/21/02

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
VCC	P183	P212	VCC(4)	VCC(4)	-
I/O	P184	P213	C10	D10	86
I/O	P185	P214	D10	E10	89
I/O	P186	P215	A9	A9	92
I/O	P187	P216	B9	B9	95
I/O	P188	P217	C9	C9	98
I/O	P189	P218	D9	D9	101
I/O	P190	P220	A8	A8	104
I/O	P191	P221	B8	B8	107
I/O	-	-	C8	C8	110
I/O	-	-	A7	D8	113
VCC	P192	P222	VCC(4)	VCC(4)	-
I/O	-	P223	A6	B7	116
I/O	-	P224	C7	C7	119
I/O	P193	P225	B6	D7	122
I/O	P194	P226	A5	A6	125
GND	P195	P227	GND(4)	GND(4)	-
I/O	P196	P228	C6	B6	128
I/O	P197	P229	B5	C6	131
I/O	P198	P230	A4	D6	134
I/O	P199	P231	C5	E6	137

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
I/O	P200	P232	B4	A5	140
I/O	P201	P233	A3	C5	143
I/O	-	-	-	D5	146
I/O	-	-	-	A4	149
I/O	P202	P234	D5	B4	152
I/O	P203	P235	C4	C4	155
I/O	P204	P236	B3	A3	158
I/O	P205	P237	B2	A2	161
I/O	P206	P238	A2	B3	164
I/O, SGCK1(1), GCK8(2)	P207	P239	C3	B2	167
VCC	P208	P240	VCC(4)	VCC(4)	-
GND	P1	P1	GND(4)	GND(4)	-
I/O, PGCK1(1), GCK1(2)	P2	P2	B1	C3	170
I/O	P3	P3	C2	C2	173
I/O	P4	P4	D2	B1	176
I/O	P5	P5	D3	C1	179
I/O, TDI	P6	P6	E4	D4	182
I/O, TCK	P7	P7	C1	D3	185
I/O	-	-	-	D2	188
I/O	-	-	-	D1	191
I/O	P8	P8	D1	E2	194
I/O	P9	P9	E3	E4	197
I/O	P10	P10	E2	E1	200
I/O	P11	P11	E1	F5	203
I/O	P12	P12	F3	F3	206
I/O	-	P13	F2	F2	209
GND	P13	P14	GND(4)	GND(4)	-
I/O	P14	P15	G3	F4	212
I/O	P15	P16	G2	F1	215
I/O, TMS	P16	P17	G1	G3	218
I/O	P17	P18	H3	G2	221
VCC	P18	P19	VCC(4)	VCC(4)	-
I/O	-	P20	H2	G4	224
I/O	-	P21	H1	H1	227
I/O	-	-	J4	H3	230
I/O	-	-	J3	H2	233
I/O	P19	P23	J2	H4	236
I/O	P20	P24	J1	J1	239
I/O	P21	P25	K2	J2	242
I/O	P22	P26	K3	J3	245
I/O	P23	P27	K1	J4	248
I/O	P24	P28	L1	K1	251