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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	768
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	113
Number of Gates	12000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx08-1tq144

SX Family FPGAs

General Description

The Actel SX family of FPGAs features a sea-of-modules architecture that delivers device performance and integration levels not currently achieved by any other FPGA architecture. SX devices greatly simplify design time, enable dramatic reductions in design costs and power consumption, and further decrease time to market for performance-intensive applications.

The Actel SX architecture features two types of logic modules, the combinatorial cell (C-cell) and the register cell (R-cell), each optimized for fast and efficient mapping of synthesized logic functions. The routing and interconnect resources are in the metal layers above the logic modules, providing optimal use of silicon. This enables the entire floor of the device to be spanned with an uninterrupted grid of fine-grained, synthesis-friendly logic modules (or "sea-of-modules"), which reduces the distance signals have to travel between logic modules. To minimize signal propagation delay, SX devices employ both local and general routing resources. The high-speed local routing resources (DirectConnect and FastConnect) enable very fast local signal propagation that is optimal for fast counters, state machines, and datapath logic. The general system of segmented routing tracks allows any logic module in the array to be connected to any other logic or I/O module. Within this system, propagation delay is minimized by limiting the number of antifuse interconnect elements to five (90 percent of connections typically use only three antifuses). The unique local and general routing structure featured in SX devices gives fast and predictable performance, allows 100 percent pin-locking with full logic utilization, enables concurrent PCB development, reduces design time, and allows designers to achieve performance goals with minimum effort.

Further complementing SX's flexible routing structure is a hardwired, constantly loaded clock network that has been tuned to provide fast clock propagation with minimal clock skew. Additionally, the high performance of the internal logic has eliminated the need to embed latches or flip-flops in the I/O cells to achieve fast clock-to-out or fast input setup times. SX devices have easy to use I/O cells that do not require HDL instantiation, facilitating design reuse and reducing design and verification time.

SX Family Architecture

The SX family architecture was designed to satisfy next-generation performance and integration requirements for production-volume designs in a broad range of applications.

Programmable Interconnect Element

The SX family provides efficient use of silicon by locating the routing interconnect resources between the Metal 2 (M2) and Metal 3 (M3) layers (Figure 1-1 on page 1-2). This completely eliminates the channels of routing and interconnect resources between logic modules (as implemented on SRAM FPGAs and previous generations of antifuse FPGAs), and enables the entire floor of the device to be spanned with an uninterrupted grid of logic modules.

Interconnection between these logic modules is achieved using The Actel patented metal-to-metal programmable antifuse interconnect elements, which are embedded between the M2 and M3 layers. The antifuses are normally open circuit and, when programmed, form a permanent low-impedance connection.

The extremely small size of these interconnect elements gives the SX family abundant routing resources and provides excellent protection against design pirating. Reverse engineering is virtually impossible because it is extremely difficult to distinguish between programmed and unprogrammed antifuses, and there is no configuration bitstream to intercept.

Additionally, the interconnect elements (i.e., the antifuses and metal tracks) have lower capacitance and lower resistance than any other device of similar capacity, leading to the fastest signal propagation in the industry.

Logic Module Design

The SX family architecture is described as a "sea-of-modules" architecture because the entire floor of the device is covered with a grid of logic modules with virtually no chip area lost to interconnect elements or routing. The Actel SX family provides two types of logic modules, the register cell (R-cell) and the combinatorial cell (C-cell).

Boundary Scan Testing (BST)

All SX devices are IEEE 1149.1 compliant. SX devices offer superior diagnostic and testing capabilities by providing Boundary Scan Testing (BST) and probing capabilities. These functions are controlled through the special test pins in conjunction with the program fuse. The functionality of each pin is described in Table 1-2. In the dedicated test mode, TCK, TDI, and TDO are dedicated pins and cannot be used as regular I/Os. In flexible mode, TMS should be set HIGH through a pull-up resistor of 10 kΩ. TMS can be pulled LOW to initiate the test sequence.

The program fuse determines whether the device is in dedicated or flexible mode. The default (fuse not blown) is flexible mode.

Table 1-2 • **Boundary Scan Pin Functionality**

Program Fuse Blown (Dedicated Test Mode)	**Program Fuse Not Blown (Flexible Mode)**
TCK, TDI, TDO are dedicated BST pins.	TCK, TDI, TDO are flexible and may be used as I/Os.
No need for pull-up resistor for TMS	Use a pull-up resistor of 10 kΩ on TMS.

Dedicated Test Mode

In Dedicated mode, all JTAG pins are reserved for BST; designers cannot use them as regular I/Os. An internal pull-up resistor is automatically enabled on both TMS and TDI pins, and the TMS pin will function as defined in the IEEE 1149.1 (JTAG) specification.

To select Dedicated mode, users need to reserve the JTAG pins in Actel's Designer software by checking the "Reserve JTAG" box in "Device Selection Wizard" (Figure 1-7). JTAG pins comply with LVTTL/TTL I/O specification regardless of whether they are used as a user I/O or a JTAG I/O. Refer to the Table 1-5 on page 1-8 for detailed specifications.

Figure 1-7 • **Device Selection Wizard**

Development Tool Support

The SX family of FPGAs is fully supported by both the Actel Libero® Integrated Design Environment (IDE) and Designer FPGA Development software. Actel Libero IDE is a design management environment, seamlessly integrating design tools while guiding the user through the design flow, managing all design and log files, and passing necessary design data among tools. Libero IDE allows users to integrate both schematic and HDL synthesis into a single flow and verify the entire design in a single environment. Libero IDE includes Synplify® for Actel from Synplicity®, ViewDraw® for Actel from Mentor Graphics®, Model*Sim*® HDL Simulator from Mentor Graphics, WaveFormer Lite™ from SynaptiCAD™, and Designer software from Actel. Refer to the Libero IDE flow diagram (located on the Actel website) for more information.

Actel Designer software is a place-and-route tool and provides a comprehensive suite of backend support tools for FPGA development. The Designer software includes timing-driven place-and-route, and a world-class integrated static timing analyzer and constraints editor. With the Designer software, a user can select and lock package pins while only minimally impacting the results of place-and-route. Additionally, the back-annotation flow is compatible with all the major simulators, and the simulation results can be cross-probed with Silicon Explorer II, Actel integrated verification and logic analysis tool. Another tool included in the Designer software is the SmartGen core generator, which easily creates popular and commonly used logic functions for implementation into your schematic or HDL design. Actel Designer software is compatible with the most popular FPGA design entry and verification tools from companies such as Mentor Graphics, Synplicity, Synopsys®, and Cadence® Design Systems. The Designer software is available for both the Windows® and UNIX® operating systems.

Probe Circuit Control Pins

The Silicon Explorer II tool uses the boundary scan ports (TDI, TCK, TMS, and TDO) to select the desired nets for verification. The selected internal nets are assigned to the PRA/PRB pins for observation. Figure 1-8 on page 1-7 illustrates the interconnection between Silicon Explorer II and the FPGA to perform in-circuit verification.

Design Considerations

The TDI, TCK, TDO, PRA, and PRB pins should not be used as input or bidirectional ports. Because these pins are active during probing, critical signals input through these pins are not available while probing. In addition, the Security Fuse should not be programmed because doing so disables the Probe Circuitry.

PCI Compliance for the SX Family

The SX family supports 3.3 V and 5.0 V PCI and is compliant with the PCI Local Bus Specification Rev. 2.1.

Table 1-6 • **A54SX16P DC Specifications (5.0 V PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
V_{CCA}	Supply Voltage for Array		3.0	3.6	V
V_{CCR}	Supply Voltage required for Internal Biasing		4.75	5.25	V
V_{CCI}	Supply Voltage for I/Os		4.75	5.25	V
V_{IH}	Input High Voltage[1]		2.0	V_{CC} + 0.5	V
V_{IL}	Input Low Voltage[1]		–0.5	0.8	V
I_{IH}	Input High Leakage Current	V_{IN} = 2.7		70	µA
I_{IL}	Input Low Leakage Current	V_{IN} = 0.5		–70	µA
V_{OH}	Output High Voltage	I_{OUT} = –2 mA	2.4		V
V_{OL}	Output Low Voltage[2]	I_{OUT} = 3 mA, 6 mA		0.55	V
C_{IN}	Input Pin Capacitance[3]			10	pF
C_{CLK}	CLK Pin Capacitance		5	12	pF
C_{IDSEL}	IDSEL Pin Capacitance[4]			8	pF

Notes:

1. *Input leakage currents include hi-Z output leakage for all bidirectional buffers with tristate outputs.*
2. *Signals without pull-up resistors must have 3 mA low output current. Signals requiring pull-up must have 6 mA; the latter include, FRAME#, IRDY#, TRDY#, DEVSEL#, STOP#, SERR#, PERR#, LOCK#, and, when used, AD[63::32], C/BE[7::4]#, PAR64, REQ64#, and ACK64#.*
3. *Absolute maximum pin capacitance for a PCI input is 10 pF (except for CLK).*
4. *Lower capacitance on this input-only pin allows for non-resistive coupling to AD[xx].*

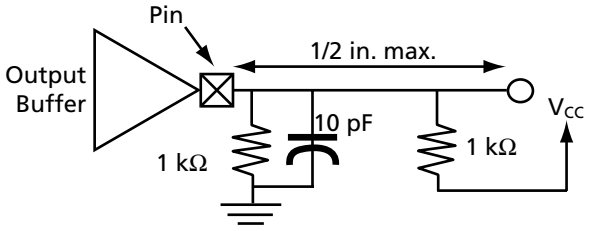
A54SX16P AC Specifications for (PCI Operation)

Table 1-7 • **A54SX16P AC Specifications for (PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \le 1.4$[1]	–44		mA
		$1.4 \le V_{OUT} < 2.4$[1, 2]	$–44 + (V_{OUT} – 1.4)/0.024$		mA
		$3.1 < V_{OUT} < V_{CC}$[1, 3]		EQ 1-1 on page 1-11	
	(Test Point)	$V_{OUT} = 3.1$[3]		–142	mA
$I_{OL(AC)}$	Switching Current High	$V_{OUT} \ge 2.2$[1]	95		mA
		$2.2 > V_{OUT} > 0.55$[1]	$V_{OUT}/0.023$		
		$0.71 > V_{OUT} > 0$[1, 3]		EQ 1-2 on page 1-11	mA
	(Test Point)	$V_{OUT} = 0.71$[3]		206	mA
I_{CL}	Low Clamp Current	$–5 < V_{IN} \le –1$	$–25 + (V_{IN} + 1)/0.015$		mA
$slew_R$	Output Rise Slew Rate	0.4 V to 2.4 V load[4]	1	5	V/ns
$slew_F$	Output Fall Slew Rate	2.4 V to 0.4 V load[4]	1	5	V/ns

Notes:

1. *Refer to the V/I curves in Figure 1-9 on page 1-11. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half-size output drivers may be used on these signals. This specification does not apply to CLK and RST#, which are system outputs. "Switching Current High" specifications are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD#, which are open drain outputs.*
2. *Note that this segment of the minimum current curve is drawn from the AC drive point directly to the DC drive point rather than toward the voltage rail (as is done in the pull-down curve). This difference is intended to allow for an optional N-channel pull-up.*
3. *Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (A and B) are provided with the respective diagrams in Figure 1-9 on page 1-11. The equation defined maxima should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.*
4. *This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per revision 2.0 of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is now required (the maximum is no longer simply a guideline). Since adherence to the maximum slew rate was not required prior to revision 2.1 of the specification, there may be components in the market for some time that have faster edge rates; therefore, motherboard designers must bear in mind that rise and fall times faster than this specification could occur, and should ensure that signal integrity modeling accounts for this. Rise slew rate does not apply to open drain outputs.*



A54SX16P AC Specifications (3.3 V PCI Operation)

Table 1-9 • **A54SX16P AC Specifications (3.3 V PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \le 0.3V_{CC}$ [1]			mA
		$0.3V_{CC} \le V_{OUT} < 0.9V_{CC}$ [1]	$-12V_{CC}$		mA
		$0.7V_{CC} < V_{OUT} < V_{CC}$ [1, 2]	$-17.1 + (V_{CC} - V_{OUT})$	EQ 1-3 on page 1-14	
	(Test Point)	$V_{OUT} = 0.7V_{CC}$ [2]		$-32V_{CC}$	mA
$I_{OL(AC)}$	Switching Current High	$V_{CC} > V_{OUT} \ge 0.6V_{CC}$ [1]			mA
		$0.6V_{CC} > V_{OUT} > 0.1V_{CC}$ [1]	$16V_{CC}$		mA
		$0.18V_{CC} > V_{OUT} > 0$ [1, 2]	$26.7V_{OUT}$	EQ 1-4 on page 1-14	mA
	(Test Point)	$V_{OUT} = 0.18V_{CC}$ [2]		$38V_{CC}$	
I_{CL}	Low Clamp Current	$-3 < V_{IN} \le -1$	$-25 + (V_{IN} + 1)/0.015$		mA
I_{CH}	High Clamp Current	$-3 < V_{IN} \le -1$	$25 + (V_{IN} - V_{OUT} - 1)/0.015$		mA
$slew_R$	Output Rise Slew Rate[3]	$0.2V_{CC}$ to $0.6V_{CC}$ load	1	4	V/ns
$slew_F$	Output Fall Slew Rate[3]	$0.6V_{CC}$ to $0.2V_{CC}$ load	1	4	V/ns

Notes:

1. *Refer to the V/I curves in Figure 1-10 on page 1-14. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half size output drivers may be used on these signals. This specification does not apply to CLK and RST# which are system outputs. "Switching Current High" specification are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD# which are open drain outputs.*
2. *Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (C and D) are provided with the respective diagrams in Figure 1-10 on page 1-14. The equation defined maxima should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.*
3. *This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per the latest revision of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is required (the maximum is no longer simply a guideline). Rise slew rate does not apply to open drain outputs.*

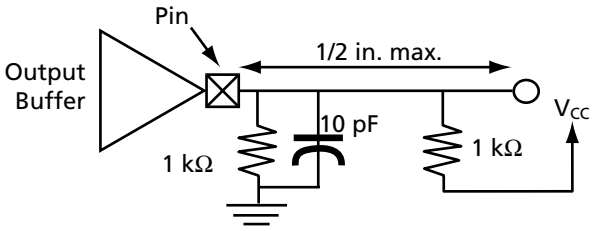


Figure 1-10 shows the 3.3 V PCI V/I curve and the minimum and maximum PCI drive characteristics of the A54SX16P device.

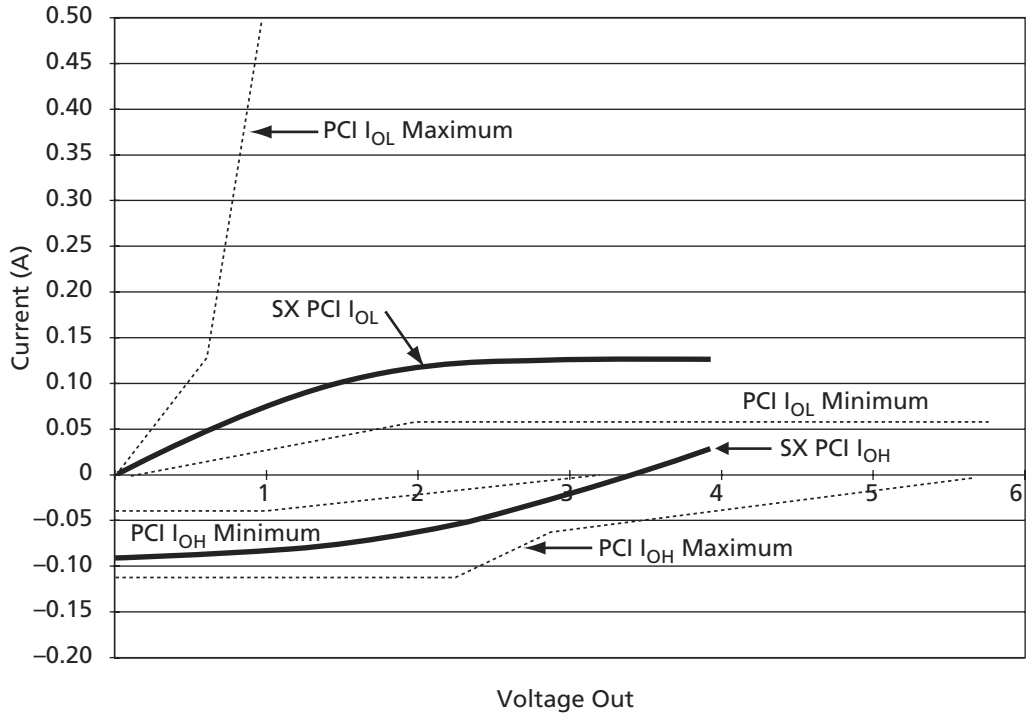


Figure 1-10 • **3.3 V PCI Curve for A54SX16P Device**

$$I_{OH} = (98.0/V_{CC}) \times (V_{OUT} - V_{CC}) \times (V_{OUT} + 0.4V_{CC})$$
$$\text{for } V_{CC} > V_{OUT} > 0.7\ V_{CC} \qquad \textit{EQ 1-3}$$

$$I_{OL} = (256/V_{CC}) \times V_{OUT} \times (V_{CC} - V_{OUT})$$
$$\text{for } 0\ V < V_{OUT} < 0.18\ V_{CC} \qquad \textit{EQ 1-4}$$

Power-Up Sequencing

Table 1-10 • **Power-Up Sequencing**

V_{CCA}	V_{CCR}	V_{CCI}	Power-Up Sequence	Comments
A54SX08, A54SX16, A54SX32				
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	Possible damage to device
A54SX16P				
3.3 V	3.3 V	3.3 V	3.3 V Only	No possible damage to device
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	Possible damage to device
3.3 V	5.0 V	5.0 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device

Note: *No inputs should be driven (high or low) before completion of power-up.*

Power-Down Sequencing

Table 1-11 • **Power-Down Sequencing**

V_{CCA}	V_{CCR}	V_{CCI}	Power-Down Sequence	Comments
A54SX08, A54SX16, A54SX32				
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	Possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device
A54SX16P				
3.3 V	3.3 V	3.3 V	3.3 V Only	No possible damage to device
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	Possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device
3.3 V	5.0 V	5.0 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device

Note: *No inputs should be driven (high or low) after the beginning of the power-down sequence.*

Figure 1-11 shows the characterized power dissipation numbers for the shift register design using frequencies ranging from 1 MHz to 200 MHz.

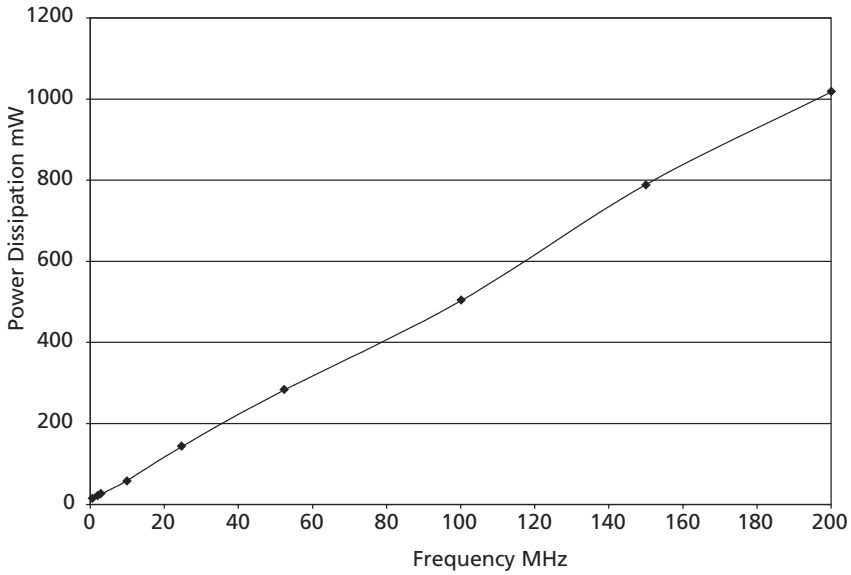


Figure 1-11 • **Power Dissipation**

Junction Temperature (T_J)

The temperature that you select in Designer Series software is the junction temperature, not ambient temperature. This is an important distinction because the heat generated from dynamic power consumption is usually hotter than the ambient temperature. Use the equation below to calculate junction temperature.

$$\text{Junction Temperature} = \Delta T + T_a$$

EQ 1-13

Where:

T_a = Ambient Temperature

ΔT = Temperature gradient between junction (silicon) and ambient

$\Delta T = \theta_{ja} \times P$

P = Power calculated from Estimating Power Consumption section

θ_{ja} = Junction to ambient of package. θ_{ja} numbers are located in the "Package Thermal Characteristics" section.

Package Thermal Characteristics

The device junction to case thermal characteristic is θ_{jc}, and the junction to ambient air characteristic is θ_{ja}. The thermal characteristics for θ_{ja} are shown with two different air flow rates.

The maximum junction temperature is 150 °C.

A sample calculation of the absolute maximum power dissipation allowed for a TQFP 176-pin package at commercial temperature and still air is as follows:

$$\text{Maximum Power Allowed} = \frac{\text{Max. junction temp. (°C)} - \text{Max. ambient temp. (°C)}}{\theta_{ja}\text{ (°C/W)}} = \frac{150°C - 70°C}{28°C/W} = 2.86\text{ W}$$

EQ 1-14

A54SX16P Timing Characteristics

Table 1-19 • **A54SX16P Timing Characteristics (Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)**

Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
C-Cell Propagation Delays[1]										
t_{PD}	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted Routing Delays[2]										
t_{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t_{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{RD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{RD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timing										
t_{RCO}	Sequential Clock-to-Q		0.9		1.1		1.3		1.4	ns
t_{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t_{PRESET}	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t_{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t_{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t_{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t_{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted Input Routing Delays[2]										
t_{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{IRD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{IRD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Note:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 10 pF loading.*

Table 1-19 • **A54SX16P Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Dedicated (Hardwired) Array Clock Network										
t_{HCKH}	Input LOW to HIGH (pad to R-Cell input)		1.2		1.4		1.5		1.8	ns
t_{HCKL}	Input HIGH to LOW (pad to R-Cell input)		1.2		1.4		1.6		1.9	ns
t_{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t_{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t_{HCKSW}	Maximum Skew		0.2		0.2		0.3		0.3	ns
t_{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f_{HMAX}	Maximum Frequency		350		320		280		240	MHz
Routed Array Clock Networks										
t_{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)		1.6		1.8		2.1		2.5	ns
t_{RCKL}	Input HIGH to LOW (Light Load) (pad to R-Cell input)		1.8		2.0		2.3		2.7	ns
t_{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)		1.8		2.1		2.5		2.8	ns
t_{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)		1.8		2.1		2.4		2.8	ns
t_{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t_{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t_{RCKSW}	Maximum Skew (light load)		0.5		0.5		0.5		0.7	ns
t_{RCKSW}	Maximum Skew (50% load)		0.5		0.6		0.7		0.8	ns
t_{RCKSW}	Maximum Skew (100% load)		0.5		0.6		0.7		0.8	ns
TTL Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		2.4		2.8		3.1		3.7	ns
t_{DHL}	Data-to-Pad HIGH to LOW		2.3		2.9		3.2		3.8	ns
t_{ENZL}	Enable-to-Pad, Z to L		3.0		3.4		3.9		4.6	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.3		3.8		4.3		5.0	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.3		2.7		3.0		3.5	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.8		3.2		3.7		4.3	ns

Note:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 10 pF loading.*

Table 1-20 • **A54SX32 Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR}= 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Dedicated (Hardwired) Array Clock Network										
t_{HCKH}	Input LOW to HIGH (pad to R-Cell input)		1.9		2.1		2.4		2.8	ns
t_{HCKL}	Input HIGH to LOW (pad to R-Cell input)		1.9		2.1		2.4		2.8	ns
t_{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t_{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t_{HCKSW}	Maximum Skew		0.3		0.4		0.4		0.5	ns
t_{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f_{HMAX}	Maximum Frequency		350		320		280		240	MHz
Routed Array Clock Networks										
t_{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)		2.4		2.7		3.0		3.5	ns
t_{RCKL}	Input HIGH to LOW (light load) (pad to R-Cell input)		2.4		2.7		3.1		3.6	ns
t_{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)		2.7		3.0		3.5		4.1	ns
t_{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)		2.7		3.1		3.6		4.2	ns
t_{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)		2.7		3.1		3.5		4.1	ns
t_{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)		2.8		3.2		3.6		4.3	ns
t_{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t_{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t_{RCKSW}	Maximum Skew (light load)		0.85		0.98		1.1		1.3	ns
t_{RCKSW}	Maximum Skew (50% load)		1.23		1.4		1.6		1.9	ns
t_{RCKSW}	Maximum Skew (100% load)		1.30		1.5		1.7		2.0	ns
TTL Output Module Timing[3]										
t_{DLH}	Data-to-Pad LOW to HIGH		1.6		1.9		2.1		2.5	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.6		1.9		2.1		2.5	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.1		2.4		2.8		3.2	ns
t_{ENZH}	Enable-to-Pad, Z to H		2.3		2.7		3.1		3.6	ns
t_{ENLZ}	Enable-to-Pad, L to Z		1.4		1.7		1.9		2.2	ns
t_{ENHZ}	Enable-to-Pad, H to Z		1.3		1.5		1.7		2.0	ns

Note:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH}. For t_{ENZL} and t_{ENZH} the loading is 5 pF.*

208-Pin PQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
73	NC	I/O	I/O
74	I/O	I/O	I/O
75	NC	I/O	I/O
76	PRB, I/O	PRB, I/O	PRB, I/O
77	GND	GND	GND
78	V_{CCA}	V_{CCA}	V_{CCA}
79	GND	GND	GND
80	V_{CCR}	V_{CCR}	V_{CCR}
81	I/O	I/O	I/O
82	HCLK	HCLK	HCLK
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	NC	I/O	I/O
86	I/O	I/O	I/O
87	I/O	I/O	I/O
88	NC	I/O	I/O
89	I/O	I/O	I/O
90	I/O	I/O	I/O
91	NC	I/O	I/O
92	I/O	I/O	I/O
93	I/O	I/O	I/O
94	NC	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	NC	I/O	I/O
98	V_{CCI}	V_{CCI}	V_{CCI}
99	I/O	I/O	I/O
100	I/O	I/O	I/O
101	I/O	I/O	I/O
102	I/O	I/O	I/O
103	TDO, I/O	TDO, I/O	TDO, I/O
104	I/O	I/O	I/O
105	GND	GND	GND
106	NC	I/O	I/O
107	I/O	I/O	I/O
108	NC	I/O	I/O

208-Pin PQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
109	I/O	I/O	I/O
110	I/O	I/O	I/O
111	I/O	I/O	I/O
112	I/O	I/O	I/O
113	I/O	I/O	I/O
114	V_{CCA}	V_{CCA}	V_{CCA}
115	V_{CCI}	V_{CCI}	V_{CCI}
116	NC	I/O	I/O
117	I/O	I/O	I/O
118	I/O	I/O	I/O
119	NC	I/O	I/O
120	I/O	I/O	I/O
121	I/O	I/O	I/O
122	NC	I/O	I/O
123	I/O	I/O	I/O
124	I/O	I/O	I/O
125	NC	I/O	I/O
126	I/O	I/O	I/O
127	I/O	I/O	I/O
128	I/O	I/O	I/O
129	GND	GND	GND
130	V_{CCA}	V_{CCA}	V_{CCA}
131	GND	GND	GND
132	V_{CCR}	V_{CCR}	V_{CCR}
133	I/O	I/O	I/O
134	I/O	I/O	I/O
135	NC	I/O	I/O
136	I/O	I/O	I/O
137	I/O	I/O	I/O
138	NC	I/O	I/O
139	I/O	I/O	I/O
140	I/O	I/O	I/O
141	NC	I/O	I/O
142	I/O	I/O	I/O
143	NC	I/O	I/O
144	I/O	I/O	I/O

Note: ** Note that Pin 65 in the A54SX32—PQ208 is a no connect (NC).*

144-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16P Function	A54SX32 Function
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	I/O	I/O	I/O
4	I/O	I/O	I/O
5	I/O	I/O	I/O
6	I/O	I/O	I/O
7	I/O	I/O	I/O
8	I/O	I/O	I/O
9	TMS	TMS	TMS
10	V_{CCI}	V_{CCI}	V_{CCI}
11	GND	GND	GND
12	I/O	I/O	I/O
13	I/O	I/O	I/O
14	I/O	I/O	I/O
15	I/O	I/O	I/O
16	I/O	I/O	I/O
17	I/O	I/O	I/O
18	I/O	I/O	I/O
19	V_{CCR}	V_{CCR}	V_{CCR}
20	V_{CCA}	V_{CCA}	V_{CCA}
21	I/O	I/O	I/O
22	I/O	I/O	I/O
23	I/O	I/O	I/O
24	I/O	I/O	I/O
25	I/O	I/O	I/O
26	I/O	I/O	I/O
27	I/O	I/O	I/O
28	GND	GND	GND
29	V_{CCI}	V_{CCI}	V_{CCI}
30	V_{CCA}	V_{CCA}	V_{CCA}
31	I/O	I/O	I/O
32	I/O	I/O	I/O
33	I/O	I/O	I/O
34	I/O	I/O	I/O
35	I/O	I/O	I/O
36	GND	GND	GND

144-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16P Function	A54SX32 Function
37	I/O	I/O	I/O
38	I/O	I/O	I/O
39	I/O	I/O	I/O
40	I/O	I/O	I/O
41	I/O	I/O	I/O
42	I/O	I/O	I/O
43	I/O	I/O	I/O
44	V_{CCI}	V_{CCI}	V_{CCI}
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	I/O	I/O	I/O
49	I/O	I/O	I/O
50	I/O	I/O	I/O
51	I/O	I/O	I/O
52	I/O	I/O	I/O
53	I/O	I/O	I/O
54	PRB, I/O	PRB, I/O	PRB, I/O
55	I/O	I/O	I/O
56	V_{CCA}	V_{CCA}	V_{CCA}
57	GND	GND	GND
58	V_{CCR}	V_{CCR}	V_{CCR}
59	I/O	I/O	I/O
60	HCLK	HCLK	HCLK
61	I/O	I/O	I/O
62	I/O	I/O	I/O
63	I/O	I/O	I/O
64	I/O	I/O	I/O
65	I/O	I/O	I/O
66	I/O	I/O	I/O
67	I/O	I/O	I/O
68	V_{CCI}	V_{CCI}	V_{CCI}
69	I/O	I/O	I/O
70	I/O	I/O	I/O
71	TDO, I/O	TDO, I/O	TDO, I/O
72	I/O	I/O	I/O

176-Pin TQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
137	I/O	I/O	I/O
138	I/O	I/O	I/O
139	I/O	I/O	I/O
140	V_{CCI}	V_{CCI}	V_{CCI}
141	I/O	I/O	I/O
142	I/O	I/O	I/O
143	I/O	I/O	I/O
144	I/O	I/O	I/O
145	I/O	I/O	I/O
146	I/O	I/O	I/O
147	I/O	I/O	I/O
148	I/O	I/O	I/O
149	I/O	I/O	I/O
150	I/O	I/O	I/O
151	I/O	I/O	I/O
152	CLKA	CLKA	CLKA
153	CLKB	CLKB	CLKB
154	V_{CCR}	V_{CCR}	V_{CCR}
155	GND	GND	GND
156	V_{CCA}	V_{CCA}	V_{CCA}

176-Pin TQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
157	PRA, I/O	PRA, I/O	PRA, I/O
158	I/O	I/O	I/O
159	I/O	I/O	I/O
160	I/O	I/O	I/O
161	I/O	I/O	I/O
162	I/O	I/O	I/O
163	I/O	I/O	I/O
164	I/O	I/O	I/O
165	I/O	I/O	I/O
166	I/O	I/O	I/O
167	I/O	I/O	I/O
168	NC	I/O	I/O
169	V_{CCI}	V_{CCI}	V_{CCI}
170	I/O	I/O	I/O
171	NC	I/O	I/O
172	NC	I/O	I/O
173	NC	I/O	I/O
174	I/O	I/O	I/O
175	I/O	I/O	I/O
176	TCK, I/O	TCK, I/O	TCK, I/O

313-Pin PBGA

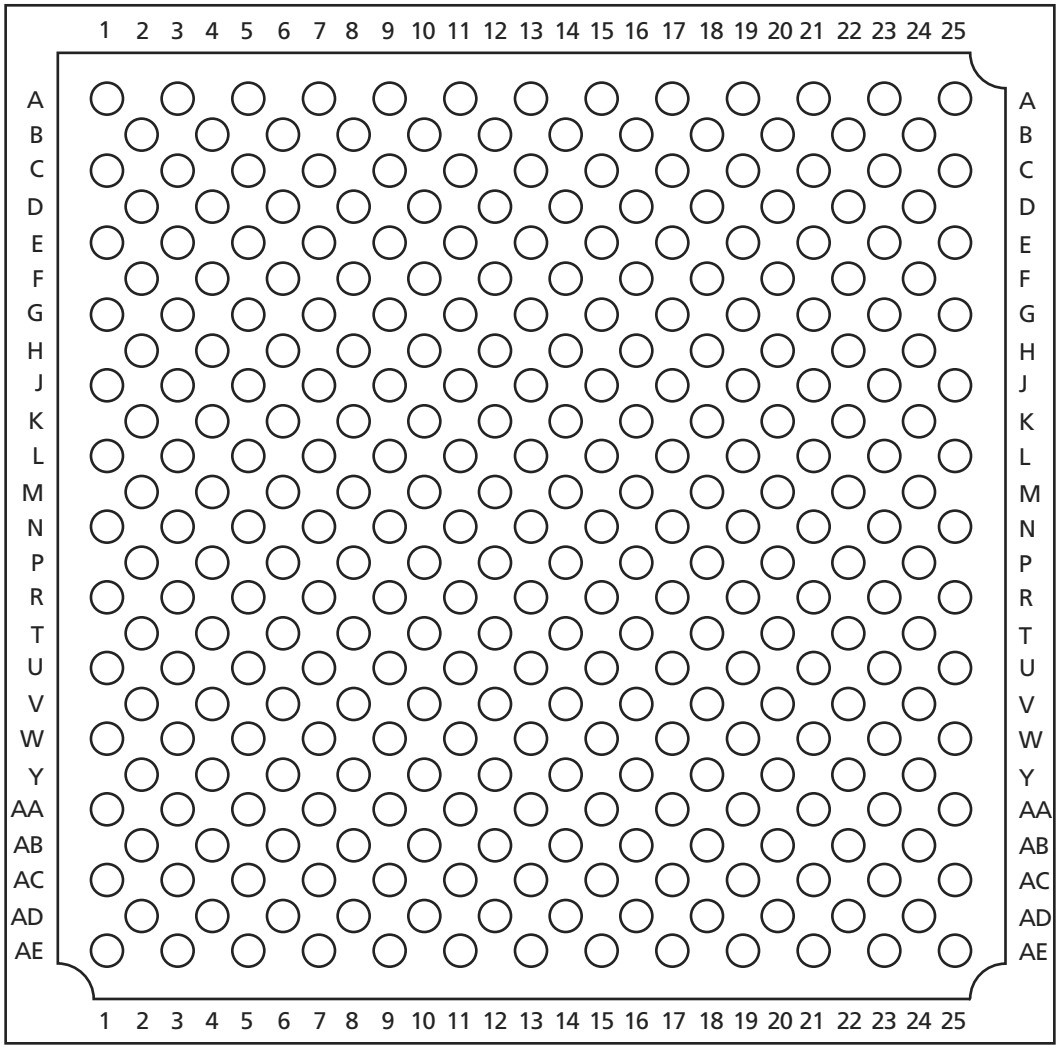


Figure 2-6 • **313-Pin PBGA (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

313-Pin PBGA	
Pin Number	**A54SX32 Function**
H20	I/O
H22	V_{CCI}
H24	I/O
J1	I/O
J3	I/O
J5	I/O
J7	NC
J9	I/O
J11	I/O
J13	CLKA
J15	I/O
J17	I/O
J19	I/O
J21	GND
J23	I/O
J25	I/O
K2	I/O
K4	I/O
K6	I/O
K8	V_{CCI}
K10	I/O
K12	I/O
K14	I/O
K16	I/O
K18	I/O
K20	V_{CCA}
K22	I/O
K24	I/O
L1	I/O
L3	I/O
L5	I/O
L7	I/O
L9	I/O
L11	I/O
L13	GND
L15	I/O
L17	I/O
L19	I/O
L21	I/O
L23	I/O

313-Pin PBGA	
Pin Number	**A54SX32 Function**
L25	I/O
M2	I/O
M4	I/O
M6	I/O
M8	I/O
M10	I/O
M12	GND
M14	GND
M16	V_{CCI}
M18	I/O
M20	I/O
M22	I/O
M24	I/O
N1	I/O
N3	V_{CCA}
N5	V_{CCR}
N7	I/O
N9	V_{CCI}
N11	GND
N13	GND
N15	GND
N17	I/O
N19	I/O
N21	I/O
N23	V_{CCR}
N25	V_{CCA}
P2	I/O
P4	I/O
P6	I/O
P8	I/O
P10	I/O
P12	GND
P14	GND
P16	I/O
P18	I/O
P20	NC
P22	I/O
P24	I/O
R1	I/O
R3	I/O

313-Pin PBGA	
Pin Number	**A54SX32 Function**
R5	I/O
R7	I/O
R9	I/O
R11	I/O
R13	GND
R15	I/O
R17	I/O
R19	I/O
R21	I/O
R23	I/O
R25	I/O
T2	I/O
T4	I/O
T6	I/O
T8	I/O
T10	I/O
T12	I/O
T14	HCLK
T16	I/O
T18	I/O
T20	I/O
T22	I/O
T24	I/O
U1	I/O
U3	I/O
U5	V_{CCI}
U7	I/O
U9	I/O
U11	I/O
U13	I/O
U15	I/O
U17	I/O
U19	I/O
U21	I/O
U23	I/O
U25	I/O
V2	V_{CCA}
V4	I/O
V6	I/O
V8	I/O

313-Pin PBGA	
Pin Number	**A54SX32 Function**
V10	I/O
V12	I/O
V14	I/O
V16	NC
V18	I/O
V20	I/O
V22	V_{CCA}
V24	V_{CCI}
W1	I/O
W3	I/O
W5	I/O
W7	NC
W9	I/O
W11	I/O
W13	V_{CCI}
W15	I/O
W17	I/O
W19	I/O
W21	I/O
W23	I/O
W25	I/O
Y2	I/O
Y4	I/O
Y6	I/O
Y8	I/O
Y10	I/O
Y12	I/O
Y14	I/O
Y16	I/O
Y18	I/O
Y20	NC
Y22	I/O
Y24	NC

329-Pin PBGA	
Pin Number	**A54SX32 Function**
A1	GND
A2	GND
A3	V_{CCI}
A4	NC
A5	I/O
A6	I/O
A7	V_{CCI}
A8	NC
A9	I/O
A10	I/O
A11	I/O
A12	I/O
A13	CLKB
A14	I/O
A15	I/O
A16	I/O
A17	I/O
A18	I/O
A19	I/O
A20	I/O
A21	NC
A22	V_{CCI}
A23	GND
AA1	V_{CCI}
AA2	I/O
AA3	GND
AA4	I/O
AA5	I/O
AA6	I/O
AA7	I/O
AA8	I/O
AA9	I/O
AA10	I/O
AA11	I/O
AA12	I/O

329-Pin PBGA	
Pin Number	**A54SX32 Function**
AA13	I/O
AA14	I/O
AA15	I/O
AA16	I/O
AA17	I/O
AA18	I/O
AA19	I/O
AA20	TDO, I/O
AA21	V_{CCI}
AA22	I/O
AA23	V_{CCI}
AB1	I/O
AB2	GND
AB3	I/O
AB4	I/O
AB5	I/O
AB6	I/O
AB7	I/O
AB8	I/O
AB9	I/O
AB10	I/O
AB11	PRB, I/O
AB12	I/O
AB13	HCLK
AB14	I/O
AB15	I/O
AB16	I/O
AB17	I/O
AB18	I/O
AB19	I/O
AB20	I/O
AB21	I/O
AB22	GND
AB23	I/O
AC1	GND

329-Pin PBGA	
Pin Number	**A54SX32 Function**
AC2	V_{CCI}
AC3	NC
AC4	I/O
AC5	I/O
AC6	I/O
AC7	I/O
AC8	I/O
AC9	V_{CCI}
AC10	I/O
AC11	I/O
AC12	I/O
AC13	I/O
AC14	I/O
AC15	NC
AC16	I/O
AC17	I/O
AC18	I/O
AC19	I/O
AC20	I/O
AC21	NC
AC22	V_{CCI}
AC23	GND
B1	V_{CCI}
B2	GND
B3	I/O
B4	I/O
B5	I/O
B6	I/O
B7	I/O
B8	I/O
B9	I/O
B10	I/O
B11	I/O
B12	PRA, I/O
B13	CLKA

329-Pin PBGA	
Pin Number	**A54SX32 Function**
B14	I/O
B15	I/O
B16	I/O
B17	I/O
B18	I/O
B19	I/O
B20	I/O
B21	I/O
B22	GND
B23	V_{CCI}
C1	NC
C2	TDI, I/O
C3	GND
C4	I/O
C5	I/O
C6	I/O
C7	I/O
C8	I/O
C9	I/O
C10	I/O
C11	I/O
C12	I/O
C13	I/O
C14	I/O
C15	I/O
C16	I/O
C17	I/O
C18	I/O
C19	I/O
C20	I/O
C21	V_{CCI}
C22	GND
C23	NC
D1	I/O
D2	I/O

329-Pin PBGA	
Pin Number	**A54SX32 Function**
T22	I/O
T23	I/O
U1	I/O
U2	I/O
U3	V_{CCA}
U4	I/O
U20	I/O
U21	V_{CCA}
U22	I/O
U23	I/O
V1	V_{CCI}
V2	I/O
V3	I/O

329-Pin PBGA	
Pin Number	**A54SX32 Function**
V4	I/O
V20	I/O
V21	I/O
V22	I/O
V23	I/O
W1	I/O
W2	I/O
W3	I/O
W4	I/O
W20	I/O
W21	I/O
W22	I/O

329-Pin PBGA	
Pin Number	**A54SX32 Function**
W23	NC
Y1	NC
Y2	I/O
Y3	I/O
Y4	GND
Y5	I/O
Y6	I/O
Y7	I/O
Y8	I/O
Y9	I/O
Y10	I/O
Y11	I/O

329-Pin PBGA	
Pin Number	**A54SX32 Function**
Y12	V_{CCA}
Y13	V_{CCR}
Y14	I/O
Y15	I/O
Y16	I/O
Y17	I/O
Y18	I/O
Y19	I/O
Y20	GND
Y21	I/O
Y22	I/O
Y23	I/O

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