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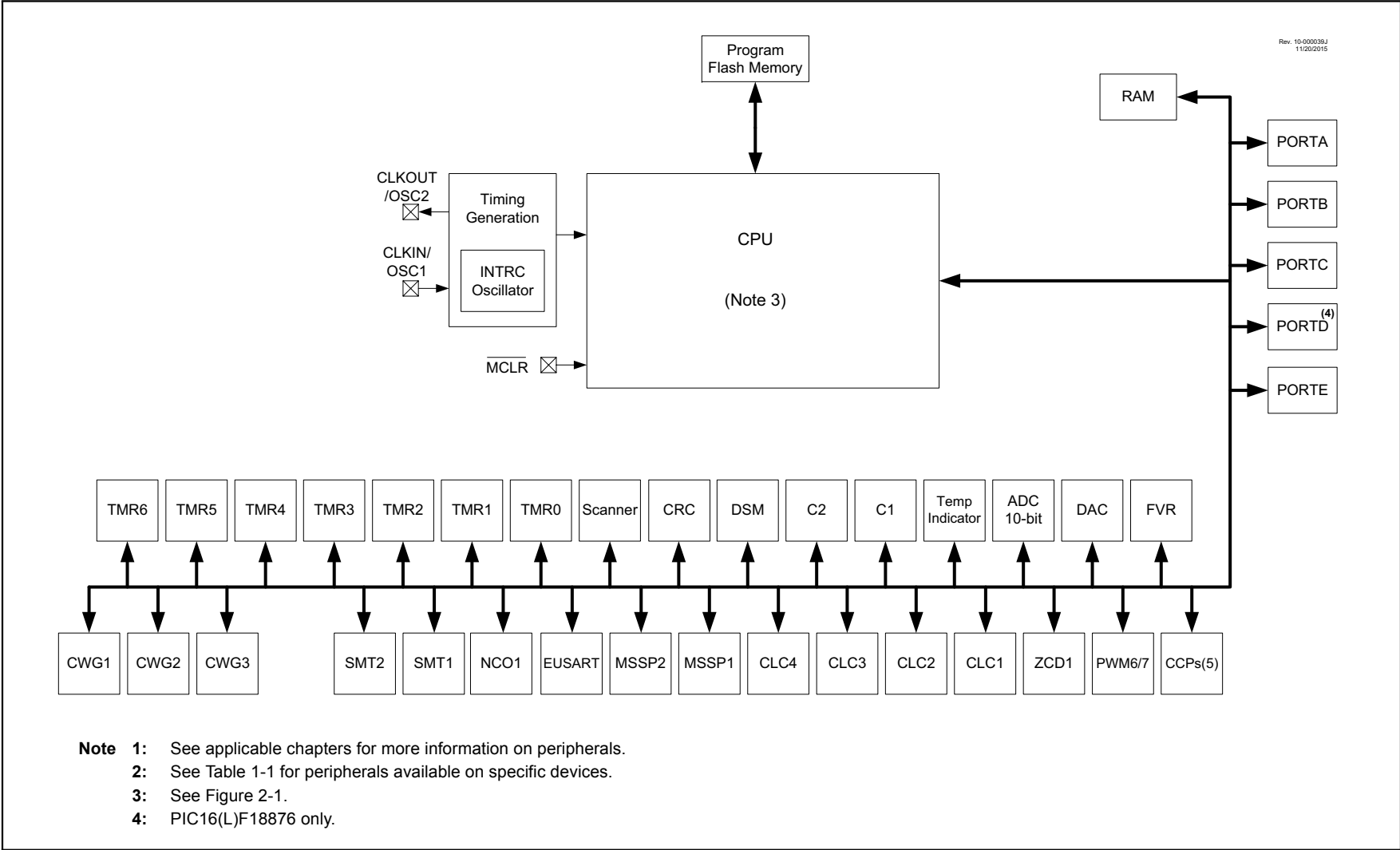
"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I ² C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	25
Program Memory Size	28KB (16K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 24x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-VQFN Exposed Pad
Supplier Device Package	28-QFN (6x6)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf18856t-i-ml

FIGURE 1-1: PIC16(L)F18856/76 BLOCK DIAGRAM



PIC16(L)F18856/76

2.0 ENHANCED MID-RANGE CPU

This family of devices contains an enhanced mid-range 8-bit CPU core. The CPU has 49 instructions. Interrupt capability includes automatic context saving. The hardware stack is 16-levels deep and has Overflow and Underflow Reset capability. Direct, Indirect, and

Relative Addressing modes are available. Two File Select Registers (FSRs) provide the ability to read program and data memory.

- Automatic Interrupt Context Saving
- 16-level Stack with Overflow and Underflow
- File Select Registers
- Instruction Set

FIGURE 2-1: CORE BLOCK DIAGRAM

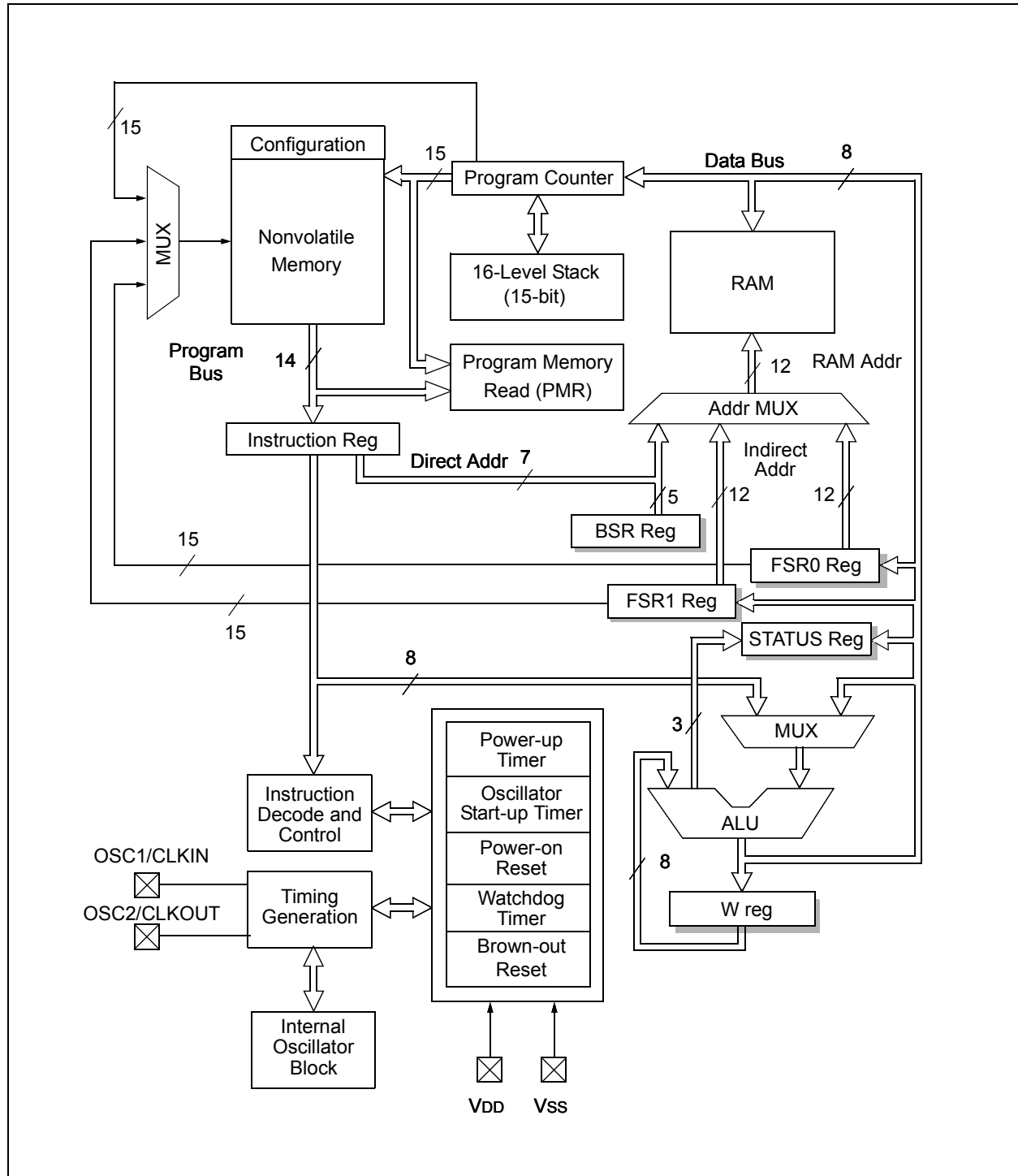


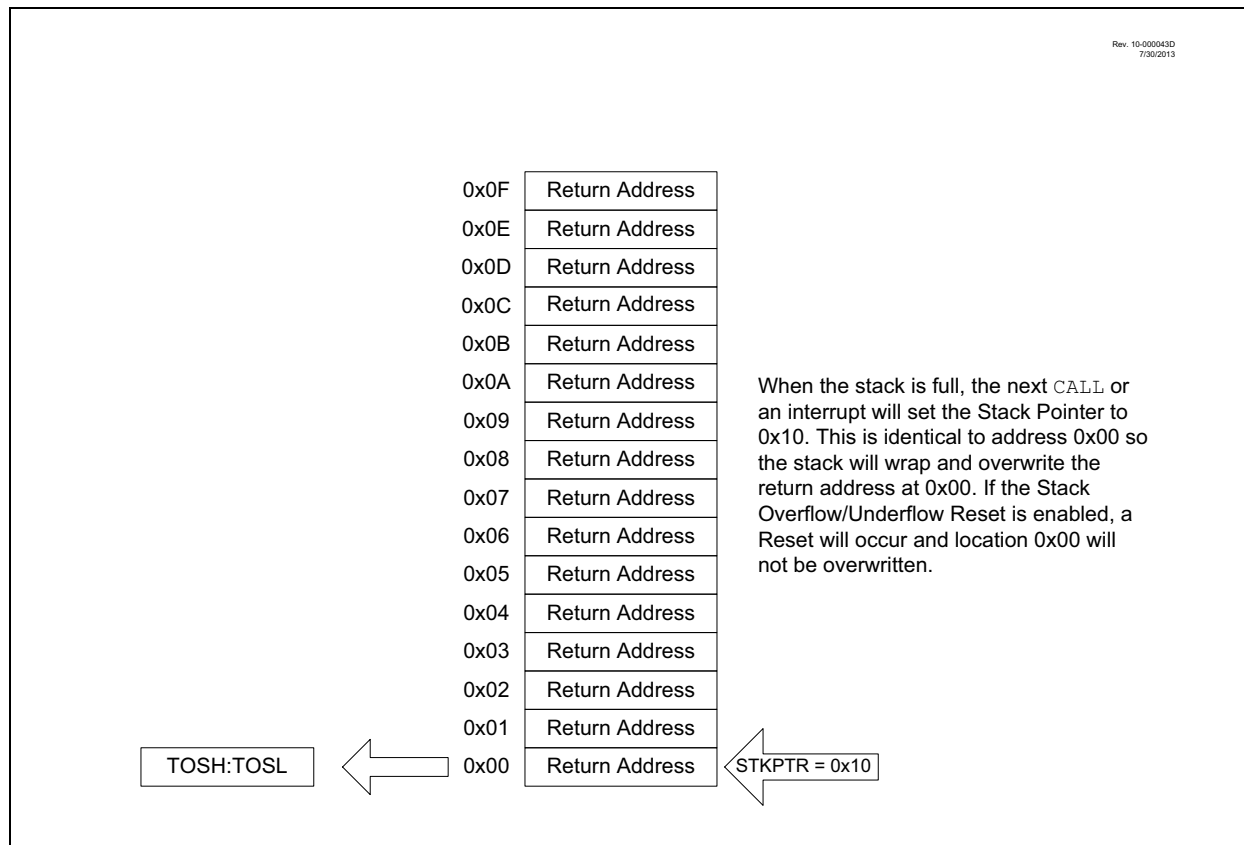
TABLE 3-5: PIC16F18856/76 MEMORY MAP BANK 8-15

BANK 8		BANK 9		BANK 10		BANK 11		BANK 12		BANK 13		BANK 14		BANK 15	
400h	Core Registers (Table 3-2)	480h	Core Registers (Table 3-2)	500h	Core Registers (Table 3-2)	580h	Core Registers (Table 3-2)	600h	Core Registers (Table 3-2)	680h	Core Registers (Table 3-2)	700h	Core Registers (Table 3-2)	780h	Core Registers (Table 3-2)
40Bh	SCANLADRL	48Bh	SMT1TMRL	50Bh	SMT2TMRL	58Bh	NCO1ACCL	60Bh	CWG1CLKCON	68Bh	CWG3CLKCON	70Bh	PIR0	78Bh	—
40Ch	SCANLADRH	48Ch	SMT1TMRH	50Ch	SMT2TMRH	58Ch	NCO1ACCH	60Ch	CWG1ISM	68Ch	CWG3ISM	70Ch	PIR1	78Ch	—
40Dh	SCANHADRL	48Dh	SMT1TMRU	50Dh	SMT2TMRU	58Dh	NCO1ACCU	60Dh	CWG1DBR	68Dh	CWG3DBR	70Dh	PIR2	78Dh	—
40Eh	SCANHADRH	48Eh	SMT1CPRL	50Eh	SMT2CPRL	58Eh	NCO1INCL	60Eh	CWG1DBF	68Eh	CWG3DBF	70Eh	PIR3	78Eh	—
40Fh	SCANHADRH	48Fh	SMT1CPRH	50Fh	SMT2CPRH	58Fh	NCO1INCH	60Fh	CWG1CON0	68Fh	CWG3CON0	70Fh	PIR4	78Fh	—
410h	SCANCON0	490h	SMT1CPRU	510h	SMT2CPRU	590h	NCO1INC	610h	CWG1CON1	690h	CWG3CON1	710h	PIR5	790h	—
411h	SCANTRIG	491h	SMT1CPWL	511h	SMT2CPWL	591h	NCO1CON	611h	CWG1AS0	691h	CWG3AS0	711h	PIR6	791h	—
412h	—	492h	SMT1CPWH	512h	SMT2CPWH	592h	NCO1CLK	612h	CWG1AS1	692h	CWG3AS1	712h	PIR7	792h	—
413h	—	493h	SMT1CPWU	513h	SMT2CPWU	593h	—	613h	CWG1STR	693h	CWG3STR	713h	PIR8	793h	—
414h	—	494h	SMT1PRL	514h	SMT2PRL	594h	—	614h	—	694h	—	714h	—	794h	—
415h	—	495h	SMT1PRH	515h	SMT2PRH	595h	—	615h	CWG2CLKCON	695h	—	715h	—	795h	—
416h	CRCDATL	496h	SMT1PRU	516h	SMT2PRU	596h	—	616h	CWG2ISM	696h	—	716h	PIE0	796h	PMD0
417h	CRCDATH	497h	SMT1CON0	517h	SMT2CON0	597h	—	617h	CWG2DBR	697h	—	717h	PIE1	797h	PMD1
418h	CRCACCL	498h	SMT1CON1	518h	SMT2CON1	598h	—	618h	CWG2DBF	698h	—	718h	PIE2	798h	PMD2
419h	CRCACCH	499h	SMT1STAT	519h	SMT2STAT	599h	—	619h	CWG2CON0	699h	—	719h	PIE3	799h	PMD3
41Ah	CRCSHIFTL	49Ah	SMT1CLK	51Ah	SMT2CLK	59Ah	—	61Ah	CWG2CON1	69Ah	—	71Ah	PIE4	79Ah	PMD4
41Bh	CRCSHIFTH	49Bh	SMT1SIG	51Bh	SMT2SIG	59Bh	—	61Bh	CWG2AS0	69Bh	—	71Bh	PIE5	79Bh	PMD5
41Ch	CRCXORL	49Ch	SMT1WIN	51Ch	SMT2WIN	59Ch	—	61Ch	CWG2AS1	69Ch	—	71Ch	PIE6	79Ch	—
41Dh	CRCXORH	49Dh	—	51Dh	—	59Dh	—	61Dh	CWG2STR	69Dh	—	71Dh	PIE7	79Dh	—
41Eh	CRCCON0	49Eh	—	51Eh	—	59Eh	—	61Eh	—	69Eh	—	71Eh	PIE8	79Eh	—
41Fh	CRCCON1	49Fh	—	51Fh	—	59Fh	—	61Fh	—	69Fh	—	71Fh	—	79Fh	—
420h	General Purpose Register 80 Bytes	4A0h	General Purpose Register 80 Bytes	520h	General Purpose Register 80 Bytes	5A0h	General Purpose Register 80 Bytes	620h	General Purpose Register 80 Bytes	6A0h	General Purpose Register 80 Bytes	720h	General Purpose Register 80 Bytes	7A0h	General Purpose Register 80 Bytes
46Fh	Common RAM Accesses 70h – 7Fh	4EFh	Common RAM Accesses 70h – 7Fh	56Fh	Common RAM Accesses 70h – 7Fh	5EFh	Common RAM Accesses 70h – 7Fh	66Fh	Common RAM Accesses 70h – 7Fh	6EFh	Common RAM Accesses 70h – 7Fh	76Fh	Common RAM Accesses 70h – 7Fh	7EFh	Common RAM Accesses 70h – 7Fh
470h	—	4F0h	—	570h	—	5F0h	—	670h	—	6F0h	—	770h	—	7F0h	—
47Fh	—	4FFh	—	57Fh	—	5FFh	—	67Fh	—	6FFh	—	77Fh	—	7FFh	—

Legend: = Unimplemented data memory locations, read as '0'.

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FIGURE 3-7: ACCESSING THE STACK EXAMPLE 4



3.4.2 OVERFLOW/UNDERFLOW RESET

If the `STVREN` bit in Configuration Words is programmed to '1', the device will be Reset if the stack is `PUSHed` beyond the sixteenth level or `POPed` beyond the first level, setting the appropriate bits (`STKOVF` or `STKUNF`, respectively) in the `PCON` register.

3.5 Indirect Addressing

The `INDFn` registers are not physical registers. Any instruction that accesses an `INDFn` register actually accesses the register at the address specified by the File Select Registers (`FSR`). If the `FSRn` address specifies one of the two `INDFn` registers, the read will return '0' and the write will not occur (though Status bits may be affected). The `FSRn` register value is created by the pair `FSRnH` and `FSRnL`.

The `FSR` registers form a 16-bit address that allows an addressing space with 65536 locations. These locations are divided into three memory regions:

- Traditional Data Memory
- Linear Data Memory
- Data EEPROM Memory
- Program Flash Memory

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6.2.1.4 4x PLL

The oscillator module contains a PLL that can be used with external clock sources to provide a system clock source. The input frequency for the PLL must fall within specifications. See the PLL Clock Timing Specifications in Table 37-9.

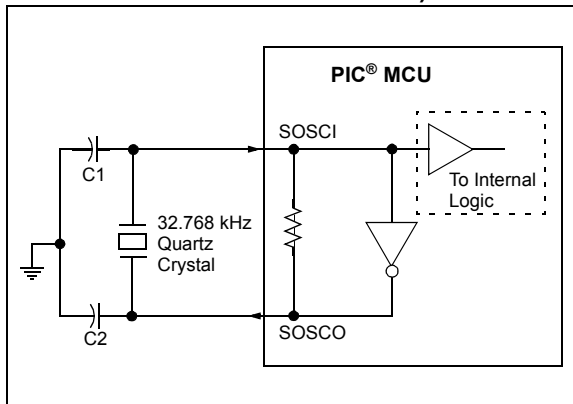
The PLL may be enabled for use by one of two methods:

1. Program the RSTOSC bits in the Configuration Word 1 to enable the EXTOSC with 4x PLL.
2. Write the NOSC bits in the OSCCON1 register to enable the EXTOSC with 4x PLL.

6.2.1.5 Secondary Oscillator

The secondary oscillator is a separate oscillator block that can be used as an alternate system clock source. The secondary oscillator is optimized for 31 kHz, and can be used with an external crystal oscillator connected to the SOSCI and SOSCO device pins, or an external clock source connected to the SOSCIN pin. The secondary oscillator can be selected during run-time using clock switching. Refer to **Section 6.3 “Clock Switching”** for more information.

FIGURE 6-5: QUARTZ CRYSTAL OPERATION (SECONDARY OSCILLATOR)



Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.

2: Always verify oscillator performance over the V_{DD} and temperature range that is expected for the application.

3: For oscillator design assistance, reference the following Microchip Application Notes:

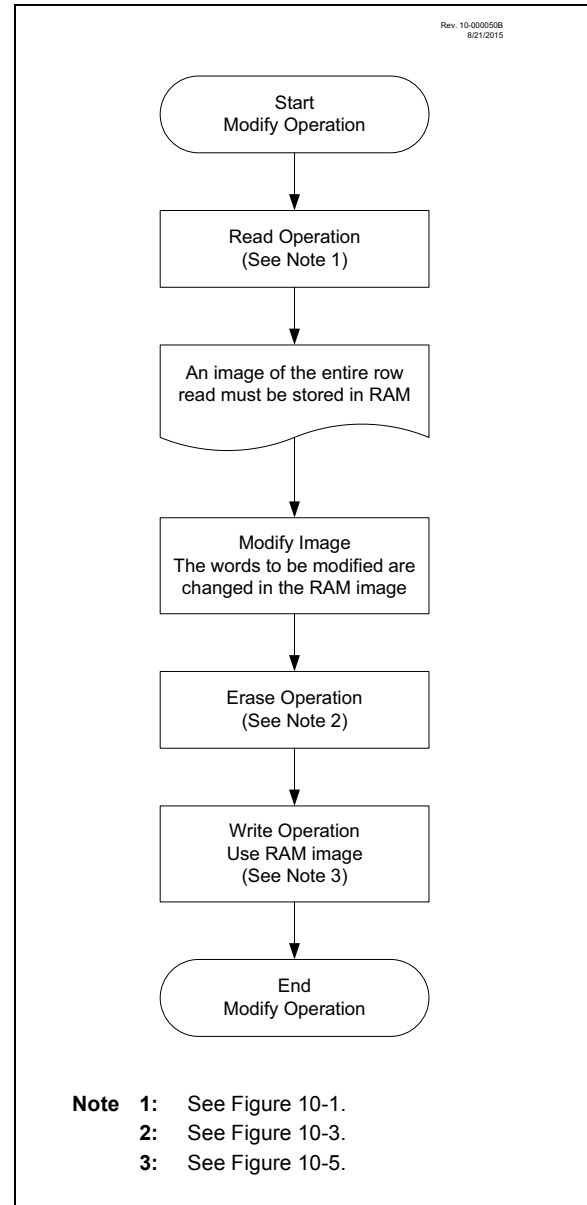
- AN826, “Crystal Oscillator Basics and Crystal Selection for *rf*PIC[®] and PIC[®] Devices” (DS00826)
- AN849, “Basic PIC[®] Oscillator Design” (DS00849)
- AN943, “Practical PIC[®] Oscillator Analysis and Design” (DS00943)
- AN949, “Making Your Oscillator Work” (DS00949)
- TB097, “Interfacing a Micro Crystal MS1V-T1K 32.768 kHz Tuning Fork Crystal to a PIC16F690/SS” (DS91097)
- AN1288, “Design Practices for Low-Power External Oscillators” (DS01288)

10.4.6 MODIFYING FLASH PROGRAM MEMORY

When modifying existing data in a program memory row, and data within that row must be preserved, it must first be read and saved in a RAM image. Program memory is modified using the following steps:

1. Load the starting address of the row to be modified.
2. Read the existing data from the row into a RAM image.
3. Modify the RAM image to contain the new data to be written into program memory.
4. Load the starting address of the row to be rewritten.
5. Erase the program memory row.
6. Load the write latches with data from the RAM image.
7. Initiate a programming operation.

FIGURE 10-6: FLASH PROGRAM MEMORY MODIFY FLOWCHART

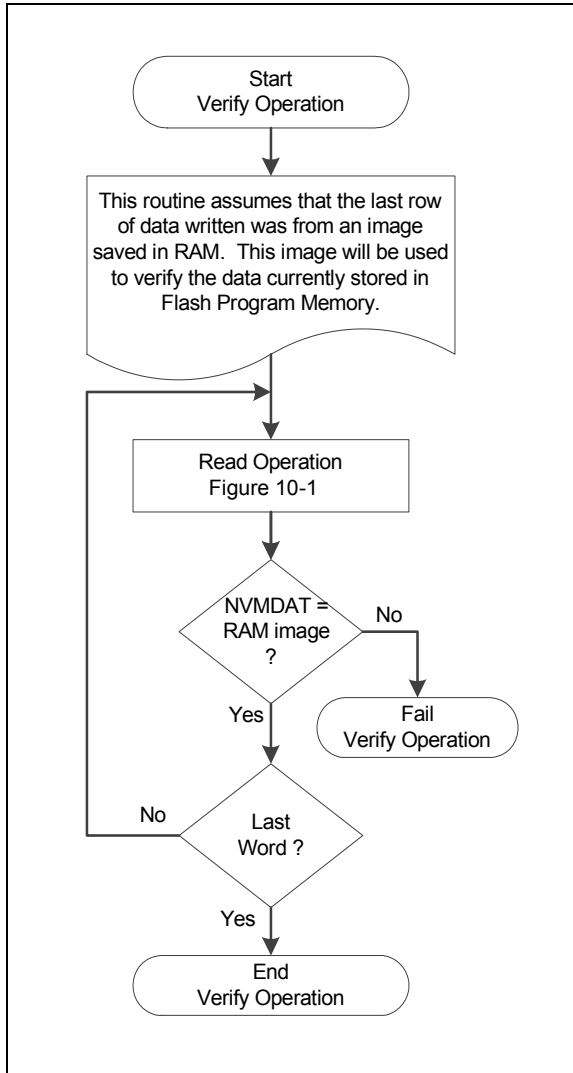


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10.4.8 WRITE VERIFY

It is considered good programming practice to verify that program memory writes agree with the intended value. Since program memory is stored as a full page then the stored program memory contents are compared with the intended data stored in RAM after the last write is complete.

FIGURE 10-8: FLASH PROGRAM MEMORY VERIFY FLOWCHART



11.7 Configuring the CRC

The following steps illustrate how to properly configure the CRC.

1. Determine if the automatic Program Memory scan will be used with the scanner or manual calculation through the SFR interface and perform the actions specified in **Section 11.4 “CRC Data Sources”**, depending on which decision was made.
2. If desired, seed a starting CRC value into the CRCACCH/L registers.
3. Program the CRCXORH/L registers with the desired generator polynomial.
4. Program the DLEN<3:0> bits of the CRCCON1 register with the length of the data word – 1 (refer to Example 11-1). This determines how many times the shifter will shift into the accumulator for each data word.
5. Program the PLEN<3:0> bits of the CRCCON1 register with the length of the polynomial – 2 (refer to Example 11-1).
6. Determine whether shifting in trailing zeros is desired and set the ACCM bit of CRCCON0 register appropriately.
7. Likewise, determine whether the MSb or LSb should be shifted first and write the SHIFTM bit of CRCCON0 register appropriately.
8. Write the CRCGO bit of the CRCCON0 register to begin the shifting process.
- 9a. If manual SFR entry is used, monitor the FULL bit of CRCCON0 register. When FULL = 0, another word of data can be written to the CRCDATH/L registers, keeping in mind that CRCDATH should be written first if the data has >8 bits, as the shifter will begin upon the CRCDATH register being written.
- 9b. If the scanner is used, the scanner will automatically stuff words into the CRCDATH/L registers as needed, as long as the SCANGO bit is set.
- 10a. If using the Flash memory scanner, monitor the SCANIF (or the SCANGO bit) for the scanner to finish pushing information into the CRCDATA registers. After the scanner is completed, monitor the CRCIF (or the BUSY bit) to determine that the CRC has been completed and the check value can be read from the CRCACC registers. If both the interrupt flags are set (or both BUSY and SCANGO bits are cleared), the completed CRC calculation can be read from the CRCACCH/L registers.
- 10b. If manual entry is used, monitor the CRCIF (or BUSY bit) to determine when the CRCACC registers will hold the check value.

11.8 Program Memory Scan Configuration

If desired, the Program Memory Scan module may be used in conjunction with the CRC module to perform a CRC calculation over a range of program memory addresses. In order to set up the Scanner to work with the CRC you need to perform the following steps:

1. Set the EN bit to enable the module. This can be performed at any point preceding the setting of the SCANGO bit, but if it gets disabled, all internal states of the Scanner are reset (registers are unaffected).
2. Choose which memory access mode is to be used (see **Section 11.10 “Scanning Modes”**) and set the MODE bits of the SCANCON0 register appropriately.
3. Based on the memory access mode, set the INTM bits of the SCANCON0 register to the appropriate interrupt mode (see **Section 11.10.5 “Interrupt Interaction”**).
4. Set the SCANLADRL/H and SCANHADRL/H registers with the beginning and ending locations in memory that are to be scanned.
5. Begin the scan by setting the SCANGO bit in the SCANCON0 register. The scanner will wait (CRCGO must be set) for the signal from the CRC that it is ready for the first Flash memory location, then begin loading data into the CRC. It will continue to do so until it either hits the configured end address or an address that is unimplemented on the device, at which point the SCANGO bit will clear, Scanner functions will cease, and the SCANIF interrupt will be triggered. Alternately, the SCANGO bit can be cleared in software if desired.

11.9 Scanner Interrupt

The scanner will trigger an interrupt when the SCANGO bit transitions from ‘1’ to ‘0’. The SCANIF interrupt flag of PIR7 is set when the last memory location is reached and the data is entered into the CRCDATA registers. The SCANIF bit can only be cleared in software. The SCAN interrupt enable is the SCANIE bit of the PIE7 register.

11.10 Scanning Modes

The memory scanner can scan in four modes: Burst, Peek, Concurrent, and Triggered. These modes are controlled by the MODE bits of the SCANCON0 register. The four modes are summarized in Table 11-1.

12.10.8 CURRENT-CONTROLLED DRIVE MODE CONTROL

The CCDPD and CCDND registers (Register 12-40 and Register 12-41) control the Current-Controlled Drive mode for both the positive-going and negative-going drivers. When a CCDPD[y] or CCDND[y] bit is set and the CCDEN bit of the CCDCON register is set, the Current-Controlled mode is enabled for the corresponding port pin. When the CCDPD[y] or CCDND[y] bit is clear, the Current-Controlled mode for the corresponding port pin is disabled. If the CCDPD[y] or CCDND[y] bit is set and the CCDEN bit is clear, operation of the port pin is undefined (see **Section 12.1.1 “Current-Controlled Drive”** for current-controlled use precautions).

12.10.9 PORTD FUNCTIONS AND OUTPUT PRIORITIES

Each pin defaults to the PORT latch data after Reset. Other output functions are selected with the peripheral pin select logic. See **Section 13.0 “Peripheral Pin Select (PPS) Module”** for more information.

Analog input functions, such as ADC and comparator inputs, are not shown in the peripheral pin select lists. Digital output functions may continue to control the pin when it is in Analog mode.

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REGISTER 12-53: CCDPE: CURRENT CONTROL DRIVE NEGATIVE PORTE REGISTER

U-0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	—	—	CCDPE2	CCDPE1	CCDPE0
bit 7					bit 0		

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7-3 **Unimplemented:** Read as '0'

bit 2-0 **CCDPE<2:0>:** RE<2:0> Current Control Drive Positive Control bits⁽¹⁾

1 = Current control source enabled

0 = Current control source disabled

Note 1: If CCDPEy is set, when CCDEN = 0 (Register 12-1), operation of the pin is undefined.

REGISTER 12-54: CCDNE: CURRENT CONTROL DRIVE NEGATIVE PORTE REGISTER

U-0	U-0	U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	—	—	CCDNE2	CCDNE1	CCDNE0
bit 7					bit 0		

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7-3 **Unimplemented:** Read as '0'

bit 2-0 **CCDNE<2:0>:** RE<2:0> Current Control Drive Negative Control bits⁽¹⁾

1 = Current control source enabled

0 = Current control source disabled

Note 1: If CCDNEy is set, when CCDEN = 0 (Register 12-1), operation of the pin is undefined.

24.2 FIXED DUTY CYCLE MODE

In Fixed Duty Cycle (FDC) mode, every time the accumulator overflows (NCO_overflow), the output is toggled. This provides a 50% duty cycle, provided that the increment value remains constant. For more information, see Figure 24-2.

The FDC mode is selected by clearing the N1PFM bit in the NCO1CON register.

24.3 PULSE FREQUENCY MODE

In Pulse Frequency (PF) mode, every time the Accumulator overflows, the output becomes active for one or more clock periods. Once the clock period expires, the output returns to an inactive state. This provides a pulsed output. The output becomes active on the rising clock edge immediately following the overflow event. For more information, see Figure 24-2.

The value of the active and inactive states depends on the polarity bit, N1POL in the NCO1CON register.

The PF mode is selected by setting the N1PFM bit in the NCO1CON register.

24.3.1 OUTPUT PULSE WIDTH CONTROL

When operating in PF mode, the active state of the output can vary in width by multiple clock periods. Various pulse widths are selected with the N1PWS<2:0> bits in the NCO1CLK register.

When the selected pulse width is greater than the Accumulator overflow time frame, then DDS operation is undefined.

24.4 OUTPUT POLARITY CONTROL

The last stage in the NCO module is the output polarity. The N1POL bit in the NCO1CON register selects the output polarity. Changing the polarity while the interrupts are enabled will cause an interrupt for the resulting output transition.

The NCO output signal is available to the following peripherals:

- CLC
- CWG
- Timer1/3/5
- Timer2/4/6
- SMT
- DSM
- Reference Clock Output

24.5 Interrupts

When the accumulator overflows (NCO_overflow), the NCO Interrupt Flag bit, NCO1IF, of the PIR7 register is set. To enable the interrupt event (NCO_interrupt), the following bits must be set:

- N1EN bit of the NCO1CON register
- NCO1IE bit of the PIE7 register
- PEIE bit of the INTCON register
- GIE bit of the INTCON register

The interrupt must be cleared by software by clearing the NCO1IF bit in the Interrupt Service Routine.

24.6 Effects of a Reset

All of the NCO registers are cleared to zero as the result of a Reset.

24.7 Operation in Sleep

The NCO module operates independently from the system clock and will continue to run during Sleep, provided that the clock source selected remains active.

The HFINTOSC remains active during Sleep when the NCO module is enabled and the HFINTOSC is selected as the clock source, regardless of the system clock source selected.

In other words, if the HFINTOSC is simultaneously selected as the system clock and the NCO clock source, when the NCO is enabled, the CPU will go idle during Sleep, but the NCO will continue to operate and the HFINTOSC will remain active.

This will have a direct effect on the Sleep mode current.

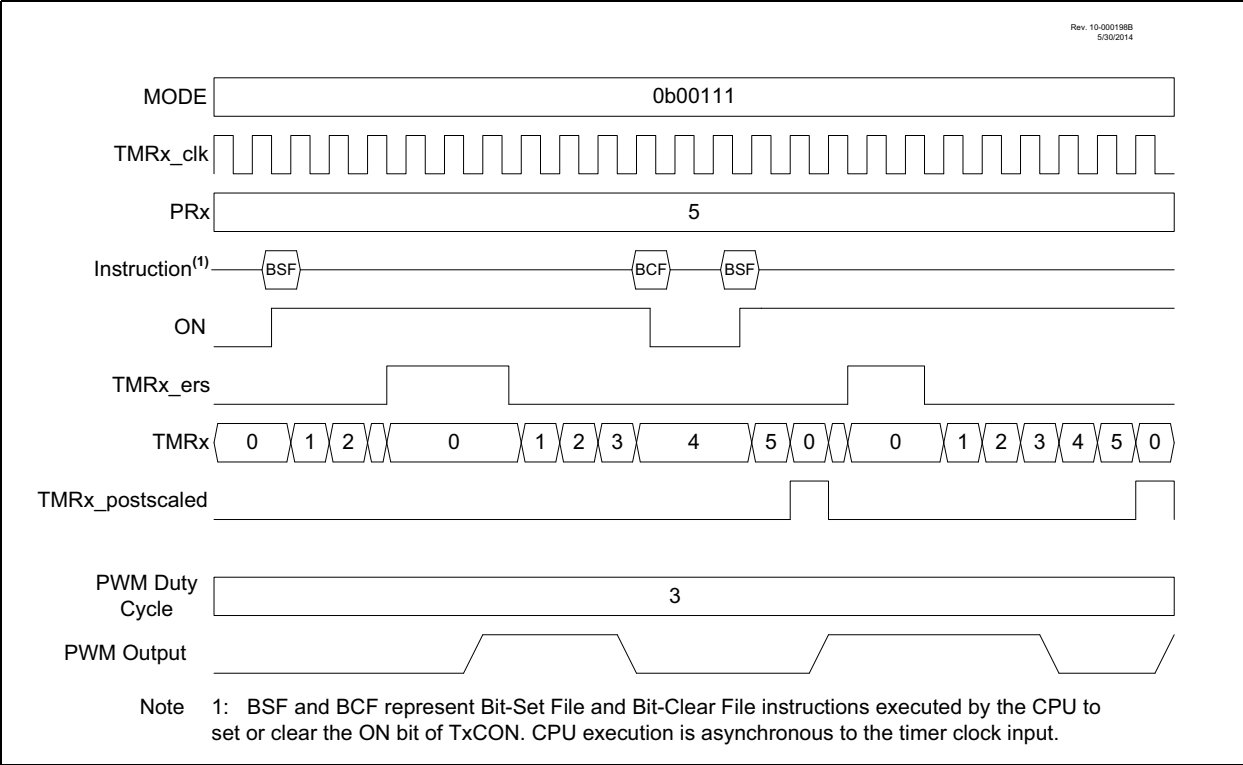
29.5.4 LEVEL-TRIGGERED HARDWARE LIMIT MODE

In the Level-Triggered Hardware Limit Timer modes the counter is reset by high or low levels of the external signal TMRx_ers, as shown in Figure 29-7. Selecting MODE<4:0> = 00110 will cause the timer to reset on a low level external signal. Selecting MODE<4:0> = 00111 will cause the timer to reset on a high level external signal. In the example, the counter is reset while TMRx_ers = 1. ON is controlled by BSF and BCF instructions. When ON = 0 the external signal is ignored.

When the CCP uses the timer as the PWM time base then the PWM output will be set high when the timer starts counting and then set low only when the timer count matches the CCPRx value. The timer is reset when either the timer count matches the PRx value or two clock periods after the external Reset signal goes true and stays true.

The timer starts counting, and the PWM output is set high, on either the clock following the PRx match or two clocks after the external Reset signal relinquishes the Reset. The PWM output will remain high until the timer counts up to match the CCPRx pulse width value. If the external Reset signal goes true while the PWM output is high then the PWM output will remain high until the Reset signal is released allowing the timer to count up to match the CCPRx value.

FIGURE 29-7: LEVEL-TRIGGERED HARDWARE LIMIT MODE TIMING DIAGRAM (MODE = 00111)



PIC16(L)F18856/76

REGISTER 29-2: TxCON: TIMER2/4/6 CONTROL REGISTER

R/W/HC-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
ON ⁽¹⁾	CKPS<2:0>			OUTPS<3:0>			
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

HC = Bit is cleared by hardware

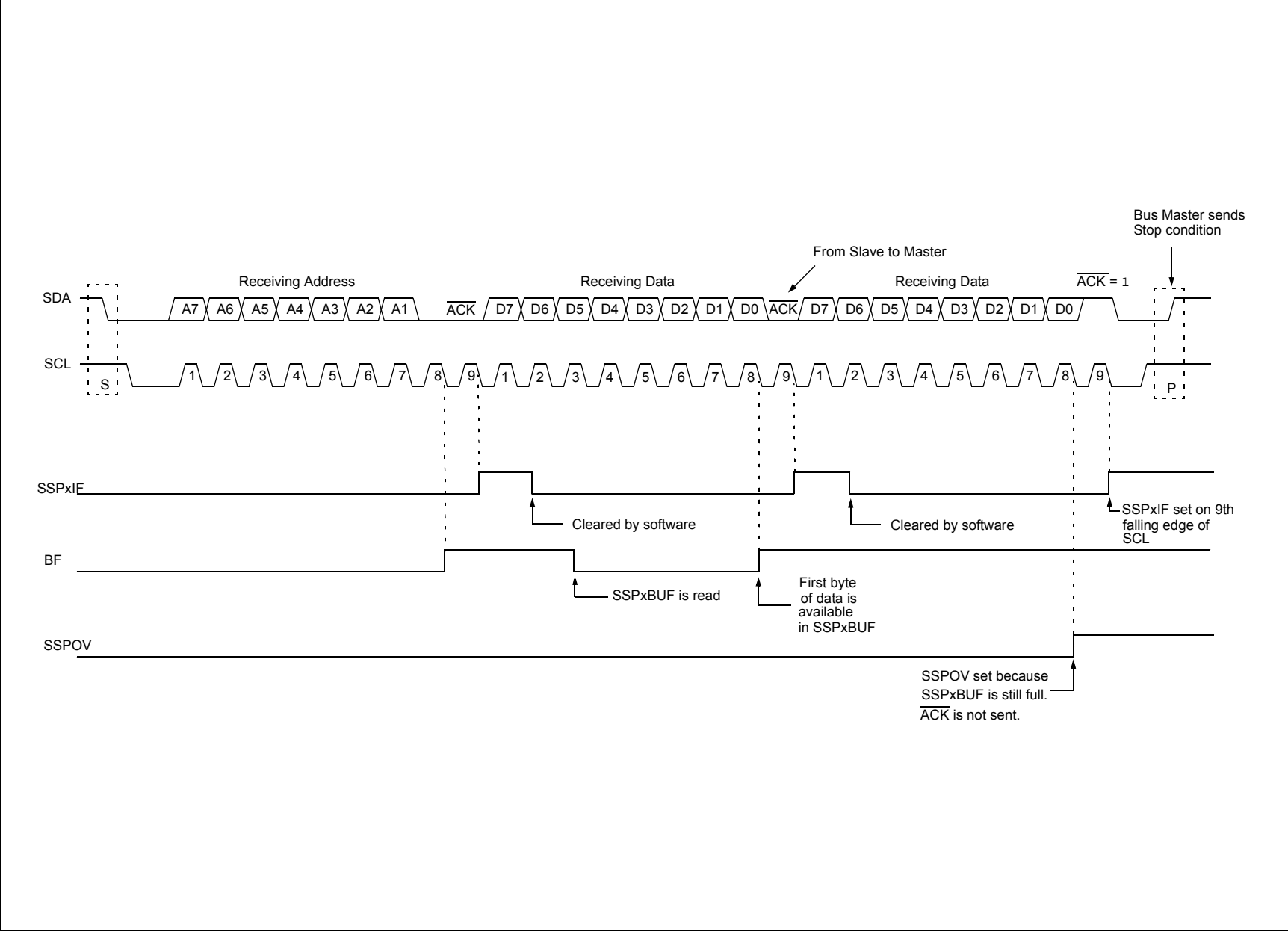
bit 7 **ON:** Timerx On bit
1 = Timerx is on
0 = Timerx is off: all counters and state machines are reset

bit 6-4 **CKPS<2:0>:** Timer2-type Clock Prescale Select bits
111 = 1:128 Prescaler
110 = 1:64 Prescaler
101 = 1:32 Prescaler
100 = 1:16 Prescaler
011 = 1:8 Prescaler
010 = 1:4 Prescaler
001 = 1:2 Prescaler
000 = 1:1 Prescaler

bit 3-0 **OUTPS<3:0>:** Timerx Output Postscaler Select bits
1111 = 1:16 Postscaler
1110 = 1:15 Postscaler
1101 = 1:14 Postscaler
1100 = 1:13 Postscaler
1011 = 1:12 Postscaler
1010 = 1:11 Postscaler
1001 = 1:10 Postscaler
1000 = 1:9 Postscaler
0111 = 1:8 Postscaler
0110 = 1:7 Postscaler
0101 = 1:6 Postscaler
0100 = 1:5 Postscaler
0011 = 1:4 Postscaler
0010 = 1:3 Postscaler
0001 = 1:2 Postscaler
0000 = 1:1 Postscaler

Note 1: In certain modes, the ON bit will be auto-cleared by hardware. See **Section 29.5 “Operation Examples”**.

FIGURE 31-14: I²C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 0, AHEN = 0, DHEN = 0)



32.6.3 PERIOD AND DUTY-CYCLE MODE

In Duty-Cycle mode, either the duty cycle or period (depending on polarity) of the SMTx_signal can be acquired relative to the SMT clock. The CPW register is updated on a falling edge of the signal, and the CPR register is updated on a rising edge of the signal, along with the SMTxTMR resetting to 0x0001. In addition, the SMTxGO bit is reset on a rising edge when the SMT is in Single Acquisition mode. See Figure 32-6 and Figure 32-7.

33.1.2.8 Asynchronous Reception Setup:

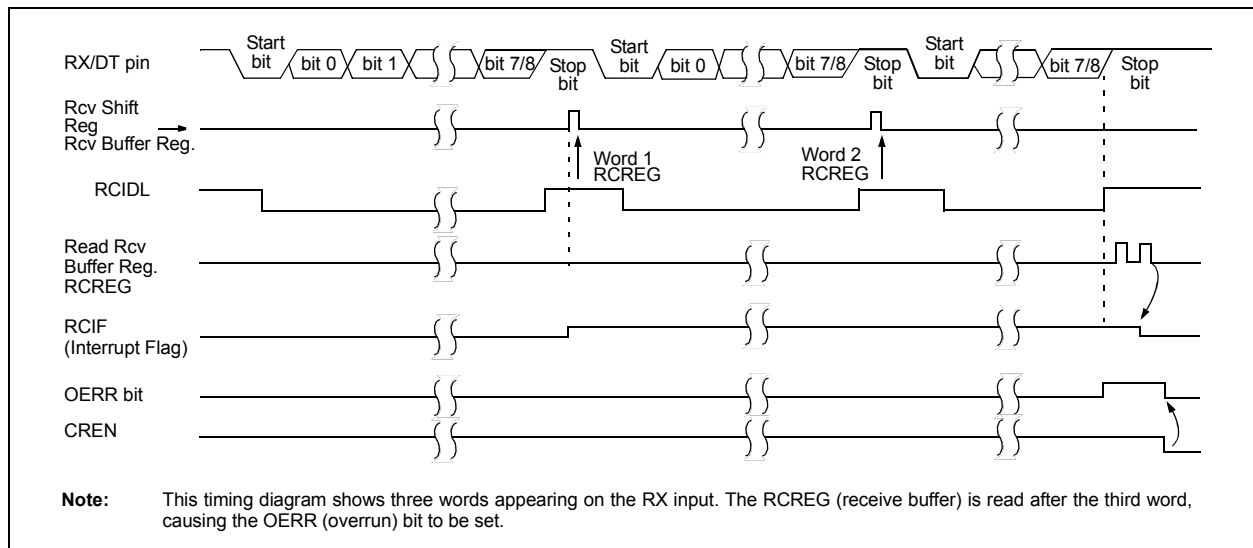
1. Initialize the SPBRGH, SPBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see **Section 33.3 “EUSART Baud Rate Generator (BRG)”**).
2. Clear the ANSEL bit for the RX pin (if applicable).
3. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
4. If interrupts are desired, set the RCIE bit of the PIE3 register and the GIE and PEIE bits of the INTCON register.
5. If 9-bit reception is desired, set the RX9 bit.
6. Enable reception by setting the CREN bit.
7. The RCIF interrupt flag bit will be set when a character is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
8. Read the RC1STA register to get the error flags and, if 9-bit data reception is enabled, the ninth data bit.
9. Get the received eight Least Significant data bits from the receive buffer by reading the RCREG register.
10. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.

33.1.2.9 9-bit Address Detection Mode Setup

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

1. Initialize the SPBRGH, SPBRGL register pair and the BRGH and BRG16 bits to achieve the desired baud rate (see **Section 33.3 “EUSART Baud Rate Generator (BRG)”**).
2. Clear the ANSEL bit for the RX pin (if applicable).
3. Enable the serial port by setting the SPEN bit. The SYNC bit must be clear for asynchronous operation.
4. If interrupts are desired, set the RCIE bit of the PIE3 register and the GIE and PEIE bits of the INTCON register.
5. Enable 9-bit reception by setting the RX9 bit.
6. Enable address detection by setting the ADDEN bit.
7. Enable reception by setting the CREN bit.
8. The RCIF interrupt flag bit will be set when a character with the ninth bit set is transferred from the RSR to the receive buffer. An interrupt will be generated if the RCIE interrupt enable bit was also set.
9. Read the RC1STA register to get the error flags. The ninth data bit will always be set.
10. Get the received eight Least Significant data bits from the receive buffer by reading the RCREG register. Software determines if this is the device's address.
11. If an overrun occurred, clear the OERR flag by clearing the CREN receiver enable bit.
12. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and generate interrupts.

FIGURE 33-5: ASYNCHRONOUS RECEPTION



PIC16(L)F18856/76

FIGURE 37-1: VOLTAGE FREQUENCY GRAPH, $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$, PIC16F18856/76 ONLY

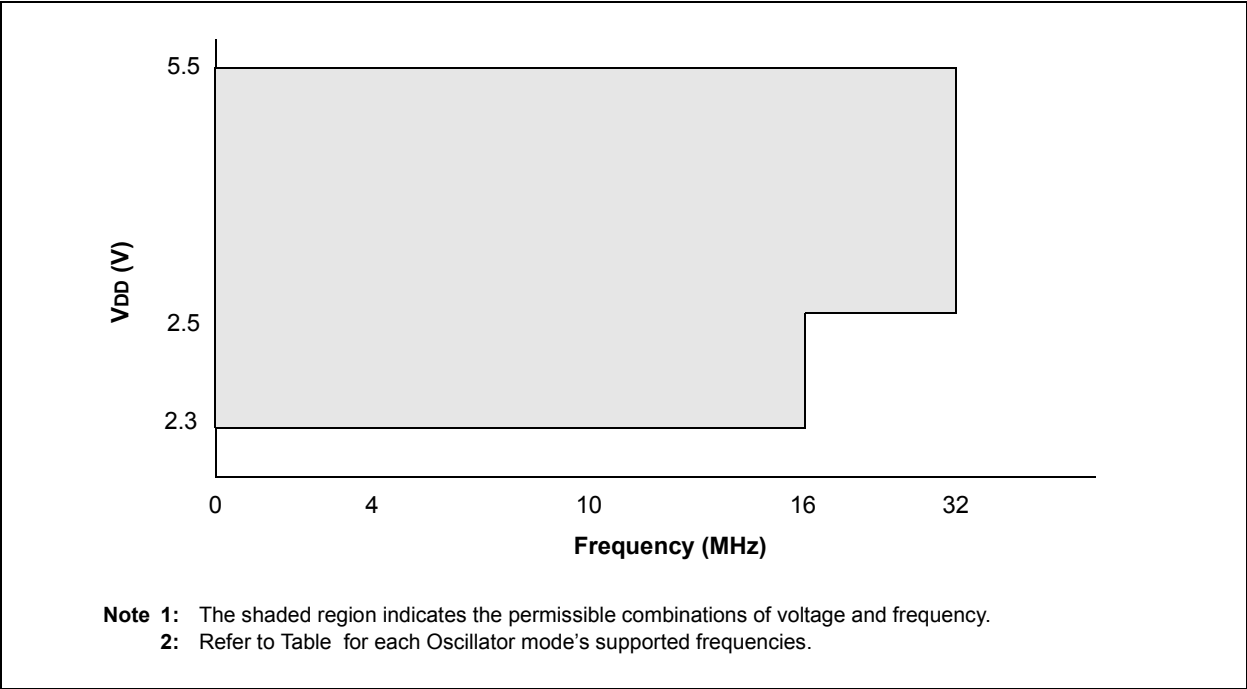
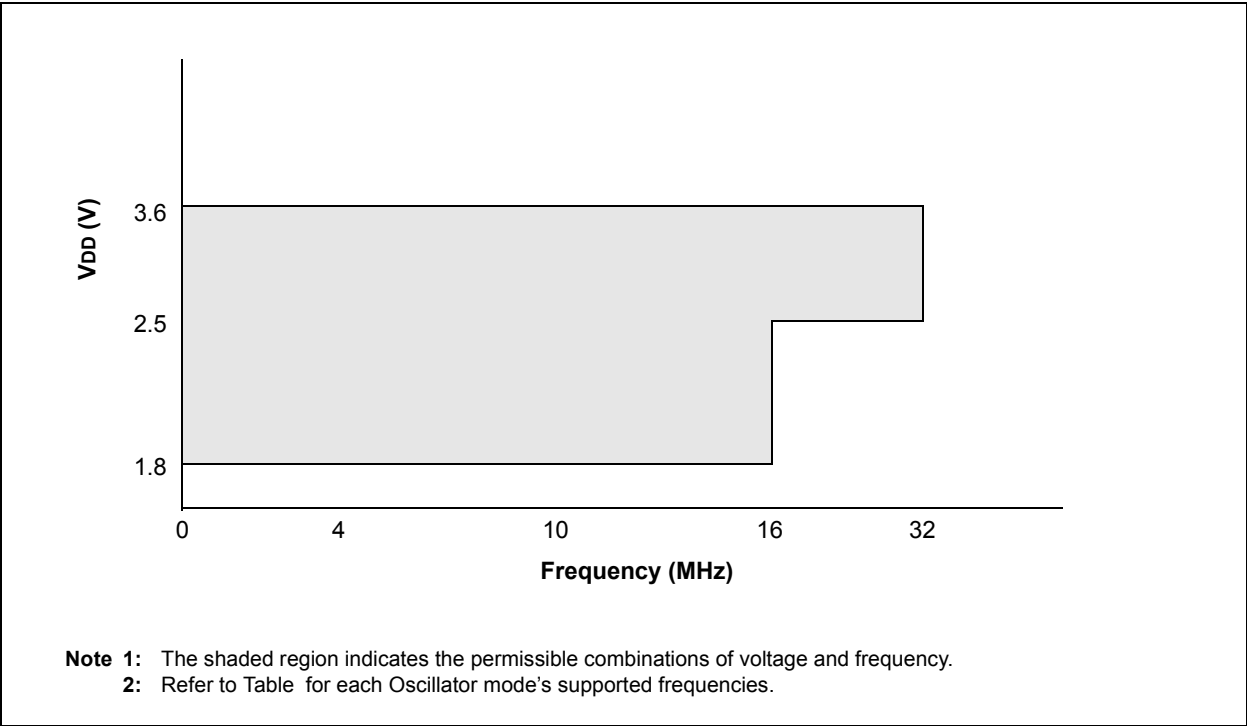


FIGURE 37-2: VOLTAGE FREQUENCY GRAPH, $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$, PIC16LF18856/76 ONLY



PIC16(L)F18856/76

TABLE 37-14: COMPARATOR SPECIFICATIONS

Operating Conditions (unless otherwise stated) V _{DD} = 3.0V, T _A = 25°C							
Param. No.	Sym.	Characteristics	Min.	Typ.	Max.	Units	Comments
CM01	V _{IOFF}	Input Offset Voltage	—	—	±30	mV	V _{ICM} = V _{DD} /2
CM02	V _{ICM}	Input Common Mode Range	GND	—	V _{DD}	V	
CM03	CMRR	Common Mode Input Rejection Ratio	—	50	—	dB	
CM04	V _{HYST}	Comparator Hysteresis	15	25	35	mV	
CM05	T _{RESP} ⁽¹⁾	Response Time, Rising Edge	—	300	600	ns	
		Response Time, Falling Edge	—	220	500	ns	

* These parameters are characterized but not tested.

Note 1: Response time measured with one comparator input at V_{DD}/2, while the other input transitions from V_{SS} to V_{DD}.

Note 2: A mode change includes changing any of the control register values, including module enable.

TABLE 37-15: 5-BIT DAC SPECIFICATIONS

Standard Operating Conditions (unless otherwise stated) V _{DD} = 3.0V, T _A = 25°C							
Param. No.	Sym.	Characteristics	Min.	Typ.	Max.	Units	Comments
DSB01	V _{LSB}	Step Size	—	(V _{DACREF+} - V _{DACREF-}) / 32	—	V	
DSB01	V _{ACC}	Absolute Accuracy	—	—	± 0.5	LSb	
DSB03*	R _{UNIT}	Unit Resistor Value	—	5000	—	Ω	
DSB04*	T _{ST}	Settling Time ⁽¹⁾	—	—	10	μs	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Settling time measured while DACR<4:0> transitions from '00000' to '01111'.

TABLE 37-16: FIXED VOLTAGE REFERENCE (FVR) SPECIFICATIONS

Standard Operating Conditions (unless otherwise stated)							
Param. No.	Symbol	Characteristic	Min.	Typ.	Max.	Units	Conditions
FVR01	V _{FVR1}	1x Gain (1.024V)	-4	—	+4	%	V _{DD} ≥ 2.5V, -40°C to 85°C
FVR02	V _{FVR2}	2x Gain (2.048V)	-4	—	+4	%	V _{DD} ≥ 2.5V, -40°C to 85°C
FVR03	V _{FVR4}	4x Gain (4.096V)	-5	—	+5	%	V _{DD} ≥ 4.75V, -40°C to 85°C
FVR04	T _{FVRST}	FVR Start-up Time	—	25	—	us	

TABLE 37-17: ZERO CROSS DETECT (ZCD) SPECIFICATIONS

Standard Operating Conditions (unless otherwise stated) V _{DD} = 3.0V, T _A = 25°C							
Param. No.	Sym.	Characteristics	Min	Typ†	Max	Units	Comments
ZC01	V _{PINZC}	Voltage on Zero Cross Pin	—	0.75	—	V	
ZC02	I _{ZCD_MAX}	Maximum source or sink current	—	—	600	μA	
ZC03	T _{RESPH}	Response Time, Rising Edge	—	1	—	μs	
	T _{RESPL}	Response Time, Falling Edge	—	1	—	μs	

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

38.0 DC AND AC CHARACTERISTICS GRAPHS AND CHARTS

The graphs and tables provided in this section are for **design guidance** and are **not tested**.

In some graphs or tables, the data presented are **outside specified operating range** (i.e., outside specified VDD range). This is for **information only** and devices are ensured to operate properly only within the specified range.

Unless otherwise noted, all graphs apply to both the L and LF devices.

Note:	The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore, outside the warranted range.
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“Typical” represents the mean of the distribution at 25°C. “Maximum”, “Max.”, “Minimum” or “Min.” represents (mean + 3σ) or (mean - 3σ) respectively, where σ is a standard deviation, over each temperature range.

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