Microsemi Corporation - A54SX08-TQG144 Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	768
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	113
Number of Gates	12000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx08-tqg144

SX Family FPGAs

General Description

The Actel SX family of FPGAs features a sea-of-modules architecture that delivers device performance and integration levels not currently achieved by any other FPGA architecture. SX devices greatly simplify design time, enable dramatic reductions in design costs and power consumption, and further decrease time to market for performance-intensive applications.

The Actel SX architecture features two types of logic modules, the combinatorial cell (C-cell) and the register cell (R-cell), each optimized for fast and efficient mapping of synthesized logic functions. The routing and interconnect resources are in the metal layers above the logic modules, providing optimal use of silicon. This enables the entire floor of the device to be spanned with an uninterrupted grid of fine-grained, synthesis-friendly logic modules (or "sea-of-modules"), which reduces the distance signals have to travel between logic modules. To minimize signal propagation delay, SX devices employ both local and general routing resources. The high-speed local routing resources (DirectConnect and FastConnect) enable very fast local signal propagation that is optimal for fast counters, state machines, and datapath logic. The general system of segmented routing tracks allows any logic module in the array to be connected to any other logic or I/O module. Within this system, propagation delay is minimized by limiting the number of antifuse interconnect elements to five (90 percent of connections typically use only three antifuses). The unique local and general routing structure featured in SX devices gives fast and predictable performance, allows 100 percent pin-locking with full logic utilization, enables concurrent PCB development, reduces design time, and allows designers to achieve performance goals with minimum effort.

Further complementing SX's flexible routing structure is a hardwired, constantly loaded clock network that has been tuned to provide fast clock propagation with minimal clock skew. Additionally, the high performance of the internal logic has eliminated the need to embed latches or flip-flops in the I/O cells to achieve fast clock-to-out or fast input setup times. SX devices have easy to use I/O cells that do not require HDL instantiation, facilitating design reuse and reducing design and verification time.

SX Family Architecture

The SX family architecture was designed to satisfy next-generation performance and integration requirements for production-volume designs in a broad range of applications.

Programmable Interconnect Element

The SX family provides efficient use of silicon by locating the routing interconnect resources between the Metal 2 (M2) and Metal 3 (M3) layers (Figure 1-1 on page 1-2). This completely eliminates the channels of routing and interconnect resources between logic modules (as implemented on SRAM FPGAs and previous generations of antifuse FPGAs), and enables the entire floor of the device to be spanned with an uninterrupted grid of logic modules.

Interconnection between these logic modules is achieved using The Actel patented metal-to-metal programmable antifuse interconnect elements, which are embedded between the M2 and M3 layers. The antifuses are normally open circuit and, when programmed, form a permanent low-impedance connection.

The extremely small size of these interconnect elements gives the SX family abundant routing resources and provides excellent protection against design pirating. Reverse engineering is virtually impossible because it is extremely difficult to distinguish between programmed and unprogrammed antifuses, and there is no configuration bitstream to intercept.

Additionally, the interconnect elements (i.e., the antifuses and metal tracks) have lower capacitance and lower resistance than any other device of similar capacity, leading to the fastest signal propagation in the industry.

Logic Module Design

The SX family architecture is described as a "sea-of-modules" architecture because the entire floor of the device is covered with a grid of logic modules with virtually no chip area lost to interconnect elements or routing. The Actel SX family provides two types of logic modules, the register cell (R-cell) and the combinatorial cell (C-cell).

Chip Architecture

The SX family chip architecture provides a unique approach to module organization and chip routing that delivers the best register/logic mix for a wide variety of new and emerging applications.

Module Organization

Actel has arranged all C-cell and R-cell logic modules into horizontal banks called *clusters*. There are two types of *clusters*: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells.

To increase design efficiency and device performance, Actel has further organized these modules into *SuperClusters* (Figure 1-4). SuperCluster 1 is a two-wide grouping of Type 1 clusters. SuperCluster 2 is a two-wide group containing one Type 1 cluster and one Type 2 cluster. SX devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flip-flops.

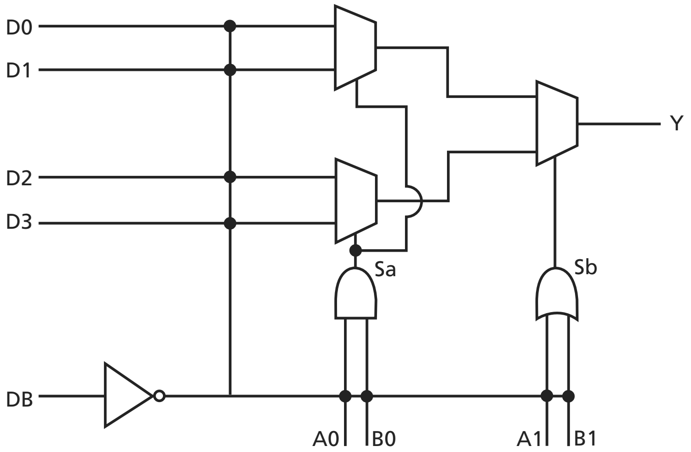

Figure 1-3 • **C-Cell**

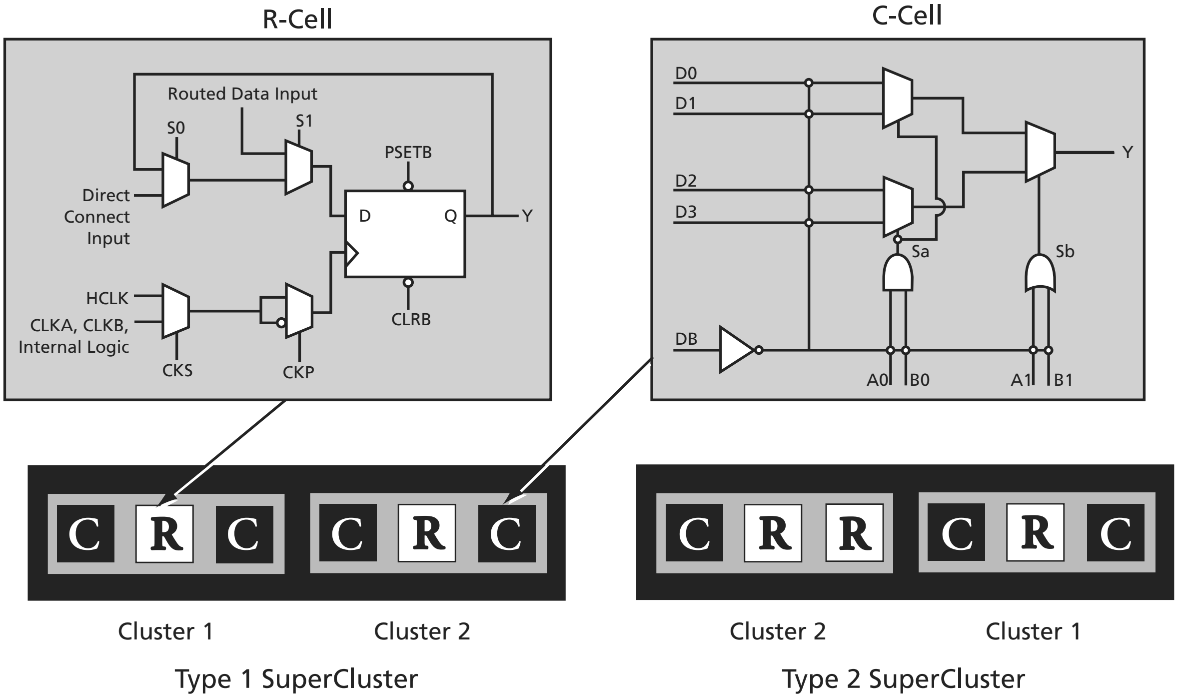

Figure 1-4 • **Cluster Organization**

Routing Resources

Clusters and SuperClusters can be connected through the use of two innovative local routing resources called *FastConnect* and *DirectConnect,* which enable extremely fast and predictable interconnection of modules within clusters and SuperClusters (Figure 1-5 and Figure 1-6). This routing architecture also dramatically reduces the number of antifuses required to complete a circuit, ensuring the highest possible performance.

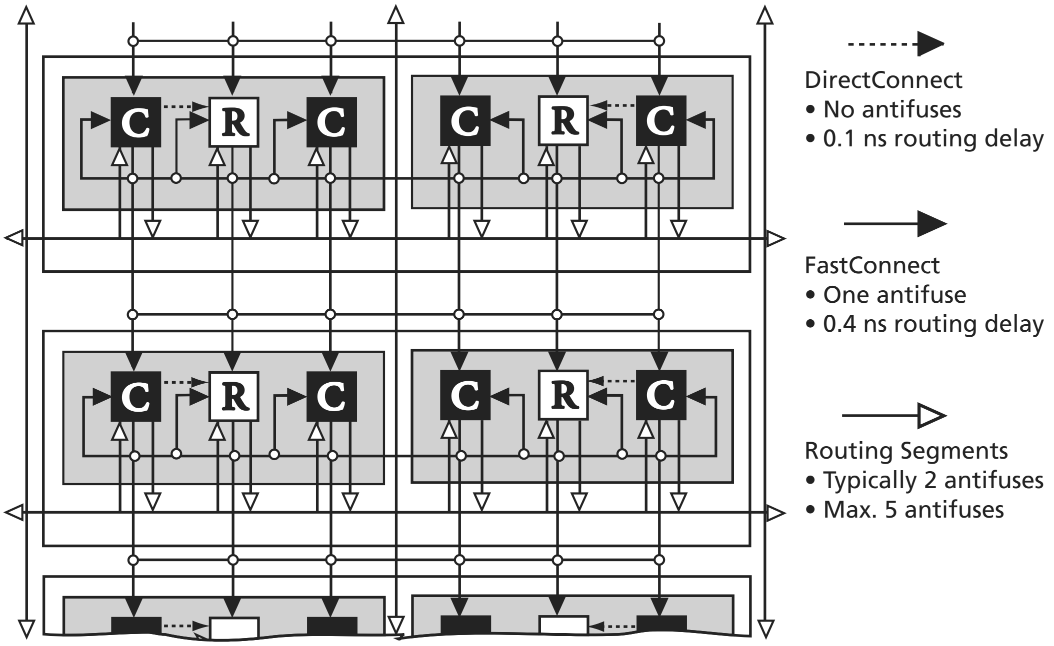

Figure 1-5 • **DirectConnect and FastConnect for Type 1 SuperClusters**

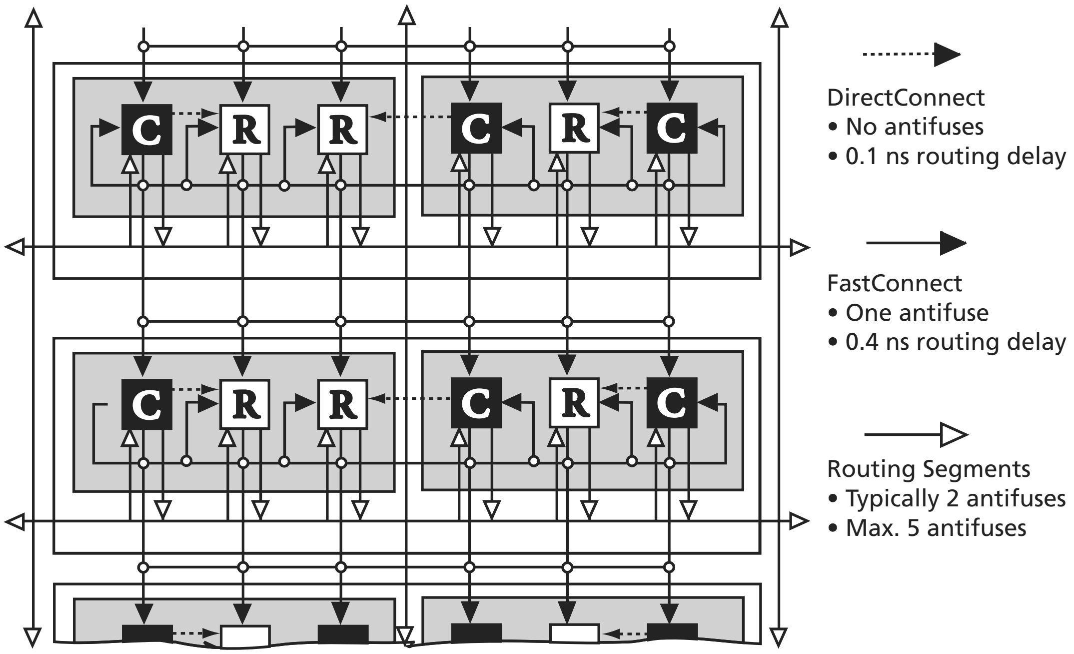

Figure 1-6 • **DirectConnect and FastConnect for Type 2 SuperClusters**

DirectConnect is a horizontal routing resource that provides connections from a C-cell to its neighboring R-cell in a given SuperCluster. DirectConnect uses a hardwired signal path requiring no programmable interconnection to achieve its fast signal propagation time of less than 0.1 ns.

FastConnect enables horizontal routing between any two logic modules within a given SuperCluster and vertical routing with the SuperCluster immediately below it. Only one programmable connection is used in a FastConnect path, delivering maximum pin-to-pin propagation of 0.4 ns.

In addition to DirectConnect and FastConnect, the architecture makes use of two globally oriented routing resources known as segmented routing and high-drive routing. The Actel segmented routing structure provides a variety of track lengths for extremely fast routing between SuperClusters. The exact combination of track lengths and antifuses within each path is chosen by the 100 percent automatic place-and-route software to minimize signal propagation delays.

The Actel high-drive routing structure provides three clock networks. The first clock, called HCLK, is hardwired from the HCLK buffer to the clock select multiplexer (MUX) in each R-cell. This provides a fast propagation path for the clock signal, enabling the 3.7 ns clock-to-out (pin-to-pin) performance of the SX devices. The hardwired clock is tuned to provide clock skew as low as 0.25 ns. The remaining two clocks (CLKA, CLKB) are global clocks that can be sourced from external pins or from internal logic signals within the SX device.

Other Architectural Features

Technology

The Actel SX family is implemented on a high-voltage twin-well CMOS process using 0.35 μ design rules. The metal-to-metal antifuse is made up of a combination of amorphous silicon and dielectric material with barrier metals and has a programmed ("on" state) resistance of 25 Ω with a capacitance of 1.0 fF for low signal impedance.

Performance

The combination of architectural features described above enables SX devices to operate with internal clock frequencies exceeding 300 MHz, enabling very fast execution of even complex logic functions. Thus, the SX family is an optimal platform upon which to integrate the functionality previously contained in multiple CPLDs. In addition, designs that previously would have required a gate array to meet performance goals can now be integrated into an SX device with dramatic improvements in cost and time to market. Using timing-driven place-and-route tools, designers can achieve highly deterministic device performance. With SX devices, designers do not need to use complicated performance-enhancing design techniques such as the use of redundant logic to reduce fanout on critical nets or the instantiation of macros in HDL code to achieve high performance.

I/O Modules

Each I/O on an SX device can be configured as an input, an output, a tristate output, or a bidirectional pin.

Even without the inclusion of dedicated I/O registers, these I/Os, in combination with array registers, can achieve clock-to-out (pad-to-pad) timing as fast as 3.7 ns. I/O cells that have embedded latches and flip-flops require instantiation in HDL code; this is a design complication not encountered in SX FPGAs. Fast pin-to-pin timing ensures that the device will have little trouble interfacing with any other device in the system, which in turn enables parallel design of system components and reduces overall design time.

Power Requirements

The SX family supports 3.3 V operation and is designed to tolerate 5.0 V inputs. (Table 1-1). Power consumption is extremely low due to the very short distances signals are required to travel to complete a circuit. Power requirements are further reduced because of the small number of low-resistance antifuses in the path. The antifuse architecture does not require active circuitry to hold a charge (as do SRAM or EPROM), making it the lowest power architecture on the market.

Table 1-1 • **Supply Voltages**

Device	V_{CCA}	V_{CCI}	V_{CCR}	Maximum Input Tolerance	Maximum Output Drive
A54SX08 A54SX16 A54SX32	3.3 V	3.3 V	5.0 V	5.0 V	3.3 V
A54SX16-P*	3.3 V	3.3 V	3.3 V	3.3 V	3.3 V
	3.3 V	3.3 V	5.0 V	5.0 V	3.3 V
	3.3 V	5.0 V	5.0 V	5.0 V	5.0 V

Note: **A54SX16-P has three different entries because it is capable of both a 3.3 V and a 5.0 V drive.*

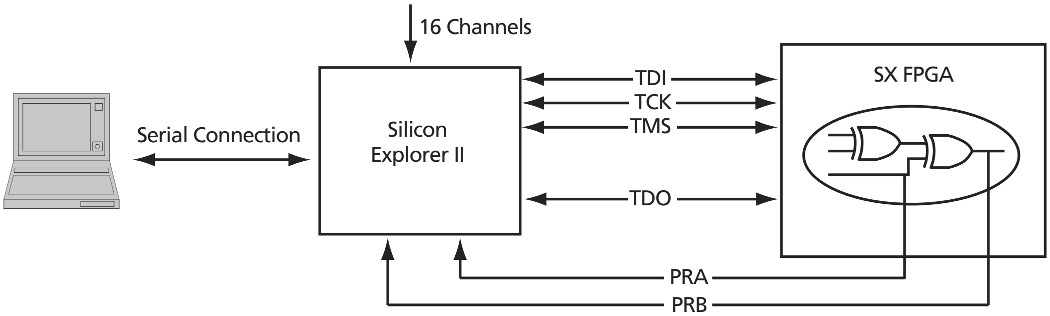

Figure 1-8 • **Probe Setup**

Programming

Device programming is supported through Silicon Sculptor series of programmers. In particular, Silicon Sculptor II are compact, robust, single-site and multi-site device programmer for the PC.

With standalone software, Silicon Sculptor II allows concurrent programming of multiple units from the same PC, ensuring the fastest programming times possible. Each fuse is subsequently verified by Silicon Sculptor II to insure correct programming. In addition, integrity tests ensure that no extra fuses are programmed. Silicon Sculptor II also provides extensive hardware self-testing capability.

The procedure for programming an SX device using Silicon Sculptor II are as follows:

1. Load the .AFM file
2. Select the device to be programmed
3. Begin programming

When the design is ready to go to production, Actel offers device volume-programming services either through distribution partners or via in-house programming from the factory.

For more details on programming SX devices, refer to the *Programming Antifuse Devices* application note and the *Silicon Sculptor II User's Guide*.

3.3 V / 5 V Operating Conditions

Table 1-3 • **Absolute Maximum Ratings**[1]

Symbol	Parameter	Limits	Units
V_{CCR}[2]	DC Supply Voltage[3]	–0.3 to + 6.0	V
V_{CCA}[2]	DC Supply Voltage	–0.3 to + 4.0	V
V_{CCI}[2]	DC Supply Voltage (A54SX08, A54SX16, A54SX32)	–0.3 to + 4.0	V
V_{CCI}[2]	DC Supply Voltage (A54SX16P)	–0.3 to + 6.0	V
V_I	Input Voltage	–0.5 to + 5.5	V
V_O	Output Voltage	–0.5 to + 3.6	V
I_{IO}	I/O Source Sink Current[3]	–30 to + 5.0	mA
T_{STG}	Storage Temperature	–65 to +150	°C

Notes:

1. *Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. Exposure to absolute maximum rated conditions for extended periods may affect device reliability. Device should not be operated outside the Recommended Operating Conditions.*
2. *V_{CCR} in the A54SX16P must be greater than or equal to V_{CCI} during power-up and power-down sequences and during normal operation.*
3. *Device inputs are normally high impedance and draw extremely low current. However, when input voltage is greater than V_{CC} + 0.5 V or less than GND – 0.5 V, the internal protection diodes will forward-bias and can draw excessive current.*

PCI Compliance for the SX Family

The SX family supports 3.3 V and 5.0 V PCI and is compliant with the PCI Local Bus Specification Rev. 2.1.

Table 1-6 • **A54SX16P DC Specifications (5.0 V PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
V_{CCA}	Supply Voltage for Array		3.0	3.6	V
V_{CCR}	Supply Voltage required for Internal Biasing		4.75	5.25	V
V_{CCI}	Supply Voltage for I/Os		4.75	5.25	V
V_{IH}	Input High Voltage[1]		2.0	V_{CC} + 0.5	V
V_{IL}	Input Low Voltage[1]		–0.5	0.8	V
I_{IH}	Input High Leakage Current	V_{IN} = 2.7		70	µA
I_{IL}	Input Low Leakage Current	V_{IN} = 0.5		–70	µA
V_{OH}	Output High Voltage	I_{OUT} = –2 mA	2.4		V
V_{OL}	Output Low Voltage[2]	I_{OUT} = 3 mA, 6 mA		0.55	V
C_{IN}	Input Pin Capacitance[3]			10	pF
C_{CLK}	CLK Pin Capacitance		5	12	pF
C_{IDSEL}	IDSEL Pin Capacitance[4]			8	pF

Notes:

1. *Input leakage currents include hi-Z output leakage for all bidirectional buffers with tristate outputs.*
2. *Signals without pull-up resistors must have 3 mA low output current. Signals requiring pull-up must have 6 mA; the latter include, FRAME#, IRDY#, TRDY#, DEVSEL#, STOP#, SERR#, PERR#, LOCK#, and, when used, AD[63::32], C/BE[7::4]#, PAR64, REQ64#, and ACK64#.*
3. *Absolute maximum pin capacitance for a PCI input is 10 pF (except for CLK).*
4. *Lower capacitance on this input-only pin allows for non-resistive coupling to AD[xx].*

A54SX16P AC Specifications for (PCI Operation)

Table 1-7 • **A54SX16P AC Specifications for (PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \le 1.4$[1]	–44		mA
		$1.4 \le V_{OUT} < 2.4$[1, 2]	$-44 + (V_{OUT} - 1.4)/0.024$		mA
		$3.1 < V_{OUT} < V_{CC}$[1, 3]		EQ 1-1 on page 1-11	
	(Test Point)	$V_{OUT} = 3.1$[3]		–142	mA
$I_{OL(AC)}$	Switching Current High	$V_{OUT} \ge 2.2$[1]	95		mA
		$2.2 > V_{OUT} > 0.55$[1]	$V_{OUT}/0.023$		
		$0.71 > V_{OUT} > 0$[1, 3]		EQ 1-2 on page 1-11	mA
	(Test Point)	$V_{OUT} = 0.71$[3]		206	mA
I_{CL}	Low Clamp Current	$-5 < V_{IN} \le -1$	$-25 + (V_{IN} + 1)/0.015$		mA
$slew_R$	Output Rise Slew Rate	0.4 V to 2.4 V load[4]	1	5	V/ns
$slew_F$	Output Fall Slew Rate	2.4 V to 0.4 V load[4]	1	5	V/ns

Notes:

1. *Refer to the V/I curves in Figure 1-9 on page 1-11. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half-size output drivers may be used on these signals. This specification does not apply to CLK and RST#, which are system outputs. "Switching Current High" specifications are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD#, which are open drain outputs.*
2. *Note that this segment of the minimum current curve is drawn from the AC drive point directly to the DC drive point rather than toward the voltage rail (as is done in the pull-down curve). This difference is intended to allow for an optional N-channel pull-up.*
3. *Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (A and B) are provided with the respective diagrams in Figure 1-9 on page 1-11. The equation defined maxima should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.*
4. *This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per revision 2.0 of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is now required (the maximum is no longer simply a guideline). Since adherence to the maximum slew rate was not required prior to revision 2.1 of the specification, there may be components in the market for some time that have faster edge rates; therefore, motherboard designers must bear in mind that rise and fall times faster than this specification could occur, and should ensure that signal integrity modeling accounts for this. Rise slew rate does not apply to open drain outputs.*

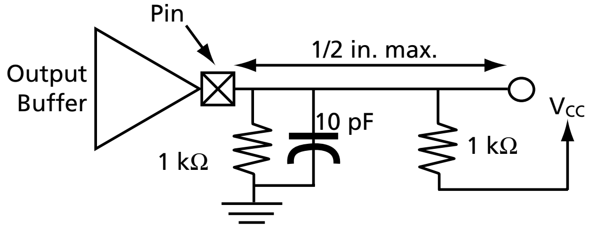

Figure 1-9 shows the 5.0 V PCI V/I curve and the minimum and maximum PCI drive characteristics of the A54SX16P device.

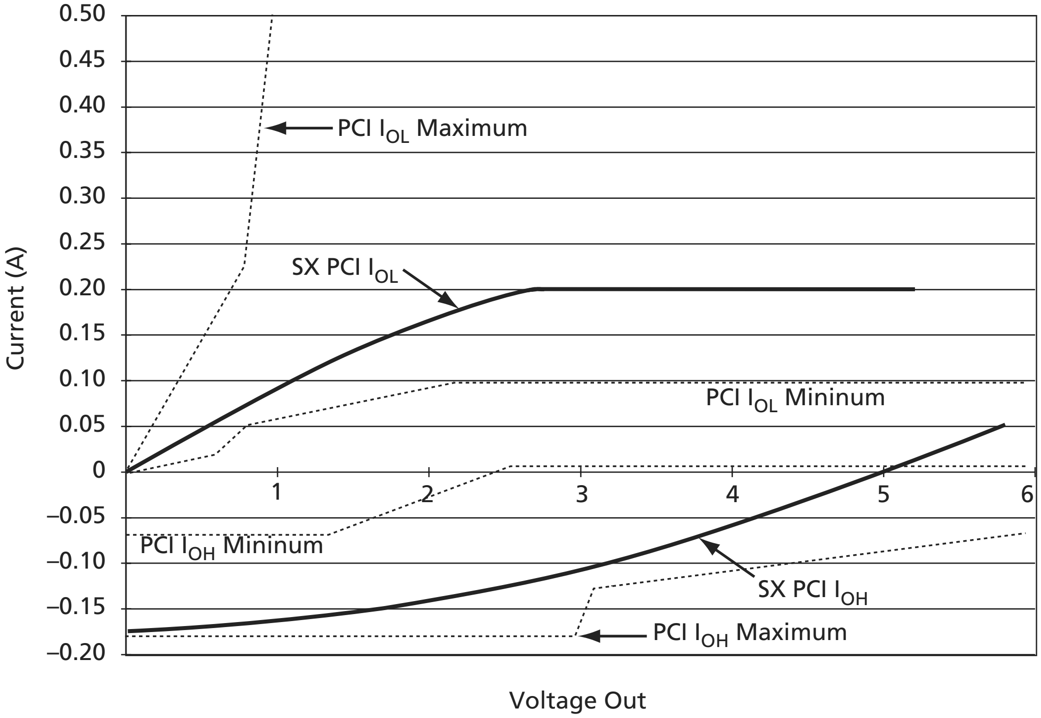

Figure 1-9 • **5.0 V PCI Curve for A54SX16P Device**

$I_{OH} = 11.9 \times (V_{OUT} - 5.25) \times (V_{OUT} + 2.45)$

for $V_{CC} > V_{OUT} > 3.1$ V

EQ 1-1

$I_{OL} = 78.5 \times V_{OUT} \times (4.4 - V_{OUT})$

for 0 V $< V_{OUT} <$ 0.71 V

EQ 1-2

Power-Up Sequencing

Table 1-10 • **Power-Up Sequencing**

V_{CCA}	V_{CCR}	V_{CCI}	Power-Up Sequence	Comments
A54SX08, A54SX16, A54SX32				
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	Possible damage to device
A54SX16P				
3.3 V	3.3 V	3.3 V	3.3 V Only	No possible damage to device
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	Possible damage to device
3.3 V	5.0 V	5.0 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device

Note: *No inputs should be driven (high or low) before completion of power-up.*

Power-Down Sequencing

Table 1-11 • **Power-Down Sequencing**

V_{CCA}	V_{CCR}	V_{CCI}	Power-Down Sequence	Comments
A54SX08, A54SX16, A54SX32				
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	Possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device
A54SX16P				
3.3 V	3.3 V	3.3 V	3.3 V Only	No possible damage to device
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	Possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device
3.3 V	5.0 V	5.0 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device

Note: *No inputs should be driven (high or low) after the beginning of the power-down sequence.*

Register Cell Timing Characteristics

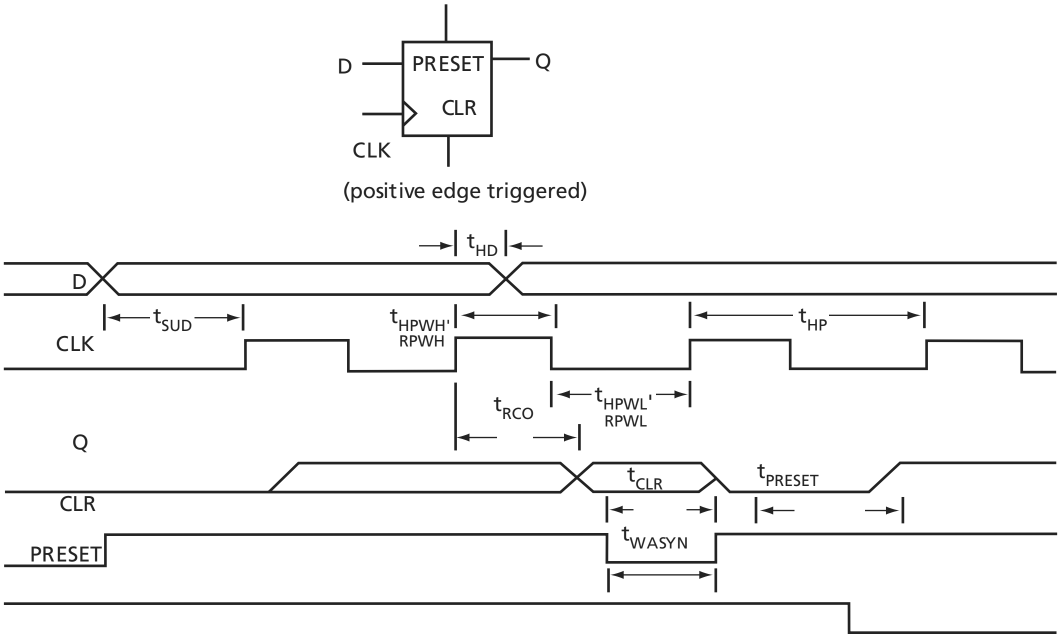

Figure 1-17 • **Flip-Flops**

Timing Characteristics

Timing characteristics for SX devices fall into three categories: family-dependent, device-dependent, and design-dependent. The input and output buffer characteristics are common to all SX family members. Internal routing delays are device-dependent. Design dependency means actual delays are not determined until after placement and routing of the user's design is complete. Delay values may then be determined by using the DirectTime Analyzer utility or performing simulation with post-layout delays.

Critical Nets and Typical Nets

Propagation delays are expressed only for typical nets, which are used for initial design performance evaluation. Critical net delays can then be applied to the most time-critical paths. Critical nets are determined by net property assignment prior to placement and routing. Up to 6% of the nets in a design may be designated as critical, while 90% of the nets in a design are typical.

Long Tracks

Some nets in the design use long tracks. Long tracks are special routing resources that span multiple rows, columns, or modules. Long tracks employ three and sometimes five antifuse connections. This increases capacitance and resistance, resulting in longer net delays for macros connected to long tracks. Typically up to 6 percent of nets in a fully utilized device require long tracks. Long tracks contribute approximately 4 ns to 8.4 ns delay. This additional delay is represented statistically in higher fanout (FO = 24) routing delays in the datasheet specifications section.

Timing Derating

SX devices are manufactured in a CMOS process. Therefore, device performance varies according to temperature, voltage, and process variations. Minimum timing parameters reflect maximum operating voltage, minimum operating temperature, and best-case processing. Maximum timing parameters reflect minimum operating voltage, maximum operating temperature, and worst-case processing.

A54SX16 Timing Characteristics

Table 1-18 • **A54SX16 Timing Characteristics**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

		'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		
Parameter	**Description**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
C-Cell Propagation Delays[1]										
t_{PD}	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted Routing Delays[2]										
t_{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t_{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{RD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{RD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timing										
t_{RCO}	Sequential Clock-to-Q		0.8		1.1		1.2		1.4	ns
t_{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t_{PRESET}	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t_{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t_{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t_{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t_{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted Input Routing Delays[2]										
t_{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{IRD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{IRD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Notes:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + $t_{PDn,}$ or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH}. For t_{ENZL} and $t_{ENZH,}$ the loading is 5 pF.*

Table 1-19 • **A54SX16P Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

		'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		
Parameter	**Description**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
Dedicated (Hardwired) Array Clock Network										
t_{HCKH}	Input LOW to HIGH (pad to R-Cell input)		1.2		1.4		1.5		1.8	ns
t_{HCKL}	Input HIGH to LOW (pad to R-Cell input)		1.2		1.4		1.6		1.9	ns
t_{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t_{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t_{HCKSW}	Maximum Skew		0.2		0.2		0.3		0.3	ns
t_{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f_{HMAX}	Maximum Frequency		350		320		280		240	MHz
Routed Array Clock Networks										
t_{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)		1.6		1.8		2.1		2.5	ns
t_{RCKL}	Input HIGH to LOW (Light Load) (pad to R-Cell input)		1.8		2.0		2.3		2.7	ns
t_{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)		1.8		2.1		2.5		2.8	ns
t_{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)		1.8		2.1		2.4		2.8	ns
t_{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t_{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t_{RCKSW}	Maximum Skew (light load)		0.5		0.5		0.5		0.7	ns
t_{RCKSW}	Maximum Skew (50% load)		0.5		0.6		0.7		0.8	ns
t_{RCKSW}	Maximum Skew (100% load)		0.5		0.6		0.7		0.8	ns
TTL Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		2.4		2.8		3.1		3.7	ns
t_{DHL}	Data-to-Pad HIGH to LOW		2.3		2.9		3.2		3.8	ns
t_{ENZL}	Enable-to-Pad, Z to L		3.0		3.4		3.9		4.6	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.3		3.8		4.3		5.0	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.3		2.7		3.0		3.5	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.8		3.2		3.7		4.3	ns

Note:

1. *For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 10 pF loading.*

Table 1-19 • **A54SX16P Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

		'–3' Speed		**'–2' Speed**		**'–1' Speed**		**'Std' Speed**		
Parameter	**Description**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
TTL/PCI Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		1.5		1.7		2.0		2.3	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.9		2.2		2.4		2.9	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.3		2.6		3.0		3.5	ns
t_{ENZH}	Enable-to-Pad, Z to H		1.5		1.7		1.9		2.3	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.7		3.1		3.5		4.1	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.9		3.3		3.7		4.4	ns
PCI Output Module Timing[3]										
t_{DLH}	Data-to-Pad LOW to HIGH		1.8		2.0		2.3		2.7	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.7		2.0		2.2		2.6	ns
t_{ENZL}	Enable-to-Pad, Z to L		0.8		1.0		1.1		1.3	ns
t_{ENZH}	Enable-to-Pad, Z to H		1.2		1.2		1.5		1.8	ns
t_{ENLZ}	Enable-to-Pad, L to Z		1.0		1.1		1.3		1.5	ns
t_{ENHZ}	Enable-to-Pad, H to Z		1.1		1.3		1.5		1.7	ns
TTL Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		2.1		2.5		2.8		3.3	ns
t_{DHL}	Data-to-Pad HIGH to LOW		2.0		2.3		2.6		3.1	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.5		2.9		3.2		3.8	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.0		3.5		3.9		4.6	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.3		2.7		3.1		3.6	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.9		3.3		3.7		4.4	ns

Note:

1. *For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 10 pF loading.*

Package Pin Assignments

84-Pin PLCC

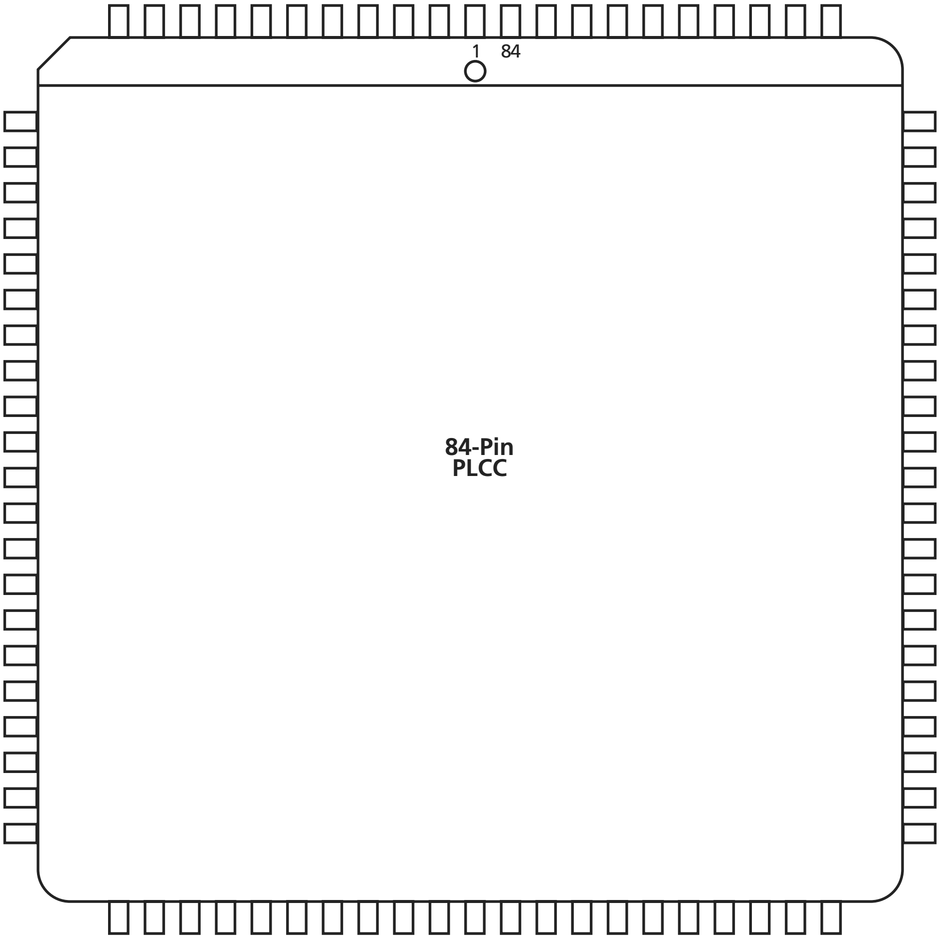

Figure 2-1 • **84-Pin PLCC (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

208-Pin PQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
73	NC	I/O	I/O
74	I/O	I/O	I/O
75	NC	I/O	I/O
76	PRB, I/O	PRB, I/O	PRB, I/O
77	GND	GND	GND
78	V_{CCA}	V_{CCA}	V_{CCA}
79	GND	GND	GND
80	V_{CCR}	V_{CCR}	V_{CCR}
81	I/O	I/O	I/O
82	HCLK	HCLK	HCLK
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	NC	I/O	I/O
86	I/O	I/O	I/O
87	I/O	I/O	I/O
88	NC	I/O	I/O
89	I/O	I/O	I/O
90	I/O	I/O	I/O
91	NC	I/O	I/O
92	I/O	I/O	I/O
93	I/O	I/O	I/O
94	NC	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	NC	I/O	I/O
98	V_{CCI}	V_{CCI}	V_{CCI}
99	I/O	I/O	I/O
100	I/O	I/O	I/O
101	I/O	I/O	I/O
102	I/O	I/O	I/O
103	TDO, I/O	TDO, I/O	TDO, I/O
104	I/O	I/O	I/O
105	GND	GND	GND
106	NC	I/O	I/O
107	I/O	I/O	I/O
108	NC	I/O	I/O

208-Pin PQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
109	I/O	I/O	I/O
110	I/O	I/O	I/O
111	I/O	I/O	I/O
112	I/O	I/O	I/O
113	I/O	I/O	I/O
114	V_{CCA}	V_{CCA}	V_{CCA}
115	V_{CCI}	V_{CCI}	V_{CCI}
116	NC	I/O	I/O
117	I/O	I/O	I/O
118	I/O	I/O	I/O
119	NC	I/O	I/O
120	I/O	I/O	I/O
121	I/O	I/O	I/O
122	NC	I/O	I/O
123	I/O	I/O	I/O
124	I/O	I/O	I/O
125	NC	I/O	I/O
126	I/O	I/O	I/O
127	I/O	I/O	I/O
128	I/O	I/O	I/O
129	GND	GND	GND
130	V_{CCA}	V_{CCA}	V_{CCA}
131	GND	GND	GND
132	V_{CCR}	V_{CCR}	V_{CCR}
133	I/O	I/O	I/O
134	I/O	I/O	I/O
135	NC	I/O	I/O
136	I/O	I/O	I/O
137	I/O	I/O	I/O
138	NC	I/O	I/O
139	I/O	I/O	I/O
140	I/O	I/O	I/O
141	NC	I/O	I/O
142	I/O	I/O	I/O
143	NC	I/O	I/O
144	I/O	I/O	I/O

Note: ** Note that Pin 65 in the A54SX32—PQ208 is a no connect (NC).*

100-Pin VQFP

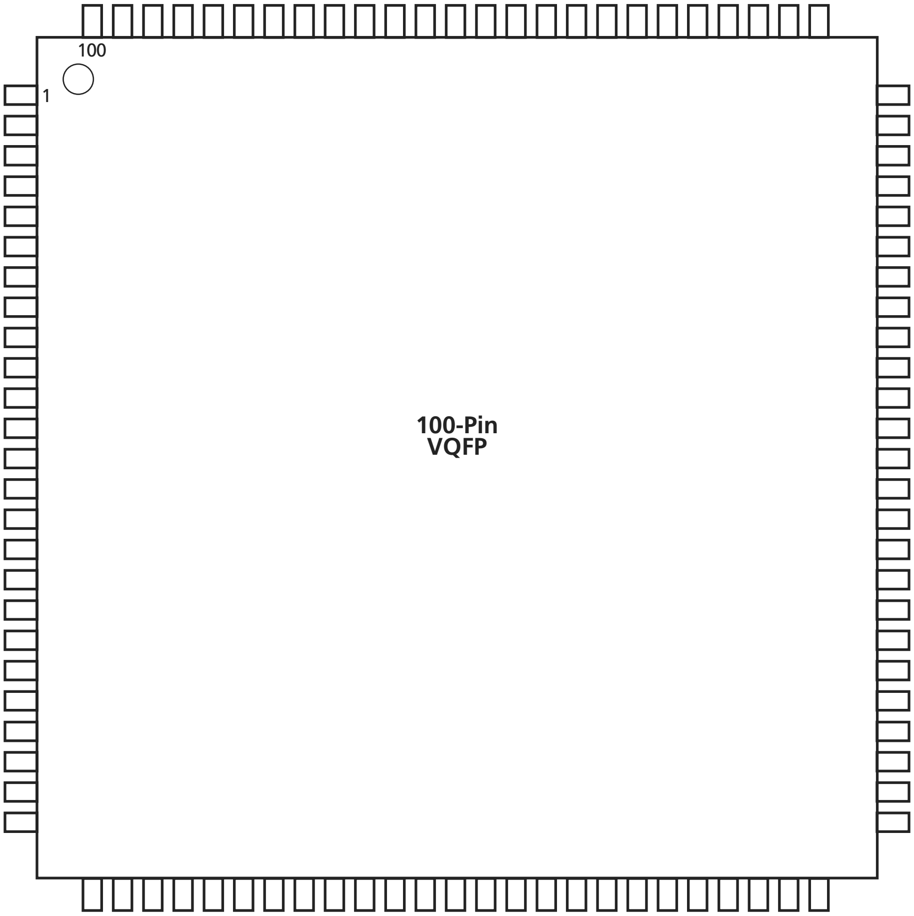

Figure 2-5 • **100-Pin VQFP (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

329-Pin PBGA

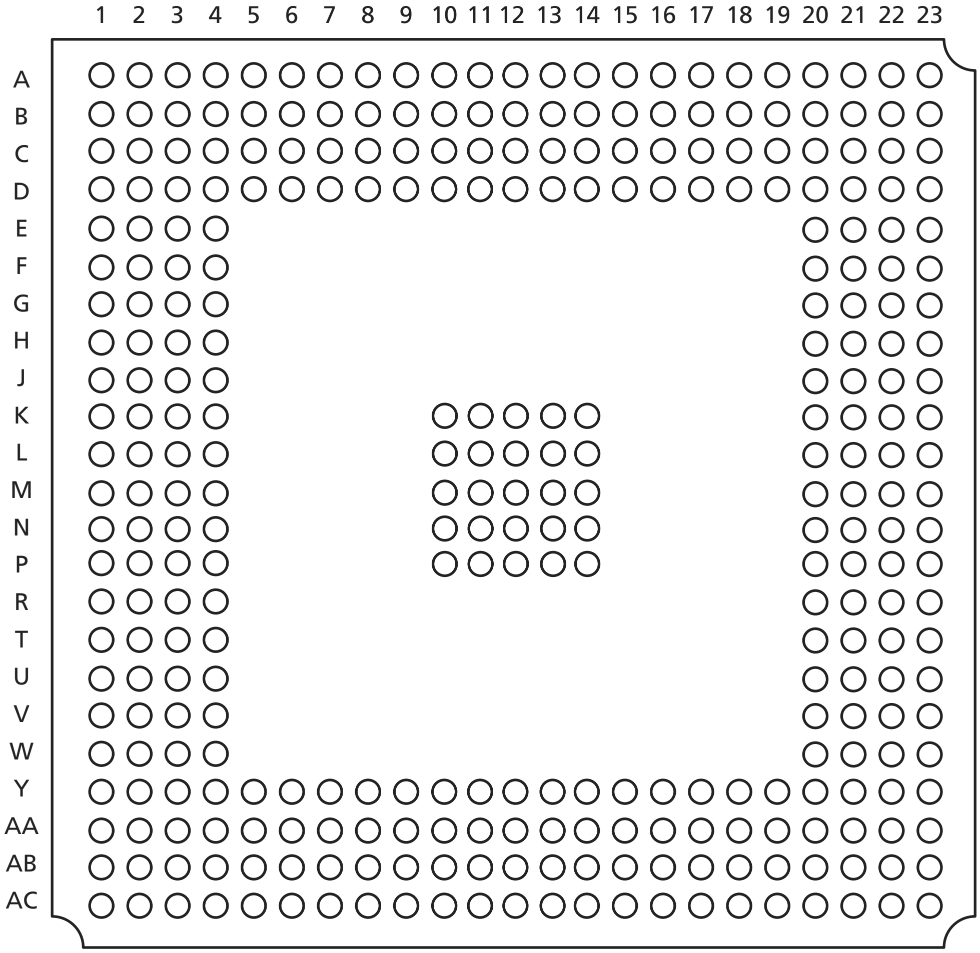

Figure 2-7 • **329-Pin PBGA (Top View)**

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

144-Pin FBGA	
Pin Number	**A54SX08 Function**
A1	I/O
A2	I/O
A3	I/O
A4	I/O
A5	V_{CCA}
A6	GND
A7	CLKA
A8	I/O
A9	I/O
A10	I/O
A11	I/O
A12	I/O
B1	I/O
B2	GND
B3	I/O
B4	I/O
B5	I/O
B6	I/O
B7	CLKB
B8	I/O
B9	I/O
B10	I/O
B11	GND
B12	I/O
C1	I/O
C2	I/O
C3	TCK, I/O
C4	I/O
C5	I/O
C6	PRA, I/O
C7	I/O
C8	I/O
C9	I/O
C10	I/O
C11	I/O
C12	I/O

144-Pin FBGA	
Pin Number	**A54SX08 Function**
D1	I/O
D2	V_{CCI}
D3	TDI, I/O
D4	I/O
D5	I/O
D6	I/O
D7	I/O
D8	I/O
D9	I/O
D10	I/O
D11	I/O
D12	I/O
E1	I/O
E2	I/O
E3	I/O
E4	I/O
E5	TMS
E6	V_{CCI}
E7	V_{CCI}
E8	V_{CCI}
E9	V_{CCA}
E10	I/O
E11	GND
E12	I/O
F1	I/O
F2	I/O
F3	V_{CCR}
F4	I/O
F5	GND
F6	GND
F7	GND
F8	V_{CCI}
F9	I/O
F10	GND
F11	I/O
F12	I/O

144-Pin FBGA	
Pin Number	**A54SX08 Function**
G1	I/O
G2	GND
G3	I/O
G4	I/O
G5	GND
G6	GND
G7	GND
G8	V_{CCI}
G9	I/O
G10	I/O
G11	I/O
G12	I/O
H1	I/O
H2	I/O
H3	I/O
H4	I/O
H5	V_{CCA}
H6	V_{CCA}
H7	V_{CCI}
H8	V_{CCI}
H9	V_{CCA}
H10	I/O
H11	I/O
H12	V_{CCR}
J1	I/O
J2	I/O
J3	I/O
J4	I/O
J5	I/O
J6	PRB, I/O
J7	I/O
J8	I/O
J9	I/O
J10	I/O
J11	I/O
J12	V_{CCA}

144-Pin FBGA	
Pin Number	**A54SX08 Function**
K1	I/O
K2	I/O
K3	I/O
K4	I/O
K5	I/O
K6	I/O
K7	GND
K8	I/O
K9	I/O
K10	GND
K11	I/O
K12	I/O
L1	GND
L2	I/O
L3	I/O
L4	I/O
L5	I/O
L6	I/O
L7	HCLK
L8	I/O
L9	I/O
L10	I/O
L11	I/O
L12	I/O
M1	I/O
M2	I/O
M3	I/O
M4	I/O
M5	I/O
M6	I/O
M7	V_{CCA}
M8	I/O
M9	I/O
M10	I/O
M11	TDO, I/O
M12	I/O

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