Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.
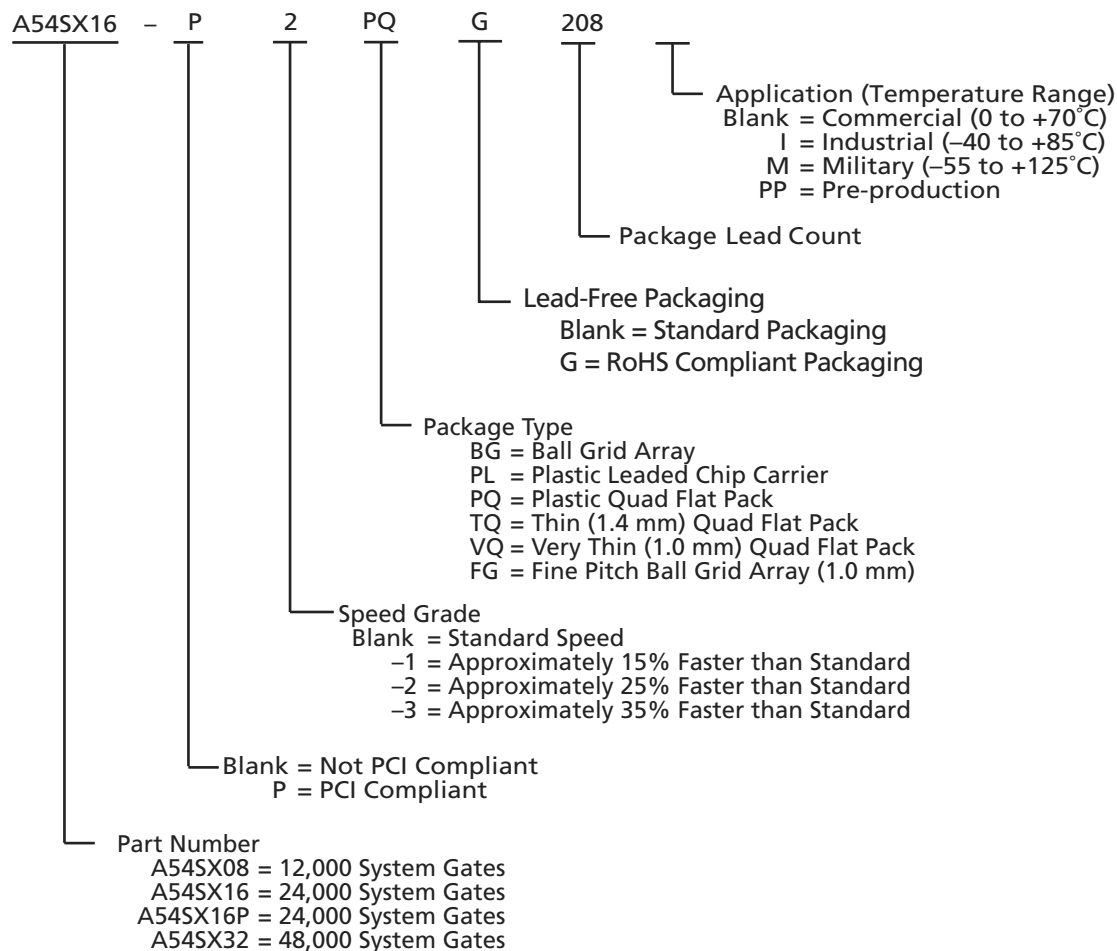
Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	768
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	81
Number of Gates	12000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	100-TQFP
Supplier Device Package	100-VQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx08-1vqg100

Ordering Information

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A54SX16  –  P   2   PQ   G   208
```

Application (Temperature Range)
Blank = Commercial (0 to +70°C)
I = Industrial (–40 to +85°C)
M = Military (–55 to +125°C)
PP = Pre-production

Package Lead Count

Lead-Free Packaging
Blank = Standard Packaging
G = RoHS Compliant Packaging

Package Type
BG = Ball Grid Array
PL = Plastic Leaded Chip Carrier
PQ = Plastic Quad Flat Pack
TQ = Thin (1.4 mm) Quad Flat Pack
VQ = Very Thin (1.0 mm) Quad Flat Pack
FG = Fine Pitch Ball Grid Array (1.0 mm)

Speed Grade
Blank = Standard Speed
–1 = Approximately 15% Faster than Standard
–2 = Approximately 25% Faster than Standard
–3 = Approximately 35% Faster than Standard

Blank = Not PCI Compliant
P = PCI Compliant

Part Number
A54SX08 = 12,000 System Gates
A54SX16 = 24,000 System Gates
A54SX16P = 24,000 System Gates
A54SX32 = 48,000 System Gates

Plastic Device Resources

Device	User I/Os (including clock buffers)							
	PLCC 84-Pin	VQFP 100-Pin	PQFP 208-Pin	TQFP 144-Pin	TQFP 176-Pin	PBGA 313-Pin	PBGA 329-Pin	FBGA 144-Pin
A54SX08	69	81	130	113	128	–	–	111
A54SX16	–	81	175	–	147	–	–	–
A54SX16P	–	81	175	113	147	–	–	–
A54SX32	–	–	174	113	147	249	249	–

Note: *Package Definitions (Consult your local Actel sales representative for product availability):*
PLCC = Plastic Leaded Chip Carrier
PQFP = Plastic Quad Flat Pack
TQFP = Thin Quad Flat Pack
VQFP = Very Thin Quad Flat Pack
PBGA = Plastic Ball Grid Array
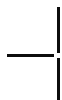*FBGA = Fine Pitch (1.0 mm) Ball Grid Array*

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The R-cell contains a flip-flop featuring asynchronous clear, asynchronous preset, and clock enable (using the S0 and S1 lines) control signals (Figure 1-2). The R-cell registers feature programmable clock polarity selectable on a register-by-register basis. This provides additional flexibility while allowing mapping of synthesized functions into the SX FPGA. The clock source for the R-cell can be chosen from either the hardwired clock or the routed clock.
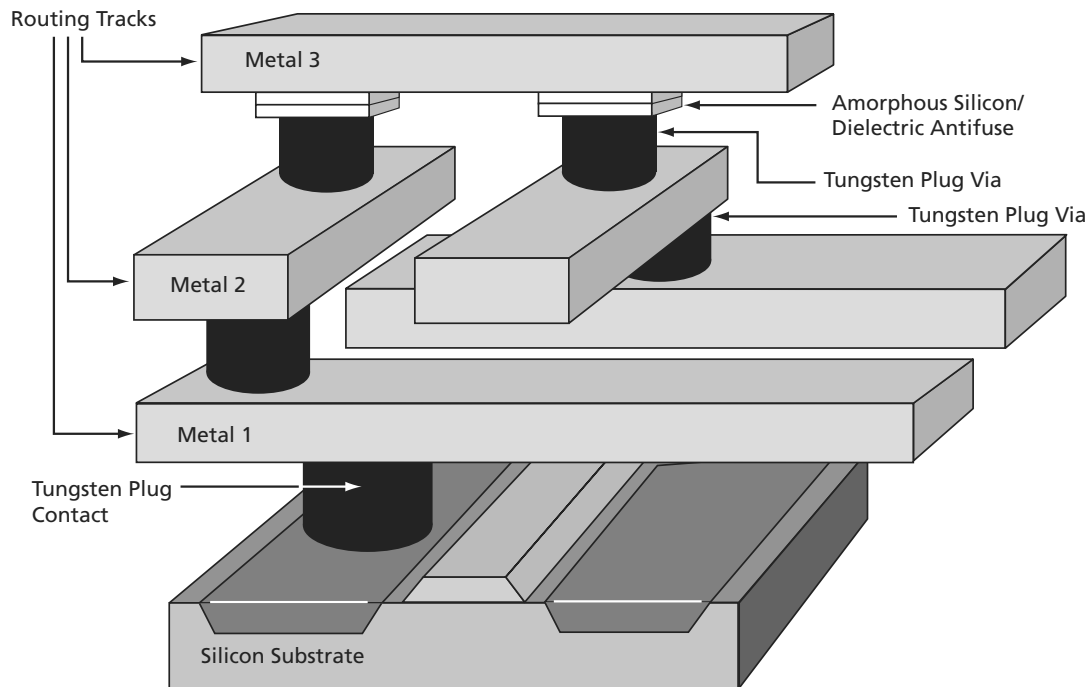

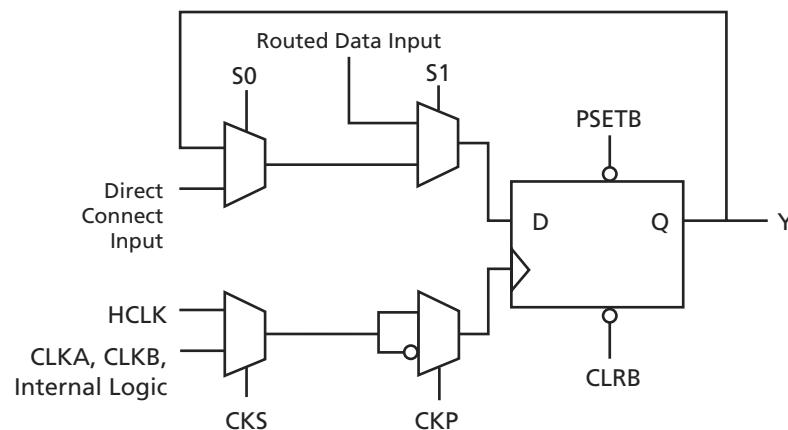*Figure 1-1* • **SX Family Interconnect Elements**

Figure 1-2 • **R-Cell**

The C-cell implements a range of combinatorial functions up to 5-inputs (Figure 1-3 on page 1-3). Inclusion of the DB input and its associated inverter function dramatically increases the number of combinatorial functions that can be implemented in a single module from 800 options in previous architectures to more than 4,000 in the SX architecture. An example of the improved flexibility enabled by the inversion capability is the ability to integrate a 3-input exclusive-OR function into a single C-cell. This facilitates construction of 9-bit parity-tree functions with 2 ns propagation delays. At the same time, the C-cell structure is extremely synthesis friendly, simplifying the overall design and reducing synthesis time.

Chip Architecture

The SX family chip architecture provides a unique approach to module organization and chip routing that delivers the best register/logic mix for a wide variety of new and emerging applications.

Module Organization

Actel has arranged all C-cell and R-cell logic modules into horizontal banks called *clusters*. There are two types of *clusters*: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells.

To increase design efficiency and device performance, Actel has further organized these modules into *SuperClusters* (Figure 1-4). SuperCluster 1 is a two-wide grouping of Type 1 clusters. SuperCluster 2 is a two-wide group containing one Type 1 cluster and one Type 2 cluster. SX devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flip-flops.
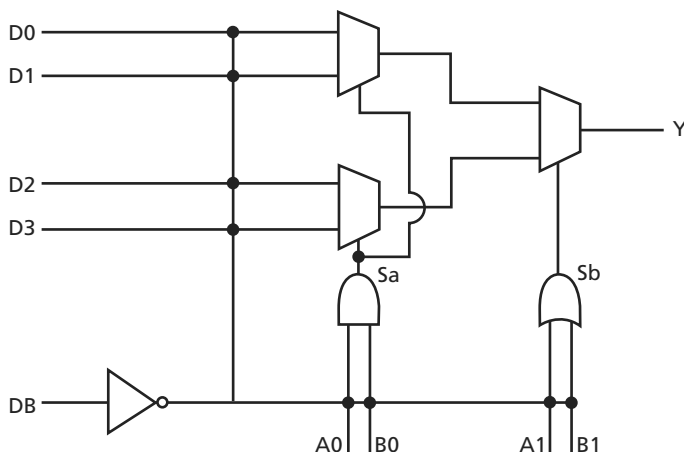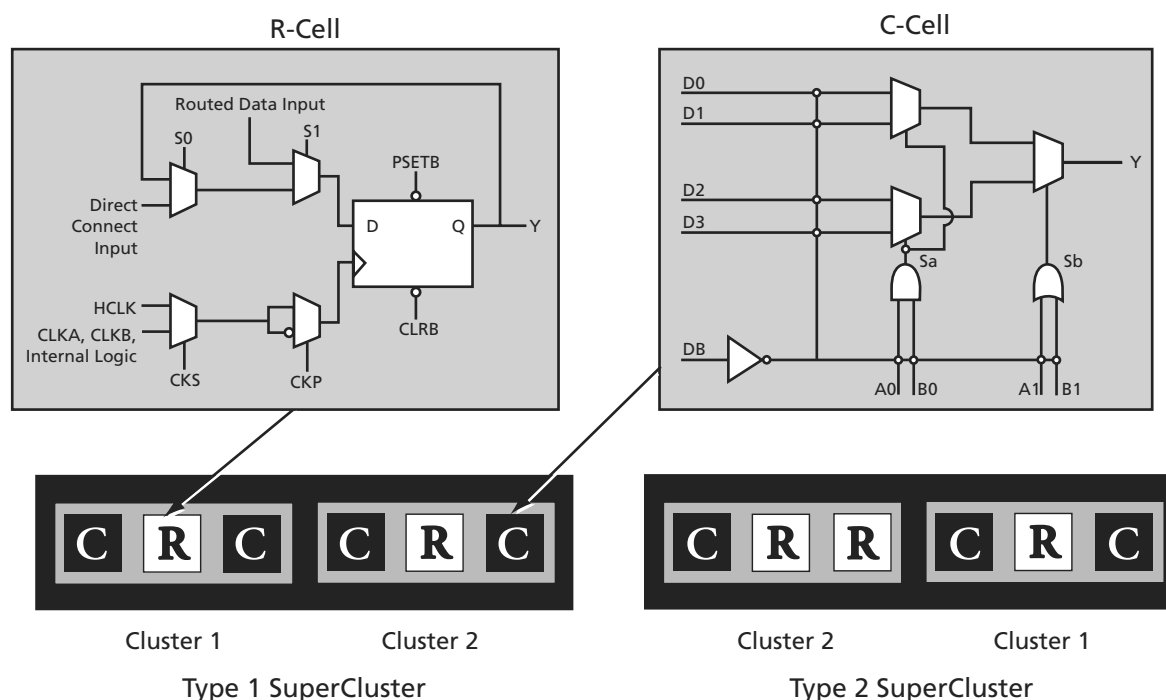

Figure 1-3 • **C-Cell**

Figure 1-4 • **Cluster Organization**

Routing Resources

Clusters and SuperClusters can be connected through the use of two innovative local routing resources called *FastConnect* and *DirectConnect,* which enable extremely fast and predictable interconnection of modules within clusters and SuperClusters (Figure 1-5 and Figure 1-6). This routing architecture also dramatically reduces the number of antifuses required to complete a circuit, ensuring the highest possible performance.

DirectConnect
• No antifuses
• 0.1 ns routing delay

FastConnect
• One antifuse
• 0.4 ns routing delay

Routing Segments
• Typically 2 antifuses
• Max. 5 antifuses
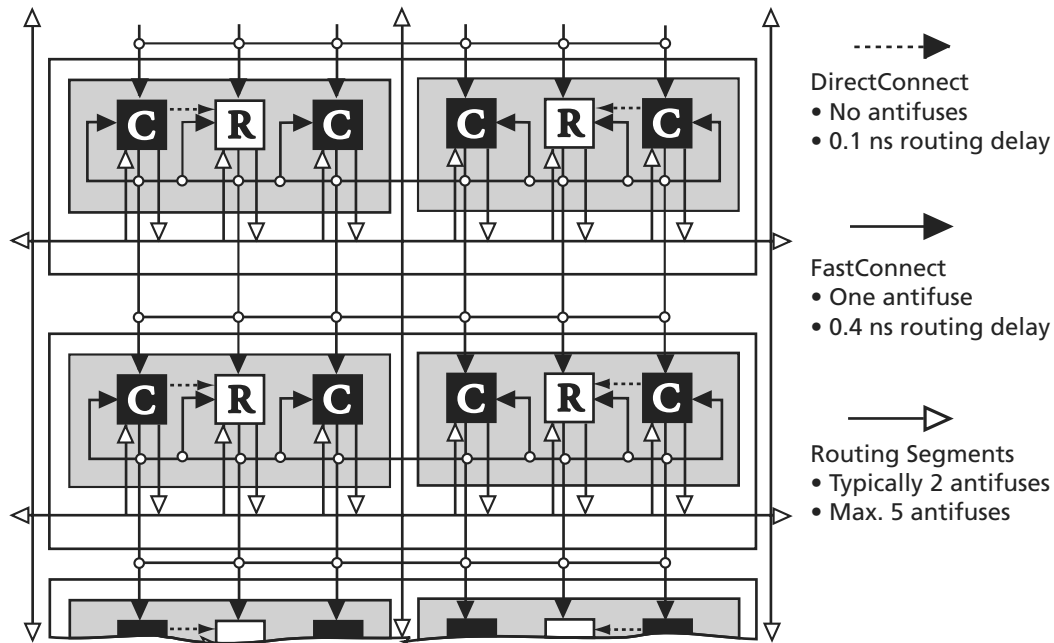
Figure 1-5 • **DirectConnect and FastConnect for Type 1 SuperClusters**

DirectConnect
• No antifuses
• 0.1 ns routing delay

FastConnect
• One antifuse
• 0.4 ns routing delay

Routing Segments
• Typically 2 antifuses
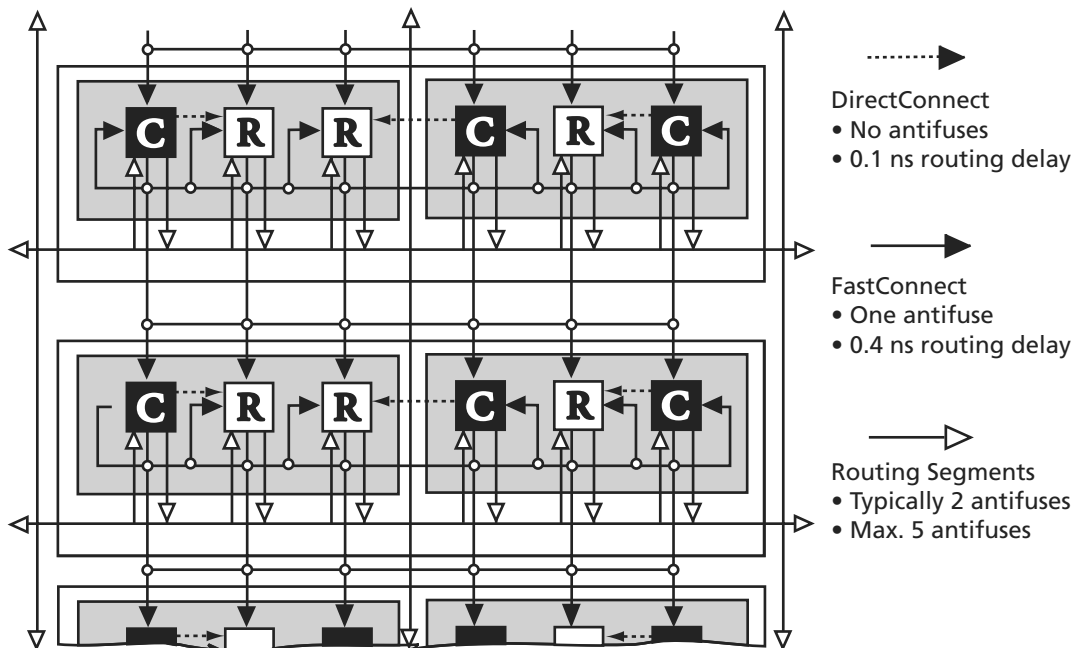• Max. 5 antifuses

Figure 1-6 • **DirectConnect and FastConnect for Type 2 SuperClusters**

DirectConnect is a horizontal routing resource that provides connections from a C-cell to its neighboring R-cell in a given SuperCluster. DirectConnect uses a hardwired signal path requiring no programmable interconnection to achieve its fast signal propagation time of less than 0.1 ns.

FastConnect enables horizontal routing between any two logic modules within a given SuperCluster and vertical routing with the SuperCluster immediately below it. Only one programmable connection is used in a FastConnect path, delivering maximum pin-to-pin propagation of 0.4 ns.

In addition to DirectConnect and FastConnect, the architecture makes use of two globally oriented routing resources known as segmented routing and high-drive routing. The Actel segmented routing structure provides a variety of track lengths for extremely fast routing between SuperClusters. The exact combination of track lengths and antifuses within each path is chosen by the 100 percent automatic place-and-route software to minimize signal propagation delays.

The Actel high-drive routing structure provides three clock networks. The first clock, called HCLK, is hardwired from the HCLK buffer to the clock select multiplexer (MUX) in each R-cell. This provides a fast propagation path for the clock signal, enabling the 3.7 ns clock-to-out (pin-to-pin) performance of the SX devices. The hardwired clock is tuned to provide clock skew as low as 0.25 ns. The remaining two clocks (CLKA, CLKB) are global clocks that can be sourced from external pins or from internal logic signals within the SX device.

Other Architectural Features

Technology

The Actel SX family is implemented on a high-voltage twin-well CMOS process using 0.35 μ design rules. The metal-to-metal antifuse is made up of a combination of amorphous silicon and dielectric material with barrier metals and has a programmed ("on" state) resistance of 25 Ω with a capacitance of 1.0 fF for low signal impedance.

Performance

The combination of architectural features described above enables SX devices to operate with internal clock frequencies exceeding 300 MHz, enabling very fast execution of even complex logic functions. Thus, the SX family is an optimal platform upon which to integrate the functionality previously contained in multiple CPLDs. In addition, designs that previously would have required a gate array to meet performance goals can now be integrated into an SX device with dramatic improvements in cost and time to market. Using timing-driven place-and-route tools, designers can achieve highly deterministic device performance. With SX devices, designers do not need to use complicated performance-enhancing design techniques such as the use of redundant logic to reduce fanout on critical nets or the instantiation of macros in HDL code to achieve high performance.

I/O Modules

Each I/O on an SX device can be configured as an input, an output, a tristate output, or a bidirectional pin.

Even without the inclusion of dedicated I/O registers, these I/Os, in combination with array registers, can achieve clock-to-out (pad-to-pad) timing as fast as 3.7 ns. I/O cells that have embedded latches and flip-flops require instantiation in HDL code; this is a design complication not encountered in SX FPGAs. Fast pin-to-pin timing ensures that the device will have little trouble interfacing with any other device in the system, which in turn enables parallel design of system components and reduces overall design time.

Power Requirements

The SX family supports 3.3 V operation and is designed to tolerate 5.0 V inputs. (Table 1-1). Power consumption is extremely low due to the very short distances signals are required to travel to complete a circuit. Power requirements are further reduced because of the small number of low-resistance antifuses in the path. The antifuse architecture does not require active circuitry to hold a charge (as do SRAM or EPROM), making it the lowest power architecture on the market.

Table 1-1 •   **Supply Voltages**

Device	V_{CCA}	V_{CCI}	V_{CCR}	Maximum Input Tolerance	Maximum Output Drive
A54SX08 A54SX16 A54SX32	3.3 V	3.3 V	5.0 V	5.0 V	3.3 V
A54SX16-P*	3.3 V	3.3 V	3.3 V	3.3 V	3.3 V
	3.3 V	3.3 V	5.0 V	5.0 V	3.3 V
	3.3 V	5.0 V	5.0 V	5.0 V	5.0 V

Note: *A54SX16-P has three different entries because it is capable of both a 3.3 V and a 5.0 V drive.*

Boundary Scan Testing (BST)

All SX devices are IEEE 1149.1 compliant. SX devices offer superior diagnostic and testing capabilities by providing Boundary Scan Testing (BST) and probing capabilities. These functions are controlled through the special test pins in conjunction with the program fuse. The functionality of each pin is described in Table 1-2. In the dedicated test mode, TCK, TDI, and TDO are dedicated pins and cannot be used as regular I/Os. In flexible mode, TMS should be set HIGH through a pull-up resistor of 10 kΩ. TMS can be pulled LOW to initiate the test sequence.

The program fuse determines whether the device is in dedicated or flexible mode. The default (fuse not blown) is flexible mode.

Table 1-2 • **Boundary Scan Pin Functionality**

Program Fuse Blown (Dedicated Test Mode)	Program Fuse Not Blown (Flexible Mode)
TCK, TDI, TDO are dedicated BST pins.	TCK, TDI, TDO are flexible and may be used as I/Os.
No need for pull-up resistor for TMS	Use a pull-up resistor of 10 kΩ on TMS.

Dedicated Test Mode

In Dedicated mode, all JTAG pins are reserved for BST; designers cannot use them as regular I/Os. An internal pull-up resistor is automatically enabled on both TMS and TDI pins, and the TMS pin will function as defined in the IEEE 1149.1 (JTAG) specification.

To select Dedicated mode, users need to reserve the JTAG pins in Actel's Designer software by checking the "Reserve JTAG" box in "Device Selection Wizard" (Figure 1-7). JTAG pins comply with LVTTL/TTL I/O specification regardless of whether they are used as a user I/O or a JTAG I/O. Refer to the Table 1-5 on page 1-8 for detailed specifications.

Figure 1-7 • **Device Selection Wizard**

Development Tool Support

The SX family of FPGAs is fully supported by both the Actel Libero® Integrated Design Environment (IDE) and Designer FPGA Development software. Actel Libero IDE is a design management environment, seamlessly integrating design tools while guiding the user through the design flow, managing all design and log files, and passing necessary design data among tools. Libero IDE allows users to integrate both schematic and HDL synthesis into a single flow and verify the entire design in a single environment. Libero IDE includes Synplify® for Actel from Synplicity®, ViewDraw® for Actel from Mentor Graphics®, Model*Sim*® HDL Simulator from Mentor Graphics, WaveFormer Lite™ from SynaptiCAD™, and Designer software from Actel. Refer to the Libero IDE flow diagram (located on the Actel website) for more information.

Actel Designer software is a place-and-route tool and provides a comprehensive suite of backend support tools for FPGA development. The Designer software includes timing-driven place-and-route, and a world-class integrated static timing analyzer and constraints editor. With the Designer software, a user can select and lock package pins while only minimally impacting the results of place-and-route. Additionally, the back-annotation flow is compatible with all the major simulators, and the simulation results can be cross-probed with Silicon Explorer II, Actel integrated verification and logic analysis tool. Another tool included in the Designer software is the SmartGen core generator, which easily creates popular and commonly used logic functions for implementation into your schematic or HDL design. Actel Designer software is compatible with the most popular FPGA design entry and verification tools from companies such as Mentor Graphics, Synplicity, Synopsys®, and Cadence® Design Systems. The Designer software is available for both the Windows® and UNIX® operating systems.

Probe Circuit Control Pins

The Silicon Explorer II tool uses the boundary scan ports (TDI, TCK, TMS, and TDO) to select the desired nets for verification. The selected internal nets are assigned to the PRA/PRB pins for observation. Figure 1-8 on page 1-7 illustrates the interconnection between Silicon Explorer II and the FPGA to perform in-circuit verification.

Design Considerations

The TDI, TCK, TDO, PRA, and PRB pins should not be used as input or bidirectional ports. Because these pins are active during probing, critical signals input through these pins are not available while probing. In addition, the Security Fuse should not be programmed because doing so disables the Probe Circuitry.

Figure 1-9 shows the 5.0 V PCI V/I curve and the minimum and maximum PCI drive characteristics of the A54SX16P device.
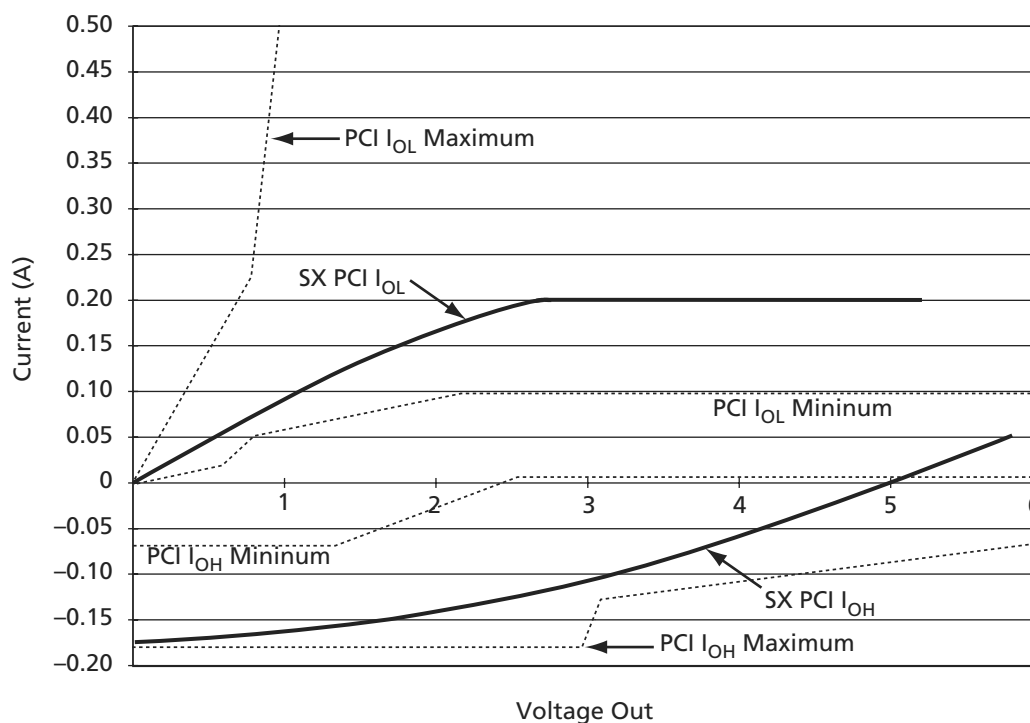

Figure 1-9 • **5.0 V PCI Curve for A54SX16P Device**

$I_{OH} = 11.9 \times (V_{OUT} - 5.25) \times (V_{OUT} + 2.45)$

for $V_{CC} > V_{OUT} > 3.1$ V

EQ 1-1

$I_{OL} = 78.5 \times V_{OUT} \times (4.4 - V_{OUT})$

for 0 V $< V_{OUT} < 0.71$ V

EQ 1-2

Evaluating Power in SX Devices

A critical element of system reliability is the ability of electronic devices to safely dissipate the heat generated during operation. The thermal characteristics of a circuit depend on the device and package used, the operating temperature, the operating current, and the system's ability to dissipate heat.

You should complete a power evaluation early in the design process to help identify potential heat-related problems in the system and to prevent the system from exceeding the device's maximum allowed junction temperature.

The actual power dissipated by most applications is significantly lower than the power the package can dissipate. However, a thermal analysis should be performed for all projects. To perform a power evaluation, follow these steps:

1. Estimate the power consumption of the application.
2. Calculate the maximum power allowed for the device and package.
3. Compare the estimated power and maximum power values.

Estimating Power Consumption

The total power dissipation for the SX family is the sum of the DC power dissipation and the AC power dissipation. Use EQ 1-5 to calculate the estimated power consumption of your application.

$$P_{Total} = P_{DC} + P_{AC}$$

EQ 1-5

DC Power Dissipation

The power due to standby current is typically a small component of the overall power. The Standby power is shown in Table 1-12 for commercial, worst-case conditions (70°C).

*Table 1-12 • **Standby Power***

I_{CC}	V_{CC}	Power
4 mA	3.6 V	14.4 mW

The DC power dissipation is defined in EQ 1-6.

$$P_{DC} = (I_{standby}) \times V_{CCA} + (I_{standby}) \times V_{CCR} + (I_{standby}) \times V_{CCI} + xV_{OL} \times I_{OL} + y(V_{CCI} - V_{OH}) \times V_{OH}$$

EQ 1-6

AC Power Dissipation

The power dissipation of the SX Family is usually dominated by the dynamic power dissipation. Dynamic power dissipation is a function of frequency, equivalent capacitance, and power supply voltage. The AC power dissipation is defined in EQ 1-7 and EQ 1-8.

$$P_{AC} = P_{Module} + P_{RCLKA\ Net} + P_{RCLKB\ Net} + P_{HCLK\ Net} + P_{Output\ Buffer} + P_{Input\ Buffer}$$

EQ 1-7

$$P_{AC} = V_{CCA}^2 \times [(m \times C_{EQM} \times f_m)_{Module} + (n \times C_{EQI} \times f_n)_{Input\ Buffer} + (p \times (C_{EQO} + C_L) \times f_p)_{Output\ Buffer} + (0.5 \times (q_1 \times C_{EQCR} \times f_{q1}) + (r_1 \times f_{q1}))_{RCLKA} + (0.5 \times (q2 \times CEQCR \times f_{q2}) + (r2 \times f_{q2}))RCLKB + (0.5 \times (s_1 \times C_{EQHV} \times f_{s1}) + (C_{EQHF} \times f_{s1}))_{HCLK}]$$

EQ 1-8

Definition of Terms Used in Formula

m   =   Number of logic modules switching at f_m

n   =   Number of input buffers switching at f_n

p   =   Number of output buffers switching at f_p

q_1 =   Number of clock loads on the first routed array clock

q_2 =   Number of clock loads on the second routed array clock

x   =   Number of I/Os at logic low

y   =   Number of I/Os at logic high

r_1 =   Fixed capacitance due to first routed array clock

r_2 =   Fixed capacitance due to second routed array clock

s_1 =   Number of clock loads on the dedicated array clock

C_{EQM} =   Equivalent capacitance of logic modules in pF

C_{EQI} =   Equivalent capacitance of input buffers in pF

C_{EQO} =   Equivalent capacitance of output buffers in pF

C_{EQCR} =   Equivalent capacitance of routed array clock in pF

C_{EQHV} =   Variable capacitance of dedicated array clock

C_{EQHF} =   Fixed capacitance of dedicated array clock

C_L =   Output lead capacitance in pF

f_m =   Average logic module switching rate in MHz

f_n =   Average input buffer switching rate in MHz

f_p =   Average output buffer switching rate in MHz

f_{q1} =   Average first routed array clock rate in MHz

f_{q2} =   Average second routed array clock rate in MHz

f_{s1} =   Average dedicated array clock rate in MHz

Figure 1-11 shows the characterized power dissipation numbers for the shift register design using frequencies ranging from 1 MHz to 200 MHz.
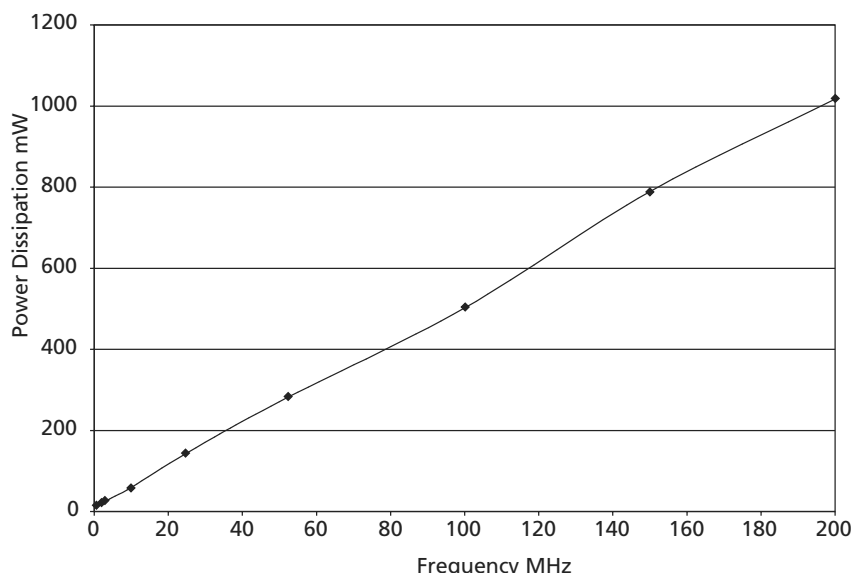
Figure 1-11 • **Power Dissipation**

Junction Temperature (T$_J$)

The temperature that you select in Designer Series software is the junction temperature, not ambient temperature. This is an important distinction because the heat generated from dynamic power consumption is usually hotter than the ambient temperature. Use the equation below to calculate junction temperature.

$$\text{Junction Temperature} = \Delta T + T_a$$

EQ 1-13

Where:

T$_a$ = Ambient Temperature

ΔT = Temperature gradient between junction (silicon) and ambient

$\Delta T = \theta_{ja} \times P$

P = Power calculated from Estimating Power Consumption section

θ_{ja} = Junction to ambient of package. θ_{ja} numbers are located in the "Package Thermal Characteristics" section.

Package Thermal Characteristics

The device junction to case thermal characteristic is θ_{jc}, and the junction to ambient air characteristic is θ_{ja}. The thermal characteristics for θ_{ja} are shown with two different air flow rates.

The maximum junction temperature is 150 °C.

A sample calculation of the absolute maximum power dissipation allowed for a TQFP 176-pin package at commercial temperature and still air is as follows:

$$\text{Maximum Power Allowed} = \frac{\text{Max. junction temp. (°C)} - \text{Max. ambient temp. (°C)}}{\theta_{ja} \text{ (°C/W)}} = \frac{150°C - 70°C}{28°C/W} = 2.86 \text{ W}$$

EQ 1-14

Figure 1-13 • **Output Buffer Delays**

Figure 1-14 • **AC Test Loads**

Figure 1-15 • **Input Buffer Delays**

Figure 1-16 • **C-Cell Delays**

A54SX16P Timing Characteristics

Table 1-19 • **A54SX16P Timing Characteristics**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA}, V_{CCI} = 3.0 V, T_J = 70°C)

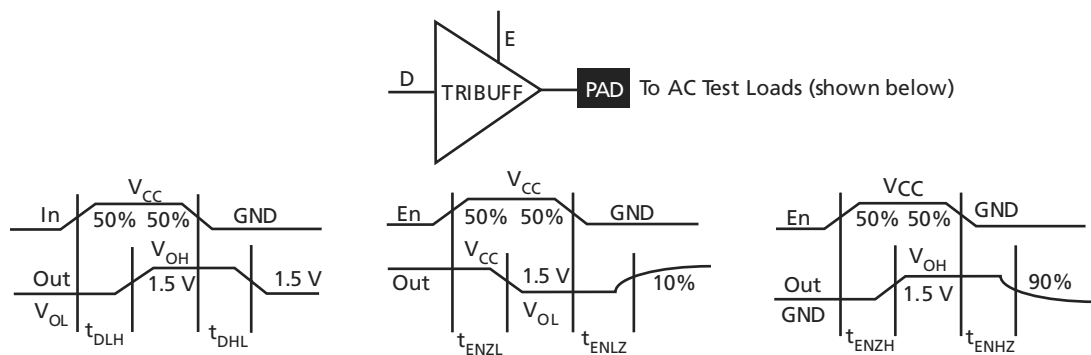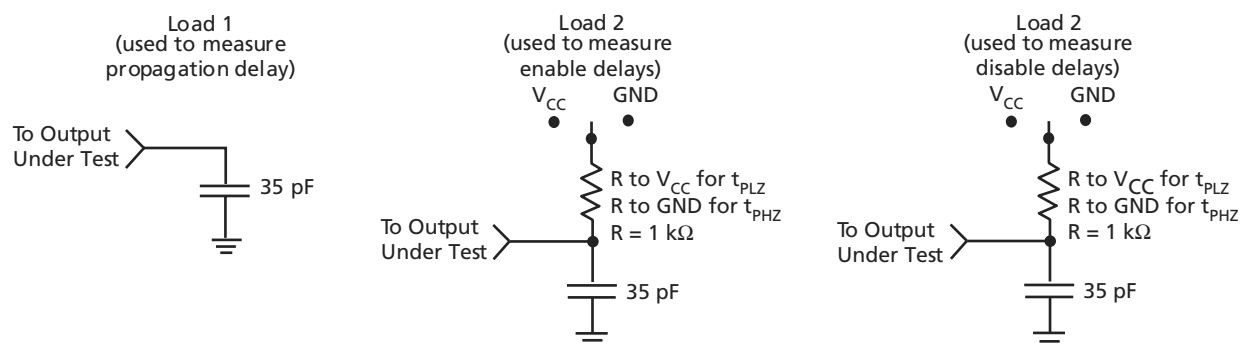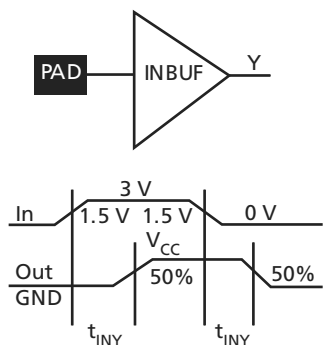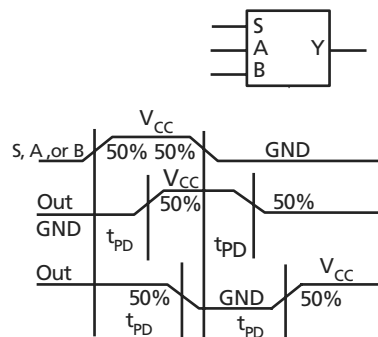| Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
C-Cell Propagation Delays[1]										
t_{PD}	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted Routing Delays[2]										
t_{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t_{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{RD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{RD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timing										
t_{RCO}	Sequential Clock-to-Q		0.9		1.1		1.3		1.4	ns
t_{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t_{PRESET}	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t_{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t_{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t_{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t_{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted Input Routing Delays[2]										
t_{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{IRD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{IRD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Note:

1. *For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.*

2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*

3. *Delays based on 10 pF loading.*

Table 1-19 • **A54SX16P Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V$_{CCR}$ = 4.75 V, V$_{CCA}$,V$_{CCI}$ = 3.0 V, T$_J$ = 70°C)

Parameter	Description	'–3' Speed Min.	'–3' Speed Max.	'–2' Speed Min.	'–2' Speed Max.	'–1' Speed Min.	'–1' Speed Max.	'Std' Speed Min.	'Std' Speed Max.	Units
Dedicated (Hardwired) Array Clock Network										
t$_{HCKH}$	Input LOW to HIGH (pad to R-Cell input)		1.2		1.4		1.5		1.8	ns
t$_{HCKL}$	Input HIGH to LOW (pad to R-Cell input)		1.2		1.4		1.6		1.9	ns
t$_{HPWH}$	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t$_{HPWL}$	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t$_{HCKSW}$	Maximum Skew		0.2		0.2		0.3		0.3	ns
t$_{HP}$	Minimum Period	2.7		3.1		3.6		4.2		ns
f$_{HMAX}$	Maximum Frequency		350		320		280		240	MHz
Routed Array Clock Networks										
t$_{RCKH}$	Input LOW to HIGH (light load) (pad to R-Cell input)		1.6		1.8		2.1		2.5	ns
t$_{RCKL}$	Input HIGH to LOW (Light Load) (pad to R-Cell input)		1.8		2.0		2.3		2.7	ns
t$_{RCKH}$	Input LOW to HIGH (50% load) (pad to R-Cell input)		1.8		2.1		2.5		2.8	ns
t$_{RCKL}$	Input HIGH to LOW (50% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t$_{RCKH}$	Input LOW to HIGH (100% load) (pad to R-Cell input)		1.8		2.1		2.4		2.8	ns
t$_{RCKL}$	Input HIGH to LOW (100% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t$_{RPWH}$	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t$_{RPWL}$	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t$_{RCKSW}$	Maximum Skew (light load)		0.5		0.5		0.5		0.7	ns
t$_{RCKSW}$	Maximum Skew (50% load)		0.5		0.6		0.7		0.8	ns
t$_{RCKSW}$	Maximum Skew (100% load)		0.5		0.6		0.7		0.8	ns
TTL Output Module Timing										
t$_{DLH}$	Data-to-Pad LOW to HIGH		2.4		2.8		3.1		3.7	ns
t$_{DHL}$	Data-to-Pad HIGH to LOW		2.3		2.9		3.2		3.8	ns
t$_{ENZL}$	Enable-to-Pad, Z to L		3.0		3.4		3.9		4.6	ns
t$_{ENZH}$	Enable-to-Pad, Z to H		3.3		3.8		4.3		5.0	ns
t$_{ENLZ}$	Enable-to-Pad, L to Z		2.3		2.7		3.0		3.5	ns
t$_{ENHZ}$	Enable-to-Pad, H to Z		2.8		3.2		3.7		4.3	ns

Note:

1. *For dual-module macros, use t$_{PD}$ + t$_{RD1}$ + t$_{PDn}$, t$_{RCO}$ + t$_{RD1}$ + t$_{PDn}$, or t$_{PD1}$ + t$_{RD1}$ + t$_{SUD}$, whichever is appropriate.*

2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*

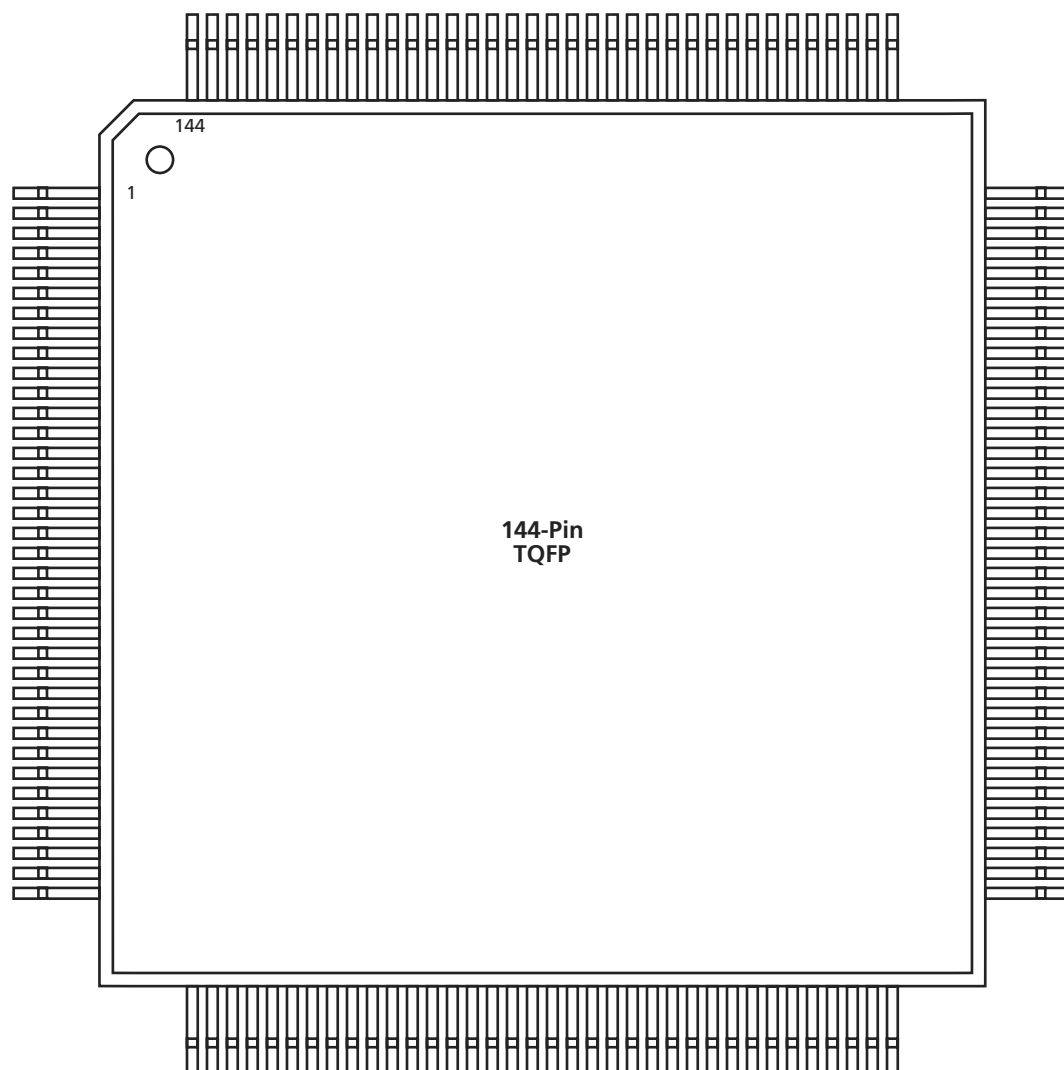3. *Delays based on 10 pF loading.*

144-Pin TQFP

Figure 2-3 • **144-Pin TQFP (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at
http://www.actel.com/products/rescenter/package/index.html.

144-Pin TQFP				144-Pin TQFP			
Pin Number	**A54SX08 Function**	**A54SX16P Function**	**A54SX32 Function**	**Pin Number**	**A54SX08 Function**	**A54SX16P Function**	**A54SX32 Function**
1	GND	GND	GND	37	I/O	I/O	I/O
2	TDI, I/O	TDI, I/O	TDI, I/O	38	I/O	I/O	I/O
3	I/O	I/O	I/O	39	I/O	I/O	I/O
4	I/O	I/O	I/O	40	I/O	I/O	I/O
5	I/O	I/O	I/O	41	I/O	I/O	I/O
6	I/O	I/O	I/O	42	I/O	I/O	I/O
7	I/O	I/O	I/O	43	I/O	I/O	I/O
8	I/O	I/O	I/O	44	V_{CCI}	V_{CCI}	V_{CCI}
9	TMS	TMS	TMS	45	I/O	I/O	I/O
10	V_{CCI}	V_{CCI}	V_{CCI}	46	I/O	I/O	I/O
11	GND	GND	GND	47	I/O	I/O	I/O
12	I/O	I/O	I/O	48	I/O	I/O	I/O
13	I/O	I/O	I/O	49	I/O	I/O	I/O
14	I/O	I/O	I/O	50	I/O	I/O	I/O
15	I/O	I/O	I/O	51	I/O	I/O	I/O
16	I/O	I/O	I/O	52	I/O	I/O	I/O
17	I/O	I/O	I/O	53	I/O	I/O	I/O
18	I/O	I/O	I/O	54	PRB, I/O	PRB, I/O	PRB, I/O
19	V_{CCR}	V_{CCR}	V_{CCR}	55	I/O	I/O	I/O
20	V_{CCA}	V_{CCA}	V_{CCA}	56	V_{CCA}	V_{CCA}	V_{CCA}
21	I/O	I/O	I/O	57	GND	GND	GND
22	I/O	I/O	I/O	58	V_{CCR}	V_{CCR}	V_{CCR}
23	I/O	I/O	I/O	59	I/O	I/O	I/O
24	I/O	I/O	I/O	60	HCLK	HCLK	HCLK
25	I/O	I/O	I/O	61	I/O	I/O	I/O
26	I/O	I/O	I/O	62	I/O	I/O	I/O
27	I/O	I/O	I/O	63	I/O	I/O	I/O
28	GND	GND	GND	64	I/O	I/O	I/O
29	V_{CCI}	V_{CCI}	V_{CCI}	65	I/O	I/O	I/O
30	V_{CCA}	V_{CCA}	V_{CCA}	66	I/O	I/O	I/O
31	I/O	I/O	I/O	67	I/O	I/O	I/O
32	I/O	I/O	I/O	68	V_{CCI}	V_{CCI}	V_{CCI}
33	I/O	I/O	I/O	69	I/O	I/O	I/O
34	I/O	I/O	I/O	70	I/O	I/O	I/O
35	I/O	I/O	I/O	71	TDO, I/O	TDO, I/O	TDO, I/O
36	GND	GND	GND	72	I/O	I/O	I/O

176-Pin TQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	NC	I/O	I/O
4	I/O	I/O	I/O
5	I/O	I/O	I/O
6	I/O	I/O	I/O
7	I/O	I/O	I/O
8	I/O	I/O	I/O
9	I/O	I/O	I/O
10	TMS	TMS	TMS
11	V_{CCI}	V_{CCI}	V_{CCI}
12	NC	I/O	I/O
13	I/O	I/O	I/O
14	I/O	I/O	I/O
15	I/O	I/O	I/O
16	I/O	I/O	I/O
17	I/O	I/O	I/O
18	I/O	I/O	I/O
19	I/O	I/O	I/O
20	I/O	I/O	I/O
21	GND	GND	GND
22	V_{CCA}	V_{CCA}	V_{CCA}
23	GND	GND	GND
24	I/O	I/O	I/O
25	I/O	I/O	I/O
26	I/O	I/O	I/O
27	I/O	I/O	I/O
28	I/O	I/O	I/O
29	I/O	I/O	I/O
30	I/O	I/O	I/O
31	I/O	I/O	I/O
32	V_{CCI}	V_{CCI}	V_{CCI}
33	V_{CCA}	V_{CCA}	V_{CCA}
34	I/O	I/O	I/O

| 176-Pin TQFP			
Pin Number	**A54SX08 Function**	**A54SX16, A54SX16P Function**	**A54SX32 Function**
35	I/O	I/O	I/O
36	I/O	I/O	I/O
37	I/O	I/O	I/O
38	I/O	I/O	I/O
39	I/O	I/O	I/O
40	NC	I/O	I/O
41	I/O	I/O	I/O
42	NC	I/O	I/O
43	I/O	I/O	I/O
44	GND	GND	GND
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	I/O	I/O	I/O
49	I/O	I/O	I/O
50	I/O	I/O	I/O
51	I/O	I/O	I/O
52	V_{CCI}	V_{CCI}	V_{CCI}
53	I/O	I/O	I/O
54	NC	I/O	I/O
55	I/O	I/O	I/O
56	I/O	I/O	I/O
57	NC	I/O	I/O
58	I/O	I/O	I/O
59	I/O	I/O	I/O
60	I/O	I/O	I/O
61	I/O	I/O	I/O
62	I/O	I/O	I/O
63	I/O	I/O	I/O
64	PRB, I/O	PRB, I/O	PRB, I/O
65	GND	GND	GND
66	V_{CCA}	V_{CCA}	V_{CCA}
67	V_{CCR}	V_{CCR}	V_{CCR}
68	I/O	I/O	I/O

100-Pin VQFP
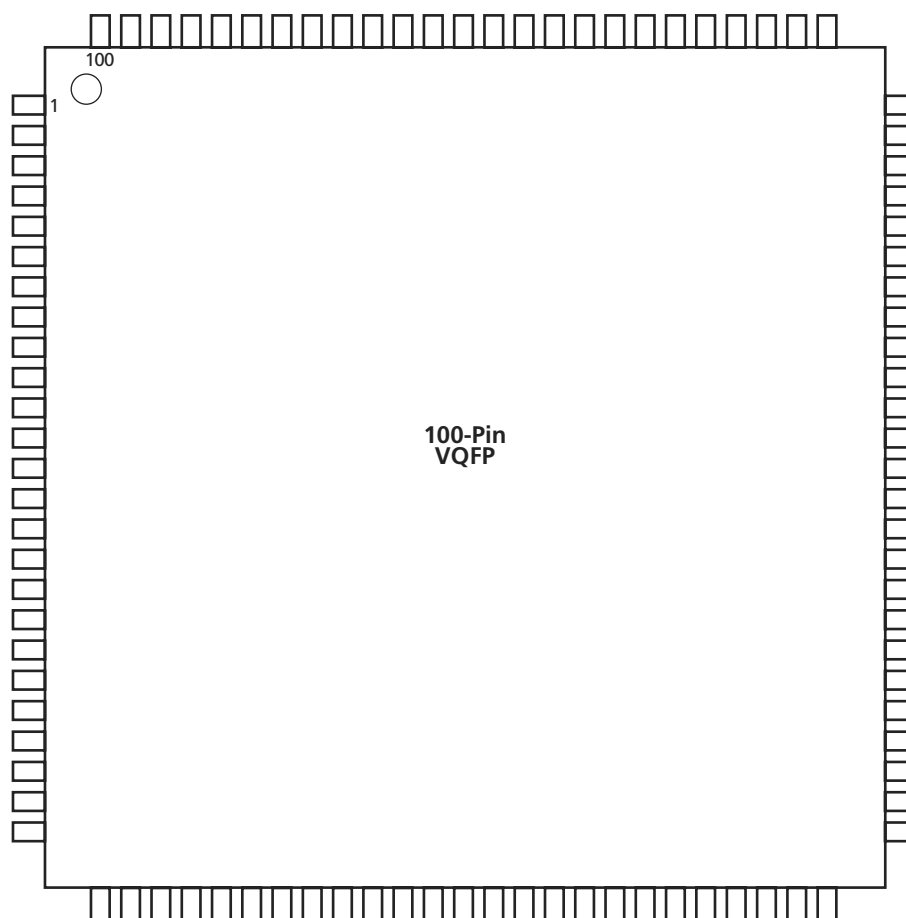
100-Pin
VQFP

Figure 2-5 • **100-Pin VQFP (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at
http://www.actel.com/products/rescenter/package/index.html.

| 313-Pin PBGA		313-Pin PBGA		313-Pin PBGA		313-Pin PBGA	
Pin Number	A54SX32 Function	Pin Number	A54SX32 Function	Pin Number	A54SX32 Function	Pin Number	A54SX32 Function
H20	I/O	L25	I/O	R5	I/O	V10	I/O
H22	V_{CCI}	M2	I/O	R7	I/O	V12	I/O
H24	I/O	M4	I/O	R9	I/O	V14	I/O
J1	I/O	M6	I/O	R11	I/O	V16	NC
J3	I/O	M8	I/O	R13	GND	V18	I/O
J5	I/O	M10	I/O	R15	I/O	V20	I/O
J7	NC	M12	GND	R17	I/O	V22	V_{CCA}
J9	I/O	M14	GND	R19	I/O	V24	V_{CCI}
J11	I/O	M16	V_{CCI}	R21	I/O	W1	I/O
J13	CLKA	M18	I/O	R23	I/O	W3	I/O
J15	I/O	M20	I/O	R25	I/O	W5	I/O
J17	I/O	M22	I/O	T2	I/O	W7	NC
J19	I/O	M24	I/O	T4	I/O	W9	I/O
J21	GND	N1	I/O	T6	I/O	W11	I/O
J23	I/O	N3	V_{CCA}	T8	I/O	W13	V_{CCI}
J25	I/O	N5	V_{CCR}	T10	I/O	W15	I/O
K2	I/O	N7	I/O	T12	I/O	W17	I/O
K4	I/O	N9	V_{CCI}	T14	HCLK	W19	I/O
K6	I/O	N11	GND	T16	I/O	W21	I/O
K8	V_{CCI}	N13	GND	T18	I/O	W23	I/O
K10	I/O	N15	GND	T20	I/O	W25	I/O
K12	I/O	N17	I/O	T22	I/O	Y2	I/O
K14	I/O	N19	I/O	T24	I/O	Y4	I/O
K16	I/O	N21	I/O	U1	I/O	Y6	I/O
K18	I/O	N23	V_{CCR}	U3	I/O	Y8	I/O
K20	V_{CCA}	N25	V_{CCA}	U5	V_{CCI}	Y10	I/O
K22	I/O	P2	I/O	U7	I/O	Y12	I/O
K24	I/O	P4	I/O	U9	I/O	Y14	I/O
L1	I/O	P6	I/O	U11	I/O	Y16	I/O
L3	I/O	P8	I/O	U13	I/O	Y18	I/O
L5	I/O	P10	I/O	U15	I/O	Y20	NC
L7	I/O	P12	GND	U17	I/O	Y22	I/O
L9	I/O	P14	GND	U19	I/O	Y24	NC
L11	I/O	P16	I/O	U21	I/O		
L13	GND	P18	I/O	U23	I/O		
L15	I/O	P20	NC	U25	I/O		
L17	I/O	P22	I/O	V2	V_{CCA}		
L19	I/O	P24	I/O	V4	I/O		
L21	I/O	R1	I/O	V6	I/O		
L23	I/O	R3	I/O	V8	I/O		

144-Pin FBGA
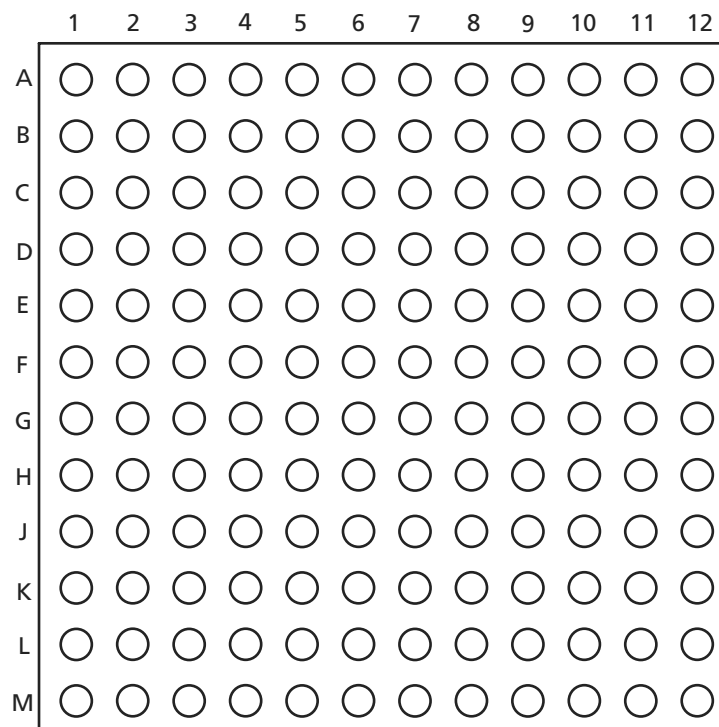
Figure 2-8 • **144-Pin FBGA (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at
http://www.actel.com/products/rescenter/package/index.html.