Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	768
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	69
Number of Gates	12000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	84-LCC (J-Lead)
Supplier Device Package	84-PLCC (29.31x29.31)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx08-1plg84

SX Family FPGAs

General Description

The Actel SX family of FPGAs features a sea-of-modules architecture that delivers device performance and integration levels not currently achieved by any other FPGA architecture. SX devices greatly simplify design time, enable dramatic reductions in design costs and power consumption, and further decrease time to market for performance-intensive applications.

The Actel SX architecture features two types of logic modules, the combinatorial cell (C-cell) and the register cell (R-cell), each optimized for fast and efficient mapping of synthesized logic functions. The routing and interconnect resources are in the metal layers above the logic modules, providing optimal use of silicon. This enables the entire floor of the device to be spanned with an uninterrupted grid of fine-grained, synthesis-friendly logic modules (or "sea-of-modules"), which reduces the distance signals have to travel between logic modules. To minimize signal propagation delay, SX devices employ both local and general routing resources. The high-speed local routing resources (DirectConnect and FastConnect) enable very fast local signal propagation that is optimal for fast counters, state machines, and datapath logic. The general system of segmented routing tracks allows any logic module in the array to be connected to any other logic or I/O module. Within this system, propagation delay is minimized by limiting the number of antifuse interconnect elements to five (90 percent of connections typically use only three antifuses). The unique local and general routing structure featured in SX devices gives fast and predictable performance, allows 100 percent pin-locking with full logic utilization, enables concurrent PCB development, reduces design time, and allows designers to achieve performance goals with minimum effort.

Further complementing SX's flexible routing structure is a hardwired, constantly loaded clock network that has been tuned to provide fast clock propagation with minimal clock skew. Additionally, the high performance of the internal logic has eliminated the need to embed latches or flip-flops in the I/O cells to achieve fast clock-to-out or fast input setup times. SX devices have easy to use I/O cells that do not require HDL instantiation, facilitating design reuse and reducing design and verification time.

SX Family Architecture

The SX family architecture was designed to satisfy next-generation performance and integration requirements for production-volume designs in a broad range of applications.

Programmable Interconnect Element

The SX family provides efficient use of silicon by locating the routing interconnect resources between the Metal 2 (M2) and Metal 3 (M3) layers (Figure 1-1 on page 1-2). This completely eliminates the channels of routing and interconnect resources between logic modules (as implemented on SRAM FPGAs and previous generations of antifuse FPGAs), and enables the entire floor of the device to be spanned with an uninterrupted grid of logic modules.

Interconnection between these logic modules is achieved using The Actel patented metal-to-metal programmable antifuse interconnect elements, which are embedded between the M2 and M3 layers. The antifuses are normally open circuit and, when programmed, form a permanent low-impedance connection.

The extremely small size of these interconnect elements gives the SX family abundant routing resources and provides excellent protection against design pirating. Reverse engineering is virtually impossible because it is extremely difficult to distinguish between programmed and unprogrammed antifuses, and there is no configuration bitstream to intercept.

Additionally, the interconnect elements (i.e., the antifuses and metal tracks) have lower capacitance and lower resistance than any other device of similar capacity, leading to the fastest signal propagation in the industry.

Logic Module Design

The SX family architecture is described as a "sea-of-modules" architecture because the entire floor of the device is covered with a grid of logic modules with virtually no chip area lost to interconnect elements or routing. The Actel SX family provides two types of logic modules, the register cell (R-cell) and the combinatorial cell (C-cell).

The R-cell contains a flip-flop featuring asynchronous clear, asynchronous preset, and clock enable (using the S0 and S1 lines) control signals (Figure 1-2). The R-cell registers feature programmable clock polarity selectable on a register-by-register basis. This provides additional flexibility while allowing mapping of synthesized functions into the SX FPGA. The clock source for the R-cell can be chosen from either the hardwired clock or the routed clock.

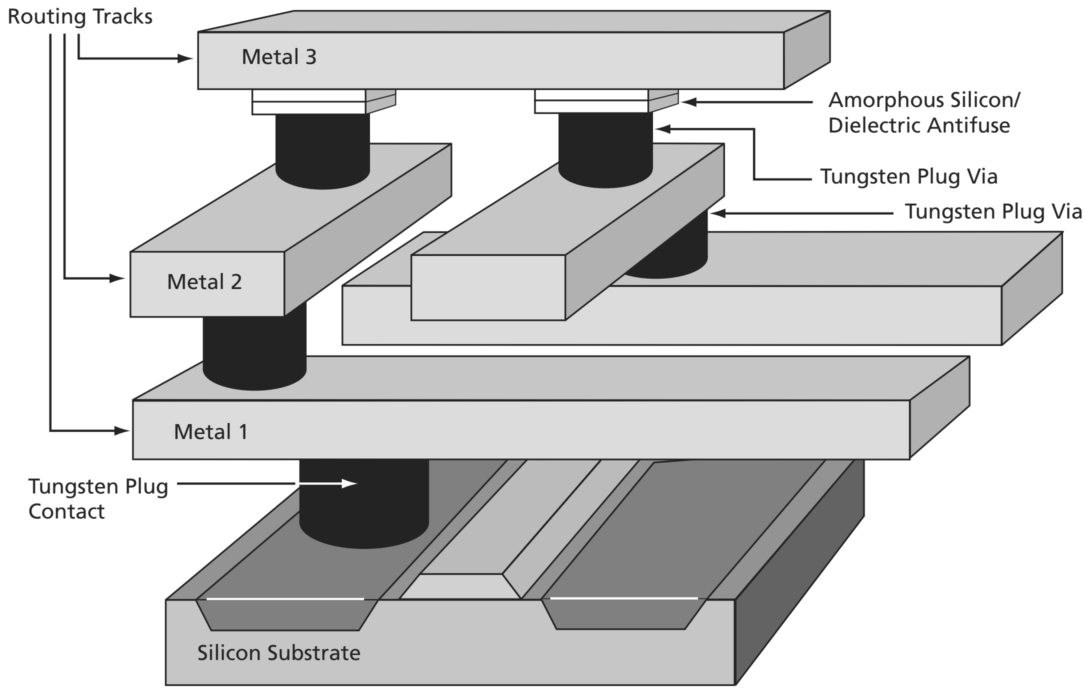


Figure 1-1 • **SX Family Interconnect Elements**

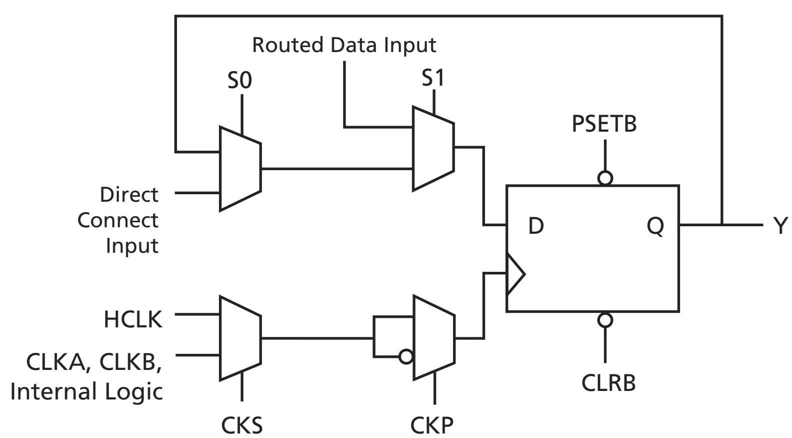


Figure 1-2 • **R-Cell**

The C-cell implements a range of combinatorial functions up to 5-inputs (Figure 1-3 on page 1-3). Inclusion of the DB input and its associated inverter function dramatically increases the number of combinatorial functions that can be implemented in a single module from 800 options in previous architectures to more than 4,000 in the SX architecture. An example of the improved flexibility enabled by the inversion capability is the ability to integrate a 3-input exclusive-OR function into a single C-cell. This facilitates construction of 9-bit parity-tree functions with 2 ns propagation delays. At the same time, the C-cell structure is extremely synthesis friendly, simplifying the overall design and reducing synthesis time.

Chip Architecture

The SX family chip architecture provides a unique approach to module organization and chip routing that delivers the best register/logic mix for a wide variety of new and emerging applications.

Module Organization

Actel has arranged all C-cell and R-cell logic modules into horizontal banks called *clusters*. There are two types of *clusters*: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells.

To increase design efficiency and device performance, Actel has further organized these modules into *SuperClusters* (Figure 1-4). SuperCluster 1 is a two-wide grouping of Type 1 clusters. SuperCluster 2 is a two-wide group containing one Type 1 cluster and one Type 2 cluster. SX devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flip-flops.

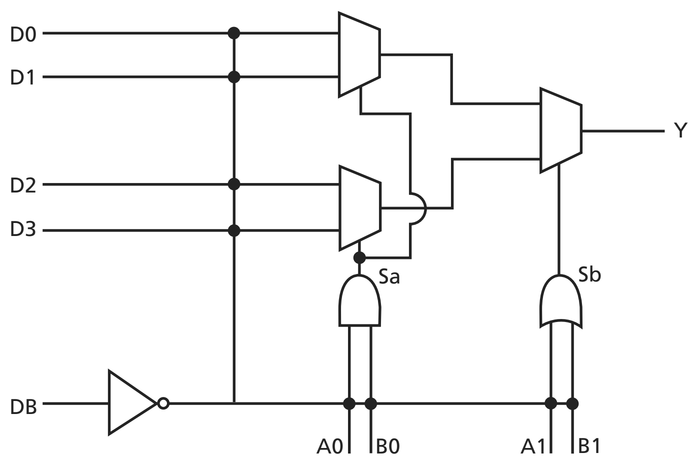


Figure 1-3 • **C-Cell**

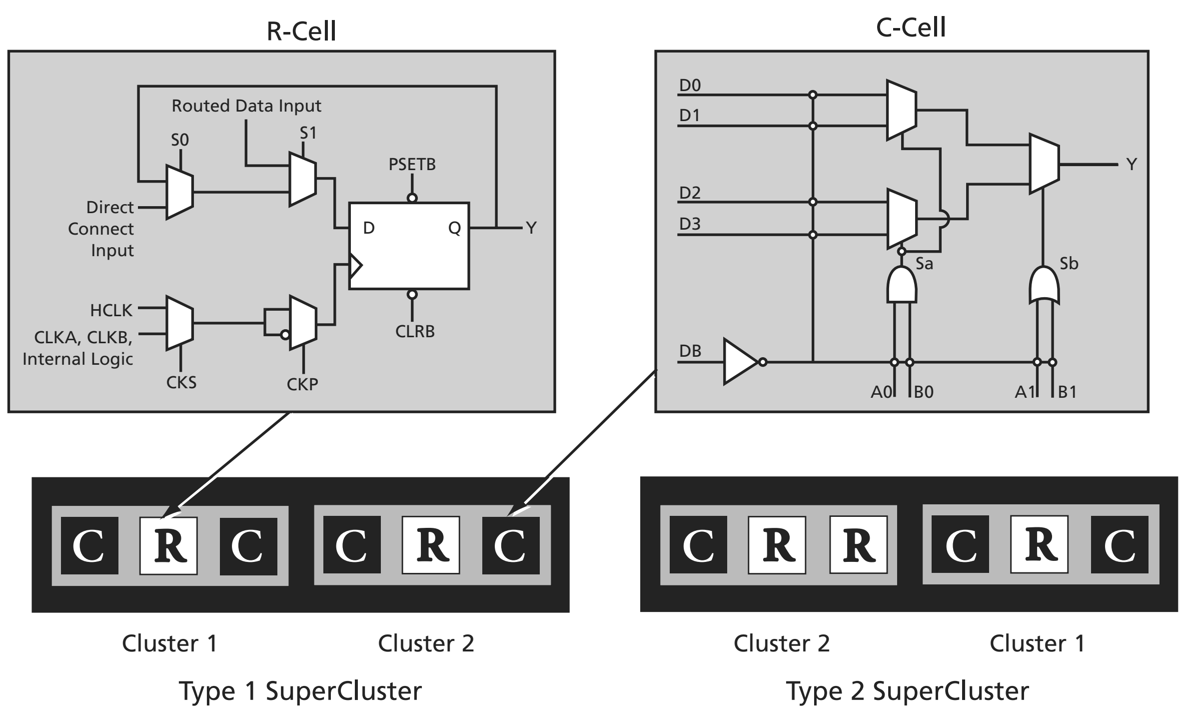


Figure 1-4 • **Cluster Organization**

Routing Resources

Clusters and SuperClusters can be connected through the use of two innovative local routing resources called *FastConnect* and *DirectConnect,* which enable extremely fast and predictable interconnection of modules within clusters and SuperClusters (Figure 1-5 and Figure 1-6). This routing architecture also dramatically reduces the number of antifuses required to complete a circuit, ensuring the highest possible performance.

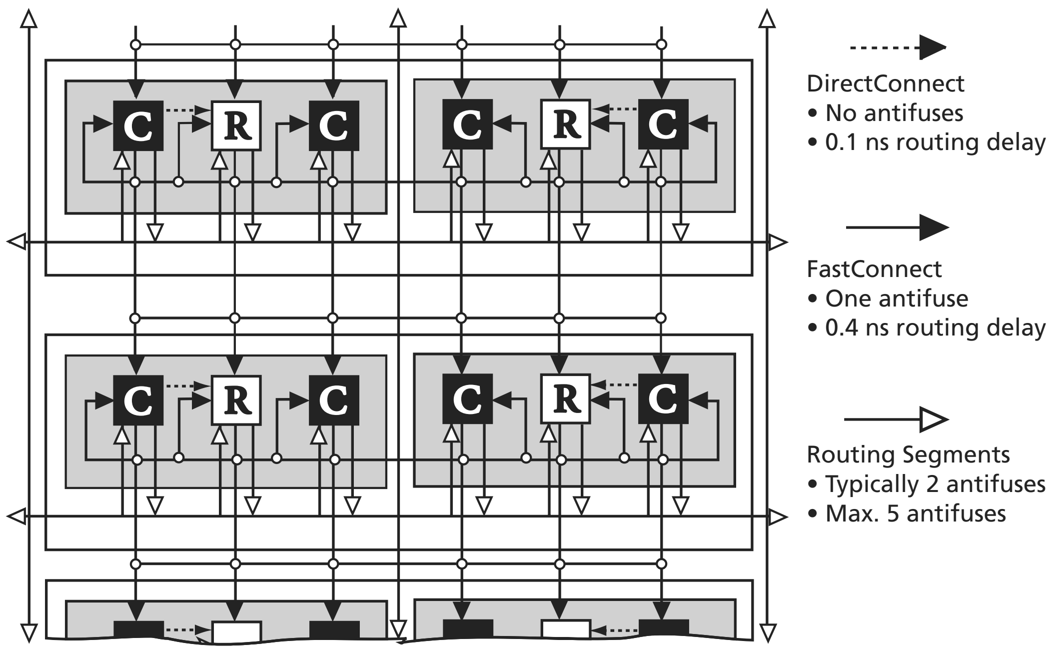


Figure 1-5 • **DirectConnect and FastConnect for Type 1 SuperClusters**

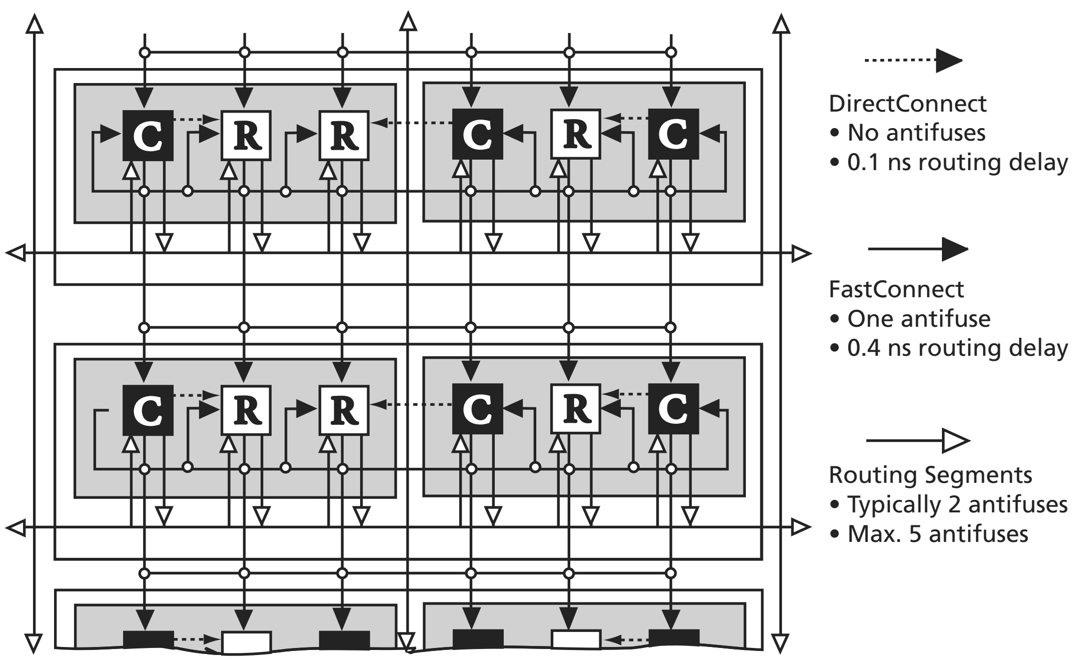


Figure 1-6 • **DirectConnect and FastConnect for Type 2 SuperClusters**

Boundary Scan Testing (BST)

All SX devices are IEEE 1149.1 compliant. SX devices offer superior diagnostic and testing capabilities by providing Boundary Scan Testing (BST) and probing capabilities. These functions are controlled through the special test pins in conjunction with the program fuse. The functionality of each pin is described in Table 1-2. In the dedicated test mode, TCK, TDI, and TDO are dedicated pins and cannot be used as regular I/Os. In flexible mode, TMS should be set HIGH through a pull-up resistor of 10 kΩ. TMS can be pulled LOW to initiate the test sequence.

The program fuse determines whether the device is in dedicated or flexible mode. The default (fuse not blown) is flexible mode.

Table 1-2 • **Boundary Scan Pin Functionality**

Program Fuse Blown (Dedicated Test Mode)	Program Fuse Not Blown (Flexible Mode)
TCK, TDI, TDO are dedicated BST pins.	TCK, TDI, TDO are flexible and may be used as I/Os.
No need for pull-up resistor for TMS	Use a pull-up resistor of 10 kΩ on TMS.

Dedicated Test Mode

In Dedicated mode, all JTAG pins are reserved for BST; designers cannot use them as regular I/Os. An internal pull-up resistor is automatically enabled on both TMS and TDI pins, and the TMS pin will function as defined in the IEEE 1149.1 (JTAG) specification.

To select Dedicated mode, users need to reserve the JTAG pins in Actel's Designer software by checking the "Reserve JTAG" box in "Device Selection Wizard" (Figure 1-7). JTAG pins comply with LVTTL/TTL I/O specification regardless of whether they are used as a user I/O or a JTAG I/O. Refer to the Table 1-5 on page 1-8 for detailed specifications.

Figure 1-7 • **Device Selection Wizard**

Development Tool Support

The SX family of FPGAs is fully supported by both the Actel Libero® Integrated Design Environment (IDE) and Designer FPGA Development software. Actel Libero IDE is a design management environment, seamlessly integrating design tools while guiding the user through the design flow, managing all design and log files, and passing necessary design data among tools. Libero IDE allows users to integrate both schematic and HDL synthesis into a single flow and verify the entire design in a single environment. Libero IDE includes Synplify® for Actel from Synplicity®, ViewDraw® for Actel from Mentor Graphics®, Model*Sim*® HDL Simulator from Mentor Graphics, WaveFormer Lite™ from SynaptiCAD™, and Designer software from Actel. Refer to the Libero IDE flow diagram (located on the Actel website) for more information.

Actel Designer software is a place-and-route tool and provides a comprehensive suite of backend support tools for FPGA development. The Designer software includes timing-driven place-and-route, and a world-class integrated static timing analyzer and constraints editor. With the Designer software, a user can select and lock package pins while only minimally impacting the results of place-and-route. Additionally, the back-annotation flow is compatible with all the major simulators, and the simulation results can be cross-probed with Silicon Explorer II, Actel integrated verification and logic analysis tool. Another tool included in the Designer software is the SmartGen core generator, which easily creates popular and commonly used logic functions for implementation into your schematic or HDL design. Actel Designer software is compatible with the most popular FPGA design entry and verification tools from companies such as Mentor Graphics, Synplicity, Synopsys®, and Cadence® Design Systems. The Designer software is available for both the Windows® and UNIX® operating systems.

Probe Circuit Control Pins

The Silicon Explorer II tool uses the boundary scan ports (TDI, TCK, TMS, and TDO) to select the desired nets for verification. The selected internal nets are assigned to the PRA/PRB pins for observation. Figure 1-8 on page 1-7 illustrates the interconnection between Silicon Explorer II and the FPGA to perform in-circuit verification.

Design Considerations

The TDI, TCK, TDO, PRA, and PRB pins should not be used as input or bidirectional ports. Because these pins are active during probing, critical signals input through these pins are not available while probing. In addition, the Security Fuse should not be programmed because doing so disables the Probe Circuitry.

Table 1-4 • **Recommended Operating Conditions**

Parameter	Commercial	Industrial	Military	Units
Temperature Range*	0 to + 70	–40 to + 85	–55 to +125	°C
3.3 V Power Supply Tolerance	±10	±10	±10	$\%V_{CC}$
5.0 V Power Supply Tolerance	±5	±10	±10	$\%V_{CC}$

Note: **Ambient temperature (T_A) is used for commercial and industrial; case temperature (T_C) is used for military.*

Table 1-5 • **Electrical Specifications**

		Commercial		Industrial		
Symbol	**Parameter**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
V_{OH}	(I_{OH} = –20 µA) (CMOS) (I_{OH} = –8 mA) (TTL) (I_{OH} = –6 mA) (TTL)	(V_{CCI} – 0.1) 2.4	V_{CCI} V_{CCI}	(V_{CCI} – 0.1) 2.4	V_{CCI} V_{CCI}	V
V_{OL}	(I_{OL}= 20 µA) (CMOS) (I_{OL} = 12 mA) (TTL) (I_{OL} = 8 mA) (TTL)		0.10 0.50		 0.50	V
V_{IL}			0.8		0.8	V
V_{IH}		2.0		2.0		V
t_R, t_F	Input Transition Time t_R, t_F		50		50	ns
C_{IO}	C_{IO} I/O Capacitance		10		10	pF
I_{CC}	Standby Current, I_{CC}		4.0		4.0	mA
$I_{CC(D)}$	$I_{CC(D)}$ $I_{Dynamic}$ V_{CC} Supply Current	See "Evaluating Power in SX Devices" on page 1-16.				

PCI Compliance for the SX Family

The SX family supports 3.3 V and 5.0 V PCI and is compliant with the PCI Local Bus Specification Rev. 2.1.

Table 1-6 • **A54SX16P DC Specifications (5.0 V PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
V_{CCA}	Supply Voltage for Array		3.0	3.6	V
V_{CCR}	Supply Voltage required for Internal Biasing		4.75	5.25	V
V_{CCI}	Supply Voltage for I/Os		4.75	5.25	V
V_{IH}	Input High Voltage[1]		2.0	V_{CC} + 0.5	V
V_{IL}	Input Low Voltage[1]		–0.5	0.8	V
I_{IH}	Input High Leakage Current	V_{IN} = 2.7		70	µA
I_{IL}	Input Low Leakage Current	V_{IN} = 0.5		–70	µA
V_{OH}	Output High Voltage	I_{OUT} = –2 mA	2.4		V
V_{OL}	Output Low Voltage[2]	I_{OUT} = 3 mA, 6 mA		0.55	V
C_{IN}	Input Pin Capacitance[3]			10	pF
C_{CLK}	CLK Pin Capacitance		5	12	pF
C_{IDSEL}	IDSEL Pin Capacitance[4]			8	pF

Notes:

1. *Input leakage currents include hi-Z output leakage for all bidirectional buffers with tristate outputs.*
2. *Signals without pull-up resistors must have 3 mA low output current. Signals requiring pull-up must have 6 mA; the latter include, FRAME#, IRDY#, TRDY#, DEVSEL#, STOP#, SERR#, PERR#, LOCK#, and, when used, AD[63::32], C/BE[7::4]#, PAR64, REQ64#, and ACK64#.*
3. *Absolute maximum pin capacitance for a PCI input is 10 pF (except for CLK).*
4. *Lower capacitance on this input-only pin allows for non-resistive coupling to AD[xx].*

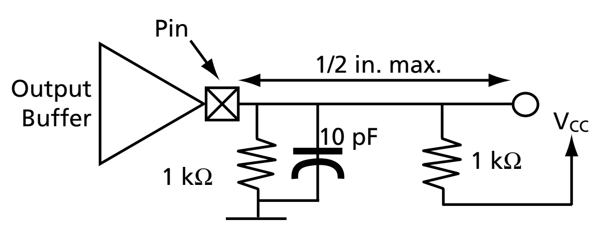
A54SX16P AC Specifications for (PCI Operation)

Table 1-7 • **A54SX16P AC Specifications for (PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \le 1.4$ [1]	–44		mA
		$1.4 \le V_{OUT} < 2.4$ [1, 2]	$-44 + (V_{OUT} - 1.4)/0.024$		mA
		$3.1 < V_{OUT} < V_{CC}$ [1, 3]		EQ 1-1 on page 1-11	
	(Test Point)	$V_{OUT} = 3.1$ [3]		–142	mA
$I_{OL(AC)}$	Switching Current High	$V_{OUT} \ge 2.2$ [1]	95		mA
		$2.2 > V_{OUT} > 0.55$ [1]	$V_{OUT}/0.023$		
		$0.71 > V_{OUT} > 0$ [1, 3]		EQ 1-2 on page 1-11	mA
	(Test Point)	$V_{OUT} = 0.71$ [3]		206	mA
I_{CL}	Low Clamp Current	$-5 < V_{IN} \le -1$	$-25 + (V_{IN} + 1)/0.015$		mA
$slew_R$	Output Rise Slew Rate	0.4 V to 2.4 V load[4]	1	5	V/ns
$slew_F$	Output Fall Slew Rate	2.4 V to 0.4 V load[4]	1	5	V/ns

Notes:

1. *Refer to the V/I curves in Figure 1-9 on page 1-11. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half-size output drivers may be used on these signals. This specification does not apply to CLK and RST#, which are system outputs. "Switching Current High" specifications are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD#, which are open drain outputs.*
2. *Note that this segment of the minimum current curve is drawn from the AC drive point directly to the DC drive point rather than toward the voltage rail (as is done in the pull-down curve). This difference is intended to allow for an optional N-channel pull-up.*
3. *Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (A and B) are provided with the respective diagrams in Figure 1-9 on page 1-11. The equation defined maxima should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.*
4. *This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per revision 2.0 of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is now required (the maximum is no longer simply a guideline). Since adherence to the maximum slew rate was not required prior to revision 2.1 of the specification, there may be components in the market for some time that have faster edge rates; therefore, motherboard designers must bear in mind that rise and fall times faster than this specification could occur, and should ensure that signal integrity modeling accounts for this. Rise slew rate does not apply to open drain outputs.*



Power-Up Sequencing

Table 1-10 • **Power-Up Sequencing**

V_{CCA}	V_{CCR}	V_{CCI}	Power-Up Sequence	Comments
A54SX08, A54SX16, A54SX32				
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	Possible damage to device
A54SX16P				
3.3 V	3.3 V	3.3 V	3.3 V Only	No possible damage to device
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	Possible damage to device
3.3 V	5.0 V	5.0 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device

Note: *No inputs should be driven (high or low) before completion of power-up.*

Power-Down Sequencing

Table 1-11 • **Power-Down Sequencing**

V_{CCA}	V_{CCR}	V_{CCI}	Power-Down Sequence	Comments
A54SX08, A54SX16, A54SX32				
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	Possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device
A54SX16P				
3.3 V	3.3 V	3.3 V	3.3 V Only	No possible damage to device
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	Possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device
3.3 V	5.0 V	5.0 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device

Note: *No inputs should be driven (high or low) after the beginning of the power-down sequence.*

Step 1: Define Terms Used in Formula

	V_{CCA}	3.3
Module		
Number of logic modules switching at f_m (Used 50%)	m	264
Average logic modules switching rate f_m (MHz) (Guidelines: f/10)	f_m	20
Module capacitance C_{EQM} (pF)	C_{EQM}	4.0
Input Buffer		
Number of input buffers switching at f_n	n	1
Average input switching rate f_n (MHz) (Guidelines: f/5)	f_n	40
Input buffer capacitance C_{EQI} (pF)	C_{EQI}	3.4
Output Buffer		
Number of output buffers switching at f_p	p	1
Average output buffers switching rate f_p(MHz) (Guidelines: f/10)	f_p	20
Output buffers buffer capacitance C_{EQO} (pF)	C_{EQO}	4.7
Output Load capacitance C_L (pF)	C_L	35
RCLKA		
Number of Clock loads q_1	q_1	528
Capacitance of routed array clock (pF)	C_{EQCR}	1.6
Average clock rate (MHz)	f_{q1}	200
Fixed capacitance (pF)	r_1	138
RCLKB		
Number of Clock loads q_2	q_2	0
Capacitance of routed array clock (pF)	C_{EQCR}	1.6
Average clock rate (MHz)	f_{q2}	0
Fixed capacitance (pF)	r_2	138
HCLK		
Number of Clock loads	s_1	0
Variable capacitance of dedicated array clock (pF)	C_{EQHV}	0.61 5
Fixed capacitance of dedicated array clock (pF)	C_{EQHF}	96
Average clock rate (MHz)	f_{s1}	0

Step 2: Calculate Dynamic Power Consumption

$V_{CCA} \times V_{CCA}$	10.89
$m \times f_m \times C_{EQM}$	0.02112
$n \times f_n \times C_{EQI}$	0.000136
$p \times f_p \times (C_{EQO}+C_L)$	0.000794
$0.5\,(q_1 \times C_{EQCR} \times f_{q1}) + (r_1 \times f_{q1})$	0.11208
$0.5(q_2 \times C_{EQCR} \times f_{q2}) + (r_2 \times f_{q2})$	0
$0.5\,(s_1 \times C_{EQHV} \times f_{s1}) + (C_{EQHF} \times f_{s1})$	0

$P_{AC} = 1.461$ W

Step 3: Calculate DC Power Dissipation

DC Power Dissipation

$$P_{DC} = (I_{standby}) \times V_{CCA} + (I_{standby}) \times V_{CCR} + (I_{standby}) \times V_{CCI} + X \times V_{OL} \times I_{OL} + Y(V_{CCI} - V_{OH}) \times V_{OH}$$

EQ 1-12

For a rough estimate of DC Power Dissipation, only use $P_{DC} = (I_{standby}) \times V_{CCA}$. The rest of the formula provides a very small number that can be considered negligible.

$P_{DC} = (I_{standby}) \times V_{CCA}$

$P_{DC} = .55\text{ mA} \times 3.3\text{ V}$

$P_{DC} = 0.001815$ W

Step 4: Calculate Total Power Consumption

$P_{Total} = P_{AC} + P_{DC}$

$P_{Total} = 1.461 + 0.001815$

$P_{Total} = 1.4628$ W

Step 5: Compare Estimated Power Consumption against Characterized Power Consumption

The estimated total power consumption for this design is 1.46 W. The characterized power consumption for this design at 200 MHz is 1.0164 W.

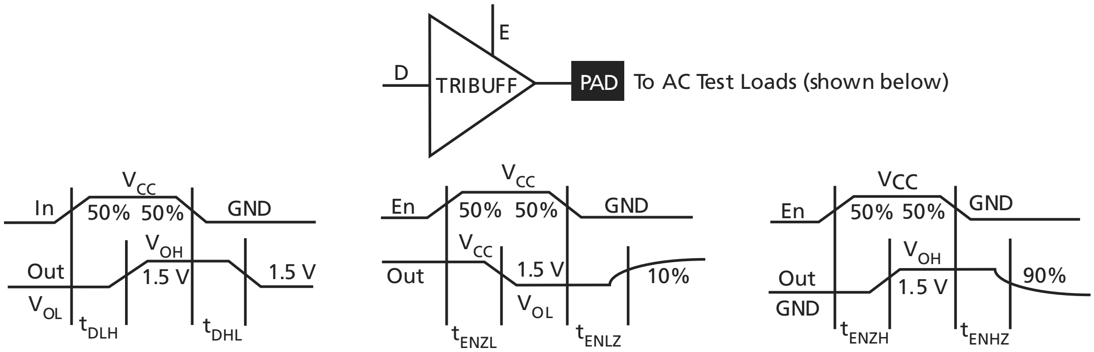


Figure 1-13 • **Output Buffer Delays**

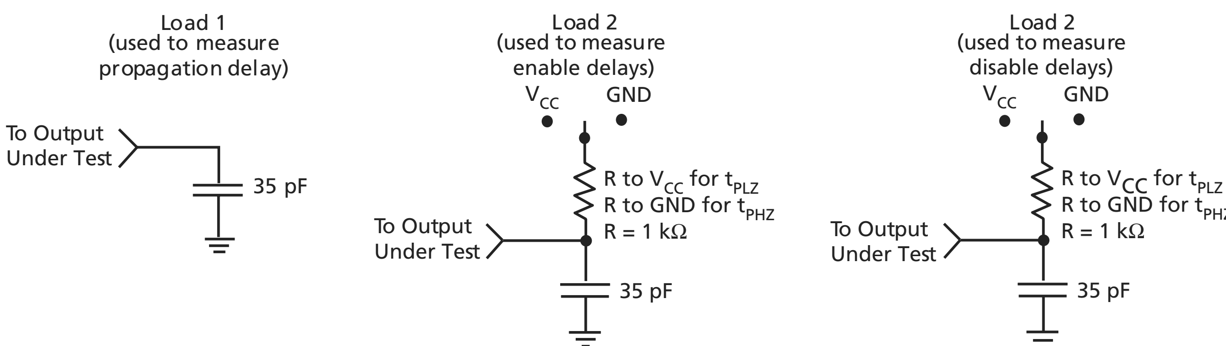


Figure 1-14 • **AC Test Loads**

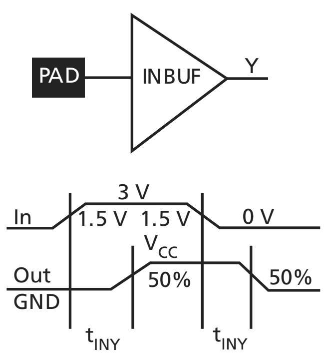


Figure 1-15 • **Input Buffer Delays**

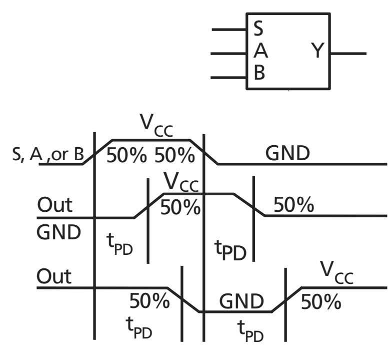


Figure 1-16 • **C-Cell Delays**

Register Cell Timing Characteristics

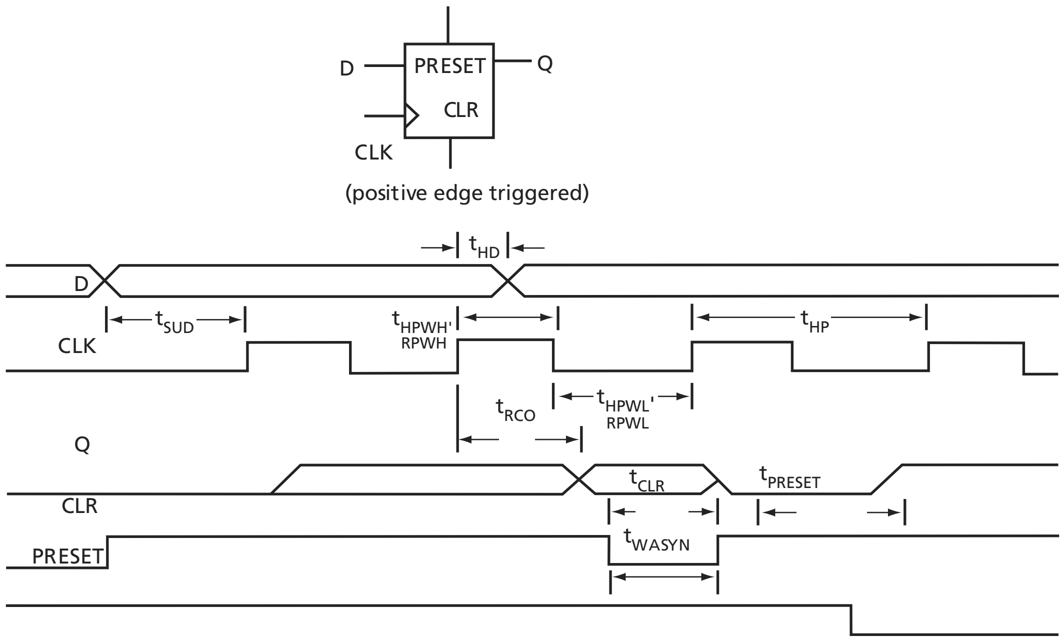


Figure 1-17 • **Flip-Flops**

Timing Characteristics

Timing characteristics for SX devices fall into three categories: family-dependent, device-dependent, and design-dependent. The input and output buffer characteristics are common to all SX family members. Internal routing delays are device-dependent. Design dependency means actual delays are not determined until after placement and routing of the user's design is complete. Delay values may then be determined by using the DirectTime Analyzer utility or performing simulation with post-layout delays.

Critical Nets and Typical Nets

Propagation delays are expressed only for typical nets, which are used for initial design performance evaluation. Critical net delays can then be applied to the most time-critical paths. Critical nets are determined by net property assignment prior to placement and routing. Up to 6% of the nets in a design may be designated as critical, while 90% of the nets in a design are typical.

Long Tracks

Some nets in the design use long tracks. Long tracks are special routing resources that span multiple rows, columns, or modules. Long tracks employ three and sometimes five antifuse connections. This increases capacitance and resistance, resulting in longer net delays for macros connected to long tracks. Typically up to 6 percent of nets in a fully utilized device require long tracks. Long tracks contribute approximately 4 ns to 8.4 ns delay. This additional delay is represented statistically in higher fanout (FO = 24) routing delays in the datasheet specifications section.

Timing Derating

SX devices are manufactured in a CMOS process. Therefore, device performance varies according to temperature, voltage, and process variations. Minimum timing parameters reflect maximum operating voltage, minimum operating temperature, and best-case processing. Maximum timing parameters reflect minimum operating voltage, maximum operating temperature, and worst-case processing.

A54SX08 Timing Characteristics

Table 1-17 • **A54SX08 Timing Characteristics (Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)**

		'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		
Parameter	**Description**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
C-Cell Propagation Delays[1]										
t_{PD}	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted Routing Delays[2]										
t_{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t_{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{RD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{RD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timing										
t_{RCO}	Sequential Clock-to-Q		0.8		1.1		1.2		1.4	ns
t_{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t_{PRESET}	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t_{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t_{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t_{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t_{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Input Module Predicted Routing Delays[2]										
t_{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{IRD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{IRD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Note:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*

208-Pin PQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
145	V_{CCA}	V_{CCA}	V_{CCA}
146	GND	GND	GND
147	I/O	I/O	I/O
148	V_{CCI}	V_{CCI}	V_{CCI}
149	I/O	I/O	I/O
150	I/O	I/O	I/O
151	I/O	I/O	I/O
152	I/O	I/O	I/O
153	I/O	I/O	I/O
154	I/O	I/O	I/O
155	NC	I/O	I/O
156	NC	I/O	I/O
157	GND	GND	GND
158	I/O	I/O	I/O
159	I/O	I/O	I/O
160	I/O	I/O	I/O
161	I/O	I/O	I/O
162	I/O	I/O	I/O
163	I/O	I/O	I/O
164	V_{CCI}	V_{CCI}	V_{CCI}
165	I/O	I/O	I/O
166	I/O	I/O	I/O
167	NC	I/O	I/O
168	I/O	I/O	I/O
169	I/O	I/O	I/O
170	NC	I/O	I/O
171	I/O	I/O	I/O
172	I/O	I/O	I/O
173	NC	I/O	I/O
174	I/O	I/O	I/O
175	I/O	I/O	I/O
176	NC	I/O	I/O
177	I/O	I/O	I/O
178	I/O	I/O	I/O
179	I/O	I/O	I/O
180	CLKA	CLKA	CLKA

208-Pin PQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
181	CLKB	CLKB	CLKB
182	V_{CCR}	V_{CCR}	V_{CCR}
183	GND	GND	GND
184	V_{CCA}	V_{CCA}	V_{CCA}
185	GND	GND	GND
186	PRA, I/O	PRA, I/O	PRA, I/O
187	I/O	I/O	I/O
188	I/O	I/O	I/O
189	NC	I/O	I/O
190	I/O	I/O	I/O
191	I/O	I/O	I/O
192	NC	I/O	I/O
193	I/O	I/O	I/O
194	I/O	I/O	I/O
195	NC	I/O	I/O
196	I/O	I/O	I/O
197	I/O	I/O	I/O
198	NC	I/O	I/O
199	I/O	I/O	I/O
200	I/O	I/O	I/O
201	V_{CCI}	V_{CCI}	V_{CCI}
202	NC	I/O	I/O
203	NC	I/O	I/O
204	I/O	I/O	I/O
205	NC	I/O	I/O
206	I/O	I/O	I/O
207	I/O	I/O	I/O
208	TCK, I/O	TCK, I/O	TCK, I/O

Note: ** Note that Pin 65 in the A54SX32—PQ208 is a no connect (NC).*

144-Pin TQFP

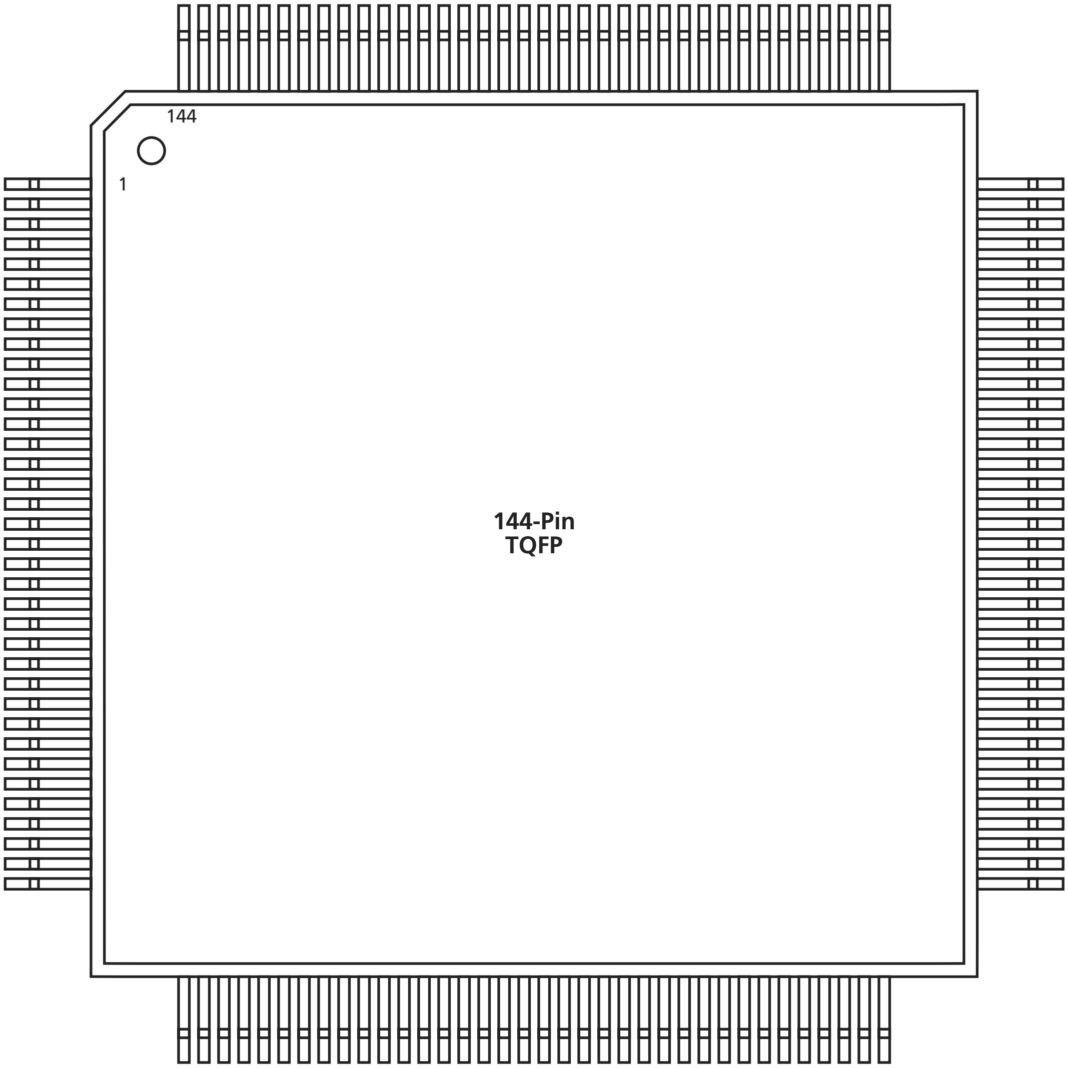


Figure 2-3 • **144-Pin TQFP (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

176-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
69	HCLK	HCLK	HCLK
70	I/O	I/O	I/O
71	I/O	I/O	I/O
72	I/O	I/O	I/O
73	I/O	I/O	I/O
74	I/O	I/O	I/O
75	I/O	I/O	I/O
76	I/O	I/O	I/O
77	I/O	I/O	I/O
78	I/O	I/O	I/O
79	NC	I/O	I/O
80	I/O	I/O	I/O
81	NC	I/O	I/O
82	V_{CCI}	V_{CCI}	V_{CCI}
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	I/O	I/O	I/O
86	I/O	I/O	I/O
87	TDO, I/O	TDO, I/O	TDO, I/O
88	I/O	I/O	I/O
89	GND	GND	GND
90	NC	I/O	I/O
91	NC	I/O	I/O
92	I/O	I/O	I/O
93	I/O	I/O	I/O
94	I/O	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	I/O	I/O	I/O
98	V_{CCA}	V_{CCA}	V_{CCA}
99	V_{CCI}	V_{CCI}	V_{CCI}
100	I/O	I/O	I/O
101	I/O	I/O	I/O
102	I/O	I/O	I/O

176-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
103	I/O	I/O	I/O
104	I/O	I/O	I/O
105	I/O	I/O	I/O
106	I/O	I/O	I/O
107	I/O	I/O	I/O
108	GND	GND	GND
109	V_{CCA}	V_{CCA}	V_{CCA}
110	GND	GND	GND
111	I/O	I/O	I/O
112	I/O	I/O	I/O
113	I/O	I/O	I/O
114	I/O	I/O	I/O
115	I/O	I/O	I/O
116	I/O	I/O	I/O
117	I/O	I/O	I/O
118	NC	I/O	I/O
119	I/O	I/O	I/O
120	NC	I/O	I/O
121	NC	I/O	I/O
122	V_{CCA}	V_{CCA}	V_{CCA}
123	GND	GND	GND
124	V_{CCI}	V_{CCI}	V_{CCI}
125	I/O	I/O	I/O
126	I/O	I/O	I/O
127	I/O	I/O	I/O
128	I/O	I/O	I/O
129	I/O	I/O	I/O
130	I/O	I/O	I/O
131	NC	I/O	I/O
132	NC	I/O	I/O
133	GND	GND	GND
134	I/O	I/O	I/O
135	I/O	I/O	I/O
136	I/O	I/O	I/O

100-Pin VQFP

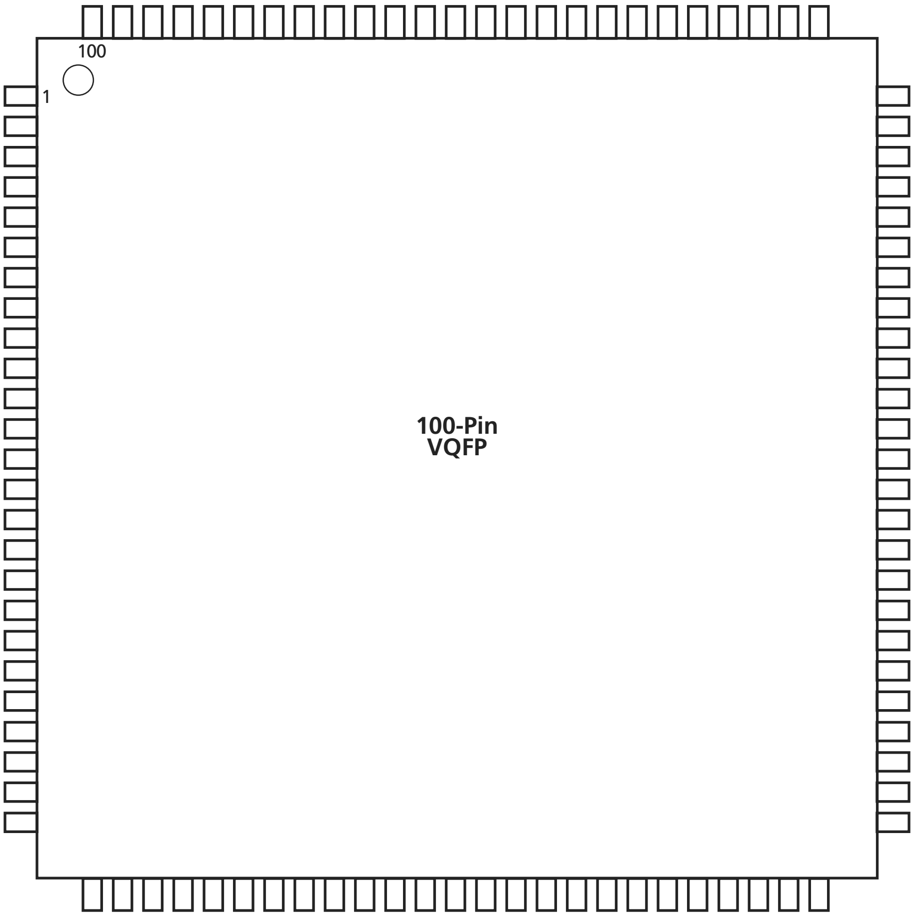


Figure 2-5 • **100-Pin VQFP (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

100-Pin VQFP		
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**
1	GND	GND
2	TDI, I/O	TDI, I/O
3	I/O	I/O
4	I/O	I/O
5	I/O	I/O
6	I/O	I/O
7	TMS	TMS
8	V_{CCI}	V_{CCI}
9	GND	GND
10	I/O	I/O
11	I/O	I/O
12	I/O	I/O
13	I/O	I/O
14	I/O	I/O
15	I/O	I/O
16	I/O	I/O
17	I/O	I/O
18	I/O	I/O
19	I/O	I/O
20	V_{CCI}	V_{CCI}
21	I/O	I/O
22	I/O	I/O
23	I/O	I/O
24	I/O	I/O
25	I/O	I/O
26	I/O	I/O
27	I/O	I/O
28	I/O	I/O
29	I/O	I/O
30	I/O	I/O
31	I/O	I/O
32	I/O	I/O
33	I/O	I/O
34	PRB, I/O	PRB, I/O
35	V_{CCA}	V_{CCA}
36	GND	GND
37	V_{CCR}	V_{CCR}
38	I/O	I/O
39	HCLK	HCLK
40	I/O	I/O
41	I/O	I/O
42	I/O	I/O
43	I/O	I/O
44	V_{CCI}	V_{CCI}
45	I/O	I/O
46	I/O	I/O
47	I/O	I/O
48	I/O	I/O
49	TDO, I/O	TDO, I/O
50	I/O	I/O
51	GND	GND
52	I/O	I/O
53	I/O	I/O
54	I/O	I/O
55	I/O	I/O
56	I/O	I/O
57	V_{CCA}	V_{CCA}
58	V_{CCI}	V_{CCI}
59	I/O	I/O
60	I/O	I/O
61	I/O	I/O
62	I/O	I/O
63	I/O	I/O
64	I/O	I/O
65	I/O	I/O
66	I/O	I/O
67	V_{CCA}	V_{CCA}
68	GND	GND
69	GND	GND
70	I/O	I/O
71	I/O	I/O
72	I/O	I/O
73	I/O	I/O
74	I/O	I/O
75	I/O	I/O
76	I/O	I/O
77	I/O	I/O
78	I/O	I/O
79	I/O	I/O
80	I/O	I/O
81	I/O	I/O
82	V_{CCI}	V_{CCI}
83	I/O	I/O
84	I/O	I/O
85	I/O	I/O
86	I/O	I/O
87	CLKA	CLKA
88	CLKB	CLKB
89	V_{CCR}	V_{CCR}
90	V_{CCA}	V_{CCA}
91	GND	GND
92	PRA, I/O	PRA, I/O
93	I/O	I/O
94	I/O	I/O
95	I/O	I/O
96	I/O	I/O
97	I/O	I/O
98	I/O	I/O
99	I/O	I/O
100	TCK, I/O	TCK, I/O

313-Pin PBGA

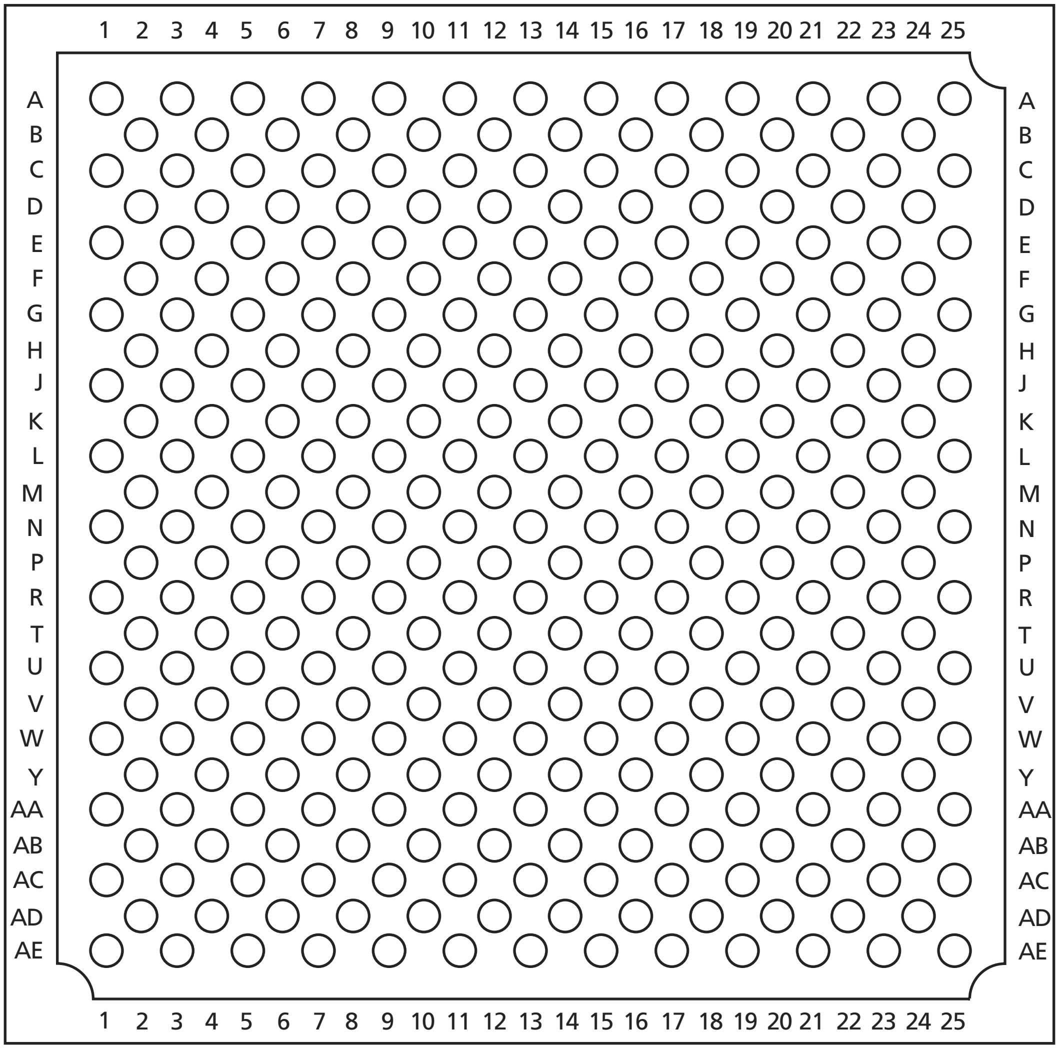


Figure 2-6 • **313-Pin PBGA (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

329-Pin PBGA

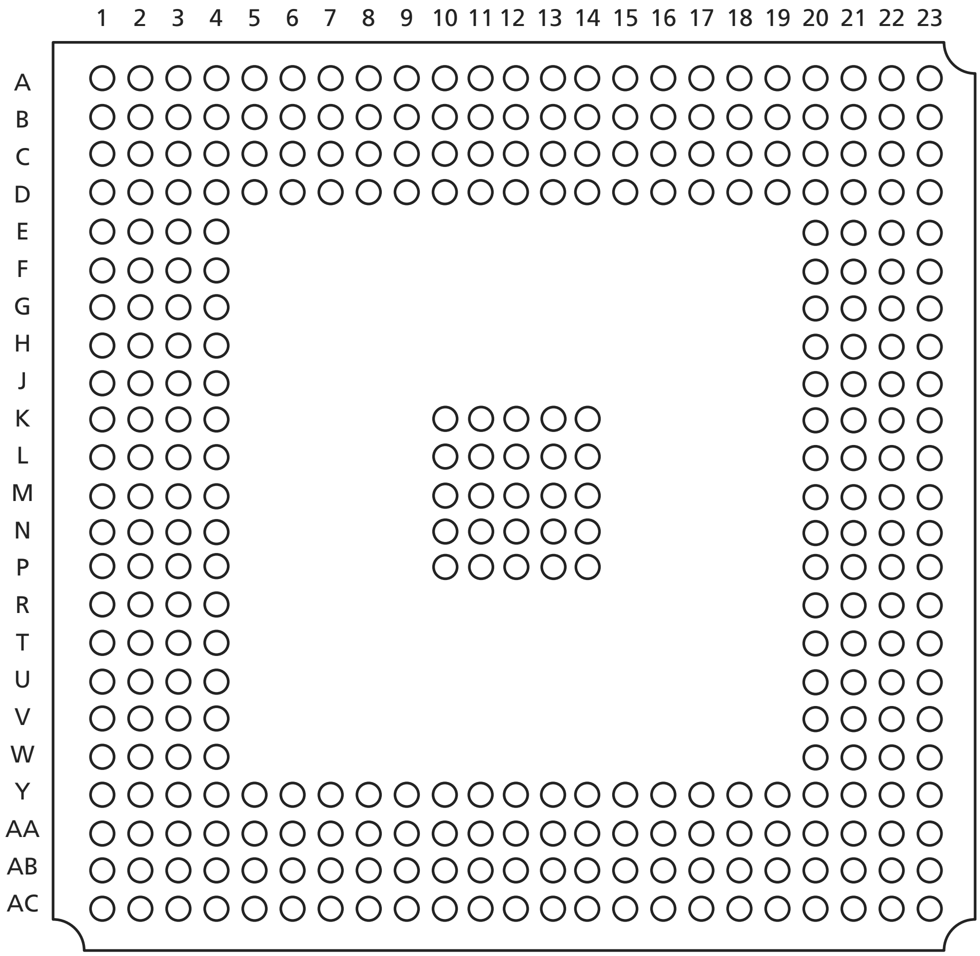


Figure 2-7 • **329-Pin PBGA (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.