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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

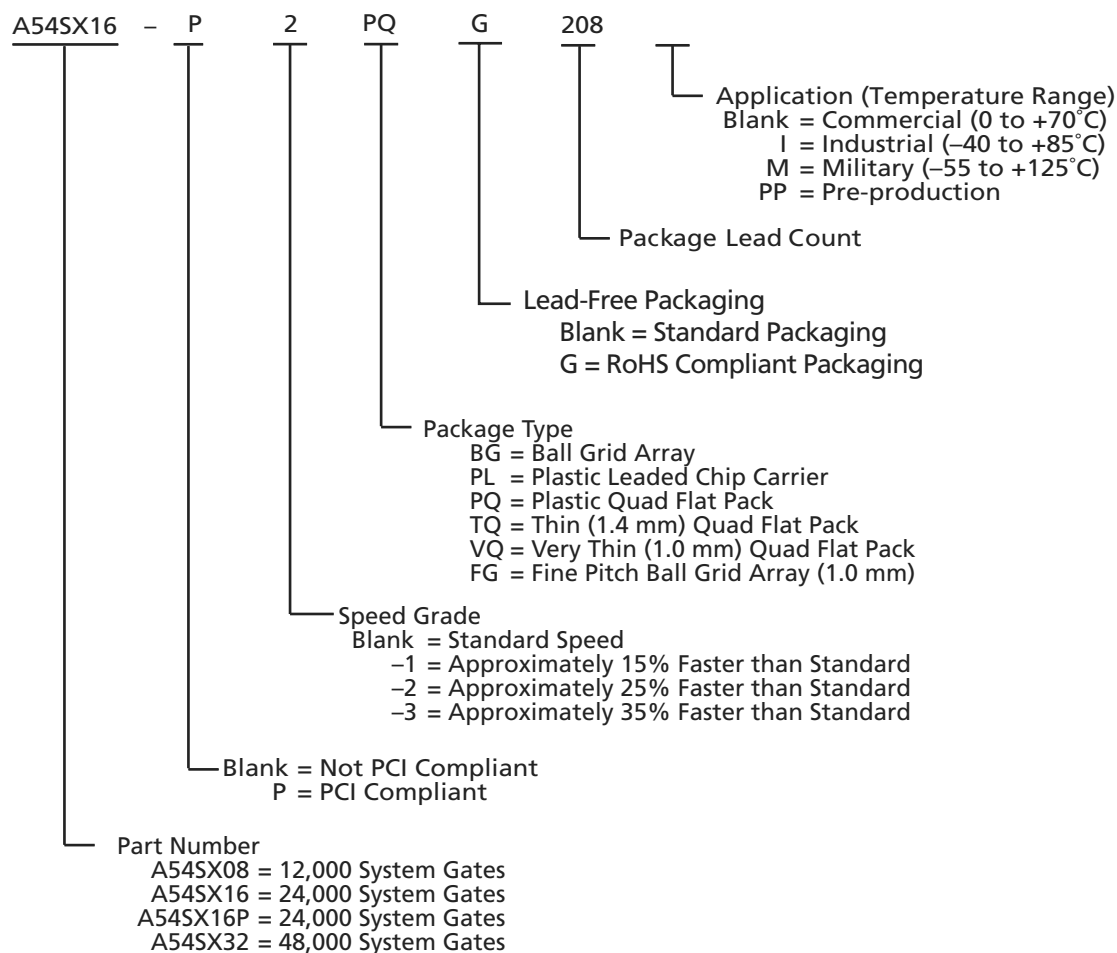
Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	768
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	81
Number of Gates	12000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	100-TQFP
Supplier Device Package	100-VQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx08-vq100

Ordering Information



Plastic Device Resources

Device	User I/Os (including clock buffers)							
	PLCC 84-Pin	VQFP 100-Pin	PQFP 208-Pin	TQFP 144-Pin	TQFP 176-Pin	PBGA 313-Pin	PBGA 329-Pin	FBGA 144-Pin
A54SX08	69	81	130	113	128	–	–	111
A54SX16	–	81	175	–	147	–	–	–
A54SX16P	–	81	175	113	147	–	–	–
A54SX32	–	–	174	113	147	249	249	–

Note: Package Definitions (Consult your local Actel sales representative for product availability):

PLCC = Plastic Leaded Chip Carrier

PQFP = Plastic Quad Flat Pack

TQFP = Thin Quad Flat Pack

VQFP = Very Thin Quad Flat Pack

PBGA = Plastic Ball Grid Array

FBGA = Fine Pitch (1.0 mm) Ball Grid Array

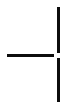


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SX Family FPGAs

General Description

The Actel SX family of FPGAs features a sea-of-modules architecture that delivers device performance and integration levels not currently achieved by any other FPGA architecture. SX devices greatly simplify design time, enable dramatic reductions in design costs and power consumption, and further decrease time to market for performance-intensive applications.

The Actel SX architecture features two types of logic modules, the combinatorial cell (C-cell) and the register cell (R-cell), each optimized for fast and efficient mapping of synthesized logic functions. The routing and interconnect resources are in the metal layers above the logic modules, providing optimal use of silicon. This enables the entire floor of the device to be spanned with an uninterrupted grid of fine-grained, synthesis-friendly logic modules (or “sea-of-modules”), which reduces the distance signals have to travel between logic modules. To minimize signal propagation delay, SX devices employ both local and general routing resources. The high-speed local routing resources (DirectConnect and FastConnect) enable very fast local signal propagation that is optimal for fast counters, state machines, and datapath logic. The general system of segmented routing tracks allows any logic module in the array to be connected to any other logic or I/O module. Within this system, propagation delay is minimized by limiting the number of antifuse interconnect elements to five (90 percent of connections typically use only three antifuses). The unique local and general routing structure featured in SX devices gives fast and predictable performance, allows 100 percent pin-locking with full logic utilization, enables concurrent PCB development, reduces design time, and allows designers to achieve performance goals with minimum effort.

Further complementing SX’s flexible routing structure is a hardwired, constantly loaded clock network that has been tuned to provide fast clock propagation with minimal clock skew. Additionally, the high performance of the internal logic has eliminated the need to embed latches or flip-flops in the I/O cells to achieve fast clock-to-out or fast input setup times. SX devices have easy to use I/O cells that do not require HDL instantiation, facilitating design reuse and reducing design and verification time.

SX Family Architecture

The SX family architecture was designed to satisfy next-generation performance and integration requirements for production-volume designs in a broad range of applications.

Programmable Interconnect Element

The SX family provides efficient use of silicon by locating the routing interconnect resources between the Metal 2 (M2) and Metal 3 (M3) layers (Figure 1-1 on page 1-2). This completely eliminates the channels of routing and interconnect resources between logic modules (as implemented on SRAM FPGAs and previous generations of antifuse FPGAs), and enables the entire floor of the device to be spanned with an uninterrupted grid of logic modules.

Interconnection between these logic modules is achieved using The Actel patented metal-to-metal programmable antifuse interconnect elements, which are embedded between the M2 and M3 layers. The antifuses are normally open circuit and, when programmed, form a permanent low-impedance connection.

The extremely small size of these interconnect elements gives the SX family abundant routing resources and provides excellent protection against design pirating. Reverse engineering is virtually impossible because it is extremely difficult to distinguish between programmed and unprogrammed antifuses, and there is no configuration bitstream to intercept.

Additionally, the interconnect elements (i.e., the antifuses and metal tracks) have lower capacitance and lower resistance than any other device of similar capacity, leading to the fastest signal propagation in the industry.

Logic Module Design

The SX family architecture is described as a “sea-of-modules” architecture because the entire floor of the device is covered with a grid of logic modules with virtually no chip area lost to interconnect elements or routing. The Actel SX family provides two types of logic modules, the register cell (R-cell) and the combinatorial cell (C-cell).

The R-cell contains a flip-flop featuring asynchronous clear, asynchronous preset, and clock enable (using the S0 and S1 lines) control signals (Figure 1-2). The R-cell registers feature programmable clock polarity selectable on a register-by-register basis. This provides additional

flexibility while allowing mapping of synthesized functions into the SX FPGA. The clock source for the R-cell can be chosen from either the hardwired clock or the routed clock.

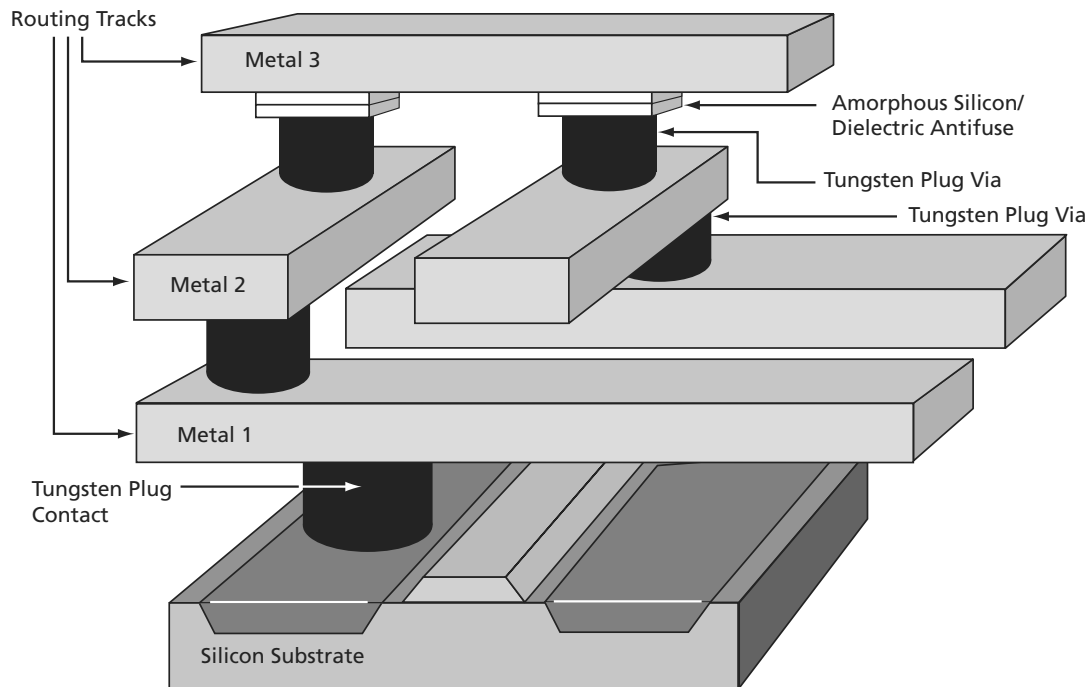


Figure 1-1 • SX Family Interconnect Elements

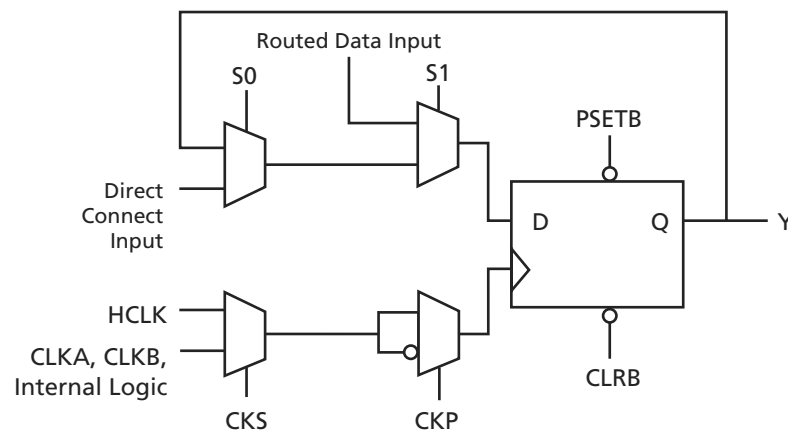


Figure 1-2 • R-Cell

The C-cell implements a range of combinatorial functions up to 5-inputs (Figure 1-3 on page 1-3). Inclusion of the DB input and its associated inverter function dramatically increases the number of combinatorial functions that can be implemented in a single module from 800 options in previous architectures to more than 4,000 in the SX architecture. An example of the improved flexibility

enabled by the inversion capability is the ability to integrate a 3-input exclusive-OR function into a single C-cell. This facilitates construction of 9-bit parity-tree functions with 2 ns propagation delays. At the same time, the C-cell structure is extremely synthesis friendly, simplifying the overall design and reducing synthesis time.

Chip Architecture

The SX family chip architecture provides a unique approach to module organization and chip routing that delivers the best register/logic mix for a wide variety of new and emerging applications.

Module Organization

Actel has arranged all C-cell and R-cell logic modules into horizontal banks called *clusters*. There are two types of *clusters*: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells.

To increase design efficiency and device performance, Actel has further organized these modules into *SuperClusters* (Figure 1-4). SuperCluster 1 is a two-wide grouping of Type 1 clusters. SuperCluster 2 is a two-wide group containing one Type 1 cluster and one Type 2 cluster. SX devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flip-flops.

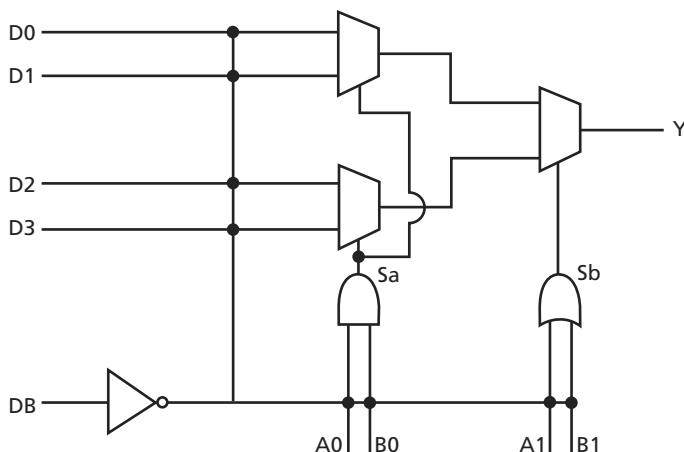


Figure 1-3 • C-Cell

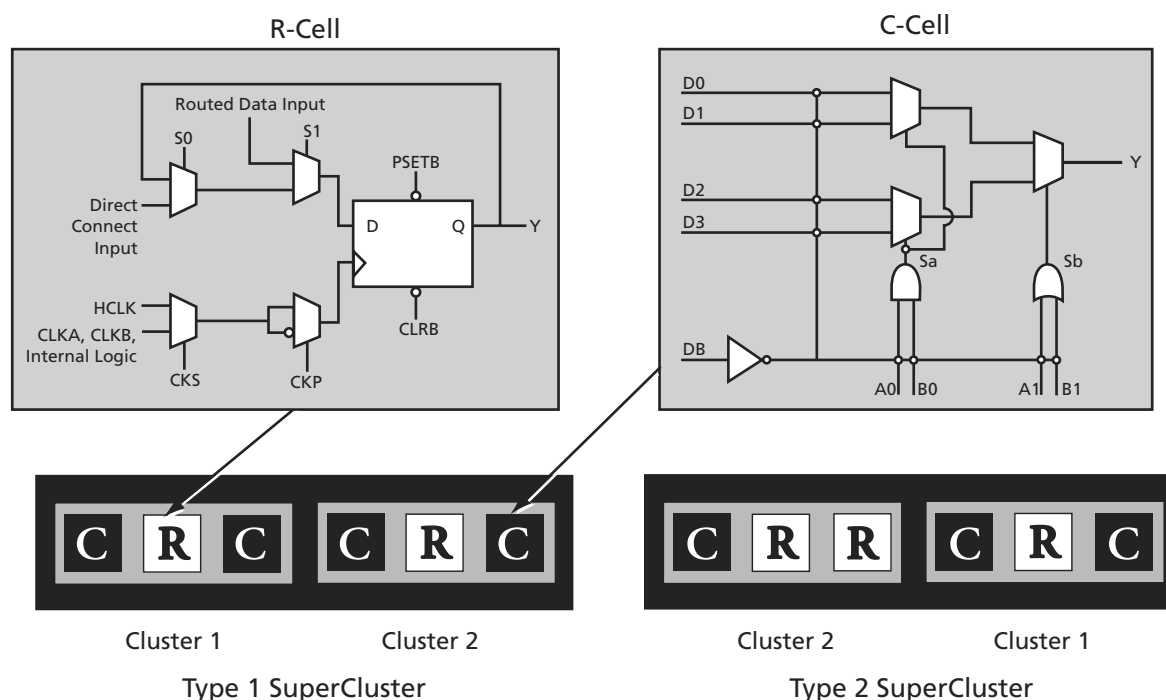


Figure 1-4 • Cluster Organization

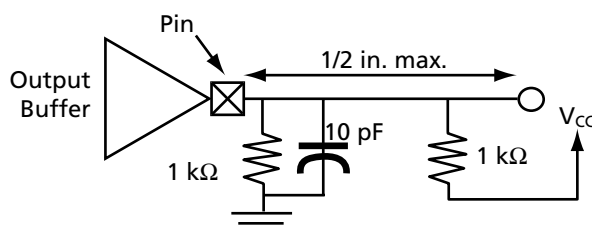
A54SX16P AC Specifications (3.3 V PCI Operation)

Table 1-9 • A54SX16P AC Specifications (3.3 V PCI Operation)

Symbol	Parameter	Condition	Min.	Max.	Units
$I_{OH(AC)}$	Switching Current High	$0 < V_{OUT} \leq 0.3V_{CC}^1$			mA
		$0.3V_{CC} \leq V_{OUT} < 0.9V_{CC}^1$	$-12V_{CC}$		mA
		$0.7V_{CC} < V_{OUT} < V_{CC}^{1,2}$	$-17.1 + (V_{CC} - V_{OUT})$	EQ 1-3 on page 1-14	
	(Test Point)	$V_{OUT} = 0.7V_{CC}^2$		$-32V_{CC}$	mA
$I_{OL(AC)}$	Switching Current High	$V_{CC} > V_{OUT} \geq 0.6V_{CC}^1$			mA
		$0.6V_{CC} > V_{OUT} > 0.1V_{CC}^1$	$16V_{CC}$		mA
		$0.18V_{CC} > V_{OUT} > 0^{1,2}$	$26.7V_{OUT}$	EQ 1-4 on page 1-14	mA
	(Test Point)	$V_{OUT} = 0.18V_{CC}^2$		$38V_{CC}$	
I_{CL}	Low Clamp Current	$-3 < V_{IN} \leq -1$	$-25 + (V_{IN} + 1)/0.015$		mA
I_{CH}	High Clamp Current	$-3 < V_{IN} \leq -1$	$25 + (V_{IN} - V_{OUT} - 1)/0.015$		mA
$slew_R$	Output Rise Slew Rate ³	0.2V _{CC} to 0.6V _{CC} load	1	4	V/ns
$slew_F$	Output Fall Slew Rate ³	0.6V _{CC} to 0.2V _{CC} load	1	4	V/ns

Notes:

1. Refer to the V/I curves in Figure 1-10 on page 1-14. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half size output drivers may be used on these signals. This specification does not apply to CLK and RST# which are system outputs. "Switching Current High" specification are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD# which are open drain outputs.
2. Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (C and D) are provided with the respective diagrams in Figure 1-10 on page 1-14. The equation defined maxima should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.
3. This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per the latest revision of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is required (the maximum is no longer simply a guideline). Rise slew rate does not apply to open drain outputs.



Power-Up Sequencing

Table 1-10 • Power-Up Sequencing

V _{CCA}	V _{CCR}	V _{CCI}	Power-Up Sequence	Comments
A54SX08, A54SX16, A54SX32				
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	Possible damage to device
A54SX16P				
3.3 V	3.3 V	3.3 V	3.3 V Only	No possible damage to device
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	Possible damage to device
3.3 V	5.0 V	5.0 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device

Note: No inputs should be driven (high or low) before completion of power-up.

Power-Down Sequencing

Table 1-11 • Power-Down Sequencing

V _{CCA}	V _{CCR}	V _{CCI}	Power-Down Sequence	Comments
A54SX08, A54SX16, A54SX32				
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	Possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device
A54SX16P				
3.3 V	3.3 V	3.3 V	3.3 V Only	No possible damage to device
3.3 V	5.0 V	3.3 V	5.0 V First 3.3 V Second	Possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device
3.3 V	5.0 V	5.0 V	5.0 V First 3.3 V Second	No possible damage to device
			3.3 V First 5.0 V Second	No possible damage to device

Note: No inputs should be driven (high or low) after the beginning of the power-down sequence.

Table 1-13 shows capacitance values for various devices.

Table 1-13 • Capacitance Values for Devices

	A545X08	A545X16	A545X16P	A545X32
C_{EQM} (pF)	4.0	4.0	4.0	4.0
C_{EQI} (pF)	3.4	3.4	3.4	3.4
C_{EQO} (pF)	4.7	4.7	4.7	4.7
C_{EQCR} (pF)	1.6	1.6	1.6	1.6
C_{EQHV}	0.615	0.615	0.615	0.615
C_{EQHF}	60	96	96	140
r_1 (pF)	87	138	138	171
r_2 (pF)	87	138	138	171

Table 1-14 • Power Consumption Guidelines

Description	Power Consumption Guideline
Logic Modules (m)	20% of modules
Inputs Switching (n)	# inputs/4
Outputs Switching (p)	# outputs/4
First Routed Array Clock Loads (q_1)	20% of register cells
Second Routed Array Clock Loads (q_2)	20% of register cells
Load Capacitance (C_L)	35 pF
Average Logic Module Switching Rate (f_m)	$f/10$
Average Input Switching Rate (f_n)	$f/5$
Average Output Switching Rate (f_p)	$f/10$
Average First Routed Array Clock Rate (f_{q1})	$f/2$
Average Second Routed Array Clock Rate (f_{q2})	$f/2$
Average Dedicated Array Clock Rate (f_{s1})	f
Dedicated Clock Array Clock Loads (s_1)	20% of regular modules

Follow the steps below to estimate power consumption. The values provided for the sample calculation below are for the shift register design above. This method for estimating power consumption is conservative and the actual power consumption of your design may be less than the estimated power consumption.

The total power dissipation for the SX family is the sum of the AC power dissipation and the DC power dissipation.

$$P_{\text{Total}} = P_{\text{AC}} (\text{dynamic power}) + P_{\text{DC}} (\text{static power})$$

EQ 1-9

Guidelines for Calculating Power Consumption

The power consumption guidelines are meant to represent worst-case scenarios so that they can be generally used to predict the upper limits of power dissipation. These guidelines are shown in Table 1-14.

Sample Power Calculation

One of the designs used to characterize the SX family was a 528 bit serial-in, serial-out shift register. The design utilized 100 percent of the dedicated flip-flops of an A545X16P device. A pattern of 0101... was clocked into the device at frequencies ranging from 1 MHz to 200 MHz. Shifting in a series of 0101... caused 50 percent of the flip-flops to toggle from low to high at every clock cycle.

AC Power Dissipation

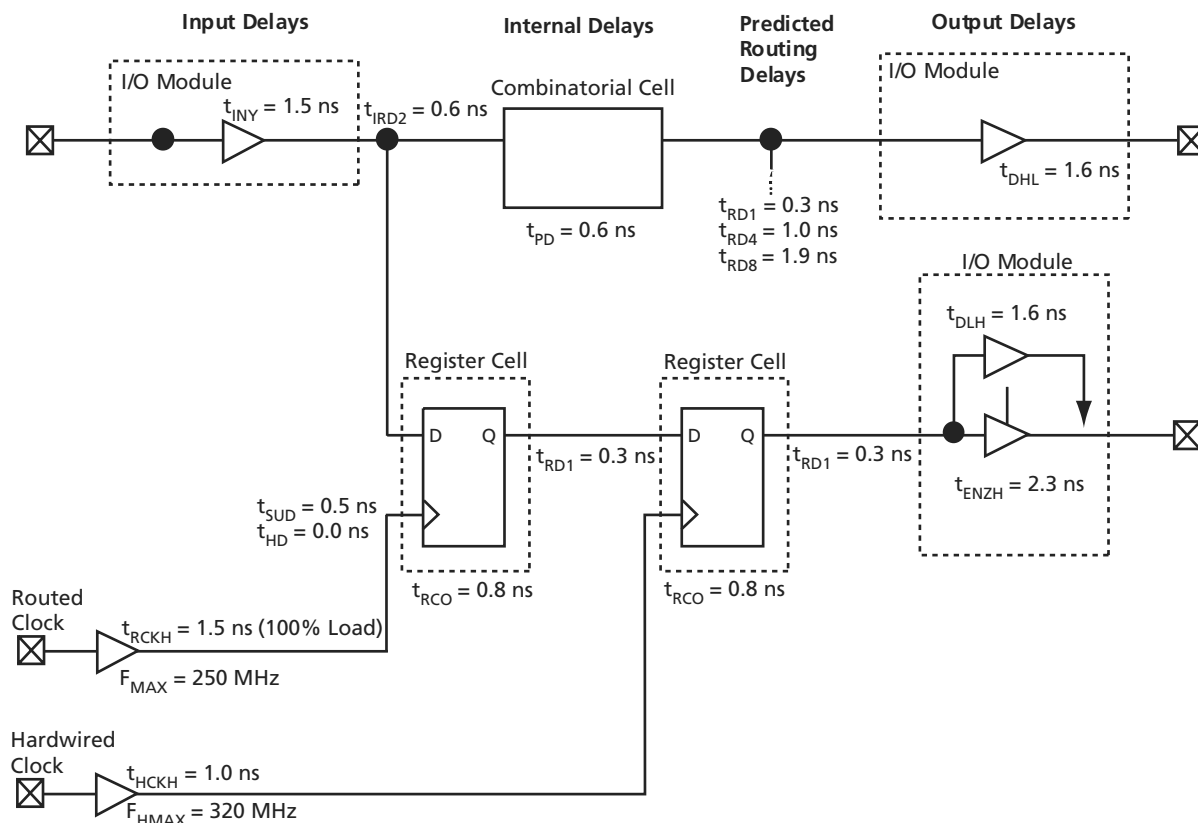
$$P_{\text{AC}} = P_{\text{Module}} + P_{\text{RCLKA Net}} + P_{\text{RCLKB Net}} + P_{\text{HCLK Net}} + P_{\text{Output Buffer}} + P_{\text{Input Buffer}}$$

EQ 1-10

$$P_{\text{AC}} = V_{\text{CCA}}^2 \times [(m \times C_{\text{EQM}} \times f_m)_{\text{Module}} + (n \times C_{\text{EQI}} \times f_n)_{\text{Input Buffer}} + (p \times (C_{\text{EQO}} + C_L) \times f_p)_{\text{Output Buffer}} + (0.5 (q_1 \times C_{\text{EQCR}} \times f_{q1}) + (r_1 \times f_{q1}))_{\text{RCLKA}} + (0.5 (q_2 \times C_{\text{EQCR}} \times f_{q2}) + (r_2 \times f_{q2}))_{\text{RCLKB}} + (0.5 (s_1 \times C_{\text{EQHV}} \times f_{s1}) + (C_{\text{EQHF}} \times f_{s1}))_{\text{HCLK}}]$$

EQ 1-11

SX Timing Model



Note: Values shown for A54SX08-3, worst-case commercial conditions.

Figure 1-12 • SX Timing Model

Hardwired Clock

$$\begin{aligned} \text{External Setup} &= t_{INY} + t_{IRD1} + t_{SUD} - t_{HCKH} \\ &= 1.5 + 0.3 + 0.5 - 1.0 = 1.3 \text{ ns} \end{aligned}$$

EQ 1-15

Clock-to-Out (Pin-to-Pin)

$$\begin{aligned} &= t_{HCKH} + t_{RCO} + t_{RD1} + t_{DHL} \\ &= 1.0 + 0.8 + 0.3 + 1.6 = 3.7 \text{ ns} \end{aligned}$$

EQ 1-16

Routed Clock

$$\begin{aligned} \text{External Setup} &= t_{INY} + t_{IRD1} + t_{SUD} - t_{RCKH} \\ &= 1.5 + 0.3 + 0.5 - 1.5 = 0.8 \text{ ns} \end{aligned}$$

EQ 1-17

Clock-to-Out (Pin-to-Pin)

$$\begin{aligned} &= t_{RCKH} + t_{RCO} + t_{RD1} + t_{DHL} \\ &= 1.52 + 0.8 + 0.3 + 1.6 = 4.2 \text{ ns} \end{aligned}$$

EQ 1-18

Register Cell Timing Characteristics

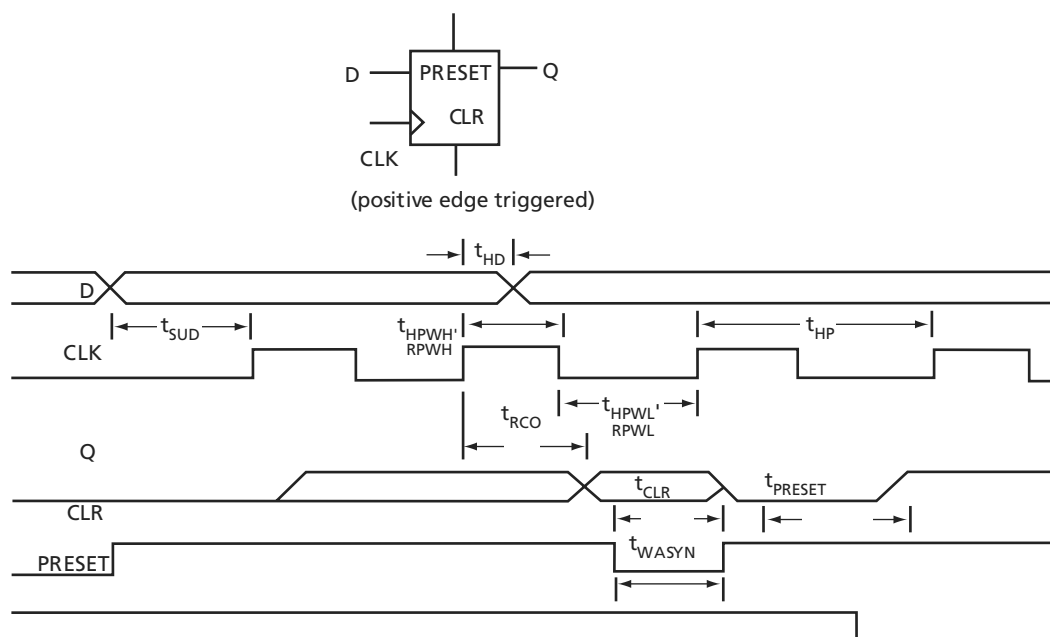


Figure 1-17 • Flip-Flops

Timing Characteristics

Timing characteristics for SX devices fall into three categories: family-dependent, device-dependent, and design-dependent. The input and output buffer characteristics are common to all SX family members. Internal routing delays are device-dependent. Design dependency means actual delays are not determined until after placement and routing of the user's design is complete. Delay values may then be determined by using the DirectTime Analyzer utility or performing simulation with post-layout delays.

Critical Nets and Typical Nets

Propagation delays are expressed only for typical nets, which are used for initial design performance evaluation. Critical net delays can then be applied to the most time-critical paths. Critical nets are determined by net property assignment prior to placement and routing. Up to 6% of the nets in a design may be designated as critical, while 90% of the nets in a design are typical.

Long Tracks

Some nets in the design use long tracks. Long tracks are special routing resources that span multiple rows, columns, or modules. Long tracks employ three and sometimes five antifuse connections. This increases capacitance and resistance, resulting in longer net delays for macros connected to long tracks. Typically up to 6 percent of nets in a fully utilized device require long tracks. Long tracks contribute approximately 4 ns to 8.4 ns delay. This additional delay is represented statistically in higher fanout ($FO = 24$) routing delays in the datasheet specifications section.

Timing Derating

SX devices are manufactured in a CMOS process. Therefore, device performance varies according to temperature, voltage, and process variations. Minimum timing parameters reflect maximum operating voltage, minimum operating temperature, and best-case processing. Maximum timing parameters reflect minimum operating voltage, maximum operating temperature, and worst-case processing.

A54SX08 Timing Characteristics

Table 1-17 • A54SX08 Timing Characteristics
(Worst-Case Commercial Conditions, $V_{CCR} = 4.75\text{ V}$, $V_{CCA}, V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	'-3' Speed		'-2' Speed		'-1' Speed		'Std' Speed		
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
C-Cell Propagation Delays ¹										
t _{PD}	Internal Array Module	0.6		0.7		0.8		0.9		ns
Predicted Routing Delays ²										
t _{DC}	FO = 1 Routing Delay, Direct Connect	0.1		0.1		0.1		0.1		ns
t _{FC}	FO = 1 Routing Delay, Fast Connect	0.3		0.4		0.4		0.5		ns
t _{RD1}	FO = 1 Routing Delay	0.3		0.4		0.4		0.5		ns
t _{RD2}	FO = 2 Routing Delay	0.6		0.7		0.8		0.9		ns
t _{RD3}	FO = 3 Routing Delay	0.8		0.9		1.0		1.2		ns
t _{RD4}	FO = 4 Routing Delay	1.0		1.2		1.4		1.6		ns
t _{RD8}	FO = 8 Routing Delay	1.9		2.2		2.5		2.9		ns
t _{RD12}	FO = 12 Routing Delay	2.8		3.2		3.7		4.3		ns
R-Cell Timing										
t _{RCO}	Sequential Clock-to-Q	0.8		1.1		1.2		1.4		ns
t _{CLR}	Asynchronous Clear-to-Q	0.5		0.6		0.7		0.8		ns
t _{PRESET}	Asynchronous Preset-to-Q	0.7		0.8		0.9		1.0		ns
t _{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t _{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t _{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t _{INYH}	Input Data Pad-to-Y HIGH	1.5		1.7		1.9		2.2		ns
t _{INYL}	Input Data Pad-to-Y LOW	1.5		1.7		1.9		2.2		ns
Input Module Predicted Routing Delays ²										
t _{IRD1}	FO = 1 Routing Delay	0.3		0.4		0.4		0.5		ns
t _{IRD2}	FO = 2 Routing Delay	0.6		0.7		0.8		0.9		ns
t _{IRD3}	FO = 3 Routing Delay	0.8		0.9		1.0		1.2		ns
t _{IRD4}	FO = 4 Routing Delay	1.0		1.2		1.4		1.6		ns
t _{IRD8}	FO = 8 Routing Delay	1.9		2.2		2.5		2.9		ns
t _{IRD12}	FO = 12 Routing Delay	2.8		3.2		3.7		4.3		ns

Note:

- For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.
- Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

Table 1-18 • A54SX16 Timing Characteristics (Continued)
(Worst-Case Commercial Conditions, $V_{CCR} = 4.75\text{ V}$, $V_{CCA}, V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	'-3' Speed		'-2' Speed		'-1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Dedicated (Hardwired) Array Clock Network										
t _{HCKH}	Input LOW to HIGH (pad to R-Cell input)	1.2		1.4		1.5		1.8		ns
t _{HCKL}	Input HIGH to LOW (pad to R-Cell input)	1.2		1.4		1.6		1.9		ns
t _{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t _{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t _{HCKSW}	Maximum Skew	0.2		0.2		0.3		0.3		ns
t _{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f _{HMAX}	Maximum Frequency	350		320		280		240		MHz
Routed Array Clock Networks										
t _{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)	1.6		1.8		2.1		2.5		ns
t _{RCKL}	Input HIGH to LOW (light load) (pad to R-Cell input)	1.8		2.0		2.3		2.7		ns
t _{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)	1.8		2.1		2.5		2.8		ns
t _{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)	2.0		2.2		2.5		3.0		ns
t _{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)	1.8		2.1		2.4		2.8		ns
t _{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)	2.0		2.2		2.5		3.0		ns
t _{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t _{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t _{RCKSW}	Maximum Skew (light load)	0.5		0.5		0.5		0.7		ns
t _{RCKSW}	Maximum Skew (50% load)	0.5		0.6		0.7		0.8		ns
t _{RCKSW}	Maximum Skew (100% load)	0.5		0.6		0.7		0.8		ns
TTL Output Module Timing ³										
t _{DLH}	Data-to-Pad LOW to HIGH	1.6		1.9		2.1		2.5		ns
t _{DHL}	Data-to-Pad HIGH to LOW	1.6		1.9		2.1		2.5		ns
t _{ENZL}	Enable-to-Pad, Z to L	2.1		2.4		2.8		3.2		ns
t _{ENZH}	Enable-to-Pad, Z to H	2.3		2.7		3.1		3.6		ns
t _{ENLZ}	Enable-to-Pad, L to Z	1.4		1.7		1.9		2.2		ns
t _{ENHZ}	Enable-to-Pad, H to Z	1.3		1.5		1.7		2.0		ns

Notes:

- For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.
- Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.
- Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH} . For t_{ENZL} and t_{ENZH} , the loading is 5 pF.

A54SX16P Timing Characteristics

Table 1-19 • **A54SX16P Timing Characteristics**
(Worst-Case Commercial Conditions, $V_{CCR} = 4.75\text{ V}$, $V_{CCA}, V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	'-3' Speed		'-2' Speed		'-1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
C-Cell Propagation Delays ¹										
t _{PD}	Internal Array Module	0.6		0.7		0.8		0.9		ns
Predicted Routing Delays ²										
t _{DC}	FO = 1 Routing Delay, Direct Connect	0.1		0.1		0.1		0.1		ns
t _{FC}	FO = 1 Routing Delay, Fast Connect	0.3		0.4		0.4		0.5		ns
t _{RD1}	FO = 1 Routing Delay	0.3		0.4		0.4		0.5		ns
t _{RD2}	FO = 2 Routing Delay	0.6		0.7		0.8		0.9		ns
t _{RD3}	FO = 3 Routing Delay	0.8		0.9		1.0		1.2		ns
t _{RD4}	FO = 4 Routing Delay	1.0		1.2		1.4		1.6		ns
t _{RD8}	FO = 8 Routing Delay	1.9		2.2		2.5		2.9		ns
t _{RD12}	FO = 12 Routing Delay	2.8		3.2		3.7		4.3		ns
R-Cell Timing										
t _{RCO}	Sequential Clock-to-Q	0.9		1.1		1.3		1.4		ns
t _{CLR}	Asynchronous Clear-to-Q	0.5		0.6		0.7		0.8		ns
t _{PRESET}	Asynchronous Preset-to-Q	0.7		0.8		0.9		1.0		ns
t _{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t _{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t _{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t _{INYH}	Input Data Pad-to-Y HIGH	1.5		1.7		1.9		2.2		ns
t _{INYL}	Input Data Pad-to-Y LOW	1.5		1.7		1.9		2.2		ns
Predicted Input Routing Delays ²										
t _{IRD1}	FO = 1 Routing Delay	0.3		0.4		0.4		0.5		ns
t _{IRD2}	FO = 2 Routing Delay	0.6		0.7		0.8		0.9		ns
t _{IRD3}	FO = 3 Routing Delay	0.8		0.9		1.0		1.2		ns
t _{IRD4}	FO = 4 Routing Delay	1.0		1.2		1.4		1.6		ns
t _{IRD8}	FO = 8 Routing Delay	1.9		2.2		2.5		2.9		ns
t _{IRD12}	FO = 12 Routing Delay	2.8		3.2		3.7		4.3		ns

Note:

1. For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.
3. Delays based on 10 pF loading.

Table 1-20 • A54SX32 Timing Characteristics (Continued)
(Worst-Case Commercial Conditions, $V_{CCR} = 4.75\text{ V}$, $V_{CCA}, V_{CCI} = 3.0\text{ V}$, $T_J = 70^\circ\text{C}$)

Parameter	Description	'-3' Speed		'-2' Speed		'-1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Dedicated (Hardwired) Array Clock Network										
t _{HCKH}	Input LOW to HIGH (pad to R-Cell input)	1.9		2.1		2.4		2.8		ns
t _{HCKL}	Input HIGH to LOW (pad to R-Cell input)	1.9		2.1		2.4		2.8		ns
t _{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t _{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t _{HCKSW}	Maximum Skew	0.3		0.4		0.4		0.5		ns
t _{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f _{HMAX}	Maximum Frequency	350		320		280		240		MHz
Routed Array Clock Networks										
t _{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)	2.4		2.7		3.0		3.5		ns
t _{RCKL}	Input HIGH to LOW (light load) (pad to R-Cell input)	2.4		2.7		3.1		3.6		ns
t _{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)	2.7		3.0		3.5		4.1		ns
t _{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)	2.7		3.1		3.6		4.2		ns
t _{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)	2.7		3.1		3.5		4.1		ns
t _{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)	2.8		3.2		3.6		4.3		ns
t _{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t _{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t _{RCKSW}	Maximum Skew (light load)	0.85		0.98		1.1		1.3		ns
t _{RCKSW}	Maximum Skew (50% load)	1.23		1.4		1.6		1.9		ns
t _{RCKSW}	Maximum Skew (100% load)	1.30		1.5		1.7		2.0		ns
TTL Output Module Timing ³										
t _{DLH}	Data-to-Pad LOW to HIGH	1.6		1.9		2.1		2.5		ns
t _{DHL}	Data-to-Pad HIGH to LOW	1.6		1.9		2.1		2.5		ns
t _{ENZL}	Enable-to-Pad, Z to L	2.1		2.4		2.8		3.2		ns
t _{ENZH}	Enable-to-Pad, Z to H	2.3		2.7		3.1		3.6		ns
t _{ENLZ}	Enable-to-Pad, L to Z	1.4		1.7		1.9		2.2		ns
t _{ENHZ}	Enable-to-Pad, H to Z	1.3		1.5		1.7		2.0		ns

Note:

- For dual-module macros, use $t_{PD} + t_{RD1} + t_{PDn}$, $t_{RCO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.
- Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.
- Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH} . For t_{ENZL} and t_{ENZH} the loading is 5 pF.

Pin Description

CLKA/B Clock A and B

These pins are 3.3 V / 5.0 V PCI/TTL clock inputs for clock distribution networks. The clock input is buffered prior to clocking the R-cells. If not used, this pin must be set LOW or HIGH on the board. It must not be left floating. (For A54SX72A, these clocks can be configured as bidirectional.)

GND Ground

LOW supply voltage.

HCLK Dedicated (hardwired) Array Clock

This pin is the 3.3 V / 5.0 V PCI/TTL clock input for sequential modules. This input is directly wired to each R-cell and offers clock speeds independent of the number of R-cells being driven. If not used, this pin must be set LOW or HIGH on the board. It must not be left floating.

I/O Input/Output

The I/O pin functions as an input, output, tristate, or bidirectional buffer. Based on certain configurations, input and output levels are compatible with standard TTL, LVTTTL, 3.3 V PCI or 5.0 V PCI specifications. Unused I/O pins are automatically tristated by the Designer Series software.

NC No Connection

This pin is not connected to circuitry within the device.

PRA, I/O Probe A

The Probe A pin is used to output data from any user-defined design node within the device. This independent diagnostic pin can be used in conjunction with the Probe B pin to allow real-time diagnostic output of any signal path within the device. The Probe A pin can be used as a user-defined I/O when verification has been completed. The pin's probe capabilities can be permanently disabled to protect programmed design confidentiality.

PRB, I/O Probe B

The Probe B pin is used to output data from any node within the device. This diagnostic pin can be used in conjunction with the Probe A pin to allow real-time diagnostic output of any signal path within the device. The Probe B pin can be used as a user-defined I/O when verification has been completed. The pin's probe capabilities can be permanently disabled to protect programmed design confidentiality.

TCK Test Clock

Test clock input for diagnostic probe and device programming. In flexible mode, TCK becomes active when the TMS pin is set LOW (refer to Table 1-2 on page 1-6). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDI Test Data Input

Serial input for boundary scan testing and diagnostic probe. In flexible mode, TDI is active when the TMS pin is set LOW (refer to Table 1-2 on page 1-6). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDO Test Data Output

Serial output for boundary scan testing. In flexible mode, TDO is active when the TMS pin is set LOW (refer to Table 1-2 on page 1-6). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TMS Test Mode Select

The TMS pin controls the use of the IEEE 1149.1 Boundary Scan pins (TCK, TDI, TDO). In flexible mode when the TMS pin is set LOW, the TCK, TDI, and TDO pins are boundary scan pins (refer to Table 1-2 on page 1-6). Once the boundary scan pins are in test mode, they will remain in that mode until the internal boundary scan state machine reaches the "logic reset" state. At this point, the boundary scan pins will be released and will function as regular I/O pins. The "logic reset" state is reached 5 TCK cycles after the TMS pin is set HIGH. In dedicated test mode, TMS functions as specified in the IEEE 1149.1 specifications.

V_{CCI} Supply Voltage

Supply voltage for I/Os. See Table 1-1 on page 1-5.

V_{CCA} Supply Voltage

Supply voltage for Array. See Table 1-1 on page 1-5.

V_{CCR} Supply Voltage

Supply voltage for input tolerance (required for internal biasing). See Table 1-1 on page 1-5.

176-Pin TQFP

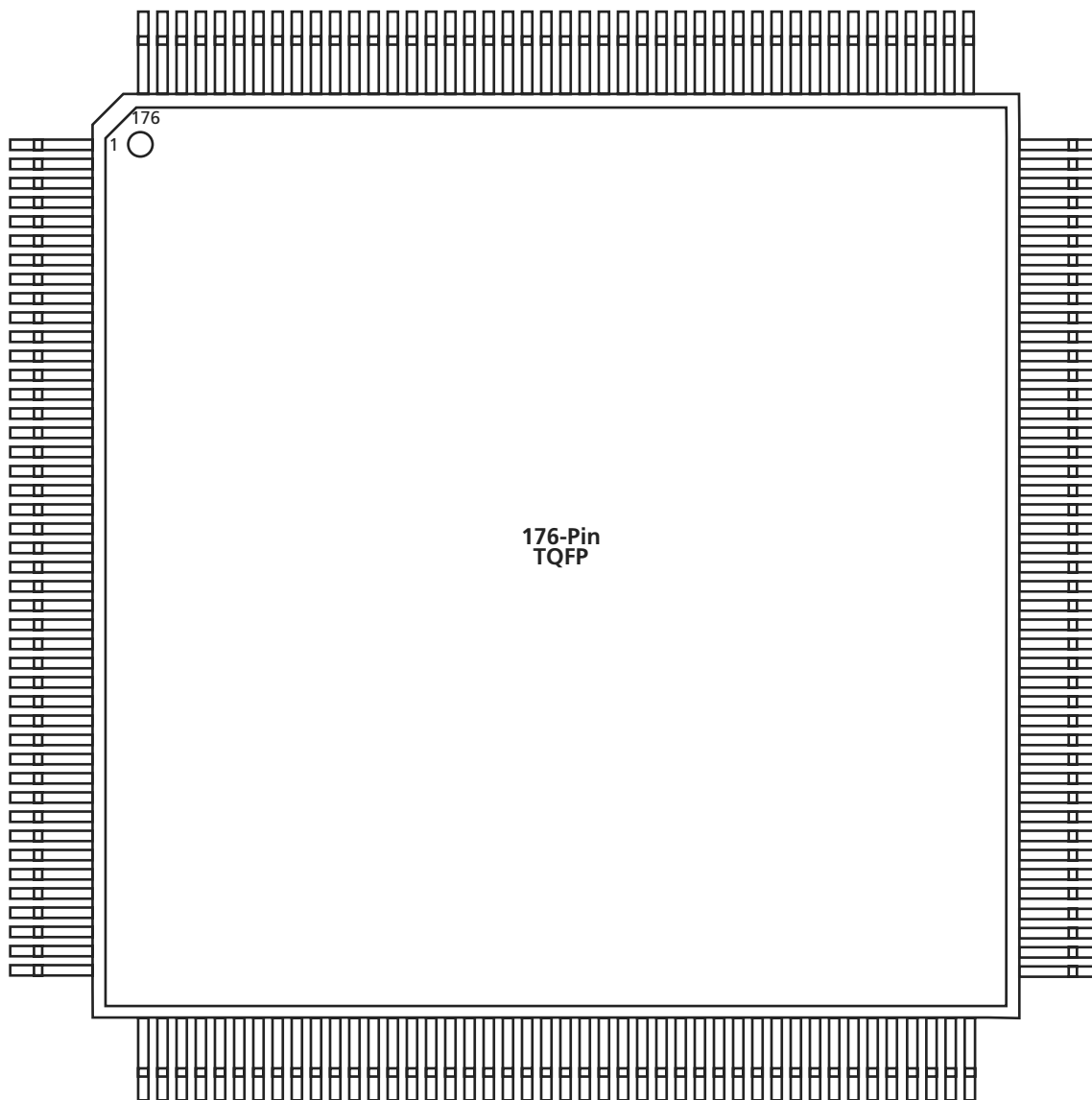


Figure 2-4 • 176-Pin TQFP (Top View)

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at <http://www.actel.com/products/rescenter/package/index.html>.

176-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
69	HCLK	HCLK	HCLK
70	I/O	I/O	I/O
71	I/O	I/O	I/O
72	I/O	I/O	I/O
73	I/O	I/O	I/O
74	I/O	I/O	I/O
75	I/O	I/O	I/O
76	I/O	I/O	I/O
77	I/O	I/O	I/O
78	I/O	I/O	I/O
79	NC	I/O	I/O
80	I/O	I/O	I/O
81	NC	I/O	I/O
82	V _{CCI}	V _{CCI}	V _{CCI}
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	I/O	I/O	I/O
86	I/O	I/O	I/O
87	TDO, I/O	TDO, I/O	TDO, I/O
88	I/O	I/O	I/O
89	GND	GND	GND
90	NC	I/O	I/O
91	NC	I/O	I/O
92	I/O	I/O	I/O
93	I/O	I/O	I/O
94	I/O	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	I/O	I/O	I/O
98	V _{CCA}	V _{CCA}	V _{CCA}
99	V _{CCI}	V _{CCI}	V _{CCI}
100	I/O	I/O	I/O
101	I/O	I/O	I/O
102	I/O	I/O	I/O

176-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
103	I/O	I/O	I/O
104	I/O	I/O	I/O
105	I/O	I/O	I/O
106	I/O	I/O	I/O
107	I/O	I/O	I/O
108	GND	GND	GND
109	V _{CCA}	V _{CCA}	V _{CCA}
110	GND	GND	GND
111	I/O	I/O	I/O
112	I/O	I/O	I/O
113	I/O	I/O	I/O
114	I/O	I/O	I/O
115	I/O	I/O	I/O
116	I/O	I/O	I/O
117	I/O	I/O	I/O
118	NC	I/O	I/O
119	I/O	I/O	I/O
120	NC	I/O	I/O
121	NC	I/O	I/O
122	V _{CCA}	V _{CCA}	V _{CCA}
123	GND	GND	GND
124	V _{CCI}	V _{CCI}	V _{CCI}
125	I/O	I/O	I/O
126	I/O	I/O	I/O
127	I/O	I/O	I/O
128	I/O	I/O	I/O
129	I/O	I/O	I/O
130	I/O	I/O	I/O
131	NC	I/O	I/O
132	NC	I/O	I/O
133	GND	GND	GND
134	I/O	I/O	I/O
135	I/O	I/O	I/O
136	I/O	I/O	I/O

100-Pin VQFP		
Pin Number	A545X08 Function	A545X16, A545X16P Function
1	GND	GND
2	TDI, I/O	TDI, I/O
3	I/O	I/O
4	I/O	I/O
5	I/O	I/O
6	I/O	I/O
7	TMS	TMS
8	V _{CCI}	V _{CCI}
9	GND	GND
10	I/O	I/O
11	I/O	I/O
12	I/O	I/O
13	I/O	I/O
14	I/O	I/O
15	I/O	I/O
16	I/O	I/O
17	I/O	I/O
18	I/O	I/O
19	I/O	I/O
20	V _{CCI}	V _{CCI}
21	I/O	I/O
22	I/O	I/O
23	I/O	I/O
24	I/O	I/O
25	I/O	I/O
26	I/O	I/O
27	I/O	I/O
28	I/O	I/O
29	I/O	I/O
30	I/O	I/O
31	I/O	I/O
32	I/O	I/O
33	I/O	I/O
34	PRB, I/O	PRB, I/O

100-Pin VQFP		
Pin Number	A545X08 Function	A545X16, A545X16P Function
35	V _{CCA}	V _{CCA}
36	GND	GND
37	V _{CCR}	V _{CCR}
38	I/O	I/O
39	HCLK	HCLK
40	I/O	I/O
41	I/O	I/O
42	I/O	I/O
43	I/O	I/O
44	V _{CCI}	V _{CCI}
45	I/O	I/O
46	I/O	I/O
47	I/O	I/O
48	I/O	I/O
49	TDO, I/O	TDO, I/O
50	I/O	I/O
51	GND	GND
52	I/O	I/O
53	I/O	I/O
54	I/O	I/O
55	I/O	I/O
56	I/O	I/O
57	V _{CCA}	V _{CCA}
58	V _{CCI}	V _{CCI}
59	I/O	I/O
60	I/O	I/O
61	I/O	I/O
62	I/O	I/O
63	I/O	I/O
64	I/O	I/O
65	I/O	I/O
66	I/O	I/O
67	V _{CCA}	V _{CCA}
68	GND	GND

100-Pin VQFP		
Pin Number	A545X08 Function	A545X16, A545X16P Function
69	GND	GND
70	I/O	I/O
71	I/O	I/O
72	I/O	I/O
73	I/O	I/O
74	I/O	I/O
75	I/O	I/O
76	I/O	I/O
77	I/O	I/O
78	I/O	I/O
79	I/O	I/O
80	I/O	I/O
81	I/O	I/O
82	V _{CCI}	V _{CCI}
83	I/O	I/O
84	I/O	I/O
85	I/O	I/O
86	I/O	I/O
87	CLKA	CLKA
88	CLKB	CLKB
89	V _{CCR}	V _{CCR}
90	V _{CCA}	V _{CCA}
91	GND	GND
92	PRA, I/O	PRA, I/O
93	I/O	I/O
94	I/O	I/O
95	I/O	I/O
96	I/O	I/O
97	I/O	I/O
98	I/O	I/O
99	I/O	I/O
100	TCK, I/O	TCK, I/O

329-Pin PBGA

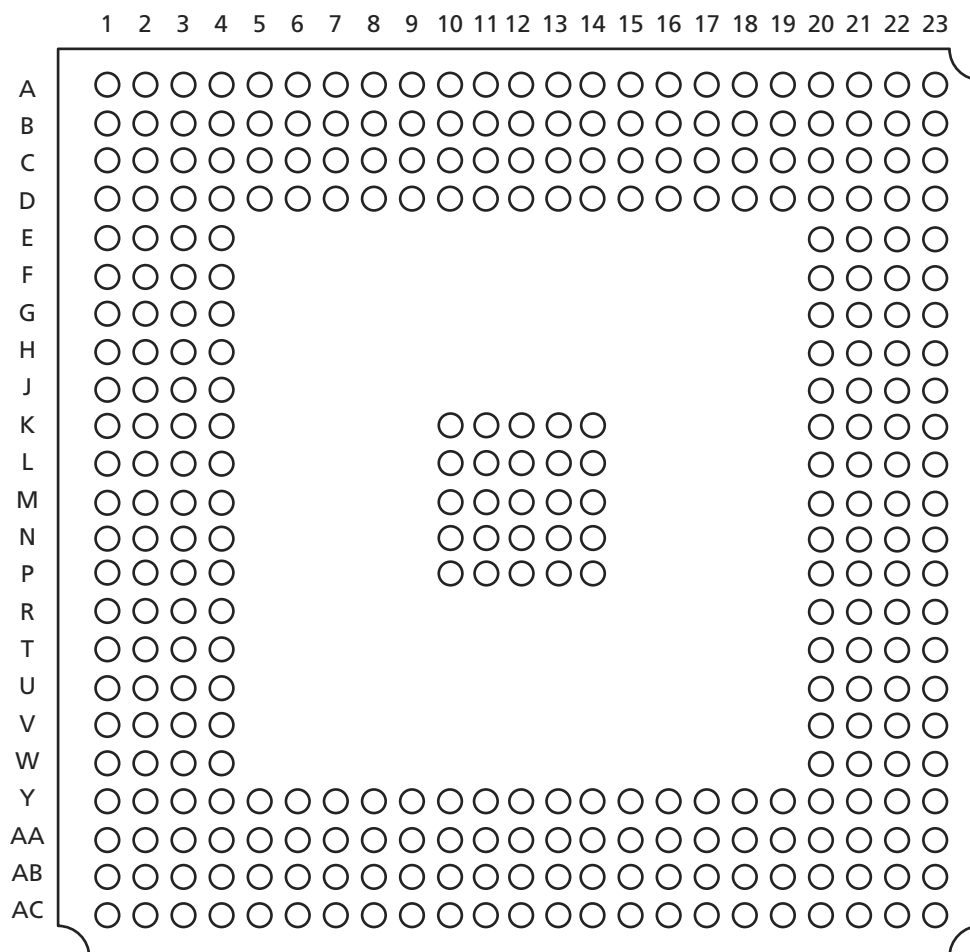


Figure 2-7 • 329-Pin PBGA (Top View)

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at <http://www.actel.com/products/rescenter/package/index.html>.