AMD Xilinx - XCS30XL-4PQG208C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	576
Number of Logic Elements/Cells	1368
Total RAM Bits	18432
Number of I/O	169
Number of Gates	30000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcs30xl-4pqg208c

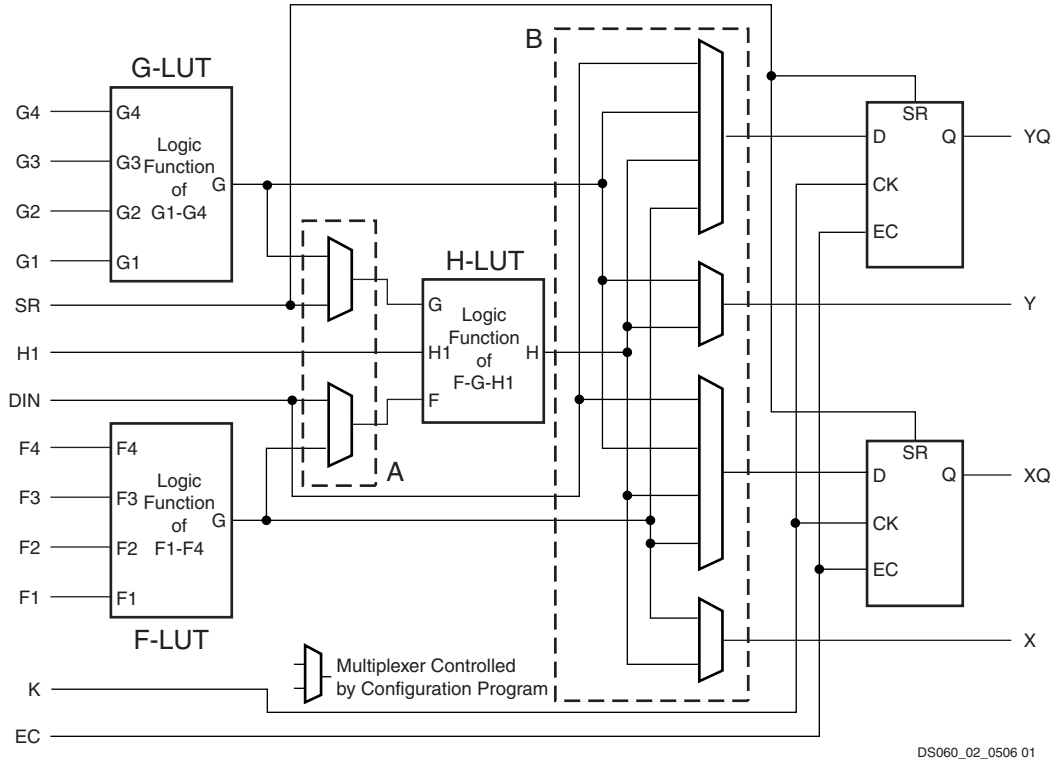


Figure 2: **Spartan/XL Simplified CLB Logic Diagram (some features not shown)**

A CLB can implement any of the following functions:

- Any function of up to four variables, plus any second function of up to four unrelated variables, plus any third function of up to three unrelated variables

  Note: When three separate functions are generated, one of the function outputs must be captured in a flip-flop internal to the CLB. Only two unregistered function generator outputs are available from the CLB.
- Any single function of five variables
- Any function of four variables together with some functions of six variables
- Some functions of up to nine variables.

Implementing wide functions in a single block reduces both the number of blocks required and the delay in the signal path, achieving both increased capacity and speed.

The versatility of the CLB function generators significantly improves system speed. In addition, the design-software tools can deal with each function generator independently. This flexibility improves cell usage.

Flip-Flops

Each CLB contains two flip-flops that can be used to register (store) the function generator outputs. The flip-flops and function generators can also be used independently (see Figure 2). The CLB input DIN can be used as a direct input to either of the two flip-flops. H1 can also drive either flip-flop via the H-LUT with a slight additional delay.

The two flip-flops have common clock (CK), clock enable (EC) and set/reset (SR) inputs. Internally both flip-flops are also controlled by a global initialization signal (GSR) which is described in detail in **Global Signals: GSR and GTS**, page 20.

Latches (Spartan-XL Family Only)

The Spartan-XL family CLB storage elements can also be configured as latches. The two latches have common clock (K) and clock enable (EC) inputs. Functionality of the storage element is described in Table 2.

Table 4: **Supported Sources for Spartan/XL Inputs**

Source	Spartan Inputs		Spartan-XL Inputs
	5V, TTL	5V, CMOS	3.3V CMOS
Any device, V_{CC} = 3.3V, CMOS outputs	√	Unreliable Data	√
Spartan family, V_{CC} = 5V, TTL outputs	√		√
Any device, V_{CC} = 5V, TTL outputs ($V_{OH} \leq 3.7V$)	√		√
Any device, V_{CC} = 5V, CMOS outputs	√	√	√ (default mode)

Spartan-XL Family V_{CC} Clamping

Spartan-XL FPGAs have an optional clamping diode connected from each I/O to V_{CC}. When enabled they clamp ringing transients back to the 3.3V supply rail. This clamping action is required in 3.3V PCI applications. V_{CC} clamping is a global option affecting all I/O pins.

Spartan-XL devices are fully 5V TTL I/O compatible if V_{CC} clamping is not enabled. With V_{CC} clamping enabled, the Spartan-XL devices will begin to clamp input voltages to one diode voltage drop above V_{CC}. If enabled, TTL I/O compatibility is maintained but full 5V I/O tolerance is sacrificed. The user may select either 5V tolerance (default) or 3.3V PCI compatibility. In both cases negative voltage is clamped to one diode voltage drop below ground.

Spartan-XL devices are compatible with TTL, LVTTL, PCI 3V, PCI 5V and LVCMOS signalling. The various standards are illustrated in Table 5.

Table 5: **I/O Standards Supported by Spartan-XL FPGAs**

Signaling Standard	VCC Clamping	Output Drive	$V_{IH\ MAX}$	$V_{IH\ MIN}$	$V_{IL\ MAX}$	$V_{OH\ MIN}$	$V_{OL\ MAX}$
TTL	Not allowed	12/24 mA	5.5	2.0	0.8	2.4	0.4
LVTTL	OK	12/24 mA	3.6	2.0	0.8	2.4	0.4
PCI5V	Not allowed	24 mA	5.5	2.0	0.8	2.4	0.4
PCI3V	Required	12 mA	3.6	50% of V_{CC}	30% of V_{CC}	90% of V_{CC}	10% of V_{CC}
LVCMOS 3V	OK	12/24 mA	3.6	50% of V_{CC}	30% of V_{CC}	90% of V_{CC}	10% of V_{CC}

Additional Fast Capture Input Latch (Spartan-XL Family Only)

The Spartan-XL family OB has an additional optional latch on the input. This latch is clocked by the clock used for the output flip-flop rather than the input clock. Therefore, two different clocks can be used to clock the two input storage elements. This additional latch allows the fast capture of input data, which is then synchronized to the internal clock by the IOB flip-flop or latch.

To place the Fast Capture latch in a design, use one of the special library symbols, ILFFX or ILFLX. ILFFX is a transparent-Low Fast Capture latch followed by an active High input flip-flop. ILFLX is a transparent Low Fast Capture latch followed by a transparent High input latch. Any of the clock inputs can be inverted before driving the library element, and the inverter is absorbed into the IOB.

IOB Output Signal Path

Output signals can be optionally inverted within the IOB, and can pass directly to the output buffer or be stored in an edge-triggered flip-flop and then to the output buffer. The functionality of this flip-flop is shown in Table 6.

Table 6: **Output Flip-Flop Functionality**

Mode	Clock	Clock Enable	T	D	Q
Power-Up or GSR	X	X	0*	X	SR
Flip-Flop	X	0	0*	X	Q
	⌐	1*	0*	D	D
	X	X	1	X	Z
	0	X	0*	X	Q

Legend:

X	Don't care
⌐	Rising edge (clock not inverted).
SR	Set or Reset value. Reset is default.
0*	Input is Low or unconnected (default value)
1*	Input is High or unconnected (default value)
Z	3-state

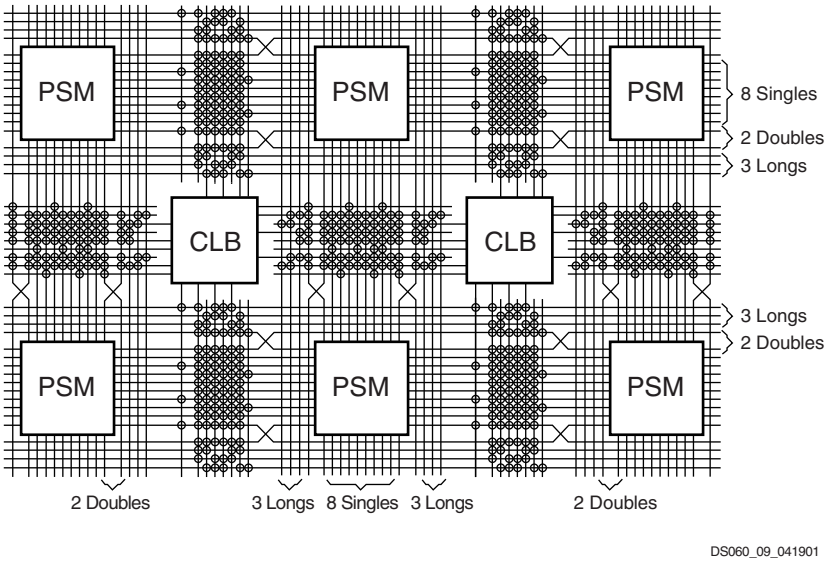


Figure 8: **Spartan/XL CLB Routing Channels and Interface Block Diagram**

CLB Interface

A block diagram of the CLB interface signals is shown in Figure 9. The input signals to the CLB are distributed evenly on all four sides providing maximum routing flexibility. In general, the entire architecture is symmetrical and regular. It is well suited to established placement and routing algorithms. Inputs, outputs, and function generators can freely swap positions within a CLB to avoid routing congestion during the placement and routing operation. The exceptions are the clock (K) input and CIN/COUT signals. The K input is routed to dedicated global vertical lines as well as four single-length lines and is on the left side of the CLB. The CIN/COUT signals are routed through dedicated interconnects which do not interfere with the general routing structure. The output signals from the CLB are available to drive both vertical and horizontal channels.

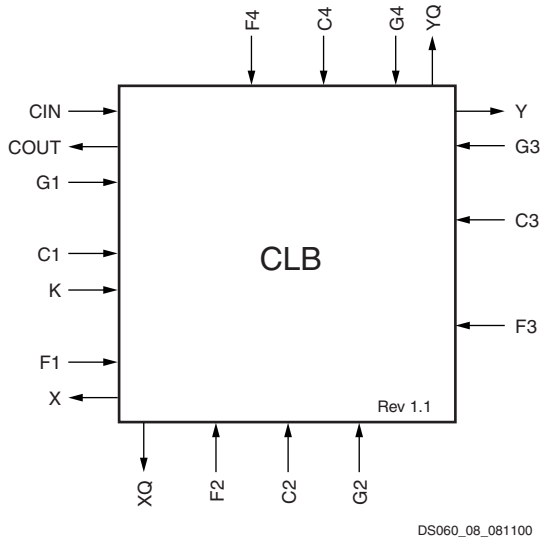


Figure 9: **CLB Interconnect Signals**

Programmable Switch Matrices

The horizontal and vertical single- and double-length lines intersect at a box called a programmable switch matrix (PSM). Each PSM consists of programmable pass transistors used to establish connections between the lines (see Figure 10).

For example, a single-length signal entering on the right side of the switch matrix can be routed to a single-length line on the top, left, or bottom sides, or any combination thereof, if multiple branches are required. Similarly, a double-length signal can be routed to a double-length line on any or all of the other three edges of the programmable switch matrix.

Single-Length Lines

Single-length lines provide the greatest interconnect flexibility and offer fast routing between adjacent blocks. There are eight vertical and eight horizontal single-length lines associated with each CLB. These lines connect the switching matrices that are located in every row and column of CLBs. Single-length lines are connected by way of the programmable switch matrices, as shown in Figure 10. Routing connectivity is shown in Figure 8.

Single-length lines incur a delay whenever they go through a PSM. Therefore, they are not suitable for routing signals for long distances. They are normally used to conduct signals within a localized area and to provide the branching for nets with fanout greater than one.

CLB signals from which they are originally derived are shown in Table 10.

Table 10: **Dual-Port RAM Signals**

RAM Signal	Function	CLB Signal
D	Data In	DIN
A[3:0]	Read Address for Single-Port. Write Address for Single-Port and Dual-Port.	F[4:1]
DPRA[3:0]	Read Address for Dual-Port	G[4:1]
WE	Write Enable	SR
WCLK	Clock	K
SPO	Single Port Out (addressed by A[3:0])	F_{OUT}
DPO	Dual Port Out (addressed by DPRA[3:0])	G_{OUT}

The RAM16X1D primitive used to instantiate the dual-port RAM consists of an upper and a lower 16 x 1 memory array. The address port labeled A[3:0] supplies both the read and write addresses for the lower memory array, which behaves the same as the 16 x 1 single-port RAM array described previously. Single Port Out (SPO) serves as the data output for the lower memory. Therefore, SPO reflects the data at address A[3:0].

The other address port, labeled DPRA[3:0] for Dual Port Read Address, supplies the read address for the upper memory. The write address for this memory, however, comes from the address A[3:0]. Dual Port Out (DPO) serves as the data output for the upper memory. Therefore, DPO reflects the data at address DPRA[3:0].

By using A[3:0] for the write address and DPRA[3:0] for the read address, and reading only the DPO output, a FIFO that can read and write simultaneously is easily generated. The simultaneous read/write capability possible with the dual-port RAM can provide twice the effective data throughput of a single-port RAM alternating read and write operations.

The timing relationships for the dual-port RAM mode are shown in Figure 13.

Note that write operations to RAM are synchronous (edge-triggered); however, data access is asynchronous.

Initializing RAM at FPGA Configuration

Both RAM and ROM implementations in the Spartan/XL families are initialized during device configuration. The initial contents are defined via an INIT attribute or property attached to the RAM or ROM symbol, as described in the library guide. If not defined, all RAM contents are initialized to zeros, by default.

RAM initialization occurs only during device configuration. The RAM content is not affected by GSR.

More Information on Using RAM Inside CLBs

Three application notes are available from Xilinx that discuss synchronous (edge-triggered) RAM: *"Xilinx Edge-Triggered and Dual-Port RAM Capability," "Implementing FIFOs in Xilinx RAM,"* and *"Synchronous and Asynchronous FIFO Designs."* All three application notes apply to both the Spartan and the Spartan-XL families.

Fast Carry Logic

Each CLB F-LUT and G-LUT contains dedicated arithmetic logic for the fast generation of carry and borrow signals. This extra output is passed on to the function generator in the adjacent CLB. The carry chain is independent of normal routing resources. (See Figure 15.)

Dedicated fast carry logic greatly increases the efficiency and performance of adders, subtractors, accumulators, comparators and counters. It also opens the door to many new applications involving arithmetic operation, where the previous generations of FPGAs were not fast enough or too inefficient. High-speed address offset calculations in microprocessor or graphics systems, and high-speed addition in digital signal processing are two typical applications.

The two 4-input function generators can be configured as a 2-bit adder with built-in hidden carry that can be expanded to any length. This dedicated carry circuitry is so fast and efficient that conventional speed-up methods like carry generate/propagate are meaningless even at the 16-bit level, and of marginal benefit at the 32-bit level. This fast carry logic is one of the more significant features of the Spartan

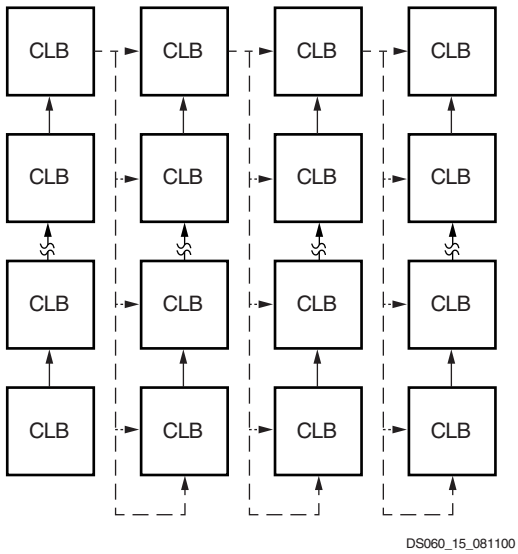
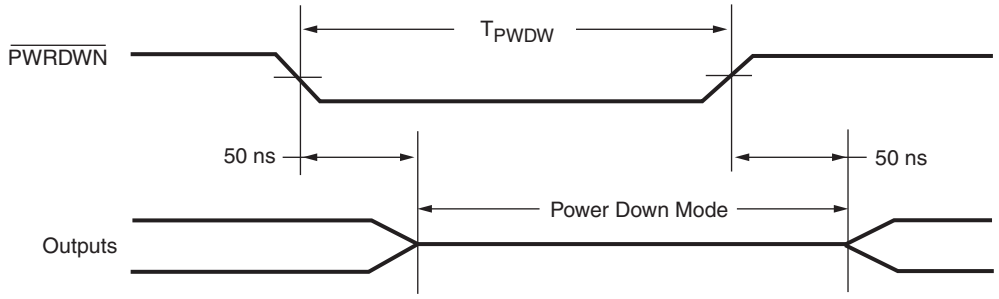


Figure 15: **Available Spartan/XL Carry Propagation Paths**



Description	Symbol	Min
Power Down Time	T_{PWD}	50 ns
Power Down Pulse Width	T_{PWDW}	50 ns

DS060_23_041901

Figure 23: **$\overline{\text{PWRDWN}}$ Pulse Timing**

Power-down retains the configuration, but loses all data stored in the device flip-flops. All inputs are interpreted as Low, but the internal combinatorial logic is fully functional. Make sure that the combination of all inputs Low and all flip-flops set or reset in your design will not generate internal oscillations, or create permanent bus contention by activating internal bus drivers with conflicting data onto the same long line.

During configuration, the $\overline{\text{PWRDWN}}$ pin must be High. If the Power Down state is entered before or during configuration, the device will restart configuration once the $\overline{\text{PWRDWN}}$ signal is removed. Note that the configuration pins are affected by Power Down and may not reflect their normal function. If there is an external pull-up resistor on the DONE pin, it will be High during Power Down even if the device is not yet configured. Similarly, if $\overline{\text{PWRDWN}}$ is asserted before configuration is completed, the $\overline{\text{INIT}}$ pin will not indicate status information.

Note that the $\overline{\text{PWRDWN}}$ pin is not part of the Boundary Scan chain. Therefore, the Spartan-XL family has a separate set of BSDL files than the 5V Spartan family. Boundary scan logic is not usable during Power Down.

Configuration and Test

Configuration is the process of loading design-specific programming data into one or more FPGAs to define the functional operation of the internal blocks and their interconnections. This is somewhat like loading the command registers of a programmable peripheral chip. Spartan/XL devices use several hundred bits of configuration data per CLB and its associated interconnects. Each configuration bit defines the state of a static memory cell that controls either a function look-up table bit, a multiplexer input, or an interconnect pass transistor. The Xilinx development system translates the design into a netlist file. It automatically partitions, places and routes the logic and generates the configuration data in PROM format.

Configuration Mode Control

5V Spartan devices have two configuration modes.

- MODE = 1 sets Slave Serial mode
- MODE = 0 sets Master Serial mode

3V Spartan-XL devices have three configuration modes.

- M1/M0 = 11 sets Slave Serial mode
- M1/M0 = 10 sets Master Serial mode
- M1/M0 = 0X sets Express mode

In addition to these modes, the device can be configured through the Boundary Scan logic (See "Configuration Through the Boundary Scan Pins" on page 37.).

The Mode pins are sampled prior to starting configuration to determine the configuration mode. After configuration, these pin are unused. The Mode pins have a weak pull-up resistor turned on during configuration. With the Mode pins High, Slave Serial mode is selected, which is the most popular configuration mode. Therefore, for the most common configuration mode, the Mode pins can be left unconnected. If the Master Serial mode is desired, the MODE/M0 pin should be connected directly to GND, or through a pull-down resistor of 1 KΩ or less.

During configuration, some of the I/O pins are used temporarily for the configuration process. All pins used during con-

Master Serial Mode

The Master serial mode uses an internal oscillator to generate a Configuration Clock (CCLK) for driving potential slave devices and the Xilinx serial-configuration PROM (SPROM). The CCLK speed is selectable as either 1 MHz (default) or 8 MHz. Configuration always starts at the default slow frequency, then can switch to the higher frequency during the first frame. Frequency tolerance is –50% to +25%.

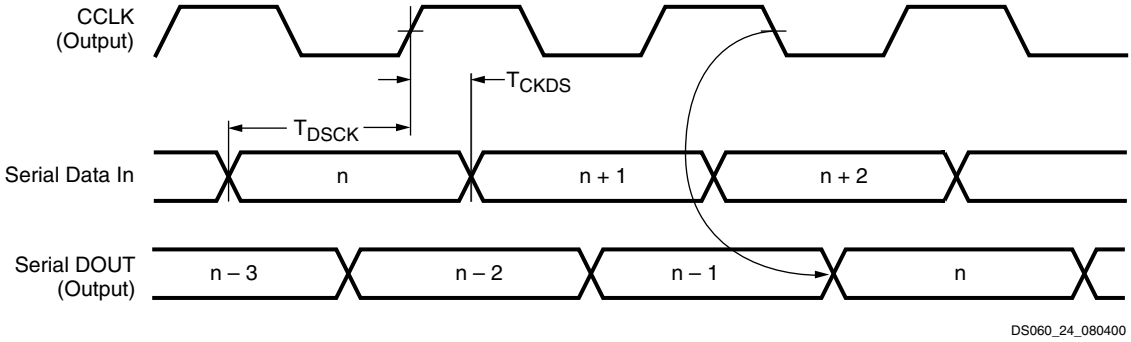
In Master Serial mode, the CCLK output of the device drives a Xilinx SPROM that feeds the FPGA DIN input. Each rising edge of the CCLK output increments the Serial PROM internal address counter. The next data bit is put on the SPROM data output, connected to the FPGA DIN pin. The FPGA accepts this data on the subsequent rising CCLK edge.

When used in a daisy-chain configuration the Master Serial FPGA is placed as the first device in the chain and is referred to as the lead FPGA. The lead FPGA presents the preamble data, and all data that overflows the lead device, on its DOUT pin. There is an internal pipeline delay of 1.5 CCLK periods, which means that DOUT changes on the falling CCLK edge, and the next FPGA in the daisy chain accepts data on the subsequent rising CCLK edge. See the timing diagram in Figure 24.

In the bitstream generation software, the user can specify Fast Configuration Rate, which, starting several bits into the first frame, increases the CCLK frequency by a factor of eight. For actual timing values please refer to the specification section. Be sure that the serial PROM and slaves are fast enough to support this data rate. Earlier families such as the XC3000 series do not support the Fast Configuration Rate option.

The SPROM CE input can be driven from either $\overline{LDC}$ or DONE. Using $\overline{LDC}$ avoids potential contention on the DIN pin, if this pin is configured as user I/O, but $\overline{LDC}$ is then restricted to be a permanently High user output after configuration. Using DONE can also avoid contention on DIN, provided the Early DONE option is invoked.

Figure 25 shows a full master/slave system. The leftmost device is in Master Serial mode, all other devices in the chain are in Slave Serial mode.



	Symbol	Description	Min	Units
CCLK	T_{DSCK}	DIN setup	20	ns
	T_{CKDS}	DIN hold	0	ns

Notes:
1. At power-up, V_{CC} must rise from 2.0V to V_{CC} min in less than 25 ms, otherwise delay configuration by pulling $\overline{PROGRAM}$ Low until V_{CC} is valid.
2. Master Serial mode timing is based on testing in slave mode.

Figure 24: **Master Serial Mode Programming Switching Characteristics**

Slave Serial Mode

In Slave Serial mode, the FPGA receives serial configuration data on the rising edge of CCLK and, after loading its configuration, passes additional data out, resynchronized on the next falling edge of CCLK.

In this mode, an external signal drives the CCLK input of the FPGA (most often from a Master Serial device). The serial configuration bitstream must be available at the DIN input of the lead FPGA a short setup time before each rising CCLK edge.

The lead FPGA then presents the preamble data—and all data that overflows the lead device—on its DOUT pin. There is an internal delay of 0.5 CCLK periods, which means that DOUT changes on the falling CCLK edge, and the next FPGA in the daisy chain accepts data on the subsequent rising CCLK edge.

Figure 25 shows a full master/slave system. A Spartan/XL device in Slave Serial mode should be connected as shown in the third device from the left.

to the DONE pin. User I/Os for each device become active after the DONE pin for that device goes High. (The exact timing is determined by development system options.) Since the DONE pin is open-drain and does not drive a High value, tying the DONE pins of all devices together prevents all devices in the chain from going High until the last device in the chain has completed its configuration cycle. If the DONE pin of a device is left unconnected, the device becomes active as soon as that device has been configured. Only devices supporting Express mode can be used to form an Express mode daisy chain.

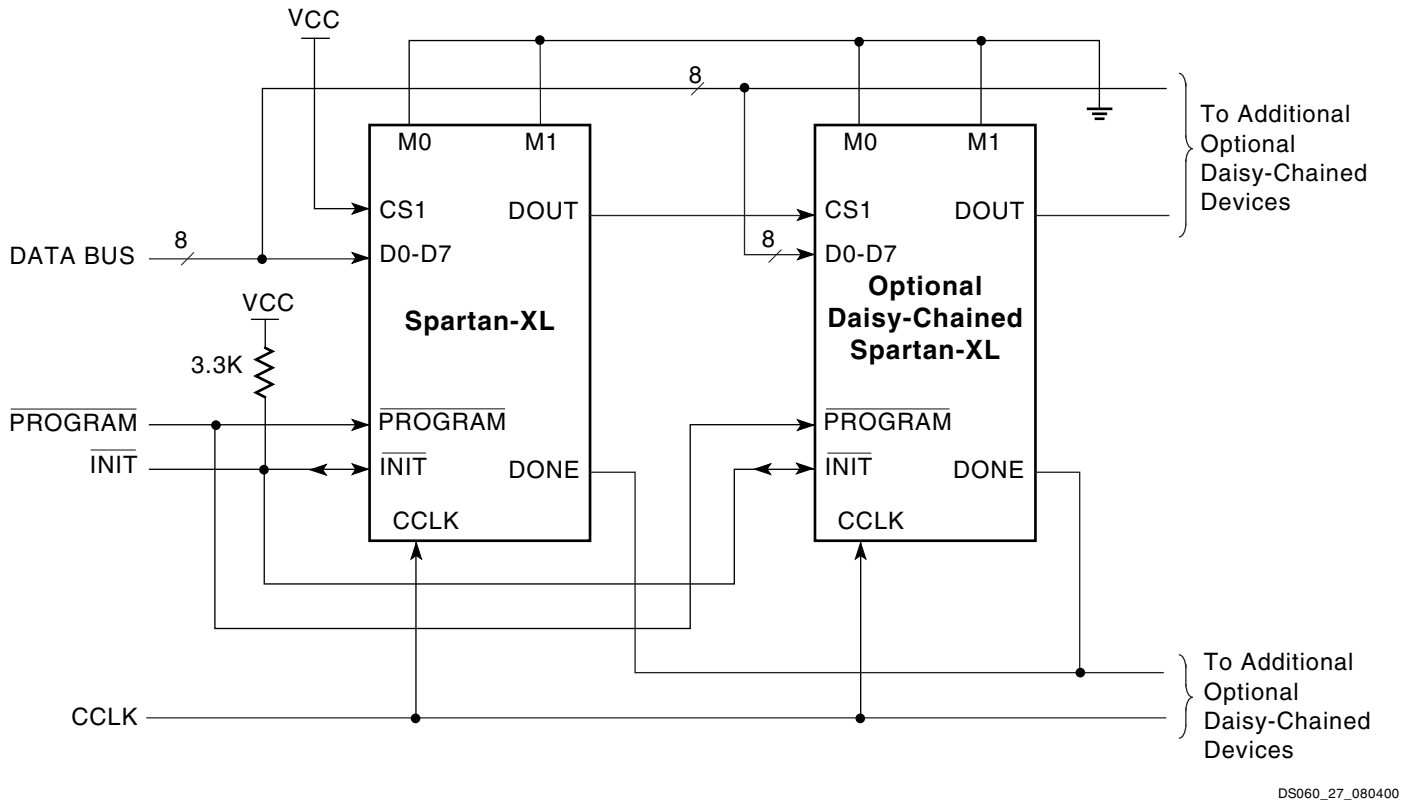


Figure 27: **Express Mode Circuit Diagram**

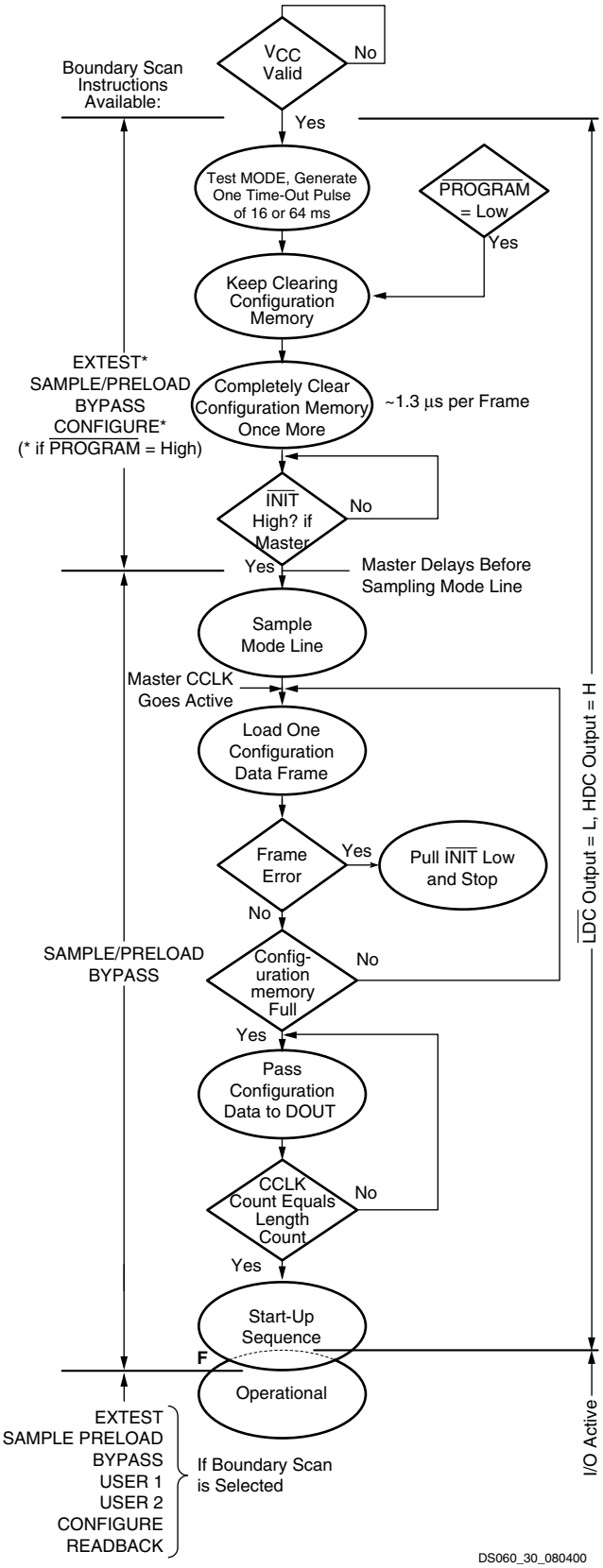


Figure 30: **Power-up Configuration Sequence**

Configuration

The 0010 preamble code indicates that the following 24 bits represent the length count for serial modes. The length count is the total number of configuration clocks needed to load the complete configuration data. (Four additional configuration clocks are required to complete the configuration process, as discussed below.) After the preamble and the length count have been passed through to any device in the daisy chain, its DOUT is held High to prevent frame start bits from reaching any daisy-chained devices. In Spartan-XL family Express mode, the length count bits are ignored, and DOUT is held Low, to disable the next device in the pseudo daisy chain.

A specific configuration bit, early in the first frame of a master device, controls the configuration-clock rate and can increase it by a factor of eight. Therefore, if a fast configuration clock is selected by the bitstream, the slower clock rate is used until this configuration bit is detected.

Each frame has a start field followed by the frame-configuration data bits and a frame error field. If a frame data error is detected, the FPGA halts loading, and signals the error by pulling the open-drain $\overline{\text{INIT}}$ pin Low. After all configuration frames have been loaded into an FPGA using a serial mode, DOUT again follows the input data so that the remaining data is passed on to the next device. In Spartan-XL family Express mode, when the first device is fully programmed, DOUT goes High to enable the next device in the chain.

Delaying Configuration After Power-Up

There are two methods of delaying configuration after power-up: put a logic Low on the $\overline{\text{PROGRAM}}$ input, or pull the bidirectional $\overline{\text{INIT}}$ pin Low, using an open-collector (open-drain) driver. (See Figure 30.)

A Low on the $\overline{\text{PROGRAM}}$ input is the more radical approach, and is recommended when the power-supply rise time is excessive or poorly defined. As long as $\overline{\text{PROGRAM}}$ is Low, the FPGA keeps clearing its configuration memory. When $\overline{\text{PROGRAM}}$ goes High, the configuration memory is cleared one more time, followed by the beginning of configuration, provided the $\overline{\text{INIT}}$ input is not externally held Low. Note that a Low on the $\overline{\text{PROGRAM}}$ input automatically forces a Low on the $\overline{\text{INIT}}$ output. The Spartan/XL FPGA $\overline{\text{PROGRAM}}$ pin has a permanent weak pull-up.

Avoid holding $\overline{\text{PROGRAM}}$ Low for more than 500 µs. The 500 µs maximum limit is only a recommendation, not a requirement. The only effect of holding $\overline{\text{PROGRAM}}$ Low for more than 500 µs is an increase in current, measured at about 40 mA in the XCS40XL. This increased current cannot damage the device. This applies only during reconfiguration, not during power-up. The $\overline{\text{INIT}}$ pin can also be held Low to delay reconfiguration, and the same characteristics apply as for the $\overline{\text{PROGRAM}}$ pin.

Using an open-collector or open-drain driver to hold $\overline{\text{INIT}}$ Low before the beginning of configuration causes the FPGA

to wait after completing the configuration memory clear operation. When $\overline{INIT}$ is no longer held Low externally, the device determines its configuration mode by capturing the state of the Mode pins, and is ready to start the configuration process. A master device waits up to an additional 300 µs to make sure that any slaves in the optional daisy chain have seen that $\overline{INIT}$ is High.

For more details on Configuration, refer to the Xilinx Application Note "FPGA Configuration Guidelines" (XAPP090).

Start-Up

Start-up is the transition from the configuration process to the intended user operation. This transition involves a change from one clock source to another, and a change from interfacing parallel or serial configuration data where most outputs are 3-stated, to normal operation with I/O pins active in the user system. Start-up must make sure that the user logic 'wakes up' gracefully, that the outputs become active without causing contention with the configuration signals, and that the internal flip-flops are released from the Global Set/Reset (GSR) at the right time.

Start-Up Initiation

Two conditions have to be met in order for the start-up sequence to begin:

- The chip's internal memory must be full, and
- The configuration length count must be met, exactly.

In all configuration modes except Express mode, Spartan/XL devices read the expected length count from the bitstream and store it in an internal register. The length count varies according to the number of devices and the composition of the daisy chain. Each device also counts the number of CCLKs during configuration.

In Express mode, there is no length count. The start-up sequence for each device begins when the device has received its quota of configuration data. Wiring the DONE pins of several devices together delays start-up of all devices until all are fully configured.

Start-Up Events

The device can be programmed to control three start-up events.

- The release of the open-drain DONE output
- The termination of the Global Three-State and the change of configuration-related pins to the user function, activating all IOBs.
- The termination of the Global Set/Reset initialization of all CLB and IOB storage elements.

Figure 31 describes start-up timing in detail. The three events — DONE going High, the internal GSR being de-activated, and the user I/O going active — can all occur in any arbitrary sequence. This relative timing is selected by options in the bitstream generation software. Heavy lines in Figure 31 show the default timing. The thin lines indicate all other possible timing options. The start-up logic must be clocked until the "F" (Finished) state is reached.

The default option, and the most practical one, is for DONE to go High first, disconnecting the configuration data source and avoiding any contention when the I/Os become active one clock later. GSR is then released another clock period later to make sure that user operation starts from stable internal conditions. This is the most common sequence, shown with heavy lines in Figure 31, but the designer can modify it to meet particular requirements.

Start-Up Clock

Normally, the start-up sequence is controlled by the internal device oscillator (CCLK), which is asynchronous to the system clock. As a configuration option, they can be triggered by an on-chip user net called UCLK. This user net can be accessed by placing the STARTUP library symbol, and the start-up modes are known as UCLK_NOSYNC or UCLK_SYNC. This allows the device to wake up in synchronism with the user system.

DONE Pin

Note that DONE is an open-drain output and does not go High unless an internal pull-up is activated or an external pull-up is attached. The internal pull-up is activated as the default by the bitstream generation software.

The DONE pin can also be wire-ANDed with DONE pins of other FPGAs or with other external signals, and can then be used as input to the start-up control logic. This is called "Start-up Timing Synchronous to Done In" and is selected by either CCLK_SYNC or UCLK_SYNC. When DONE is not used as an input, the operation is called "Start-up Timing Not Synchronous to DONE In," and is selected by either CCLK_NOSYNC or UCLK_NOSYNC. Express mode configuration always uses either CCLK_SYNC or UCLK_SYNC timing, while the other configuration modes can use any of the four timing sequences.

When the UCLK_SYNC option is enabled, the user can externally hold the open-drain DONE output Low, and thus stall all further progress in the start-up sequence until DONE is released and has gone High. This option can be used to force synchronization of several FPGAs to a common user clock, or to guarantee that all devices are successfully configured before any I/Os go active.

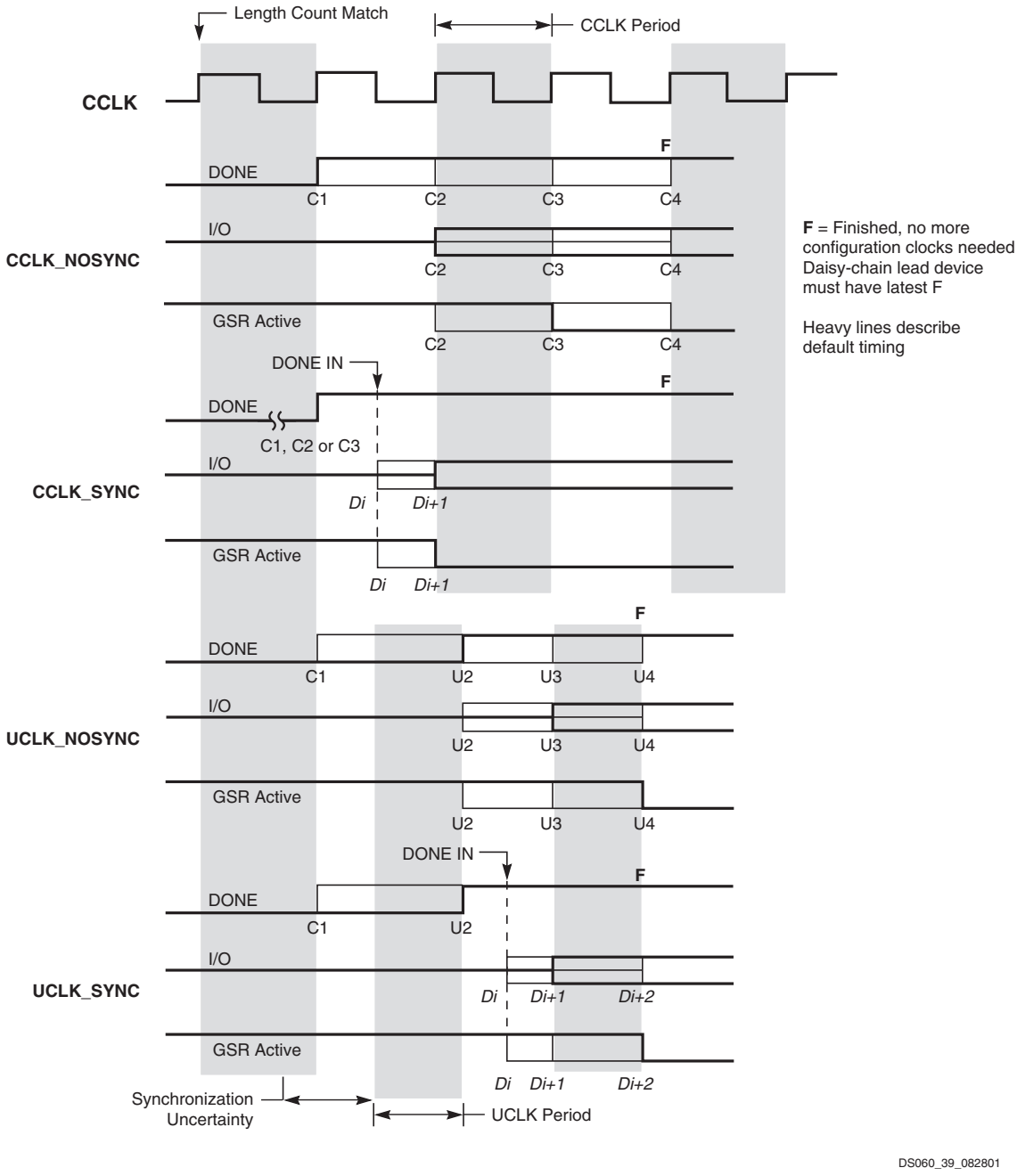


Figure 31: **Start-up Timing**

Configuration Through the Boundary Scan Pins

Spartan/XL devices can be configured through the boundary scan pins. The basic procedure is as follows:

- Power up the FPGA with $\overline{\text{INIT}}$ held Low (or drive the $\overline{\text{PROGRAM}}$ pin Low for more than 300 ns followed by a High while holding $\overline{\text{INIT}}$ Low). Holding $\overline{\text{INIT}}$ Low allows enough time to issue the CONFIG command to the FPGA. The pin can be used as I/O after configuration if a resistor is used to hold $\overline{\text{INIT}}$ Low.
- Issue the CONFIG command to the TMS input.
- Wait for $\overline{\text{INIT}}$ to go High.
- Sequence the boundary scan Test Access Port to the SHIFT-DR state.
- Toggle TCK to clock data into TDI pin.

The user must account for all TCK clock cycles after $\overline{\text{INIT}}$ goes High, as all of these cycles affect the Length Count compare.

For more detailed information, refer to the Xilinx application note, "*Boundary Scan in FPGA Devices*." This application note applies to Spartan and Spartan-XL devices.

Readback Switching Characteristics Guidelines

The following guidelines reflect worst-case values over the recommended operating conditions.

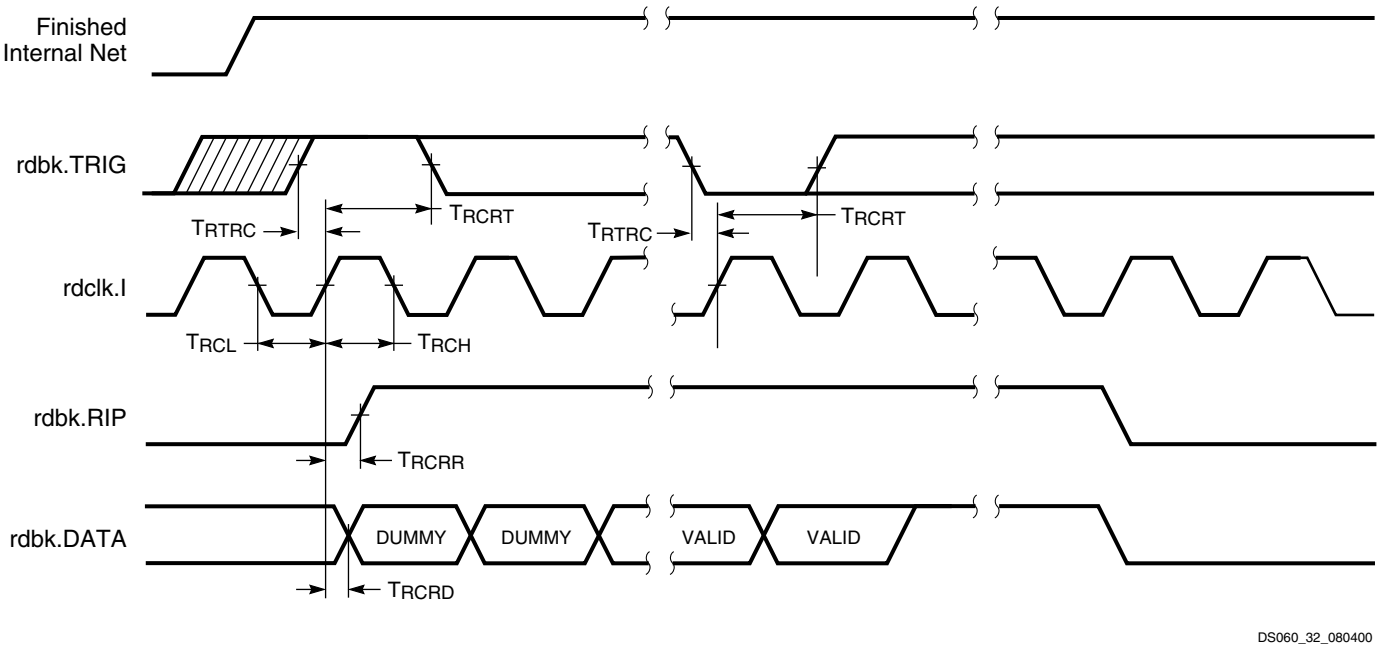


Figure 33: **Spartan and Spartan-XL Readback Timing Diagram**

Spartan and Spartan-XL Readback Switching Characteristics

Symbol		Description	Min	Max	Units
T_{RTRC}	rdbk.TRIG	rdbk.TRIG setup to initiate and abort Readback	200	-	ns
T_{RCRT}		rdbk.TRIG hold to initiate and abort Readback	50	-	ns
T_{RCRD}	rdclk.I	rdbk.DATA delay	-	250	ns
T_{RCRR}		rdbk.RIP delay	-	250	ns
T_{RCH}		High time	250	500	ns
T_{RCL}		Low time	250	500	ns

Notes:
1. Timing parameters apply to all speed grades.
2. If rdbk.TRIG is High prior to Finished, Finished will trigger the first Readback.

Spartan-XL Family Global Buffer Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.

When fewer vertical clock lines are connected, the clock distribution is faster; when multiple clock lines per column are driven from the same global clock, the delay is longer. For more specific, more precise, and worst-case guaranteed data, reflecting the actual routing structure, use the values provided by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. These path delays, provided as a guideline, have been extracted from the static timing analyzer report. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature).

Symbol	Description	Device	Speed Grade		Units
			-5	-4	
			Max	Max	
T_{GLS}	From pad through buffer, to any clock K	XCS05XL	1.4	1.5	ns
		XCS10XL	1.7	1.8	ns
		XCS20XL	2.0	2.1	ns
		XCS30XL	2.3	2.5	ns
		XCS40XL	2.6	2.8	ns

Spartan-XL Family CLB Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values apply to all Spartan-XL devices and expressed in nanoseconds unless otherwise noted.

Symbol	Description	Speed Grade				Units
		-5		-4		
		Min	Max	Min	Max	
Clocks						
T_{CH}	Clock High time	2.0	-	2.3	-	ns
T_{CL}	Clock Low time	2.0	-	2.3	-	ns
Combinatorial Delays						
T_{ILO}	F/G inputs to X/Y outputs	-	1.0	-	1.1	ns
T_{IHO}	F/G inputs via H to X/Y outputs	-	1.7	-	2.0	ns
T_{ITO}	F/G inputs via transparent latch to Q outputs	-	1.5	-	1.8	ns
T_{HH1O}	C inputs via H1 via H to X/Y outputs	-	1.5	-	1.8	ns
Sequential Delays						
T_{CKO}	Clock K to Flip-Flop or latch outputs Q	-	1.2	-	1.4	ns
Setup Time before Clock K						
T_{ICK}	F/G inputs	0.6	-	0.7	-	ns
T_{IHCK}	F/G inputs via H	1.3	-	1.6	-	ns
Hold Time after Clock K						
	All Hold times, all devices	0.0	-	0.0	-	ns
Set/Reset Direct						
T_{RPW}	Width (High)	2.5	-	2.8	-	ns
T_{RIO}	Delay from C inputs via S/R, going High to Q	-	2.3	-	2.7	ns
Global Set/Reset						
T_{MRW}	Minimum GSR Pulse Width	10.5	-	11.5	-	ns
T_{MRQ}	Delay from GSR input to any Q	See page 60 for T_{RRI} values per device.				
F_{TOG}	Toggle Frequency (MHz) (for export control purposes)	-	250	-	217	MHz

Spartan-XL Family CLB RAM Synchronous (Edge-Triggered) Write Operation Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values apply to all Spartan-XL devices and are expressed in nanoseconds unless otherwise noted.

Symbol	Single Port RAM	Size(1)	Speed Grade				Units
			-5		-4		
			Min	Max	Min	Max	
Write Operation							
T_{WCS}	Address write cycle time (clock K period)	16x2	7.7	-	8.4	-	ns
T_{WCTS}		32x1	7.7	-	8.4	-	ns
T_{WPS}	Clock K pulse width (active edge)	16x2	3.1	-	3.6	-	ns
T_{WPTS}		32x1	3.1	-	3.6	-	ns
T_{ASS}	Address setup time before clock K	16x2	1.3	-	1.5	-	ns
T_{ASTS}		32x1	1.5	-	1.7	-	ns
T_{DSS}	DIN setup time before clock K	16x2	1.5	-	1.7	-	ns
T_{DSTS}		32x1	1.8	-	2.1	-	ns
T_{WSS}	WE setup time before clock K	16x2	1.4	-	1.6	-	ns
T_{WSTS}		32x1	1.3	-	1.5	-	ns
	All hold times after clock K	16x2	0.0	-	0.0	-	ns
T_{WOS}	Data valid after clock K	32x1	-	4.5	-	5.3	ns
T_{WOTS}		16x2	-	5.4	-	6.3	ns
Read Operation							
T_{RC}	Address read cycle time	16x2	2.6	-	3.1	-	ns
T_{RCT}		32x1	3.8	-	5.5	-	ns
T_{ILO}	Data Valid after address change (no Write Enable)	16x2	-	1.0	-	1.1	ns
T_{IHO}		32x1	-	1.7	-	2.0	ns
T_{ICK}	Address setup time before clock K	16x2	0.6	-	0.7	-	ns
T_{IHCK}		32x1	1.3	-	1.6	-	ns

Notes:
1. Timing for 16 x 1 RAM option is identical to 16 x 2 RAM timing.

Spartan-XL Family IOB Output Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. These path delays, provided as a guideline, have been extracted from the static timing analyzer report. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values are expressed in nanoseconds unless otherwise noted.

Symbol	Description	Device	Speed Grade				Units
			-5		-4		
			Min	Max	Min	Max	
Propagation Delays							
T_{OKPOF}	Clock (OK) to Pad, fast	All devices	-	3.2	-	3.7	ns
T_{OPF}	Output (O) to Pad, fast	All devices	-	2.5	-	2.9	ns
T_{TSHZ}	3-state to Pad High-Z (slew-rate independent)	All devices	-	2.8	-	3.3	ns
T_{TSONF}	3-state to Pad active and valid, fast	All devices	-	2.6	-	3.0	ns
T_{OFPF}	Output (O) to Pad via Output MUX, fast	All devices	-	3.7	-	4.4	ns
T_{OKFPF}	Select (OK) to Pad via Output MUX, fast	All devices	-	3.3	-	3.9	ns
T_{SLOW}	For Output SLOW option add	All devices	-	1.5	-	1.7	ns
Setup and Hold Times							
T_{OOK}	Output (O) to clock (OK) setup time	All devices	0.5	-	0.5	-	ns
T_{OKO}	Output (O) to clock (OK) hold time	All devices	0.0	-	0.0	-	ns
T_{ECOK}	Clock Enable (EC) to clock (OK) setup time	All devices	0.0	-	0.0	-	ns
T_{OKEC}	Clock Enable (EC) to clock (OK) hold time	All devices	0.1	-	0.2	-	ns
Global Set/Reset							
T_{MRW}	Minimum GSR pulse width	All devices	10.5	-	11.5	-	ns
T_{RPO}	Delay from GSR input to any Pad	XCS05XL	-	11.9	-	14.0	ns
		XCS10XL	-	12.4	-	14.5	ns
		XCS20XL	-	12.9	-	15.0	ns
		XCS30XL	-	13.9	-	16.0	ns
		XCS40XL	-	14.9	-	17.0	ns

Notes:

1. Output timing is measured at ~50% V_{CC} threshold, with 50 pF external capacitive loads including test fixture. Slew-rate limited output rise/fall times are approximately two times longer than fast output rise/fall times.
2. Voltage levels of unused pads, bonded or unbonded, must be valid logic levels. Each can be configured with the internal pull-up (default) or pull-down resistor, or configured as a driven output, or can be driven from an external source.

Additional XCS20/XL Package Pins

PQ208					
Not Connected Pins					
P12	P18 (1)	P33 (1)	P39	P65	P71 (1)
P86 (1)	P92	P111	P121(1)	P140(1)	P144
P165	P173(1)	P192(1)	P202	P203	-
9/16/98					

Notes:
1. 5V Spartan family only
2. 3V Spartan-XL family only
3. The “PWRDWN” on the XCS20XL is not part of the Boundary Scan chain. For the XCS20XL, subtract 1 from all Boundary Scan numbers from GCK3 on (247 and higher).
4. CS144 package discontinued by PDN2004-01

XCS30 and XCS30XL Device Pinouts

XCS30/XL Pad Name	VQ100(5)	TQ144	PQ208	PQ240	BG256(5)	CS280(2,5)	Bndry Scan
VCC	P89	P128	P183	P212	VCC(4)	C10	-
I/O	P90	P129	P184	P213	C10	D10	74
I/O	P91	P130	P185	P214	D10	E10	77
I/O	P92	P131	P186	P215	A9	A9	80
I/O	P93	P132	P187	P216	B9	B9	83
I/O	-	-	P188	P217	C9	C9	86
I/O	-	-	P189	P218	D9	D9	89
I/O	P94	P133	P190	P220	A8	A8	92
I/O	P95	P134	P191	P221	B8	B8	95
VCC	-	-	P192	P222	VCC(4)	A7	-
I/O	-	-	-	P223	A6	B7	98
I/O	-	-	-	P224	C7	C7	101
I/O	-	P135	P193	P225	B6	D7	104
I/O	-	P136	P194	P226	A5	A6	107
GND	-	P137	P195	P227	GND(4)	GND(4)	-
I/O	-	-	P196	P228	C6	B6	110
I/O	-	-	P197	P229	B5	C6	113
I/O	-	-	P198	P230	A4	D6	116
I/O	-	-	P199	P231	C5	E6	119
I/O	P96	P138	P200	P232	B4	A5	122
I/O	P97	P139	P201	P233	A3	C5	125
I/O	-	-	P202	P234	D5	B4	128
I/O	-	-	P203	P235	C4	C4	131
I/O	-	P140	P204	P236	B3	A3	134
I/O	-	P141	P205	P237	B2	A2	137
I/O	P98	P142	P206	P238	A2	B3	140
I/O, SGCK1(1), GCK8(2)	P99	P143	P207	P239	C3	B2	143
VCC	P100	P144	P208	P240	VCC(4)	A1	-
GND	P1	P1	P1	P1	GND(4)	GND(4)	-
I/O, PGCK1(1), GCK1(2)	P2	P2	P2	P2	B1	C3	146
I/O	P3	P3	P3	P3	C2	C2	149
I/O	-	P4	P4	P4	D2	B1	152

XCS30 and XCS30XL Device Pinouts *(Continued)*

XCS30/XL Pad Name	VQ100[5]	TQ144	PQ208	PQ240	BG256[5]	CS280[2,5]	Bndry Scan
I/O	-	P5	P5	P5	D3	C1	155
I/O, TDI	P4	P6	P6	P6	E4	D4	158
I/O, TCK	P5	P7	P7	P7	C1	D3	161
I/O	-	-	P8	P8	D1	E2	164
I/O	-	-	P9	P9	E3	E4	167
I/O	-	-	P10	P10	E2	E1	170
I/O	-	-	P11	P11	E1	F5	173
I/O	-	-	P12	P12	F3	F3	176
I/O	-	-	-	P13	F2	F2	179
GND	-	P8	P13	P14	GND[4]	GND[4]	-
I/O	-	P9	P14	P15	G3	F4	182
I/O	-	P10	P15	P16	G2	F1	185
I/O, TMS	P6	P11	P16	P17	G1	G3	188
I/O	P7	P12	P17	P18	H3	G2	191
VCC	-	-	P18	P19	VCC[4]	G1	-
I/O	-	-	-	P20	H2	G4	194
I/O	-	-	-	P21	H1	H1	197
I/O	-	-	P19	P23	J2	H4	200
I/O	-	-	P20	P24	J1	J1	203
I/O	-	P13	P21	P25	K2	J2	206
I/O	P8	P14	P22	P26	K3	J3	209
I/O	P9	P15	P23	P27	K1	J4	212
I/O	P10	P16	P24	P28	L1	K1	215
GND	P11	P17	P25	P29	GND[4]	GND[4]	-
VCC	P12	P18	P26	P30	VCC[4]	K2	-
I/O	P13	P19	P27	P31	L2	K3	218
I/O	P14	P20	P28	P32	L3	K4	221
I/O	P15	P21	P29	P33	L4	K5	224
I/O	-	P22	P30	P34	M1	L1	227
I/O	-	-	P31	P35	M2	L2	230
I/O	-	-	P32	P36	M3	L3	233
I/O	-	-	-	P38	N1	M2	236
I/O	-	-	-	P39	N2	M3	239
VCC	-	-	P33	P40	VCC[4]	M4	-
I/O	P16	P23	P34	P41	P1	N1	242
I/O	P17	P24	P35	P42	P2	N2	245
I/O	-	P25	P36	P43	R1	N3	248
I/O	-	P26	P37	P44	P3	N4	251
GND	-	P27	P38	P45	GND[4]	GND[4]	-
I/O	-	-	-	P46	T1	P1	254
I/O	-	-	P39	P47	R3	P2	257
I/O	-	-	P40	P48	T2	P3	260
I/O	-	-	P41	P49	U1	P4	263
I/O	-	-	P42	P50	T3	P5	266
I/O	-	-	P43	P51	U2	R1	269

XCS30 and XCS30XL Device Pinouts *(Continued)*

XCS30/XL Pad Name	VQ100[5]	TQ144	PQ208	PQ240	BG256[5]	CS280[2,5]	Bndry Scan
I/O	P18	P28	P44	P52	V1	T1	272
I/O	P19	P29	P45	P53	T4	T2	275
I/O	-	P30	P46	P54	U3	T3	278
I/O	-	P31	P47	P55	V2	U1	281
I/O	P20	P32	P48	P56	W1	V1	284
I/O, SGCK2[1], GCK2[2]	P21	P33	P49	P57	V3	U2	287
Not Connected[1], M1[2]	P22	P34	P50	P58	W2	V2	290
GND	P23	P35	P51	P59	GND[4]	GND[4]	-
MODE[1], M0[2]	P24	P36	P52	P60	Y1	W1	293
VCC	P25	P37	P53	P61	VCC[4]	U3	-
Not Connected [1], $\overline{\text{PWRDWN}}$[2]	P26	P38	P54	P62	W3	V3	294 [1]
I/O, PGCK2[1], GCK3[2]	P27	P39	P55	P63	Y2	W2	295 [3]
I/O (HDC)	P28	P40	P56	P64	W4	W3	298 [3]
I/O	-	P41	P57	P65	V4	T4	301 [3]
I/O	-	P42	P58	P66	U5	U4	304 [3]
I/O	P29	P43	P59	P67	Y3	V4	307 [3]
I/O ($\overline{\text{LDC}}$)	P30	P44	P60	P68	Y4	W4	310 [3]
I/O	-	-	P61	P69	V5	T5	313 [3]
I/O	-	-	P62	P70	W5	W5	316 [3]
I/O	-	-	P63	P71	Y5	R6	319 [3]
I/O	-	-	P64	P72	V6	U6	322 [3]
I/O	-	-	P65	P73	W6	V6	325 [3]
I/O	-	-	-	P74	Y6	T6	328 [3]
GND	-	P45	P66	P75	GND[4]	GND[4]	-
I/O	-	P46	P67	P76	W7	W6	331 [3]
I/O	-	P47	P68	P77	Y7	U7	334 [3]
I/O	P31	P48	P69	P78	V8	V7	337 [3]
I/O	P32	P49	P70	P79	W8	W7	340 [3]
VCC	-	-	P71	P80	VCC[4]	T7	-
I/O	-	-	P72	P81	Y8	W8	343 [3]
I/O	-	-	P73	P82	U9	U8	346 [3]
I/O	-	-	-	P84	Y9	W9	349 [3]
I/O	-	-	-	P85	W10	V9	352 [3]
I/O	P33	P50	P74	P86	V10	U9	355 [3]
I/O	P34	P51	P75	P87	Y10	T9	358 [3]
I/O	P35	P52	P76	P88	Y11	W10	361 [3]
I/O ($\overline{\text{INIT}}$)	P36	P53	P77	P89	W11	V10	364 [3]
VCC	P37	P54	P78	P90	VCC[4]	U10	-
GND	P38	P55	P79	P91	GND[4]	GND[4]	-
I/O	P39	P56	P80	P92	V11	T10	367 [3]
I/O	P40	P57	P81	P93	U11	R10	370 [3]
I/O	P41	P58	P82	P94	Y12	W11	373 [3]
I/O	P42	P59	P83	P95	W12	V11	376 [3]
I/O	-	-	P84	P96	V12	U11	379 [3]

XCS30 and XCS30XL Device Pinouts *(Continued)*

XCS30/XL Pad Name	VQ100[5]	TQ144	PQ208	PQ240	BG256[5]	CS280[2,5]	Bndry Scan
I/O	-	-	P85	P97	U12	T11	382 [3]
I/O	-	-	-	P99	V13	U12	385 [3]
I/O	-	-	-	P100	Y14	T12	388 [3]
VCC	-	-	P86	P101	VCC[4]	W13	-
I/O	P43	P60	P87	P102	Y15	V13	391 [3]
I/O	P44	P61	P88	P103	V14	U13	394 [3]
I/O	-	P62	P89	P104	W15	T13	397 [3]
I/O	-	P63	P90	P105	Y16	W14	400 [3]
GND	-	P64	P91	P106	GND[4]	GND[4]	-
I/O	-	-	-	P107	V15	V14	403 [3]
I/O	-	-	P92	P108	W16	U14	406 [3]
I/O	-	-	P93	P109	Y17	T14	409 [3]
I/O	-	-	P94	P110	V16	R14	412 [3]
I/O	-	-	P95	P111	W17	W15	415 [3]
I/O	-	-	P96	P112	Y18	U15	418 [3]
I/O	P45	P65	P97	P113	U16	V16	421 [3]
I/O	P46	P66	P98	P114	V17	U16	424 [3]
I/O	-	P67	P99	P115	W18	W17	427 [3]
I/O	-	P68	P100	P116	Y19	W18	430 [3]
I/O	P47	P69	P101	P117	V18	V17	433 [3]
I/O, SGCK3[1], GCK4[2]	P48	P70	P102	P118	W19	V18	436 [3]
GND	P49	P71	P103	P119	GND[4]	GND[4]	-
DONE	P50	P72	P104	P120	Y20	W19	-
VCC	P51	P73	P105	P121	VCC[4]	U17	-
$\overline{\text{PROGRAM}}$	P52	P74	P106	P122	V19	U18	-
I/O (D7[2])	P53	P75	P107	P123	U19	V19	439 [3]
I/O, PGCK3[1], GCK5[2]	P54	P76	P108	P124	U18	U19	442 [3]
I/O	-	P77	P109	P125	T17	T16	445 [3]
I/O	-	P78	P110	P126	V20	T17	448 [3]
I/O	-	-	-	P127	U20	T18	451 [3]
I/O	-	-	P111	P128	T18	T19	454 [3]
I/O (D6[2])	P55	P79	P112	P129	T19	R16	457 [3]
I/O	P56	P80	P113	P130	T20	R19	460 [3]
I/O	-	-	P114	P131	R18	P15	463 [3]
I/O	-	-	P115	P132	R19	P17	466 [3]
I/O	-	-	P116	P133	R20	P18	469 [3]
I/O	-	-	P117	P134	P18	P16	472 [3]
GND	-	P81	P118	P135	GND[4]	GND[4]	-
I/O	-	-	-	P136	P20	P19	475 [3]
I/O	-	-	-	P137	N18	N17	478 [3]
I/O	-	P82	P119	P138	N19	N18	481 [3]
I/O	-	P83	P120	P139	N20	N19	484 [3]
VCC	-	-	P121	P140	VCC[4]	N16	-
I/O (D5[2])	P57	P84	P122	P141	M17	M19	487 [3]
I/O	P58	P85	P123	P142	M18	M17	490 [3]

CS280

VCC Pins					
E5	E7	E8	E9	E11	E12
E13	G5	G15	H5	H15	J5
J15	L5	L15	M5	M15	N5
N15	R7	R8	R9	R11	R12
R13	-	-	-	-	-
Not Connected Pins					
A4	A12	C8	C12	C15	D1
D2	D5	D8	D17	D18	E15
H2	H3	H18	H19	L4	M1
M16	M18	R2	R4	R5	R15
R17	T8	T15	U5	V8	V12
W12	W16	-	-	-	-
Not Connected Pins (VCC in XCS40XL)					
B5	B15	E3	E18	R3	R18
V5	V15	-	-	-	-

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XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
VCC	P183	P212	VCC(4)	VCC(4)	-
I/O	P184	P213	C10	D10	86
I/O	P185	P214	D10	E10	89
I/O	P186	P215	A9	A9	92
I/O	P187	P216	B9	B9	95
I/O	P188	P217	C9	C9	98
I/O	P189	P218	D9	D9	101
I/O	P190	P220	A8	A8	104
I/O	P191	P221	B8	B8	107
I/O	-	-	C8	C8	110
I/O	-	-	A7	D8	113
VCC	P192	P222	VCC(4)	VCC(4)	-
I/O	-	P223	A6	B7	116
I/O	-	P224	C7	C7	119
I/O	P193	P225	B6	D7	122
I/O	P194	P226	A5	A6	125
GND	P195	P227	GND(4)	GND(4)	-
I/O	P196	P228	C6	B6	128
I/O	P197	P229	B5	C6	131
I/O	P198	P230	A4	D6	134
I/O	P199	P231	C5	E6	137

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
I/O	P200	P232	B4	A5	140
I/O	P201	P233	A3	C5	143
I/O	-	-	-	D5	146
I/O	-	-	-	A4	149
I/O	P202	P234	D5	B4	152
I/O	P203	P235	C4	C4	155
I/O	P204	P236	B3	A3	158
I/O	P205	P237	B2	A2	161
I/O	P206	P238	A2	B3	164
I/O, SGCK1(1), GCK8(2)	P207	P239	C3	B2	167
VCC	P208	P240	VCC(4)	VCC(4)	-
GND	P1	P1	GND(4)	GND(4)	-
I/O, PGCK1(1), GCK1(2)	P2	P2	B1	C3	170
I/O	P3	P3	C2	C2	173
I/O	P4	P4	D2	B1	176
I/O	P5	P5	D3	C1	179
I/O, TDI	P6	P6	E4	D4	182
I/O, TCK	P7	P7	C1	D3	185
I/O	-	-	-	D2	188
I/O	-	-	-	D1	191
I/O	P8	P8	D1	E2	194
I/O	P9	P9	E3	E4	197
I/O	P10	P10	E2	E1	200
I/O	P11	P11	E1	F5	203
I/O	P12	P12	F3	F3	206
I/O	-	P13	F2	F2	209
GND	P13	P14	GND(4)	GND(4)	-
I/O	P14	P15	G3	F4	212
I/O	P15	P16	G2	F1	215
I/O, TMS	P16	P17	G1	G3	218
I/O	P17	P18	H3	G2	221
VCC	P18	P19	VCC(4)	VCC(4)	-
I/O	-	P20	H2	G4	224
I/O	-	P21	H1	H1	227
I/O	-	-	J4	H3	230
I/O	-	-	J3	H2	233
I/O	P19	P23	J2	H4	236
I/O	P20	P24	J1	J1	239
I/O	P21	P25	K2	J2	242
I/O	P22	P26	K3	J3	245
I/O	P23	P27	K1	J4	248
I/O	P24	P28	L1	K1	251