Microsemi Corporation - A54SX08-1TQ176I Datasheet

Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	768
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	128
Number of Gates	12000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	176-LQFP
Supplier Device Package	176-TQFP (24x24)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx08-1tq176i

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Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

The R-cell contains a flip-flop featuring asynchronous clear, asynchronous preset, and clock enable (using the S0 and S1 lines) control signals (Figure 1-2). The R-cell registers feature programmable clock polarity selectable on a register-by-register basis. This provides additional flexibility while allowing mapping of synthesized functions into the SX FPGA. The clock source for the R-cell can be chosen from either the hardwired clock or the routed clock.

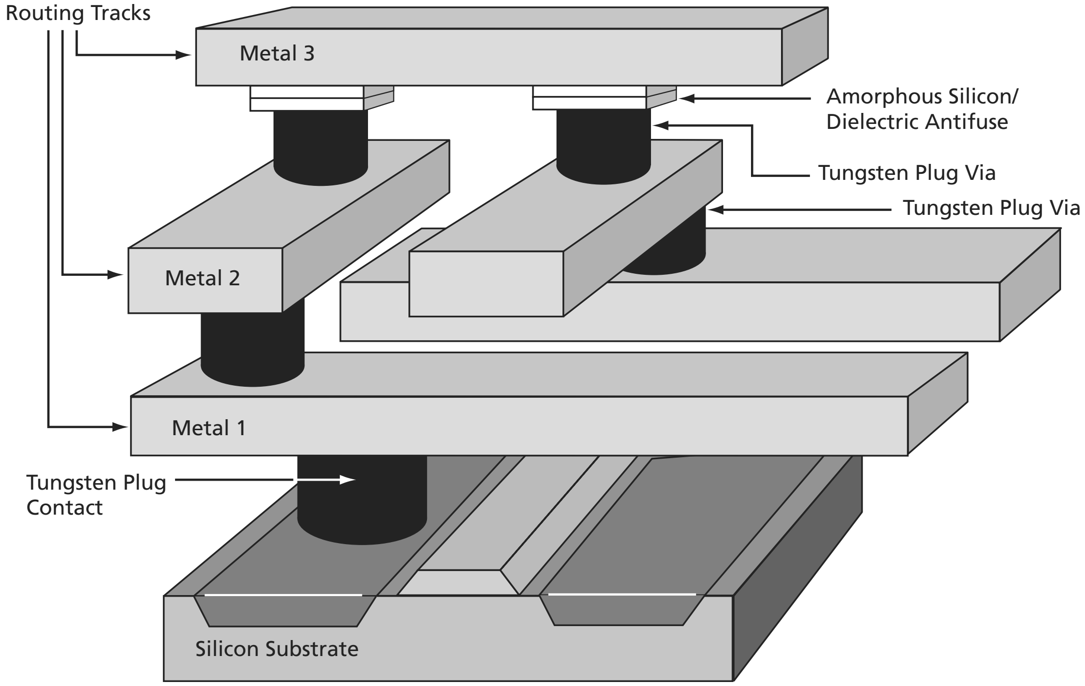


Figure 1-1 • **SX Family Interconnect Elements**

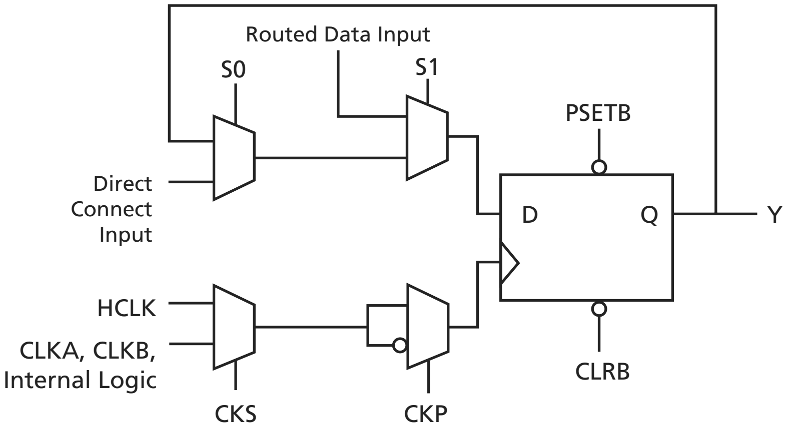


Figure 1-2 • **R-Cell**

The C-cell implements a range of combinatorial functions up to 5-inputs (Figure 1-3 on page 1-3). Inclusion of the DB input and its associated inverter function dramatically increases the number of combinatorial functions that can be implemented in a single module from 800 options in previous architectures to more than 4,000 in the SX architecture. An example of the improved flexibility enabled by the inversion capability is the ability to integrate a 3-input exclusive-OR function into a single C-cell. This facilitates construction of 9-bit parity-tree functions with 2 ns propagation delays. At the same time, the C-cell structure is extremely synthesis friendly, simplifying the overall design and reducing synthesis time.

Chip Architecture

The SX family chip architecture provides a unique approach to module organization and chip routing that delivers the best register/logic mix for a wide variety of new and emerging applications.

Module Organization

Actel has arranged all C-cell and R-cell logic modules into horizontal banks called *clusters*. There are two types of *clusters*: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells.

To increase design efficiency and device performance, Actel has further organized these modules into *SuperClusters* (Figure 1-4). SuperCluster 1 is a two-wide grouping of Type 1 clusters. SuperCluster 2 is a two-wide group containing one Type 1 cluster and one Type 2 cluster. SX devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flip-flops.

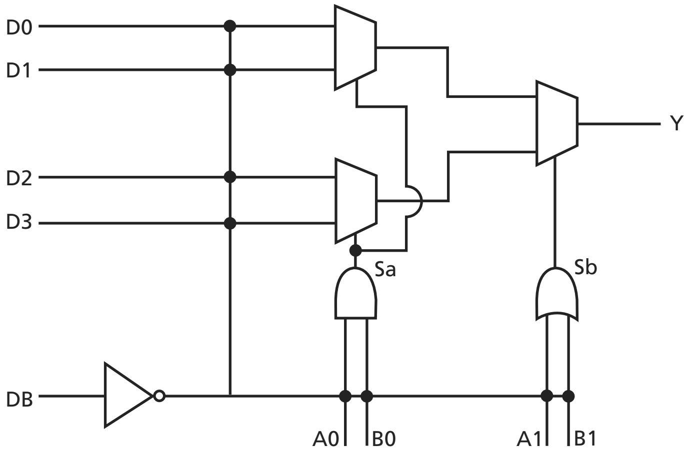


Figure 1-3 • **C-Cell**

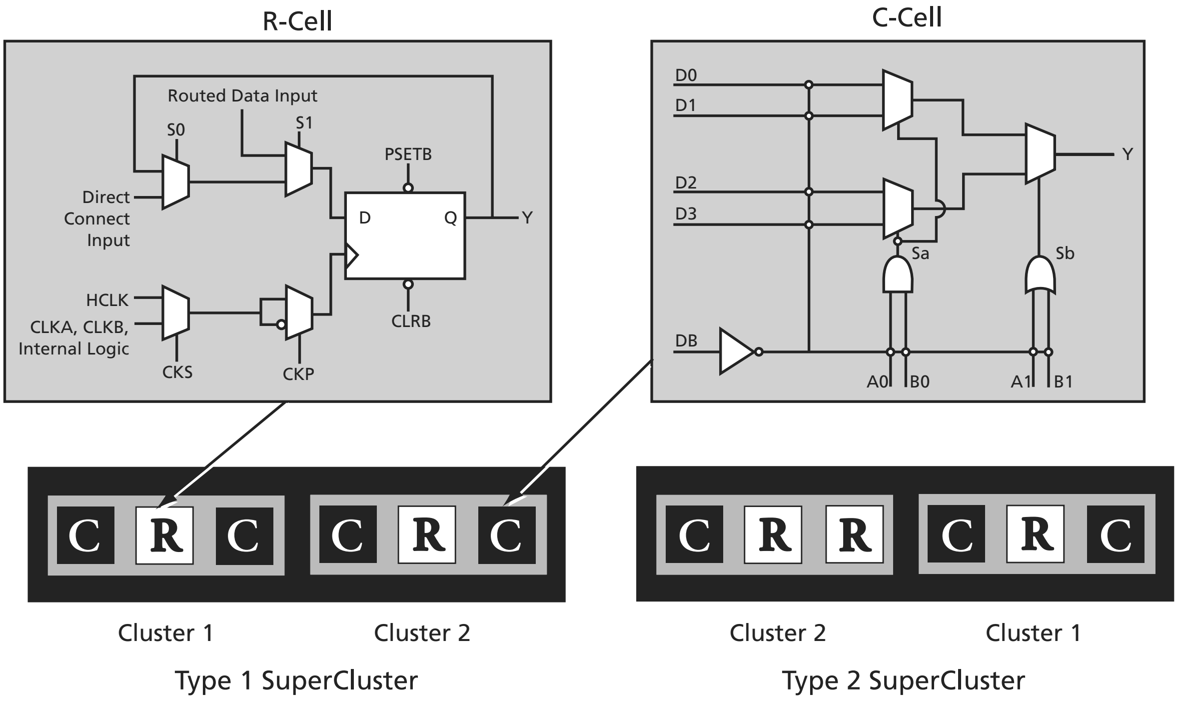


Figure 1-4 • **Cluster Organization**

Routing Resources

Clusters and SuperClusters can be connected through the use of two innovative local routing resources called *FastConnect* and *DirectConnect,* which enable extremely fast and predictable interconnection of modules within clusters and SuperClusters (Figure 1-5 and Figure 1-6). This routing architecture also dramatically reduces the number of antifuses required to complete a circuit, ensuring the highest possible performance.

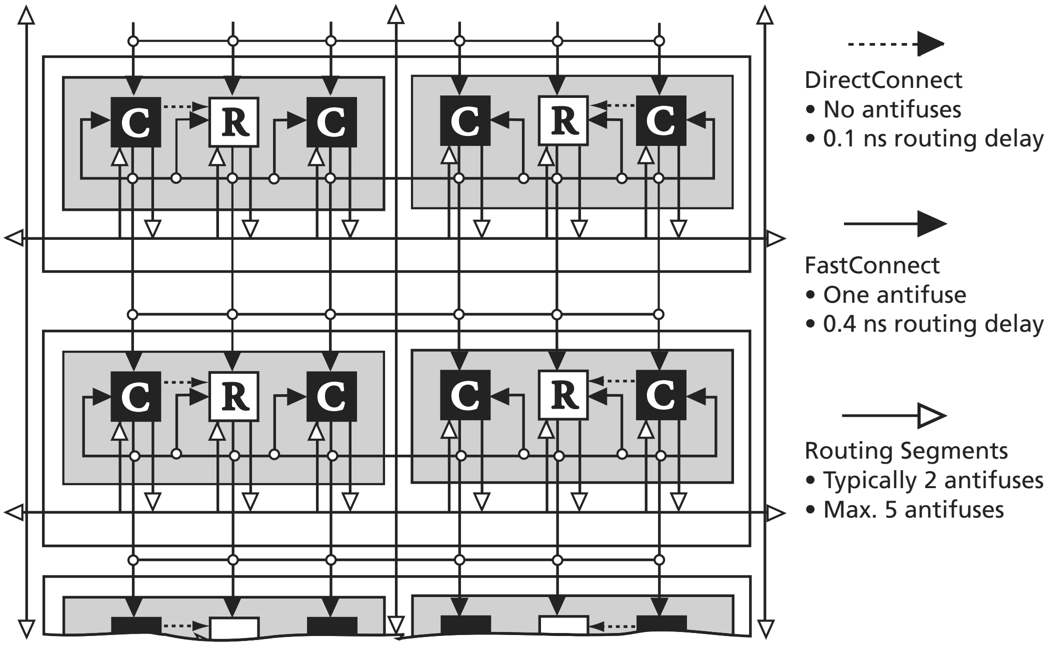


Figure 1-5 • **DirectConnect and FastConnect for Type 1 SuperClusters**

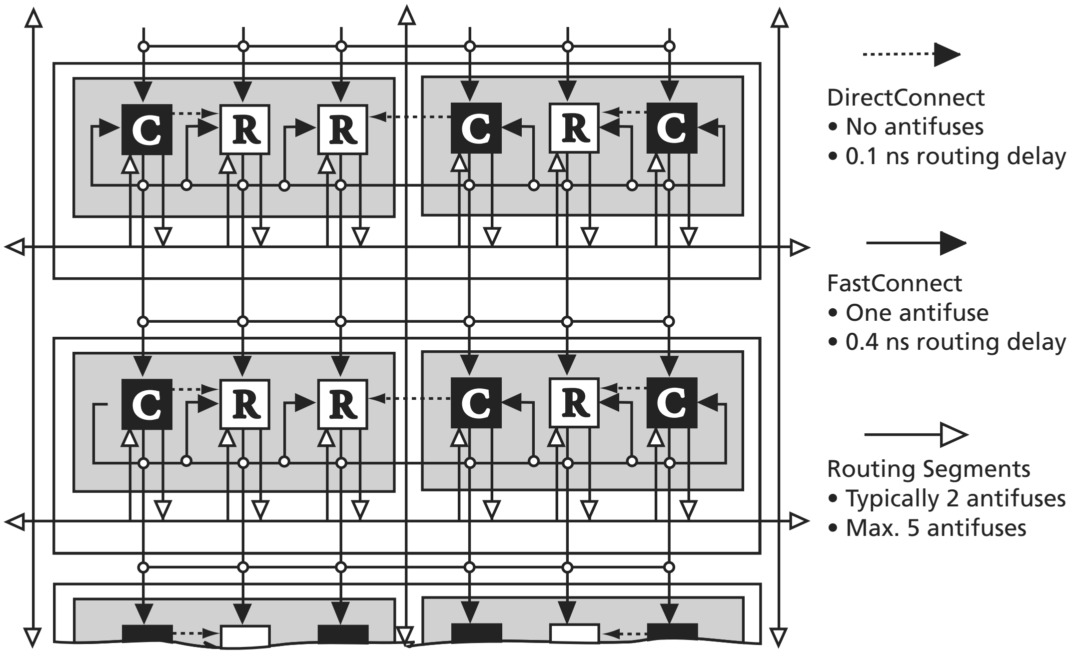


Figure 1-6 • **DirectConnect and FastConnect for Type 2 SuperClusters**

Boundary Scan Testing (BST)

All SX devices are IEEE 1149.1 compliant. SX devices offer superior diagnostic and testing capabilities by providing Boundary Scan Testing (BST) and probing capabilities. These functions are controlled through the special test pins in conjunction with the program fuse. The functionality of each pin is described in Table 1-2. In the dedicated test mode, TCK, TDI, and TDO are dedicated pins and cannot be used as regular I/Os. In flexible mode, TMS should be set HIGH through a pull-up resistor of 10 kΩ. TMS can be pulled LOW to initiate the test sequence.

The program fuse determines whether the device is in dedicated or flexible mode. The default (fuse not blown) is flexible mode.

Table 1-2 • **Boundary Scan Pin Functionality**

Program Fuse Blown (Dedicated Test Mode)	**Program Fuse Not Blown (Flexible Mode)**
TCK, TDI, TDO are dedicated BST pins.	TCK, TDI, TDO are flexible and may be used as I/Os.
No need for pull-up resistor for TMS	Use a pull-up resistor of 10 kΩ on TMS.

Dedicated Test Mode

In Dedicated mode, all JTAG pins are reserved for BST; designers cannot use them as regular I/Os. An internal pull-up resistor is automatically enabled on both TMS and TDI pins, and the TMS pin will function as defined in the IEEE 1149.1 (JTAG) specification.

To select Dedicated mode, users need to reserve the JTAG pins in Actel's Designer software by checking the "Reserve JTAG" box in "Device Selection Wizard" (Figure 1-7). JTAG pins comply with LVTTL/TTL I/O specification regardless of whether they are used as a user I/O or a JTAG I/O. Refer to the Table 1-5 on page 1-8 for detailed specifications.

Figure 1-7 • **Device Selection Wizard**

Development Tool Support

The SX family of FPGAs is fully supported by both the Actel Libero® Integrated Design Environment (IDE) and Designer FPGA Development software. Actel Libero IDE is a design management environment, seamlessly integrating design tools while guiding the user through the design flow, managing all design and log files, and passing necessary design data among tools. Libero IDE allows users to integrate both schematic and HDL synthesis into a single flow and verify the entire design in a single environment. Libero IDE includes Synplify® for Actel from Synplicity®, ViewDraw® for Actel from Mentor Graphics®, Model*Sim*® HDL Simulator from Mentor Graphics, WaveFormer Lite™ from SynaptiCAD™, and Designer software from Actel. Refer to the Libero IDE flow diagram (located on the Actel website) for more information.

Actel Designer software is a place-and-route tool and provides a comprehensive suite of backend support tools for FPGA development. The Designer software includes timing-driven place-and-route, and a world-class integrated static timing analyzer and constraints editor. With the Designer software, a user can select and lock package pins while only minimally impacting the results of place-and-route. Additionally, the back-annotation flow is compatible with all the major simulators, and the simulation results can be cross-probed with Silicon Explorer II, Actel integrated verification and logic analysis tool. Another tool included in the Designer software is the SmartGen core generator, which easily creates popular and commonly used logic functions for implementation into your schematic or HDL design. Actel Designer software is compatible with the most popular FPGA design entry and verification tools from companies such as Mentor Graphics, Synplicity, Synopsys®, and Cadence® Design Systems. The Designer software is available for both the Windows® and UNIX® operating systems.

Probe Circuit Control Pins

The Silicon Explorer II tool uses the boundary scan ports (TDI, TCK, TMS, and TDO) to select the desired nets for verification. The selected internal nets are assigned to the PRA/PRB pins for observation. Figure 1-8 on page 1-7 illustrates the interconnection between Silicon Explorer II and the FPGA to perform in-circuit verification.

Design Considerations

The TDI, TCK, TDO, PRA, and PRB pins should not be used as input or bidirectional ports. Because these pins are active during probing, critical signals input through these pins are not available while probing. In addition, the Security Fuse should not be programmed because doing so disables the Probe Circuitry.

PCI Compliance for the SX Family

The SX family supports 3.3 V and 5.0 V PCI and is compliant with the PCI Local Bus Specification Rev. 2.1.

Table 1-6 • **A54SX16P DC Specifications (5.0 V PCI Operation)**

Symbol	**Parameter**	**Condition**	**Min.**	**Max.**	**Units**
V_{CCA}	Supply Voltage for Array		3.0	3.6	V
V_{CCR}	Supply Voltage required for Internal Biasing		4.75	5.25	V
V_{CCI}	Supply Voltage for I/Os		4.75	5.25	V
V_{IH}	Input High Voltage[1]		2.0	V_{CC} + 0.5	V
V_{IL}	Input Low Voltage[1]		–0.5	0.8	V
I_{IH}	Input High Leakage Current	V_{IN} = 2.7		70	µA
I_{IL}	Input Low Leakage Current	V_{IN} = 0.5		–70	µA
V_{OH}	Output High Voltage	I_{OUT} = –2 mA	2.4		V
V_{OL}	Output Low Voltage[2]	I_{OUT} = 3 mA, 6 mA		0.55	V
C_{IN}	Input Pin Capacitance[3]			10	pF
C_{CLK}	CLK Pin Capacitance		5	12	pF
C_{IDSEL}	IDSEL Pin Capacitance[4]			8	pF

Notes:

1. *Input leakage currents include hi-Z output leakage for all bidirectional buffers with tristate outputs.*
2. *Signals without pull-up resistors must have 3 mA low output current. Signals requiring pull-up must have 6 mA; the latter include, FRAME#, IRDY#, TRDY#, DEVSEL#, STOP#, SERR#, PERR#, LOCK#, and, when used, AD[63::32], C/BE[7::4]#, PAR64, REQ64#, and ACK64#.*
3. *Absolute maximum pin capacitance for a PCI input is 10 pF (except for CLK).*
4. *Lower capacitance on this input-only pin allows for non-resistive coupling to AD[xx].*

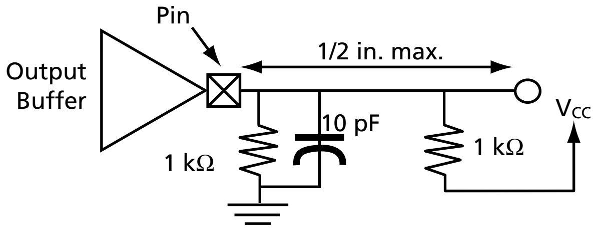
A54SX16P AC Specifications (3.3 V PCI Operation)

Table 1-9 • **A54SX16P AC Specifications (3.3 V PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \le 0.3V_{CC}$[1]			mA
		$0.3V_{CC} \le V_{OUT} < 0.9V_{CC}$[1]	$-12V_{CC}$		mA
		$0.7V_{CC} < V_{OUT} < V_{CC}$[1, 2]	$-17.1 + (V_{CC} - V_{OUT})$	EQ 1-3 on page 1-14	
	(Test Point)	$V_{OUT} = 0.7V_{CC}$[2]		$-32V_{CC}$	mA
$I_{OL(AC)}$	Switching Current High	$V_{CC} > V_{OUT} \ge 0.6V_{CC}$[1]			mA
		$0.6V_{CC} > V_{OUT} > 0.1V_{CC}$[1]	$16V_{CC}$		mA
		$0.18V_{CC} > V_{OUT} > 0$[1, 2]	$26.7V_{OUT}$	EQ 1-4 on page 1-14	mA
	(Test Point)	$V_{OUT} = 0.18V_{CC}$[2]		$38V_{CC}$	
I_{CL}	Low Clamp Current	$-3 < V_{IN} \le -1$	$-25 + (V_{IN} + 1)/0.015$		mA
I_{CH}	High Clamp Current	$-3 < V_{IN} \le -1$	$25 + (V_{IN} - V_{OUT} - 1)/0.015$		mA
$slew_R$	Output Rise Slew Rate[3]	$0.2V_{CC}$ to $0.6V_{CC}$ load	1	4	V/ns
$slew_F$	Output Fall Slew Rate[3]	$0.6V_{CC}$ to $0.2V_{CC}$ load	1	4	V/ns

Notes:

1. *Refer to the V/I curves in Figure 1-10 on page 1-14. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half size output drivers may be used on these signals. This specification does not apply to CLK and RST# which are system outputs. "Switching Current High" specification are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD# which are open drain outputs.*
2. *Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (C and D) are provided with the respective diagrams in Figure 1-10 on page 1-14. The equation defined maxima should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.*
3. *This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per the latest revision of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is required (the maximum is no longer simply a guideline). Rise slew rate does not apply to open drain outputs.*



Evaluating Power in SX Devices

A critical element of system reliability is the ability of electronic devices to safely dissipate the heat generated during operation. The thermal characteristics of a circuit depend on the device and package used, the operating temperature, the operating current, and the system's ability to dissipate heat.

You should complete a power evaluation early in the design process to help identify potential heat-related problems in the system and to prevent the system from exceeding the device's maximum allowed junction temperature.

The actual power dissipated by most applications is significantly lower than the power the package can dissipate. However, a thermal analysis should be performed for all projects. To perform a power evaluation, follow these steps:

1. Estimate the power consumption of the application.
2. Calculate the maximum power allowed for the device and package.
3. Compare the estimated power and maximum power values.

Estimating Power Consumption

The total power dissipation for the SX family is the sum of the DC power dissipation and the AC power dissipation. Use EQ 1-5 to calculate the estimated power consumption of your application.

$$P_{Total} = P_{DC} + P_{AC}$$

EQ 1-5

DC Power Dissipation

The power due to standby current is typically a small component of the overall power. The Standby power is shown in Table 1-12 for commercial, worst-case conditions (70°C).

Table 1-12 • **Standby Power**

I_{CC}	V_{CC}	Power
4 mA	3.6 V	14.4 mW

The DC power dissipation is defined in EQ 1-6.

$$P_{DC} = (I_{standby}) \times V_{CCA} + (I_{standby}) \times V_{CCR} + (I_{standby}) \times V_{CCI} + xV_{OL} \times I_{OL} + y(V_{CCI} - V_{OH}) \times V_{OH}$$

EQ 1-6

AC Power Dissipation

The power dissipation of the SX Family is usually dominated by the dynamic power dissipation. Dynamic power dissipation is a function of frequency, equivalent capacitance, and power supply voltage. The AC power dissipation is defined in EQ 1-7 and EQ 1-8.

$$P_{AC} = P_{Module} + P_{RCLKA\ Net} + P_{RCLKB\ Net} + P_{HCLK\ Net} + P_{Output\ Buffer} + P_{Input\ Buffer}$$

EQ 1-7

$$\begin{aligned}P_{AC} = V_{CCA}^2 \times [&(m \times C_{EQM} \times f_m)_{Module} + \\ &(n \times C_{EQI} \times f_n)_{Input\ Buffer} + (p \times (C_{EQO} + C_L) \times f_p)_{Output\ Buffer} + \\ &(0.5 \times (q_1 \times C_{EQCR} \times f_{q1}) + (r_1 \times f_{q1}))_{RCLKA} + \\ &(0.5 \times (q2 \times CEQCR \times f_{q2}) + (r2 \times f_{q2}))RCLKB + \\ &(0.5 \times (s_1 \times C_{EQHV} \times f_{s1}) + (C_{EQHF} \times f_{s1}))_{HCLK}]\end{aligned}$$

EQ 1-8

Definition of Terms Used in Formula

m = Number of logic modules switching at f_m

n = Number of input buffers switching at f_n

p = Number of output buffers switching at f_p

q_1 = Number of clock loads on the first routed array clock

q_2 = Number of clock loads on the second routed array clock

x = Number of I/Os at logic low

y = Number of I/Os at logic high

r_1 = Fixed capacitance due to first routed array clock

r_2 = Fixed capacitance due to second routed array clock

s_1 = Number of clock loads on the dedicated array clock

C_{EQM} = Equivalent capacitance of logic modules in pF

C_{EQI} = Equivalent capacitance of input buffers in pF

C_{EQO} = Equivalent capacitance of output buffers in pF

C_{EQCR} = Equivalent capacitance of routed array clock in pF

C_{EQHV} = Variable capacitance of dedicated array clock

C_{EQHF} = Fixed capacitance of dedicated array clock

C_L = Output lead capacitance in pF

f_m = Average logic module switching rate in MHz

f_n = Average input buffer switching rate in MHz

f_p = Average output buffer switching rate in MHz

f_{q1} = Average first routed array clock rate in MHz

f_{q2} = Average second routed array clock rate in MHz

f_{s1} = Average dedicated array clock rate in MHz

Table 1-13 shows capacitance values for various devices.

Table 1-13 • **Capacitance Values for Devices**

	A54SX08	A54SX16	A54SX16P	A54SX32
C_{EQM} (pF)	4.0	4.0	4.0	4.0
C_{EQI} (pF)	3.4	3.4	3.4	3.4
C_{EQO} (pF)	4.7	4.7	4.7	4.7
C_{EQCR} (pF)	1.6	1.6	1.6	1.6
C_{EQHV}	0.615	0.615	0.615	0.615
C_{EQHF}	60	96	96	140
r_1 (pF)	87	138	138	171
r_2 (pF)	87	138	138	171

Guidelines for Calculating Power Consumption

The power consumption guidelines are meant to represent worst-case scenarios so that they can be generally used to predict the upper limits of power dissipation. These guidelines are shown in Table 1-14.

Sample Power Calculation

One of the designs used to characterize the SX family was a 528 bit serial-in, serial-out shift register. The design utilized 100 percent of the dedicated flip-flops of an A54SX16P device. A pattern of 0101… was clocked into the device at frequencies ranging from 1 MHz to 200 MHz. Shifting in a series of 0101… caused 50 percent of the flip-flops to toggle from low to high at every clock cycle.

Table 1-14 • **Power Consumption Guidelines**

Description	Power Consumption Guideline
Logic Modules (m)	20% of modules
Inputs Switching (n)	# inputs/4
Outputs Switching (p)	# outputs/4
First Routed Array Clock Loads (q_1)	20% of register cells
Second Routed Array Clock Loads (q_2)	20% of register cells
Load Capacitance (C_L)	35 pF
Average Logic Module Switching Rate (f_m)	f/10
Average Input Switching Rate (f_n)	f/5
Average Output Switching Rate (f_p)	f/10
Average First Routed Array Clock Rate (f_{q1})	f/2
Average Second Routed Array Clock Rate (f_{q2})	f/2
Average Dedicated Array Clock Rate (f_{s1})	f
Dedicated Clock Array Clock Loads (s_1)	20% of regular modules

Follow the steps below to estimate power consumption. The values provided for the sample calculation below are for the shift register design above. This method for estimating power consumption is conservative and the actual power consumption of your design may be less than the estimated power consumption.

The total power dissipation for the SX family is the sum of the AC power dissipation and the DC power dissipation.

$$P_{Total} = P_{AC}\text{ (dynamic power)} + P_{DC}\text{ (static power)}$$

EQ 1-9

AC Power Dissipation

$$P_{AC} = P_{Module} + P_{RCLKA\ Net} + P_{RCLKB\ Net} + P_{HCLK\ Net} + P_{Output\ Buffer} + P_{Input\ Buffer}$$

EQ 1-10

$$\begin{aligned}P_{AC} = V_{CCA}^2 \times [&(m \times C_{EQM} \times f_m)_{Module} + \\ &(n \times C_{EQI} \times f_n)_{Input\ Buffer} + (p \times (C_{EQO} + C_L) \times f_p)_{Output\ Buffer} + \\ &(0.5\ (q_1 \times C_{EQCR} \times f_{q1}) + (r_1 \times f_{q1}))_{RCLKA} + \\ &(0.5\ (q_2 \times C_{EQCR} \times f_{q2}) + (r_2 \times f_{q2}))_{RCLKB} + \\ &(0.5\ (s_1 \times C_{EQHV} \times f_{s1}) + (C_{EQHF} \times f_{s1}))_{HCLK}]\end{aligned}$$

EQ 1-11

Table 1-15 • **Package Thermal Characteristics**

Package Type	**Pin Count**	θ_{jc}	θ_{ja} **Still Air**	θ_{ja} **300 ft/min.**	**Units**
Plastic Leaded Chip Carrier (PLCC)	84	12	32	22	°C/W
Thin Quad Flat Pack (TQFP)	144	11	32	24	°C/W
Thin Quad Flat Pack (TQFP)	176	11	28	21	°C/W
Very Thin Quad Flatpack (VQFP)	100	10	38	32	°C/W
Plastic Quad Flat Pack (PQFP) without Heat Spreader	208	8	30	23	°C/W
Plastic Quad Flat Pack (PQFP) with Heat Spreader	208	3.8	20	17	°C/W
Plastic Ball Grid Array (PBGA)	272	3	20	14.5	°C/W
Plastic Ball Grid Array (PBGA)	313	3	23	17	°C/W
Plastic Ball Grid Array (PBGA)	329	3	18	13.5	°C/W
Fine Pitch Ball Grid Array (FBGA)	144	3.8	38.8	26.7	°C/W

Note: *SX08 does not have a heat spreader.*

Table 1-16 • **Temperature and Voltage Derating Factors***

V_{CCA}	**Junction Temperature**						
	–55	**–40**	**0**	**25**	**70**	**85**	**125**
3.0	0.75	0.78	0.87	0.89	1.00	1.04	1.16
3.3	0.70	0.73	0.82	0.83	0.93	0.97	1.08
3.6	0.66	0.69	0.77	0.78	0.87	0.92	1.02

Note: **Normalized to worst-case commercial, T_J = 70°C, V_{CCA} = 3.0 V*

A54SX16 Timing Characteristics

Table 1-18 • **A54SX16 Timing Characteristics**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

		'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		
Parameter	**Description**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
C-Cell Propagation Delays[1]										
t_{PD}	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted Routing Delays[2]										
t_{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t_{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{RD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{RD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timing										
t_{RCO}	Sequential Clock-to-Q		0.8		1.1		1.2		1.4	ns
t_{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t_{PRESET}	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t_{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t_{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t_{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t_{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted Input Routing Delays[2]										
t_{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{IRD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{IRD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Notes:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH}. For t_{ENZL} and t_{ENZH}, the loading is 5 pF.*

Table 1-18 • **A54SX16 Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

		'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		
Parameter	**Description**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
Dedicated (Hardwired) Array Clock Network										
t_{HCKH}	Input LOW to HIGH (pad to R-Cell input)		1.2		1.4		1.5		1.8	ns
t_{HCKL}	Input HIGH to LOW (pad to R-Cell input)		1.2		1.4		1.6		1.9	ns
t_{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t_{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t_{HCKSW}	Maximum Skew		0.2		0.2		0.3		0.3	ns
t_{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f_{HMAX}	Maximum Frequency		350		320		280		240	MHz
Routed Array Clock Networks										
t_{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)		1.6		1.8		2.1		2.5	ns
t_{RCKL}	Input HIGH to LOW (light load) (pad to R-Cell input)		1.8		2.0		2.3		2.7	ns
t_{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)		1.8		2.1		2.5		2.8	ns
t_{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)		1.8		2.1		2.4		2.8	ns
t_{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t_{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t_{RCKSW}	Maximum Skew (light load)		0.5		0.5		0.5		0.7	ns
t_{RCKSW}	Maximum Skew (50% load)		0.5		0.6		0.7		0.8	ns
t_{RCKSW}	Maximum Skew (100% load)		0.5		0.6		0.7		0.8	ns
TTL Output Module Timing[3]										
t_{DLH}	Data-to-Pad LOW to HIGH		1.6		1.9		2.1		2.5	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.6		1.9		2.1		2.5	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.1		2.4		2.8		3.2	ns
t_{ENZH}	Enable-to-Pad, Z to H		2.3		2.7		3.1		3.6	ns
t_{ENLZ}	Enable-to-Pad, L to Z		1.4		1.7		1.9		2.2	ns
t_{ENHZ}	Enable-to-Pad, H to Z		1.3		1.5		1.7		2.0	ns

Notes:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH}. For t_{ENZL} and t_{ENZH}, the loading is 5 pF.*

A54SX16P Timing Characteristics

Table 1-19 • **A54SX16P Timing Characteristics**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

		'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		
Parameter	**Description**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
C-Cell Propagation Delays[1]										
t_{PD}	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted Routing Delays[2]										
t_{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t_{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{RD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{RD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timing										
t_{RCO}	Sequential Clock-to-Q		0.9		1.1		1.3		1.4	ns
t_{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t_{PRESET}	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t_{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t_{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t_{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t_{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted Input Routing Delays[2]										
t_{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{IRD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{IRD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Note:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 10 pF loading.*

Table 1-19 • **A54SX16P Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

		'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		
Parameter	**Description**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
TTL/PCI Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		1.5		1.7		2.0		2.3	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.9		2.2		2.4		2.9	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.3		2.6		3.0		3.5	ns
t_{ENZH}	Enable-to-Pad, Z to H		1.5		1.7		1.9		2.3	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.7		3.1		3.5		4.1	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.9		3.3		3.7		4.4	ns
PCI Output Module Timing[3]										
t_{DLH}	Data-to-Pad LOW to HIGH		1.8		2.0		2.3		2.7	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.7		2.0		2.2		2.6	ns
t_{ENZL}	Enable-to-Pad, Z to L		0.8		1.0		1.1		1.3	ns
t_{ENZH}	Enable-to-Pad, Z to H		1.2		1.2		1.5		1.8	ns
t_{ENLZ}	Enable-to-Pad, L to Z		1.0		1.1		1.3		1.5	ns
t_{ENHZ}	Enable-to-Pad, H to Z		1.1		1.3		1.5		1.7	ns
TTL Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		2.1		2.5		2.8		3.3	ns
t_{DHL}	Data-to-Pad HIGH to LOW		2.0		2.3		2.6		3.1	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.5		2.9		3.2		3.8	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.0		3.5		3.9		4.6	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.3		2.7		3.1		3.6	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.9		3.3		3.7		4.4	ns

Note:

1. *For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 10 pF loading.*

208-Pin PQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
145	V_{CCA}	V_{CCA}	V_{CCA}
146	GND	GND	GND
147	I/O	I/O	I/O
148	V_{CCI}	V_{CCI}	V_{CCI}
149	I/O	I/O	I/O
150	I/O	I/O	I/O
151	I/O	I/O	I/O
152	I/O	I/O	I/O
153	I/O	I/O	I/O
154	I/O	I/O	I/O
155	NC	I/O	I/O
156	NC	I/O	I/O
157	GND	GND	GND
158	I/O	I/O	I/O
159	I/O	I/O	I/O
160	I/O	I/O	I/O
161	I/O	I/O	I/O
162	I/O	I/O	I/O
163	I/O	I/O	I/O
164	V_{CCI}	V_{CCI}	V_{CCI}
165	I/O	I/O	I/O
166	I/O	I/O	I/O
167	NC	I/O	I/O
168	I/O	I/O	I/O
169	I/O	I/O	I/O
170	NC	I/O	I/O
171	I/O	I/O	I/O
172	I/O	I/O	I/O
173	NC	I/O	I/O
174	I/O	I/O	I/O
175	I/O	I/O	I/O
176	NC	I/O	I/O
177	I/O	I/O	I/O
178	I/O	I/O	I/O
179	I/O	I/O	I/O
180	CLKA	CLKA	CLKA

208-Pin PQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
181	CLKB	CLKB	CLKB
182	V_{CCR}	V_{CCR}	V_{CCR}
183	GND	GND	GND
184	V_{CCA}	V_{CCA}	V_{CCA}
185	GND	GND	GND
186	PRA, I/O	PRA, I/O	PRA, I/O
187	I/O	I/O	I/O
188	I/O	I/O	I/O
189	NC	I/O	I/O
190	I/O	I/O	I/O
191	I/O	I/O	I/O
192	NC	I/O	I/O
193	I/O	I/O	I/O
194	I/O	I/O	I/O
195	NC	I/O	I/O
196	I/O	I/O	I/O
197	I/O	I/O	I/O
198	NC	I/O	I/O
199	I/O	I/O	I/O
200	I/O	I/O	I/O
201	V_{CCI}	V_{CCI}	V_{CCI}
202	NC	I/O	I/O
203	NC	I/O	I/O
204	I/O	I/O	I/O
205	NC	I/O	I/O
206	I/O	I/O	I/O
207	I/O	I/O	I/O
208	TCK, I/O	TCK, I/O	TCK, I/O

Note: ** Note that Pin 65 in the A54SX32—PQ208 is a no connect (NC).*

176-Pin TQFP

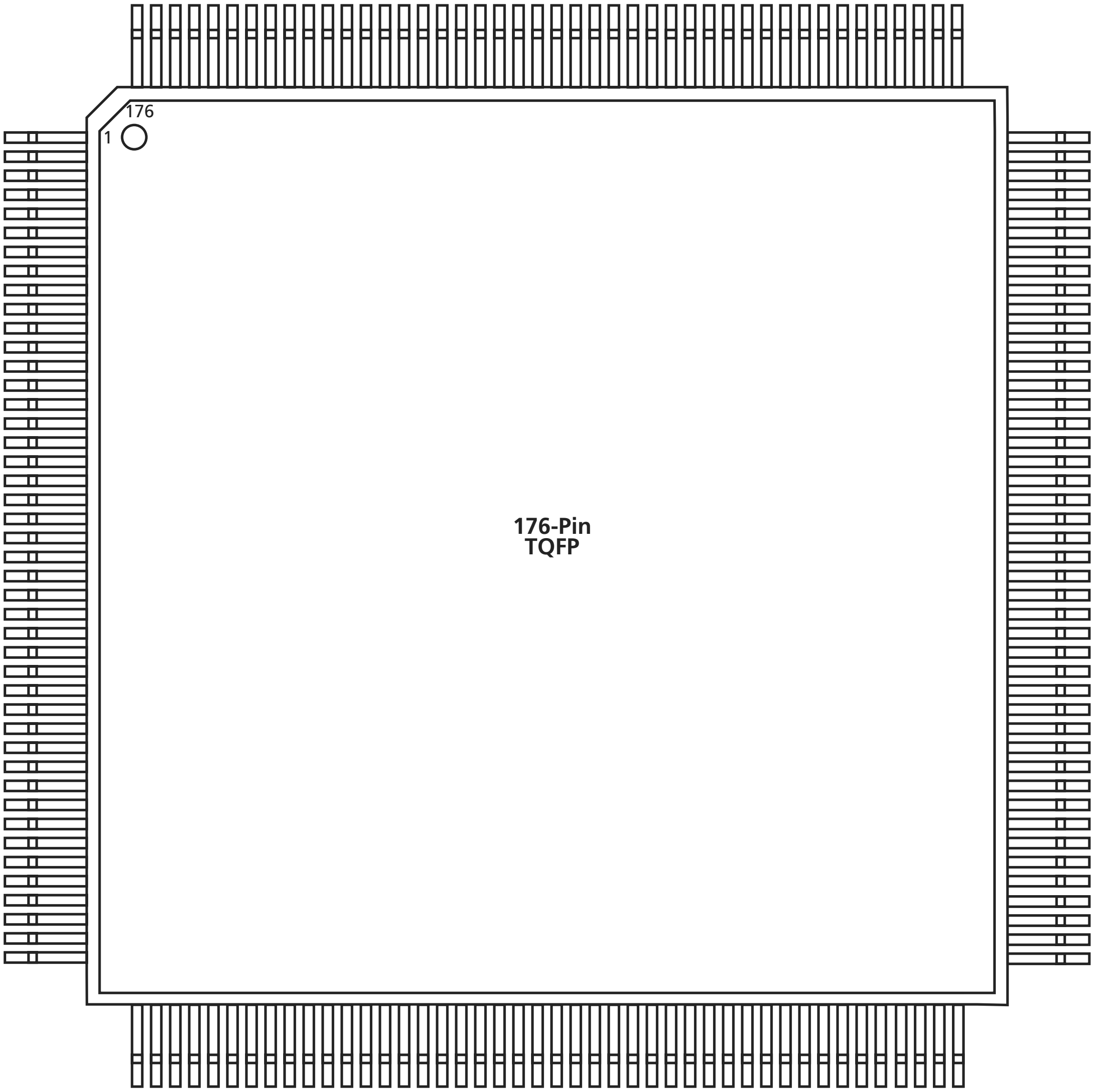


Figure 2-4 • **176-Pin TQFP (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

176-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	NC	I/O	I/O
4	I/O	I/O	I/O
5	I/O	I/O	I/O
6	I/O	I/O	I/O
7	I/O	I/O	I/O
8	I/O	I/O	I/O
9	I/O	I/O	I/O
10	TMS	TMS	TMS
11	V_{CCI}	V_{CCI}	V_{CCI}
12	NC	I/O	I/O
13	I/O	I/O	I/O
14	I/O	I/O	I/O
15	I/O	I/O	I/O
16	I/O	I/O	I/O
17	I/O	I/O	I/O
18	I/O	I/O	I/O
19	I/O	I/O	I/O
20	I/O	I/O	I/O
21	GND	GND	GND
22	V_{CCA}	V_{CCA}	V_{CCA}
23	GND	GND	GND
24	I/O	I/O	I/O
25	I/O	I/O	I/O
26	I/O	I/O	I/O
27	I/O	I/O	I/O
28	I/O	I/O	I/O
29	I/O	I/O	I/O
30	I/O	I/O	I/O
31	I/O	I/O	I/O
32	V_{CCI}	V_{CCI}	V_{CCI}
33	V_{CCA}	V_{CCA}	V_{CCA}
34	I/O	I/O	I/O
35	I/O	I/O	I/O
36	I/O	I/O	I/O
37	I/O	I/O	I/O
38	I/O	I/O	I/O
39	I/O	I/O	I/O
40	NC	I/O	I/O
41	I/O	I/O	I/O
42	NC	I/O	I/O
43	I/O	I/O	I/O
44	GND	GND	GND
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	I/O	I/O	I/O
49	I/O	I/O	I/O
50	I/O	I/O	I/O
51	I/O	I/O	I/O
52	V_{CCI}	V_{CCI}	V_{CCI}
53	I/O	I/O	I/O
54	NC	I/O	I/O
55	I/O	I/O	I/O
56	I/O	I/O	I/O
57	NC	I/O	I/O
58	I/O	I/O	I/O
59	I/O	I/O	I/O
60	I/O	I/O	I/O
61	I/O	I/O	I/O
62	I/O	I/O	I/O
63	I/O	I/O	I/O
64	PRB, I/O	PRB, I/O	PRB, I/O
65	GND	GND	GND
66	V_{CCA}	V_{CCA}	V_{CCA}
67	V_{CCR}	V_{CCR}	V_{CCR}
68	I/O	I/O	I/O

329-Pin PBGA

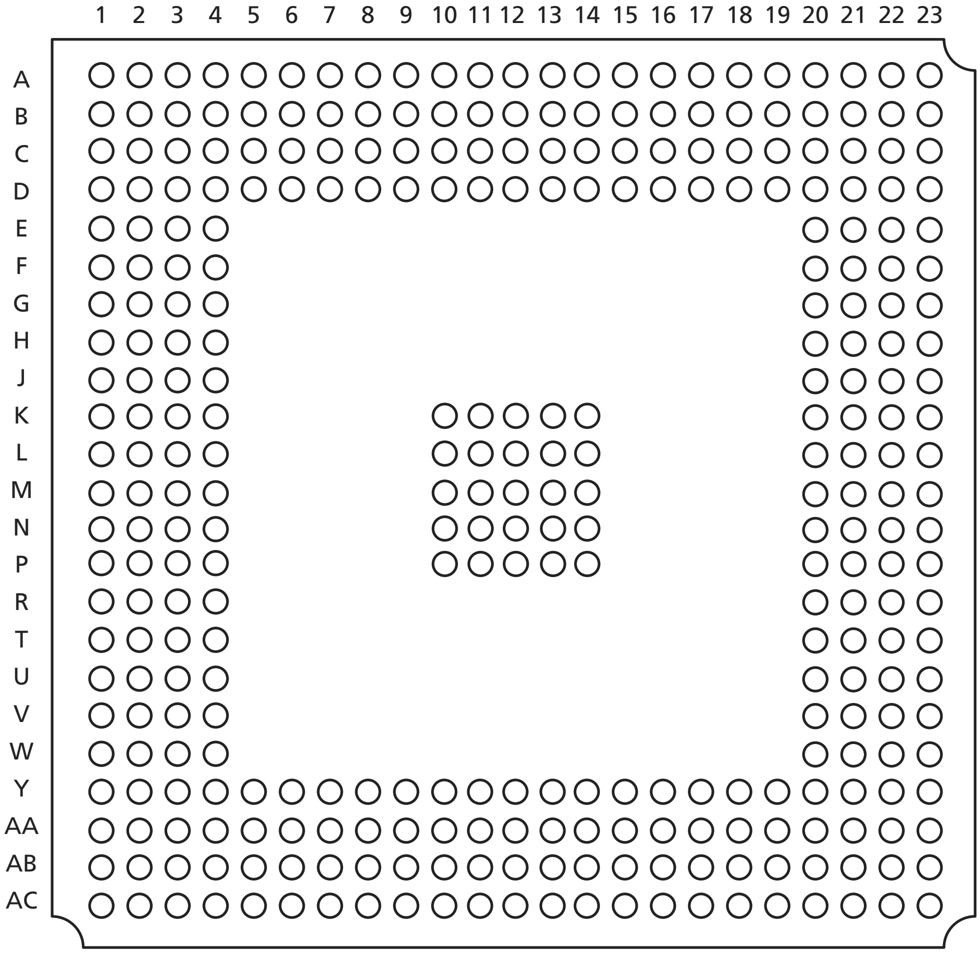


Figure 2-7 • **329-Pin PBGA (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

329-Pin PBGA	
Pin Number	A54SX32 Function
D3	I/O
D4	TCK, I/O
D5	I/O
D6	I/O
D7	I/O
D8	I/O
D9	I/O
D10	I/O
D11	V_{CCA}
D12	V_{CCR}
D13	I/O
D14	I/O
D15	I/O
D16	I/O
D17	I/O
D18	I/O
D19	I/O
D20	I/O
D21	I/O
D22	I/O
D23	I/O
E1	V_{CCI}
E2	I/O
E3	I/O
E4	I/O
E20	I/O
E21	I/O
E22	I/O
E23	I/O
F1	I/O
F2	TMS
F3	I/O
F4	I/O
F20	I/O
F21	I/O
F22	I/O
F23	I/O
G1	I/O
G2	I/O
G3	I/O
G4	I/O
G20	I/O
G21	I/O
G22	I/O
G23	GND
H1	I/O
H2	I/O
H3	I/O
H4	I/O
H20	V_{CCA}
H21	I/O
H22	I/O
H23	I/O
J1	NC
J2	I/O
J3	I/O
J4	I/O
J20	I/O
J21	I/O
J22	I/O
J23	I/O
K1	I/O
K2	I/O
K3	I/O
K4	I/O
K10	GND
K11	GND
K12	GND
K13	GND
K14	GND
K20	I/O
K21	I/O
K22	I/O
K23	I/O
L1	I/O
L2	I/O
L3	I/O
L4	V_{CCR}
L10	GND
L11	GND
L12	GND
L13	GND
L14	GND
L20	V_{CCR}
L21	I/O
L22	I/O
L23	NC
M1	I/O
M2	I/O
M3	I/O
M4	V_{CCA}
M10	GND
M11	GND
M12	GND
M13	GND
M14	GND
M20	V_{CCA}
M21	I/O
M22	I/O
M23	V_{CCI}
N1	I/O
N2	I/O
N3	I/O
N4	I/O
N10	GND
N11	GND
N12	GND
N13	GND
N14	GND
N20	NC
N21	I/O
N22	I/O
N23	I/O
P1	I/O
P2	I/O
P3	I/O
P4	I/O
P10	GND
P11	GND
P12	GND
P13	GND
P14	GND
P20	I/O
P21	I/O
P22	I/O
P23	I/O
R1	I/O
R2	I/O
R3	I/O
R4	I/O
R20	I/O
R21	I/O
R22	I/O
R23	I/O
T1	I/O
T2	I/O
T3	I/O
T4	I/O
T20	I/O
T21	I/O

144-Pin FBGA

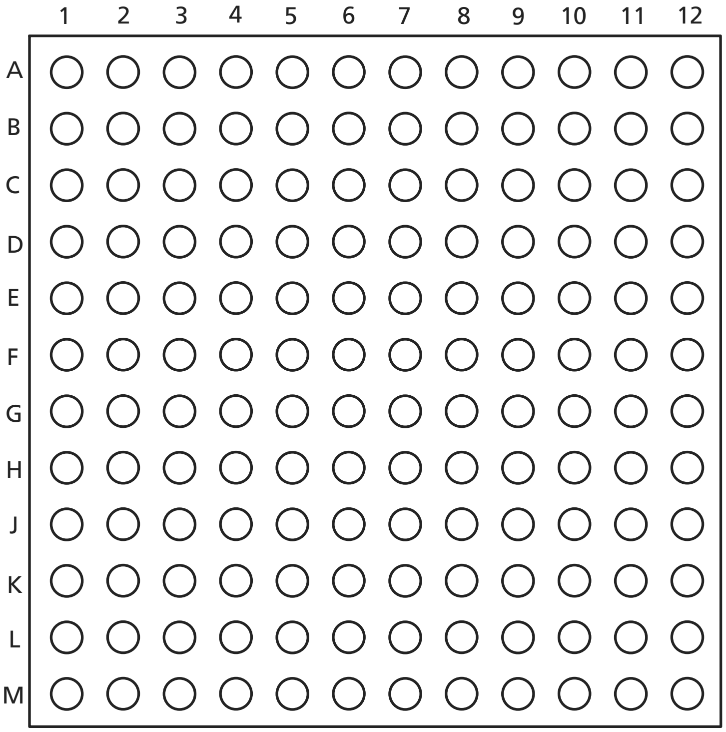


Figure 2-8 • **144-Pin FBGA (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

Datasheet Information

List of Changes

The following table lists critical changes that were made in the current version of the document.

Previous Version	Changes in Current Version (v3.2)	Page
v3.1 (June 2003)	The "Ordering Information" was updated to include RoHS information.	1-ii
	The Product Plan was removed since all products have been released.	N/A
	Information concerning the TRST pin in the "Probe Circuit Control Pins" section was removed.	1-6
	The "Dedicated Test Mode" section is new.	1-6
	The "Programming" section is new.	1-7
	A note was added to the "Power-Up Sequencing" table.	1-15
	A note was added to the "Power-Down Sequencing" table. The 3.3 V comments were updated for the following devices: A54SX08, A54SX16, A54SX32.	1-15
	U11 and U13 were added to the "313-Pin PBGA" table.	2-17
v3.0.1	Storage temperature in Table 1-3 was updated.	1-7
	Table 1-1 was updated.	1-5

Datasheet Categories

In order to provide the latest information to designers, some datasheets are published before data has been fully characterized. Datasheets are designated as "Product Brief," "Advanced," "Production," and "Datasheet Supplement." The definitions of these categories are as follows:

Product Brief

The product brief is a summarized version of a datasheet (advanced or production) containing general product information. This brief gives an overview of specific device and family information.

Advanced

This datasheet version contains initial estimated information based on simulation, other products, devices, or speed grades. This information can be used as estimates, but not for production.

Unmarked (production)

This datasheet version contains information that is considered to be final.

Datasheet Supplement

The datasheet supplement gives specific device information for a derivative family that differs from the general family datasheet. The supplement is to be used in conjunction with the datasheet to obtain more detailed information and for specifications that do not differ between the two families.

International Traffic in Arms Regulations (ITAR) and Export Administration Regulations (EAR)

The products described in this datasheet are subject to the International Traffic in Arms Regulations (ITAR) or the Export Administration Regulations (EAR). They may require an approved export license prior to their export. An export can include a release or disclosure to a foreign national inside or outside the United States.