AMD Xilinx - XCS30XL-5PQ208C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	576
Number of Logic Elements/Cells	1368
Total RAM Bits	18432
Number of I/O	169
Number of Gates	30000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcs30xl-5pq208c

General Overview

Spartan series FPGAs are implemented with a regular, flexible, programmable architecture of Configurable Logic Blocks (CLBs), interconnected by a powerful hierarchy of versatile routing resources (routing channels), and surrounded by a perimeter of programmable Input/Output Blocks (IOBs), as seen in Figure 1. They have generous routing resources to accommodate the most complex interconnect patterns.

The devices are customized by loading configuration data into internal static memory cells. Re-programming is possible an unlimited number of times. The values stored in these memory cells determine the logic functions and interconnections implemented in the FPGA. The FPGA can either actively read its configuration data from an external serial PROM (Master Serial mode), or the configuration data can be written into the FPGA from an external device (Slave Serial mode).

Spartan series FPGAs can be used where hardware must be adapted to different user applications. FPGAs are ideal for shortening design and development cycles, and also offer a cost-effective solution for production rates well beyond 50,000 systems per month.

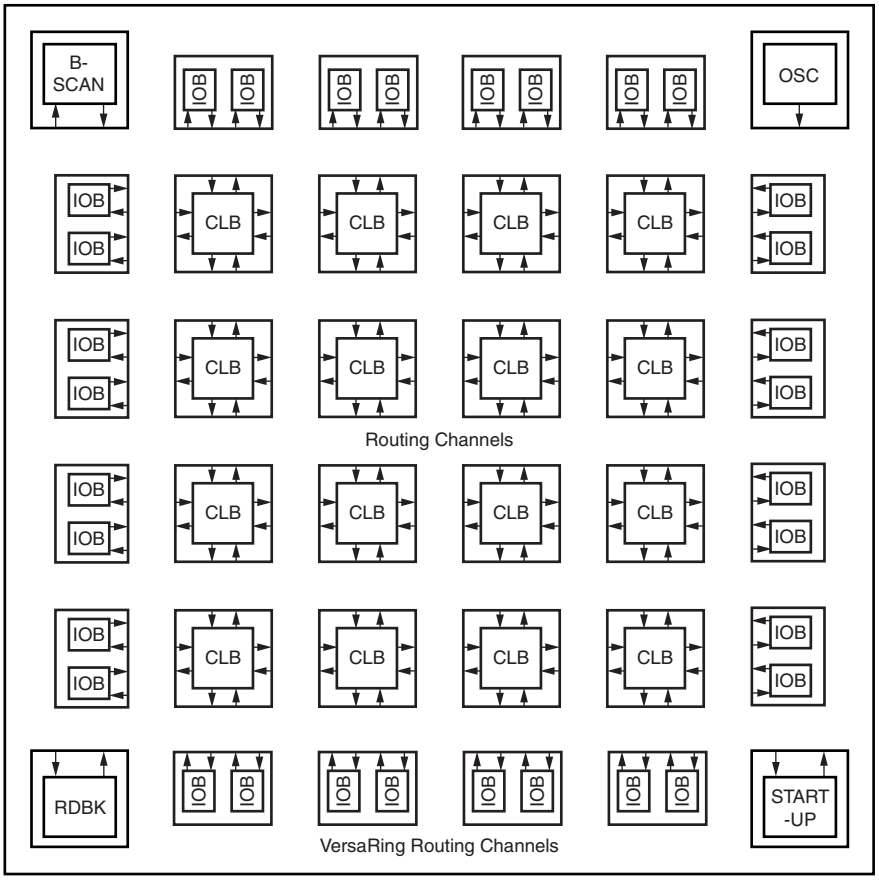


Figure 1: **Basic FPGA Block Diagram**

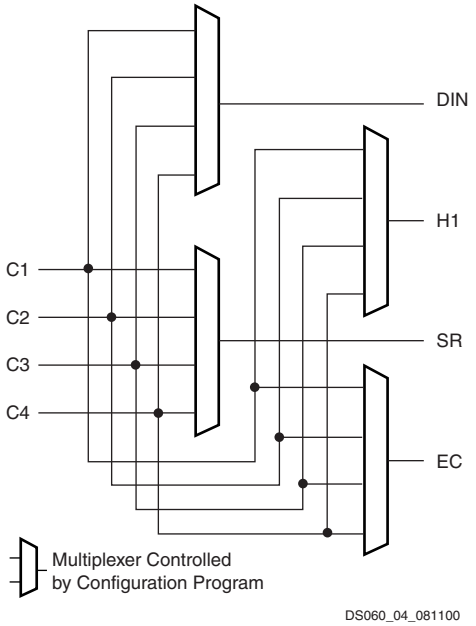


Figure 4: **CLB Control Signal Interface**

The four internal control signals are:

- EC: Enable Clock
- SR: Asynchronous Set/Reset or H function generator Input 0
- DIN: Direct In or H function generator Input 2
- H1: H function generator Input 1.

Input/Output Blocks (IOBs)

User-configurable input/output blocks (IOBs) provide the interface between external package pins and the internal logic. Each IOB controls one package pin and can be configured for input, output, or bidirectional signals. Figure 6 shows a simplified functional block diagram of the Spartan/XL FPGA IOB.

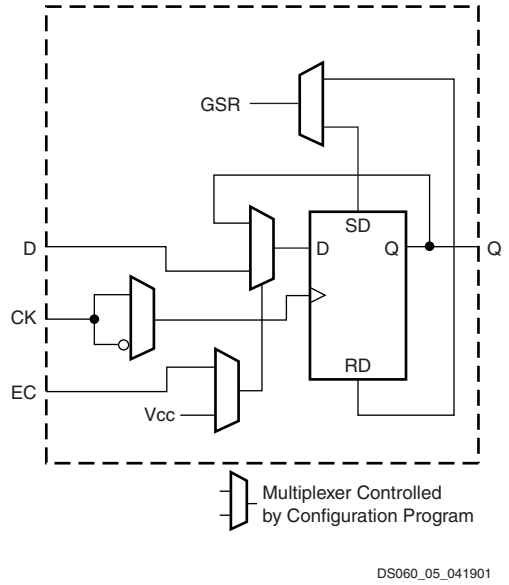


Figure 5: **IOB Flip-Flop/Latch Functional Block Diagram**

IOB Input Signal Path

The input signal to the IOB can be configured to either go directly to the routing channels (via I1 and I2 in Figure 6) or to the input register. The input register can be programmed as either an edge-triggered flip-flop or a level-sensitive latch. The functionality of this register is shown in Table 3, and a simplified block diagram of the register can be seen in Figure 5.

Table 3: **Input Register Functionality**

Mode	CK	EC	D	Q
Power-Up or GSR	X	X	X	SR
Flip-Flop	⌐	1*	D	D
	0	X	X	Q
Latch	1	1*	X	Q
	0	1*	D	D
Both	X	0	X	Q

Legend:

X	Don't care.
⌐	Rising edge (clock not inverted).
SR	Set or Reset value. Reset is default.
0*	Input is Low or unconnected (default value)
1*	Input is High or unconnected (default value)

The register choice is made by placing the appropriate library symbol. For example, IFD is the basic input flip-flop (rising edge triggered), and ILD is the basic input latch (transparent-High). Variations with inverted clocks are also available. The clock signal inverter is also shown in Figure 5 on the CK line.

The Spartan family IOB data input path has a one-tap delay element: either the delay is inserted (default), or it is not. The Spartan-XL family IOB data input path has a two-tap delay element, with choices of a full delay, a partial delay, or no delay. The added delay guarantees a zero hold time with respect to clocks routed through the global clock buffers. (See **Global Nets and Buffers**, page 12 for a description of the global clock buffers in the Spartan/XL families.) For a shorter input register setup time, with positive hold-time, attach a NODELAY attribute or property to the flip-flop.The output of the input register goes to the routing channels (via I1 and I2 in Figure 6). The I1 and I2 signals that exit the IOB can each carry either the direct or registered input signal.

The 5V Spartan family input buffers can be globally configured for either TTL (1.2V) or CMOS (VCC/2) thresholds, using an option in the bitstream generation software. The Spartan family output levels are also configurable; the two global adjustments of input threshold and output level are independent. The inputs of Spartan devices can be driven by the outputs of any 3.3V device, if the Spartan family inputs are in TTL mode. Input and output thresholds are TTL on all configuration pins until the configuration has been loaded into the device and specifies how they are to be used. Spartan-XL family inputs are TTL compatible and 3.3V CMOS compatible.

Supported sources for Spartan/XL device inputs are shown in Table 4.

Spartan-XL family I/Os are fully 5V tolerant even though the V_{CC} is 3.3V. This allows 5V signals to directly connect to the Spartan-XL family inputs without damage, as shown in Table 4. In addition, the 3.3V V_{CC} can be applied before or after 5V signals are applied to the I/Os. This makes the Spartan-XL devices immune to power supply sequencing problems.

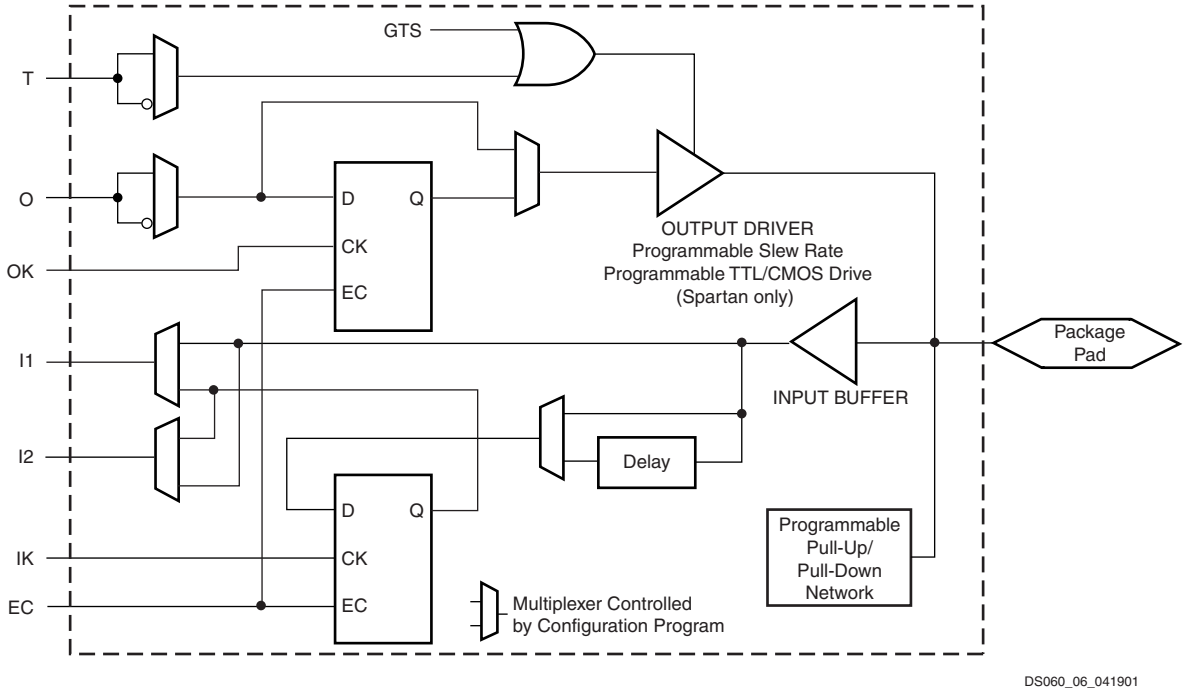


Figure 6: **Simplified Spartan/XL IOB Block Diagram**

Output Multiplexer/2-Input Function Generator (Spartan-XL Family Only)

The output path in the Spartan-XL family IOB contains an additional multiplexer not available in the Spartan family IOB. The multiplexer can also be configured as a 2-input function generator, implementing a pass gate, AND gate, OR gate, or XOR gate, with 0, 1, or 2 inverted inputs.

When configured as a multiplexer, this feature allows two output signals to time-share the same output pad, effectively doubling the number of device outputs without requiring a larger, more expensive package. The select input is the pin used for the output flip-flop clock, OK.

When the multiplexer is configured as a 2-input function generator, logic can be implemented within the IOB itself. Combined with a Global buffer, this arrangement allows very high-speed gating of a single signal. For example, a wide decoder can be implemented in CLBs, and its output gated with a Read or Write Strobe driven by a global buffer.

The user can specify that the IOB function generator be used by placing special library symbols beginning with the letter "O." For example, a 2-input AND gate in the IOB function generator is called OAND2. Use the symbol input pin labeled "F" for the signal on the critical path. This signal is placed on the OK pin — the IOB input with the shortest delay to the function generator. Two examples are shown in Figure 7.

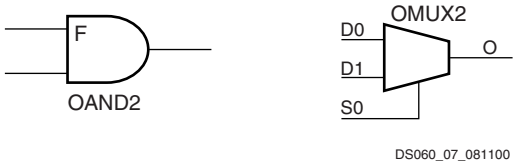


Figure 7: **AND and MUX Symbols in Spartan-XL IOB**

Output Buffer

An active High 3-state signal can be used to place the output buffer in a high-impedance state, implementing 3-state outputs or bidirectional I/O. Under configuration control, the output (O) and output 3-state (T) signals can be inverted. The polarity of these signals is independently configured for each IOB (see Figure 6, page 7). An output can be configured as open-drain (open-collector) by tying the 3-state pin (T) to the output signal, and the input pin (I) to Ground.

By default, a 5V Spartan device output buffer pull-up structure is configured as a TTL-like totem-pole. The High driver is an n-channel pull-up transistor, pulling to a voltage one transistor threshold below V_{CC}. Alternatively, the outputs can be globally configured as CMOS drivers, with additional p-channel pull-up transistors pulling to V_{CC}. This option, applied using the bitstream generation software, applies to all outputs on the device. It is not individually programmable.

All Spartan-XL device outputs are configured as CMOS drivers, therefore driving rail-to-rail. The Spartan-XL family outputs are individually programmable for 12 mA or 24 mA output drive.

Any 5V Spartan device with its outputs configured in TTL mode can drive the inputs of any typical 3.3V device. Supported destinations for Spartan/XL device outputs are shown in Table 7.

Three-State Register (Spartan-XL Family Only)

Spartan-XL devices incorporate an optional register controlling the three-state enable in the IOBs. The use of the three-state control register can significantly improve output enable and disable time.

Output Slew Rate

The slew rate of each output buffer is, by default, reduced, to minimize power bus transients when switching non-critical signals. For critical signals, attach a FAST attribute or property to the output buffer or flip-flop.

Spartan/XL devices have a feature called "Soft Start-up," designed to reduce ground bounce when all outputs are turned on simultaneously at the end of configuration. When the configuration process is finished and the device starts up, the first activation of the outputs is automatically slew-rate limited. Immediately following the initial activation of the I/O, the slew rate of the individual outputs is determined by the individual configuration option for each IOB.

Pull-up and Pull-down Network

Programmable pull-up and pull-down resistors are used for tying unused pins to V_{CC} or Ground to minimize power consumption and reduce noise sensitivity. The configurable pull-up resistor is a p-channel transistor that pulls to V_{CC}. The configurable pull-down resistor is an n-channel transistor that pulls to Ground. The value of these resistors is typically 20 KΩ – 100 KΩ (See "Spartan Family DC Characteristics Over Operating Conditions" on page 43.).

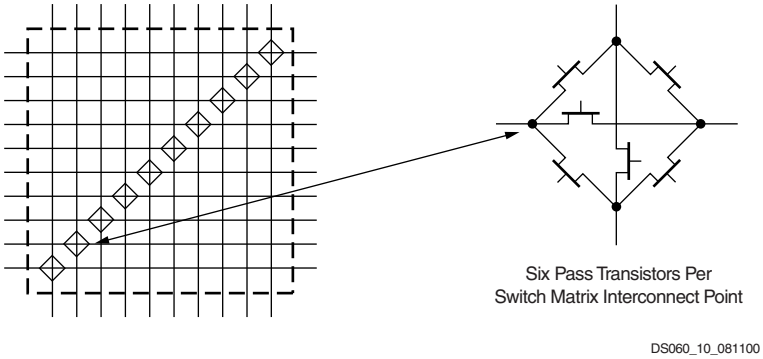


Figure 10: **Programmable Switch Matrix**

Double-Length Lines

The double-length lines consist of a grid of metal segments, each twice as long as the single-length lines: they run past two CLBs before entering a PSM. Double-length lines are grouped in pairs with the PSMs staggered, so that each line goes through a PSM at every other row or column of CLBs (see Figure 8).

There are four vertical and four horizontal double-length lines associated with each CLB. These lines provide faster signal routing over intermediate distances, while retaining routing flexibility.

Longlines

Longlines form a grid of metal interconnect segments that run the entire length or width of the array. Longlines are intended for high fan-out, time-critical signal nets, or nets that are distributed over long distances.

Each Spartan/XL device longline has a programmable splitter switch at its center. This switch can separate the line into two independent routing channels, each running half the width or height of the array.

Routing connectivity of the longlines is shown in Figure 8. The longlines also interface to some 3-state buffers which is described later in **3-State Long Line Drivers**, page 19.

I/O Routing

Spartan/XL devices have additional routing around the IOB ring. This routing is called a VersaRing. The VersaRing facilitates pin-swapping and redesign without affecting board layout. Included are eight double-length lines, and four longlines.

Global Nets and Buffers

The Spartan/XL devices have dedicated global networks. These networks are designed to distribute clocks and other high fanout control signals throughout the devices with minimal skew.

Four vertical longlines in each CLB column are driven exclusively by special global buffers. These longlines are in addition to the vertical longlines used for standard interconnect. In the 5V Spartan devices, the four global lines can be driven by either of two types of global buffers; Primary Global buffers (BUFGP) or Secondary Global buffers (BUFGS). Each of these lines can be accessed by one particular Primary Global buffer, or by any of the Secondary Global buffers, as shown in Figure 11. In the 3V Spartan-XL devices, the four global lines can be driven by any of the eight Global Low-Skew Buffers (BUFGLS). The clock pins of every CLB and IOB can also be sourced from local interconnect.

- The 16 x 1 single-port configuration contains a RAM array with 16 locations, each one-bit wide. One 4-bit address decoder determines the RAM location for write and read operations. There is one input for writing data and one output for reading data, all at the selected address.
- The (16 x 1) x 2 single-port configuration combines two 16 x 1 single-port configurations (each according to the preceding description). There is one data input, one data output and one address decoder for each array. These arrays can be addressed independently.
- The 32 x 1 single-port configuration contains a RAM array with 32 locations, each one-bit wide. There is one data input, one data output, and one 5-bit address decoder.
- The dual-port mode 16 x 1 configuration contains a RAM array with 16 locations, each one-bit wide. There are two 4-bit address decoders, one for each port. One port consists of an input for writing and an output for reading, all at a selected address. The other port consists of one output for reading from an independently selected address.

The appropriate choice of RAM configuration mode for a given design should be based on timing and resource requirements, desired functionality, and the simplicity of the design process. Selection criteria include the following: Whereas the 32 x 1 single-port, the (16 x 1) x 2 single-port, and the 16 x 1 dual-port configurations each use one entire CLB, the 16 x 1 single-port configuration uses only one half of a CLB. Due to its simultaneous read/write capability, the dual-port RAM can transfer twice as much data as the single-port RAM, which permits only one data operation at any given time.

CLB memory configuration options are selected by using the appropriate library symbol in the design entry.

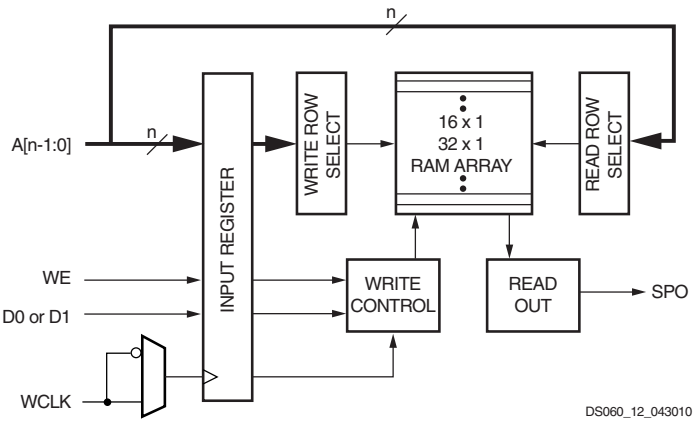
Single-Port Mode

There are three CLB memory configurations for the single-port RAM: 16 x 1, (16 x 1) x 2, and 32 x 1, the functional organization of which is shown in Figure 12.

The single-port RAM signals and the CLB signals (Figure 2, page 4) from which they are originally derived are shown in Table 9.

Table 9: **Single-Port RAM Signals**

RAM Signal	Function	CLB Signal
D0 or D1	Data In	DIN or H1
A[3:0]	Address	F[4:1] or G[4:1]
A4 (32 x 1 only)	Address	H1
WE	Write Enable	SR
WCLK	Clock	K
SPO	Single Port Out (Data Out)	F_{OUT} or G_{OUT}



Notes:
1. The (16 x 1) x 2 configuration combines two 16 x 1 single-port RAMs, each with its own independent address bus and data input. The same WE and WCLK signals are connected to both RAMs.
2. n = 4 for the 16 x 1 and (16 x 1) x 2 configurations. n = 5 for the 32 x 1 configuration.

Figure 12: **Logic Diagram for the Single-Port RAM**

Writing data to the single-port RAM is essentially the same as writing to a data register. It is an edge-triggered (synchronous) operation performed by applying an address to the A inputs and data to the D input during the active edge of WCLK while WE is High.

The timing relationships are shown in Figure 13. The High logic level on WE enables the input data register for writing. The active edge of WCLK latches the address, input data, and WE signals. Then, an internal write pulse is generated that loads the data into the memory cell.

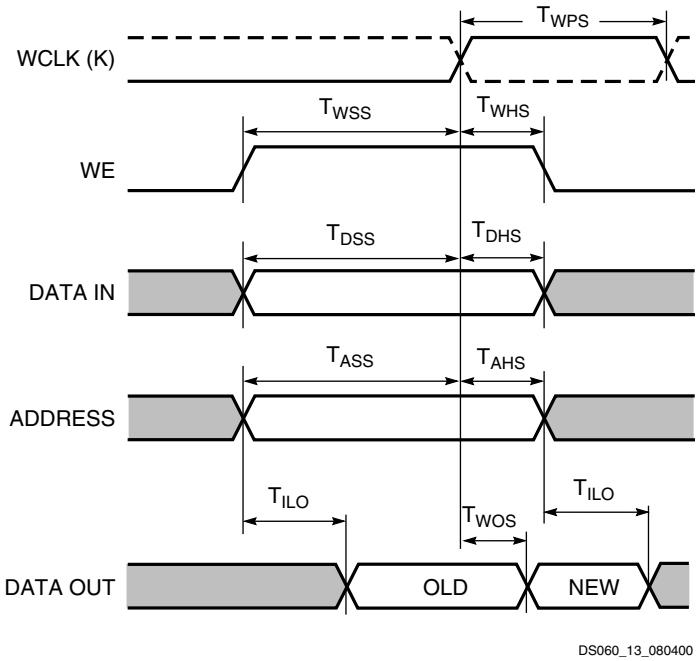


Figure 13: **Data Write and Access Timing for RAM**

WCLK can be configured as active on either the rising edge (default) or the falling edge. While the WCLK input to the RAM accepts the same signal as the clock input to the associated CLB's flip-flops, the sense of this WCLK input can be inverted with respect to the sense of the flip-flop clock inputs. Consequently, within the same CLB, data at the RAM SPO line can be stored in a flip-flop with either the same or the inverse clock polarity used to write data to the RAM.

The WE input is active High and cannot be inverted within the CLB.

Allowing for settling time, the data on the SPO output reflects the contents of the RAM location currently addressed. When the address changes, following the asynchronous delay T_{ILO}, the data stored at the new address location will appear on SPO. If the data at a particular RAM address is overwritten, after the delay T_{WOS}, the new data will appear on SPO.

Dual-Port Mode

In dual-port mode, the function generators (F-LUT and G-LUT) are used to create a 16 x 1 dual-port memory. Of the two data ports available, one permits read and write operations at the address specified by A[3:0] while the second provides only for read operations at the address specified independently by DPRA[3:0]. As a result, simultaneous read/write operations at different addresses (or even at the same address) are supported.

The functional organization of the 16 x 1 dual-port RAM is shown in Figure 14. The dual-port RAM signals and the

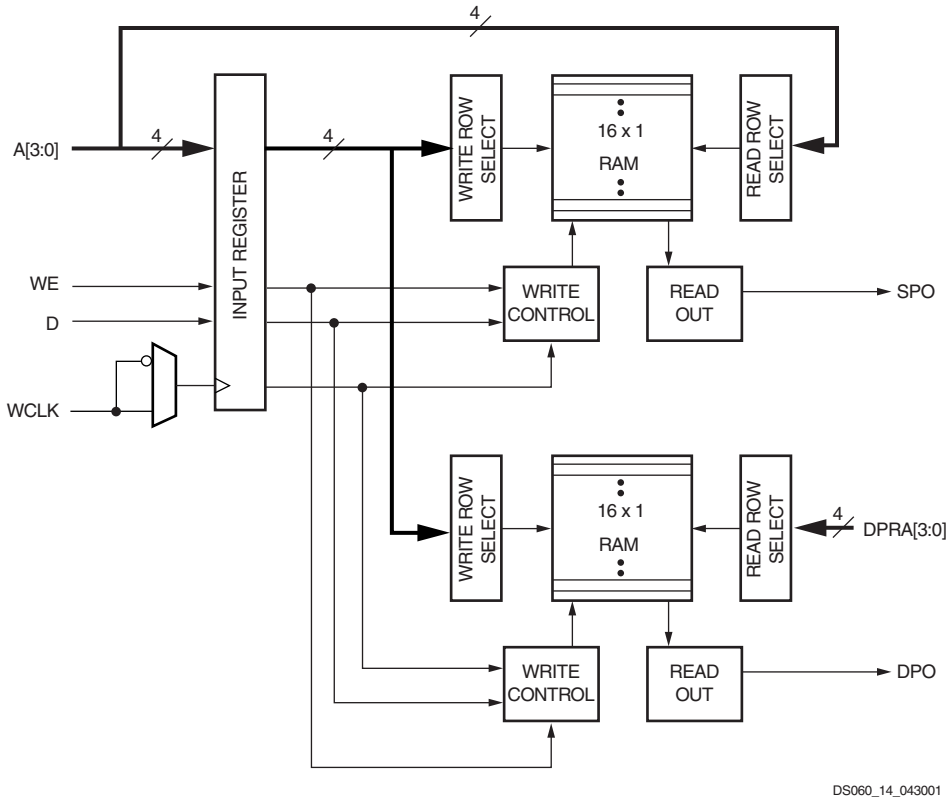


Figure 14: **Logic Diagram for the Dual-Port RAM**

CLB signals from which they are originally derived are shown in Table 10.

Table 10: **Dual-Port RAM Signals**

RAM Signal	Function	CLB Signal
D	Data In	DIN
A[3:0]	Read Address for Single-Port. Write Address for Single-Port and Dual-Port.	F[4:1]
DPRA[3:0]	Read Address for Dual-Port	G[4:1]
WE	Write Enable	SR
WCLK	Clock	K
SPO	Single Port Out (addressed by A[3:0])	F_{OUT}
DPO	Dual Port Out (addressed by DPRA[3:0])	G_{OUT}

The RAM16X1D primitive used to instantiate the dual-port RAM consists of an upper and a lower 16 x 1 memory array. The address port labeled A[3:0] supplies both the read and write addresses for the lower memory array, which behaves the same as the 16 x 1 single-port RAM array described previously. Single Port Out (SPO) serves as the data output for the lower memory. Therefore, SPO reflects the data at address A[3:0].

The other address port, labeled DPRA[3:0] for Dual Port Read Address, supplies the read address for the upper memory. The write address for this memory, however, comes from the address A[3:0]. Dual Port Out (DPO) serves as the data output for the upper memory. Therefore, DPO reflects the data at address DPRA[3:0].

By using A[3:0] for the write address and DPRA[3:0] for the read address, and reading only the DPO output, a FIFO that can read and write simultaneously is easily generated. The simultaneous read/write capability possible with the dual-port RAM can provide twice the effective data throughput of a single-port RAM alternating read and write operations.

The timing relationships for the dual-port RAM mode are shown in Figure 13.

Note that write operations to RAM are synchronous (edge-triggered); however, data access is asynchronous.

Initializing RAM at FPGA Configuration

Both RAM and ROM implementations in the Spartan/XL families are initialized during device configuration. The initial contents are defined via an INIT attribute or property attached to the RAM or ROM symbol, as described in the library guide. If not defined, all RAM contents are initialized to zeros, by default.

RAM initialization occurs only during device configuration. The RAM content is not affected by GSR.

More Information on Using RAM Inside CLBs

Three application notes are available from Xilinx that discuss synchronous (edge-triggered) RAM: *"Xilinx Edge-Triggered and Dual-Port RAM Capability," "Implementing FIFOs in Xilinx RAM,"* and *"Synchronous and Asynchronous FIFO Designs."* All three application notes apply to both the Spartan and the Spartan-XL families.

Fast Carry Logic

Each CLB F-LUT and G-LUT contains dedicated arithmetic logic for the fast generation of carry and borrow signals. This extra output is passed on to the function generator in the adjacent CLB. The carry chain is independent of normal routing resources. (See Figure 15.)

Dedicated fast carry logic greatly increases the efficiency and performance of adders, subtractors, accumulators, comparators and counters. It also opens the door to many new applications involving arithmetic operation, where the previous generations of FPGAs were not fast enough or too inefficient. High-speed address offset calculations in microprocessor or graphics systems, and high-speed addition in digital signal processing are two typical applications.

The two 4-input function generators can be configured as a 2-bit adder with built-in hidden carry that can be expanded to any length. This dedicated carry circuitry is so fast and efficient that conventional speed-up methods like carry generate/propagate are meaningless even at the 16-bit level, and of marginal benefit at the 32-bit level. This fast carry logic is one of the more significant features of the Spartan

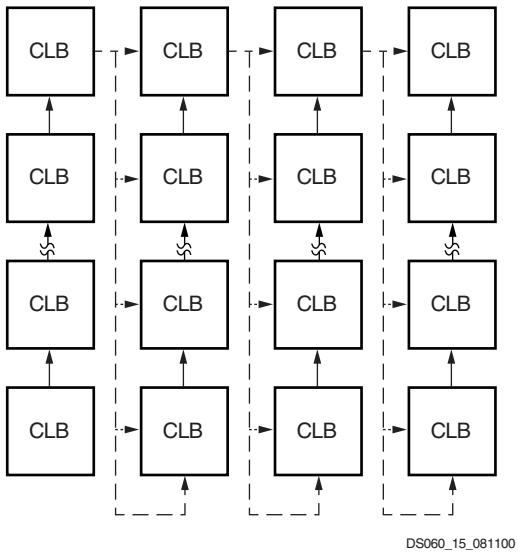


Figure 15: **Available Spartan/XL Carry Propagation Paths**

On-Chip Oscillator

Spartan/XL devices include an internal oscillator. This oscillator is used to clock the power-on time-out, for configuration memory clearing, and as the source of CCLK in Master configuration mode. The oscillator runs at a nominal 8 MHz frequency that varies with process, V_{CC}, and temperature. The output frequency falls between 4 MHz and 10 MHz.

The oscillator output is optionally available after configuration. Any two of four resynchronized taps of a built-in divider are also available. These taps are at the fourth, ninth, fourteenth and nineteenth bits of the divider. Therefore, if the primary oscillator output is running at the nominal 8 MHz, the user has access to an 8-MHz clock, plus any two of 500 kHz, 16 kHz, 490 Hz and 15 Hz. These frequencies can vary by as much as -50% or +25%.

These signals can be accessed by placing the OSC4 library element in a schematic or in HDL code. The oscillator is automatically disabled after configuration if the OSC4 symbol is not used in the design.

Global Signals: GSR and GTS

Global Set/Reset

A separate Global Set/Reset line, as shown in Figure 3, page 5 for the CLB and Figure 5, page 6 for the IOB, sets or clears each flip-flop during power-up, reconfiguration, or when a dedicated Reset net is driven active. This global net (GSR) does not compete with other routing resources; it uses a dedicated distribution network.

Each flip-flop is configured as either globally set or reset in the same way that the local set/reset (SR) is specified. Therefore, if a flip-flop is set by SR, it is also set by GSR. Similarly, if in reset mode, it is reset by both SR and GSR.

GSR can be driven from any user-programmable pin as a global reset input. To use this global net, place an input pad and input buffer in the schematic or HDL code, driving the GSR pin of the STARTUP symbol. (See Figure 19.) A specific pin location can be assigned to this input using a LOC attribute or property, just as with any other user-programmable pad. An inverter can optionally be inserted after the input buffer to invert the sense of the GSR signal. Alternatively, GSR can be driven from any internal node.

Global 3-State

A separate Global 3-state line (GTS) as shown in Figure 6, page 7 forces all FPGA outputs to the high-impedance state, unless boundary scan is enabled and is executing an EXTEST instruction. GTS does not compete with other routing resources; it uses a dedicated distribution network.

GTS can be driven from any user-programmable pin as a global 3-state input. To use this global net, place an input pad and input buffer in the schematic or HDL code, driving the GTS pin of the STARTUP symbol. This is similar to what is shown in Figure 19 for GSR except the IBUF would be connected to GTS. A specific pin location can be assigned to this input using a LOC attribute or property, just as with any other user-programmable pad. An inverter can optionally be inserted after the input buffer to invert the sense of the Global 3-state signal. Alternatively, GTS can be driven from any internal node.

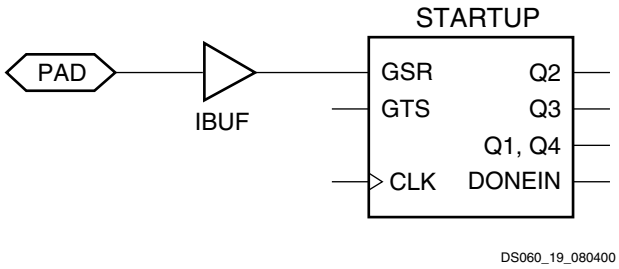


Figure 19: **Symbols for Global Set/Reset**

Boundary Scan

The "bed of nails" has been the traditional method of testing electronic assemblies. This approach has become less appropriate, due to closer pin spacing and more sophisticated assembly methods like surface-mount technology and multi-layer boards. The IEEE Boundary Scan Standard 1149.1 was developed to facilitate board-level testing of electronic assemblies. Design and test engineers can embed a standard test logic structure in their device to achieve high fault coverage for I/O and internal logic. This structure is easily implemented with a four-pin interface on any boundary scan compatible device. IEEE 1149.1-compatible devices may be serial daisy-chained together, connected in parallel, or a combination of the two.

The Spartan and Spartan-XL families implement IEEE 1149.1-compatible BYPASS, PRELOAD/SAMPLE and EXTEST boundary scan instructions. When the boundary scan configuration option is selected, three normal user I/O pins become dedicated inputs for these functions. Another user output pin becomes the dedicated boundary scan output. The details of how to enable this circuitry are covered later in this section.

By exercising these input signals, the user can serially load commands and data into these devices to control the driving of their outputs and to examine their inputs. This method is an improvement over bed-of-nails testing. It avoids the need to over-drive device outputs, and it reduces the user interface to four pins. An optional fifth pin, a reset for the control logic, is described in the standard but is not implemented in the Spartan/XL devices.

The dedicated on-chip logic implementing the IEEE 1149.1 functions includes a 16-state machine, an instruction register and a number of data registers. The functional details can be found in the IEEE 1149.1 specification and are also discussed in the Xilinx application note: "*Boundary Scan in FPGA Devices*."

to the DONE pin. User I/Os for each device become active after the DONE pin for that device goes High. (The exact timing is determined by development system options.) Since the DONE pin is open-drain and does not drive a High value, tying the DONE pins of all devices together prevents all devices in the chain from going High until the last device in the chain has completed its configuration cycle. If the DONE pin of a device is left unconnected, the device becomes active as soon as that device has been configured. Only devices supporting Express mode can be used to form an Express mode daisy chain.

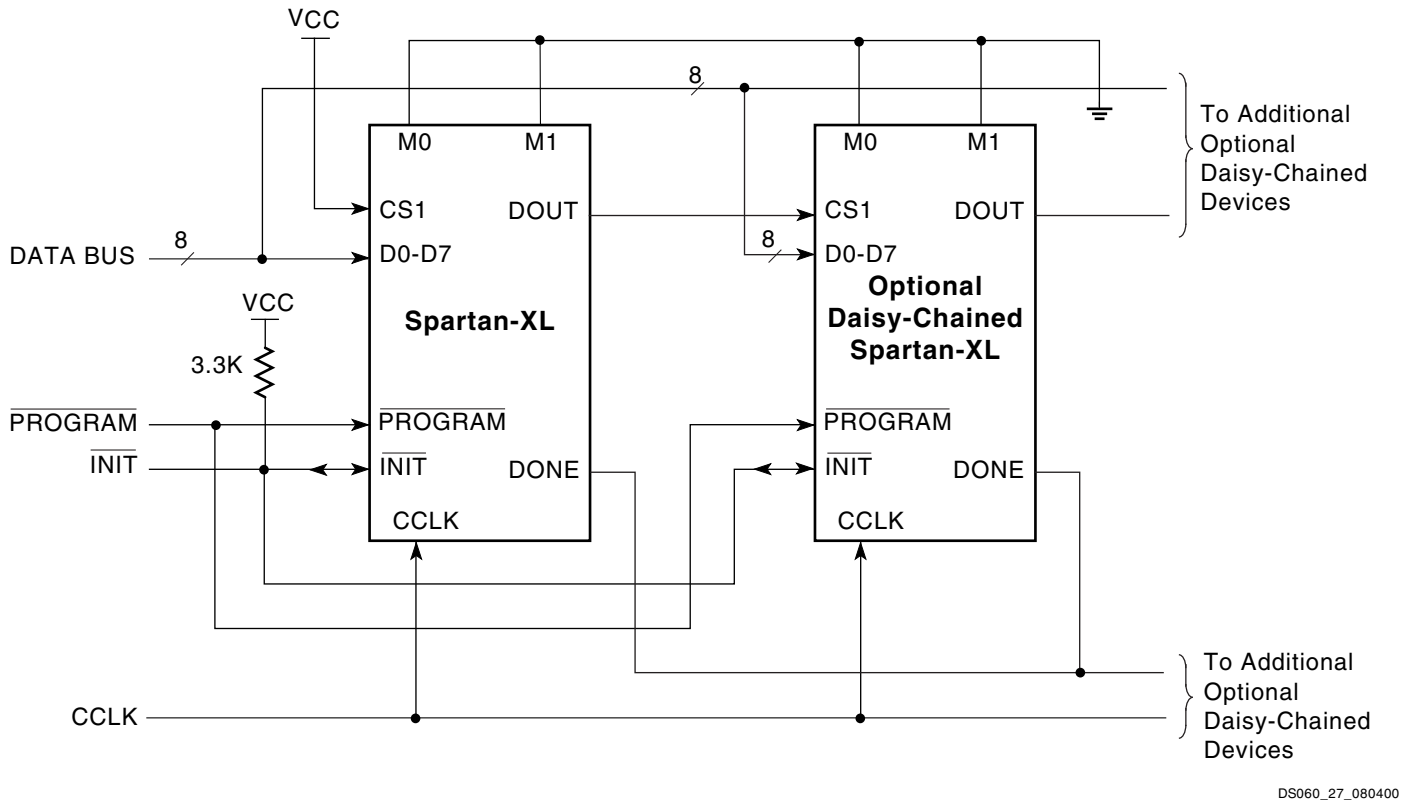


Figure 27: **Express Mode Circuit Diagram**

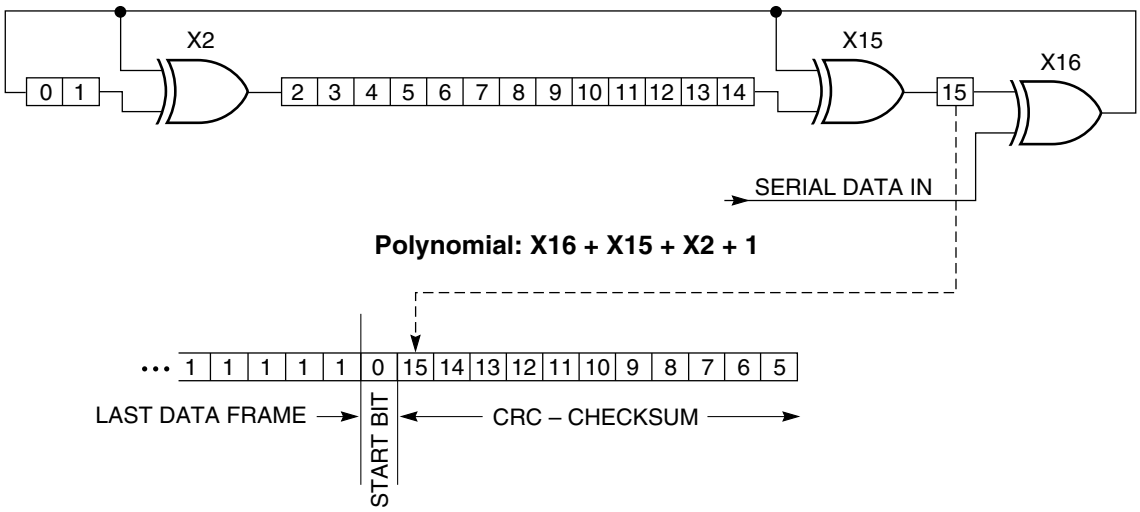


Figure 29: **Circuit for Generating CRC-16**

Configuration Sequence

There are four major steps in the Spartan/XL FPGA power-up configuration sequence.

- Configuration Memory Clear
- Initialization
- Configuration
- Start-up

The full process is illustrated in Figure 30.

Configuration Memory Clear

When power is first applied or is reapplied to an FPGA, an internal circuit forces initialization of the configuration logic. When V_{CC} reaches an operational level, and the circuit passes the write and read test of a sample pair of configuration bits, a time delay is started. This time delay is nominally 16 ms. The delay is four times as long when in Master Serial Mode to allow ample time for all slaves to reach a stable V_{CC}. When all $\overline{INIT}$ pins are tied together, as recommended, the longest delay takes precedence. Therefore, devices with different time delays can easily be mixed and matched in a daisy chain.

This delay is applied only on power-up. It is not applied when reconfiguring an FPGA by pulsing the $\overline{PROGRAM}$ pin Low. During this time delay, or as long as the $\overline{PROGRAM}$ input is asserted, the configuration logic is held in a Configuration Memory Clear state. The configuration-memory frames are consecutively initialized, using the internal oscillator.

At the end of each complete pass through the frame addressing, the power-on time-out delay circuitry and the level of the $\overline{PROGRAM}$ pin are tested. If neither is asserted, the logic initiates one additional clearing of the configuration frames and then tests the $\overline{INIT}$ input.

Initialization

During initialization and configuration, user pins HDC, $\overline{LDC}$, $\overline{INIT}$ and DONE provide status outputs for the system interface. The outputs $\overline{LDC}$, $\overline{INIT}$ and DONE are held Low and HDC is held High starting at the initial application of power.

The open drain $\overline{INIT}$ pin is released after the final initialization pass through the frame addresses. There is a deliberate delay before a Master-mode device recognizes an inactive $\overline{INIT}$. Two internal clocks after the $\overline{INIT}$ pin is recognized as High, the device samples the MODE pin to determine the configuration mode. The appropriate interface lines become active and the configuration preamble and data can be loaded.

Capacitive Load Factor

Figure 34 shows the relationship between I/O output delay and load capacitance. It allows a user to adjust the specified output delay if the load capacitance is different than 50 pF. For example, if the actual load capacitance is 120 pF, add 2.5 ns to the specified delay. If the load capacitance is 20 pF, subtract 0.8 ns from the specified output delay. Figure 34 is usable over the specified operating conditions of voltage and temperature and is independent of the output slew rate control.

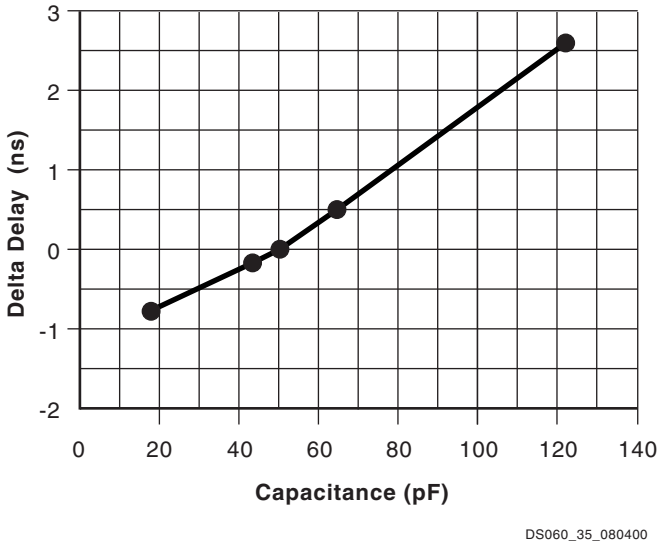


Figure 34: **Delay Factor at Various Capacitive Loads**

Spartan-XL Family DC Characteristics Over Operating Conditions

Symbol	Description		Min	Typ.	Max	Units
V_{OH}	High-level output voltage @ I_{OH} = –4.0 mA, V_{CC} min (LVTTL)		2.4	-	-	V
	High-level output voltage @ I_{OH} = –500 µA, (LVCMOS)		90% V_{CC}	-	-	V
V_{OL}	Low-level output voltage @ I_{OL} = 12.0 mA, V_{CC} min (LVTTL)(1)		-	-	0.4	V
	Low-level output voltage @ I_{OL} = 24.0 mA, V_{CC} min (LVTTL)(2)		-	-	0.4	V
	Low-level output voltage @ I_{OL} = 1500 µA, (LVCMOS)		-	-	10% V_{CC}	V
V_{DR}	Data retention supply voltage (below which configuration data may be lost)		2.5	-	-	V
I_{CCO}	Quiescent FPGA supply current(3,4)	Commercial	-	0.1	2.5	mA
		Industrial	-	0.1	5	mA
I_{CCPD}	Power Down FPGA supply current(3,5)	Commercial	-	0.1	2.5	mA
		Industrial	-	0.1	5	mA
I_L	Input or output leakage current		–10	-	10	µA
C_{IN}	Input capacitance (sample tested)		-	-	10	pF
I_{RPU}	Pad pull-up (when selected) @ V_{IN} = 0V (sample tested)		0.02	-	0.25	mA
I_{RPD}	Pad pull-down (when selected) @ V_{IN} = 3.3V (sample tested)		0.02	-	-	mA

Notes:
1. With up to 64 pins simultaneously sinking 12 mA (default mode).
2. With up to 64 pins simultaneously sinking 24 mA (with 24 mA option selected).
3. With 5V tolerance not selected, no internal oscillators, and the FPGA configured with the Tie option.
4. With no output current loads, no active input resistors, and all package pins at V_{CC} or GND.
5. With $\overline{PWRDWN}$ active.

Supply Current Requirements During Power-On

Spartan-XL FPGAs require that a minimum supply current I_{CCPO} be provided to the V_{CC} lines for a successful power on. If more current is available, the FPGA can consume more than I_{CCPO} min., though this cannot adversely affect reliability.

A maximum limit for I_{CCPO} is not specified. Be careful when using foldback/crowbar supplies and fuses. It is possible to control the magnitude of I_{CCPO} by limiting the supply current available to the FPGA. A current limit below the trip level will avoid inadvertently activating over-current protection circuits.

Symbol	Description	Min	Max	Units
I_{CCPO}	Total V_{CC} supply current required during power-on	100	-	mA
T_{CCPO}	V_{CC} ramp time(2,3)	-	50	ms

Notes:
1. The I_{CCPO} requirement applies for a brief time (commonly only a few milliseconds) when V_{CC} ramps from 0 to 3.3V.
2. The ramp time is measured from GND to V_{CC} max on a fully loaded board.
3. V_{CC} must not dip in the negative direction during power on.

Spartan-XL Family Global Buffer Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.

When fewer vertical clock lines are connected, the clock distribution is faster; when multiple clock lines per column are driven from the same global clock, the delay is longer. For more specific, more precise, and worst-case guaranteed data, reflecting the actual routing structure, use the values provided by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. These path delays, provided as a guideline, have been extracted from the static timing analyzer report. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature).

Symbol	Description	Device	Speed Grade		Units
			-5	-4	
			Max	Max	
T_{GLS}	From pad through buffer, to any clock K	XCS05XL	1.4	1.5	ns
		XCS10XL	1.7	1.8	ns
		XCS20XL	2.0	2.1	ns
		XCS30XL	2.3	2.5	ns
		XCS40XL	2.6	2.8	ns

Table 18: **Pin Descriptions *(Continued)***

Pin Name	I/O During Config.	I/O After Config.	Pin Description
SGCK1 - SGCK4 (Spartan)	Weak Pull-up (except SGCK4 is DOUT)	I or I/O	Four Secondary Global inputs each drive a dedicated internal global net with short delay and minimal skew. These internal global nets can also be driven from internal logic. If not used to drive a global net, any of these pins is a user-programmable I/O pin. The SGCK1-SGCK4 pins provide the shortest path to the four Secondary Global Buffers. Any input pad symbol connected directly to the input of a BUFGS symbol is automatically placed on one of these pins.
GCK1 - GCK8 (Spartan-XL)	Weak Pull-up (except GCK6 is DOUT)	I or I/O	Eight Global inputs each drive a dedicated internal global net with short delay and minimal skew. These internal global nets can also be driven from internal logic. If not used to drive a global net, any of these pins is a user-programmable I/O pin. The GCK1-GCK8 pins provide the shortest path to the eight Global Low-Skew Buffers. Any input pad symbol connected directly to the input of a BUFGLS symbol is automatically placed on one of these pins.
CS1 (Spartan-XL)	I	I/O	During Express configuration, CS1 is used as a serial-enable signal for daisy-chaining.
D0-D7 (Spartan-XL)	I	I/O	During Express configuration, these eight input pins receive configuration data. After configuration, they are user-programmable I/O pins.
DIN	I	I/O	During Slave Serial or Master Serial configuration, DIN is the serial configuration data input receiving data on the rising edge of CCLK. After configuration, DIN is a user-programmable I/O pin.
DOUT	O	I/O	During Slave Serial or Master Serial configuration, DOUT is the serial configuration data output that can drive the DIN of daisy-chained slave FPGAs. DOUT data changes on the falling edge of CCLK, one-and-a-half CCLK periods after it was received at the DIN input. In Spartan-XL family Express mode, DOUT is the status output that can drive the CS1 of daisy-chained FPGAs, to enable and disable downstream devices. After configuration, DOUT is a user-programmable I/O pin.
Unrestricted User-Programmable I/O Pins			
I/O	Weak Pull-up	I/O	These pins can be configured to be input and/or output after configuration is completed. Before configuration is completed, these pins have an internal high-value pull-up resistor network that defines the logic level as High.

XCS05 and XCS05XL Device Pinouts

XCS05/XL Pad Name	PC84[4]	VQ100	Bndry Scan
I/O	P70	P71	238[3]
I/O (D0[2], DIN)	P71	P72	241[3]
I/O, SGCK4[1], GCK6[2] (DOUT)	P72	P73	244[3]
CCLK	P73	P74	-
VCC	P74	P75	-
O, TDO	P75	P76	0
GND	P76	P77	-
I/O	P77	P78	2
I/O, PGCK4[1], GCK7[2]	P78	P79	5
I/O (CS1[2])	P79	P80	8
I/O	P80	P81	11
I/O	P81	P82	14
I/O	P82	P83	17
I/O	-	P84	20
I/O	-	P85	23
I/O	P83	P86	26
I/O	P84	P87	29
GND	P1	P88	-

Notes:
1. 5V Spartan family only
2. 3V Spartan-XL family only
3. The "PWRDWN" on the XCS05XL is not part of the Boundary Scan chain. For the XCS05XL, subtract 1 from all Boundary Scan numbers from GCK3 on (127 and higher).
4. PC84 package discontinued by PDN2004-01

XCS10 and XCS10XL Device Pinouts

XCS10/XL Pad Name	PC84[4]	VQ100	CS144[2,4]	TQ144	Bndry Scan
VCC	P2	P89	D7	P128	-
I/O	P3	P90	A6	P129	44
I/O	P4	P91	B6	P130	47
I/O	-	P92	C6	P131	50
I/O	-	P93	D6	P132	53
I/O	P5	P94	A5	P133	56
I/O	P6	P95	B5	P134	59
I/O	-	-	C5	P135	62
I/O	-	-	D5	P136	65
GND	-	-	A4	P137	-
I/O	P7	P96	B4	P138	68
I/O	P8	P97	C4	P139	71
I/O	-	-	A3	P140	74
I/O	-	-	B3	P141	77
I/O	P9	P98	C3	P142	80

XCS10 and XCS10XL Device Pinouts

XCS10/XL Pad Name	PC84[4]	VQ100	CS144[2,4]	TQ144	Bndry Scan
I/O, SGCK1[1] GCK8[2]	P10	P99	A2	P143	83
VCC	P11	P100	B2	P144	-
GND	P12	P1	A1	P1	-
I/O, PGCK1[1] GCK1[2]	P13	P2	B1	P2	86
I/O	P14	P3	C2	P3	89
I/O	-	-	C1	P4	92
I/O	-	-	D4	P5	95
I/O, TDI	P15	P4	D3	P6	98
I/O, TCK	P16	P5	D2	P7	101
GND	-	-	D1	P8	-
I/O	-	-	E4	P9	104
I/O	-	-	E3	P10	107
I/O, TMS	P17	P6	E2	P11	110
I/O	P18	P7	E1	P12	113
I/O	-	-	F4	P13	116
I/O	-	P8	F3	P14	119
I/O	P19	P9	F2	P15	122
I/O	P20	P10	F1	P16	125
GND	P21	P11	G2	P17	-
VCC	P22	P12	G1	P18	-
I/O	P23	P13	G3	P19	128
I/O	P24	P14	G4	P20	131
I/O	-	P15	H1	P21	134
I/O	-	-	H2	P22	137
I/O	P25	P16	H3	P23	140
I/O	P26	P17	H4	P24	143
I/O	-	-	J1	P25	146
I/O	-	-	J2	P26	149
GND	-	-	J3	P27	-
I/O	P27	P18	J4	P28	152
I/O	-	P19	K1	P29	155
I/O	-	-	K2	P30	158
I/O	-	-	K3	P31	161
I/O	P28	P20	L1	P32	164
I/O, SGCK2[1] GCK2[2]	P29	P21	L2	P33	167
Not Connect-ed[1] M1[2]	P30	P22	L3	P34	170
GND	P31	P23	M1	P35	-
MODE[1], M0[2]	P32	P24	M2	P36	173

XCS10 and XCS10XL Device Pinouts

XCS10/XL Pad Name	PC84[4]	VQ100	CS144[2,4]	TQ144	Bndry Scan
VCC	P33	P25	N1	P37	-
Not Connect-ed[1] PWRDWN[2]	P34	P26	N2	P38	174 [1]
I/O, PGCK2[1] GCK3[2]	P35	P27	M3	P39	175[3]
I/O (HDC)	P36	P28	N3	P40	178 [3]
I/O	-	-	K4	P41	181 [3]
I/O	-	-	L4	P42	184 [3]
I/O	-	P29	M4	P43	187 [3]
I/O ($\overline{LDC}$)	P37	P30	N4	P44	190 [3]
GND	-	-	K5	P45	-
I/O	-	-	L5	P46	193 [3]
I/O	-	-	M5	P47	196 [3]
I/O	P38	P31	N5	P48	199 [3]
I/O	P39	P32	K6	P49	202 [3]
I/O	-	P33	L6	P50	205 [3]
I/O	-	P34	M6	P51	208 [3]
I/O	P40	P35	N6	P52	211 [3]
I/O ($\overline{INIT}$)	P41	P36	M7	P53	214 [3]
VCC	P42	P37	N7	P54	-
GND	P43	P38	L7	P55	-
I/O	P44	P39	K7	P56	217 [3]
I/O	P45	P40	N8	P57	220 [3]
I/O	-	P41	M8	P58	223 [3]
I/O	-	P42	L8	P59	226 [3]
I/O	P46	P43	K8	P60	229 [3]
I/O	P47	P44	N9	P61	232 [3]
I/O	-	-	M9	P62	235 [3]
I/O	-	-	L9	P63	238 [3]
GND	-	-	K9	P64	-
I/O	P48	P45	N10	P65	241 [3]
I/O	P49	P46	M10	P66	244 [3]
I/O	-	-	L10	P67	247 [3]
I/O	-	-	N11	P68	250 [3]
I/O	P50	P47	M11	P69	253 [3]
I/O, SGCK3[1] GCK4[2]	P51	P48	L11	P70	256 [3]
GND	P52	P49	N12	P71	-
DONE	P53	P50	M12	P72	-
VCC	P54	P51	N13	P73	-
$\overline{PROGRAM}$	P55	P52	M13	P74	-
I/O (D7[2])	P56	P53	L12	P75	259 [3]

XCS10 and XCS10XL Device Pinouts

XCS10/XL Pad Name	PC84[4]	VQ100	CS144[2,4]	TQ144	Bndry Scan
I/O, PGCK3[1] GCK5[2]	P57	P54	L13	P76	262 [3]
I/O	-	-	K10	P77	265 [3]
I/O	-	-	K11	P78	268 [3]
I/O (D6[2])	P58	P55	K12	P79	271 [3]
I/O	-	P56	K13	P80	274 [3]
GND	-	-	J10	P81	-
I/O	-	-	J11	P82	277 [3]
I/O	-	-	J12	P83	280 [3]
I/O (D5[2])	P59	P57	J13	P84	283 [3]
I/O	P60	P58	H10	P85	286 [3]
I/O	-	P59	H11	P86	289 [3]
I/O	-	P60	H12	P87	292 [3]
I/O (D4[2])	P61	P61	H13	P88	295 [3]
I/O	P62	P62	G12	P89	298 [3]
VCC	P63	P63	G13	P90	-
GND	P64	P64	G11	P91	-
I/O (D3[2])	P65	P65	G10	P92	301 [3]
I/O	P66	P66	F13	P93	304 [3]
I/O	-	P67	F12	P94	307 [3]
I/O	-	-	F11	P95	310 [3]
I/O (D2[2])	P67	P68	F10	P96	313 [3]
I/O	P68	P69	E13	P97	316 [3]
I/O	-	-	E12	P98	319 [3]
I/O	-	-	E11	P99	322 [3]
GND	-	-	E10	P100	-
I/O (D1[2])	P69	P70	D13	P101	325 [3]
I/O	P70	P71	D12	P102	328 [3]
I/O	-	-	D11	P103	331 [3]
I/O	-	-	C13	P104	334 [3]
I/O (D0[2], DIN)	P71	P72	C12	P105	337 [3]
I/O, SGCK4[1] GCK6[2] (DOUT)	P72	P73	C11	P106	340 [3]
CCLK	P73	P74	B13	P107	-
VCC	P74	P75	B12	P108	-
O, TDO	P75	P76	A13	P109	0
GND	P76	P77	A12	P110	-
I/O	P77	P78	B11	P111	2
I/O, PGCK4[1] GCK7[2]	P78	P79	A11	P112	5
I/O	-	-	D10	P113	8
I/O	-	-	C10	P114	11
I/O (CS1[2])	P79	P80	B10	P115	14

Additional XCS20/XL Package Pins

PQ208					
Not Connected Pins					
P12	P18 (1)	P33 (1)	P39	P65	P71 (1)
P86 (1)	P92	P111	P121(1)	P140(1)	P144
P165	P173(1)	P192(1)	P202	P203	-
9/16/98					

Notes:
1. 5V Spartan family only
2. 3V Spartan-XL family only
3. The “PWRDWN” on the XCS20XL is not part of the Boundary Scan chain. For the XCS20XL, subtract 1 from all Boundary Scan numbers from GCK3 on (247 and higher).
4. CS144 package discontinued by PDN2004-01

XCS30 and XCS30XL Device Pinouts

XCS30/XL Pad Name	VQ100(5)	TQ144	PQ208	PQ240	BG256(5)	CS280(2,5)	Bndry Scan
VCC	P89	P128	P183	P212	VCC(4)	C10	-
I/O	P90	P129	P184	P213	C10	D10	74
I/O	P91	P130	P185	P214	D10	E10	77
I/O	P92	P131	P186	P215	A9	A9	80
I/O	P93	P132	P187	P216	B9	B9	83
I/O	-	-	P188	P217	C9	C9	86
I/O	-	-	P189	P218	D9	D9	89
I/O	P94	P133	P190	P220	A8	A8	92
I/O	P95	P134	P191	P221	B8	B8	95
VCC	-	-	P192	P222	VCC(4)	A7	-
I/O	-	-	-	P223	A6	B7	98
I/O	-	-	-	P224	C7	C7	101
I/O	-	P135	P193	P225	B6	D7	104
I/O	-	P136	P194	P226	A5	A6	107
GND	-	P137	P195	P227	GND(4)	GND(4)	-
I/O	-	-	P196	P228	C6	B6	110
I/O	-	-	P197	P229	B5	C6	113
I/O	-	-	P198	P230	A4	D6	116
I/O	-	-	P199	P231	C5	E6	119
I/O	P96	P138	P200	P232	B4	A5	122
I/O	P97	P139	P201	P233	A3	C5	125
I/O	-	-	P202	P234	D5	B4	128
I/O	-	-	P203	P235	C4	C4	131
I/O	-	P140	P204	P236	B3	A3	134
I/O	-	P141	P205	P237	B2	A2	137
I/O	P98	P142	P206	P238	A2	B3	140
I/O, SGCK1(1), GCK8(2)	P99	P143	P207	P239	C3	B2	143
VCC	P100	P144	P208	P240	VCC(4)	A1	-
GND	P1	P1	P1	P1	GND(4)	GND(4)	-
I/O, PGCK1(1), GCK1(2)	P2	P2	P2	P2	B1	C3	146
I/O	P3	P3	P3	P3	C2	C2	149
I/O	-	P4	P4	P4	D2	B1	152

XCS30 and XCS30XL Device Pinouts *(Continued)*

XCS30/XL Pad Name	VQ100[5]	TQ144	PQ208	PQ240	BG256[5]	CS280[2,5]	Bndry Scan
I/O	-	-	P85	P97	U12	T11	382 [3]
I/O	-	-	-	P99	V13	U12	385 [3]
I/O	-	-	-	P100	Y14	T12	388 [3]
VCC	-	-	P86	P101	VCC[4]	W13	-
I/O	P43	P60	P87	P102	Y15	V13	391 [3]
I/O	P44	P61	P88	P103	V14	U13	394 [3]
I/O	-	P62	P89	P104	W15	T13	397 [3]
I/O	-	P63	P90	P105	Y16	W14	400 [3]
GND	-	P64	P91	P106	GND[4]	GND[4]	-
I/O	-	-	-	P107	V15	V14	403 [3]
I/O	-	-	P92	P108	W16	U14	406 [3]
I/O	-	-	P93	P109	Y17	T14	409 [3]
I/O	-	-	P94	P110	V16	R14	412 [3]
I/O	-	-	P95	P111	W17	W15	415 [3]
I/O	-	-	P96	P112	Y18	U15	418 [3]
I/O	P45	P65	P97	P113	U16	V16	421 [3]
I/O	P46	P66	P98	P114	V17	U16	424 [3]
I/O	-	P67	P99	P115	W18	W17	427 [3]
I/O	-	P68	P100	P116	Y19	W18	430 [3]
I/O	P47	P69	P101	P117	V18	V17	433 [3]
I/O, SGCK3[1], GCK4[2]	P48	P70	P102	P118	W19	V18	436 [3]
GND	P49	P71	P103	P119	GND[4]	GND[4]	-
DONE	P50	P72	P104	P120	Y20	W19	-
VCC	P51	P73	P105	P121	VCC[4]	U17	-
$\overline{\text{PROGRAM}}$	P52	P74	P106	P122	V19	U18	-
I/O (D7[2])	P53	P75	P107	P123	U19	V19	439 [3]
I/O, PGCK3[1], GCK5[2]	P54	P76	P108	P124	U18	U19	442 [3]
I/O	-	P77	P109	P125	T17	T16	445 [3]
I/O	-	P78	P110	P126	V20	T17	448 [3]
I/O	-	-	-	P127	U20	T18	451 [3]
I/O	-	-	P111	P128	T18	T19	454 [3]
I/O (D6[2])	P55	P79	P112	P129	T19	R16	457 [3]
I/O	P56	P80	P113	P130	T20	R19	460 [3]
I/O	-	-	P114	P131	R18	P15	463 [3]
I/O	-	-	P115	P132	R19	P17	466 [3]
I/O	-	-	P116	P133	R20	P18	469 [3]
I/O	-	-	P117	P134	P18	P16	472 [3]
GND	-	P81	P118	P135	GND[4]	GND[4]	-
I/O	-	-	-	P136	P20	P19	475 [3]
I/O	-	-	-	P137	N18	N17	478 [3]
I/O	-	P82	P119	P138	N19	N18	481 [3]
I/O	-	P83	P120	P139	N20	N19	484 [3]
VCC	-	-	P121	P140	VCC[4]	N16	-
I/O (D5[2])	P57	P84	P122	P141	M17	M19	487 [3]
I/O	P58	P85	P123	P142	M18	M17	490 [3]

XCS30 and XCS30XL Device Pinouts *(Continued)*

XCS30/XL Pad Name	VQ100[5]	TQ144	PQ208	PQ240	BG256[5]	CS280[2,5]	Bndry Scan
I/O	-	-	P124	P144	M20	L19	493 [3]
I/O	-	-	P125	P145	L19	L18	496 [3]
I/O	P59	P86	P126	P146	L18	L17	499 [3]
I/O	P60	P87	P127	P147	L20	L16	502 [3]
I/O (D4[2])	P61	P88	P128	P148	K20	K19	505 [3]
I/O	P62	P89	P129	P149	K19	K18	508 [3]
VCC	P63	P90	P130	P150	VCC[4]	K17	-
GND	P64	P91	P131	P151	GND[4]	GND[4]	-
I/O (D3[2])	P65	P92	P132	P152	K18	K16	511 [3]
I/O	P66	P93	P133	P153	K17	K15	514 [3]
I/O	P67	P94	P134	P154	J20	J19	517 [3]
I/O	-	P95	P135	P155	J19	J18	520 [3]
I/O	-	-	P136	P156	J18	J17	523 [3]
I/O	-	-	P137	P157	J17	J16	526 [3]
I/O (D2[2])	P68	P96	P138	P159	H19	H17	529 [3]
I/O	P69	P97	P139	P160	H18	H16	532 [3]
VCC	-	-	P140	P161	VCC[4]	G19	-
I/O	-	P98	P141	P162	G19	G18	535 [3]
I/O	-	P99	P142	P163	F20	G17	538 [3]
I/O	-	-	-	P164	G18	G16	541 [3]
I/O	-	-	-	P165	F19	F19	544 [3]
GND	-	P100	P143	P166	GND[4]	GND[4]	-
I/O	-	-	-	P167	F18	F18	547 [3]
I/O	-	-	P144	P168	E19	F17	550 [3]
I/O	-	-	P145	P169	D20	F16	553 [3]
I/O	-	-	P146	P170	E18	F15	556 [3]
I/O	-	-	P147	P171	D19	E19	559 [3]
I/O	-	-	P148	P172	C20	E17	562 [3]
I/O (D1[2])	P70	P101	P149	P173	E17	E16	565 [3]
I/O	P71	P102	P150	P174	D18	D19	568 [3]
I/O	-	P103	P151	P175	C19	C19	571 [3]
I/O	-	P104	P152	P176	B20	B19	574 [3]
I/O (D0[2], DIN)	P72	P105	P153	P177	C18	C18	577 [3]
I/O, SGCK4[1], GCK6[2] (DOUT)	P73	P106	P154	P178	B19	B18	580 [3]
CCLK	P74	P107	P155	P179	A20	A19	-
VCC	P75	P108	P156	P180	VCC[4]	C17	-
O, TDO	P76	P109	P157	P181	A19	B17	0
GND	P77	P110	P158	P182	GND[4]	GND[4]	-
I/O	P78	P111	P159	P183	B18	A18	2
I/O, PGCK4[1], GCK7[2]	P79	P112	P160	P184	B17	A17	5
I/O	-	P113	P161	P185	C17	D16	8
I/O	-	P114	P162	P186	D16	C16	11
I/O (CS1)[2]	P80	P115	P163	P187	A18	B16	14
I/O	P81	P116	P164	P188	A17	A16	17
I/O	-	-	P165	P189	C16	D15	20