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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1452
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	180
Number of Gates	24000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	-55°C ~ 125°C (TJ)
Package / Case	256-BFCQFP Exposed Pad and Tie Bar
Supplier Device Package	256-CQFP (75x75)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/5962-9956903qxc

SX Family FPGAs

General Description

The Actel SX family of FPGAs features a sea-of-modules architecture that delivers device performance and integration levels not currently achieved by any other FPGA architecture. SX devices greatly simplify design time, enable dramatic reductions in design costs and power consumption, and further decrease time to market for performance-intensive applications.

The Actel SX architecture features two types of logic modules, the combinatorial cell (C-cell) and the register cell (R-cell), each optimized for fast and efficient mapping of synthesized logic functions. The routing and interconnect resources are in the metal layers above the logic modules, providing optimal use of silicon. This enables the entire floor of the device to be spanned with an uninterrupted grid of fine-grained, synthesis-friendly logic modules (or "sea-of-modules"), which reduces the distance signals have to travel between logic modules. To minimize signal propagation delay, SX devices employ both local and general routing resources. The high-speed local routing resources (DirectConnect and FastConnect) enable very fast local signal propagation that is optimal for fast counters, state machines, and datapath logic. The general system of segmented routing tracks allows any logic module in the array to be connected to any other logic or I/O module. Within this system, propagation delay is minimized by limiting the number of antifuse interconnect elements to five (90 percent of connections typically use only three antifuses). The unique local and general routing structure featured in SX devices gives fast and predictable performance, allows 100 percent pin-locking with full logic utilization, enables concurrent PCB development, reduces design time, and allows designers to achieve performance goals with minimum effort.

Further complementing SX's flexible routing structure is a hardwired, constantly loaded clock network that has been tuned to provide fast clock propagation with minimal clock skew. Additionally, the high performance of the internal logic has eliminated the need to embed latches or flip-flops in the I/O cells to achieve fast clock-to-out or fast input setup times. SX devices have easy to use I/O cells that do not require HDL instantiation, facilitating design reuse and reducing design and verification time.

SX Family Architecture

The SX family architecture was designed to satisfy next-generation performance and integration requirements for production-volume designs in a broad range of applications.

Programmable Interconnect Element

The SX family provides efficient use of silicon by locating the routing interconnect resources between the Metal 2 (M2) and Metal 3 (M3) layers (Figure 1-1 on page 1-2). This completely eliminates the channels of routing and interconnect resources between logic modules (as implemented on SRAM FPGAs and previous generations of antifuse FPGAs), and enables the entire floor of the device to be spanned with an uninterrupted grid of logic modules.

Interconnection between these logic modules is achieved using The Actel patented metal-to-metal programmable antifuse interconnect elements, which are embedded between the M2 and M3 layers. The antifuses are normally open circuit and, when programmed, form a permanent low-impedance connection.

The extremely small size of these interconnect elements gives the SX family abundant routing resources and provides excellent protection against design pirating. Reverse engineering is virtually impossible because it is extremely difficult to distinguish between programmed and unprogrammed antifuses, and there is no configuration bitstream to intercept.

Additionally, the interconnect elements (i.e., the antifuses and metal tracks) have lower capacitance and lower resistance than any other device of similar capacity, leading to the fastest signal propagation in the industry.

Logic Module Design

The SX family architecture is described as a "sea-of-modules" architecture because the entire floor of the device is covered with a grid of logic modules with virtually no chip area lost to interconnect elements or routing. The Actel SX family provides two types of logic modules, the register cell (R-cell) and the combinatorial cell (C-cell).

Chip Architecture

The SX family chip architecture provides a unique approach to module organization and chip routing that delivers the best register/logic mix for a wide variety of new and emerging applications.

Module Organization

Actel has arranged all C-cell and R-cell logic modules into horizontal banks called *clusters*. There are two types of *clusters*: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells.

To increase design efficiency and device performance, Actel has further organized these modules into *SuperClusters* (Figure 1-4). SuperCluster 1 is a two-wide grouping of Type 1 clusters. SuperCluster 2 is a two-wide group containing one Type 1 cluster and one Type 2 cluster. SX devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flip-flops.

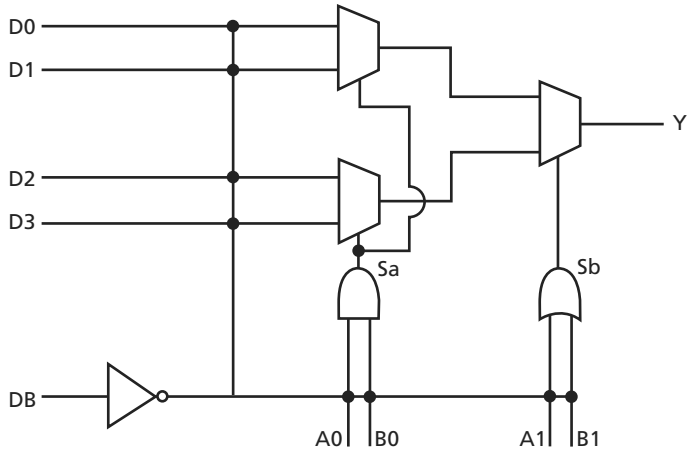


Figure 1-3 • **C-Cell**

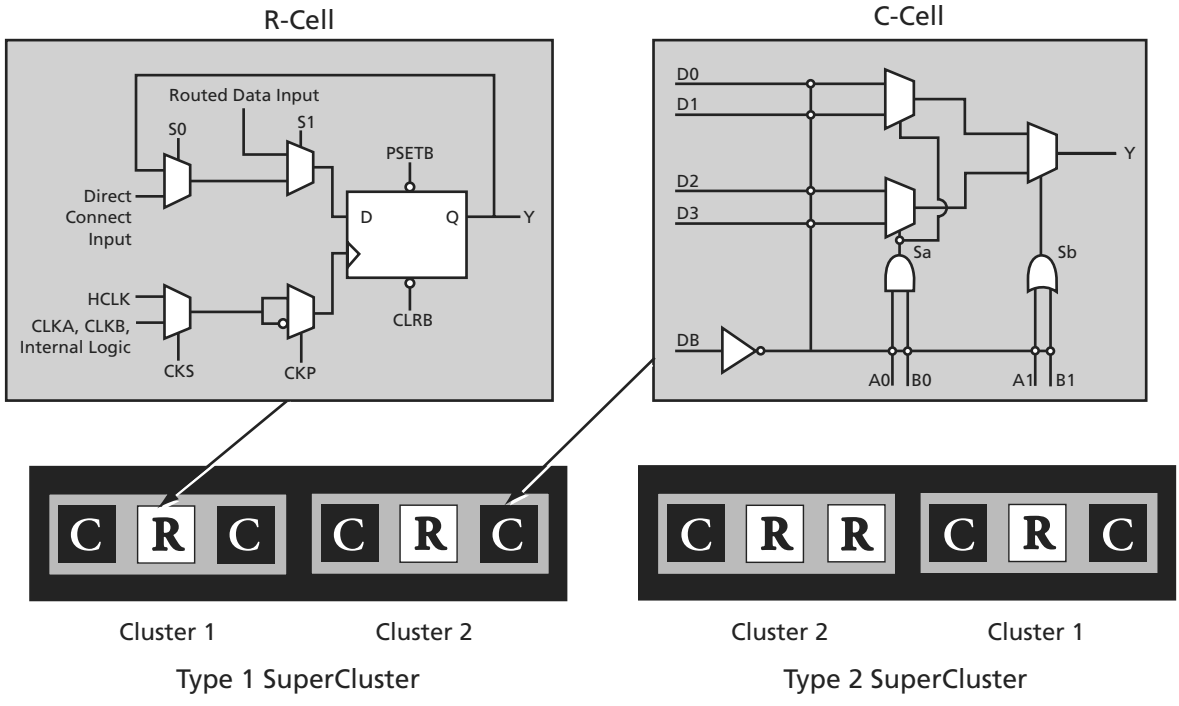


Figure 1-4 • **Cluster Organization**

Routing Resources

Clusters and SuperClusters can be connected through the use of two innovative local routing resources called *FastConnect* and *DirectConnect,* which enable extremely fast and predictable interconnection of modules within clusters and SuperClusters (Figure 1-5 and Figure 1-6). This routing architecture also dramatically reduces the number of antifuses required to complete a circuit, ensuring the highest possible performance.

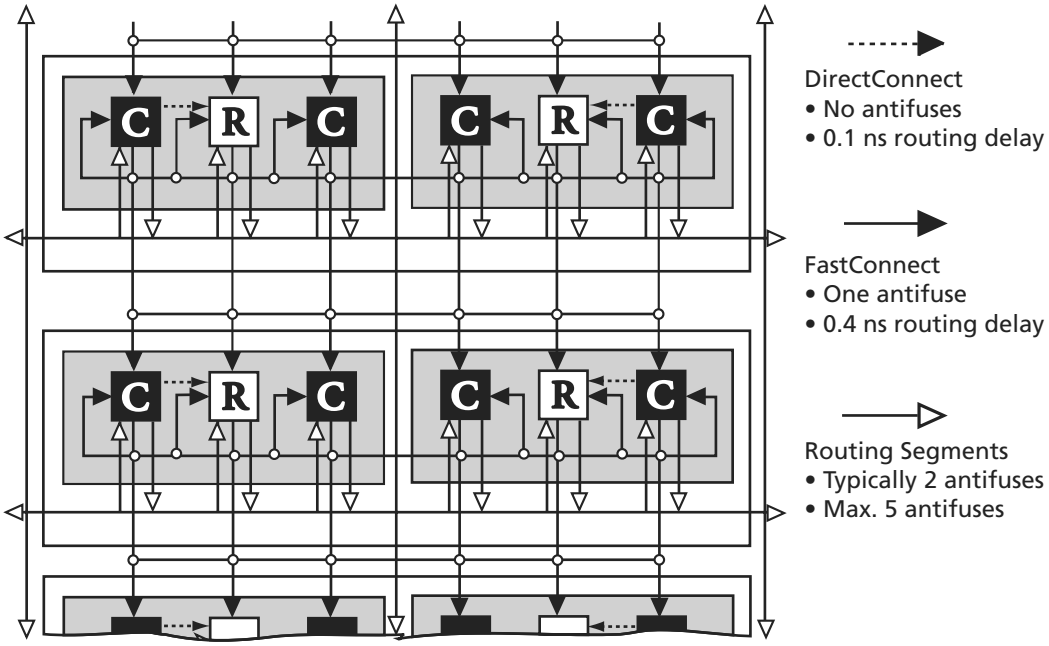


Figure 1-5 • **DirectConnect and FastConnect for Type 1 SuperClusters**

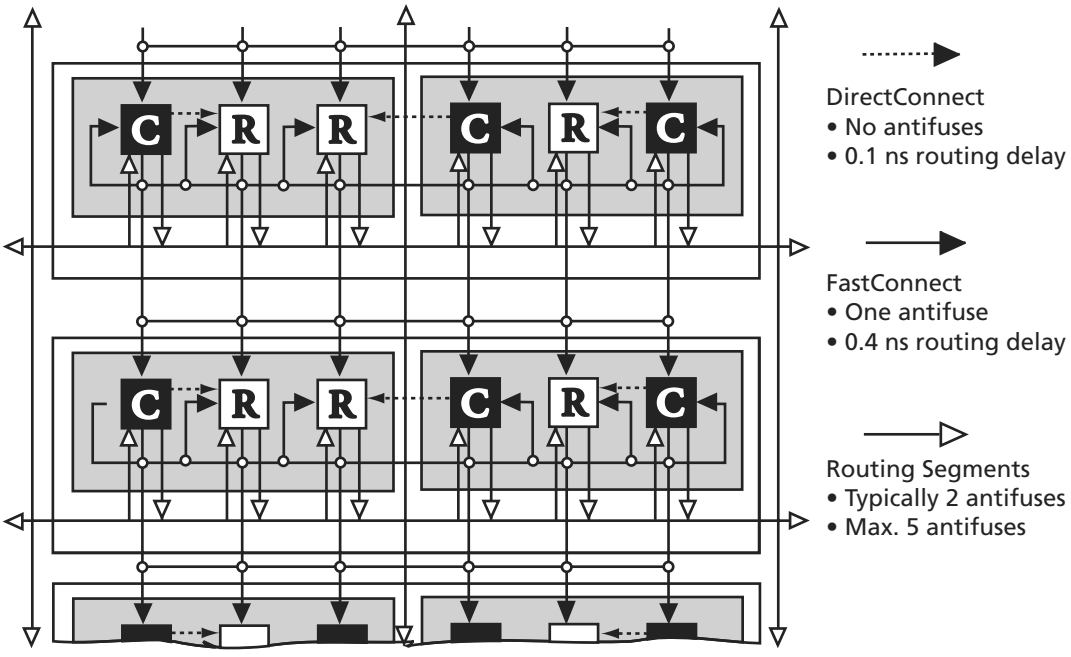


Figure 1-6 • **DirectConnect and FastConnect for Type 2 SuperClusters**

Boundary Scan Testing (BST)

All SX devices are IEEE 1149.1 compliant. SX devices offer superior diagnostic and testing capabilities by providing Boundary Scan Testing (BST) and probing capabilities. These functions are controlled through the special test pins in conjunction with the program fuse. The functionality of each pin is described in Table 1-2. In the dedicated test mode, TCK, TDI, and TDO are dedicated pins and cannot be used as regular I/Os. In flexible mode, TMS should be set HIGH through a pull-up resistor of 10 kΩ. TMS can be pulled LOW to initiate the test sequence.

The program fuse determines whether the device is in dedicated or flexible mode. The default (fuse not blown) is flexible mode.

Table 1-2 • **Boundary Scan Pin Functionality**

Program Fuse Blown (Dedicated Test Mode)	**Program Fuse Not Blown (Flexible Mode)**
TCK, TDI, TDO are dedicated BST pins.	TCK, TDI, TDO are flexible and may be used as I/Os.
No need for pull-up resistor for TMS	Use a pull-up resistor of 10 kΩ on TMS.

Dedicated Test Mode

In Dedicated mode, all JTAG pins are reserved for BST; designers cannot use them as regular I/Os. An internal pull-up resistor is automatically enabled on both TMS and TDI pins, and the TMS pin will function as defined in the IEEE 1149.1 (JTAG) specification.

To select Dedicated mode, users need to reserve the JTAG pins in Actel's Designer software by checking the "Reserve JTAG" box in "Device Selection Wizard" (Figure 1-7). JTAG pins comply with LVTTL/TTL I/O specification regardless of whether they are used as a user I/O or a JTAG I/O. Refer to the Table 1-5 on page 1-8 for detailed specifications.

Figure 1-7 • **Device Selection Wizard**

Development Tool Support

The SX family of FPGAs is fully supported by both the Actel Libero® Integrated Design Environment (IDE) and Designer FPGA Development software. Actel Libero IDE is a design management environment, seamlessly integrating design tools while guiding the user through the design flow, managing all design and log files, and passing necessary design data among tools. Libero IDE allows users to integrate both schematic and HDL synthesis into a single flow and verify the entire design in a single environment. Libero IDE includes Synplify® for Actel from Synplicity®, ViewDraw® for Actel from Mentor Graphics®, Model*Sim*® HDL Simulator from Mentor Graphics, WaveFormer Lite™ from SynaptiCAD™, and Designer software from Actel. Refer to the Libero IDE flow diagram (located on the Actel website) for more information.

Actel Designer software is a place-and-route tool and provides a comprehensive suite of backend support tools for FPGA development. The Designer software includes timing-driven place-and-route, and a world-class integrated static timing analyzer and constraints editor. With the Designer software, a user can select and lock package pins while only minimally impacting the results of place-and-route. Additionally, the back-annotation flow is compatible with all the major simulators, and the simulation results can be cross-probed with Silicon Explorer II, Actel integrated verification and logic analysis tool. Another tool included in the Designer software is the SmartGen core generator, which easily creates popular and commonly used logic functions for implementation into your schematic or HDL design. Actel Designer software is compatible with the most popular FPGA design entry and verification tools from companies such as Mentor Graphics, Synplicity, Synopsys®, and Cadence® Design Systems. The Designer software is available for both the Windows® and UNIX® operating systems.

Probe Circuit Control Pins

The Silicon Explorer II tool uses the boundary scan ports (TDI, TCK, TMS, and TDO) to select the desired nets for verification. The selected internal nets are assigned to the PRA/PRB pins for observation. Figure 1-8 on page 1-7 illustrates the interconnection between Silicon Explorer II and the FPGA to perform in-circuit verification.

Design Considerations

The TDI, TCK, TDO, PRA, and PRB pins should not be used as input or bidirectional ports. Because these pins are active during probing, critical signals input through these pins are not available while probing. In addition, the Security Fuse should not be programmed because doing so disables the Probe Circuitry.

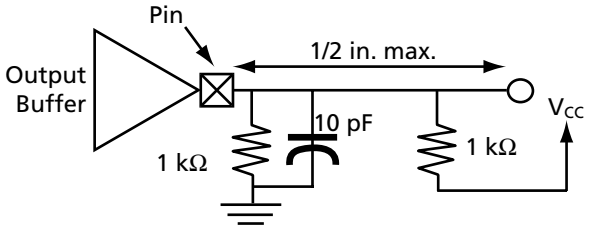
A54SX16P AC Specifications for (PCI Operation)

Table 1-7 • **A54SX16P AC Specifications for (PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \le 1.4$[1]	–44		mA
		$1.4 \le V_{OUT} < 2.4$[1, 2]	$–44 + (V_{OUT} – 1.4)/0.024$		mA
		$3.1 < V_{OUT} < V_{CC}$[1, 3]		EQ 1-1 on page 1-11	
	(Test Point)	$V_{OUT} = 3.1$[3]		–142	mA
$I_{OL(AC)}$	Switching Current High	$V_{OUT} \ge 2.2$[1]	95		mA
		$2.2 > V_{OUT} > 0.55$[1]	$V_{OUT}/0.023$		
		$0.71 > V_{OUT} > 0$[1, 3]		EQ 1-2 on page 1-11	mA
	(Test Point)	$V_{OUT} = 0.71$[3]		206	mA
I_{CL}	Low Clamp Current	$–5 < V_{IN} \le –1$	$–25 + (V_{IN} + 1)/0.015$		mA
$slew_R$	Output Rise Slew Rate	0.4 V to 2.4 V load[4]	1	5	V/ns
$slew_F$	Output Fall Slew Rate	2.4 V to 0.4 V load[4]	1	5	V/ns

Notes:

1. *Refer to the V/I curves in Figure 1-9 on page 1-11. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half-size output drivers may be used on these signals. This specification does not apply to CLK and RST#, which are system outputs. "Switching Current High" specifications are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD#, which are open drain outputs.*
2. *Note that this segment of the minimum current curve is drawn from the AC drive point directly to the DC drive point rather than toward the voltage rail (as is done in the pull-down curve). This difference is intended to allow for an optional N-channel pull-up.*
3. *Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (A and B) are provided with the respective diagrams in Figure 1-9 on page 1-11. The equation defined maxima should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.*
4. *This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per revision 2.0 of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is now required (the maximum is no longer simply a guideline). Since adherence to the maximum slew rate was not required prior to revision 2.1 of the specification, there may be components in the market for some time that have faster edge rates; therefore, motherboard designers must bear in mind that rise and fall times faster than this specification could occur, and should ensure that signal integrity modeling accounts for this. Rise slew rate does not apply to open drain outputs.*



Evaluating Power in SX Devices

A critical element of system reliability is the ability of electronic devices to safely dissipate the heat generated during operation. The thermal characteristics of a circuit depend on the device and package used, the operating temperature, the operating current, and the system's ability to dissipate heat.

You should complete a power evaluation early in the design process to help identify potential heat-related problems in the system and to prevent the system from exceeding the device's maximum allowed junction temperature.

The actual power dissipated by most applications is significantly lower than the power the package can dissipate. However, a thermal analysis should be performed for all projects. To perform a power evaluation, follow these steps:

1. Estimate the power consumption of the application.
2. Calculate the maximum power allowed for the device and package.
3. Compare the estimated power and maximum power values.

Estimating Power Consumption

The total power dissipation for the SX family is the sum of the DC power dissipation and the AC power dissipation. Use EQ 1-5 to calculate the estimated power consumption of your application.

$$P_{Total} = P_{DC} + P_{AC}$$

EQ 1-5

DC Power Dissipation

The power due to standby current is typically a small component of the overall power. The Standby power is shown in Table 1-12 for commercial, worst-case conditions (70°C).

Table 1-12 • **Standby Power**

I_{CC}	V_{CC}	Power
4 mA	3.6 V	14.4 mW

The DC power dissipation is defined in EQ 1-6.

$$P_{DC} = (I_{standby}) \times V_{CCA} + (I_{standby}) \times V_{CCR} + (I_{standby}) \times V_{CCI} + xV_{OL} \times I_{OL} + y(V_{CCI} - V_{OH}) \times V_{OH}$$

EQ 1-6

AC Power Dissipation

The power dissipation of the SX Family is usually dominated by the dynamic power dissipation. Dynamic power dissipation is a function of frequency, equivalent capacitance, and power supply voltage. The AC power dissipation is defined in EQ 1-7 and EQ 1-8.

$$P_{AC} = P_{Module} + P_{RCLKA\ Net} + P_{RCLKB\ Net} + P_{HCLK\ Net} + P_{Output\ Buffer} + P_{Input\ Buffer}$$

EQ 1-7

$$P_{AC} = V_{CCA}^2 \times [(m \times C_{EQM} \times f_m)_{Module} + (n \times C_{EQI} \times f_n)_{Input\ Buffer} + (p \times (C_{EQO} + C_L) \times f_p)_{Output\ Buffer} + (0.5 \times (q_1 \times C_{EQCR} \times f_{q1}) + (r_1 \times f_{q1}))_{RCLKA} + (0.5 \times (q2 \times CEQCR \times f_{q2}) + (r2 \times f_{q2}))RCLKB + (0.5 \times (s_1 \times C_{EQHV} \times f_{s1}) + (C_{EQHF} \times f_{s1}))_{HCLK}]$$

EQ 1-8

Definition of Terms Used in Formula

m = Number of logic modules switching at f_m
n = Number of input buffers switching at f_n
p = Number of output buffers switching at f_p
q_1 = Number of clock loads on the first routed array clock
q_2 = Number of clock loads on the second routed array clock
x = Number of I/Os at logic low
y = Number of I/Os at logic high
r_1 = Fixed capacitance due to first routed array clock
r_2 = Fixed capacitance due to second routed array clock
s_1 = Number of clock loads on the dedicated array clock
C_{EQM} = Equivalent capacitance of logic modules in pF
C_{EQI} = Equivalent capacitance of input buffers in pF
C_{EQO} = Equivalent capacitance of output buffers in pF
C_{EQCR} = Equivalent capacitance of routed array clock in pF
C_{EQHV} = Variable capacitance of dedicated array clock
C_{EQHF} = Fixed capacitance of dedicated array clock
C_L = Output lead capacitance in pF
f_m = Average logic module switching rate in MHz
f_n = Average input buffer switching rate in MHz
f_p = Average output buffer switching rate in MHz
f_{q1} = Average first routed array clock rate in MHz
f_{q2} = Average second routed array clock rate in MHz
f_{s1} = Average dedicated array clock rate in MHz

Table 1-13 shows capacitance values for various devices.

Table 1-13 • **Capacitance Values for Devices**

	A54SX08	A54SX16	A54SX16P	A54SX32
C_{EQM} (pF)	4.0	4.0	4.0	4.0
C_{EQI} (pF)	3.4	3.4	3.4	3.4
C_{EQO} (pF)	4.7	4.7	4.7	4.7
C_{EQCR} (pF)	1.6	1.6	1.6	1.6
C_{EQHV}	0.615	0.615	0.615	0.615
C_{EQHF}	60	96	96	140
r_1 (pF)	87	138	138	171
r_2 (pF)	87	138	138	171

Guidelines for Calculating Power Consumption

The power consumption guidelines are meant to represent worst-case scenarios so that they can be generally used to predict the upper limits of power dissipation. These guidelines are shown in Table 1-14.

Sample Power Calculation

One of the designs used to characterize the SX family was a 528 bit serial-in, serial-out shift register. The design utilized 100 percent of the dedicated flip-flops of an A54SX16P device. A pattern of 0101… was clocked into the device at frequencies ranging from 1 MHz to 200 MHz. Shifting in a series of 0101… caused 50 percent of the flip-flops to toggle from low to high at every clock cycle.

Table 1-14 • **Power Consumption Guidelines**

Description	Power Consumption Guideline
Logic Modules (m)	20% of modules
Inputs Switching (n)	# inputs/4
Outputs Switching (p)	# outputs/4
First Routed Array Clock Loads (q_1)	20% of register cells
Second Routed Array Clock Loads (q_2)	20% of register cells
Load Capacitance (C_L)	35 pF
Average Logic Module Switching Rate (f_m)	f/10
Average Input Switching Rate (f_n)	f/5
Average Output Switching Rate (f_p)	f/10
Average First Routed Array Clock Rate (f_{q1})	f/2
Average Second Routed Array Clock Rate (f_{q2})	f/2
Average Dedicated Array Clock Rate (f_{s1})	f
Dedicated Clock Array Clock Loads (s_1)	20% of regular modules

Follow the steps below to estimate power consumption. The values provided for the sample calculation below are for the shift register design above. This method for estimating power consumption is conservative and the actual power consumption of your design may be less than the estimated power consumption.

The total power dissipation for the SX family is the sum of the AC power dissipation and the DC power dissipation.

$$P_{Total} = P_{AC} \text{ (dynamic power)} + P_{DC} \text{ (static power)}$$

EQ 1-9

AC Power Dissipation

$$P_{AC} = P_{Module} + P_{RCLKA\ Net} + P_{RCLKB\ Net} + P_{HCLK\ Net} + P_{Output\ Buffer} + P_{Input\ Buffer}$$

EQ 1-10

$$P_{AC} = V_{CCA}^2 \times [(m \times C_{EQM} \times f_m)_{Module} + (n \times C_{EQI} \times f_n)_{Input\ Buffer} + (p \times (C_{EQO} + C_L) \times f_p)_{Output\ Buffer} + (0.5\ (q_1 \times C_{EQCR} \times f_{q1}) + (r_1 \times f_{q1}))_{RCLKA} + (0.5\ (q_2 \times C_{EQCR} \times f_{q2}) + (r_2 \times f_{q2}))_{RCLKB} + (0.5\ (s_1 \times C_{EQHV} \times f_{s1}) + (C_{EQHF} \times f_{s1}))_{HCLK}]$$

EQ 1-11

Table 1-15 • **Package Thermal Characteristics**

Package Type	Pin Count	θ_{jc}	θ_{ja} Still Air	θ_{ja} 300 ft/min.	Units
Plastic Leaded Chip Carrier (PLCC)	84	12	32	22	°C/W
Thin Quad Flat Pack (TQFP)	144	11	32	24	°C/W
Thin Quad Flat Pack (TQFP)	176	11	28	21	°C/W
Very Thin Quad Flatpack (VQFP)	100	10	38	32	°C/W
Plastic Quad Flat Pack (PQFP) without Heat Spreader	208	8	30	23	°C/W
Plastic Quad Flat Pack (PQFP) with Heat Spreader	208	3.8	20	17	°C/W
Plastic Ball Grid Array (PBGA)	272	3	20	14.5	°C/W
Plastic Ball Grid Array (PBGA)	313	3	23	17	°C/W
Plastic Ball Grid Array (PBGA)	329	3	18	13.5	°C/W
Fine Pitch Ball Grid Array (FBGA)	144	3.8	38.8	26.7	°C/W

Note: *SX08 does not have a heat spreader.*

Table 1-16 • **Temperature and Voltage Derating Factors***

	Junction Temperature						
V_{CCA}	–55	–40	0	25	70	85	125
3.0	0.75	0.78	0.87	0.89	1.00	1.04	1.16
3.3	0.70	0.73	0.82	0.83	0.93	0.97	1.08
3.6	0.66	0.69	0.77	0.78	0.87	0.92	1.02

Note: **Normalized to worst-case commercial, T_J = 70°C, V_{CCA} = 3.0 V*

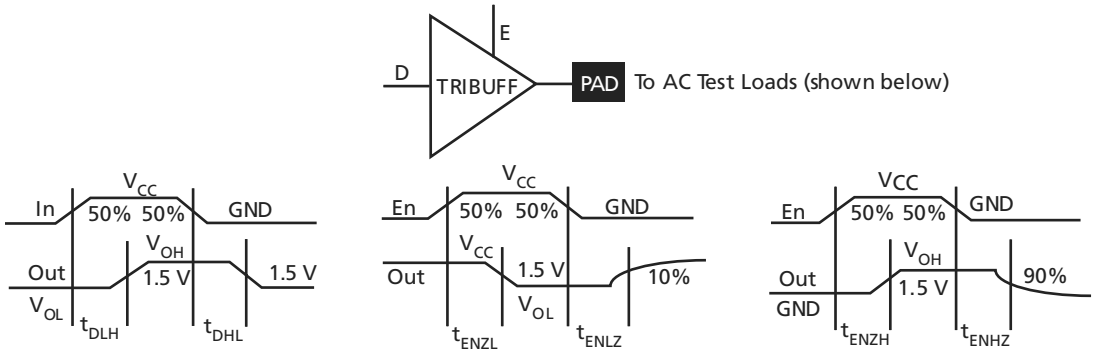


Figure 1-13 • **Output Buffer Delays**

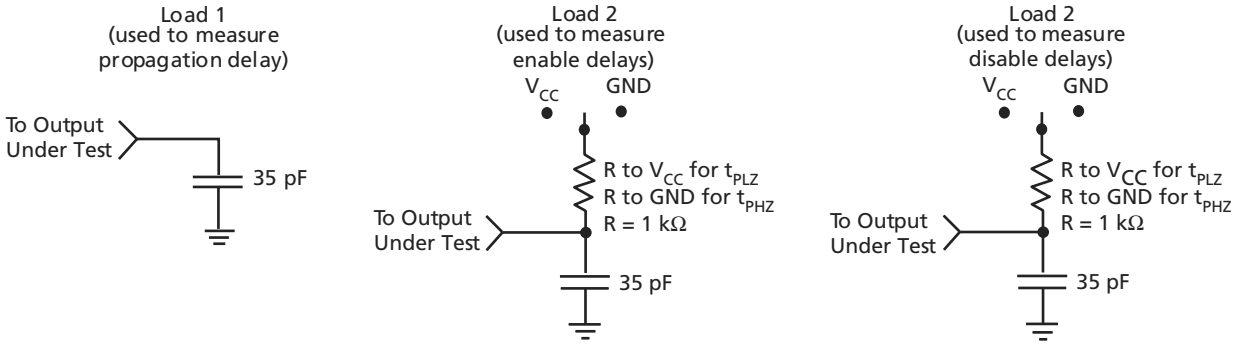


Figure 1-14 • **AC Test Loads**

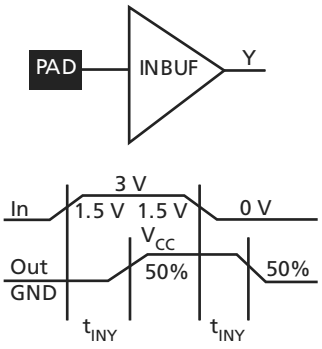


Figure 1-15 • **Input Buffer Delays**

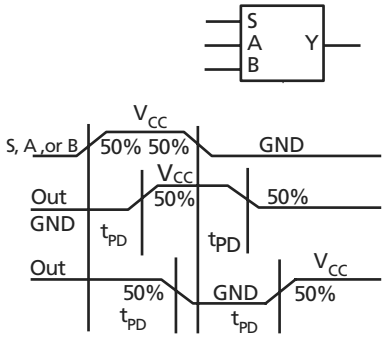


Figure 1-16 • **C-Cell Delays**

A54SX16 Timing Characteristics

Table 1-18 • **A54SX16 Timing Characteristics (Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)**

Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
C-Cell Propagation Delays[1]										
t_{PD}	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted Routing Delays[2]										
t_{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t_{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{RD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{RD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timing										
t_{RCO}	Sequential Clock-to-Q		0.8		1.1		1.2		1.4	ns
t_{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t_{PRESET}	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t_{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t_{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t_{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t_{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted Input Routing Delays[2]										
t_{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{IRD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{IRD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Notes:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH}. For t_{ENZL} and t_{ENZH}, the loading is 5 pF.*

Table 1-19 • **A54SX16P Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
TTL/PCI Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		1.5		1.7		2.0		2.3	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.9		2.2		2.4		2.9	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.3		2.6		3.0		3.5	ns
t_{ENZH}	Enable-to-Pad, Z to H		1.5		1.7		1.9		2.3	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.7		3.1		3.5		4.1	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.9		3.3		3.7		4.4	ns
PCI Output Module Timing[3]										
t_{DLH}	Data-to-Pad LOW to HIGH		1.8		2.0		2.3		2.7	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.7		2.0		2.2		2.6	ns
t_{ENZL}	Enable-to-Pad, Z to L		0.8		1.0		1.1		1.3	ns
t_{ENZH}	Enable-to-Pad, Z to H		1.2		1.2		1.5		1.8	ns
t_{ENLZ}	Enable-to-Pad, L to Z		1.0		1.1		1.3		1.5	ns
t_{ENHZ}	Enable-to-Pad, H to Z		1.1		1.3		1.5		1.7	ns
TTL Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		2.1		2.5		2.8		3.3	ns
t_{DHL}	Data-to-Pad HIGH to LOW		2.0		2.3		2.6		3.1	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.5		2.9		3.2		3.8	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.0		3.5		3.9		4.6	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.3		2.7		3.1		3.6	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.9		3.3		3.7		4.4	ns

Note:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 10 pF loading.*

208-Pin PQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	I/O	I/O	I/O
4	NC	I/O	I/O
5	I/O	I/O	I/O
6	NC	I/O	I/O
7	I/O	I/O	I/O
8	I/O	I/O	I/O
9	I/O	I/O	I/O
10	I/O	I/O	I/O
11	TMS	TMS	TMS
12	V_{CCI}	V_{CCI}	V_{CCI}
13	I/O	I/O	I/O
14	NC	I/O	I/O
15	I/O	I/O	I/O
16	I/O	I/O	I/O
17	NC	I/O	I/O
18	I/O	I/O	I/O
19	I/O	I/O	I/O
20	NC	I/O	I/O
21	I/O	I/O	I/O
22	I/O	I/O	I/O
23	NC	I/O	I/O
24	I/O	I/O	I/O
25	V_{CCR}	V_{CCR}	V_{CCR}
26	GND	GND	GND
27	V_{CCA}	V_{CCA}	V_{CCA}
28	GND	GND	GND
29	I/O	I/O	I/O
30	I/O	I/O	I/O
31	NC	I/O	I/O
32	I/O	I/O	I/O
33	I/O	I/O	I/O
34	I/O	I/O	I/O
35	NC	I/O	I/O
36	I/O	I/O	I/O

208-Pin PQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
37	I/O	I/O	I/O
38	I/O	I/O	I/O
39	NC	I/O	I/O
40	V_{CCI}	V_{CCI}	V_{CCI}
41	V_{CCA}	V_{CCA}	V_{CCA}
42	I/O	I/O	I/O
43	I/O	I/O	I/O
44	I/O	I/O	I/O
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	NC	I/O	I/O
49	I/O	I/O	I/O
50	NC	I/O	I/O
51	I/O	I/O	I/O
52	GND	GND	GND
53	I/O	I/O	I/O
54	I/O	I/O	I/O
55	I/O	I/O	I/O
56	I/O	I/O	I/O
57	I/O	I/O	I/O
58	I/O	I/O	I/O
59	I/O	I/O	I/O
60	V_{CCI}	V_{CCI}	V_{CCI}
61	NC	I/O	I/O
62	I/O	I/O	I/O
63	I/O	I/O	I/O
64	NC	I/O	I/O
65*	I/O	I/O	NC*
66	I/O	I/O	I/O
67	NC	I/O	I/O
68	I/O	I/O	I/O
69	I/O	I/O	I/O
70	NC	I/O	I/O
71	I/O	I/O	I/O
72	I/O	I/O	I/O

Note: ** Note that Pin 65 in the A54SX32—PQ208 is a no connect (NC).*

208-Pin PQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
73	NC	I/O	I/O
74	I/O	I/O	I/O
75	NC	I/O	I/O
76	PRB, I/O	PRB, I/O	PRB, I/O
77	GND	GND	GND
78	V_{CCA}	V_{CCA}	V_{CCA}
79	GND	GND	GND
80	V_{CCR}	V_{CCR}	V_{CCR}
81	I/O	I/O	I/O
82	HCLK	HCLK	HCLK
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	NC	I/O	I/O
86	I/O	I/O	I/O
87	I/O	I/O	I/O
88	NC	I/O	I/O
89	I/O	I/O	I/O
90	I/O	I/O	I/O
91	NC	I/O	I/O
92	I/O	I/O	I/O
93	I/O	I/O	I/O
94	NC	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	NC	I/O	I/O
98	V_{CCI}	V_{CCI}	V_{CCI}
99	I/O	I/O	I/O
100	I/O	I/O	I/O
101	I/O	I/O	I/O
102	I/O	I/O	I/O
103	TDO, I/O	TDO, I/O	TDO, I/O
104	I/O	I/O	I/O
105	GND	GND	GND
106	NC	I/O	I/O
107	I/O	I/O	I/O
108	NC	I/O	I/O

208-Pin PQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
109	I/O	I/O	I/O
110	I/O	I/O	I/O
111	I/O	I/O	I/O
112	I/O	I/O	I/O
113	I/O	I/O	I/O
114	V_{CCA}	V_{CCA}	V_{CCA}
115	V_{CCI}	V_{CCI}	V_{CCI}
116	NC	I/O	I/O
117	I/O	I/O	I/O
118	I/O	I/O	I/O
119	NC	I/O	I/O
120	I/O	I/O	I/O
121	I/O	I/O	I/O
122	NC	I/O	I/O
123	I/O	I/O	I/O
124	I/O	I/O	I/O
125	NC	I/O	I/O
126	I/O	I/O	I/O
127	I/O	I/O	I/O
128	I/O	I/O	I/O
129	GND	GND	GND
130	V_{CCA}	V_{CCA}	V_{CCA}
131	GND	GND	GND
132	V_{CCR}	V_{CCR}	V_{CCR}
133	I/O	I/O	I/O
134	I/O	I/O	I/O
135	NC	I/O	I/O
136	I/O	I/O	I/O
137	I/O	I/O	I/O
138	NC	I/O	I/O
139	I/O	I/O	I/O
140	I/O	I/O	I/O
141	NC	I/O	I/O
142	I/O	I/O	I/O
143	NC	I/O	I/O
144	I/O	I/O	I/O

Note: ** Note that Pin 65 in the A54SX32—PQ208 is a no connect (NC).*

144-Pin TQFP			
Pin Number	**A54SX08 Function**	**A54SX16P Function**	**A54SX32 Function**
73	GND	GND	GND
74	I/O	I/O	I/O
75	I/O	I/O	I/O
76	I/O	I/O	I/O
77	I/O	I/O	I/O
78	I/O	I/O	I/O
79	V_{CCA}	V_{CCA}	V_{CCA}
80	V_{CCI}	V_{CCI}	V_{CCI}
81	GND	GND	GND
82	I/O	I/O	I/O
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	I/O	I/O	I/O
86	I/O	I/O	I/O
87	I/O	I/O	I/O
88	I/O	I/O	I/O
89	V_{CCA}	V_{CCA}	V_{CCA}
90	V_{CCR}	V_{CCR}	V_{CCR}
91	I/O	I/O	I/O
92	I/O	I/O	I/O
93	I/O	I/O	I/O
94	I/O	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	I/O	I/O	I/O
98	V_{CCA}	V_{CCA}	V_{CCA}
99	GND	GND	GND
100	I/O	I/O	I/O
101	GND	GND	GND
102	V_{CCI}	V_{CCI}	V_{CCI}
103	I/O	I/O	I/O
104	I/O	I/O	I/O
105	I/O	I/O	I/O
106	I/O	I/O	I/O
107	I/O	I/O	I/O
108	I/O	I/O	I/O

144-Pin TQFP			
Pin Number	**A54SX08 Function**	**A54SX16P Function**	**A54SX32 Function**
109	GND	GND	GND
110	I/O	I/O	I/O
111	I/O	I/O	I/O
112	I/O	I/O	I/O
113	I/O	I/O	I/O
114	I/O	I/O	I/O
115	V_{CCI}	V_{CCI}	V_{CCI}
116	I/O	I/O	I/O
117	I/O	I/O	I/O
118	I/O	I/O	I/O
119	I/O	I/O	I/O
120	I/O	I/O	I/O
121	I/O	I/O	I/O
122	I/O	I/O	I/O
123	I/O	I/O	I/O
124	I/O	I/O	I/O
125	CLKA	CLKA	CLKA
126	CLKB	CLKB	CLKB
127	V_{CCR}	V_{CCR}	V_{CCR}
128	GND	GND	GND
129	V_{CCA}	V_{CCA}	V_{CCA}
130	I/O	I/O	I/O
131	PRA, I/O	PRA, I/O	PRA, I/O
132	I/O	I/O	I/O
133	I/O	I/O	I/O
134	I/O	I/O	I/O
135	I/O	I/O	I/O
136	I/O	I/O	I/O
137	I/O	I/O	I/O
138	I/O	I/O	I/O
139	I/O	I/O	I/O
140	V_{CCI}	V_{CCI}	V_{CCI}
141	I/O	I/O	I/O
142	I/O	I/O	I/O
143	I/O	I/O	I/O
144	TCK, I/O	TCK, I/O	TCK, I/O

176-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	NC	I/O	I/O
4	I/O	I/O	I/O
5	I/O	I/O	I/O
6	I/O	I/O	I/O
7	I/O	I/O	I/O
8	I/O	I/O	I/O
9	I/O	I/O	I/O
10	TMS	TMS	TMS
11	V_{CCI}	V_{CCI}	V_{CCI}
12	NC	I/O	I/O
13	I/O	I/O	I/O
14	I/O	I/O	I/O
15	I/O	I/O	I/O
16	I/O	I/O	I/O
17	I/O	I/O	I/O
18	I/O	I/O	I/O
19	I/O	I/O	I/O
20	I/O	I/O	I/O
21	GND	GND	GND
22	V_{CCA}	V_{CCA}	V_{CCA}
23	GND	GND	GND
24	I/O	I/O	I/O
25	I/O	I/O	I/O
26	I/O	I/O	I/O
27	I/O	I/O	I/O
28	I/O	I/O	I/O
29	I/O	I/O	I/O
30	I/O	I/O	I/O
31	I/O	I/O	I/O
32	V_{CCI}	V_{CCI}	V_{CCI}
33	V_{CCA}	V_{CCA}	V_{CCA}
34	I/O	I/O	I/O

176-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
35	I/O	I/O	I/O
36	I/O	I/O	I/O
37	I/O	I/O	I/O
38	I/O	I/O	I/O
39	I/O	I/O	I/O
40	NC	I/O	I/O
41	I/O	I/O	I/O
42	NC	I/O	I/O
43	I/O	I/O	I/O
44	GND	GND	GND
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	I/O	I/O	I/O
49	I/O	I/O	I/O
50	I/O	I/O	I/O
51	I/O	I/O	I/O
52	V_{CCI}	V_{CCI}	V_{CCI}
53	I/O	I/O	I/O
54	NC	I/O	I/O
55	I/O	I/O	I/O
56	I/O	I/O	I/O
57	NC	I/O	I/O
58	I/O	I/O	I/O
59	I/O	I/O	I/O
60	I/O	I/O	I/O
61	I/O	I/O	I/O
62	I/O	I/O	I/O
63	I/O	I/O	I/O
64	PRB, I/O	PRB, I/O	PRB, I/O
65	GND	GND	GND
66	V_{CCA}	V_{CCA}	V_{CCA}
67	V_{CCR}	V_{CCR}	V_{CCR}
68	I/O	I/O	I/O

329-Pin PBGA	
Pin Number	**A54SX32 Function**
D3	I/O
D4	TCK, I/O
D5	I/O
D6	I/O
D7	I/O
D8	I/O
D9	I/O
D10	I/O
D11	V_{CCA}
D12	V_{CCR}
D13	I/O
D14	I/O
D15	I/O
D16	I/O
D17	I/O
D18	I/O
D19	I/O
D20	I/O
D21	I/O
D22	I/O
D23	I/O
E1	V_{CCI}
E2	I/O
E3	I/O
E4	I/O
E20	I/O
E21	I/O
E22	I/O
E23	I/O
F1	I/O
F2	TMS
F3	I/O
F4	I/O
F20	I/O
F21	I/O

329-Pin PBGA	
Pin Number	**A54SX32 Function**
F22	I/O
F23	I/O
G1	I/O
G2	I/O
G3	I/O
G4	I/O
G20	I/O
G21	I/O
G22	I/O
G23	GND
H1	I/O
H2	I/O
H3	I/O
H4	I/O
H20	V_{CCA}
H21	I/O
H22	I/O
H23	I/O
J1	NC
J2	I/O
J3	I/O
J4	I/O
J20	I/O
J21	I/O
J22	I/O
J23	I/O
K1	I/O
K2	I/O
K3	I/O
K4	I/O
K10	GND
K11	GND
K12	GND
K13	GND
K14	GND

329-Pin PBGA	
Pin Number	**A54SX32 Function**
K20	I/O
K21	I/O
K22	I/O
K23	I/O
L1	I/O
L2	I/O
L3	I/O
L4	V_{CCR}
L10	GND
L11	GND
L12	GND
L13	GND
L14	GND
L20	V_{CCR}
L21	I/O
L22	I/O
L23	NC
M1	I/O
M2	I/O
M3	I/O
M4	V_{CCA}
M10	GND
M11	GND
M12	GND
M13	GND
M14	GND
M20	V_{CCA}
M21	I/O
M22	I/O
M23	V_{CCI}
N1	I/O
N2	I/O
N3	I/O
N4	I/O
N10	GND

329-Pin PBGA	
Pin Number	**A54SX32 Function**
N11	GND
N12	GND
N13	GND
N14	GND
N20	NC
N21	I/O
N22	I/O
N23	I/O
P1	I/O
P2	I/O
P3	I/O
P4	I/O
P10	GND
P11	GND
P12	GND
P13	GND
P14	GND
P20	I/O
P21	I/O
P22	I/O
P23	I/O
R1	I/O
R2	I/O
R3	I/O
R4	I/O
R20	I/O
R21	I/O
R22	I/O
R23	I/O
T1	I/O
T2	I/O
T3	I/O
T4	I/O
T20	I/O
T21	I/O

144-Pin FBGA

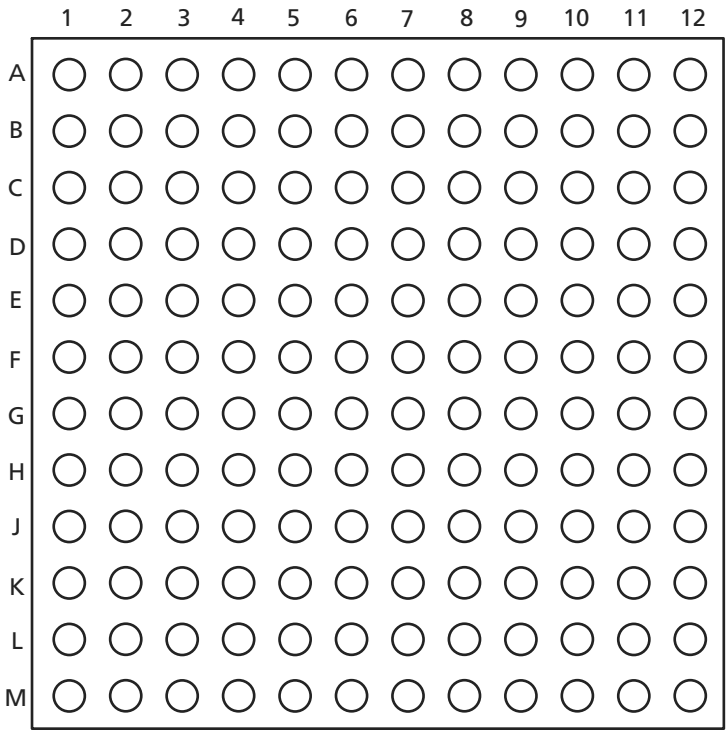


Figure 2-8 • **144-Pin FBGA (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

144-Pin FBGA	
Pin Number	**A54SX08 Function**
A1	I/O
A2	I/O
A3	I/O
A4	I/O
A5	V_{CCA}
A6	GND
A7	CLKA
A8	I/O
A9	I/O
A10	I/O
A11	I/O
A12	I/O
B1	I/O
B2	GND
B3	I/O
B4	I/O
B5	I/O
B6	I/O
B7	CLKB
B8	I/O
B9	I/O
B10	I/O
B11	GND
B12	I/O
C1	I/O
C2	I/O
C3	TCK, I/O
C4	I/O
C5	I/O
C6	PRA, I/O
C7	I/O
C8	I/O
C9	I/O
C10	I/O
C11	I/O
C12	I/O

144-Pin FBGA	
Pin Number	**A54SX08 Function**
D1	I/O
D2	V_{CCI}
D3	TDI, I/O
D4	I/O
D5	I/O
D6	I/O
D7	I/O
D8	I/O
D9	I/O
D10	I/O
D11	I/O
D12	I/O
E1	I/O
E2	I/O
E3	I/O
E4	I/O
E5	TMS
E6	V_{CCI}
E7	V_{CCI}
E8	V_{CCI}
E9	V_{CCA}
E10	I/O
E11	GND
E12	I/O
F1	I/O
F2	I/O
F3	V_{CCR}
F4	I/O
F5	GND
F6	GND
F7	GND
F8	V_{CCI}
F9	I/O
F10	GND
F11	I/O
F12	I/O

144-Pin FBGA	
Pin Number	**A54SX08 Function**
G1	I/O
G2	GND
G3	I/O
G4	I/O
G5	GND
G6	GND
G7	GND
G8	V_{CCI}
G9	I/O
G10	I/O
G11	I/O
G12	I/O
H1	I/O
H2	I/O
H3	I/O
H4	I/O
H5	V_{CCA}
H6	V_{CCA}
H7	V_{CCI}
H8	V_{CCI}
H9	V_{CCA}
H10	I/O
H11	I/O
H12	V_{CCR}
J1	I/O
J2	I/O
J3	I/O
J4	I/O
J5	I/O
J6	PRB, I/O
J7	I/O
J8	I/O
J9	I/O
J10	I/O
J11	I/O
J12	V_{CCA}

144-Pin FBGA	
Pin Number	**A54SX08 Function**
K1	I/O
K2	I/O
K3	I/O
K4	I/O
K5	I/O
K6	I/O
K7	GND
K8	I/O
K9	I/O
K10	GND
K11	I/O
K12	I/O
L1	GND
L2	I/O
L3	I/O
L4	I/O
L5	I/O
L6	I/O
L7	HCLK
L8	I/O
L9	I/O
L10	I/O
L11	I/O
L12	I/O
M1	I/O
M2	I/O
M3	I/O
M4	I/O
M5	I/O
M6	I/O
M7	V_{CCA}
M8	I/O
M9	I/O
M10	I/O
M11	TDO, I/O
M12	I/O

www.actel.com

Actel Corporation

2061 Stierlin Court
Mountain View, CA
94043-4655 USA
Phone 650.318.4200
Fax 650.318.4600

Actel Europe Ltd.

Dunlop House, Riverside Way
Camberley, Surrey GU15 3YL
United Kingdom
Phone +44 (0) 1276 401 450
Fax +44 (0) 1276 401 490

Actel Japan
www.jp.actel.com

EXOS Ebisu Bldg. 4F
1-24-14 Ebisu Shibuya-ku
Tokyo 150 Japan
Phone +81.03.3445.7671
Fax +81.03.3445.7668

Actel Hong Kong
www.actel.com.cn

Suite 2114, Two Pacific Place
88 Queensway, Admiralty
Hong Kong
Phone +852 2185 6460
Fax +852 2185 6488

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