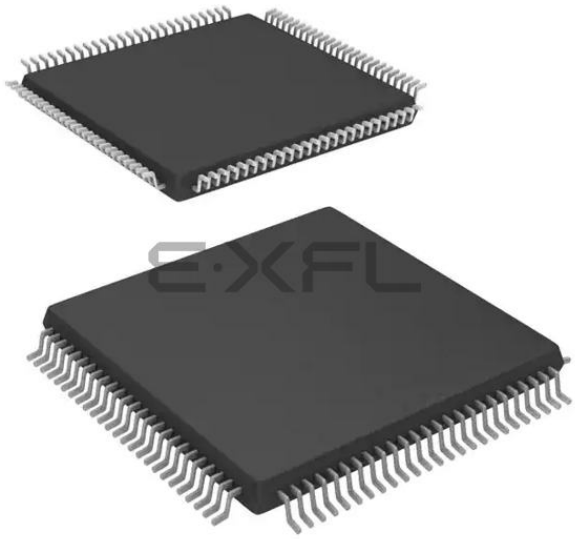

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1452
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	81
Number of Gates	24000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	100-TQFP
Supplier Device Package	100-VQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx16-1vqg100

Routing Resources

Clusters and SuperClusters can be connected through the use of two innovative local routing resources called *FastConnect* and *DirectConnect,* which enable extremely fast and predictable interconnection of modules within clusters and SuperClusters (Figure 1-5 and Figure 1-6). This routing architecture also dramatically reduces the number of antifuses required to complete a circuit, ensuring the highest possible performance.

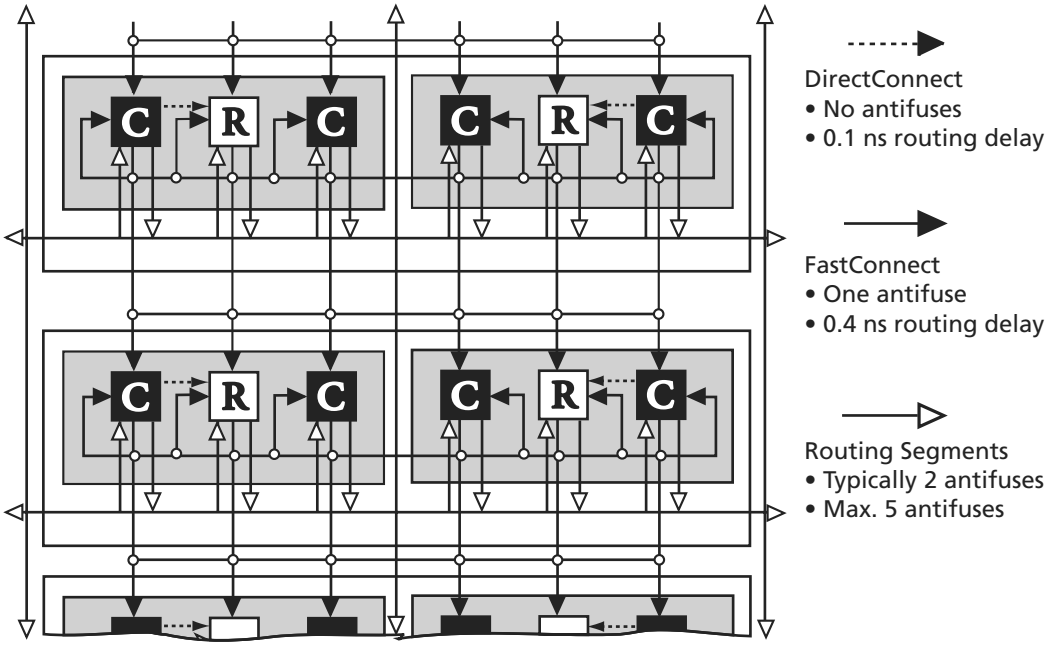


Figure 1-5 • **DirectConnect and FastConnect for Type 1 SuperClusters**

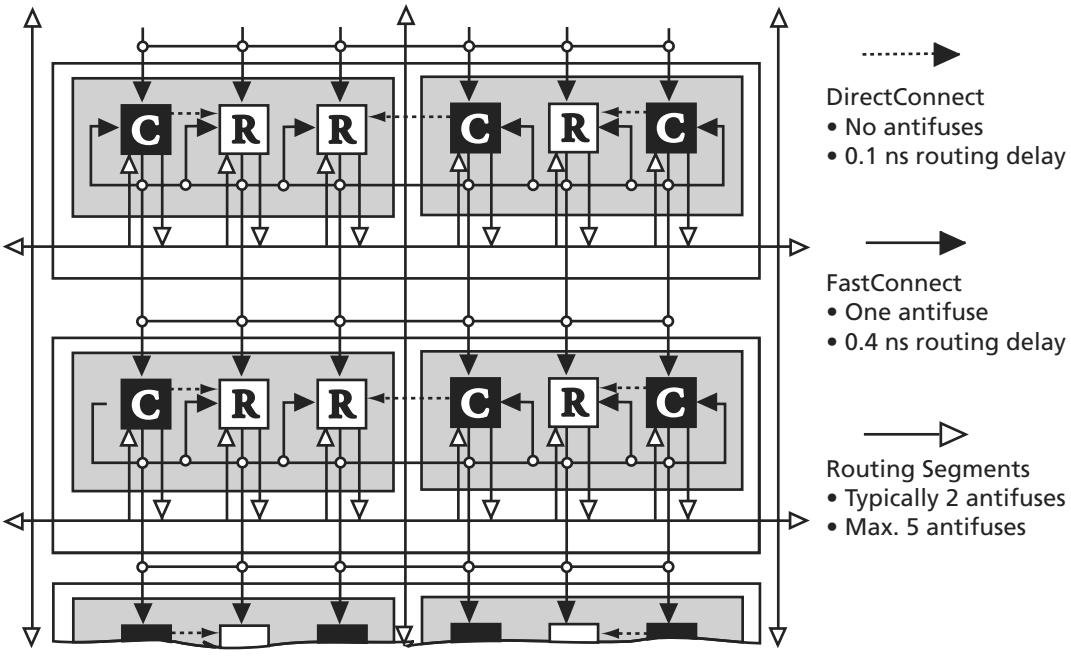


Figure 1-6 • **DirectConnect and FastConnect for Type 2 SuperClusters**

Boundary Scan Testing (BST)

All SX devices are IEEE 1149.1 compliant. SX devices offer superior diagnostic and testing capabilities by providing Boundary Scan Testing (BST) and probing capabilities. These functions are controlled through the special test pins in conjunction with the program fuse. The functionality of each pin is described in Table 1-2. In the dedicated test mode, TCK, TDI, and TDO are dedicated pins and cannot be used as regular I/Os. In flexible mode, TMS should be set HIGH through a pull-up resistor of 10 kΩ. TMS can be pulled LOW to initiate the test sequence.

The program fuse determines whether the device is in dedicated or flexible mode. The default (fuse not blown) is flexible mode.

Table 1-2 • **Boundary Scan Pin Functionality**

Program Fuse Blown (Dedicated Test Mode)	**Program Fuse Not Blown (Flexible Mode)**
TCK, TDI, TDO are dedicated BST pins.	TCK, TDI, TDO are flexible and may be used as I/Os.
No need for pull-up resistor for TMS	Use a pull-up resistor of 10 kΩ on TMS.

Dedicated Test Mode

In Dedicated mode, all JTAG pins are reserved for BST; designers cannot use them as regular I/Os. An internal pull-up resistor is automatically enabled on both TMS and TDI pins, and the TMS pin will function as defined in the IEEE 1149.1 (JTAG) specification.

To select Dedicated mode, users need to reserve the JTAG pins in Actel's Designer software by checking the "Reserve JTAG" box in "Device Selection Wizard" (Figure 1-7). JTAG pins comply with LVTTL/TTL I/O specification regardless of whether they are used as a user I/O or a JTAG I/O. Refer to the Table 1-5 on page 1-8 for detailed specifications.

Figure 1-7 • **Device Selection Wizard**

Development Tool Support

The SX family of FPGAs is fully supported by both the Actel Libero® Integrated Design Environment (IDE) and Designer FPGA Development software. Actel Libero IDE is a design management environment, seamlessly integrating design tools while guiding the user through the design flow, managing all design and log files, and passing necessary design data among tools. Libero IDE allows users to integrate both schematic and HDL synthesis into a single flow and verify the entire design in a single environment. Libero IDE includes Synplify® for Actel from Synplicity®, ViewDraw® for Actel from Mentor Graphics®, Model*Sim*® HDL Simulator from Mentor Graphics, WaveFormer Lite™ from SynaptiCAD™, and Designer software from Actel. Refer to the Libero IDE flow diagram (located on the Actel website) for more information.

Actel Designer software is a place-and-route tool and provides a comprehensive suite of backend support tools for FPGA development. The Designer software includes timing-driven place-and-route, and a world-class integrated static timing analyzer and constraints editor. With the Designer software, a user can select and lock package pins while only minimally impacting the results of place-and-route. Additionally, the back-annotation flow is compatible with all the major simulators, and the simulation results can be cross-probed with Silicon Explorer II, Actel integrated verification and logic analysis tool. Another tool included in the Designer software is the SmartGen core generator, which easily creates popular and commonly used logic functions for implementation into your schematic or HDL design. Actel Designer software is compatible with the most popular FPGA design entry and verification tools from companies such as Mentor Graphics, Synplicity, Synopsys®, and Cadence® Design Systems. The Designer software is available for both the Windows® and UNIX® operating systems.

Probe Circuit Control Pins

The Silicon Explorer II tool uses the boundary scan ports (TDI, TCK, TMS, and TDO) to select the desired nets for verification. The selected internal nets are assigned to the PRA/PRB pins for observation. Figure 1-8 on page 1-7 illustrates the interconnection between Silicon Explorer II and the FPGA to perform in-circuit verification.

Design Considerations

The TDI, TCK, TDO, PRA, and PRB pins should not be used as input or bidirectional ports. Because these pins are active during probing, critical signals input through these pins are not available while probing. In addition, the Security Fuse should not be programmed because doing so disables the Probe Circuitry.

Table 1-4 • **Recommended Operating Conditions**

Parameter	Commercial	Industrial	Military	Units
Temperature Range*	0 to + 70	–40 to + 85	–55 to +125	°C
3.3 V Power Supply Tolerance	±10	±10	±10	%V_{CC}
5.0 V Power Supply Tolerance	±5	±10	±10	%V_{CC}

Note: **Ambient temperature (T_A) is used for commercial and industrial; case temperature (T_C) is used for military.*

Table 1-5 • **Electrical Specifications**

		Commercial		Industrial		
Symbol	**Parameter**	**Min.**	**Max.**	**Min.**	**Max.**	**Units**
V_{OH}	(I_{OH} = –20 µA) (CMOS) (I_{OH} = –8 mA) (TTL) (I_{OH} = –6 mA) (TTL)	(V_{CCI} – 0.1) 2.4	V_{CCI} V_{CCI}	(V_{CCI} – 0.1) 2.4	V_{CCI} V_{CCI}	V
V_{OL}	(I_{OL}= 20 µA) (CMOS) (I_{OL} = 12 mA) (TTL) (I_{OL} = 8 mA) (TTL)		0.10 0.50		 0.50	V
V_{IL}			0.8		0.8	V
V_{IH}		2.0		2.0		V
t_R, t_F	Input Transition Time t_R, t_F		50		50	ns
C_{IO}	C_{IO} I/O Capacitance		10		10	pF
I_{CC}	Standby Current, I_{CC}		4.0		4.0	mA
$I_{CC(D)}$	$I_{CC(D)}$ $I_{Dynamic}$ V_{CC} Supply Current	See "Evaluating Power in SX Devices" on page 1-16.				

A54SX16P DC Specifications (3.3 V PCI Operation)

Table 1-8 • **A54SX16P DC Specifications (3.3 V PCI Operation)**

Symbol	Parameter	Condition	Min.	Max.	Units
V_{CCA}	Supply Voltage for Array		3.0	3.6	V
V_{CCR}	Supply Voltage required for Internal Biasing		3.0	3.6	V
V_{CCI}	Supply Voltage for I/Os		3.0	3.6	V
V_{IH}	Input High Voltage		$0.5V_{CC}$	V_{CC} + 0.5	V
V_{IL}	Input Low Voltage		–0.5	$0.3V_{CC}$	V
I_{IPU}	Input Pull-up Voltage[1]		$0.7V_{CC}$		V
I_{IL}	Input Leakage Current[2]	$0 < V_{IN} < V_{CC}$		±10	µA
V_{OH}	Output High Voltage	I_{OUT} = –500 µA	$0.9V_{CC}$		V
V_{OL}	Output Low Voltage	I_{OUT} = 1500 µA		$0.1V_{CC}$	V
C_{IN}	Input Pin Capacitance[3]			10	pF
C_{CLK}	CLK Pin Capacitance		5	12	pF
C_{IDSEL}	IDSEL Pin Capacitance[4]			8	pF

Notes:

1. *This specification should be guaranteed by design. It is the minimum voltage to which pull-up resistors are calculated to pull a floated network. Applications sensitive to static power utilization should assure that the input buffer is conducting minimum current at this input voltage.*
2. *Input leakage currents include hi-Z output leakage for all bidirectional buffers with tristate outputs.*
3. *Absolute maximum pin capacitance for a PCI input is 10 pF (except for CLK).*
4. *Lower capacitance on this input-only pin allows for non-resistive coupling to AD[xx].*

Evaluating Power in SX Devices

A critical element of system reliability is the ability of electronic devices to safely dissipate the heat generated during operation. The thermal characteristics of a circuit depend on the device and package used, the operating temperature, the operating current, and the system's ability to dissipate heat.

You should complete a power evaluation early in the design process to help identify potential heat-related problems in the system and to prevent the system from exceeding the device's maximum allowed junction temperature.

The actual power dissipated by most applications is significantly lower than the power the package can dissipate. However, a thermal analysis should be performed for all projects. To perform a power evaluation, follow these steps:

1. Estimate the power consumption of the application.
2. Calculate the maximum power allowed for the device and package.
3. Compare the estimated power and maximum power values.

Estimating Power Consumption

The total power dissipation for the SX family is the sum of the DC power dissipation and the AC power dissipation. Use EQ 1-5 to calculate the estimated power consumption of your application.

$$P_{Total} = P_{DC} + P_{AC}$$

EQ 1-5

DC Power Dissipation

The power due to standby current is typically a small component of the overall power. The Standby power is shown in Table 1-12 for commercial, worst-case conditions (70°C).

Table 1-12 • **Standby Power**

I_{CC}	V_{CC}	Power
4 mA	3.6 V	14.4 mW

The DC power dissipation is defined in EQ 1-6.

$$P_{DC} = (I_{standby}) \times V_{CCA} + (I_{standby}) \times V_{CCR} + (I_{standby}) \times V_{CCI} + xV_{OL} \times I_{OL} + y(V_{CCI} - V_{OH}) \times V_{OH}$$

EQ 1-6

AC Power Dissipation

The power dissipation of the SX Family is usually dominated by the dynamic power dissipation. Dynamic power dissipation is a function of frequency, equivalent capacitance, and power supply voltage. The AC power dissipation is defined in EQ 1-7 and EQ 1-8.

$$P_{AC} = P_{Module} + P_{RCLKA\ Net} + P_{RCLKB\ Net} + P_{HCLK\ Net} + P_{Output\ Buffer} + P_{Input\ Buffer}$$

EQ 1-7

$$P_{AC} = V_{CCA}^2 \times [(m \times C_{EQM} \times f_m)_{Module} + (n \times C_{EQI} \times f_n)_{Input\ Buffer} + (p \times (C_{EQO} + C_L) \times f_p)_{Output\ Buffer} + (0.5 \times (q_1 \times C_{EQCR} \times f_{q1}) + (r_1 \times f_{q1}))_{RCLKA} + (0.5 \times (q2 \times CEQCR \times f_{q2}) + (r2 \times f_{q2}))RCLKB + (0.5 \times (s_1 \times C_{EQHV} \times f_{s1}) + (C_{EQHF} \times f_{s1}))_{HCLK}]$$

EQ 1-8

Definition of Terms Used in Formula

m = Number of logic modules switching at f_m
n = Number of input buffers switching at f_n
p = Number of output buffers switching at f_p
q_1 = Number of clock loads on the first routed array clock
q_2 = Number of clock loads on the second routed array clock
x = Number of I/Os at logic low
y = Number of I/Os at logic high
r_1 = Fixed capacitance due to first routed array clock
r_2 = Fixed capacitance due to second routed array clock
s_1 = Number of clock loads on the dedicated array clock
C_{EQM} = Equivalent capacitance of logic modules in pF
C_{EQI} = Equivalent capacitance of input buffers in pF
C_{EQO} = Equivalent capacitance of output buffers in pF
C_{EQCR} = Equivalent capacitance of routed array clock in pF
C_{EQHV} = Variable capacitance of dedicated array clock
C_{EQHF} = Fixed capacitance of dedicated array clock
C_L = Output lead capacitance in pF
f_m = Average logic module switching rate in MHz
f_n = Average input buffer switching rate in MHz
f_p = Average output buffer switching rate in MHz
f_{q1} = Average first routed array clock rate in MHz
f_{q2} = Average second routed array clock rate in MHz
f_{s1} = Average dedicated array clock rate in MHz

Table 1-13 shows capacitance values for various devices.

Table 1-13 • **Capacitance Values for Devices**

	A54SX08	A54SX16	A54SX16P	A54SX32
C_{EQM} (pF)	4.0	4.0	4.0	4.0
C_{EQI} (pF)	3.4	3.4	3.4	3.4
C_{EQO} (pF)	4.7	4.7	4.7	4.7
C_{EQCR} (pF)	1.6	1.6	1.6	1.6
C_{EQHV}	0.615	0.615	0.615	0.615
C_{EQHF}	60	96	96	140
r_1 (pF)	87	138	138	171
r_2 (pF)	87	138	138	171

Guidelines for Calculating Power Consumption

The power consumption guidelines are meant to represent worst-case scenarios so that they can be generally used to predict the upper limits of power dissipation. These guidelines are shown in Table 1-14.

Sample Power Calculation

One of the designs used to characterize the SX family was a 528 bit serial-in, serial-out shift register. The design utilized 100 percent of the dedicated flip-flops of an A54SX16P device. A pattern of 0101… was clocked into the device at frequencies ranging from 1 MHz to 200 MHz. Shifting in a series of 0101… caused 50 percent of the flip-flops to toggle from low to high at every clock cycle.

Table 1-14 • **Power Consumption Guidelines**

Description	Power Consumption Guideline
Logic Modules (m)	20% of modules
Inputs Switching (n)	# inputs/4
Outputs Switching (p)	# outputs/4
First Routed Array Clock Loads (q_1)	20% of register cells
Second Routed Array Clock Loads (q_2)	20% of register cells
Load Capacitance (C_L)	35 pF
Average Logic Module Switching Rate (f_m)	f/10
Average Input Switching Rate (f_n)	f/5
Average Output Switching Rate (f_p)	f/10
Average First Routed Array Clock Rate (f_{q1})	f/2
Average Second Routed Array Clock Rate (f_{q2})	f/2
Average Dedicated Array Clock Rate (f_{s1})	f
Dedicated Clock Array Clock Loads (s_1)	20% of regular modules

Follow the steps below to estimate power consumption. The values provided for the sample calculation below are for the shift register design above. This method for estimating power consumption is conservative and the actual power consumption of your design may be less than the estimated power consumption.

The total power dissipation for the SX family is the sum of the AC power dissipation and the DC power dissipation.

$$P_{Total} = P_{AC} \text{ (dynamic power)} + P_{DC} \text{ (static power)}$$

EQ 1-9

AC Power Dissipation

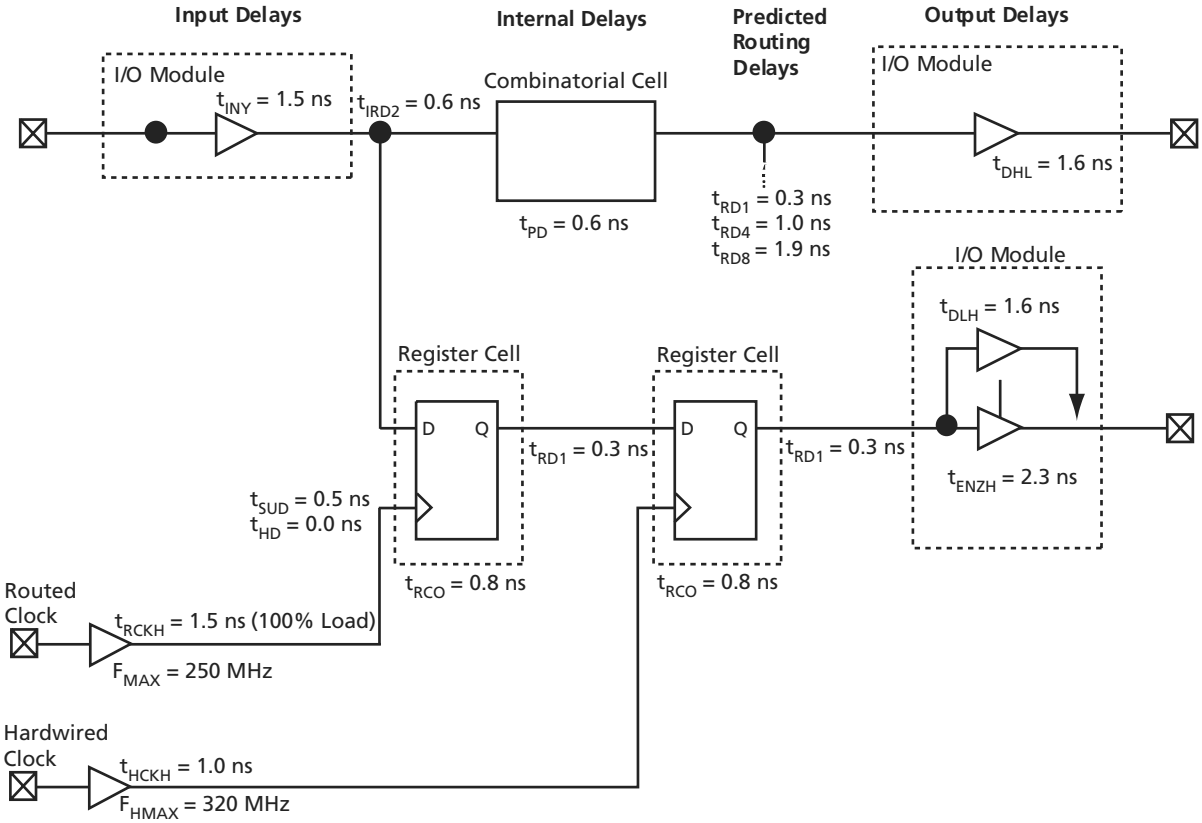
$$P_{AC} = P_{Module} + P_{RCLKA\ Net} + P_{RCLKB\ Net} + P_{HCLK\ Net} + P_{Output\ Buffer} + P_{Input\ Buffer}$$

EQ 1-10

$$P_{AC} = V_{CCA}^2 \times [(m \times C_{EQM} \times f_m)_{Module} + (n \times C_{EQI} \times f_n)_{Input\ Buffer} + (p \times (C_{EQO} + C_L) \times f_p)_{Output\ Buffer} + (0.5\ (q_1 \times C_{EQCR} \times f_{q1}) + (r_1 \times f_{q1}))_{RCLKA} + (0.5\ (q_2 \times C_{EQCR} \times f_{q2}) + (r_2 \times f_{q2}))_{RCLKB} + (0.5\ (s_1 \times C_{EQHV} \times f_{s1}) + (C_{EQHF} \times f_{s1}))_{HCLK}]$$

EQ 1-11

SX Timing Model



Note: *Values shown for A54SX08-3, worst-case commercial conditions.*

Figure 1-12 • **SX Timing Model**

Hardwired Clock

External Setup = $t_{INY} + t_{IRD1} + t_{SUD} - t_{HCKH}$
= 1.5 + 0.3 + 0.5 – 1.0 = 1.3 ns

EQ 1-15

Clock-to-Out (Pin-to-Pin)

= $t_{HCKH} + t_{RCO} + t_{RD1} + t_{DHL}$
= 1.0 + 0.8 + 0.3 + 1.6 = 3.7 ns

EQ 1-16

Routed Clock

External Setup = $t_{INY} + t_{IRD1} + t_{SUD} - t_{RCKH}$
= 1.5 + 0.3 + 0.5 – 1.5 = 0.8 ns

EQ 1-17

Clock-to-Out (Pin-to-Pin)

= $t_{RCKH} + t_{RCO} + t_{RD1} + t_{DHL}$
= 1.52+ 0.8 + 0.3 + 1.6 = 4.2 ns

EQ 1-18

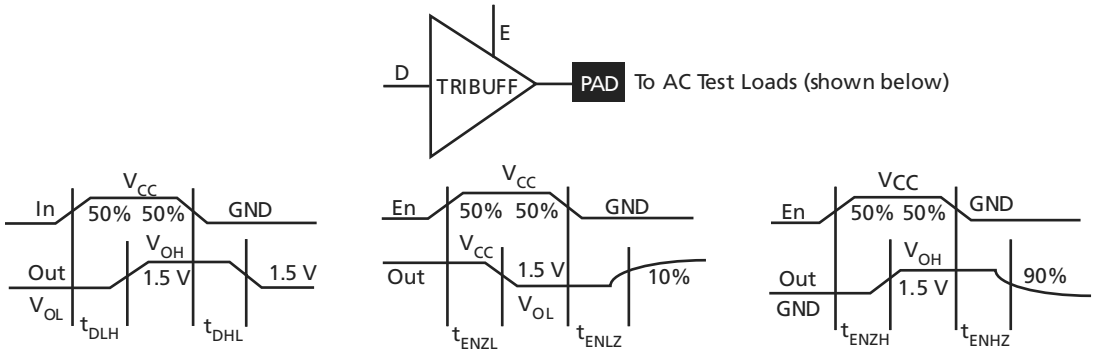


Figure 1-13 • **Output Buffer Delays**

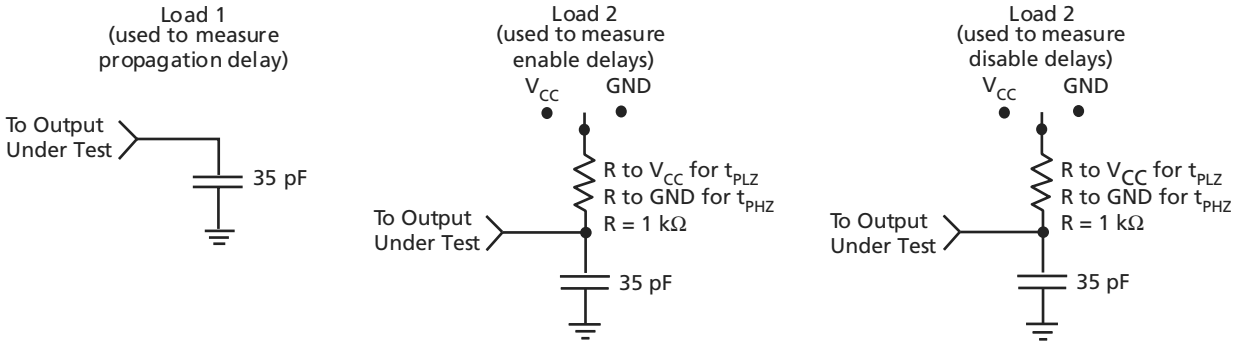


Figure 1-14 • **AC Test Loads**

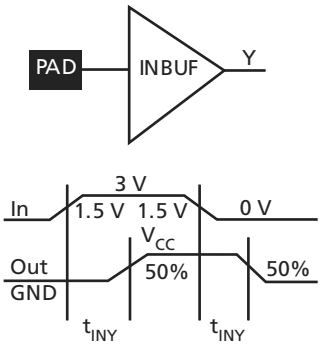


Figure 1-15 • **Input Buffer Delays**

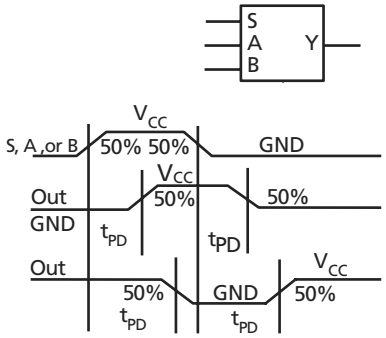


Figure 1-16 • **C-Cell Delays**

Register Cell Timing Characteristics

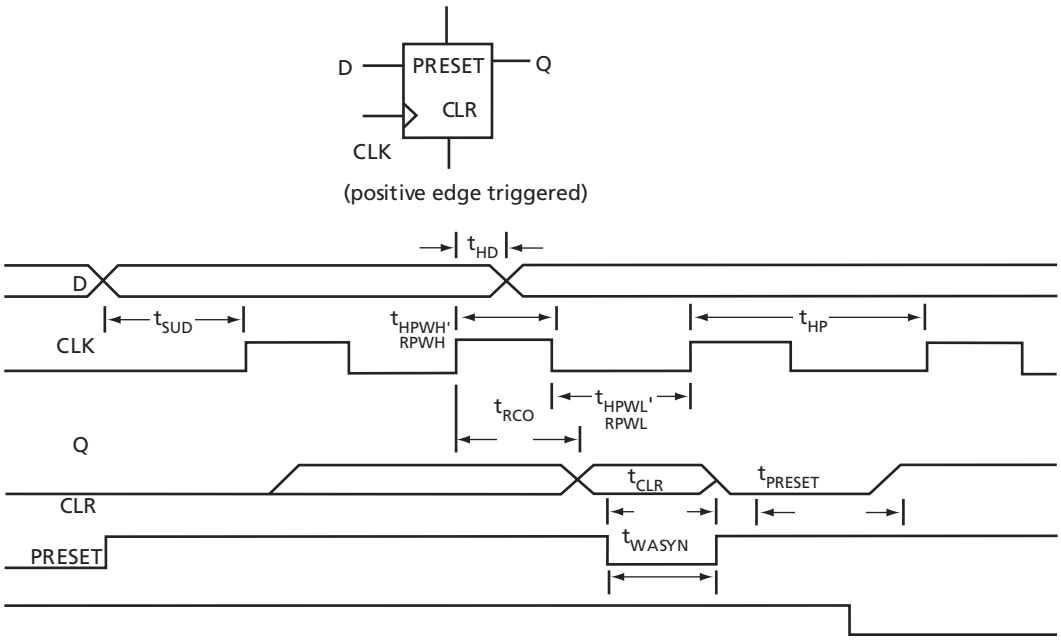


Figure 1-17 • **Flip-Flops**

Timing Characteristics

Timing characteristics for SX devices fall into three categories: family-dependent, device-dependent, and design-dependent. The input and output buffer characteristics are common to all SX family members. Internal routing delays are device-dependent. Design dependency means actual delays are not determined until after placement and routing of the user's design is complete. Delay values may then be determined by using the DirectTime Analyzer utility or performing simulation with post-layout delays.

Critical Nets and Typical Nets

Propagation delays are expressed only for typical nets, which are used for initial design performance evaluation. Critical net delays can then be applied to the most time-critical paths. Critical nets are determined by net property assignment prior to placement and routing. Up to 6% of the nets in a design may be designated as critical, while 90% of the nets in a design are typical.

Long Tracks

Some nets in the design use long tracks. Long tracks are special routing resources that span multiple rows, columns, or modules. Long tracks employ three and sometimes five antifuse connections. This increases capacitance and resistance, resulting in longer net delays for macros connected to long tracks. Typically up to 6 percent of nets in a fully utilized device require long tracks. Long tracks contribute approximately 4 ns to 8.4 ns delay. This additional delay is represented statistically in higher fanout (FO = 24) routing delays in the datasheet specifications section.

Timing Derating

SX devices are manufactured in a CMOS process. Therefore, device performance varies according to temperature, voltage, and process variations. Minimum timing parameters reflect maximum operating voltage, minimum operating temperature, and best-case processing. Maximum timing parameters reflect minimum operating voltage, maximum operating temperature, and worst-case processing.

A54SX08 Timing Characteristics

Table 1-17 • **A54SX08 Timing Characteristics (Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)**

Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
C-Cell Propagation Delays[1]										
t_{PD}	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted Routing Delays[2]										
t_{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t_{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t_{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{RD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{RD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timing										
t_{RCO}	Sequential Clock-to-Q		0.8		1.1		1.2		1.4	ns
t_{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t_{PRESET}	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t_{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t_{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t_{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Module Propagation Delays										
t_{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t_{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Input Module Predicted Routing Delays[2]										
t_{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t_{IRD2}	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t_{IRD3}	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t_{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t_{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t_{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Note:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*

Table 1-18 • **A54SX16 Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Dedicated (Hardwired) Array Clock Network										
t_{HCKH}	Input LOW to HIGH (pad to R-Cell input)		1.2		1.4		1.5		1.8	ns
t_{HCKL}	Input HIGH to LOW (pad to R-Cell input)		1.2		1.4		1.6		1.9	ns
t_{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t_{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t_{HCKSW}	Maximum Skew		0.2		0.2		0.3		0.3	ns
t_{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f_{HMAX}	Maximum Frequency		350		320		280		240	MHz
Routed Array Clock Networks										
t_{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)		1.6		1.8		2.1		2.5	ns
t_{RCKL}	Input HIGH to LOW (light load) (pad to R-Cell input)		1.8		2.0		2.3		2.7	ns
t_{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)		1.8		2.1		2.5		2.8	ns
t_{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)		1.8		2.1		2.4		2.8	ns
t_{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t_{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t_{RCKSW}	Maximum Skew (light load)		0.5		0.5		0.5		0.7	ns
t_{RCKSW}	Maximum Skew (50% load)		0.5		0.6		0.7		0.8	ns
t_{RCKSW}	Maximum Skew (100% load)		0.5		0.6		0.7		0.8	ns
TTL Output Module Timing[3]										
t_{DLH}	Data-to-Pad LOW to HIGH		1.6		1.9		2.1		2.5	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.6		1.9		2.1		2.5	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.1		2.4		2.8		3.2	ns
t_{ENZH}	Enable-to-Pad, Z to H		2.3		2.7		3.1		3.6	ns
t_{ENLZ}	Enable-to-Pad, L to Z		1.4		1.7		1.9		2.2	ns
t_{ENHZ}	Enable-to-Pad, H to Z		1.3		1.5		1.7		2.0	ns

Notes:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH}. For t_{ENZL} and t_{ENZH}, the loading is 5 pF.*

Table 1-19 • **A54SX16P Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Dedicated (Hardwired) Array Clock Network										
t_{HCKH}	Input LOW to HIGH (pad to R-Cell input)		1.2		1.4		1.5		1.8	ns
t_{HCKL}	Input HIGH to LOW (pad to R-Cell input)		1.2		1.4		1.6		1.9	ns
t_{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t_{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t_{HCKSW}	Maximum Skew		0.2		0.2		0.3		0.3	ns
t_{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f_{HMAX}	Maximum Frequency		350		320		280		240	MHz
Routed Array Clock Networks										
t_{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)		1.6		1.8		2.1		2.5	ns
t_{RCKL}	Input HIGH to LOW (Light Load) (pad to R-Cell input)		1.8		2.0		2.3		2.7	ns
t_{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)		1.8		2.1		2.5		2.8	ns
t_{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)		1.8		2.1		2.4		2.8	ns
t_{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t_{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t_{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t_{RCKSW}	Maximum Skew (light load)		0.5		0.5		0.5		0.7	ns
t_{RCKSW}	Maximum Skew (50% load)		0.5		0.6		0.7		0.8	ns
t_{RCKSW}	Maximum Skew (100% load)		0.5		0.6		0.7		0.8	ns
TTL Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		2.4		2.8		3.1		3.7	ns
t_{DHL}	Data-to-Pad HIGH to LOW		2.3		2.9		3.2		3.8	ns
t_{ENZL}	Enable-to-Pad, Z to L		3.0		3.4		3.9		4.6	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.3		3.8		4.3		5.0	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.3		2.7		3.0		3.5	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.8		3.2		3.7		4.3	ns

Note:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 10 pF loading.*

Table 1-19 • **A54SX16P Timing Characteristics (Continued)**
(Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

Parameter	Description	'–3' Speed		'–2' Speed		'–1' Speed		'Std' Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
TTL/PCI Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		1.5		1.7		2.0		2.3	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.9		2.2		2.4		2.9	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.3		2.6		3.0		3.5	ns
t_{ENZH}	Enable-to-Pad, Z to H		1.5		1.7		1.9		2.3	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.7		3.1		3.5		4.1	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.9		3.3		3.7		4.4	ns
PCI Output Module Timing[3]										
t_{DLH}	Data-to-Pad LOW to HIGH		1.8		2.0		2.3		2.7	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.7		2.0		2.2		2.6	ns
t_{ENZL}	Enable-to-Pad, Z to L		0.8		1.0		1.1		1.3	ns
t_{ENZH}	Enable-to-Pad, Z to H		1.2		1.2		1.5		1.8	ns
t_{ENLZ}	Enable-to-Pad, L to Z		1.0		1.1		1.3		1.5	ns
t_{ENHZ}	Enable-to-Pad, H to Z		1.1		1.3		1.5		1.7	ns
TTL Output Module Timing										
t_{DLH}	Data-to-Pad LOW to HIGH		2.1		2.5		2.8		3.3	ns
t_{DHL}	Data-to-Pad HIGH to LOW		2.0		2.3		2.6		3.1	ns
t_{ENZL}	Enable-to-Pad, Z to L		2.5		2.9		3.2		3.8	ns
t_{ENZH}	Enable-to-Pad, Z to H		3.0		3.5		3.9		4.6	ns
t_{ENLZ}	Enable-to-Pad, L to Z		2.3		2.7		3.1		3.6	ns
t_{ENHZ}	Enable-to-Pad, H to Z		2.9		3.3		3.7		4.4	ns

Note:

1. *For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn}, t_{RCO} + t_{RD1} + t_{PDn}, or t_{PD1} + t_{RD1} + t_{SUD}, whichever is appropriate.*
2. *Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.*
3. *Delays based on 10 pF loading.*

Pin Description

CLKA/B Clock A and B

These pins are 3.3 V / 5.0 V PCI/TTL clock inputs for clock distribution networks. The clock input is buffered prior to clocking the R-cells. If not used, this pin must be set LOW or HIGH on the board. It must not be left floating. (For A54SX72A, these clocks can be configured as bidirectional.)

GND Ground

LOW supply voltage.

HCLK Dedicated (hardwired) Array Clock

This pin is the 3.3 V / 5.0 V PCI/TTL clock input for sequential modules. This input is directly wired to each R-cell and offers clock speeds independent of the number of R-cells being driven. If not used, this pin must be set LOW or HIGH on the board. It must not be left floating.

I/O Input/Output

The I/O pin functions as an input, output, tristate, or bidirectional buffer. Based on certain configurations, input and output levels are compatible with standard TTL, LVTTL, 3.3 V PCI or 5.0 V PCI specifications. Unused I/O pins are automatically tristated by the Designer Series software.

NC No Connection

This pin is not connected to circuitry within the device.

PRA, I/O Probe A

The Probe A pin is used to output data from any user-defined design node within the device. This independent diagnostic pin can be used in conjunction with the Probe B pin to allow real-time diagnostic output of any signal path within the device. The Probe A pin can be used as a user-defined I/O when verification has been completed. The pin's probe capabilities can be permanently disabled to protect programmed design confidentiality.

PRB, I/O Probe B

The Probe B pin is used to output data from any node within the device. This diagnostic pin can be used in conjunction with the Probe A pin to allow real-time diagnostic output of any signal path within the device. The Probe B pin can be used as a user-defined I/O when verification has been completed. The pin's probe capabilities can be permanently disabled to protect programmed design confidentiality.

TCK Test Clock

Test clock input for diagnostic probe and device programming. In flexible mode, TCK becomes active when the TMS pin is set LOW (refer to Table 1-2 on page 1-6). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDI Test Data Input

Serial input for boundary scan testing and diagnostic probe. In flexible mode, TDI is active when the TMS pin is set LOW (refer to Table 1-2 on page 1-6). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDO Test Data Output

Serial output for boundary scan testing. In flexible mode, TDO is active when the TMS pin is set LOW (refer to Table 1-2 on page 1-6). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TMS Test Mode Select

The TMS pin controls the use of the IEEE 1149.1 Boundary Scan pins (TCK, TDI, TDO). In flexible mode when the TMS pin is set LOW, the TCK, TDI, and TDO pins are boundary scan pins (refer to Table 1-2 on page 1-6). Once the boundary scan pins are in test mode, they will remain in that mode until the internal boundary scan state machine reaches the "logic reset" state. At this point, the boundary scan pins will be released and will function as regular I/O pins. The "logic reset" state is reached 5 TCK cycles after the TMS pin is set HIGH. In dedicated test mode, TMS functions as specified in the IEEE 1149.1 specifications.

V_{CCI} Supply Voltage

Supply voltage for I/Os. See Table 1-1 on page 1-5.

V_{CCA} Supply Voltage

Supply voltage for Array. See Table 1-1 on page 1-5.

V_{CCR} Supply Voltage

Supply voltage for input tolerance (required for internal biasing). See Table 1-1 on page 1-5.

144-Pin TQFP			
Pin Number	**A54SX08 Function**	**A54SX16P Function**	**A54SX32 Function**
73	GND	GND	GND
74	I/O	I/O	I/O
75	I/O	I/O	I/O
76	I/O	I/O	I/O
77	I/O	I/O	I/O
78	I/O	I/O	I/O
79	V_{CCA}	V_{CCA}	V_{CCA}
80	V_{CCI}	V_{CCI}	V_{CCI}
81	GND	GND	GND
82	I/O	I/O	I/O
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	I/O	I/O	I/O
86	I/O	I/O	I/O
87	I/O	I/O	I/O
88	I/O	I/O	I/O
89	V_{CCA}	V_{CCA}	V_{CCA}
90	V_{CCR}	V_{CCR}	V_{CCR}
91	I/O	I/O	I/O
92	I/O	I/O	I/O
93	I/O	I/O	I/O
94	I/O	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	I/O	I/O	I/O
98	V_{CCA}	V_{CCA}	V_{CCA}
99	GND	GND	GND
100	I/O	I/O	I/O
101	GND	GND	GND
102	V_{CCI}	V_{CCI}	V_{CCI}
103	I/O	I/O	I/O
104	I/O	I/O	I/O
105	I/O	I/O	I/O
106	I/O	I/O	I/O
107	I/O	I/O	I/O
108	I/O	I/O	I/O

144-Pin TQFP			
Pin Number	**A54SX08 Function**	**A54SX16P Function**	**A54SX32 Function**
109	GND	GND	GND
110	I/O	I/O	I/O
111	I/O	I/O	I/O
112	I/O	I/O	I/O
113	I/O	I/O	I/O
114	I/O	I/O	I/O
115	V_{CCI}	V_{CCI}	V_{CCI}
116	I/O	I/O	I/O
117	I/O	I/O	I/O
118	I/O	I/O	I/O
119	I/O	I/O	I/O
120	I/O	I/O	I/O
121	I/O	I/O	I/O
122	I/O	I/O	I/O
123	I/O	I/O	I/O
124	I/O	I/O	I/O
125	CLKA	CLKA	CLKA
126	CLKB	CLKB	CLKB
127	V_{CCR}	V_{CCR}	V_{CCR}
128	GND	GND	GND
129	V_{CCA}	V_{CCA}	V_{CCA}
130	I/O	I/O	I/O
131	PRA, I/O	PRA, I/O	PRA, I/O
132	I/O	I/O	I/O
133	I/O	I/O	I/O
134	I/O	I/O	I/O
135	I/O	I/O	I/O
136	I/O	I/O	I/O
137	I/O	I/O	I/O
138	I/O	I/O	I/O
139	I/O	I/O	I/O
140	V_{CCI}	V_{CCI}	V_{CCI}
141	I/O	I/O	I/O
142	I/O	I/O	I/O
143	I/O	I/O	I/O
144	TCK, I/O	TCK, I/O	TCK, I/O

176-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	NC	I/O	I/O
4	I/O	I/O	I/O
5	I/O	I/O	I/O
6	I/O	I/O	I/O
7	I/O	I/O	I/O
8	I/O	I/O	I/O
9	I/O	I/O	I/O
10	TMS	TMS	TMS
11	V_{CCI}	V_{CCI}	V_{CCI}
12	NC	I/O	I/O
13	I/O	I/O	I/O
14	I/O	I/O	I/O
15	I/O	I/O	I/O
16	I/O	I/O	I/O
17	I/O	I/O	I/O
18	I/O	I/O	I/O
19	I/O	I/O	I/O
20	I/O	I/O	I/O
21	GND	GND	GND
22	V_{CCA}	V_{CCA}	V_{CCA}
23	GND	GND	GND
24	I/O	I/O	I/O
25	I/O	I/O	I/O
26	I/O	I/O	I/O
27	I/O	I/O	I/O
28	I/O	I/O	I/O
29	I/O	I/O	I/O
30	I/O	I/O	I/O
31	I/O	I/O	I/O
32	V_{CCI}	V_{CCI}	V_{CCI}
33	V_{CCA}	V_{CCA}	V_{CCA}
34	I/O	I/O	I/O

176-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
35	I/O	I/O	I/O
36	I/O	I/O	I/O
37	I/O	I/O	I/O
38	I/O	I/O	I/O
39	I/O	I/O	I/O
40	NC	I/O	I/O
41	I/O	I/O	I/O
42	NC	I/O	I/O
43	I/O	I/O	I/O
44	GND	GND	GND
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	I/O	I/O	I/O
49	I/O	I/O	I/O
50	I/O	I/O	I/O
51	I/O	I/O	I/O
52	V_{CCI}	V_{CCI}	V_{CCI}
53	I/O	I/O	I/O
54	NC	I/O	I/O
55	I/O	I/O	I/O
56	I/O	I/O	I/O
57	NC	I/O	I/O
58	I/O	I/O	I/O
59	I/O	I/O	I/O
60	I/O	I/O	I/O
61	I/O	I/O	I/O
62	I/O	I/O	I/O
63	I/O	I/O	I/O
64	PRB, I/O	PRB, I/O	PRB, I/O
65	GND	GND	GND
66	V_{CCA}	V_{CCA}	V_{CCA}
67	V_{CCR}	V_{CCR}	V_{CCR}
68	I/O	I/O	I/O

100-Pin VQFP		
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function
1	GND	GND
2	TDI, I/O	TDI, I/O
3	I/O	I/O
4	I/O	I/O
5	I/O	I/O
6	I/O	I/O
7	TMS	TMS
8	V_{CCI}	V_{CCI}
9	GND	GND
10	I/O	I/O
11	I/O	I/O
12	I/O	I/O
13	I/O	I/O
14	I/O	I/O
15	I/O	I/O
16	I/O	I/O
17	I/O	I/O
18	I/O	I/O
19	I/O	I/O
20	V_{CCI}	V_{CCI}
21	I/O	I/O
22	I/O	I/O
23	I/O	I/O
24	I/O	I/O
25	I/O	I/O
26	I/O	I/O
27	I/O	I/O
28	I/O	I/O
29	I/O	I/O
30	I/O	I/O
31	I/O	I/O
32	I/O	I/O
33	I/O	I/O
34	PRB, I/O	PRB, I/O

100-Pin VQFP		
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function
35	V_{CCA}	V_{CCA}
36	GND	GND
37	V_{CCR}	V_{CCR}
38	I/O	I/O
39	HCLK	HCLK
40	I/O	I/O
41	I/O	I/O
42	I/O	I/O
43	I/O	I/O
44	V_{CCI}	V_{CCI}
45	I/O	I/O
46	I/O	I/O
47	I/O	I/O
48	I/O	I/O
49	TDO, I/O	TDO, I/O
50	I/O	I/O
51	GND	GND
52	I/O	I/O
53	I/O	I/O
54	I/O	I/O
55	I/O	I/O
56	I/O	I/O
57	V_{CCA}	V_{CCA}
58	V_{CCI}	V_{CCI}
59	I/O	I/O
60	I/O	I/O
61	I/O	I/O
62	I/O	I/O
63	I/O	I/O
64	I/O	I/O
65	I/O	I/O
66	I/O	I/O
67	V_{CCA}	V_{CCA}
68	GND	GND

100-Pin VQFP		
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function
69	GND	GND
70	I/O	I/O
71	I/O	I/O
72	I/O	I/O
73	I/O	I/O
74	I/O	I/O
75	I/O	I/O
76	I/O	I/O
77	I/O	I/O
78	I/O	I/O
79	I/O	I/O
80	I/O	I/O
81	I/O	I/O
82	V_{CCI}	V_{CCI}
83	I/O	I/O
84	I/O	I/O
85	I/O	I/O
86	I/O	I/O
87	CLKA	CLKA
88	CLKB	CLKB
89	V_{CCR}	V_{CCR}
90	V_{CCA}	V_{CCA}
91	GND	GND
92	PRA, I/O	PRA, I/O
93	I/O	I/O
94	I/O	I/O
95	I/O	I/O
96	I/O	I/O
97	I/O	I/O
98	I/O	I/O
99	I/O	I/O
100	TCK, I/O	TCK, I/O

329-Pin PBGA	
Pin Number	**A54SX32 Function**
D3	I/O
D4	TCK, I/O
D5	I/O
D6	I/O
D7	I/O
D8	I/O
D9	I/O
D10	I/O
D11	V_{CCA}
D12	V_{CCR}
D13	I/O
D14	I/O
D15	I/O
D16	I/O
D17	I/O
D18	I/O
D19	I/O
D20	I/O
D21	I/O
D22	I/O
D23	I/O
E1	V_{CCI}
E2	I/O
E3	I/O
E4	I/O
E20	I/O
E21	I/O
E22	I/O
E23	I/O
F1	I/O
F2	TMS
F3	I/O
F4	I/O
F20	I/O
F21	I/O

329-Pin PBGA	
Pin Number	**A54SX32 Function**
F22	I/O
F23	I/O
G1	I/O
G2	I/O
G3	I/O
G4	I/O
G20	I/O
G21	I/O
G22	I/O
G23	GND
H1	I/O
H2	I/O
H3	I/O
H4	I/O
H20	V_{CCA}
H21	I/O
H22	I/O
H23	I/O
J1	NC
J2	I/O
J3	I/O
J4	I/O
J20	I/O
J21	I/O
J22	I/O
J23	I/O
K1	I/O
K2	I/O
K3	I/O
K4	I/O
K10	GND
K11	GND
K12	GND
K13	GND
K14	GND

329-Pin PBGA	
Pin Number	**A54SX32 Function**
K20	I/O
K21	I/O
K22	I/O
K23	I/O
L1	I/O
L2	I/O
L3	I/O
L4	V_{CCR}
L10	GND
L11	GND
L12	GND
L13	GND
L14	GND
L20	V_{CCR}
L21	I/O
L22	I/O
L23	NC
M1	I/O
M2	I/O
M3	I/O
M4	V_{CCA}
M10	GND
M11	GND
M12	GND
M13	GND
M14	GND
M20	V_{CCA}
M21	I/O
M22	I/O
M23	V_{CCI}
N1	I/O
N2	I/O
N3	I/O
N4	I/O
N10	GND

329-Pin PBGA	
Pin Number	**A54SX32 Function**
N11	GND
N12	GND
N13	GND
N14	GND
N20	NC
N21	I/O
N22	I/O
N23	I/O
P1	I/O
P2	I/O
P3	I/O
P4	I/O
P10	GND
P11	GND
P12	GND
P13	GND
P14	GND
P20	I/O
P21	I/O
P22	I/O
P23	I/O
R1	I/O
R2	I/O
R3	I/O
R4	I/O
R20	I/O
R21	I/O
R22	I/O
R23	I/O
T1	I/O
T2	I/O
T3	I/O
T4	I/O
T20	I/O
T21	I/O