AMD Xilinx - XCS40XL-5PQ240C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	784
Number of Logic Elements/Cells	1862
Total RAM Bits	25088
Number of I/O	192
Number of Gates	40000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	240-BFQFP
Supplier Device Package	240-PQFP (32x32)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcs40xl-5pq240c

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Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

The register choice is made by placing the appropriate library symbol. For example, IFD is the basic input flip-flop (rising edge triggered), and ILD is the basic input latch (transparent-High). Variations with inverted clocks are also available. The clock signal inverter is also shown in Figure 5 on the CK line.

The Spartan family IOB data input path has a one-tap delay element: either the delay is inserted (default), or it is not. The Spartan-XL family IOB data input path has a two-tap delay element, with choices of a full delay, a partial delay, or no delay. The added delay guarantees a zero hold time with respect to clocks routed through the global clock buffers. (See **Global Nets and Buffers**, page 12 for a description of the global clock buffers in the Spartan/XL families.) For a shorter input register setup time, with positive hold-time, attach a NODELAY attribute or property to the flip-flop.The output of the input register goes to the routing channels (via I1 and I2 in Figure 6). The I1 and I2 signals that exit the IOB can each carry either the direct or registered input signal.

The 5V Spartan family input buffers can be globally configured for either TTL (1.2V) or CMOS (VCC/2) thresholds, using an option in the bitstream generation software. The Spartan family output levels are also configurable; the two global adjustments of input threshold and output level are independent. The inputs of Spartan devices can be driven by the outputs of any 3.3V device, if the Spartan family inputs are in TTL mode. Input and output thresholds are TTL on all configuration pins until the configuration has been loaded into the device and specifies how they are to be used. Spartan-XL family inputs are TTL compatible and 3.3V CMOS compatible.

Supported sources for Spartan/XL device inputs are shown in Table 4.

Spartan-XL family I/Os are fully 5V tolerant even though the V_{CC} is 3.3V. This allows 5V signals to directly connect to the Spartan-XL family inputs without damage, as shown in Table 4. In addition, the 3.3V V_{CC} can be applied before or after 5V signals are applied to the I/Os. This makes the Spartan-XL devices immune to power supply sequencing problems.

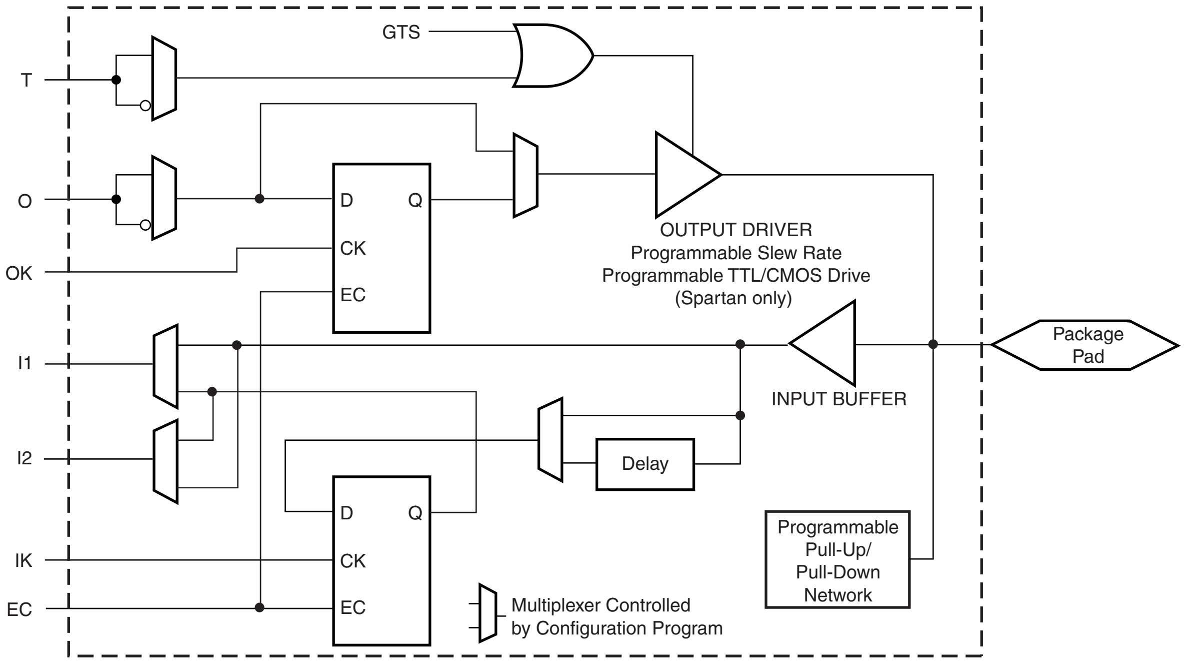

Figure 6: **Simplified Spartan/XL IOB Block Diagram**

Table 4: **Supported Sources for Spartan/XL Inputs**

	Spartan Inputs		Spartan-XL Inputs
Source	**5V, TTL**	**5V, CMOS**	**3.3V CMOS**
Any device, V_{CC} = 3.3V, CMOS outputs	√	Unreliable Data	√
Spartan family, V_{CC} = 5V, TTL outputs	√		√
Any device, V_{CC} = 5V, TTL outputs ($V_{OH} \leq 3.7V$)	√		√
Any device, V_{CC} = 5V, CMOS outputs	√	√	√ (default mode)

Spartan-XL Family V_{CC} Clamping

Spartan-XL FPGAs have an optional clamping diode connected from each I/O to V_{CC}. When enabled they clamp ringing transients back to the 3.3V supply rail. This clamping action is required in 3.3V PCI applications. V_{CC} clamping is a global option affecting all I/O pins.

Spartan-XL devices are fully 5V TTL I/O compatible if V_{CC} clamping is not enabled. With V_{CC} clamping enabled, the Spartan-XL devices will begin to clamp input voltages to one diode voltage drop above V_{CC}. If enabled, TTL I/O compatibility is maintained but full 5V I/O tolerance is sacrificed. The user may select either 5V tolerance (default) or 3.3V PCI compatibility. In both cases negative voltage is clamped to one diode voltage drop below ground.

Spartan-XL devices are compatible with TTL, LVTTL, PCI 3V, PCI 5V and LVCMOS signalling. The various standards are illustrated in Table 5.

Table 5: **I/O Standards Supported by Spartan-XL FPGAs**

Signaling Standard	**VCC Clamping**	**Output Drive**	$V_{IH\ MAX}$	$V_{IH\ MIN}$	$V_{IL\ MAX}$	$V_{OH\ MIN}$	$V_{OL\ MAX}$
TTL	Not allowed	12/24 mA	5.5	2.0	0.8	2.4	0.4
LVTTL	OK	12/24 mA	3.6	2.0	0.8	2.4	0.4
PCI5V	Not allowed	24 mA	5.5	2.0	0.8	2.4	0.4
PCI3V	Required	12 mA	3.6	50% of V_{CC}	30% of V_{CC}	90% of V_{CC}	10% of V_{CC}
LVCMOS 3V	OK	12/24 mA	3.6	50% of V_{CC}	30% of V_{CC}	90% of V_{CC}	10% of V_{CC}

Additional Fast Capture Input Latch (Spartan-XL Family Only)

The Spartan-XL family OB has an additional optional latch on the input. This latch is clocked by the clock used for the output flip-flop rather than the input clock. Therefore, two different clocks can be used to clock the two input storage elements. This additional latch allows the fast capture of input data, which is then synchronized to the internal clock by the IOB flip-flop or latch.

To place the Fast Capture latch in a design, use one of the special library symbols, ILFFX or ILFLX. ILFFX is a transparent-Low Fast Capture latch followed by an active High input flip-flop. ILFLX is a transparent Low Fast Capture latch followed by a transparent High input latch. Any of the clock inputs can be inverted before driving the library element, and the inverter is absorbed into the IOB.

IOB Output Signal Path

Output signals can be optionally inverted within the IOB, and can pass directly to the output buffer or be stored in an edge-triggered flip-flop and then to the output buffer. The functionality of this flip-flop is shown in Table 6.

Table 6: **Output Flip-Flop Functionality**

Mode	**Clock**	**Clock Enable**	**T**	**D**	**Q**
Power-Up or GSR	X	X	0*	X	SR
Flip-Flop	X	0	0*	X	Q
	↑	1*	0*	D	D
	X	X	1	X	Z
	0	X	0*	X	Q

Legend:

X	Don't care
↑	Rising edge (clock not inverted).
SR	Set or Reset value. Reset is default.
0*	Input is Low or unconnected (default value)
1*	Input is High or unconnected (default value)
Z	3-state

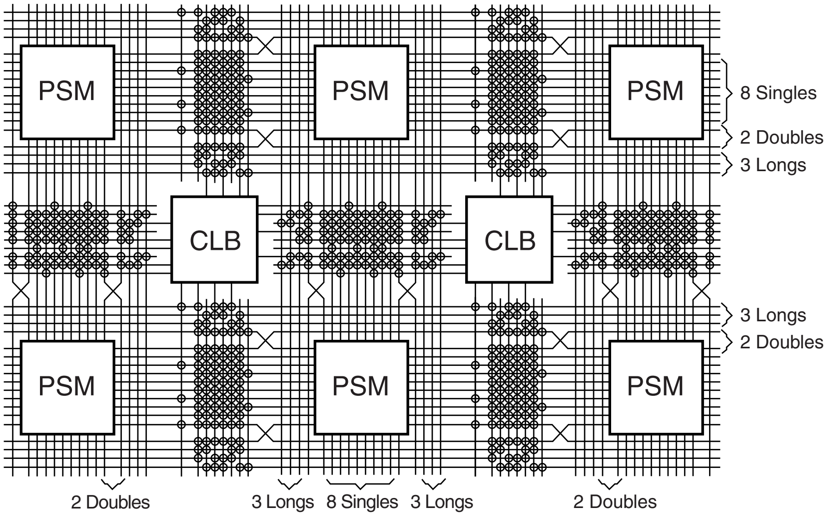

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Figure 8: **Spartan/XL CLB Routing Channels and Interface Block Diagram**

CLB Interface

A block diagram of the CLB interface signals is shown in Figure 9. The input signals to the CLB are distributed evenly on all four sides providing maximum routing flexibility. In general, the entire architecture is symmetrical and regular. It is well suited to established placement and routing algorithms. Inputs, outputs, and function generators can freely swap positions within a CLB to avoid routing congestion during the placement and routing operation. The exceptions are the clock (K) input and CIN/COUT signals. The K input is routed to dedicated global vertical lines as well as four single-length lines and is on the left side of the CLB. The CIN/COUT signals are routed through dedicated interconnects which do not interfere with the general routing structure. The output signals from the CLB are available to drive both vertical and horizontal channels.

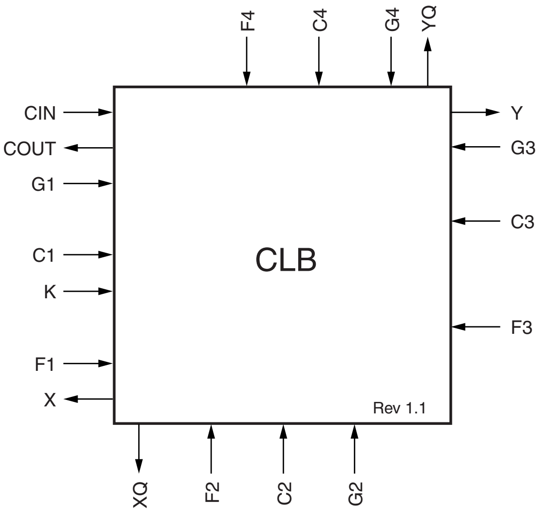

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Figure 9: **CLB Interconnect Signals**

Programmable Switch Matrices

The horizontal and vertical single- and double-length lines intersect at a box called a programmable switch matrix (PSM). Each PSM consists of programmable pass transistors used to establish connections between the lines (see Figure 10).

For example, a single-length signal entering on the right side of the switch matrix can be routed to a single-length line on the top, left, or bottom sides, or any combination thereof, if multiple branches are required. Similarly, a double-length signal can be routed to a double-length line on any or all of the other three edges of the programmable switch matrix.

Single-Length Lines

Single-length lines provide the greatest interconnect flexibility and offer fast routing between adjacent blocks. There are eight vertical and eight horizontal single-length lines associated with each CLB. These lines connect the switching matrices that are located in every row and column of CLBs. Single-length lines are connected by way of the programmable switch matrices, as shown in Figure 10. Routing connectivity is shown in Figure 8.

Single-length lines incur a delay whenever they go through a PSM. Therefore, they are not suitable for routing signals for long distances. They are normally used to conduct signals within a localized area and to provide the branching for nets with fanout greater than one.

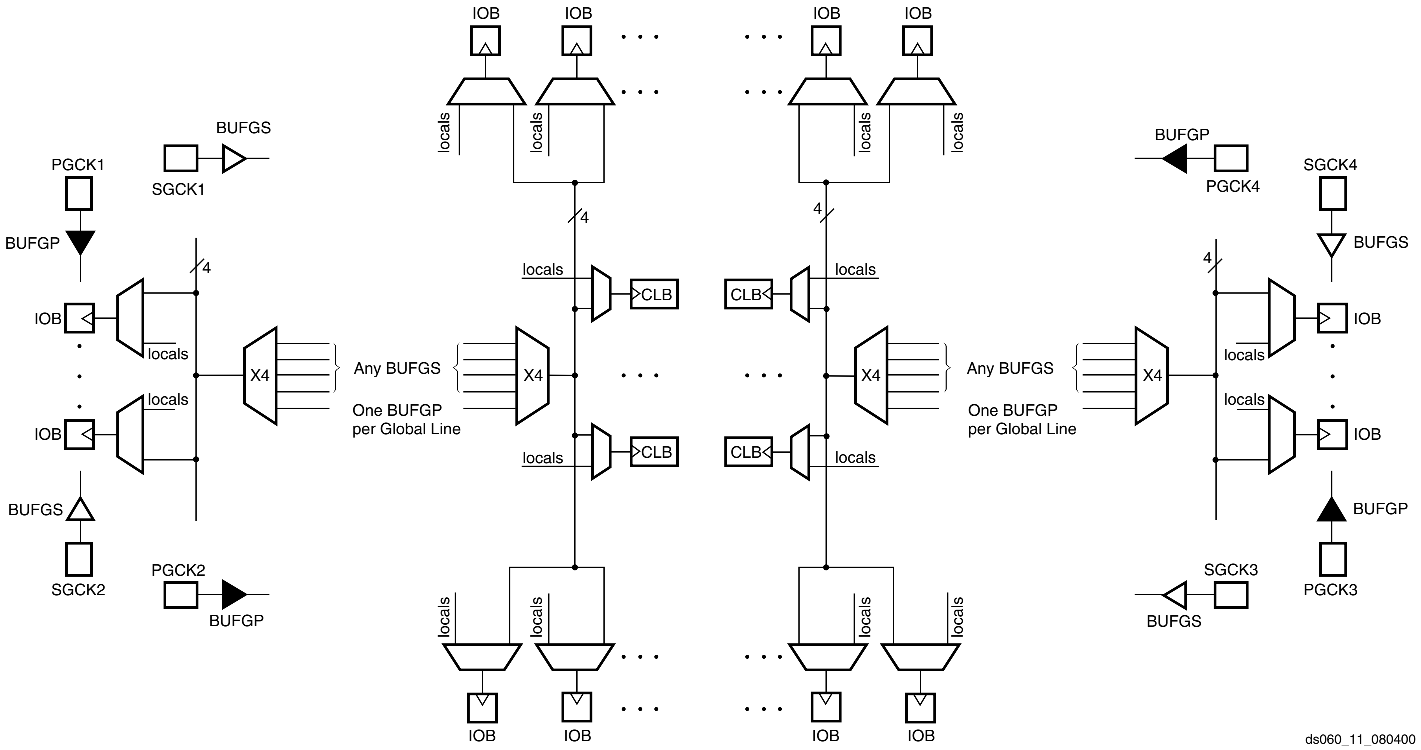


Figure 11: **5V Spartan Family Global Net Distribution**

The four Primary Global buffers offer the shortest delay and negligible skew. Four Secondary Global buffers have slightly longer delay and slightly more skew due to potentially heavier loading, but offer greater flexibility when used to drive non-clock CLB inputs. The eight Global Low-Skew buffers in the Spartan-XL devices combine short delay, negligible skew, and flexibility.

The Primary Global buffers must be driven by the semi-dedicated pads (PGCK1-4). The Secondary Global buffers can be sourced by either semi-dedicated pads (SGCK1-4) or internal nets. Each corner of the device has one Primary buffer and one Secondary buffer. The Spartan-XL family has eight global low-skew buffers, two in each corner. All can be sourced by either semi-dedicated pads (GCK1-8) or internal nets.

Using the library symbol called BUFG results in the software choosing the appropriate clock buffer, based on the timing requirements of the design. A global buffer should be specified for all timing-sensitive global signal distribution. To use a global buffer, place a BUFGP (primary buffer), BUFGS (secondary buffer), BUFGLS (Spartan-XL family global low-skew buffer), or BUFG (any buffer type) element in a schematic or in HDL code.

Advanced Features Description

Distributed RAM

Optional modes for each CLB allow the function generators (F-LUT and G-LUT) to be used as Random Access Memory (RAM).

Read and write operations are significantly faster for this on-chip RAM than for off-chip implementations. This speed advantage is due to the relatively short signal propagation delays within the FPGA.

Memory Configuration Overview

There are two available memory configuration modes: single-port RAM and dual-port RAM. For both these modes, write operations are synchronous (edge-triggered), while read operations are asynchronous. In the single-port mode, a single CLB can be configured as either a 16 x 1, (16 x 1) x 2, or 32 x 1 RAM array. In the dual-port mode, a single CLB can be configured only as one 16 x 1 RAM array. The different CLB memory configurations are summarized in Table 8. Any of these possibilities can be individually programmed into a Spartan/XL FPGA CLB.

Table 8: **CLB Memory Configurations**

Mode	16 x 1	(16 x 1) x 2	32 x 1
Single-Port	√	√	√
Dual-Port	√	–	–

- The 16 x 1 single-port configuration contains a RAM array with 16 locations, each one-bit wide. One 4-bit address decoder determines the RAM location for write and read operations. There is one input for writing data and one output for reading data, all at the selected address.
- The (16 x 1) x 2 single-port configuration combines two 16 x 1 single-port configurations (each according to the preceding description). There is one data input, one data output and one address decoder for each array. These arrays can be addressed independently.
- The 32 x 1 single-port configuration contains a RAM array with 32 locations, each one-bit wide. There is one data input, one data output, and one 5-bit address decoder.
- The dual-port mode 16 x 1 configuration contains a RAM array with 16 locations, each one-bit wide. There are two 4-bit address decoders, one for each port. One port consists of an input for writing and an output for reading, all at a selected address. The other port consists of one output for reading from an independently selected address.

The appropriate choice of RAM configuration mode for a given design should be based on timing and resource requirements, desired functionality, and the simplicity of the design process. Selection criteria include the following: Whereas the 32 x 1 single-port, the (16 x 1) x 2 single-port, and the 16 x 1 dual-port configurations each use one entire CLB, the 16 x 1 single-port configuration uses only one half of a CLB. Due to its simultaneous read/write capability, the dual-port RAM can transfer twice as much data as the single-port RAM, which permits only one data operation at any given time.

CLB memory configuration options are selected by using the appropriate library symbol in the design entry.

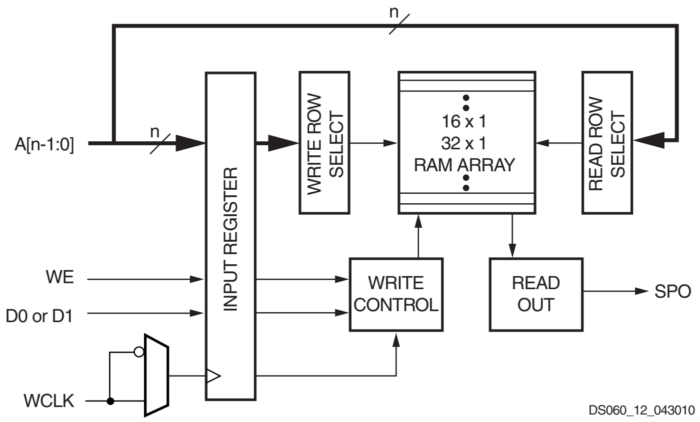
Single-Port Mode

There are three CLB memory configurations for the single-port RAM: 16 x 1, (16 x 1) x 2, and 32 x 1, the functional organization of which is shown in Figure 12.

The single-port RAM signals and the CLB signals (Figure 2, page 4) from which they are originally derived are shown in Table 9.

Table 9: **Single-Port RAM Signals**

RAM Signal	Function	CLB Signal
D0 or D1	Data In	DIN or H1
A[3:0]	Address	F[4:1] or G[4:1]
A4 (32 x 1 only)	Address	H1
WE	Write Enable	SR
WCLK	Clock	K
SPO	Single Port Out (Data Out)	F_{OUT} or G_{OUT}



Notes:

1. The (16 x 1) x 2 configuration combines two 16 x 1 single-port RAMs, each with its own independent address bus and data input. The same WE and WCLK signals are connected to both RAMs.
2. n = 4 for the 16 x 1 and (16 x 1) x 2 configurations. n = 5 for the 32 x 1 configuration.

Figure 12: **Logic Diagram for the Single-Port RAM**

Writing data to the single-port RAM is essentially the same as writing to a data register. It is an edge-triggered (synchronous) operation performed by applying an address to the A inputs and data to the D input during the active edge of WCLK while WE is High.

The timing relationships are shown in Figure 13. The High logic level on WE enables the input data register for writing. The active edge of WCLK latches the address, input data, and WE signals. Then, an internal write pulse is generated that loads the data into the memory cell.

and Spartan-XL families, speeding up arithmetic and counting functions.

The carry chain in 5V Spartan devices can run either up or down. At the top and bottom of the columns where there are no CLBs above and below, the carry is propagated to the right. The default is always to propagate up the column, as shown in the figures. The carry chain in Spartan-XL devices can only run up the column, providing even higher speed.

Figure 16, page 18 shows a Spartan/XL FPGA CLB with dedicated fast carry logic. The carry logic shares operand and control inputs with the function generators. The carry outputs connect to the function generators, where they are combined with the operands to form the sums.

Figure 17, page 19 shows the details of the Spartan/XL FPGA carry logic. This diagram shows the contents of the box labeled "CARRY LOGIC" in Figure 16.

The fast carry logic can be accessed by placing special library symbols, or by using Xilinx Relationally Placed Macros (RPMs) that already include these symbols.

Figure 20 is a diagram of the Spartan/XL FPGA boundary scan logic. It includes three bits of Data Register per IOB, the IEEE 1149.1 Test Access Port controller, and the Instruction Register with decodes.

Spartan/XL devices can also be configured through the boundary scan logic. See **Configuration Through the Boundary Scan Pins**, page 37.

Data Registers

The primary data register is the boundary scan register. For each IOB pin in the FPGA, bonded or not, it includes three bits for In, Out and 3-state Control. Non-IOB pins have appropriate partial bit population for In or Out only. $\overline{\text{PROGRAM}}$, CCLK and DONE are not included in the boundary scan register. Each EXTEST CAPTURE-DR state captures all In, Out, and 3-state pins.

The data register also includes the following non-pin bits: TDO.T, and TDO.O, which are always bits 0 and 1 of the data register, respectively, and BSCANT.UPD, which is always the last bit of the data register. These three boundary scan bits are special-purpose Xilinx test signals.

The other standard data register is the single flip-flop BYPASS register. It synchronizes data being passed through the FPGA to the next downstream boundary scan device.

The FPGA provides two additional data registers that can be specified using the BSCAN macro. The FPGA provides two user pins (BSCAN.SEL1 and BSCAN.SEL2) which are the decodes of two user instructions. For these instructions, two corresponding pins (BSCAN.TDO1 and BSCAN.TDO2) allow user scan data to be shifted out on TDO. The data register clock (BSCAN.DRCK) is available for control of test logic which the user may wish to implement with CLBs. The NAND of TCK and RUN-TEST-IDLE is also provided (BSCAN.IDLE).

Instruction Set

The Spartan/XL FPGA boundary scan instruction set also includes instructions to configure the device and read back the configuration data. The instruction set is coded as shown in Table 12.

Even if the boundary scan symbol is used in a design, the input pins TMS, TCK, and TDI can still be used as inputs to be routed to internal logic. Care must be taken not to force the chip into an undesired boundary scan state by inadvertently applying boundary scan input patterns to these pins. The simplest way to prevent this is to keep TMS High, and then apply whatever signal is desired to TDI and TCK.

Avoiding Inadvertent Boundary Scan

If TMS or TCK is used as user I/O, care must be taken to ensure that at least one of these pins is held constant during configuration. In some applications, a situation may occur where TMS or TCK is driven during configuration. This may cause the device to go into boundary scan mode and disrupt the configuration process.

To prevent activation of boundary scan during configuration, do either of the following:

- TMS: Tie High to put the Test Access Port controller in a benign RESET state.
- TCK: Tie High or Low—do not toggle this clock input.

For more information regarding boundary scan, refer to the Xilinx Application Note, "*Boundary Scan in FPGA Devices*. "

Boundary Scan Enhancements (Spartan-XL Family Only)

Spartan-XL devices have improved boundary scan functionality and performance in the following areas:

IDCODE: The IDCODE register is supported. By using the IDCODE, the device connected to the JTAG port can be determined. The use of the IDCODE enables selective configuration dependent on the FPGA found.

The IDCODE register has the following binary format:

```
vvvv:ffff:fffa:aaaa:aaaa:cccc:cccc:ccc1
```

where

c = the company code (49h for Xilinx)

a = the array dimension in CLBs (ranges from 0Ah for XCS05XL to 1Ch for XCS40XL)

f = the family code (02h for Spartan-XL family)

v = the die version number

Table 13: **IDCODEs Assigned to Spartan-XL FPGAs**

FPGA	IDCODE
XCS05XL	0040A093h
XCS10XL	0040E093h
XCS20XL	00414093h
XCS30XL	00418093h
XCS40XL	0041C093h

Configuration State: The configuration state is available to JTAG controllers.

Configuration Disable: The JTAG port can be prevented from configuring the FPGA.

TCK Startup: TCK can now be used to clock the start-up block in addition to other user clocks.

CCLK Holdoff: Changed the requirement for Boundary Scan Configure or EXTEST to be issued prior to the release of INIT pin and CCLK cycling.

Reissue Configure: The Boundary Scan Configure can be reissued to recover from an unfinished attempt to configure the device.

Bypass FF: Bypass FF and IOB is modified to provide DRCLOCK only during BYPASS for the bypass flip-flop, and during EXTEST or SAMPLE/PRELOAD for the IOB register.

Power-Down (Spartan-XL Family Only)

All Spartan/XL devices use a combination of efficient segmented routing and advanced process technology to provide low power consumption under all conditions. The 3.3V Spartan-XL family adds a dedicated active Low power-down pin ($\overline{\text{PWRDWN}}$) to reduce supply current to 100 µA typical. The $\overline{\text{PWRDWN}}$ pin takes advantage of one of the unused No Connect locations on the 5V Spartan device. The user must de-select the "5V Tolerant I/Os" option in the Configuration Options to achieve the specified Power Down current. The $\overline{\text{PWRDWN}}$ pin has a default internal pull-up resistor, allowing it to be left unconnected if unused.

V_{CC} must continue to be supplied during Power-down, and configuration data is maintained. When the $\overline{\text{PWRDWN}}$ pin is pulled Low, the input and output buffers are disabled. The inputs are internally forced to a logic Low level, including the MODE pins, DONE, CCLK, and TDO, and all internal pull-up resistors are turned off. The $\overline{\text{PROGRAM}}$ pin is not affected by Power Down. The GSR net is asserted during Power Down, initializing all the flip-flops to their start-up state.

$\overline{\text{PWRDWN}}$ has a minimum pulse width of 50 ns (Figure 23). On entering the Power-down state, the inputs will be disabled and the flip-flops set/reset, and then the outputs are disabled about 10 ns later. The user may prefer to assert the GTS or GSR signals before $\overline{\text{PWRDWN}}$ to affect the order of events. When the $\overline{\text{PWRDWN}}$ signal is returned High, the inputs will be enabled first, followed immediately by the release of the GSR signal initializing the flip-flops. About 10 ns later, the outputs will be enabled. Allow 50 ns after the release of $\overline{\text{PWRDWN}}$ before using the device.

to the DONE pin. User I/Os for each device become active after the DONE pin for that device goes High. (The exact timing is determined by development system options.) Since the DONE pin is open-drain and does not drive a High value, tying the DONE pins of all devices together prevents all devices in the chain from going High until the last device in the chain has completed its configuration cycle. If the DONE pin of a device is left unconnected, the device becomes active as soon as that device has been configured. Only devices supporting Express mode can be used to form an Express mode daisy chain.

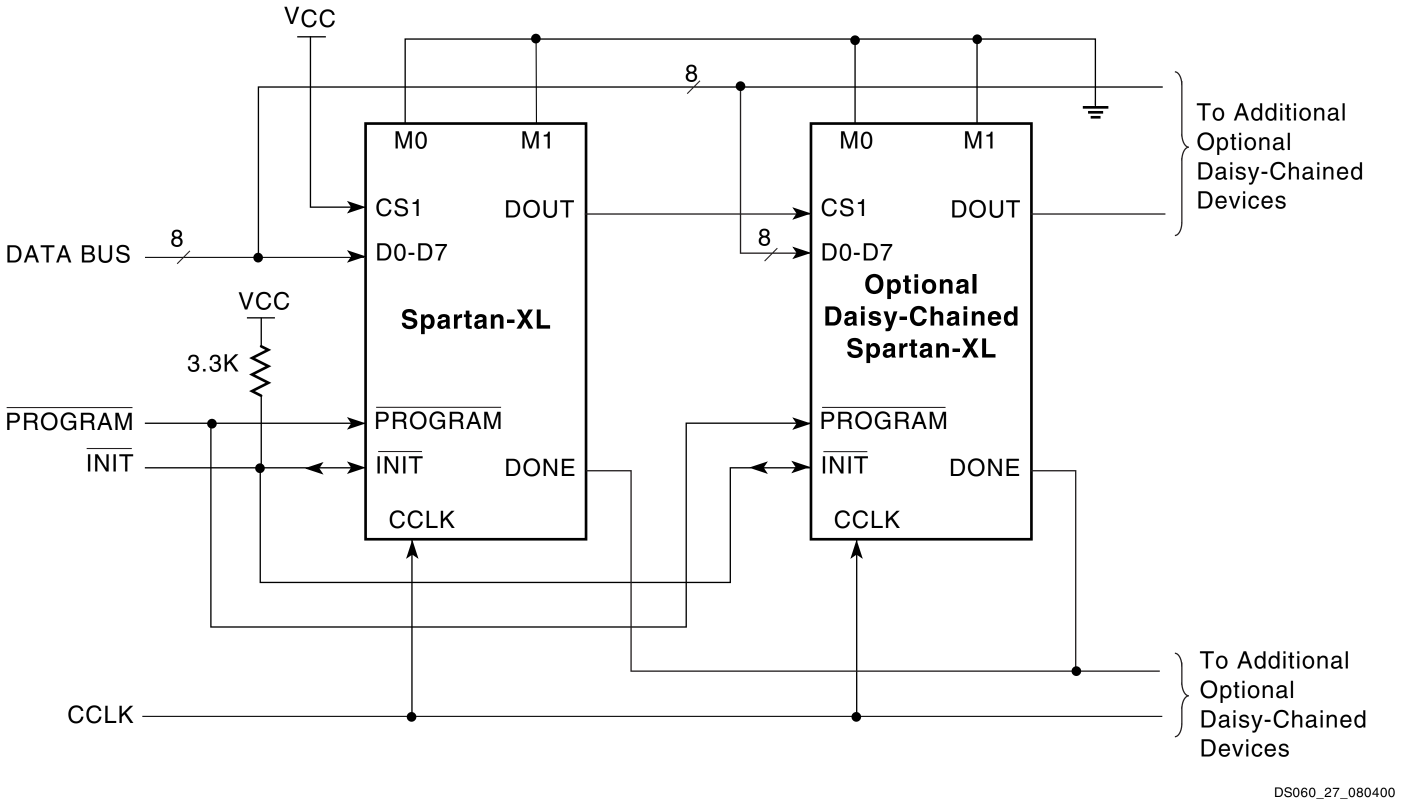


Figure 27: **Express Mode Circuit Diagram**

Table 16: **Spartan/XL Data Stream Formats**

Data Type	Serial Modes (D0...)	Express Mode (D0-D7) (Spartan-XL only)
Fill Byte	11111111b	FFFFh
Preamble Code	0010b	11110010b
Length Count	COUNT[23:0]	COUNT[23:0](1)
Fill Bits	1111b	-
Field Check Code	-	11010010b
Start Field	0b	11111110b(2)
Data Frame	DATA[n–1:0]	DATA[n–1:0]
CRC or Constant Field Check	xxxx (CRC) or 0110b	11010010b
Extend Write Cycle	-	FFD2FFFFFFh
Postamble	01111111b	-
Start-Up Bytes(3)	FFh	FFFFFFFFFFFFFFh

Legend:

Unshaded	Once per bitstream
Light	Once per data frame
Dark	Once per device

Notes:

1. Not used by configuration logic.
2. 11111111b for XCS40XL only.
3. Development system may add more start-up bytes.

A selection of CRC or non-CRC error checking is allowed by the bitstream generation software. The Spartan-XL family Express mode only supports non-CRC error checking. The non-CRC error checking tests for a designated end-of-frame field for each frame. For CRC error checking, the software calculates a running CRC and inserts a unique four-bit partial check at the end of each frame. The 11-bit CRC check of the last frame of an FPGA includes the last seven data bits.

Detection of an error results in the suspension of data loading before DONE goes High, and the pulling down of the $\overline{\text{INIT}}$ pin. In Master serial mode, CCLK continues to operate externally. The user must detect $\overline{\text{INIT}}$ and initialize a new configuration by pulsing the $\overline{\text{PROGRAM}}$ pin Low or cycling V_{CC}.

Cyclic Redundancy Check (CRC) for Configuration and Readback

The Cyclic Redundancy Check is a method of error detection in data transmission applications. Generally, the transmitting system performs a calculation on the serial bitstream. The result of this calculation is tagged onto the data stream as additional check bits. The receiving system performs an identical calculation on the bitstream and compares the result with the received checksum.

Each data frame of the configuration bitstream has four error bits at the end, as shown in Table 16. If a frame data error is detected during the loading of the FPGA, the configuration process with a potentially corrupted bitstream is terminated. The FPGA pulls the $\overline{\text{INIT}}$ pin Low and goes into a Wait state.

Table 17: **Spartan/XL Program Data**

Device	XCS05		XCS10		XCS20		XCS30		XCS40	
Max System Gates	5,000		10,000		20,000		30,000		40,000	
CLBs (Row x Col.)	100 (10 x 10)		196 (14 x 14)		400 (20 x 20)		576 (24 x 24)		784 (28 x 28)	
IOBs	80		112		160		192		205(4)	
Part Number	XCS05	XCS05XL	XCS10	XCS10XL	XCS20	XCS20XL	XCS30	XCS30XL	XCS40	XCS40XL
Supply Voltage	5V	3.3V	5V	3.3V	5V	3.3V	5V	3.3V	5V	3.3V
Bits per Frame	126	127	166	167	226	227	266	267	306	307
Frames	428	429	572	573	788	789	932	933	1,076	1,077
Program Data	53,936	54,491	94,960	95,699	178,096	179,111	247,920	249,119	329,264	330,647
PROM Size (bits)	53,984	54,544	95,008	95,752	178,144	179,160	247,968	249,168	329,312	330,696
Express Mode PROM Size (bits)	-	79,072	-	128,488	-	221,056	-	298,696	-	387,856

Notes:

1. Bits per Frame = (10 x number of rows) + 7 for the top + 13 for the bottom + 1 + 1 start bit + 4 error check bits (+1 for Spartan-XL device)
   Number of Frames = (36 x number of columns) + 26 for the left edge + 41 for the right edge + 1 (+ 1 for Spartan-XL device)
   Program Data = (Bits per Frame x Number of Frames) + 8 postamble bits
   PROM Size = Program Data + 40 (header) + 8, rounded up to the nearest byte
2. The user can add more "1" bits as leading dummy bits in the header, or, if CRC = off, as trailing dummy bits at the end of any frame, following the four error check bits. However, the Length Count value **must** be adjusted for all such extra "one" bits, even for extra leading ones at the beginning of the header.
3. Express mode adds 57 (XCS05XL, XCS10XL), or 53 (XCS20XL, XCS30XL, XCS40XL) bits per frame, + additional start-up bits.
4. XCS40XL provided 224 max I/O in CS280 package discontinued by PDN2004-01.

During Readback, 11 bits of the 16-bit checksum are added to the end of the Readback data stream. The checksum is computed using the CRC-16 CCITT polynomial, as shown in Figure 29. The checksum consists of the 11 most significant bits of the 16-bit code. A change in the checksum indicates a change in the Readback bitstream. A comparison to a previous checksum is meaningful only if the readback data is independent of the current device state. CLB outputs should not be included (Readback Capture option not used), and if RAM is present, the RAM content must be unchanged.

Statistically, one error out of 2048 might go undetected.

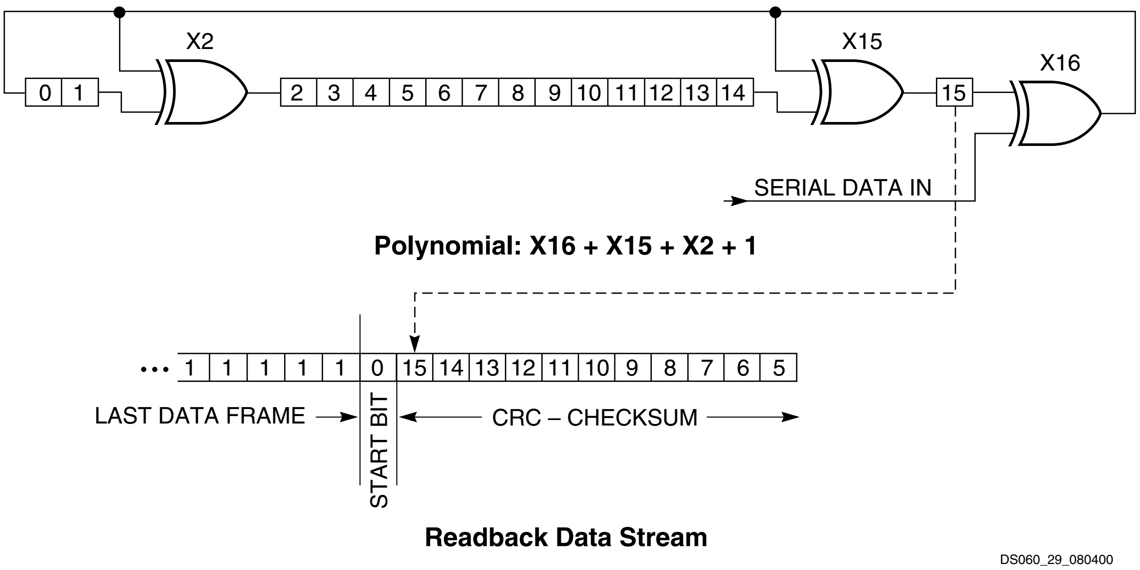


Figure 29: **Circuit for Generating CRC-16**

Configuration Sequence

There are four major steps in the Spartan/XL FPGA power-up configuration sequence.

- Configuration Memory Clear
- Initialization
- Configuration
- Start-up

The full process is illustrated in Figure 30.

Configuration Memory Clear

When power is first applied or is reapplied to an FPGA, an internal circuit forces initialization of the configuration logic. When V_{CC} reaches an operational level, and the circuit passes the write and read test of a sample pair of configuration bits, a time delay is started. This time delay is nominally 16 ms. The delay is four times as long when in Master Serial Mode to allow ample time for all slaves to reach a stable V_{CC}. When all $\overline{\text{INIT}}$ pins are tied together, as recommended, the longest delay takes precedence. Therefore, devices with different time delays can easily be mixed and matched in a daisy chain.

This delay is applied only on power-up. It is not applied when reconfiguring an FPGA by pulsing the $\overline{\text{PROGRAM}}$ pin Low. During this time delay, or as long as the $\overline{\text{PROGRAM}}$ input is asserted, the configuration logic is held in a Configuration Memory Clear state. The configuration-memory frames are consecutively initialized, using the internal oscillator.

At the end of each complete pass through the frame addressing, the power-on time-out delay circuitry and the level of the $\overline{\text{PROGRAM}}$ pin are tested. If neither is asserted, the logic initiates one additional clearing of the configuration frames and then tests the $\overline{\text{INIT}}$ input.

Initialization

During initialization and configuration, user pins HDC, $\overline{\text{LDC}}$, $\overline{\text{INIT}}$ and DONE provide status outputs for the system interface. The outputs $\overline{\text{LDC}}$, $\overline{\text{INIT}}$ and DONE are held Low and HDC is held High starting at the initial application of power.

The open drain $\overline{\text{INIT}}$ pin is released after the final initialization pass through the frame addresses. There is a deliberate delay before a Master-mode device recognizes an inactive $\overline{\text{INIT}}$. Two internal clocks after the $\overline{\text{INIT}}$ pin is recognized as High, the device samples the MODE pin to determine the configuration mode. The appropriate interface lines become active and the configuration preamble and data can be loaded.

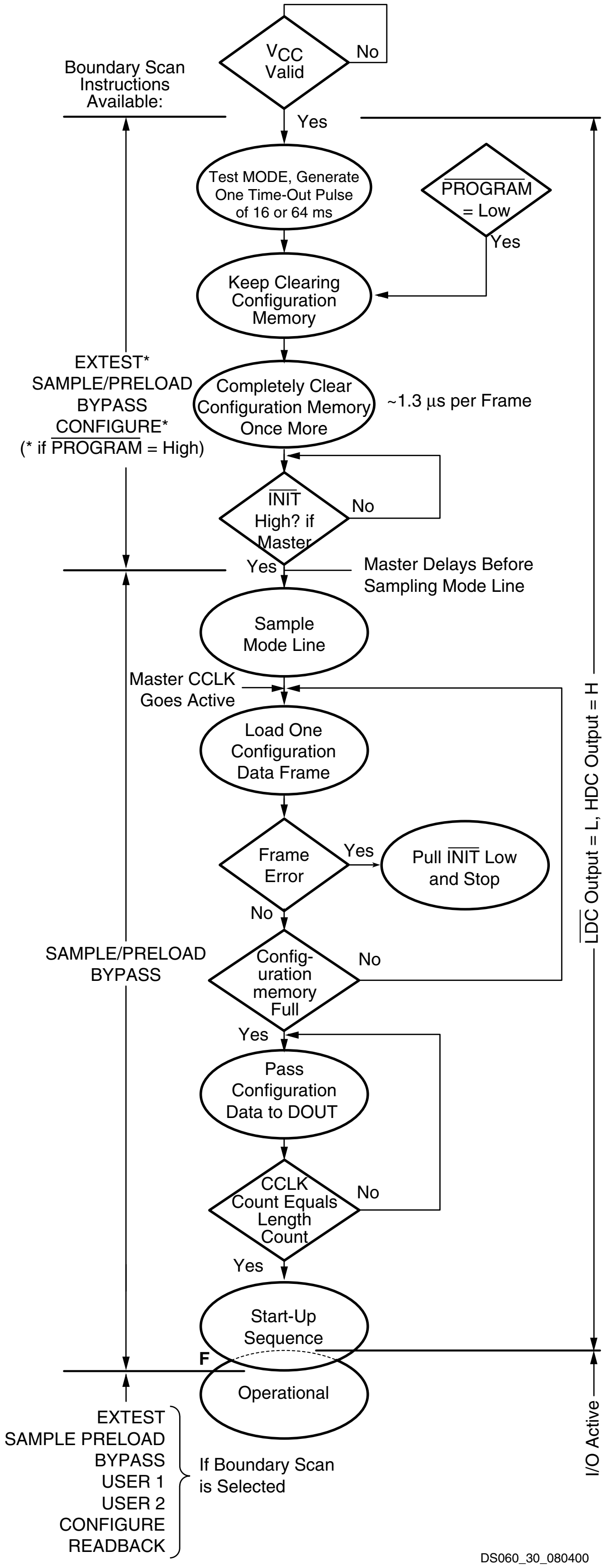


Figure 30: **Power-up Configuration Sequence**

Configuration

The 0010 preamble code indicates that the following 24 bits represent the length count for serial modes. The length count is the total number of configuration clocks needed to load the complete configuration data. (Four additional configuration clocks are required to complete the configuration process, as discussed below.) After the preamble and the length count have been passed through to any device in the daisy chain, its DOUT is held High to prevent frame start bits from reaching any daisy-chained devices. In Spartan-XL family Express mode, the length count bits are ignored, and DOUT is held Low, to disable the next device in the pseudo daisy chain.

A specific configuration bit, early in the first frame of a master device, controls the configuration-clock rate and can increase it by a factor of eight. Therefore, if a fast configuration clock is selected by the bitstream, the slower clock rate is used until this configuration bit is detected.

Each frame has a start field followed by the frame-configuration data bits and a frame error field. If a frame data error is detected, the FPGA halts loading, and signals the error by pulling the open-drain INIT pin Low. After all configuration frames have been loaded into an FPGA using a serial mode, DOUT again follows the input data so that the remaining data is passed on to the next device. In Spartan-XL family Express mode, when the first device is fully programmed, DOUT goes High to enable the next device in the chain.

Delaying Configuration After Power-Up

There are two methods of delaying configuration after power-up: put a logic Low on the PROGRAM input, or pull the bidirectional INIT pin Low, using an open-collector (open-drain) driver. (See Figure 30.)

A Low on the PROGRAM input is the more radical approach, and is recommended when the power-supply rise time is excessive or poorly defined. As long as PROGRAM is Low, the FPGA keeps clearing its configuration memory. When PROGRAM goes High, the configuration memory is cleared one more time, followed by the beginning of configuration, provided the INIT input is not externally held Low. Note that a Low on the PROGRAM input automatically forces a Low on the INIT output. The Spartan/XL FPGA PROGRAM pin has a permanent weak pull-up.

Avoid holding PROGRAM Low for more than 500 µs. The 500 µs maximum limit is only a recommendation, not a requirement. The only effect of holding PROGRAM Low for more than 500 µs is an increase in current, measured at about 40 mA in the XCS40XL. This increased current cannot damage the device. This applies only during reconfiguration, not during power-up. The INIT pin can also be held Low to delay reconfiguration, and the same characteristics apply as for the PROGRAM pin.

Using an open-collector or open-drain driver to hold INIT Low before the beginning of configuration causes the FPGA

Spartan Family Detailed Specifications

Definition of Terms

In the following tables, some specifications may be designated as Advance or Preliminary. These terms are defined as follows:

Advance: Initial estimates based on simulation and/or extrapolation from other speed grades, devices, or families. Values are subject to change. Use as estimates, not for production.

Preliminary: Based on preliminary characterization. Further changes are not expected.

Unmarked: Specifications not identified as either Advance or Preliminary are to be considered Final.

Notwithstanding the definition of the above terms, all specifications are subject to change without notice.

Except for pin-to-pin input and output parameters, the AC parameter delay specifications included in this document are derived from measuring internal test patterns. All specifications are representative of worst-case supply voltage and junction temperature conditions. The parameters included are common to popular designs and typical applications.

Spartan Family Absolute Maximum Ratings(1)

Symbol	Description		Value	Units
V_{CC}	Supply voltage relative to GND		–0.5 to +7.0	V
V_{IN}	Input voltage relative to GND(2,3)		–0.5 to V_{CC} +0.5	V
V_{TS}	Voltage applied to 3-state output(2,3)		–0.5 to V_{CC} +0.5	V
T_{STG}	Storage temperature (ambient)		–65 to +150	°C
T_J	Junction temperature	Plastic packages	+125	°C

Notes:

1. Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those listed under Operating Conditions is not implied. Exposure to Absolute Maximum Ratings conditions for extended periods of time may affect device reliability.
2. Maximum DC overshoot (above V_{CC}) or undershoot (below GND) must be limited to either 0.5V or 10 mA, whichever is easier to achieve.
3. Maximum AC (during transitions) conditions are as follows; the device pins may undershoot to –2.0V or overshoot to +7.0V, provided this overshoot or undershoot lasts no more than 11 ns with a forcing current no greater than 100 mA.
4. For soldering guidelines, see the Package Information on the Xilinx website.

Spartan Family Recommended Operating Conditions

Symbol	Description		Min	Max	Units
V_{CC}	Supply voltage relative to GND, T_J = 0°C to +85°C	Commercial	4.75	5.25	V
	Supply voltage relative to GND, T_J = –40°C to +100°C(1)	Industrial	4.5	5.5	V
V_{IH}	High-level input voltage(2)	TTL inputs	2.0	V_{CC}	V
		CMOS inputs	70%	100%	V_{CC}
V_{IL}	Low-level input voltage(2)	TTL inputs	0	0.8	V
		CMOS inputs	0	20%	V_{CC}
T_{IN}	Input signal transition time		-	250	ns

Notes:

1. At junction temperatures above those listed as Recommended Operating Conditions, all delay parameters increase by 0.35% per °C.
2. Input and output measurement thresholds are: 1.5V for TTL and 2.5V for CMOS.

Spartan Family IOB Input Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. These path delays, provided as a guideline, have been extracted from the static timing analyzer report. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature).

Symbol	Description	Device	Speed Grade -4 Min	Speed Grade -4 Max	Speed Grade -3 Min	Speed Grade -3 Max	Units
Setup Times - TTL Inputs(1)							
T_{ECIK}	Clock Enable (EC) to Clock (IK), no delay	All devices	1.6	-	2.1	-	ns
T_{PICK}	Pad to Clock (IK), no delay	All devices	1.5	-	2.0	-	ns
Hold Times							
T_{IKEC}	Clock Enable (EC) to Clock (IK), no delay	All devices	0.0	-	0.9	-	ns
	All Other Hold Times	All devices	0.0	-	0.0	-	ns
Propagation Delays - TTL Inputs(1)							
T_{PID}	Pad to I1, I2	All devices	-	1.5	-	2.0	ns
T_{PLI}	Pad to I1, I2 via transparent input latch, no delay	All devices	-	2.8	-	3.6	ns
T_{IKRI}	Clock (IK) to I1, I2 (flip-flop)	All devices	-	2.7	-	2.8	ns
T_{IKLI}	Clock (IK) to I1, I2 (latch enable, active Low)	All devices	-	3.2	-	3.9	ns
Delay Adder for Input with Delay Option							
T_{Delay}	$T_{ECIKD} = T_{ECIK} + T_{Delay}$ $T_{PICKD} = T_{PICK} + T_{Delay}$ $T_{PDLI} = T_{PLI} + T_{Delay}$	XCS05	3.6	-	4.0	-	ns
		XCS10	3.7	-	4.1	-	ns
		XCS20	3.8	-	4.2	-	ns
		XCS30	4.5	-	5.0	-	ns
		XCS40	5.5	-	5.5	-	ns
Global Set/Reset							
T_{MRW}	Minimum GSR pulse width	All devices	11.5	-	13.5	-	ns
T_{RRI}	Delay from GSR input to any Q	XCS05	-	9.0	-	11.3	ns
		XCS10	-	9.5	-	11.9	ns
		XCS20	-	10.0	-	12.5	ns
		XCS30	-	10.5	-	13.1	ns
		XCS40	-	11.0	-	13.8	ns

Notes:

1. Delay adder for CMOS Inputs option: for -3 speed grade, add 0.4 ns; for -4 speed grade, add 0.2 ns.
2. Input pad setup and hold times are specified with respect to the internal clock (IK). For setup and hold times with respect to the clock input, see the pin-to-pin parameters in the Pin-to-Pin Input Parameters table.
3. Voltage levels of unused pads, bonded or unbonded, must be valid logic levels. Each can be configured with the internal pull-up (default) or pull-down resistor, or configured as a driven output, or can be driven from an external source.

Spartan Family IOB Output Switching Characteristic Guidelines

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. These path delays, provided as a guideline, have been extracted from the static timing analyzer report. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values are expressed in nanoseconds unless otherwise noted.

Symbol	Description	Device	Speed Grade				Units
			-4		-3		
			Min	Max	Min	Max	
Clocks							
T_{CH}	Clock High	All devices	3.0	-	4.0	-	ns
T_{CL}	Clock Low	All devices	3.0	-	4.0	-	ns
Propagation Delays - TTL Outputs$^{(1,2)}$							
T_{OKPOF}	Clock (OK) to Pad, fast	All devices	-	3.3	-	4.5	ns
T_{OKPOS}	Clock (OK to Pad, slew-rate limited	All devices	-	6.9	-	7.0	ns
T_{OPF}	Output (O) to Pad, fast	All devices	-	3.6	-	4.8	ns
T_{OPS}	Output (O) to Pad, slew-rate limited	All devices	-	7.2	-	7.3	ns
T_{TSHZ}	3-state to Pad High-Z (slew-rate independent)	All devices	-	3.0	-	3.8	ns
T_{TSONF}	3-state to Pad active and valid, fast	All devices	-	6.0	-	7.3	ns
T_{TSONS}	3-state to Pad active and valid, slew-rate limited	All devices	-	9.6	-	9.8	ns
Setup and Hold Times							
T_{OOK}	Output (O) to clock (OK) setup time	All devices	2.5	-	3.8	-	ns
T_{OKO}	Output (O) to clock (OK) hold time	All devices	0.0	-	0.0	-	ns
T_{ECOK}	Clock Enable (EC) to clock (OK) setup time	All devices	2.0	-	2.7	-	ns
T_{OKEC}	Clock Enable (EC) to clock (OK) hold time	All devices	0.0	-	0.5	-	ns
Global Set/Reset							
T_{MRW}	Minimum GSR pulse width	All devices	11.5		13.5		ns
T_{RPO}	Delay from GSR input to any Pad	XCS05	-	12.0	-	15.0	ns
		XCS10	-	12.5	-	15.7	ns
		XCS20	-	13.0	-	16.2	ns
		XCS30	-	13.5	-	16.9	ns
		XCS40	-	14.0	-	17.5	ns

Notes:

1. Delay adder for CMOS Outputs option (with fast slew rate option): for -3 speed grade, add 1.0 ns; for -4 speed grade, add 0.8 ns.
2. Delay adder for CMOS Outputs option (with slow slew rate option): for -3 speed grade, add 2.0 ns; for -4 speed grade, add 1.5 ns.
3. Output timing is measured at ~50% V_{CC} threshold, with 50 pF external capacitive loads including test fixture. Slew-rate limited output rise/fall times are approximately two times longer than fast output rise/fall times.
4. Voltage levels of unused pads, bonded or unbonded, must be valid logic levels. Each can be configured with the internal pull-up (default) or pull-down resistor, or configured as a driven output, or can be driven from an external source.

Pin Descriptions

There are three types of pins in the Spartan/XL devices:

- Permanently dedicated pins
- User I/O pins that can have special functions
- Unrestricted user-programmable I/O pins.

Before and during configuration, all outputs not used for the configuration process are 3-stated with the I/O pull-up resistor network activated. After configuration, if an IOB is unused it is configured as an input with the I/O pull-up resistor network remaining activated.

Any user I/O can be configured to drive the Global Set/Reset net GSR or the global three-state net GTS. See **Global Signals: GSR and GTS**, page 20 for more information.

Device pins for Spartan/XL devices are described in Table 18.

Some Spartan-XL devices are available in Pb-free package options. The Pb-free package options have the same pinouts as the standard package options.

Table 18: **Pin Descriptions**

Pin Name	I/O During Config.	I/O After Config.	Pin Description
Permanently Dedicated Pins			
V_{CC}	X	X	Eight or more (depending on package) connections to the nominal +5V supply voltage (+3.3V for Spartan-XL devices). All must be connected, and each must be decoupled with a 0.01 –0.1 μF capacitor to Ground.
GND	X	X	Eight or more (depending on package type) connections to Ground. All must be connected.
CCLK	I or O	I	During configuration, Configuration Clock (CCLK) is an output in Master mode and is an input in Slave mode. After configuration, CCLK has a weak pull-up resistor and can be selected as the Readback Clock. There is no CCLK High or Low time restriction on Spartan/XL devices, except during Readback. See **Violating the Maximum High and Low Time Specification for the Readback Clock**, page 39 for an explanation of this exception.
DONE	I/O	O	DONE is a bidirectional signal with an optional internal pull-up resistor. As an open-drain output, it indicates the completion of the configuration process. As an input, a Low level on DONE can be configured to delay the global logic initialization and the enabling of outputs. The optional pull-up resistor is selected as an option in the program that creates the configuration bitstream. The resistor is included by default.
$\overline{\text{PROGRAM}}$	I	I	$\overline{\text{PROGRAM}}$ is an active Low input that forces the FPGA to clear its configuration memory. It is used to initiate a configuration cycle. When $\overline{\text{PROGRAM}}$ goes High, the FPGA finishes the current clear cycle and executes another complete clear cycle, before it goes into a WAIT state and releases $\overline{\text{INIT}}$. The $\overline{\text{PROGRAM}}$ pin has a permanent weak pull-up, so it need not be externally pulled up to VCC.
MODE (Spartan) M0, M1 (Spartan-XL)	I	X	The Mode input(s) are sampled after $\overline{\text{INIT}}$ goes High to determine the configuration mode to be used. During configuration, these pins have a weak pull-up resistor. For the most popular configuration mode, Slave Serial, the mode pins can be left unconnected. For Master Serial mode, connect the Mode/M0 pin directly to system ground.

XCS30 and XCS30XL Device Pinouts *(Continued)*

XCS30/XL Pad Name	VQ100(5)	TQ144	PQ208	PQ240	BG256(5)	CS280(2,5)	Bndry Scan
I/O	-	P5	P5	P5	D3	C1	155
I/O, TDI	P4	P6	P6	P6	E4	D4	158
I/O, TCK	P5	P7	P7	P7	C1	D3	161
I/O	-	-	P8	P8	D1	E2	164
I/O	-	-	P9	P9	E3	E4	167
I/O	-	-	P10	P10	E2	E1	170
I/O	-	-	P11	P11	E1	F5	173
I/O	-	-	P12	P12	F3	F3	176
I/O	-	-	-	P13	F2	F2	179
GND	-	P8	P13	P14	GND(4)	GND(4)	-
I/O	-	P9	P14	P15	G3	F4	182
I/O	-	P10	P15	P16	G2	F1	185
I/O, TMS	P6	P11	P16	P17	G1	G3	188
I/O	P7	P12	P17	P18	H3	G2	191
VCC	-	-	P18	P19	VCC(4)	G1	-
I/O	-	-	-	P20	H2	G4	194
I/O	-	-	-	P21	H1	H1	197
I/O	-	-	P19	P23	J2	H4	200
I/O	-	-	P20	P24	J1	J1	203
I/O	-	P13	P21	P25	K2	J2	206
I/O	P8	P14	P22	P26	K3	J3	209
I/O	P9	P15	P23	P27	K1	J4	212
I/O	P10	P16	P24	P28	L1	K1	215
GND	P11	P17	P25	P29	GND(4)	GND(4)	-
VCC	P12	P18	P26	P30	VCC(4)	K2	-
I/O	P13	P19	P27	P31	L2	K3	218
I/O	P14	P20	P28	P32	L3	K4	221
I/O	P15	P21	P29	P33	L4	K5	224
I/O	-	P22	P30	P34	M1	L1	227
I/O	-	-	P31	P35	M2	L2	230
I/O	-	-	P32	P36	M3	L3	233
I/O	-	-	-	P38	N1	M2	236
I/O	-	-	-	P39	N2	M3	239
VCC	-	-	P33	P40	VCC(4)	M4	-
I/O	P16	P23	P34	P41	P1	N1	242
I/O	P17	P24	P35	P42	P2	N2	245
I/O	-	P25	P36	P43	R1	N3	248
I/O	-	P26	P37	P44	P3	N4	251
GND	-	P27	P38	P45	GND(4)	GND(4)	-
I/O	-	-	-	P46	T1	P1	254
I/O	-	-	P39	P47	R3	P2	257
I/O	-	-	P40	P48	T2	P3	260
I/O	-	-	P41	P49	U1	P4	263
I/O	-	-	P42	P50	T3	P5	266
I/O	-	-	P43	P51	U2	R1	269

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
I/O	P90	P105	Y16	W14	466 (3)
GND	P91	P106	GND(4)	GND(4)	-
I/O	-	P107	V15	V14	469 (3)
I/O	P92	P108	W16	U14	472 (3)
I/O	P93	P109	Y17	T14	475 (3)
I/O	P94	P110	V16	R14	478 (3)
I/O	P95	P111	W17	W15	481 (3)
I/O	P96	P112	Y18	U15	484 (3)
I/O	-	-	-	T15	487 (3)
I/O	-	-	-	W16	490 (3)
I/O	P97	P113	U16	V16	493 (3)
I/O	P98	P114	V17	U16	496 (3)
I/O	P99	P115	W18	W17	499 (3)
I/O	P100	P116	Y19	W18	502 (3)
I/O	P101	P117	V18	V17	505 (3)
I/O, SGCK3(1), GCK4(2)	P102	P118	W19	V18	508 (3)
GND	P103	P119	GND(4)	GND(4)	-
DONE	P104	P120	Y20	W19	-
VCC	P105	P121	VCC(4)	VCC(4)	-
$\overline{\text{PROGRAM}}$	P106	P122	V19	U18	-
I/O (D7(2))	P107	P123	U19	V19	511 (3)
I/O, PGCK3(1), GCK5(2)	P108	P124	U18	U19	514 (3)
I/O	P109	P125	T17	T16	517 (3)
I/O	P110	P126	V20	T17	520 (3)
I/O	-	P127	U20	T18	523 (3)
I/O	P111	P128	T18	T19	526 (3)
I/O	-	-	-	R15	529 (3)
I/O	-	-	-	R17	523 (3)
I/O (D6(2))	P112	P129	T19	R16	535 (3)
I/O	P113	P130	T20	R19	538 (3)
I/O	P114	P131	R18	P15	541 (3)
I/O	P115	P132	R19	P17	544 (3)
I/O	P116	P133	R20	P18	547 (3)
I/O	P117	P134	P18	P16	550 (3)
GND	P118	P135	GND(4)	GND(4)	-
I/O	-	P136	P20	P19	553 (3)
I/O	-	P137	N18	N17	556 (3)
I/O	P119	P138	N19	N18	559 (3)
I/O	P120	P139	N20	N19	562 (3)
VCC	P121	P140	VCC(4)	VCC(4)	-
I/O (D5(2))	P122	P141	M17	M19	565 (3)
I/O	P123	P142	M18	M17	568 (3)

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
I/O	-	-	-	M18	571 (3)
I/O	-	-	M19	M16	574 (3)
I/O	P124	P144	M20	L19	577 (3)
I/O	P125	P145	L19	L18	580 (3)
I/O	P126	P146	L18	L17	583 (3)
I/O	P127	P147	L20	L16	586 (3)
I/O (D4(2))	P128	P148	K20	K19	589 (3)
I/O	P129	P149	K19	K18	592 (3)
VCC	P130	P150	VCC(4)	VCC(4)	-
GND	P131	P151	GND(4)	GND(4)	-
I/O (D3(2))	P132	P152	K18	K16	595 (3)
I/O	P133	P153	K17	K15	598 (3)
I/O	P134	P154	J20	J19	601 (3)
I/O	P135	P155	J19	J18	604 (3)
I/O	P136	P156	J18	J17	607 (3)
I/O	P137	P157	J17	J16	610 (3)
I/O	-	-	H20	H19	613 (3)
I/O	-	-	-	H18	616 (3)
I/O (D2(2))	P138	P159	H19	H17	619 (3)
I/O	P139	P160	H18	H16	622 (3)
VCC	P140	P161	VCC(4)	VCC(4)	-
I/O	P141	P162	G19	G18	625 (3)
I/O	P142	P163	F20	G17	628 (3)
I/O	-	P164	G18	G16	631 (3)
I/O	-	P165	F19	F19	634 (3)
GND	P143	P166	GND(4)	GND(4)	-
I/O	-	P167	F18	F18	637 (3)
I/O	P144	P168	E19	F17	640 (3)
I/O	P145	P169	D20	F16	643 (3)
I/O	P146	P170	E18	F15	646 (3)
I/O	P147	P171	D19	E19	649 (3)
I/O	P148	P172	C20	E17	652 (3)
I/O (D1(2))	P149	P173	E17	E16	655 (3)
I/O	P150	P174	D18	D19	658 (3)
I/O	-	-	-	D18	661 (3)
I/O	-	-	-	D17	664 (3)
I/O	P151	P175	C19	C19	667 (3)
I/O	P152	P176	B20	B19	670 (3)
I/O (D0(2), DIN)	P153	P177	C18	C18	673 (3)
I/O, SGCK4(1), GCK6(2) (DOUT)	P154	P178	B19	B18	676 (3)
CCLK	P155	P179	A20	A19	-
VCC	P156	P180	VCC(4)	VCC(4)	-

XCS40 and XCS40XL Device Pinouts

XCS40/XL Pad Name	PQ208	PQ240	BG256	CS280(2,5)	Bndry Scan
O, TDO	P157	P181	A19	B17	0
GND	P158	P182	GND(4)	GND(4)	-
I/O	P159	P183	B18	A18	2
I/O, PGCK4(1), GCK7(2)	P160	P184	B17	A17	5
I/O	P161	P185	C17	D16	8
I/O	P162	P186	D16	C16	11
I/O (CS1(2))	P163	P187	A18	B16	14
I/O	P164	P188	A17	A16	17
I/O	-	-	-	E15	20
I/O	-	-	-	C15	23
I/O	P165	P189	C16	D15	26
I/O	-	P190	B16	A15	29
I/O	P166	P191	A16	E14	32
I/O	P167	P192	C15	C14	35
I/O	P168	P193	B15	B14	38
I/O	P169	P194	A15	D14	41
GND	P170	P196	GND(4)	GND(4)	-
I/O	P171	P197	B14	A14	44
I/O	P172	P198	A14	C13	47
I/O	-	P199	C13	B13	50
I/O	-	P200	B13	A13	53
VCC	P173	P201	VCC(4)	VCC(4)	-
I/O	-	-	A13	A12	56
I/O	-	-	D12	C12	59
I/O	P174	P202	C12	B12	62
I/O	P175	P203	B12	D12	65
I/O	P176	P205	A12	A11	68
I/O	P177	P206	B11	B11	71
I/O	P178	P207	C11	C11	74
I/O	P179	P208	A11	D11	77
I/O	P180	P209	A10	A10	80
I/O	P181	P210	B10	B10	83
GND	P182	P211	GND(4)	GND(4)	-

2/8/00

Notes:

1. 5V Spartan family only
2. 3V Spartan-XL family only
3. The “PWRDWN” on the XCS40XL is not part of the Boundary Scan chain. For the XCS40XL, subtract 1 from all Boundary Scan numbers from GCK3 on (343 and higher).
4. Pads labeled GND(4) or V_{CC}(4) are internally bonded to Ground or V_{CC} planes within the package.
5. CS280 package discontinued by PDN2004-01

Additional XCS40/XL Package Pins

PQ240

GND Pins					
P22	P37	P83	P98	P143	P158
P204	P219	-	-	-	-
Not Connected Pins					
P195	-	-	-	-	-

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BG256

VCC Pins					
C14	D6	D7	D11	D14	D15
E20	F1	F4	F17	G4	G17
K4	L17	P4	P17	P19	R2
R4	R17	U6	U7	U10	U14
U15	V7	W20	-	-	-
GND Pins					
A1	B7	D4	D8	D13	D17
G20	H4	H17	N3	N4	N17
U4	U8	U13	U17	W14	-

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CS280

VCC Pins					
A1	A7	B5	B15	C10	C17
D13	E3	E18	G1	G19	K2
K17	M4	N16	R3	R18	T7
U3	U10	U17	V5	V15	W13
GND Pins					
E5	E7	E8	E9	E11	E12
E13	G5	G15	H5	H15	J5
J15	L5	L15	M5	M15	N5
N15	R7	R8	R9	R11	R12
R13	-	-	-	-	-

5/19/99