AMD Xilinx - XCS30-3VQ100C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	576
Number of Logic Elements/Cells	1368
Total RAM Bits	18432
Number of I/O	77
Number of Gates	30000
Voltage - Supply	4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	100-TQFP
Supplier Device Package	100-VQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcs30-3vq100c

General Overview

Spartan series FPGAs are implemented with a regular, flexible, programmable architecture of Configurable Logic Blocks (CLBs), interconnected by a powerful hierarchy of versatile routing resources (routing channels), and surrounded by a perimeter of programmable Input/Output Blocks (IOBs), as seen in Figure 1. They have generous routing resources to accommodate the most complex interconnect patterns.

The devices are customized by loading configuration data into internal static memory cells. Re-programming is possible an unlimited number of times. The values stored in these memory cells determine the logic functions and interconnections implemented in the FPGA. The FPGA can either actively read its configuration data from an external serial PROM (Master Serial mode), or the configuration data can be written into the FPGA from an external device (Slave Serial mode).

Spartan series FPGAs can be used where hardware must be adapted to different user applications. FPGAs are ideal for shortening design and development cycles, and also offer a cost-effective solution for production rates well beyond 50,000 systems per month.

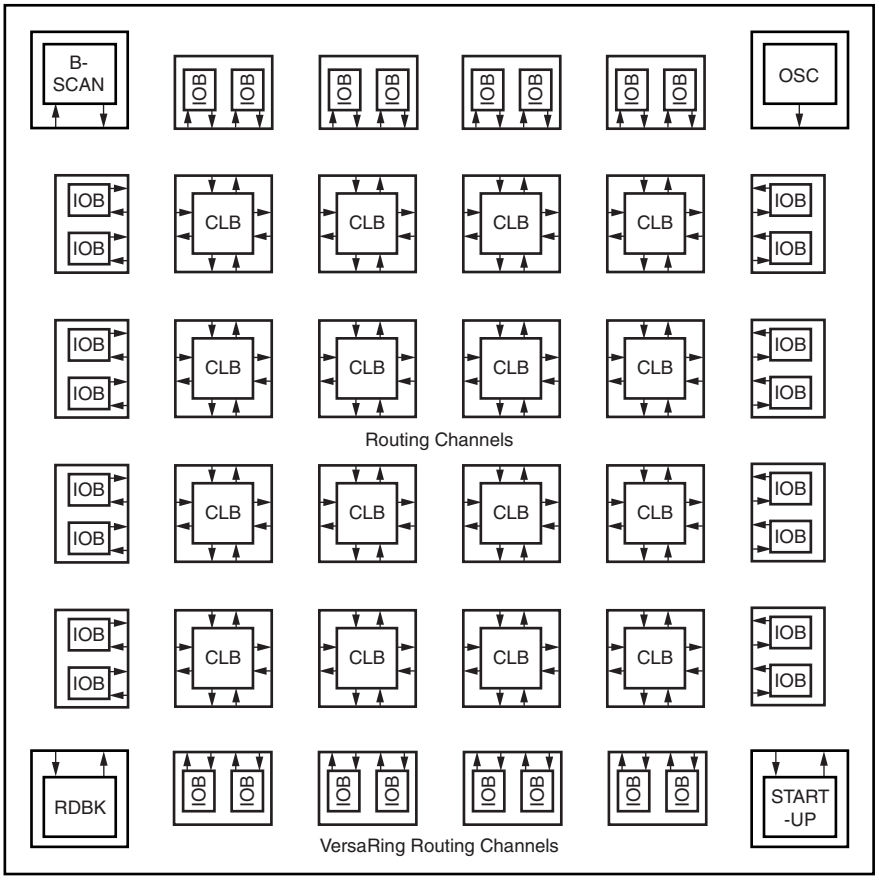

Figure 1: **Basic FPGA Block Diagram**

Spartan and Spartan-XL devices provide system clock rates exceeding 80 MHz and internal performance in excess of 150 MHz. In addition to the conventional benefit of high volume programmable logic solutions, Spartan series FPGAs also offer on-chip edge-triggered single-port and dual-port RAM, clock enables on all flip-flops, fast carry logic, and many other features.

The Spartan/XL families leverage the highly successful XC4000 architecture with many of that family's features and benefits. Technology advancements have been derived from the XC4000XLA process developments.

Logic Functional Description

The Spartan series uses a standard FPGA structure as shown in Figure 1, page 2. The FPGA consists of an array of configurable logic blocks (CLBs) placed in a matrix of routing channels. The input and output of signals is achieved through a set of input/output blocks (IOBs) forming a ring around the CLBs and routing channels.

- CLBs provide the functional elements for implementing the user's logic.
- IOBs provide the interface between the package pins and internal signal lines.
- Routing channels provide paths to interconnect the inputs and outputs of the CLBs and IOBs.

The functionality of each circuit block is customized during configuration by programming internal static memory cells. The values stored in these memory cells determine the logic functions and interconnections implemented in the FPGA.

Configurable Logic Blocks (CLBs)

The CLBs are used to implement most of the logic in an FPGA. The principal CLB elements are shown in the simplified block diagram in Figure 2. There are three look-up tables (LUT) which are used as logic function generators, two flip-flops and two groups of signal steering multiplexers. There are also some more advanced features provided by the CLB which will be covered in the **Advanced Features Description**, page 13.

Function Generators

Two 16 x 1 memory look-up tables (F-LUT and G-LUT) are used to implement 4-input function generators, each offering unrestricted logic implementation of any Boolean function of up to four independent input signals (F1 to F4 or G1 to G4). Using memory look-up tables the propagation delay is independent of the function implemented.

A third 3-input function generator (H-LUT) can implement any Boolean function of its three inputs. Two of these inputs are controlled by programmable multiplexers (see box "A" of Figure 2). These inputs can come from the F-LUT or G-LUT outputs or from CLB inputs. The third input always comes from a CLB input. The CLB can, therefore, implement certain functions of up to nine inputs, like parity checking. The three LUTs in the CLB can also be combined to do any arbitrarily defined Boolean function of five inputs.

The register choice is made by placing the appropriate library symbol. For example, IFD is the basic input flip-flop (rising edge triggered), and ILD is the basic input latch (transparent-High). Variations with inverted clocks are also available. The clock signal inverter is also shown in Figure 5 on the CK line.

The Spartan family IOB data input path has a one-tap delay element: either the delay is inserted (default), or it is not. The Spartan-XL family IOB data input path has a two-tap delay element, with choices of a full delay, a partial delay, or no delay. The added delay guarantees a zero hold time with respect to clocks routed through the global clock buffers. (See **Global Nets and Buffers**, page 12 for a description of the global clock buffers in the Spartan/XL families.) For a shorter input register setup time, with positive hold-time, attach a NODELAY attribute or property to the flip-flop.The output of the input register goes to the routing channels (via I1 and I2 in Figure 6). The I1 and I2 signals that exit the IOB can each carry either the direct or registered input signal.

The 5V Spartan family input buffers can be globally configured for either TTL (1.2V) or CMOS (VCC/2) thresholds, using an option in the bitstream generation software. The Spartan family output levels are also configurable; the two global adjustments of input threshold and output level are independent. The inputs of Spartan devices can be driven by the outputs of any 3.3V device, if the Spartan family inputs are in TTL mode. Input and output thresholds are TTL on all configuration pins until the configuration has been loaded into the device and specifies how they are to be used. Spartan-XL family inputs are TTL compatible and 3.3V CMOS compatible.

Supported sources for Spartan/XL device inputs are shown in Table 4.

Spartan-XL family I/Os are fully 5V tolerant even though the V_{CC} is 3.3V. This allows 5V signals to directly connect to the Spartan-XL family inputs without damage, as shown in Table 4. In addition, the 3.3V V_{CC} can be applied before or after 5V signals are applied to the I/Os. This makes the Spartan-XL devices immune to power supply sequencing problems.

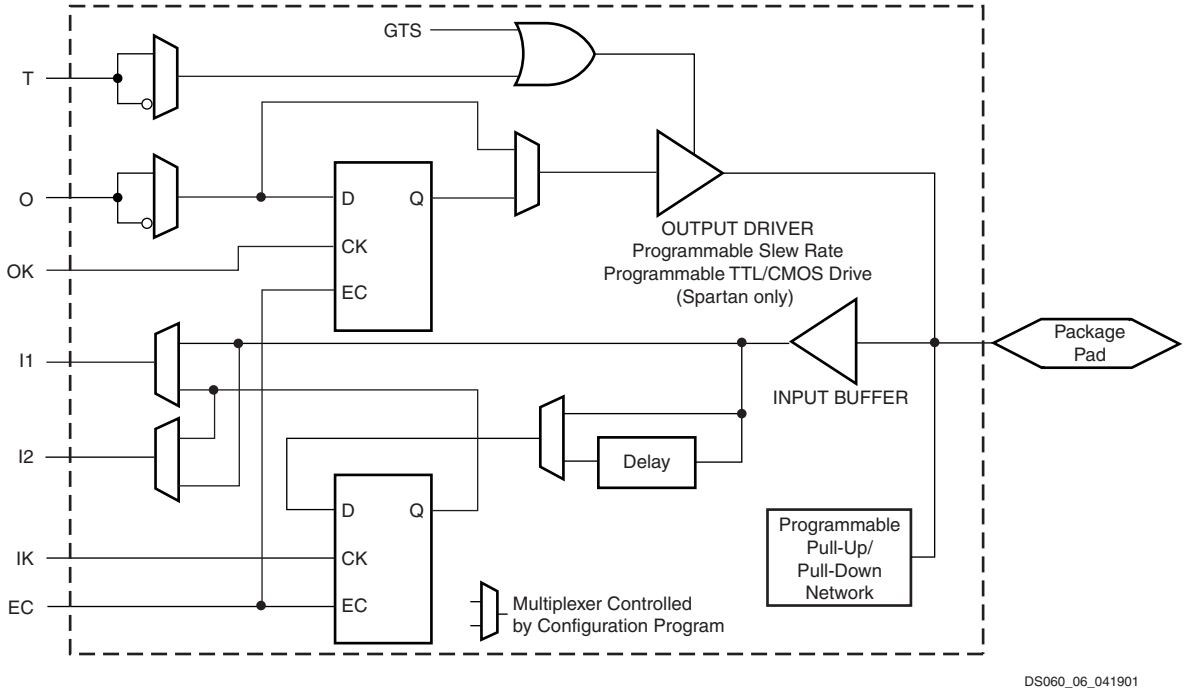

Figure 6: **Simplified Spartan/XL IOB Block Diagram**

This high value makes them unsuitable as wired-AND pull-up resistors.

Table 7: **Supported Destinations for Spartan/XL Outputs**

Destination	Spartan-XL Outputs	Spartan Outputs	
	3.3V, CMOS	5V, TTL	5V, CMOS
Any device, V_{CC} = 3.3V, CMOS-threshold inputs	√	√	Some(1)
Any device, V_{CC} = 5V, TTL-threshold inputs	√	√	√
Any device, V_{CC} = 5V, CMOS-threshold inputs	Unreliable Data		√

Notes:
1. Only if destination device has 5V tolerant inputs.

After configuration, voltage levels of unused pads, bonded or unbonded, must be valid logic levels, to reduce noise sensitivity and avoid excess current. Therefore, by default, unused pads are configured with the internal pull-up resistor active. Alternatively, they can be individually configured with the pull-down resistor, or as a driven output, or to be driven by an external source. To activate the internal pull-up, attach the PULLUP library component to the net attached to the pad. To activate the internal pull-down, attach the PULLDOWN library component to the net attached to the pad.

Set/Reset

As with the CLB registers, the GSR signal can be used to set or clear the input and output registers, depending on the value of the INIT attribute or property. The two flip-flops can be individually configured to set or clear on reset and after configuration. Other than the global GSR net, no user-controlled set/reset signal is available to the I/O flip-flops (Figure 5). The choice of set or reset applies to both the initial state of the flip-flop and the response to the GSR pulse.

Independent Clocks

Separate clock signals are provided for the input (IK) and output (OK) flip-flops. The clock can be independently inverted for each flip-flop within the IOB, generating either falling-edge or rising-edge triggered flip-flops. The clock inputs for each IOB are independent.

Common Clock Enables

The input and output flip-flops in each IOB have a common clock enable input (see EC signal in Figure 5), which through configuration, can be activated individually for the input or output flip-flop, or both. This clock enable operates exactly like the EC signal on the Spartan/XL FPGA CLB. It cannot be inverted within the IOB.

Routing Channel Description

All internal routing channels are composed of metal segments with programmable switching points and switching matrices to implement the desired routing. A structured, hierarchical matrix of routing channels is provided to achieve efficient automated routing.

This section describes the routing channels available in Spartan/XL devices. Figure 8 shows a general block diagram of the CLB routing channels. The implementation software automatically assigns the appropriate resources based on the density and timing requirements of the design. The following description of the routing channels is for information only and is simplified with some minor details omitted. For an exact interconnect description the designer should open a design in the FPGA Editor and review the actual connections in this tool.

The routing channels will be discussed as follows;

- CLB routing channels which run along each row and column of the CLB array.
- IOB routing channels which form a ring (called a VersaRing) around the outside of the CLB array. It connects the I/O with the CLB routing channels.
- Global routing consists of dedicated networks primarily designed to distribute clocks throughout the device with minimum delay and skew. Global routing can also be used for other high-fanout signals.

CLB Routing Channels

The routing channels around the CLB are derived from three types of interconnects; single-length, double-length, and longlines. At the intersection of each vertical and horizontal routing channel is a signal steering matrix called a Programmable Switch Matrix (PSM). Figure 8 shows the basic routing channel configuration showing single-length lines, double-length lines and longlines as well as the CLBs and PSMs. The CLB to routing channel interface is shown as well as how the PSMs interface at the channel intersections.

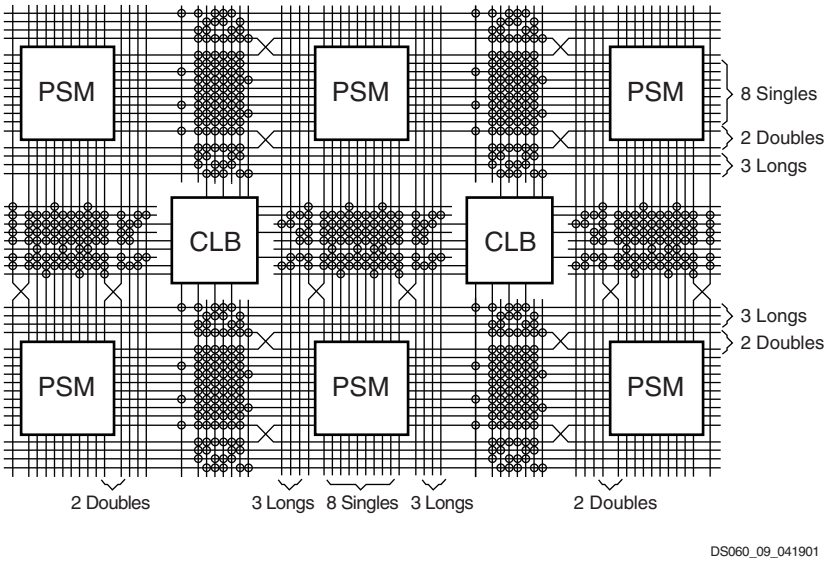

Figure 8: **Spartan/XL CLB Routing Channels and Interface Block Diagram**

CLB Interface

A block diagram of the CLB interface signals is shown in Figure 9. The input signals to the CLB are distributed evenly on all four sides providing maximum routing flexibility. In general, the entire architecture is symmetrical and regular. It is well suited to established placement and routing algorithms. Inputs, outputs, and function generators can freely swap positions within a CLB to avoid routing congestion during the placement and routing operation. The exceptions are the clock (K) input and CIN/COUT signals. The K input is routed to dedicated global vertical lines as well as four single-length lines and is on the left side of the CLB. The CIN/COUT signals are routed through dedicated interconnects which do not interfere with the general routing structure. The output signals from the CLB are available to drive both vertical and horizontal channels.

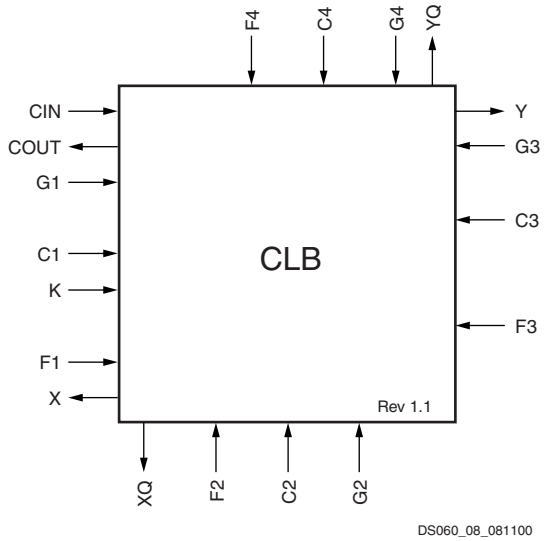

Figure 9: **CLB Interconnect Signals**

Programmable Switch Matrices

The horizontal and vertical single- and double-length lines intersect at a box called a programmable switch matrix (PSM). Each PSM consists of programmable pass transistors used to establish connections between the lines (see Figure 10).

For example, a single-length signal entering on the right side of the switch matrix can be routed to a single-length line on the top, left, or bottom sides, or any combination thereof, if multiple branches are required. Similarly, a double-length signal can be routed to a double-length line on any or all of the other three edges of the programmable switch matrix.

Single-Length Lines

Single-length lines provide the greatest interconnect flexibility and offer fast routing between adjacent blocks. There are eight vertical and eight horizontal single-length lines associated with each CLB. These lines connect the switching matrices that are located in every row and column of CLBs. Single-length lines are connected by way of the programmable switch matrices, as shown in Figure 10. Routing connectivity is shown in Figure 8.

Single-length lines incur a delay whenever they go through a PSM. Therefore, they are not suitable for routing signals for long distances. They are normally used to conduct signals within a localized area and to provide the branching for nets with fanout greater than one.

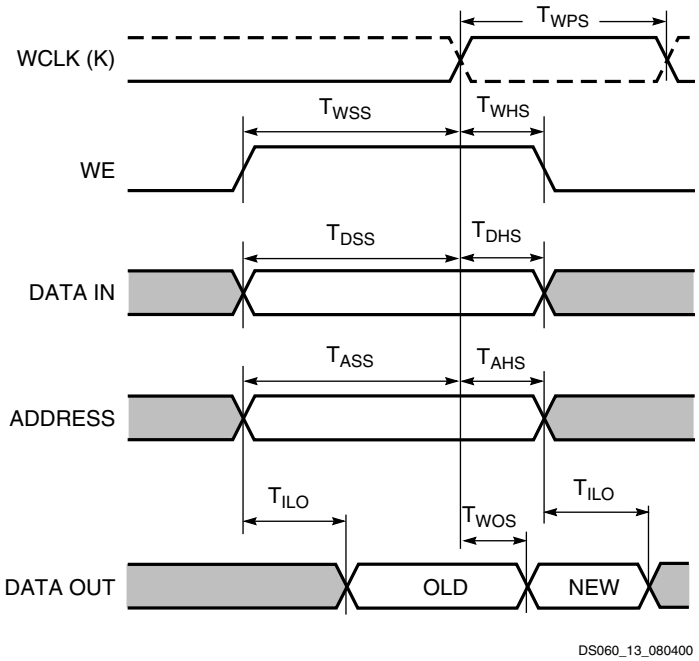

Figure 13: **Data Write and Access Timing for RAM**

WCLK can be configured as active on either the rising edge (default) or the falling edge. While the WCLK input to the RAM accepts the same signal as the clock input to the associated CLB's flip-flops, the sense of this WCLK input can be inverted with respect to the sense of the flip-flop clock inputs. Consequently, within the same CLB, data at the RAM SPO line can be stored in a flip-flop with either the same or the inverse clock polarity used to write data to the RAM.

The WE input is active High and cannot be inverted within the CLB.

Allowing for settling time, the data on the SPO output reflects the contents of the RAM location currently addressed. When the address changes, following the asynchronous delay T_{ILO}, the data stored at the new address location will appear on SPO. If the data at a particular RAM address is overwritten, after the delay T_{WOS}, the new data will appear on SPO.

Dual-Port Mode

In dual-port mode, the function generators (F-LUT and G-LUT) are used to create a 16 x 1 dual-port memory. Of the two data ports available, one permits read and write operations at the address specified by A[3:0] while the second provides only for read operations at the address specified independently by DPRA[3:0]. As a result, simultaneous read/write operations at different addresses (or even at the same address) are supported.

The functional organization of the 16 x 1 dual-port RAM is shown in Figure 14. The dual-port RAM signals and the

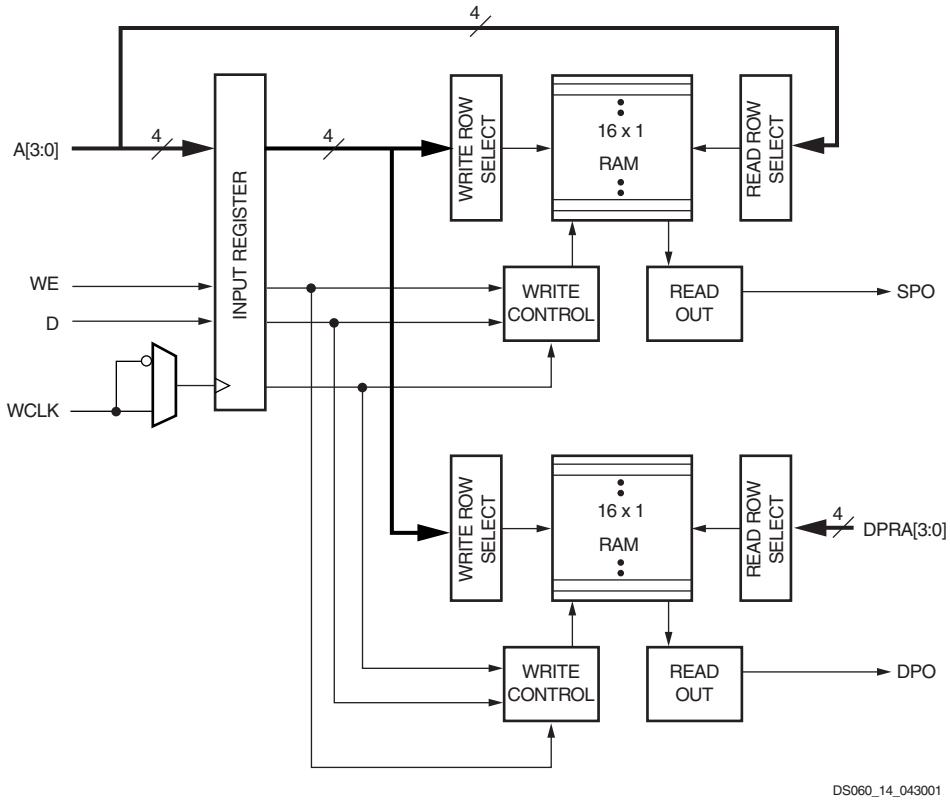

Figure 14: **Logic Diagram for the Dual-Port RAM**

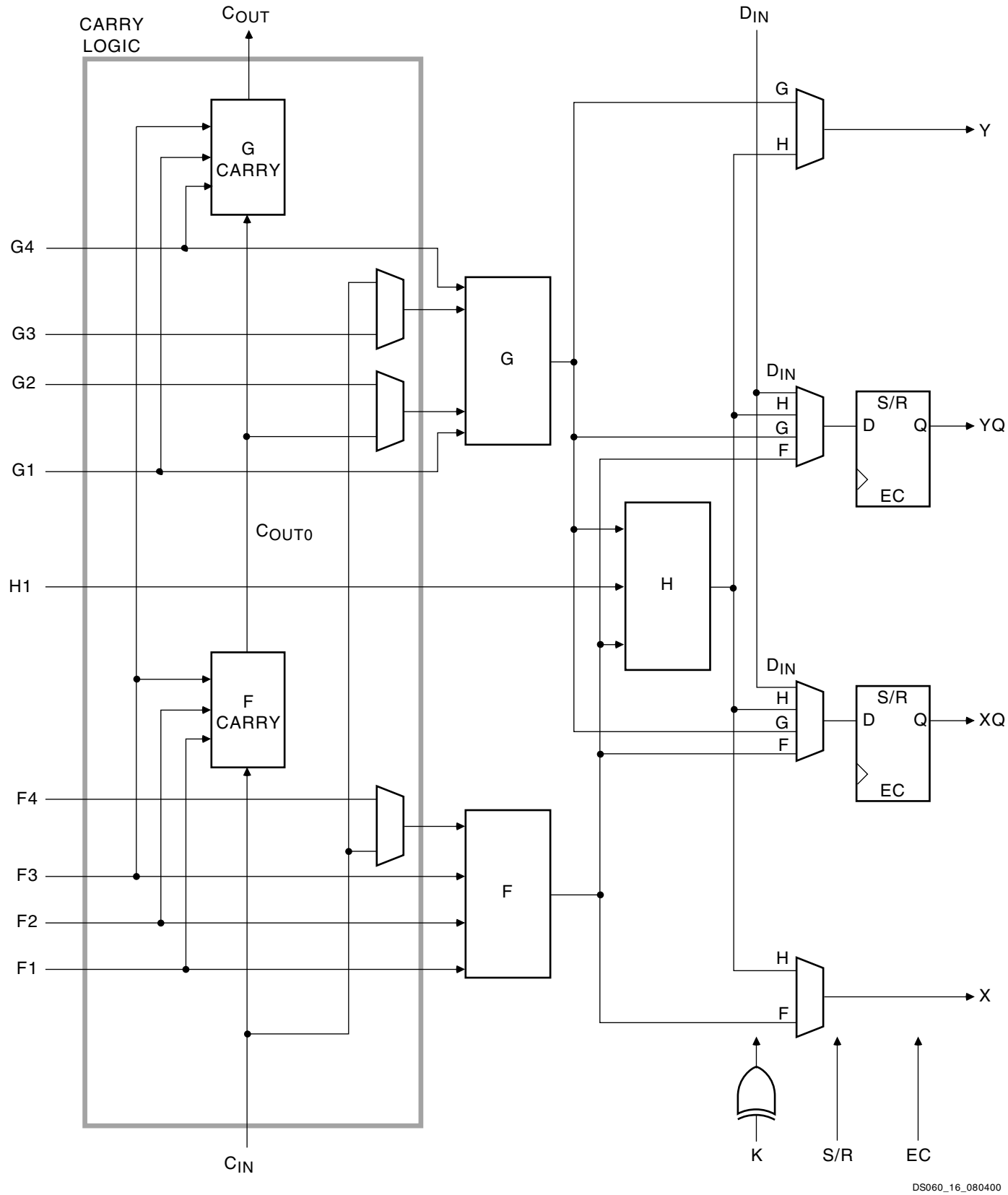

Figure 16: **Fast Carry Logic in Spartan/XL CLB**

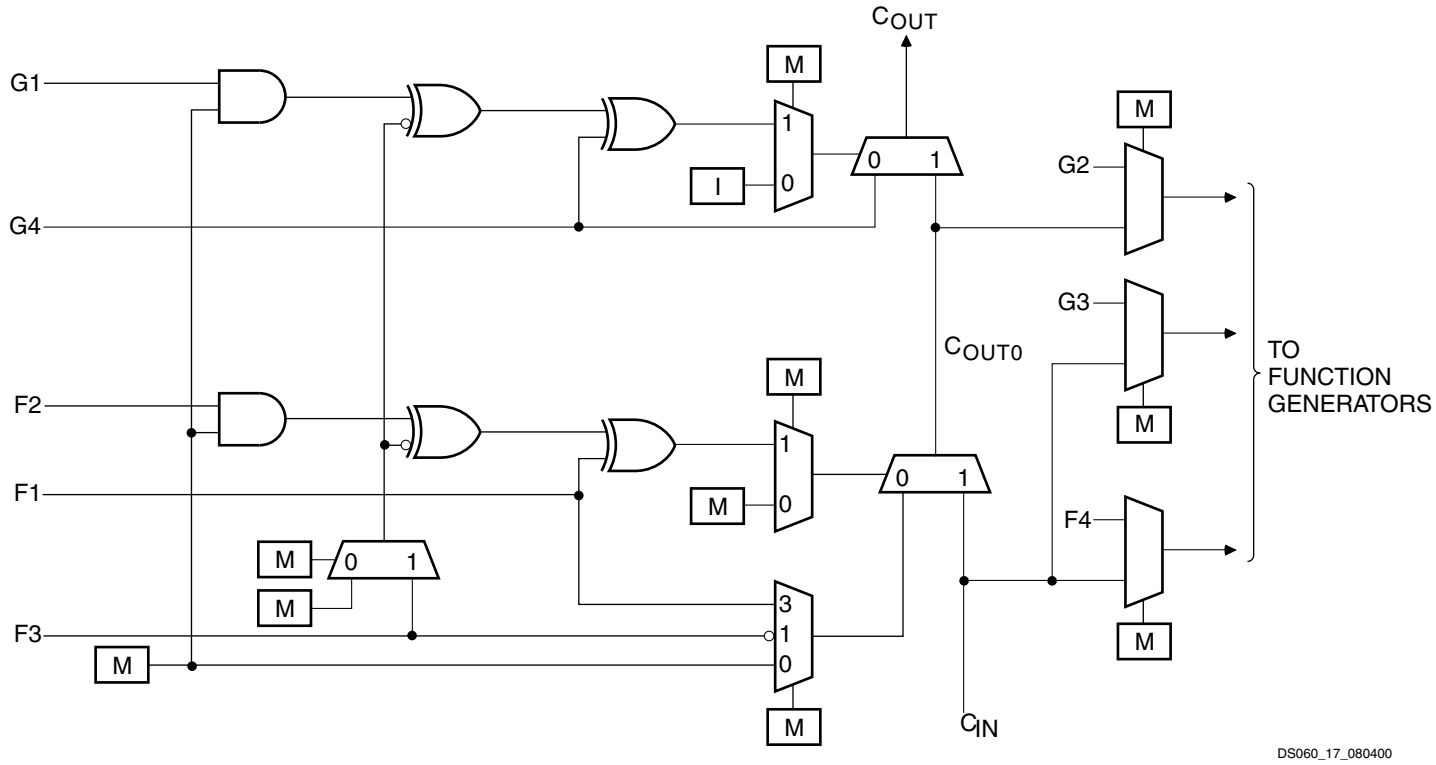

Figure 17: **Detail of Spartan/XL Dedicated Carry Logic**

3-State Long Line Drivers

A pair of 3-state buffers is associated with each CLB in the array. These 3-state buffers (BUFT) can be used to drive signals onto the nearest horizontal longlines above and below the CLB. They can therefore be used to implement multiplexed or bidirectional buses on the horizontal longlines, saving logic resources.

There is a weak keeper at each end of these two horizontal longlines. This circuit prevents undefined floating levels. However, it is overridden by any driver.

The buffer enable is an active High 3-state (i.e., an active Low enable), as shown in Table 11.

Three-State Buffer Example

Figure 18 shows how to use the 3-state buffers to implement a multiplexer. The selection is accomplished by the buffer 3-state signal.

Pay particular attention to the polarity of the T pin when using these buffers in a design. Active High 3-state (T) is identical to an active Low output enable, as shown in Table 11.

Table 11: **Three-State Buffer Functionality**

IN	T	OUT
X	1	Z
IN	0	IN

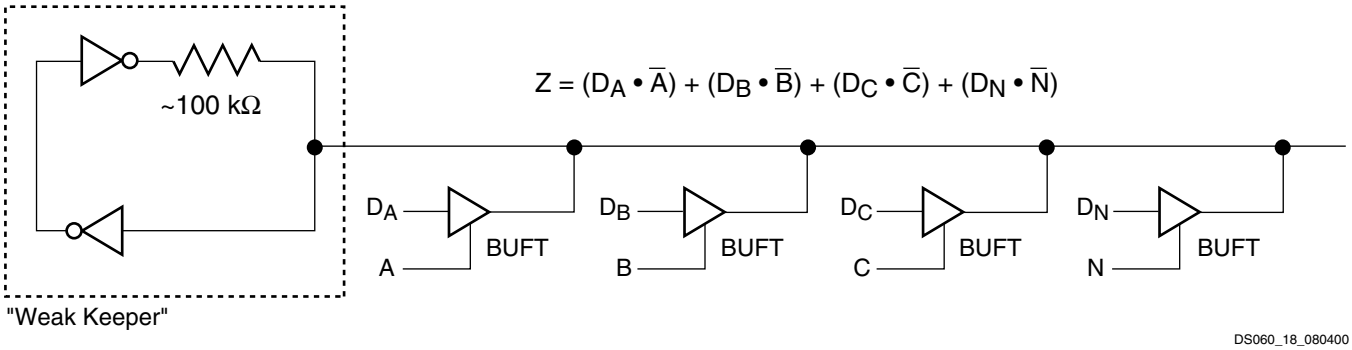

Figure 18: **3-state Buffers Implement a Multiplexer**

On-Chip Oscillator

Spartan/XL devices include an internal oscillator. This oscillator is used to clock the power-on time-out, for configuration memory clearing, and as the source of CCLK in Master configuration mode. The oscillator runs at a nominal 8 MHz frequency that varies with process, V_{CC}, and temperature. The output frequency falls between 4 MHz and 10 MHz.

The oscillator output is optionally available after configuration. Any two of four resynchronized taps of a built-in divider are also available. These taps are at the fourth, ninth, fourteenth and nineteenth bits of the divider. Therefore, if the primary oscillator output is running at the nominal 8 MHz, the user has access to an 8-MHz clock, plus any two of 500 kHz, 16 kHz, 490 Hz and 15 Hz. These frequencies can vary by as much as -50% or +25%.

These signals can be accessed by placing the OSC4 library element in a schematic or in HDL code. The oscillator is automatically disabled after configuration if the OSC4 symbol is not used in the design.

Global Signals: GSR and GTS

Global Set/Reset

A separate Global Set/Reset line, as shown in Figure 3, page 5 for the CLB and Figure 5, page 6 for the IOB, sets or clears each flip-flop during power-up, reconfiguration, or when a dedicated Reset net is driven active. This global net (GSR) does not compete with other routing resources; it uses a dedicated distribution network.

Each flip-flop is configured as either globally set or reset in the same way that the local set/reset (SR) is specified. Therefore, if a flip-flop is set by SR, it is also set by GSR. Similarly, if in reset mode, it is reset by both SR and GSR.

GSR can be driven from any user-programmable pin as a global reset input. To use this global net, place an input pad and input buffer in the schematic or HDL code, driving the GSR pin of the STARTUP symbol. (See Figure 19.) A specific pin location can be assigned to this input using a LOC attribute or property, just as with any other user-programmable pad. An inverter can optionally be inserted after the input buffer to invert the sense of the GSR signal. Alternatively, GSR can be driven from any internal node.

Global 3-State

A separate Global 3-state line (GTS) as shown in Figure 6, page 7 forces all FPGA outputs to the high-impedance state, unless boundary scan is enabled and is executing an EXTEST instruction. GTS does not compete with other routing resources; it uses a dedicated distribution network.

GTS can be driven from any user-programmable pin as a global 3-state input. To use this global net, place an input pad and input buffer in the schematic or HDL code, driving the GTS pin of the STARTUP symbol. This is similar to what is shown in Figure 19 for GSR except the IBUF would be connected to GTS. A specific pin location can be assigned to this input using a LOC attribute or property, just as with any other user-programmable pad. An inverter can optionally be inserted after the input buffer to invert the sense of the Global 3-state signal. Alternatively, GTS can be driven from any internal node.

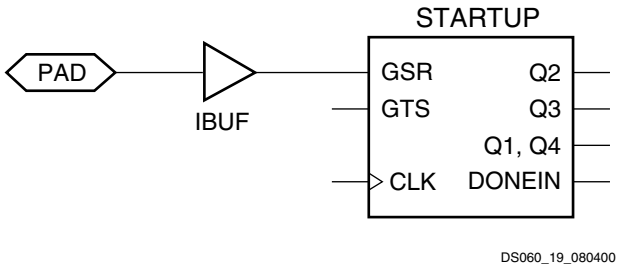

Figure 19: **Symbols for Global Set/Reset**

Boundary Scan

The "bed of nails" has been the traditional method of testing electronic assemblies. This approach has become less appropriate, due to closer pin spacing and more sophisticated assembly methods like surface-mount technology and multi-layer boards. The IEEE Boundary Scan Standard 1149.1 was developed to facilitate board-level testing of electronic assemblies. Design and test engineers can embed a standard test logic structure in their device to achieve high fault coverage for I/O and internal logic. This structure is easily implemented with a four-pin interface on any boundary scan compatible device. IEEE 1149.1-compatible devices may be serial daisy-chained together, connected in parallel, or a combination of the two.

The Spartan and Spartan-XL families implement IEEE 1149.1-compatible BYPASS, PRELOAD/SAMPLE and EXTEST boundary scan instructions. When the boundary scan configuration option is selected, three normal user I/O pins become dedicated inputs for these functions. Another user output pin becomes the dedicated boundary scan output. The details of how to enable this circuitry are covered later in this section.

By exercising these input signals, the user can serially load commands and data into these devices to control the driving of their outputs and to examine their inputs. This method is an improvement over bed-of-nails testing. It avoids the need to over-drive device outputs, and it reduces the user interface to four pins. An optional fifth pin, a reset for the control logic, is described in the standard but is not implemented in the Spartan/XL devices.

The dedicated on-chip logic implementing the IEEE 1149.1 functions includes a 16-state machine, an instruction register and a number of data registers. The functional details can be found in the IEEE 1149.1 specification and are also discussed in the Xilinx application note: "*Boundary Scan in FPGA Devices*."

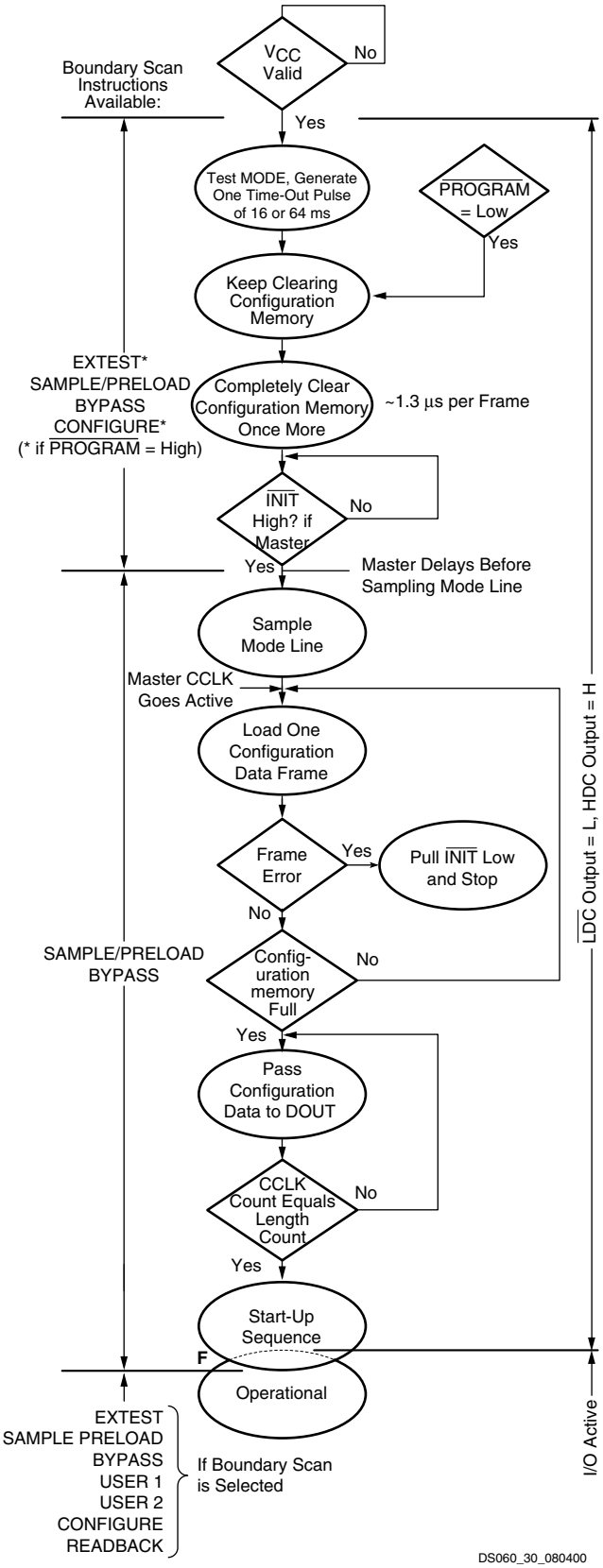

Figure 30: **Power-up Configuration Sequence**

Configuration

The 0010 preamble code indicates that the following 24 bits represent the length count for serial modes. The length count is the total number of configuration clocks needed to load the complete configuration data. (Four additional configuration clocks are required to complete the configuration process, as discussed below.) After the preamble and the length count have been passed through to any device in the daisy chain, its DOUT is held High to prevent frame start bits from reaching any daisy-chained devices. In Spartan-XL family Express mode, the length count bits are ignored, and DOUT is held Low, to disable the next device in the pseudo daisy chain.

A specific configuration bit, early in the first frame of a master device, controls the configuration-clock rate and can increase it by a factor of eight. Therefore, if a fast configuration clock is selected by the bitstream, the slower clock rate is used until this configuration bit is detected.

Each frame has a start field followed by the frame-configuration data bits and a frame error field. If a frame data error is detected, the FPGA halts loading, and signals the error by pulling the open-drain $\overline{\text{INIT}}$ pin Low. After all configuration frames have been loaded into an FPGA using a serial mode, DOUT again follows the input data so that the remaining data is passed on to the next device. In Spartan-XL family Express mode, when the first device is fully programmed, DOUT goes High to enable the next device in the chain.

Delaying Configuration After Power-Up

There are two methods of delaying configuration after power-up: put a logic Low on the $\overline{\text{PROGRAM}}$ input, or pull the bidirectional $\overline{\text{INIT}}$ pin Low, using an open-collector (open-drain) driver. (See Figure 30.)

A Low on the $\overline{\text{PROGRAM}}$ input is the more radical approach, and is recommended when the power-supply rise time is excessive or poorly defined. As long as $\overline{\text{PROGRAM}}$ is Low, the FPGA keeps clearing its configuration memory. When $\overline{\text{PROGRAM}}$ goes High, the configuration memory is cleared one more time, followed by the beginning of configuration, provided the $\overline{\text{INIT}}$ input is not externally held Low. Note that a Low on the $\overline{\text{PROGRAM}}$ input automatically forces a Low on the $\overline{\text{INIT}}$ output. The Spartan/XL FPGA $\overline{\text{PROGRAM}}$ pin has a permanent weak pull-up.

Avoid holding $\overline{\text{PROGRAM}}$ Low for more than 500 µs. The 500 µs maximum limit is only a recommendation, not a requirement. The only effect of holding $\overline{\text{PROGRAM}}$ Low for more than 500 µs is an increase in current, measured at about 40 mA in the XCS40XL. This increased current cannot damage the device. This applies only during reconfiguration, not during power-up. The $\overline{\text{INIT}}$ pin can also be held Low to delay reconfiguration, and the same characteristics apply as for the $\overline{\text{PROGRAM}}$ pin.

Using an open-collector or open-drain driver to hold $\overline{\text{INIT}}$ Low before the beginning of configuration causes the FPGA

to wait after completing the configuration memory clear operation. When $\overline{INIT}$ is no longer held Low externally, the device determines its configuration mode by capturing the state of the Mode pins, and is ready to start the configuration process. A master device waits up to an additional 300 µs to make sure that any slaves in the optional daisy chain have seen that $\overline{INIT}$ is High.

For more details on Configuration, refer to the Xilinx Application Note "FPGA Configuration Guidelines" (XAPP090).

Start-Up

Start-up is the transition from the configuration process to the intended user operation. This transition involves a change from one clock source to another, and a change from interfacing parallel or serial configuration data where most outputs are 3-stated, to normal operation with I/O pins active in the user system. Start-up must make sure that the user logic 'wakes up' gracefully, that the outputs become active without causing contention with the configuration signals, and that the internal flip-flops are released from the Global Set/Reset (GSR) at the right time.

Start-Up Initiation

Two conditions have to be met in order for the start-up sequence to begin:

- The chip's internal memory must be full, and
- The configuration length count must be met, exactly.

In all configuration modes except Express mode, Spartan/XL devices read the expected length count from the bitstream and store it in an internal register. The length count varies according to the number of devices and the composition of the daisy chain. Each device also counts the number of CCLKs during configuration.

In Express mode, there is no length count. The start-up sequence for each device begins when the device has received its quota of configuration data. Wiring the DONE pins of several devices together delays start-up of all devices until all are fully configured.

Start-Up Events

The device can be programmed to control three start-up events.

- The release of the open-drain DONE output
- The termination of the Global Three-State and the change of configuration-related pins to the user function, activating all IOBs.
- The termination of the Global Set/Reset initialization of all CLB and IOB storage elements.

Figure 31 describes start-up timing in detail. The three events — DONE going High, the internal GSR being de-activated, and the user I/O going active — can all occur in any arbitrary sequence. This relative timing is selected by options in the bitstream generation software. Heavy lines in Figure 31 show the default timing. The thin lines indicate all other possible timing options. The start-up logic must be clocked until the "F" (Finished) state is reached.

The default option, and the most practical one, is for DONE to go High first, disconnecting the configuration data source and avoiding any contention when the I/Os become active one clock later. GSR is then released another clock period later to make sure that user operation starts from stable internal conditions. This is the most common sequence, shown with heavy lines in Figure 31, but the designer can modify it to meet particular requirements.

Start-Up Clock

Normally, the start-up sequence is controlled by the internal device oscillator (CCLK), which is asynchronous to the system clock. As a configuration option, they can be triggered by an on-chip user net called UCLK. This user net can be accessed by placing the STARTUP library symbol, and the start-up modes are known as UCLK_NOSYNC or UCLK_SYNC. This allows the device to wake up in synchronism with the user system.

DONE Pin

Note that DONE is an open-drain output and does not go High unless an internal pull-up is activated or an external pull-up is attached. The internal pull-up is activated as the default by the bitstream generation software.

The DONE pin can also be wire-ANDed with DONE pins of other FPGAs or with other external signals, and can then be used as input to the start-up control logic. This is called "Start-up Timing Synchronous to Done In" and is selected by either CCLK_SYNC or UCLK_SYNC. When DONE is not used as an input, the operation is called "Start-up Timing Not Synchronous to DONE In," and is selected by either CCLK_NOSYNC or UCLK_NOSYNC. Express mode configuration always uses either CCLK_SYNC or UCLK_SYNC timing, while the other configuration modes can use any of the four timing sequences.

When the UCLK_SYNC option is enabled, the user can externally hold the open-drain DONE output Low, and thus stall all further progress in the start-up sequence until DONE is released and has gone High. This option can be used to force synchronization of several FPGAs to a common user clock, or to guarantee that all devices are successfully configured before any I/Os go active.

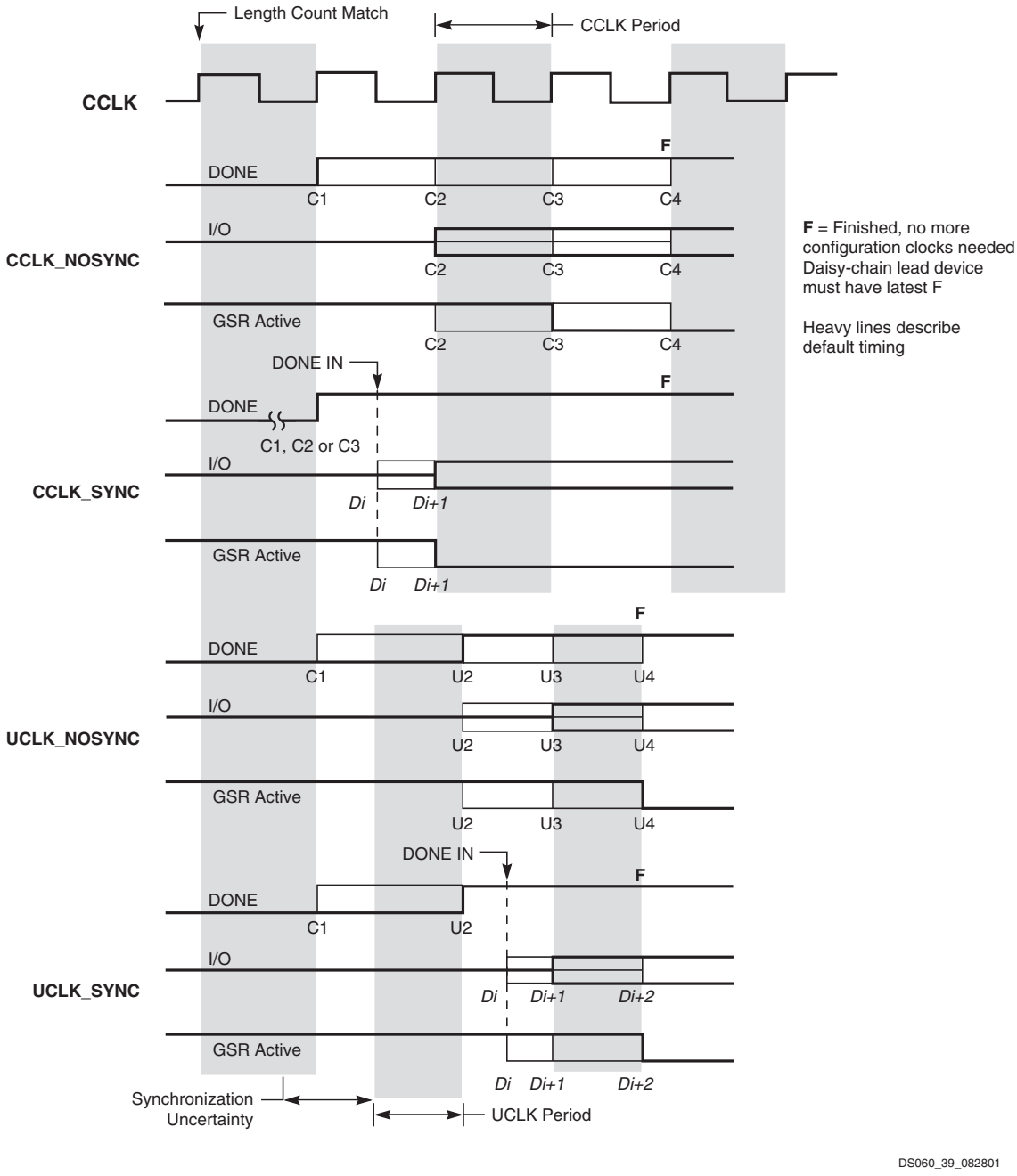

Figure 31: **Start-up Timing**

Configuration Through the Boundary Scan Pins

Spartan/XL devices can be configured through the boundary scan pins. The basic procedure is as follows:

- Power up the FPGA with $\overline{\text{INIT}}$ held Low (or drive the $\overline{\text{PROGRAM}}$ pin Low for more than 300 ns followed by a High while holding $\overline{\text{INIT}}$ Low). Holding $\overline{\text{INIT}}$ Low allows enough time to issue the CONFIG command to the FPGA. The pin can be used as I/O after configuration if a resistor is used to hold $\overline{\text{INIT}}$ Low.
- Issue the CONFIG command to the TMS input.
- Wait for $\overline{\text{INIT}}$ to go High.
- Sequence the boundary scan Test Access Port to the SHIFT-DR state.
- Toggle TCK to clock data into TDI pin.

The user must account for all TCK clock cycles after $\overline{\text{INIT}}$ goes High, as all of these cycles affect the Length Count compare.

For more detailed information, refer to the Xilinx application note, "*Boundary Scan in FPGA Devices*." This application note applies to Spartan and Spartan-XL devices.

Spartan-XL Family CLB Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values apply to all Spartan-XL devices and expressed in nanoseconds unless otherwise noted.

Symbol	Description	Speed Grade				Units
		-5		-4		
		Min	Max	Min	Max	
Clocks						
T_{CH}	Clock High time	2.0	-	2.3	-	ns
T_{CL}	Clock Low time	2.0	-	2.3	-	ns
Combinatorial Delays						
T_{ILO}	F/G inputs to X/Y outputs	-	1.0	-	1.1	ns
T_{IHO}	F/G inputs via H to X/Y outputs	-	1.7	-	2.0	ns
T_{ITO}	F/G inputs via transparent latch to Q outputs	-	1.5	-	1.8	ns
T_{HH1O}	C inputs via H1 via H to X/Y outputs	-	1.5	-	1.8	ns
Sequential Delays						
T_{CKO}	Clock K to Flip-Flop or latch outputs Q	-	1.2	-	1.4	ns
Setup Time before Clock K						
T_{ICK}	F/G inputs	0.6	-	0.7	-	ns
T_{IHCK}	F/G inputs via H	1.3	-	1.6	-	ns
Hold Time after Clock K						
	All Hold times, all devices	0.0	-	0.0	-	ns
Set/Reset Direct						
T_{RPW}	Width (High)	2.5	-	2.8	-	ns
T_{RIO}	Delay from C inputs via S/R, going High to Q	-	2.3	-	2.7	ns
Global Set/Reset						
T_{MRW}	Minimum GSR Pulse Width	10.5	-	11.5	-	ns
T_{MRQ}	Delay from GSR input to any Q	See page 60 for T_{RRI} values per device.				
F_{TOG}	Toggle Frequency (MHz) (for export control purposes)	-	250	-	217	MHz

Spartan-XL Family IOB Input Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. These path delays, provided as a guideline, have been extracted from the static timing analyzer report. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature).

Symbol	Description	Device	Speed Grade				Units
			-5		-4		
			Min	Max	Min	Max	
Setup Times							
T_{ECIK}	Clock Enable (EC) to Clock (IK)	All devices	0.0	-	0.0	-	ns
T_{PICK}	Pad to Clock (IK), no delay	All devices	1.0	-	1.2	-	ns
T_{POCK}	Pad to Fast Capture Latch Enable (OK), no delay	All devices	0.7	-	0.8	-	ns
Hold Times							
	All Hold Times	All devices	0.0	-	0.0	-	ns
Propagation Delays							
T_{PID}	Pad to I1, I2	All devices	-	0.9	-	1.1	ns
T_{PLI}	Pad to I1, I2 via transparent input latch, no delay	All devices	-	2.1	-	2.5	ns
T_{IKRI}	Clock (IK) to I1, I2 (flip-flop)	All devices	-	1.0	-	1.1	ns
T_{IKLI}	Clock (IK) to I1, I2 (latch enable, active Low)	All devices	-	1.1	-	1.2	ns
Delay Adder for Input with Full Delay Option							
T_{Delay}	$T_{PICKD} = T_{PICK} + T_{Delay}$ $T_{PDLI} = T_{PLI} + T_{Delay}$	XCS05XL	4.0	-	4.7	-	ns
		XCS10XL	4.8	-	5.6	-	ns
		XCS20XL	5.0	-	5.9	-	ns
		XCS30XL	5.5	-	6.5	-	ns
		XCS40XL	6.5	-	7.6	-	ns
Global Set/Reset							
T_{MRW}	Minimum GSR pulse width	All devices	10.5	-	11.5	-	ns
T_{RRI}	Delay from GSR input to any Q	XCS05XL	-	9.0	-	10.5	ns
		XCS10XL	-	9.5	-	11.0	ns
		XCS20XL	-	10.0	-	11.5	ns
		XCS30XL	-	11.0	-	12.5	ns
		XCS40XL	-	12.0	-	13.5	ns

Notes:

1. Input pad setup and hold times are specified with respect to the internal clock (IK). For setup and hold times with respect to the clock input, see the pin-to-pin parameters in the Pin-to-Pin Input Parameters table.
2. Voltage levels of unused pads, bonded or unbonded, must be valid logic levels. Each can be configured with the internal pull-up (default) or pull-down resistor, or configured as a driven output, or can be driven from an external source.

Pin Descriptions

There are three types of pins in the Spartan/XL devices:

- Permanently dedicated pins
- User I/O pins that can have special functions
- Unrestricted user-programmable I/O pins.

Before and during configuration, all outputs not used for the configuration process are 3-stated with the I/O pull-up resistor network activated. After configuration, if an IOB is unused it is configured as an input with the I/O pull-up resistor network remaining activated.

Any user I/O can be configured to drive the Global Set/Reset net GSR or the global three-state net GTS. See **Global Signals: GSR and GTS**, page 20 for more information.

Device pins for Spartan/XL devices are described in Table 18.

Some Spartan-XL devices are available in Pb-free package options. The Pb-free package options have the same pinouts as the standard package options.

Table 18: **Pin Descriptions**

Pin Name	I/O During Config.	I/O After Config.	Pin Description
Permanently Dedicated Pins			
V_{CC}	X	X	Eight or more (depending on package) connections to the nominal +5V supply voltage (+3.3V for Spartan-XL devices). All must be connected, and each must be decoupled with a 0.01 –0.1 µF capacitor to Ground.
GND	X	X	Eight or more (depending on package type) connections to Ground. All must be connected.
CCLK	I or O	I	During configuration, Configuration Clock (CCLK) is an output in Master mode and is an input in Slave mode. After configuration, CCLK has a weak pull-up resistor and can be selected as the Readback Clock. There is no CCLK High or Low time restriction on Spartan/XL devices, except during Readback. See **Violating the Maximum High and Low Time Specification for the Readback Clock**, page 39 for an explanation of this exception.
DONE	I/O	O	DONE is a bidirectional signal with an optional internal pull-up resistor. As an open-drain output, it indicates the completion of the configuration process. As an input, a Low level on DONE can be configured to delay the global logic initialization and the enabling of outputs. The optional pull-up resistor is selected as an option in the program that creates the configuration bitstream. The resistor is included by default.
$\overline{\text{PROGRAM}}$	I	I	$\overline{\text{PROGRAM}}$ is an active Low input that forces the FPGA to clear its configuration memory. It is used to initiate a configuration cycle. When $\overline{\text{PROGRAM}}$ goes High, the FPGA finishes the current clear cycle and executes another complete clear cycle, before it goes into a WAIT state and releases $\overline{\text{INIT}}$. The $\overline{\text{PROGRAM}}$ pin has a permanent weak pull-up, so it need not be externally pulled up to VCC.
MODE (Spartan) M0, M1 (Spartan-XL)	I	X	The Mode input(s) are sampled after $\overline{\text{INIT}}$ goes High to determine the configuration mode to be used. During configuration, these pins have a weak pull-up resistor. For the most popular configuration mode, Slave Serial, the mode pins can be left unconnected. For Master Serial mode, connect the Mode/M0 pin directly to system ground.

Device-Specific Pinout Tables

Device-specific tables include all packages for each Spartan and Spartan-XL device. They follow the pad locations around the die, and include boundary scan register locations.

Some Spartan-XL devices are available in Pb-free package options. The Pb-free package options have the same pinouts as the standard package options.

XCS05 and XCS05XL Device Pinouts

XCS05/XL Pad Name	PC84(4)	VQ100	Bndry Scan
VCC	P2	P89	-
I/O	P3	P90	32
I/O	P4	P91	35
I/O	-	P92	38
I/O	-	P93	41
I/O	P5	P94	44
I/O	P6	P95	47
I/O	P7	P96	50
I/O	P8	P97	53
I/O	P9	P98	56
I/O, SGCK1(1), GCK8(2)	P10	P99	59
VCC	P11	P100	-
GND	P12	P1	-
I/O, PGCK1(1), GCK1(2)	P13	P2	62
I/O	P14	P3	65
I/O, TDI	P15	P4	68
I/O, TCK	P16	P5	71
I/O, TMS	P17	P6	74
I/O	P18	P7	77
I/O	-	P8	83
I/O	P19	P9	86
I/O	P20	P10	89
GND	P21	P11	-
VCC	P22	P12	-
I/O	P23	P13	92
I/O	P24	P14	95
I/O	-	P15	98
I/O	P25	P16	104
I/O	P26	P17	107
I/O	P27	P18	110
I/O	-	P19	113
I/O	P28	P20	116
I/O, SGCK2(1), GCK2(2)	P29	P21	119
Not Connected(1), M1(2)	P30	P22	122
GND	P31	P23	-
MODE(1), M0 (2)	P32	P24	125
VCC	P33	P25	-

XCS05 and XCS05XL Device Pinouts

XCS05/XL Pad Name	PC84(4)	VQ100	Bndry Scan
Not Connected(1), $\overline{\text{PWRDWN}}$(2)	P34	P26	126(1)
I/O, PGCK2(1), GCK3 (2)	P35	P27	127(3)
I/O (HDC)	P36	P28	130(3)
I/O	-	P29	133(3)
I/O ($\overline{\text{LDC}}$)	P37	P30	136(3)
I/O	P38	P31	139(3)
I/O	P39	P32	142(3)
I/O	-	P33	145(3)
I/O	-	P34	148(3)
I/O	P40	P35	151(3)
I/O ($\overline{\text{INIT}}$)	P41	P36	154(3)
VCC	P42	P37	-
GND	P43	P38	-
I/O	P44	P39	157(3)
I/O	P45	P40	160(3)
I/O	-	P41	163(3)
I/O	-	P42	166(3)
I/O	P46	P43	169(3)
I/O	P47	P44	172(3)
I/O	P48	P45	175(3)
I/O	P49	P46	178(3)
I/O	P50	P47	181(3)
I/O, SGCK3(1), GCK4(2)	P51	P48	184(3)
GND	P52	P49	-
DONE	P53	P50	-
VCC	P54	P51	-
$\overline{\text{PROGRAM}}$	P55	P52	-
I/O (D7(2))	P56	P53	187(3)
I/O, PGCK3(1), GCK5(2)	P57	P54	190(3)
I/O (D6(2))	P58	P55	193(3)
I/O	-	P56	196(3)
I/O (D5(2))	P59	P57	199(3)
I/O	P60	P58	202(3)
I/O	-	P59	205(3)
I/O	-	P60	208(3)
I/O (D4(2))	P61	P61	211(3)
I/O	P62	P62	214(3)
VCC	P63	P63	-
GND	P64	P64	-
I/O (D3(2))	P65	P65	217(3)
I/O	P66	P66	220(3)
I/O	-	P67	223(3)
I/O (D2(2))	P67	P68	229(3)
I/O	P68	P69	232(3)
I/O (D1(2))	P69	P70	235(3)

XCS10 and XCS10XL Device Pinouts

XCS10/XL Pad Name	PC84[4]	VQ100	CS144[2,4]	TQ144	Bndry Scan
I/O	P80	P81	A10	P116	17
GND	-	-	C9	P118	-
I/O	-	-	B9	P119	20
I/O	-	-	A9	P120	23
I/O	P81	P82	D8	P121	26
I/O	P82	P83	C8	P122	29
I/O	-	P84	B8	P123	32
I/O	-	P85	A8	P124	35
I/O	P83	P86	B7	P125	38
I/O	P84	P87	A7	P126	41
GND	P1	P88	C7	P127	-

Notes:

1. 5V Spartan family only
2. 3V Spartan-XL family only
3. The “PWRDWN” on the XCS10XL is not part of the Boundary Scan chain. For the XCS10XL, subtract 1 from all Boundary Scan numbers from GCK3 on (175 and higher).
4. PC84 and CS144 packages discontinued by PDN2004-01

Additional XCS10/XL Package Pins

TQ144					
Not Connected Pins					
P117	-	-	-	-	-
5/5/97					

CS144					
Not Connected Pins					
D9	-	-	-	-	-
4/28/99					

XCS20 and XCS20XL Device Pinouts

XCS20/XL Pad Name	VQ100	CS144[2,4]	TQ144	PQ208	Bndry Scan
VCC	P89	D7	P128	P183	-
I/O	P90	A6	P129	P184	62
I/O	P91	B6	P130	P185	65
I/O	P92	C6	P131	P186	68
I/O	P93	D6	P132	P187	71
I/O	-	-	-	P188	74
I/O	-	-	-	P189	77
I/O	P94	A5	P133	P190	80
I/O	P95	B5	P134	P191	83
VCC[2]	-	-	-	P192	-
I/O	-	C5	P135	P193	86
I/O	-	D5	P136	P194	89
GND	-	A4	P137	P195	-
I/O	-	-	-	P196	92
I/O	-	-	-	P197	95
I/O	-	-	-	P198	98
I/O	-	-	-	P199	101
I/O	P96	B4	P138	P200	104
I/O	P97	C4	P139	P201	107
I/O	-	A3	P140	P204	110
I/O	-	B3	P141	P205	113
I/O	P98	C3	P142	P206	116

XCS20 and XCS20XL Device Pinouts

XCS20/XL Pad Name	VQ100	CS144[2,4]	TQ144	PQ208	Bndry Scan
I/O, SGCK1[1], GCK8[2]	P99	A2	P143	P207	119
VCC	P100	B2	P144	P208	-
GND	P1	A1	P1	P1	-
I/O, PGCK1[1], GCK1[2]	P2	B1	P2	P2	122
I/O	P3	C2	P3	P3	125
I/O	-	C1	P4	P4	128
I/O	-	D4	P5	P5	131
I/O, TDI	P4	D3	P6	P6	134
I/O, TCK	P5	D2	P7	P7	137
I/O	-	-	-	P8	140
I/O	-	-	-	P9	143
I/O	-	-	-	P10	146
I/O	-	-	-	P11	149
GND	-	D1	P8	P13	-
I/O	-	E4	P9	P14	152
I/O	-	E3	P10	P15	155
I/O, TMS	P6	E2	P11	P16	158
I/O	P7	E1	P12	P17	161
VCC[2]	-	-	-	P18	-
I/O	-	-	-	P19	164
I/O	-	-	-	P20	167

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
GND	P25	P29	GND(4)	GND(4)	-
VCC	P26	P30	VCC(4)	VCC(4)	-
I/O	P27	P31	L2	K3	254
I/O	P28	P32	L3	K4	257
I/O	P29	P33	L4	K5	260
I/O	P30	P34	M1	L1	263
I/O	P31	P35	M2	L2	266
I/O	P32	P36	M3	L3	269
I/O	-	-	M4	L4	272
I/O	-	-	-	M1	275
I/O	-	P38	N1	M2	278
I/O	-	P39	N2	M3	281
VCC	P33	P40	VCC(4)	VCC(4)	-
I/O	P34	P41	P1	N1	284
I/O	P35	P42	P2	N2	287
I/O	P36	P43	R1	N3	290
I/O	P37	P44	P3	N4	293
GND	P38	P45	GND(4)	GND(4)	-
I/O	-	P46	T1	P1	296
I/O	P39	P47	R3	P2	299
I/O	P40	P48	T2	P3	302
I/O	P41	P49	U1	P4	305
I/O	P42	P50	T3	P5	308
I/O	P43	P51	U2	R1	311
I/O	-	-	-	R2	314
I/O	-	-	-	R4	317
I/O	P44	P52	V1	T1	320
I/O	P45	P53	T4	T2	323
I/O	P46	P54	U3	T3	326
I/O	P47	P55	V2	U1	329
I/O	P48	P56	W1	V1	332
I/O, SGCK2(1), GCK2 (2)	P49	P57	V3	U2	335
Not Connected(1) M1(2)	P50	P58	W2	V2	338
GND	P51	P59	GND(4)	GND(4)	-
MODE(1), M0(2)	P52	P60	Y1	W1	341
VCC	P53	P61	VCC(4)	VCC(4)	-
Not Connected(1) $\overline{\text{PWRDWN}}$(2)	P54	P62	W3	V3	342(1)
I/O, PGCK2(1), GCK3(2)	P55	P63	Y2	W2	343 (3)

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
I/O (HDC)	P56	P64	W4	W3	346 (3)
I/O	P57	P65	V4	T4	349 (3)
I/O	P58	P66	U5	U4	352 (3)
I/O	P59	P67	Y3	V4	355 (3)
I/O ($\overline{\text{LDC}}$)	P60	P68	Y4	W4	358 (3)
I/O	-	-	-	R5	361 (3)
I/O	-	-	-	U5	364 (3)
I/O	P61	P69	V5	T5	367 (3)
I/O	P62	P70	W5	W5	370 (3)
I/O	P63	P71	Y5	R6	373 (3)
I/O	P64	P72	V6	U6	376 (3)
I/O	P65	P73	W6	V6	379 (3)
I/O	-	P74	Y6	T6	382 (3)
GND	P66	P75	GND(4)	GND(4)	-
I/O	P67	P76	W7	W6	385 (3)
I/O	P68	P77	Y7	U7	388 (3)
I/O	P69	P78	V8	V7	391 (3)
I/O	P70	P79	W8	W7	394 (3)
VCC	P71	P80	VCC(4)	VCC(4)	-
I/O	P72	P81	Y8	W8	397 (3)
I/O	P73	P82	U9	U8	400 (3)
I/O	-	-	V9	V8	403 (3)
I/O	-	-	W9	T8	406 (3)
I/O	-	P84	Y9	W9	409 (3)
I/O	-	P85	W10	V9	412 (3)
I/O	P74	P86	V10	U9	415 (3)
I/O	P75	P87	Y10	T9	418 (3)
I/O	P76	P88	Y11	W10	421 (3)
I/O ($\overline{\text{INIT}}$)	P77	P89	W11	V10	424 (3)
VCC	P78	P90	VCC(4)	VCC(4)	VCC(4)
GND	P79	P91	GND(4)	GND(4)	-
I/O	P80	P92	V11	T10	427 (3)
I/O	P81	P93	U11	R10	430 (3)
I/O	P82	P94	Y12	W11	433 (3)
I/O	P83	P95	W12	V11	436 (3)
I/O	P84	P96	V12	U11	439 (3)
I/O	P85	P97	U12	T11	442 (3)
I/O	-	-	Y13	W12	445 (3)
I/O	-	-	W13	V12	448 (3)
I/O	-	P99	V13	U12	451 (3)
I/O	-	P100	Y14	T12	454 (3)
VCC	P86	P101	VCC(4)	VCC(4)	-
I/O	P87	P102	Y15	V13	457 (3)
I/O	P88	P103	V14	U13	460 (3)
I/O	P89	P104	W15	T13	463 (3)

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280[2,5]	Bndry Scan
O, TDO	P157	P181	A19	B17	0
GND	P158	P182	GND[4]	GND[4]	-
I/O	P159	P183	B18	A18	2
I/O, PGCK4[1], GCK7[2]	P160	P184	B17	A17	5
I/O	P161	P185	C17	D16	8
I/O	P162	P186	D16	C16	11
I/O (CS1[2])	P163	P187	A18	B16	14
I/O	P164	P188	A17	A16	17
I/O	-	-	-	E15	20
I/O	-	-	-	C15	23
I/O	P165	P189	C16	D15	26
I/O	-	P190	B16	A15	29
I/O	P166	P191	A16	E14	32
I/O	P167	P192	C15	C14	35
I/O	P168	P193	B15	B14	38
I/O	P169	P194	A15	D14	41
GND	P170	P196	GND[4]	GND[4]	-
I/O	P171	P197	B14	A14	44
I/O	P172	P198	A14	C13	47
I/O	-	P199	C13	B13	50
I/O	-	P200	B13	A13	53
VCC	P173	P201	VCC[4]	VCC[4]	-
I/O	-	-	A13	A12	56
I/O	-	-	D12	C12	59
I/O	P174	P202	C12	B12	62
I/O	P175	P203	B12	D12	65
I/O	P176	P205	A12	A11	68
I/O	P177	P206	B11	B11	71
I/O	P178	P207	C11	C11	74
I/O	P179	P208	A11	D11	77
I/O	P180	P209	A10	A10	80
I/O	P181	P210	B10	B10	83
GND	P182	P211	GND[4]	GND[4]	-

2/8/00

Notes:

1. 5V Spartan family only
2. 3V Spartan-XL family only
3. The “PWRDWN” on the XCS40XL is not part of the Boundary Scan chain. For the XCS40XL, subtract 1 from all Boundary Scan numbers from GCK3 on (343 and higher).
4. Pads labeled GND[4] or V_{CC}[4] are internally bonded to Ground or V_{CC} planes within the package.
5. CS280 package discontinued by PDN2004-01

Additional XCS40/XL Package Pins

PQ240

GND Pins					
P22	P37	P83	P98	P143	P158
P204	P219	-	-	-	-
Not Connected Pins					
P195	-	-	-	-	-

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BG256

VCC Pins					
C14	D6	D7	D11	D14	D15
E20	F1	F4	F17	G4	G17
K4	L17	P4	P17	P19	R2
R4	R17	U6	U7	U10	U14
U15	V7	W20	-	-	-
GND Pins					
A1	B7	D4	D8	D13	D17
G20	H4	H17	N3	N4	N17
U4	U8	U13	U17	W14	-

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CS280

VCC Pins					
A1	A7	B5	B15	C10	C17
D13	E3	E18	G1	G19	K2
K17	M4	N16	R3	R18	T7
U3	U10	U17	V5	V15	W13
GND Pins					
E5	E7	E8	E9	E11	E12
E13	G5	G15	H5	H15	J5
J15	L5	L15	M5	M15	N5
N15	R7	R8	R9	R11	R12
R13	-	-	-	-	-

5/19/99

Table 20: **User I/O Chart for Spartan/XL FPGAs**

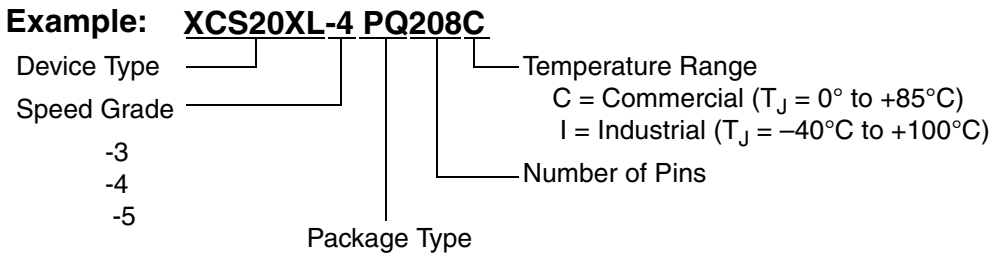
Device	Max I/O	Package Type							
		PC84(1)	VQ100(1)	CS144(1)	TQ144	PQ208	PQ240	BG256(1)	CS280(1)
XCS05	80	61(1)	77	-	-	-	-	-	-
XCS10	112	61(1)	77	-	112	-	-	-	-
XCS20	160	-	77	-	113	160	-	-	-
XCS30	192	-	77(1)	-	113	169	192	192(1)	-
XCS40	224	-	-	-	-	169	192	205	-
XCS05XL	80	61(1)	77(2)	-	-	-	-	-	-
XCS10XL	112	61(1)	77(2)	112(1)	112(2)	-	-	-	-
XCS20XL	160	-	77(2)	113(1)	113(2)	160(2)	-	-	-
XCS30XL	192	-	77(2)	-	113(2)	169(2)	192(2)	192(2)	192(1)
XCS40XL	224	-	-	-	-	169(2)	192(2)	205(2)	224(1)

6/25/08

Notes:

1. PC84, CS144, and CS280 packages, and VQ100 and BG256 packages for XCS30 only, discontinued by PDN2004-01
2. These Spartan-XL devices are available in Pb-free package options. The Pb-free packages insert a "G" in the package code. Contact Xilinx for availability.

Ordering Information



BG = Ball Grid Array
BGG = Ball Grid Array (Pb-free)
PC = Plastic Lead Chip Carrier
PQ = Plastic Quad Flat Pack
PQG = Plastic Quad Flat Pack (Pb-free)
VQ = Very Thin Quad Flat Pack
VQG = Very Thin Quad Flat Pack (Pb-free)
TQ = Thin Quad Flat Pack
TQG = Thin Quad Flat Pack (Pb-free)
CS = Chip Scale