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#### Understanding <u>Embedded - FPGAs (Field</u> <u>Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

#### **Applications of Embedded - FPGAs**

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

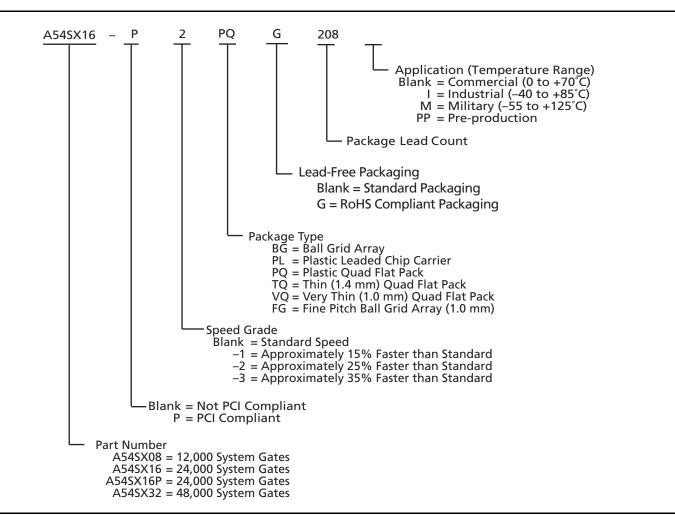
E·XFI

Details	
Product Status	Obsolete
Number of LABs/CLBs	1452
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	113
Number of Gates	24000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a54sx16p-2tq144i

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

# **Ordering Information**



## **Plastic Device Resources**

	User I/Os (including clock buffers)											
Device	PLCC 84-Pin	VQFP 100-Pin	PQFP 208-Pin	TQFP 144-Pin	TQFP 176-Pin	PBGA 313-Pin	PBGA 329-Pin	FBGA 144-Pin				
A54SX08	69	81	130	113	128	-	-	111				
A54SX16	-	81	175	-	147	-	-	-				
A54SX16P	-	81	175	113	147	-	-	-				
A54SX32	_	_	174	113	147	249	249	-				

Note: Package Definitions (Consult your local Actel sales representative for product availability):

PLCC = Plastic Leaded Chip Carrier

PQFP = Plastic Quad Flat Pack

TQFP = Thin Quad Flat Pack

VQFP = Very Thin Quad Flat Pack

PBGA = Plastic Ball Grid Array

FBGA = Fine Pitch (1.0 mm) Ball Grid Array



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### Package Pin Assignments

84-Pin PLCC	 	• •	• •	•	 	 •	 	•	 • •	 • •	•	 • •	••	 • •	•	 • •	 •	 • •	•	 		2-1
208-Pin PQFP																						
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176-Pin TQFP																						
100-Pin VQFP	 			•	 	 •	 	•	 	 	•	 • •	• •	 		 	 •	 	•	 	2	2-14
313-Pin PBGA	 				 		 	•	 	 		 		 		 		 	•	 	2	2-16
329-Pin PBGA	 			•	 	 •	 	•	 	 	•	 • •	• •	 		 	 •	 	•	 	2	2-19
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### Chip Architecture

The SX family chip architecture provides a unique approach to module organization and chip routing that delivers the best register/logic mix for a wide variety of new and emerging applications.

### **Module Organization**

Actel has arranged all C-cell and R-cell logic modules into horizontal banks called *clusters*. There are two types of *clusters*: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells. To increase design efficiency and device performance, Actel has further organized these modules into *SuperClusters* (Figure 1-4). SuperCluster 1 is a two-wide grouping of Type 1 clusters. SuperCluster 2 is a two-wide group containing one Type 1 cluster and one Type 2 cluster. SX devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flipflops.

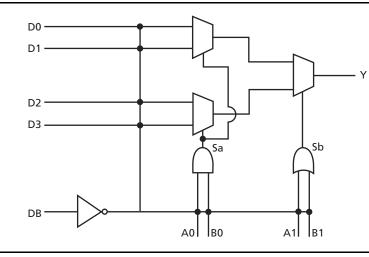
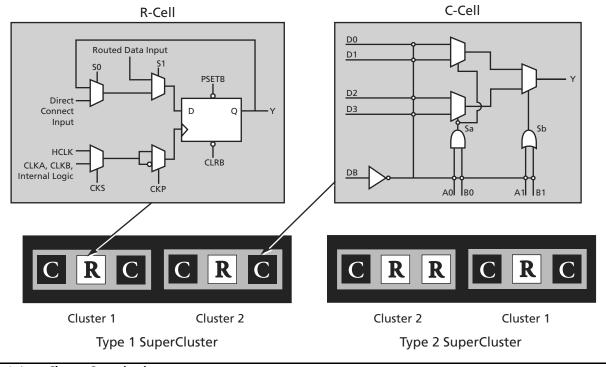


Figure 1-3 • C-Cell



*Figure 1-4* • Cluster Organization



DirectConnect is a horizontal routing resource that provides connections from a C-cell to its neighboring Rcell in a given SuperCluster. DirectConnect uses a hardwired signal path requiring no programmable interconnection to achieve its fast signal propagation time of less than 0.1 ns.

FastConnect enables horizontal routing between any two logic modules within a given SuperCluster and vertical routing with the SuperCluster immediately below it. Only one programmable connection is used in a FastConnect path, delivering maximum pin-to-pin propagation of 0.4 ns.

In addition to DirectConnect and FastConnect, the architecture makes use of two globally oriented routing resources known as segmented routing and high-drive routing. The Actel segmented routing structure provides a variety of track lengths for extremely fast routing between SuperClusters. The exact combination of track lengths and antifuses within each path is chosen by the 100 percent automatic place-and-route software to minimize signal propagation delays.

The Actel high-drive routing structure provides three clock networks. The first clock, called HCLK, is hardwired from the HCLK buffer to the clock select multiplexer (MUX) in each R-cell. This provides a fast propagation path for the clock signal, enabling the 3.7 ns clock-to-out (pin-to-pin) performance of the SX devices. The hardwired clock is tuned to provide clock skew as low as 0.25 ns. The remaining two clocks (CLKA, CLKB) are global clocks that can be sourced from external pins or from internal logic signals within the SX device.

### **Other Architectural Features**

#### Technology

The Actel SX family is implemented on a high-voltage twin-well CMOS process using 0.35  $\mu$  design rules. The metal-to-metal antifuse is made up of a combination of amorphous silicon and dielectric material with barrier metals and has a programmed ("on" state) resistance of 25  $\Omega$  with a capacitance of 1.0 fF for low signal impedance.

#### Performance

The combination of architectural features described above enables SX devices to operate with internal clock frequencies exceeding 300 MHz, enabling very fast execution of even complex logic functions. Thus, the SX family is an optimal platform upon which to integrate the functionality previously contained in multiple CPLDs. In addition, designs that previously would have required a gate array to meet performance goals can now be integrated into an SX device with dramatic improvements in cost and time to market. Using timingdriven place-and-route tools, designers can achieve highly deterministic device performance. With SX devices, designers do not need to use complicated performance-enhancing design techniques such as the use of redundant logic to reduce fanout on critical nets or the instantiation of macros in HDL code to achieve high performance.

#### I/O Modules

Each I/O on an SX device can be configured as an input, an output, a tristate output, or a bidirectional pin.

Even without the inclusion of dedicated I/O registers, these I/Os, in combination with array registers, can achieve clock-to-out (pad-to-pad) timing as fast as 3.7 ns. I/O cells that have embedded latches and flip-flops require instantiation in HDL code; this is a design complication not encountered in SX FPGAs. Fast pin-to-pin timing ensures that the device will have little trouble interfacing with any other device in the system, which in turn enables parallel design of system components and reduces overall design time.

#### **Power Requirements**

The SX family supports 3.3 V operation and is designed to tolerate 5.0 V inputs. (Table 1-1). Power consumption is extremely low due to the very short distances signals are required to travel to complete a circuit. Power requirements are further reduced because of the small number of low-resistance antifuses in the path. The antifuse architecture does not require active circuitry to hold a charge (as do SRAM or EPROM), making it the lowest power architecture on the market.

Denter		V	V		Maniana Outrat Daire
Device	V <sub>CCA</sub>	V <sub>CCI</sub>	V <sub>CCR</sub>	Maximum Input Tolerance	Maximum Output Drive
A54SX08 A54SX16 A54SX32	3.3 V	3.3 V	5.0 V	5.0 V	3.3 V
A54SX16-P*	3.3 V	3.3 V	3.3 V	3.3 V	3.3 V
	3.3 V	3.3 V	5.0 V	5.0 V	3.3 V
	3.3 V	5.0 V	5.0 V	5.0 V	5.0 V

**Note:** \*A54SX16-P has three different entries because it is capable of both a 3.3 V and a 5.0 V drive.

#### SX Family FPGAs

### Boundary Scan Testing (BST)

All SX devices are IEEE 1149.1 compliant. SX devices offer superior diagnostic and testing capabilities by providing Boundary Scan Testing (BST) and probing capabilities. These functions are controlled through the special test pins in conjunction with the program fuse. The functionality of each pin is described in Table 1-2. In the dedicated test mode, TCK, TDI, and TDO are dedicated pins and cannot be used as regular I/Os. In flexible mode, TMS should be set HIGH through a pull-up resistor of 10 k $\Omega$ . TMS can be pulled LOW to initiate the test sequence.

The program fuse determines whether the device is in dedicated or flexible mode. The default (fuse not blown) is flexible mode.

Table 1-2 •	Boundary Scan Pin Functionality
-------------	---------------------------------

Program Fuse Blown (Dedicated Test Mode)	Program Fuse Not Blown (Flexible Mode)
TCK, TDI, TDO are dedicated BST pins.	TCK, TDI, TDO are flexible and may be used as I/Os.
No need for pull-up resistor for TMS	Use a pull-up resistor of 10 $k\Omega$ on TMS.

### **Dedicated Test Mode**

In Dedicated mode, all JTAG pins are reserved for BST; designers cannot use them as regular I/Os. An internal pull-up resistor is automatically enabled on both TMS and TDI pins, and the TMS pin will function as defined in the IEEE 1149.1 (JTAG) specification.

To select Dedicated mode, users need to reserve the JTAG pins in Actel's Designer software by checking the "Reserve JTAG" box in "Device Selection Wizard" (Figure 1-7). JTAG pins comply with LVTTL/TTL I/O specification regardless of whether they are used as a user I/O or a JTAG I/O. Refer to the Table 1-5 on page 1-8 for detailed specifications.

Figure 1-7 • Device Selection Wizard

#### **Development Tool Support**

The SX family of FPGAs is fully supported by both the Actel Libero<sup>®</sup> Integrated Design Environment (IDE) and Designer FPGA Development software. Actel Libero IDE is a design management environment, seamlessly integrating design tools while guiding the user through the design flow, managing all design and log files, and passing necessary design data among tools. Libero IDE allows users to integrate both schematic and HDL synthesis into a single flow and verify the entire design in a single environment. Libero IDE includes Synplify<sup>®</sup> for Actel from Synplicity<sup>®</sup>, ViewDraw<sup>®</sup> for Actel from Mentor Graphics<sup>®</sup>, ModelSim<sup>®</sup> HDL Simulator from Mentor Graphics, WaveFormer Lite™ from SynaptiCAD<sup>™</sup>, and Designer software from Actel. Refer to the Libero IDE flow diagram (located on the Actel website) for more information.

Actel Designer software is a place-and-route tool and provides a comprehensive suite of backend support tools for FPGA development. The Designer software includes timing-driven place-and-route, and a world-class integrated static timing analyzer and constraints editor. With the Designer software, a user can select and lock package pins while only minimally impacting the results of place-and-route. Additionally, the back-annotation flow is compatible with all the major simulators, and the simulation results can be cross-probed with Silicon Explorer II, Actel integrated verification and logic analysis tool. Another tool included in the Designer software is the SmartGen core generator, which easily creates popular and commonly used logic functions for implementation into your schematic or HDL design. Actel Designer software is compatible with the most popular FPGA design entry and verification tools from companies such as Mentor Graphics, Synplicity, Synopsys<sup>®</sup>, and Cadence<sup>®</sup> Design Systems. The Designer software is available for both the Windows® and UNIX® operating systems.

#### **Probe Circuit Control Pins**

The Silicon Explorer II tool uses the boundary scan ports (TDI, TCK, TMS, and TDO) to select the desired nets for verification. The selected internal nets are assigned to the PRA/PRB pins for observation. Figure 1-8 on page 1-7 illustrates the interconnection between Silicon Explorer II and the FPGA to perform in-circuit verification.

#### **Design Considerations**

The TDI, TCK, TDO, PRA, and PRB pins should not be used as input or bidirectional ports. Because these pins are active during probing, critical signals input through these pins are not available while probing. In addition, the Security Fuse should not be programmed because doing so disables the Probe Circuitry.



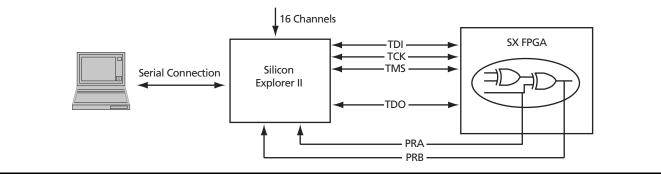


Figure 1-8 • Probe Setup

## Programming

Device programming is supported through Silicon Sculptor series of programmers. In particular, Silicon Sculptor II are compact, robust, single-site and multi-site device programmer for the PC.

With standalone software, Silicon Sculptor II allows concurrent programming of multiple units from the same PC, ensuring the fastest programming times possible. Each fuse is subsequently verified by Silicon Sculptor II to insure correct programming. In addition, integrity tests ensure that no extra fuses are programmed. Silicon Sculptor II also provides extensive hardware self-testing capability. The procedure for programming an SX device using Silicon Sculptor II are as follows:

- 1. Load the .AFM file
- 2. Select the device to be programmed
- 3. Begin programming

When the design is ready to go to production, Actel offers device volume-programming services either through distribution partners or via in-house programming from the factory.

For more details on programming SX devices, refer to the *Programming Antifuse Devices* application note and the *Silicon Sculptor II User's Guide*.

## **3.3 V / 5 V Operating Conditions** *Table 1-3* • Absolute Maximum Ratings<sup>1</sup>

Symbol	Parameter	Limits	Units
V <sub>CCR</sub> <sup>2</sup>	DC Supply Voltage <sup>3</sup>	-0.3 to + 6.0	V
V <sub>CCA</sub> <sup>2</sup>	DC Supply Voltage	-0.3 to + 4.0	V
V <sub>CCI</sub> <sup>2</sup>	DC Supply Voltage (A54SX08, A54SX16, A54SX32)	-0.3 to + 4.0	V
V <sub>CCI</sub> <sup>2</sup>	DC Supply Voltage (A54SX16P)	-0.3 to + 6.0	V
VI	Input Voltage	-0.5 to + 5.5	V
V <sub>O</sub>	Output Voltage	-0.5 to + 3.6	V
I <sub>IO</sub>	I/O Source Sink Current <sup>3</sup>	-30 to + 5.0	mA
T <sub>STG</sub>	Storage Temperature	-65 to +150	°C

Notes:

1. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum rated conditions for extended periods may affect device reliability. Device should not be operated outside the Recommended Operating Conditions.

2.  $V_{CCR}$  in the A54SX16P must be greater than or equal to  $V_{CCI}$  during power-up and power-down sequences and during normal operation.

3. Device inputs are normally high impedance and draw extremely low current. However, when input voltage is greater than  $V_{CC}$  + 0.5 V or less than GND – 0.5 V, the internal protection diodes will forward-bias and can draw excessive current.

# PCI Compliance for the SX Family

The SX family supports 3.3 V and 5.0 V PCI and is compliant with the PCI Local Bus Specification Rev. 2.1.

Table 1-6 •	A54SX16P DC Specifications (5.0 V PCI Operation)	
-------------	--------------------------------------------------	--

Symbol	Parameter	Condition	Min.	Max.	Units
V <sub>CCA</sub>	Supply Voltage for Array		3.0	3.6	V
V <sub>CCR</sub>	Supply Voltage required for Internal Biasing		4.75	5.25	V
V <sub>CCI</sub>	Supply Voltage for I/Os		4.75	5.25	V
V <sub>IH</sub>	Input High Voltage <sup>1</sup>		2.0	$V_{CC} + 0.5$	V
V <sub>IL</sub>	Input Low Voltage <sup>1</sup>		-0.5	0.8	V
I <sub>IH</sub>	Input High Leakage Current	V <sub>IN</sub> = 2.7		70	μA
IIL	Input Low Leakage Current	V <sub>IN</sub> = 0.5		-70	μA
V <sub>OH</sub>	Output High Voltage	I <sub>OUT</sub> = -2 mA	2.4		V
V <sub>OL</sub>	Output Low Voltage <sup>2</sup>	I <sub>OUT</sub> = 3 mA, 6 mA		0.55	V
C <sub>IN</sub>	Input Pin Capacitance <sup>3</sup>			10	pF
C <sub>CLK</sub>	CLK Pin Capacitance		5	12	pF
C <sub>IDSEL</sub>	IDSEL Pin Capacitance <sup>4</sup>			8	pF

Notes:

1. Input leakage currents include hi-Z output leakage for all bidirectional buffers with tristate outputs.

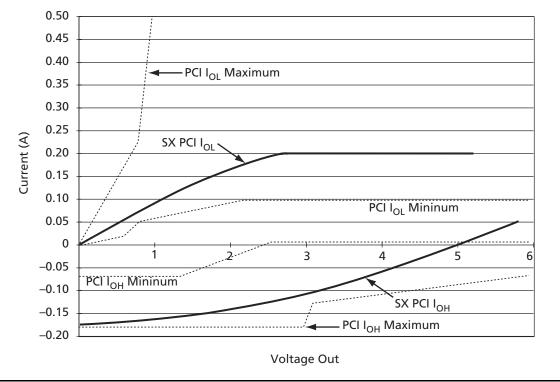
2. Signals without pull-up resistors must have 3 mA low output current. Signals requiring pull-up must have 6 mA; the latter include, FRAME#, IRDY#, TRDY#, DEVSEL#, STOP#, SERR#, PERR#, LOCK#, and, when used, AD[63::32], C/BE[7::4]#, PAR64, REQ64#, and ACK64#.

3. Absolute maximum pin capacitance for a PCI input is 10 pF (except for CLK).

4. Lower capacitance on this input-only pin allows for non-resistive coupling to AD[xx].



Figure 1-9 shows the 5.0 V PCI V/I curve and the minimum and maximum PCI drive characteristics of the A54SX16P device.



### Figure 1-9 • 5.0 V PCI Curve for A54SX16P Device

 $I_{OH} = 11.9 \times (V_{OUT} - 5.25) \times (V_{OUT} + 2.45)$ for V<sub>CC</sub> > V<sub>OUT</sub> > 3.1 V  $I_{OL} = 78.5 \times V_{OUT} \times (4.4 - V_{OUT})$  for 0 V < V\_{OUT} < 0.71 V

EQ 1-1

EQ 1-2

# A54SX16P AC Specifications (3.3 V PCI Operation)

Symbol	Parameter	Condition	Min.	Max.	Units
	Switching Current High	$0 < V_{OUT} \le 0.3 V_{CC}^{1}$			mA
1		$0.3V_{CC} \le V_{OUT} < 0.9V_{CC}^{-1}$	-12V <sub>CC</sub>		mA
IOH(AC)		$0.7V_{CC} < V_{OUT} < V_{CC}^{1, 2}$	–17.1 + (V <sub>CC</sub> – V <sub>OUT</sub> )	EQ 1-3 on page 1-14	
	(Test Point)	$V_{OUT} = 0.7 V_{CC}^2$		-32V <sub>CC</sub>	mA
	Switching Current High	$V_{CC} > V_{OUT} \ge 0.6 V_{CC}^{1}$			mA
1		$0.6V_{CC} > V_{OUT} > 0.1V_{CC}^{1}$	16V <sub>CC</sub>		mA
I <sub>OL(AC)</sub>		$0.18V_{CC} > V_{OUT} > 0^{1, 2}$	26.7V <sub>OUT</sub>	EQ 1-4 on page 1-14	mA
	(Test Point)	$V_{OUT} = 0.18 V_{CC}^2$		38V <sub>CC</sub>	
I <sub>CL</sub>	Low Clamp Current	$-3 < V_{IN} \le -1$	-25 + (V <sub>IN</sub> + 1)/0.015		mA
I <sub>CH</sub>	High Clamp Current	$-3 < V_{IN} \le -1$	25 + (V <sub>IN</sub> – V <sub>OUT</sub> – 1)/0.015		mA
slew <sub>R</sub>	Output Rise Slew Rate <sup>3</sup>	$0.2V_{CC}$ to $0.6V_{CC}$ load	1	4	V/ns
slew <sub>F</sub>	Output Fall Slew Rate <sup>3</sup>	$0.6V_{CC}$ to $0.2V_{CC}$ load	1	4	V/ns

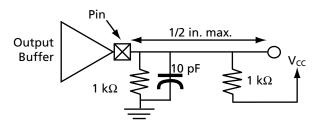
#### Table 1-9 • A54SX16P AC Specifications (3.3 V PCI Operation)

Notes:

1. Refer to the V/I curves in Figure 1-10 on page 1-14. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half size output drivers may be used on these signals. This specification does not apply to CLK and RST# which are system outputs. "Switching Current High" specification are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD# which are open drain outputs.

2. Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (C and D) are provided with the respective diagrams in Figure 1-10 on page 1-14. The equation defined maxima should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.

3. This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per the latest revision of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is required (the maximum is no longer simply a guideline). Rise slew rate does not apply to open drain outputs.



### SX Family FPGAs

#### Table 1-15 • Package Thermal Characteristics

Package Type	Pin Count	θ <sub>jc</sub>	θ <sub>ja</sub> Still Air	$^{ heta_{ja}}$ 300 ft/min.	Units
Plastic Leaded Chip Carrier (PLCC)	84	12	32	22	°C/W
Thin Quad Flat Pack (TQFP)	144	11	32	24	°C/W
Thin Quad Flat Pack (TQFP)	176	11	28	21	°C/W
Very Thin Quad Flatpack (VQFP)	100	10	38	32	°C/W
Plastic Quad Flat Pack (PQFP) without Heat Spreader	208	8	30	23	°C/W
Plastic Quad Flat Pack (PQFP) with Heat Spreader	208	3.8	20	17	°C/W
Plastic Ball Grid Array (PBGA)	272	3	20	14.5	°C/W
Plastic Ball Grid Array (PBGA)	313	3	23	17	°C/W
Plastic Ball Grid Array (PBGA)	329	3	18	13.5	°C/W
Fine Pitch Ball Grid Array (FBGA)	144	3.8	38.8	26.7	°C/W

Note: SX08 does not have a heat spreader.

#### Table 1-16 • Temperature and Voltage Derating Factors\*

			Junct	ion Temperat	ure		
V <sub>CCA</sub>	-55	-40	0	25	70	85	125
3.0	0.75	0.78	0.87	0.89	1.00	1.04	1.16
3.3	0.70	0.73	0.82	0.83	0.93	0.97	1.08
3.6	0.66	0.69	0.77	0.78	0.87	0.92	1.02

**Note:** \*Normalized to worst-case commercial,  $T_J = 70^{\circ}$ C,  $V_{CCA} = 3.0 V$ 



#### Table 1-17 A54SX08 Timing Characteristics (Continued)

(Worst-Case Commercial Conditions,	V <sub>CCR</sub> = 4.75 V, V <sub>CC</sub>	<sub>A</sub> ,V <sub>CCI</sub> = 3.0 V, T <sub>J</sub> = 70°C)
------------------------------------	--------------------------------------------	----------------------------------------------------------------

			5peed	'-2' \$	Speed	'-1' \$	5peed	'Std'	Speed	
Parameter	Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Units
Dedicated (	Hardwired) Array Clock Network									
t <sub>HCKH</sub>	Input LOW to HIGH (pad to R-Cell input)		1.0		1.1		1.3		1.5	ns
t <sub>HCKL</sub>	Input HIGH to LOW (pad to R-Cell input)		1.0		1.2		1.4		1.6	ns
t <sub>HPWH</sub>	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t <sub>HPWL</sub>	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t <sub>HCKSW</sub>	Maximum Skew		0.1		0.2		0.2		0.2	ns
t <sub>HP</sub>	Minimum Period	2.7		3.1		3.6		4.2		ns
f <sub>HMAX</sub>	Maximum Frequency		350		320		280		240	MHz
Routed Arra	Array Clock Networks									
t <sub>RCKH</sub>	Input LOW to HIGH (light load) (pad to R-Cell input)		1.3		1.5		1.7		2.0	ns
t <sub>RCKL</sub>	Input HIGH to LOW (light load) (pad to R-Cell Input)		1.4		1.6		1.8		2.1	ns
t <sub>RCKH</sub>	Input LOW to HIGH (50% load) (pad to R-Cell input)		1.4		1.7		1.9		2.2	ns
t <sub>RCKL</sub>	Input HIGH to LOW (50% load) (pad to R-Cell input)		1.5		1.7		2.0		2.3	ns
t <sub>RCKH</sub>	Input LOW to HIGH (100% load) (pad to R-Cell input)		1.5		1.7		1.9		2.2	ns
t <sub>RCKL</sub>	Input HIGH to LOW (100% load) (pad to R-Cell input)		1.5		1.8		2.0		2.3	ns
t <sub>RPWH</sub>	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t <sub>RPWL</sub>	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t <sub>RCKSW</sub>	Maximum Skew (light load)		0.1		0.2		0.2		0.2	ns
t <sub>RCKSW</sub>	Maximum Skew (50% load)		0.3		0.3		0.4		0.4	ns
t <sub>RCKSW</sub>	Maximum Skew (100% load)		0.3		0.3		0.4		0.4	ns
TTL Output	Module Timing1									
t <sub>DLH</sub>	Data-to-Pad LOW to HIGH		1.6		1.9		2.1		2.5	ns
t <sub>DHL</sub>	Data-to-Pad HIGH to LOW		1.6		1.9		2.1		2.5	ns
t <sub>ENZL</sub>	Enable-to-Pad, Z to L		2.1		2.4		2.8		3.2	ns
t <sub>ENZH</sub>	Enable-to-Pad, Z to H		2.3		2.7		3.1		3.6	ns
t <sub>ENLZ</sub>	Enable-to-Pad, L to Z		1.4		1.7		1.9		2.2	ns

Note:

1. For dual-module macros, use  $t_{PD} + t_{RD1} + t_{PDn}$ ,  $t_{RCO} + t_{RD1} + t_{PDn}$ , or  $t_{PD1} + t_{RD1} + t_{SUD}$ , whichever is appropriate.

2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

## A54SX16 Timing Characteristics

Table 1-18 • A54SX16 Timing Characteristics

(Worst-Case Commercial Conditions, V<sub>CCR</sub> = 4.75 V, V<sub>CCA</sub>, V<sub>CCI</sub> = 3.0 V, T<sub>J</sub> = 70°C)

			Speed	'-2' 9	5peed	'-1' 9	Speed	'Std'	Speed	
Parameter	Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Units
C-Cell Prop	agation Delays <sup>1</sup>									
t <sub>PD</sub>	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted R	outing Delays <sup>2</sup>									
t <sub>DC</sub>	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t <sub>FC</sub>	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t <sub>RD1</sub>	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t <sub>RD2</sub>	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t <sub>RD3</sub>	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t <sub>RD4</sub>	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t <sub>RD8</sub>	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t <sub>RD12</sub>	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timi	ng									
t <sub>RCO</sub>	Sequential Clock-to-Q		0.8		1.1		1.2		1.4	ns
t <sub>CLR</sub>	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t <sub>PRESET</sub>	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t <sub>SUD</sub>	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t <sub>HD</sub>	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t <sub>WASYN</sub>	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Mod	le Propagation Delays									
t <sub>INYH</sub>	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t <sub>INYL</sub>	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted I	nput Routing Delays <sup>2</sup>									
t <sub>IRD1</sub>	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t <sub>IRD2</sub>	FO = 2 Routing Delay		0.6		0.7		0.8		0.9	ns
t <sub>IRD3</sub>	FO = 3 Routing Delay		0.8		0.9		1.0		1.2	ns
t <sub>IRD4</sub>	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t <sub>IRD8</sub>	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t <sub>IRD12</sub>	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

#### Notes:

1. For dual-module macros, use  $t_{PD} + t_{RD1} + t_{PDn}$ ,  $t_{RCO} + t_{RD1} + t_{PDn}$ , or  $t_{PD1} + t_{RD1} + t_{SUD}$ , whichever is appropriate.

2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

3. Delays based on 35 pF loading, except  $t_{ENZL}$  and  $t_{ENZH}$ . For  $t_{ENZL}$  and  $t_{ENZH}$ , the loading is 5 pF.

## A54SX32 Timing Characteristics

Table 1-20 • A54SX32 Timing Characteristics

(Worst-Case Commercial Conditions,  $V_{CCR}$ = 4.75 V,  $V_{CCA}$ ,  $V_{CCI}$  = 3.0 V, T<sub>J</sub> = 70°C)

		'-3' 9	5peed	'-2' \$	Speed	'-1' 9	5peed	'Std'		
Parameter	Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Units
C-Cell Prop	agation Delays <sup>1</sup>									
t <sub>PD</sub>	Internal Array Module		0.6		0.7		0.8		0.9	ns
Predicted R	outing Delays <sup>2</sup>									
t <sub>DC</sub>	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t <sub>FC</sub>	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t <sub>RD1</sub>	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t <sub>RD2</sub>	FO = 2 Routing Delay		0.7		0.8		0.9		1.0	ns
t <sub>RD3</sub>	FO = 3 Routing Delay		1.0		1.2		1.4		1.6	ns
t <sub>RD4</sub>	FO = 4 Routing Delay		1.4		1.6		1.8		2.1	ns
t <sub>RD8</sub>	FO = 8 Routing Delay		2.7		3.1		3.5		4.1	ns
t <sub>RD12</sub>	FO = 12 Routing Delay		4.0		4.7		5.3		6.2	ns
R-Cell Timi	ng									
t <sub>RCO</sub>	Sequential Clock-to-Q		0.8		1.1		1.3		1.4	ns
t <sub>CLR</sub>	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t <sub>PRESET</sub>	Asynchronous Preset-to-Q		0.7		0.8		0.9		1.0	ns
t <sub>SUD</sub>	Flip-Flop Data Input Set-Up	0.5		0.6		0.7		0.8		ns
t <sub>HD</sub>	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t <sub>WASYN</sub>	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Modu	le Propagation Delays									
t <sub>INYH</sub>	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t <sub>INYL</sub>	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted I	nput Routing Delays <sup>2</sup>									
t <sub>IRD1</sub>	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t <sub>IRD2</sub>	FO = 2 Routing Delay		0.7		0.8		0.9		1.0	ns
t <sub>IRD3</sub>	FO = 3 Routing Delay		1.0		1.2		1.4		1.6	ns
t <sub>IRD4</sub>	FO = 4 Routing Delay		1.4		1.6		1.8		2.1	ns
t <sub>IRD8</sub>	FO = 8 Routing Delay		2.7		3.1		3.5		4.1	ns
t <sub>IRD12</sub>	FO = 12 Routing Delay		4.0		4.7		5.3		6.2	ns

Note:

1. For dual-module macros, use  $t_{PD} + t_{RD1} + t_{PDn}$ ,  $t_{RCO} + t_{RD1} + t_{PDn}$ , or  $t_{PD1} + t_{RD1} + t_{SUD}$ , whichever is appropriate.

2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

3. Delays based on 35 pF loading, except  $t_{ENZL}$  and  $t_{ENZH}$ . For  $t_{ENZL}$  and  $t_{ENZH}$  the loading is 5 pF.



# 208-Pin PQFP

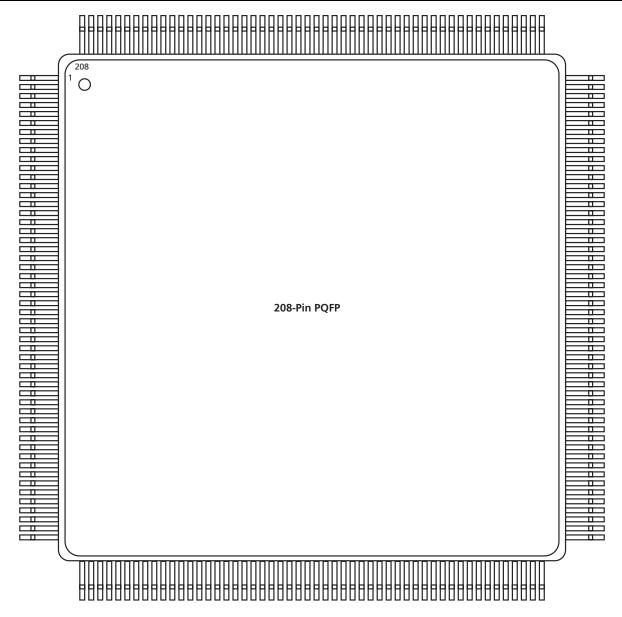


Figure 2-2 • 208-Pin PQFP (Top View)

### Note



# 144-Pin TQFP

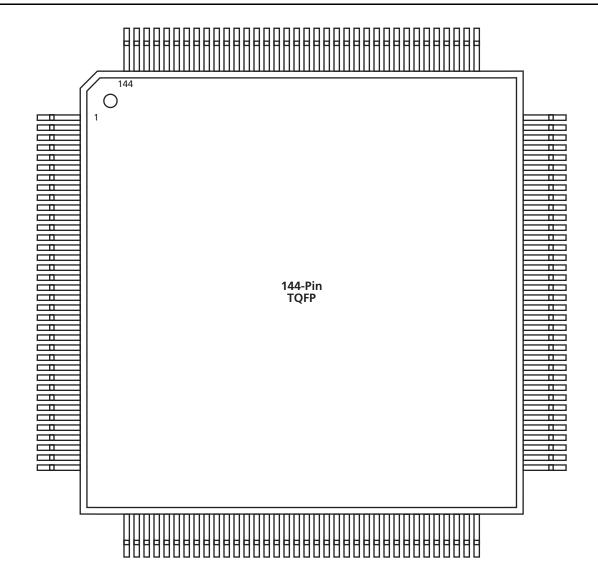


Figure 2-3 • 144-Pin TQFP (Top View)

#### Note

# 176-Pin TQFP

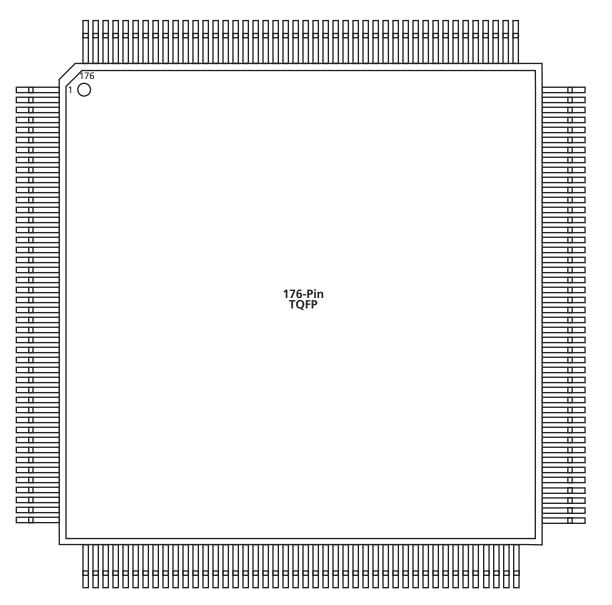


Figure 2-4 • 176-Pin TQFP (Top View)

### Note



313-Pi	n PBGA	313-Pi	n PBGA	313-Pi	n PBGA	313-Pin PBGA				
Pin Number	A54SX32 Function	Pin Number	A54SX32 Function	Pin Number	A54SX32 Function	Pin Number	A54SX32 Function			
A1	GND	AC5	I/O	B10	I/O	E15	I/O			
A3	NC	AC7	I/O	B12	I/O	E17	I/O			
A5	I/O	AC9	I/O	B14	I/O	E19	I/O			
A7	I/O	AC11	I/O	B16	I/O	E21	I/O			
A9	I/O	AC13	V <sub>CCR</sub>	B18	I/O	E23	I/O			
A11	I/O	AC15	I/O	B20	I/O	E25	I/O			
A13	V <sub>CCR</sub>	AC17	I/O	B22	I/O	F2	I/O			
A15	I/O	AC19	I/O	B24	I/O	F4	I/O			
A17	I/O	AC21	I/O	C1	TDI, I/O	F6	NC			
A19	I/O	AC23	I/O	C3	I/O	F8	I/O			
A21	I/O	AC25	NC	C5	NC	F10	NC			
A23	NC	AD2	GND	С7	I/O	F12	I/O			
A25	GND	AD4	I/O	С9	I/O	F14	I/O			
AA1	I/O	AD6	V <sub>CCI</sub>	C11	I/O	F16	NC			
AA3	I/O	AD8	I/O	C13	V <sub>CCI</sub>	F18	I/O			
AA5	NC	AD10	I/O	C15	I/O	F20	I/O			
AA7	I/O	AD12	PRB, I/O	C17	I/O	F22	I/O			
AA9	NC	AD14	I/O	C19	V <sub>CCI</sub>	F24	I/O			
AA11	I/O	AD16	I/O	C21	I/O	G1	I/O			
AA13	I/O	AD18	I/O	C23	I/O	G3	TMS			
AA15	I/O	AD20	I/O	C25	NC	G5	I/O			
AA17	I/O	AD22	NC	D2	I/O	G7	I/O			
AA19	I/O	AD24	I/O	D4	NC	G9	V <sub>CCI</sub>			
AA21	I/O	AE1	NC	D6	I/O	G11	I/O			
AA23	NC	AE3	I/O	D8	I/O	G13	CLKB			
AA25	I/O	AE5	I/O	D10	I/O	G15	I/O			
AB2	NC	AE7	I/O	D12	I/O	G17	I/O			
AB4	NC	AE9	I/O	D14	I/O	G19	I/O			
AB6	I/O	AE11	I/O	D16	I/O	G21	I/O			
AB8	I/O	AE13	V <sub>CCA</sub>	D18	I/O	G23	I/O			
AB10	I/O	AE15	I/O	D20	I/O	G25	I/O			
AB12	I/O	AE17	I/O	D22	I/O	H2	I/O			
AB14	I/O	AE19	I/O	D24	NC	H4	I/O			
AB16	I/O	AE21	I/O	E1	I/O	H6	I/O			
AB18	V <sub>CCI</sub>	AE23	TDO, I/O	E3	NC	H8	I/O			
AB20	NC	AE25	GND	E5	I/O	H10	I/O			
AB22	I/O	B2	TCK, I/O	E7	I/O	H12	PRA, I/O			
AB24	I/O	B4	I/O	E9	I/O	H14	I/O			
AC1	I/O	B6	I/O	E11	I/O	H16	I/O			
AC3	I/O	B8	I/O	E13	V <sub>CCA</sub>	H18	NC			

# 329-Pin PBGA

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	
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Figure 2-7 • 329-Pin PBGA (Top View)

#### Note

# **Datasheet Information**

# List of Changes

The following table lists critical changes that were made in the current version of the document.

<b>Previous Version</b>	Changes in Current Version (v3.2)	Page
v3.1	The "Ordering Information" was updated to include RoHS information.	1-ii
(June 2003)	The Product Plan was removed since all products have been released.	N/A
	Information concerning the TRST pin in the "Probe Circuit Control Pins" section was removed.	1-6
	The "Dedicated Test Mode" section is new.	1-6
	The "Programming" section is new.	1-7
	A note was added to the "Power-Up Sequencing" table.	1-15
	A note was added to the "Power-Down Sequencing" table. The 3.3 V comments were updated for the following devices: A54SX08, A54SX16, A54SX32.	1-15
	U11 and U13 were added to the "313-Pin PBGA" table.	2-17
v3.0.1	Storage temperature in Table 1-3 was updated.	1-7
	Table 1-1 was updated.	1-5

# **Datasheet Categories**

In order to provide the latest information to designers, some datasheets are published before data has been fully characterized. Datasheets are designated as "Product Brief," "Advanced," "Production," and "Datasheet Supplement." The definitions of these categories are as follows:

## **Product Brief**

The product brief is a summarized version of a datasheet (advanced or production) containing general product information. This brief gives an overview of specific device and family information.

## Advanced

This datasheet version contains initial estimated information based on simulation, other products, devices, or speed grades. This information can be used as estimates, but not for production.

## Unmarked (production)

This datasheet version contains information that is considered to be final.

## **Datasheet Supplement**

The datasheet supplement gives specific device information for a derivative family that differs from the general family datasheet. The supplement is to be used in conjunction with the datasheet to obtain more detailed information and for specifications that do not differ between the two families.

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The products described in this datasheet are subject to the International Traffic in Arms Regulations (ITAR) or the Export Administration Regulations (EAR). They may require an approved export license prior to their export. An export can include a release or disclosure to a foreign national inside or outside the United States.