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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

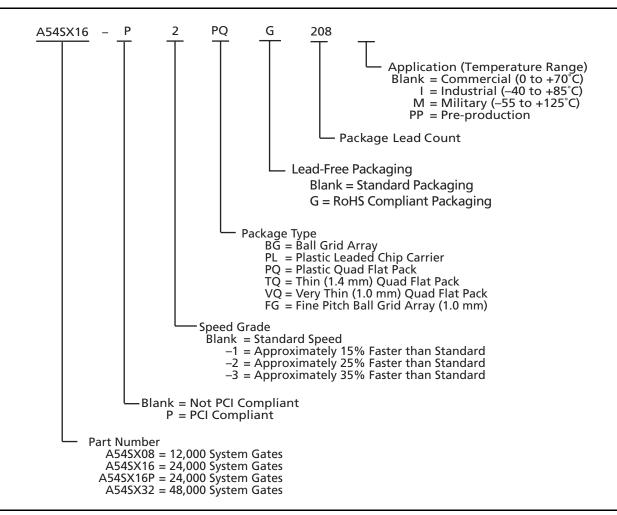
The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	2880
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	174
Number of Gates	48000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 85°C (TA)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/a54sx32-pqg208i

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Ordering Information



Plastic Device Resources

	User I/Os (including clock buffers)										
Device	PLCC 84-Pin	VQFP 100-Pin	PQFP 208-Pin	TQFP 144-Pin	TQFP 176-Pin	PBGA 313-Pin	PBGA 329-Pin	FBGA 144-Pin			
A54SX08	69	81	130	113	128	_	_	111			
A54SX16	_	81	175	-	147	_	_	_			
A54SX16P	_	81	175	113	147	_	_	_			
A54SX32	_	-	174	113	147	249	249	_			

Note: Package Definitions (Consult your local Actel sales representative for product availability):

PLCC = Plastic Leaded Chip Carrier

PQFP = Plastic Quad Flat Pack

TQFP = Thin Quad Flat Pack

VQFP = Very Thin Quad Flat Pack

PBGA = Plastic Ball Grid Array

FBGA = Fine Pitch (1.0 mm) Ball Grid Array

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General Description

The Actel SX family of FPGAs features a sea-of-modules architecture that delivers device performance and integration levels not currently achieved by any other FPGA architecture. SX devices greatly simplify design time, enable dramatic reductions in design costs and power consumption, and further decrease time to market for performance-intensive applications.

The Actel SX architecture features two types of logic modules, the combinatorial cell (C-cell) and the register cell (R-cell), each optimized for fast and efficient mapping of synthesized logic functions. The routing and interconnect resources are in the metal layers above the logic modules, providing optimal use of silicon. This enables the entire floor of the device to be spanned with an uninterrupted grid of fine-grained, synthesis-friendly logic modules (or "sea-of-modules"), which reduces the distance signals have to travel between logic modules. To minimize signal propagation delay, SX devices employ both local and general routing resources. The high-speed local routing resources (DirectConnect and FastConnect) enable very fast local signal propagation that is optimal for fast counters, state machines, and datapath logic. The general system of segmented routing tracks allows any logic module in the array to be connected to any other logic or I/O module. Within this system, propagation delay is minimized by limiting the number of antifuse interconnect elements to five (90 percent of connections typically use only three antifuses). The unique local and general routing structure featured in SX devices gives fast and predictable performance, allows 100 percent pin-locking with full logic utilization, enables concurrent PCB development, reduces design time, and allows designers to achieve performance goals with minimum effort.

Further complementing SX's flexible routing structure is a hardwired, constantly loaded clock network that has been tuned to provide fast clock propagation with minimal clock skew. Additionally, the high performance of the internal logic has eliminated the need to embed latches or flip-flops in the I/O cells to achieve fast clock-to-out or fast input setup times. SX devices have easy to use I/O cells that do not require HDL instantiation, facilitating design reuse and reducing design and verification time.

SX Family Architecture

The SX family architecture was designed to satisfy nextgeneration performance and integration requirements for production-volume designs in a broad range of applications.

Programmable Interconnect Element

The SX family provides efficient use of silicon by locating the routing interconnect resources between the Metal 2 (M2) and Metal 3 (M3) layers (Figure 1-1 on page 1-2). This completely eliminates the channels of routing and interconnect resources between logic modules (as implemented on SRAM FPGAs and previous generations of antifuse FPGAs), and enables the entire floor of the device to be spanned with an uninterrupted grid of logic modules.

Interconnection between these logic modules is achieved using The Actel patented metal-to-metal programmable antifuse interconnect elements, which are embedded between the M2 and M3 layers. The antifuses are normally open circuit and, when programmed, form a permanent low-impedance connection.

The extremely small size of these interconnect elements gives the SX family abundant routing resources and provides excellent protection against design pirating. Reverse engineering is virtually impossible because it is extremely difficult to distinguish between programmed and unprogrammed antifuses, and there is no configuration bitstream to intercept.

Additionally, the interconnect elements (i.e., the antifuses and metal tracks) have lower capacitance and lower resistance than any other device of similar capacity, leading to the fastest signal propagation in the industry.

Logic Module Design

The SX family architecture is described as a "sea-of-modules" architecture because the entire floor of the device is covered with a grid of logic modules with virtually no chip area lost to interconnect elements or routing. The Actel SX family provides two types of logic modules, the register cell (R-cell) and the combinatorial cell (C-cell).



Chip Architecture

The SX family chip architecture provides a unique approach to module organization and chip routing that delivers the best register/logic mix for a wide variety of new and emerging applications.

Module Organization

Actel has arranged all C-cell and R-cell logic modules into horizontal banks called *clusters*. There are two types of *clusters*: Type 1 contains two C-cells and one R-cell, while Type 2 contains one C-cell and two R-cells.

To increase design efficiency and device performance, Actel has further organized these modules into *SuperClusters* (Figure 1-4). SuperCluster 1 is a two-wide grouping of Type 1 clusters. SuperCluster 2 is a two-wide group containing one Type 1 cluster and one Type 2 cluster. SX devices feature more SuperCluster 1 modules than SuperCluster 2 modules because designers typically require significantly more combinatorial logic than flip-flops.

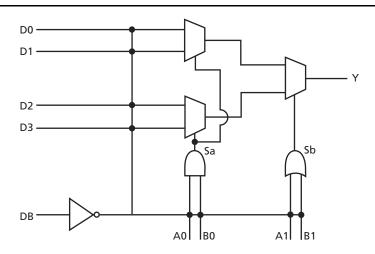


Figure 1-3 • C-Cell

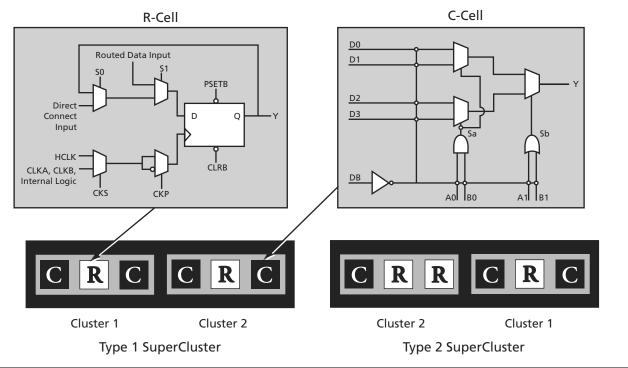


Figure 1-4 • Cluster Organization

DirectConnect is a horizontal routing resource that provides connections from a C-cell to its neighboring R-cell in a given SuperCluster. DirectConnect uses a hardwired signal path requiring no programmable interconnection to achieve its fast signal propagation time of less than 0.1 ns.

FastConnect enables horizontal routing between any two logic modules within a given SuperCluster and vertical routing with the SuperCluster immediately below it. Only one programmable connection is used in a FastConnect path, delivering maximum pin-to-pin propagation of 0.4 ns.

In addition to DirectConnect and FastConnect, the architecture makes use of two globally oriented routing resources known as segmented routing and high-drive routing. The Actel segmented routing structure provides a variety of track lengths for extremely fast routing between SuperClusters. The exact combination of track lengths and antifuses within each path is chosen by the 100 percent automatic place-and-route software to minimize signal propagation delays.

The Actel high-drive routing structure provides three clock networks. The first clock, called HCLK, is hardwired from the HCLK buffer to the clock select multiplexer (MUX) in each R-cell. This provides a fast propagation path for the clock signal, enabling the 3.7 ns clock-to-out (pin-to-pin) performance of the SX devices. The hardwired clock is tuned to provide clock skew as low as 0.25 ns. The remaining two clocks (CLKA, CLKB) are global clocks that can be sourced from external pins or from internal logic signals within the SX device.

Other Architectural Features

Technology

The Actel SX family is implemented on a high-voltage twin-well CMOS process using 0.35 μ design rules. The metal-to-metal antifuse is made up of a combination of amorphous silicon and dielectric material with barrier metals and has a programmed ("on" state) resistance of 25 Ω with a capacitance of 1.0 fF for low signal impedance.

Performance

The combination of architectural features described above enables SX devices to operate with internal clock frequencies exceeding 300 MHz, enabling very fast execution of even complex logic functions. Thus, the SX family is an optimal platform upon which to integrate the functionality previously contained in multiple CPLDs. In addition, designs that previously would have required a gate array to meet performance goals can now be integrated into an SX device with dramatic improvements in cost and time to market. Using timingdriven place-and-route tools, designers can achieve highly deterministic device performance. With SX devices, designers do not need to use complicated performance-enhancing design techniques such as the use of redundant logic to reduce fanout on critical nets or the instantiation of macros in HDL code to achieve high performance.

I/O Modules

Each I/O on an SX device can be configured as an input, an output, a tristate output, or a bidirectional pin.

Even without the inclusion of dedicated I/O registers, these I/Os, in combination with array registers, can achieve clock-to-out (pad-to-pad) timing as fast as 3.7 ns. I/O cells that have embedded latches and flip-flops require instantiation in HDL code; this is a design complication not encountered in SX FPGAs. Fast pin-to-pin timing ensures that the device will have little trouble interfacing with any other device in the system, which in turn enables parallel design of system components and reduces overall design time.

Power Requirements

The SX family supports 3.3 V operation and is designed to tolerate 5.0 V inputs. (Table 1-1). Power consumption is extremely low due to the very short distances signals are required to travel to complete a circuit. Power requirements are further reduced because of the small number of low-resistance antifuses in the path. The antifuse architecture does not require active circuitry to hold a charge (as do SRAM or EPROM), making it the lowest power architecture on the market.

Table 1-1 • Supply Voltages

Device	V _{CCA}	V _{CCI}	V _{CCR}	Maximum Input Tolerance	Maximum Output Drive
A54SX08 A54SX16 A54SX32	3.3 V	3.3 V	5.0 V	5.0 V	3.3 V
A54SX16-P*	3.3 V	3.3 V	3.3 V	3.3 V	3.3 V
	3.3 V	3.3 V	5.0 V	5.0 V	3.3 V
	3.3 V	5.0 V	5.0 V	5.0 V	5.0 V

Note: *A54SX16-P has three different entries because it is capable of both a 3.3 V and a 5.0 V drive.

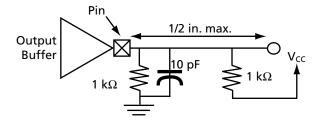
A54SX16P AC Specifications for (PCI Operation)

Table 1-7 • A54SX16P AC Specifications for (PCI Operation)

Symbol	Parameter	Condition	Min.	Max.	Units
I _{OH(AC)}	Switching Current High	$0 < V_{OUT} \le 1.4^{1}$	-44		mA
		$1.4 \le V_{OUT} < 2.4^{1, 2}$	-44 + (V _{OUT} - 1.4)/0.024		mA
		$3.1 < V_{OUT} < V_{CC}^{1, 3}$		EQ 1-1 on page 1-11	
	(Test Point)	$V_{OUT} = 3.1^3$		-142	mA
I _{OL(AC)}	Switching Current High	$V_{OUT} \ge 2.2^1$	95		mA
		$2.2 > V_{OUT} > 0.55^{1}$	V _{OUT} /0.023		
		$0.71 > V_{OUT} > 0^{1, 3}$		EQ 1-2 on page 1-11	mA
	(Test Point)	$V_{OUT} = 0.71^3$		206	mA
I _{CL}	Low Clamp Current	$-5 < V_{IN} \le -1$	-25 + (V _{IN} + 1)/0.015		mA
slew _R	Output Rise Slew Rate	0.4 V to 2.4 V load ⁴	1	5	V/ns
slew _F	Output Fall Slew Rate	2.4 V to 0.4 V load ⁴	1	5	V/ns

Notes:

- 1. Refer to the V/I curves in Figure 1-9 on page 1-11. Switching current characteristics for REQ# and GNT# are permitted to be one half of that specified here; i.e., half-size output drivers may be used on these signals. This specification does not apply to CLK and RST#, which are system outputs. "Switching Current High" specifications are not relevant to SERR#, INTA#, INTB#, INTC#, and INTD#, which are open drain outputs.
- 2. Note that this segment of the minimum current curve is drawn from the AC drive point directly to the DC drive point rather than toward the voltage rail (as is done in the pull-down curve). This difference is intended to allow for an optional N-channel pull-up.
- 3. Maximum current requirements must be met as drivers pull beyond the last step voltage. Equations defining these maximums (A and B) are provided with the respective diagrams in Figure 1-9 on page 1-11. The equation defined maxima should be met by design. In order to facilitate component testing, a maximum current test point is defined for each side of the output driver.
- 4. This parameter is to be interpreted as the cumulative edge rate across the specified range, rather than the instantaneous rate at any point within the transition range. The specified load (diagram below) is optional; i.e., the designer may elect to meet this parameter with an unloaded output per revision 2.0 of the PCI Local Bus Specification. However, adherence to both maximum and minimum parameters is now required (the maximum is no longer simply a guideline). Since adherence to the maximum slew rate was not required prior to revision 2.1 of the specification, there may be components in the market for some time that have faster edge rates; therefore, motherboard designers must bear in mind that rise and fall times faster than this specification could occur, and should ensure that signal integrity modeling accounts for this. Rise slew rate does not apply to open drain outputs.



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EQ 1-2

Figure 1-9 shows the 5.0 V PCI V/I curve and the minimum and maximum PCI drive characteristics of the A54SX16P device.

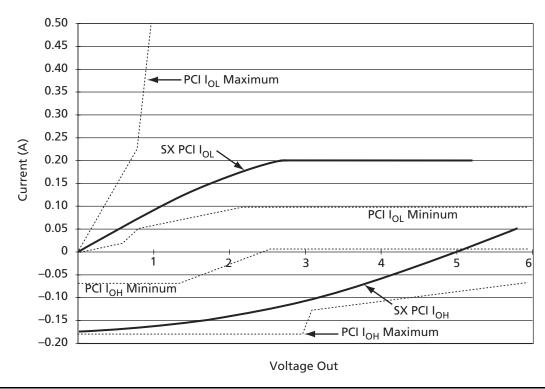


Figure 1-9 • 5.0 V PCI Curve for A54SX16P Device

$$I_{OH} = 11.9 \times (V_{OUT} - 5.25) \times (V_{OUT} + 2.45)$$

$$I_{OL} = 78.5 \times V_{OUT} \times (4.4 - V_{OUT})$$
for $V_{CC} > V_{OUT} > 3.1 \text{ V}$

$$EQ 1-1$$

Register Cell Timing Characteristics

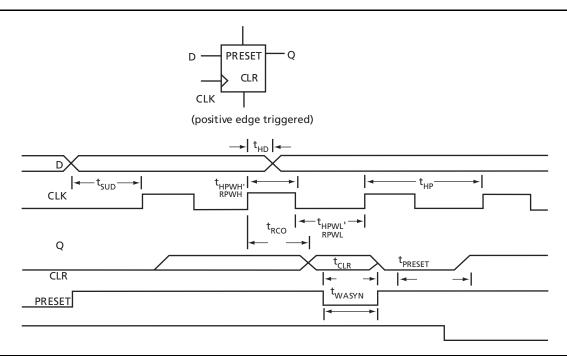


Figure 1-17 • Flip-Flops

Timing Characteristics

Timing characteristics for SX devices fall into three categories: family-dependent, device-dependent, and design-dependent. The input and output buffer characteristics are common to all SX family members. Internal routing delays are device-dependent. Design dependency means actual delays are not determined until after placement and routing of the user's design is complete. Delay values may then be determined by using the DirectTime Analyzer utility or performing simulation with post-layout delays.

Critical Nets and Typical Nets

Propagation delays are expressed only for typical nets, which are used for initial design performance evaluation. Critical net delays can then be applied to the most time-critical paths. Critical nets are determined by net property assignment prior to placement and routing. Up to 6% of the nets in a design may be designated as critical, while 90% of the nets in a design are typical.

Long Tracks

Some nets in the design use long tracks. Long tracks are special routing resources that span multiple rows, columns, or modules. Long tracks employ three and sometimes five antifuse connections. This increases capacitance and resistance, resulting in longer net delays for macros connected to long tracks. Typically up to 6 percent of nets in a fully utilized device require long tracks. Long tracks contribute approximately 4 ns to 8.4 ns delay. This additional delay is represented statistically in higher fanout (FO = 24) routing delays in the datasheet specifications section.

Timing Derating

SX devices are manufactured in a CMOS process. Therefore, device performance varies according to temperature, voltage, and process variations. Minimum timing parameters reflect maximum operating voltage, minimum operating temperature, and best-case processing. Maximum timing parameters reflect minimum operating voltage, maximum operating temperature, and worst-case processing.

Table 1-17 • A54SX08 Timing Characteristics (Continued) (Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA}, V_{CCI} = 3.0 V, T_J = 70°C)

		'-3' 9	Speed	'-2' \$	Speed	'-1' 9	Speed	'Std' Speed		
Parameter	Description	Min.	Мах.	Min.	Мах.	Min.	Мах.	Min.	Мах.	Units
Dedicated (Hardwired) Array Clock Network									
t _{HCKH}	Input LOW to HIGH (pad to R-Cell input)		1.0		1.1		1.3		1.5	ns
t_{HCKL}	Input HIGH to LOW (pad to R-Cell input)		1.0		1.2		1.4		1.6	ns
t_{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t_{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t _{HCKSW}	Maximum Skew		0.1		0.2		0.2		0.2	ns
t _{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f_{HMAX}	Maximum Frequency		350		320		280		240	MHz
Routed Arra	ay Clock Networks									
t _{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)		1.3		1.5		1.7		2.0	ns
t _{RCKL}	Input HIGH to LOW (light load) (pad to R-Cell Input)		1.4		1.6		1.8		2.1	ns
t _{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)		1.4		1.7		1.9		2.2	ns
t _{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)		1.5		1.7		2.0		2.3	ns
t _{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)		1.5		1.7		1.9		2.2	ns
t_{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)		1.5		1.8		2.0		2.3	ns
t _{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t _{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t _{RCKSW}	Maximum Skew (light load)		0.1		0.2		0.2		0.2	ns
t _{RCKSW}	Maximum Skew (50% load)		0.3		0.3		0.4		0.4	ns
t _{RCKSW}	Maximum Skew (100% load)		0.3		0.3		0.4		0.4	ns
TTL Output	Module Timing1									
t _{DLH}	Data-to-Pad LOW to HIGH		1.6		1.9		2.1		2.5	ns
t_{DHL}	Data-to-Pad HIGH to LOW		1.6		1.9		2.1		2.5	ns
t _{ENZL}	Enable-to-Pad, Z to L		2.1		2.4		2.8		3.2	ns
t _{ENZH}	Enable-to-Pad, Z to H		2.3		2.7		3.1		3.6	ns
t _{ENLZ}	Enable-to-Pad, L to Z		1.4		1.7		1.9		2.2	ns

Note:

- 1. For dual-module macros, use $t_{PD}+t_{RD1}+t_{PDn}$, $t_{RCO}+t_{RD1}+t_{PDn}$, or $t_{PD1}+t_{RD1}+t_{SUD}$, whichever is appropriate.
- 2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

A54SX16 Timing Characteristics

Table 1-18 • A54SX16 Timing Characteristics (Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA}, V_{CCI} = 3.0 V, T_J = 70°C)

	(Norse case commercial conditions, t		Speed		Speed	'-1' \$	Speed	'Std'	Speed	
Parameter	Description	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Units
C-Cell Propa	agation Delays ¹									
t _{PD}	Internal Array Module		0.6		0.7		8.0		0.9	ns
Predicted R	outing Delays ²									
t _{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t _{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t _{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t _{RD2}	FO = 2 Routing Delay		0.6		0.7		8.0		0.9	ns
t _{RD3}	FO = 3 Routing Delay		8.0		0.9		1.0		1.2	ns
t _{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t _{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t _{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timir	ıg									
t _{RCO}	Sequential Clock-to-Q		0.8		1.1		1.2		1.4	ns
t _{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		8.0	ns
t _{PRESET}	Asynchronous Preset-to-Q		0.7		8.0		0.9		1.0	ns
t _{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		8.0		ns
t _{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t _{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Modu	ile Propagation Delays									
t _{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t _{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted In	nput Routing Delays ²									
t _{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t _{IRD2}	FO = 2 Routing Delay		0.6		0.7		8.0		0.9	ns
t _{IRD3}	FO = 3 Routing Delay		8.0		0.9		1.0		1.2	ns
t _{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t _{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t _{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Notes:

- 1. For dual-module macros, use $t_{PD}+t_{RD1}+t_{PDn},\ t_{RCO}+t_{RD1}+t_{PDn},\ or\ t_{PD1}+t_{RD1}+t_{SUD},\ whichever\ is\ appropriate.$
- 2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.
- 3. Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH} . For t_{ENZL} and t_{ENZH} , the loading is 5 pF.

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Table 1-18 • A54SX16 Timing Characteristics (Continued) (Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA}, V_{CCI} = 3.0 V, T_J = 70°C)

		'-3' 9	peed	'-2' 9	Speed	'-1' 9	peed	'Std'	Speed	
Parameter	Description	Min.	Max.	Min.	Мах.	Min.	Мах.	Min.	Мах.	Units
Dedicated (Hardwired) Array Clock Network									
t _{HCKH}	Input LOW to HIGH (pad to R-Cell input)		1.2		1.4		1.5		1.8	ns
t _{HCKL}	Input HIGH to LOW (pad to R-Cell input)		1.2		1.4		1.6		1.9	ns
t _{HPWH}	Minimum Pulse Width HIGH	1.4		1.6		1.8		2.1		ns
t _{HPWL}	Minimum Pulse Width LOW	1.4		1.6		1.8		2.1		ns
t _{HCKSW}	Maximum Skew		0.2		0.2		0.3		0.3	ns
t _{HP}	Minimum Period	2.7		3.1		3.6		4.2		ns
f _{HMAX}	Maximum Frequency		350		320		280		240	MHz
Routed Arra	ay Clock Networks									
t _{RCKH}	Input LOW to HIGH (light load) (pad to R-Cell input)		1.6		1.8		2.1		2.5	ns
t _{RCKL}	Input HIGH to LOW (light load) (pad to R-Cell input)		1.8		2.0		2.3		2.7	ns
t _{RCKH}	Input LOW to HIGH (50% load) (pad to R-Cell input)		1.8		2.1		2.5		2.8	ns
t _{RCKL}	Input HIGH to LOW (50% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t _{RCKH}	Input LOW to HIGH (100% load) (pad to R-Cell input)		1.8		2.1		2.4		2.8	ns
t _{RCKL}	Input HIGH to LOW (100% load) (pad to R-Cell input)		2.0		2.2		2.5		3.0	ns
t _{RPWH}	Min. Pulse Width HIGH	2.1		2.4		2.7		3.2		ns
t _{RPWL}	Min. Pulse Width LOW	2.1		2.4		2.7		3.2		ns
t _{RCKSW}	Maximum Skew (light load)		0.5		0.5		0.5		0.7	ns
t _{RCKSW}	Maximum Skew (50% load)		0.5		0.6		0.7		8.0	ns
t _{RCKSW}	Maximum Skew (100% load)		0.5		0.6		0.7		8.0	ns
TTL Output	Module Timing ³									
t _{DLH}	Data-to-Pad LOW to HIGH		1.6		1.9		2.1		2.5	ns
t _{DHL}	Data-to-Pad HIGH to LOW		1.6		1.9		2.1		2.5	ns
t _{ENZL}	Enable-to-Pad, Z to L		2.1		2.4		2.8		3.2	ns
t _{ENZH}	Enable-to-Pad, Z to H		2.3		2.7		3.1		3.6	ns
t _{ENLZ}	Enable-to-Pad, L to Z		1.4		1.7		1.9		2.2	ns
t _{ENHZ}	Enable-to-Pad, H to Z		1.3		1.5		1.7		2.0	ns

Notes:

- 1. For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn} , t_{RCO} + t_{RD1} + t_{PDn} , or t_{PD1} + t_{RD1} + t_{SUD} , whichever is appropriate.
- 2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.
- 3. Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH} . For t_{ENZL} and t_{ENZH} , the loading is 5 pF.

A54SX16P Timing Characteristics

Table 1-19 • A54SX16P Timing Characteristics (Worst-Case Commercial Conditions, V_{CCR} = 4.75 V, V_{CCA},V_{CCI} = 3.0 V, T_J = 70°C)

		'-3' \$	Speed	'-2' \$	Speed	'-1' \$	Speed	'Std'	Speed	
Parameter	Description	Min.	Мах.	Min.	Max.	Min.	Max.	Min.	Мах.	Units
C-Cell Propa	agation Delays ¹									
t _{PD}	Internal Array Module		0.6		0.7		8.0		0.9	ns
Predicted R	outing Delays ²									
t _{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t _{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t _{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t _{RD2}	FO = 2 Routing Delay		0.6		0.7		8.0		0.9	ns
t _{RD3}	FO = 3 Routing Delay		8.0		0.9		1.0		1.2	ns
t _{RD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t _{RD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t _{RD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns
R-Cell Timir	ng									
t _{RCO}	Sequential Clock-to-Q		0.9		1.1		1.3		1.4	ns
t _{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		0.8	ns
t _{PRESET}	Asynchronous Preset-to-Q		0.7		8.0		0.9		1.0	ns
t _{SUD}	Flip-Flop Data Input Set-Up	0.5		0.5		0.7		0.8		ns
t _{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t _{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Modu	ıle Propagation Delays									
t _{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t _{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted In	nput Routing Delays ²									
t _{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t _{IRD2}	FO = 2 Routing Delay		0.6		0.7		8.0		0.9	ns
t _{IRD3}	FO = 3 Routing Delay		8.0		0.9		1.0		1.2	ns
t _{IRD4}	FO = 4 Routing Delay		1.0		1.2		1.4		1.6	ns
t _{IRD8}	FO = 8 Routing Delay		1.9		2.2		2.5		2.9	ns
t _{IRD12}	FO = 12 Routing Delay		2.8		3.2		3.7		4.3	ns

Note:

- 1. For dual-module macros, use t_{PD} + t_{RD1} + t_{PDn} , t_{RCO} + t_{RD1} + t_{PDn} , or t_{PD1} + t_{RD1} + t_{SUD} , whichever is appropriate.
- 2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.

3. Delays based on 10 pF loading.

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A54SX32 Timing Characteristics

Table 1-20 • A54SX32 Timing Characteristics (Worst-Case Commercial Conditions, V_{CCR}= 4.75 V, V_{CCA}, V_{CCI} = 3.0 V, T_J = 70°C)

		'-3' \$	Speed	'-2' 9	Speed	'-1' 9	Speed	'Std'	Speed	
Parameter	Description	Min.	Мах.	Min.	Мах.	Min.	Мах.	Min.	Мах.	Units
C-Cell Propa	agation Delays ¹									
t _{PD}	Internal Array Module		0.6		0.7		8.0		0.9	ns
Predicted R	outing Delays ²									
t _{DC}	FO = 1 Routing Delay, Direct Connect		0.1		0.1		0.1		0.1	ns
t _{FC}	FO = 1 Routing Delay, Fast Connect		0.3		0.4		0.4		0.5	ns
t _{RD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t _{RD2}	FO = 2 Routing Delay		0.7		8.0		0.9		1.0	ns
t _{RD3}	FO = 3 Routing Delay		1.0		1.2		1.4		1.6	ns
t _{RD4}	FO = 4 Routing Delay		1.4		1.6		1.8		2.1	ns
t _{RD8}	FO = 8 Routing Delay		2.7		3.1		3.5		4.1	ns
t _{RD12}	FO = 12 Routing Delay		4.0		4.7		5.3		6.2	ns
R-Cell Timir	ng									
t _{RCO}	Sequential Clock-to-Q		0.8		1.1		1.3		1.4	ns
t _{CLR}	Asynchronous Clear-to-Q		0.5		0.6		0.7		8.0	ns
t _{PRESET}	Asynchronous Preset-to-Q		0.7		8.0		0.9		1.0	ns
t _{SUD}	Flip-Flop Data Input Set-Up	0.5		0.6		0.7		0.8		ns
t _{HD}	Flip-Flop Data Input Hold	0.0		0.0		0.0		0.0		ns
t _{WASYN}	Asynchronous Pulse Width	1.4		1.6		1.8		2.1		ns
Input Modu	ıle Propagation Delays									
t _{INYH}	Input Data Pad-to-Y HIGH		1.5		1.7		1.9		2.2	ns
t _{INYL}	Input Data Pad-to-Y LOW		1.5		1.7		1.9		2.2	ns
Predicted In	nput Routing Delays ²									
t _{IRD1}	FO = 1 Routing Delay		0.3		0.4		0.4		0.5	ns
t _{IRD2}	FO = 2 Routing Delay		0.7		8.0		0.9		1.0	ns
t _{IRD3}	FO = 3 Routing Delay		1.0		1.2		1.4		1.6	ns
t _{IRD4}	FO = 4 Routing Delay		1.4		1.6		1.8		2.1	ns
t _{IRD8}	FO = 8 Routing Delay		2.7		3.1		3.5		4.1	ns
t _{IRD12}	FO = 12 Routing Delay		4.0		4.7		5.3		6.2	ns

Note:

- 1. For dual-module macros, use t_{PD} + t_{RD1} + $t_{PDn'}$ t_{RCO} + t_{RD1} + t_{PDn} , or t_{PD1} + t_{RD1} + t_{SUD} , whichever is appropriate.
- 2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual worst-case performance. Post-route timing is based on actual routing delay measurements performed on the device prior to shipment.
- 3. Delays based on 35 pF loading, except t_{ENZL} and t_{ENZH} . For t_{ENZL} and t_{ENZH} the loading is 5 pF.



Pin Description

CLKA/B Clock A and B

These pins are 3.3 V / 5.0 V PCI/TTL clock inputs for clock distribution networks. The clock input is buffered prior to clocking the R-cells. If not used, this pin must be set LOW or HIGH on the board. It must not be left floating. (For A54SX72A, these clocks can be configured as bidirectional.)

GND Ground

LOW supply voltage.

HCLK Dedicated (hardwired) Array Clock

This pin is the 3.3 V / 5.0 V PCI/TTL clock input for sequential modules. This input is directly wired to each R-cell and offers clock speeds independent of the number of R-cells being driven. If not used, this pin must be set LOW or HIGH on the board. It must not be left floating.

I/O Input/Output

The I/O pin functions as an input, output, tristate, or bidirectional buffer. Based on certain configurations, input and output levels are compatible with standard TTL, LVTTL, 3.3 V PCI or 5.0 V PCI specifications. Unused I/O pins are automatically tristated by the Designer Series software.

NC No Connection

This pin is not connected to circuitry within the device.

PRA, I/O Probe A

The Probe A pin is used to output data from any userdefined design node within the device. This independent diagnostic pin can be used in conjunction with the Probe B pin to allow real-time diagnostic output of any signal path within the device. The Probe A pin can be used as a user-defined I/O when verification has been completed. The pin's probe capabilities can be permanently disabled to protect programmed design confidentiality.

PRB. I/O Probe B

The Probe B pin is used to output data from any node within the device. This diagnostic pin can be used in conjunction with the Probe A pin to allow real-time diagnostic output of any signal path within the device. The Probe B pin can be used as a user-defined I/O when verification has been completed. The pin's probe capabilities can be permanently disabled to protect programmed design confidentiality.

TCK Test Clock

Test clock input for diagnostic probe and device programming. In flexible mode, TCK becomes active when the TMS pin is set LOW (refer to Table 1-2 on page 1-6). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDI Test Data Input

Serial input for boundary scan testing and diagnostic probe. In flexible mode, TDI is active when the TMS pin is set LOW (refer to Table 1-2 on page 1-6). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDO Test Data Output

Serial output for boundary scan testing. In flexible mode, TDO is active when the TMS pin is set LOW (refer to Table 1-2 on page 1-6). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TMS Test Mode Select

The TMS pin controls the use of the IEEE 1149.1 Boundary Scan pins (TCK, TDI, TDO). In flexible mode when the TMS pin is set LOW, the TCK, TDI, and TDO pins are boundary scan pins (refer to Table 1-2 on page 1-6). Once the boundary scan pins are in test mode, they will remain in that mode until the internal boundary scan state machine reaches the "logic reset" state. At this point, the boundary scan pins will be released and will function as regular I/O pins. The "logic reset" state is reached 5 TCK cycles after the TMS pin is set HIGH. In dedicated test mode, TMS functions as specified in the IEEE 1149.1 specifications.

V_{CCI} Supply Voltage

Supply voltage for I/Os. See Table 1-1 on page 1-5.

V_{CCA} Supply Voltage

Supply voltage for Array. See Table 1-1 on page 1-5.

V_{CCR} Supply Voltage

Supply voltage for input tolerance (required for internal biasing). See Table 1-1 on page 1-5.

	208-Pin PQFP								
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function						
145	V_{CCA}	V_{CCA}	V_{CCA}						
146	GND	GND	GND						
147	I/O	I/O	I/O						
148	V _{CCI}	V _{CCI}	V _{CCI}						
149	I/O	I/O	1/0						
150	I/O	I/O	I/O						
151	I/O	I/O	1/0						
152	I/O	I/O	1/0						
153	I/O	I/O	1/0						
154	I/O	I/O	1/0						
155	NC	I/O	I/O						
156	NC	I/O	I/O						
157	GND	GND	GND						
158	I/O	I/O	I/O						
159	I/O	1/0	I/O						
160	I/O	I/O	I/O						
161	I/O	I/O	I/O						
162	I/O	I/O	I/O						
163	I/O	I/O	I/O						
164	V _{CCI}	V _{CCI}	V _{CCI}						
165	I/O	1/0	I/O						
166	I/O	I/O	I/O						
167	NC	I/O	I/O						
168	I/O	I/O	I/O						
169	I/O	I/O	I/O						
170	NC	I/O	I/O						
171	I/O	I/O	I/O						
172	I/O	I/O	I/O						
173	NC	I/O	I/O						
174	I/O	I/O	I/O						
175	I/O	I/O	I/O						
176	NC	I/O	I/O						
177	I/O	I/O	I/O						
178	I/O	1/0	I/O						
179	I/O	1/0	I/O						
180	CLKA	CLKA	CLKA						

	208-Pi	n PQFP	
Pin Number	A54SX08 Function	A54SX16, A54SX16P Function	A54SX32 Function
181	CLKB	CLKB	CLKB
182	V_{CCR}	V_{CCR}	V_{CCR}
183	GND	GND	GND
184	V_{CCA}	V _{CCA}	V_{CCA}
185	GND	GND	GND
186	PRA, I/O	PRA, I/O	PRA, I/O
187	I/O	1/0	1/0
188	I/O	1/0	1/0
189	NC	I/O	I/O
190	I/O	I/O	I/O
191	I/O	I/O	I/O
192	NC	I/O	I/O
193	I/O	1/0	1/0
194	I/O	I/O	I/O
195	NC	I/O	I/O
196	I/O	I/O	I/O
197	I/O	1/0	I/O
198	NC	I/O	I/O
199	I/O	I/O	I/O
200	I/O	I/O	I/O
201	V _{CCI}	V _{CCI}	V _{CCI}
202	NC	I/O	I/O
203	NC	1/0	I/O
204	I/O	I/O	I/O
205	NC	1/0	I/O
206	I/O	1/0	I/O
207	I/O	1/0	I/O
208	TCK, I/O	TCK, I/O	TCK, I/O

Note: * Note that Pin 65 in the A54SX32—PQ208 is a no connect (NC).

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	144-Pi	n TQFP	
Pin Number	A54SX08 Function	A54SX16P Function	A54SX32 Function
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	I/O	1/0	I/O
4	I/O	1/0	I/O
5	I/O	1/0	I/O
6	I/O	1/0	1/0
7	I/O	1/0	I/O
8	I/O	I/O	1/0
9	TMS	TMS	TMS
10	V _{CCI}	V_{CCI}	V _{CCI}
11	GND	GND	GND
12	I/O	I/O	1/0
13	I/O	1/0	I/O
14	I/O	I/O	1/0
15	I/O	I/O	1/0
16	I/O	I/O	I/O
17	I/O	1/0	1/0
18	I/O	I/O	1/0
19	V_{CCR}	V_{CCR}	V_{CCR}
20	V_{CCA}	V_{CCA}	V_{CCA}
21	I/O	1/0	I/O
22	I/O	1/0	I/O
23	I/O	1/0	I/O
24	I/O	1/0	I/O
25	I/O	1/0	I/O
26	I/O	1/0	I/O
27	I/O	1/0	I/O
28	GND	GND	GND
29	V _{CCI}	V _{CCI}	V _{CCI}
30	V_{CCA}	V _{CCA}	V _{CCA}
31	I/O	1/0	I/O
32	I/O	1/0	I/O
33	I/O	I/O	1/0
34	I/O	I/O	1/0
35	I/O	I/O	I/O
36	GND	GND	GND

144-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16P Function	A54SX32 Function
37	I/O	1/0	I/O
38	I/O	1/0	I/O
39	I/O	1/0	I/O
40	I/O	1/0	I/O
41	I/O	1/0	I/O
42	I/O	1/0	I/O
43	I/O	1/0	I/O
44	V _{CCI}	V _{CCI}	V _{CCI}
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	I/O	I/O	I/O
49	I/O	I/O	I/O
50	I/O	1/0	I/O
51	I/O	1/0	I/O
52	I/O	1/0	I/O
53	I/O	1/0	I/O
54	PRB, I/O	PRB, I/O	PRB, I/O
55	I/O	I/O	I/O
56	V_{CCA}	V_{CCA}	V_{CCA}
57	GND	GND	GND
58	V_{CCR}	V_{CCR}	V_{CCR}
59	I/O	1/0	I/O
60	HCLK	HCLK	HCLK
61	I/O	I/O	I/O
62	I/O	1/0	I/O
63	I/O	1/0	I/O
64	I/O	1/0	I/O
65	I/O	I/O	I/O
66	I/O	I/O	I/O
67	I/O	I/O	I/O
68	V _{CCI}	V _{CCI}	V _{CCI}
69	I/O	I/O	I/O
70	I/O	1/0	I/O
71	TDO, I/O	TDO, I/O	TDO, I/O
72	I/O	I/O	I/O
		-	

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144-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16P Function	A54SX32 Function
73	GND	GND	GND
74	I/O	1/0	I/O
75	I/O	1/0	I/O
76	I/O	I/O	I/O
77	I/O	I/O	I/O
78	I/O	I/O	I/O
79	V_{CCA}	V_{CCA}	V_{CCA}
80	V _{CCI}	V _{CCI}	V _{CCI}
81	GND	GND	GND
82	I/O	I/O	I/O
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	I/O	I/O	I/O
86	I/O	1/0	I/O
87	I/O	1/0	I/O
88	I/O	1/0	I/O
89	V _{CCA}	V _{CCA}	V _{CCA}
90	V_{CCR}	V_{CCR}	V_{CCR}
91	I/O	1/0	I/O
92	I/O	1/0	I/O
93	I/O	1/0	I/O
94	I/O	1/0	I/O
95	I/O	1/0	I/O
96	I/O	1/0	I/O
97	I/O	I/O	I/O
98	V_{CCA}	V_{CCA}	V_{CCA}
99	GND	GND	GND
100	I/O	I/O	I/O
101	GND	GND	GND
102	V _{CCI}	V _{CCI}	V _{CCI}
103	I/O	I/O	I/O
104	I/O	I/O	I/O
105	I/O	1/0	I/O
106	I/O	1/0	I/O
107	I/O	1/0	I/O
108	I/O	1/0	I/O

144-Pin TQFP			
Pin Number	A54SX08 Function	A54SX16P Function	A54SX32 Function
109	GND	GND	GND
110	I/O	I/O	1/0
111	I/O	I/O	1/0
112	I/O	1/0	1/0
113	I/O	1/0	1/0
114	I/O	1/0	1/0
115	V _{CCI}	V _{CCI}	V _{CCI}
116	I/O	1/0	I/O
117	I/O	I/O	1/0
118	I/O	I/O	1/0
119	I/O	I/O	I/O
120	I/O	1/0	I/O
121	I/O	I/O	I/O
122	I/O	I/O	1/0
123	I/O	I/O	1/0
124	I/O	I/O	1/0
125	CLKA	CLKA	CLKA
126	CLKB	CLKB	CLKB
127	V_{CCR}	V_{CCR}	V_{CCR}
128	GND	GND	GND
129	V_{CCA}	V_{CCA}	V_{CCA}
130	I/O	I/O	I/O
131	PRA, I/O	PRA, I/O	PRA, I/O
132	I/O	1/0	1/0
133	I/O	1/0	1/0
134	I/O	I/O	1/0
135	I/O	I/O	1/0
136	I/O	1/0	1/0
137	I/O	1/0	I/O
138	I/O	1/0	1/0
139	I/O	1/0	I/O
140	V _{CCI}	V _{CCI}	V _{CCI}
141	I/O	I/O	I/O
142	I/O	I/O	I/O
143	I/O	1/0	I/O
144	TCK, I/O	TCK, I/O	TCK, I/O

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313-Pin PBGA

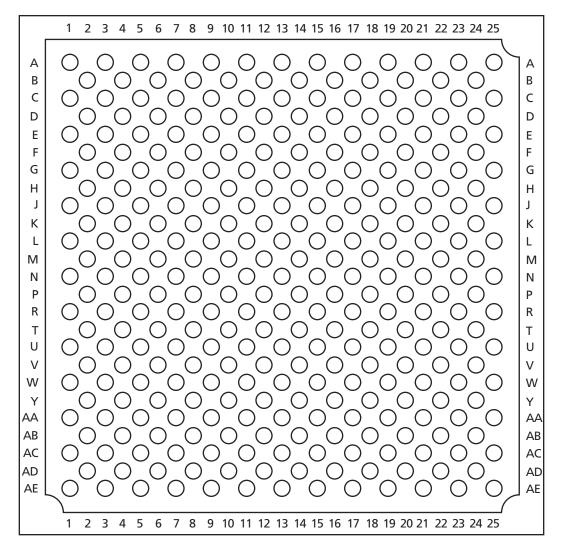


Figure 2-6 • 313-Pin PBGA (Top View)

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

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313-Pin PBGA		
Pin	A54SX32	
Number	Function	
H20	I/O	
H22	V_{CCI}	
H24	I/O	
J1	I/O	
J3	1/0	
J5	I/O	
J7	NC	
J9	I/O	
J11	1/0	
J13	CLKA	
J15	I/O	
J17	I/O	
J19	I/O	
J21	GND	
J23	I/O	
J25	I/O	
K2	I/O	
K4	I/O	
K6	I/O	
K8	V _{CCI}	
K10	I/O	
K12	I/O	
K14	I/O	
K16	I/O	
K18	I/O	
K20	V _{CCA}	
K22	I/O	
K24	I/O	
L1	I/O	
L3	I/O	
L5	I/O	
L7	I/O	
L9	I/O	
L11	I/O	
L13	GND	
L15	I/O	
L17	I/O	
L19	I/O	
L21	I/O	
L23	I/O	

_		
n PBGA		
A54SX32 Function		
I/O		
GND		
GND		
V _{CCI}		
I/O		
V_{CCA}		
V_{CCR}		
I/O		
V _{CCI}		
GND		
GND		
GND		
I/O		
I/O		
I/O		
V_{CCR}		
V_{CCA}		
I/O		
GND		
GND		
I/O		
I/O		
NC		
I/O		

313-Pin PBGA		
Pin Number	A54SX32 Function	
R5	I/O	
R7	I/O	
R9	1/0	
R11	1/0	
R13	GND	
R15	I/O	
R17	I/O	
R19	I/O	
R21	I/O	
R23	I/O	
R25	I/O	
T2	I/O	
T4	I/O	
T6	I/O	
Т8	1/0	
T10	I/O	
T12	I/O	
T14	HCLK	
T16	I/O	
T18	I/O	
T20	I/O	
T22	I/O	
T24	I/O	
U1	I/O	
U3	I/O	
U5	V _{CCI}	
U7	I/O	
U9	I/O	
U11	I/O	
U13	I/O	
U15	I/O	
U17	I/O	
U19	I/O	
U21	I/O	
U23	I/O	
U25	I/O	
V2	V_{CCA}	
V4	I/O	
V6	I/O	
V8	I/O	

313-Pin PBGA		
Pin	A54SX32	
Number	Function	
V10	I/O	
V12	I/O	
V14	I/O	
V16	NC	
V18	I/O	
V20	I/O	
V22	V_{CCA}	
V24	V _{CCI}	
W1	I/O	
W3	I/O	
W5	I/O	
W7	NC	
W9	I/O	
W11	I/O	
W13	V _{CCI}	
W15	I/O	
W17	I/O	
W19	I/O	
W21	I/O	
W23	I/O	
W25	I/O	
Y2	I/O	
Y4	I/O	
Y6	I/O	
Y8	I/O	
Y10	I/O	
Y12	I/O	
Y14	I/O	
Y16	1/0	
Y18	1/0	
Y20	NC	
Y22	I/O	
Y24	NC	

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329-Pin PBGA

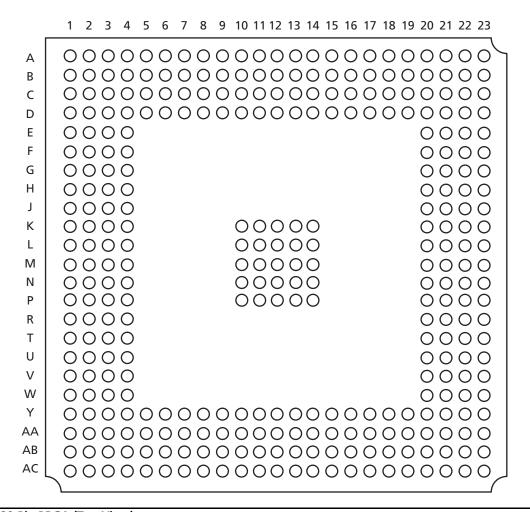


Figure 2-7 • 329-Pin PBGA (Top View)

Note

For Package Manufacturing and Environmental information, visit the Package Resource center at http://www.actel.com/products/rescenter/package/index.html.

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Datasheet Information

List of Changes

The following table lists critical changes that were made in the current version of the document.

Previous Version	Changes in Current Version (v3.2)	Page
v3.1	The "Ordering Information" was updated to include RoHS information.	1-ii
(June 2003)	The Product Plan was removed since all products have been released.	N/A
	Information concerning the TRST pin in the "Probe Circuit Control Pins" section was removed.	1-6
	The "Dedicated Test Mode" section is new.	1-6
	The "Programming" section is new.	1-7
	A note was added to the "Power-Up Sequencing" table.	1-15
	A note was added to the "Power-Down Sequencing" table. The 3.3 V comments were updated for the following devices: A54SX08, A54SX16, A54SX32.	1-15
	U11 and U13 were added to the "313-Pin PBGA" table.	2-17
v3.0.1	Storage temperature in Table 1-3 was updated.	1-7
	Table 1-1 was updated.	1-5

Datasheet Categories

In order to provide the latest information to designers, some datasheets are published before data has been fully characterized. Datasheets are designated as "Product Brief," "Advanced," "Production," and "Datasheet Supplement." The definitions of these categories are as follows:

Product Brief

The product brief is a summarized version of a datasheet (advanced or production) containing general product information. This brief gives an overview of specific device and family information.

Advanced

This datasheet version contains initial estimated information based on simulation, other products, devices, or speed grades. This information can be used as estimates, but not for production.

Unmarked (production)

This datasheet version contains information that is considered to be final.

Datasheet Supplement

The datasheet supplement gives specific device information for a derivative family that differs from the general family datasheet. The supplement is to be used in conjunction with the datasheet to obtain more detailed information and for specifications that do not differ between the two families.

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