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Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	2 Core, 32-Bit
Speed	1.5GHz
Co-Processors/DSP	Signal Processing; SPE, Security; SEC
RAM Controllers	DDR2, DDR3
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.5V, 1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	1023-BFBGA, FCBGA
Supplier Device Package	1023-FCBGA (33x33)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8572elpxavne

- Supports fully nested interrupt delivery
- Interrupts can be routed to external pin for external processing.
- Interrupts can be routed to the e500 core's standard or critical interrupt inputs.
- Interrupt summary registers allow fast identification of interrupt source.
- Integrated security engine (SEC) optimized to process all the algorithms associated with IPsec, IKE, SSL/TLS, SRTP, 802.16e, and 3GPP
 - Four crypto-channels, each supporting multi-command descriptor chains
 - Dynamic assignment of crypto-execution units through an integrated controller
 - Buffer size of 256 bytes for each execution unit, with flow control for large data sizes
 - PKEU—public key execution unit
 - RSA and Diffie-Hellman; programmable field size up to 4096 bits
 - Elliptic curve cryptography with F_{2^m} and $F(p)$ modes and programmable field size up to 1023 bits
 - DEU—Data Encryption Standard execution unit
 - DES, 3DES
 - Two key (K1, K2, K1) or three key (K1, K2, K3)
 - ECB, CBC and OFB-64 modes for both DES and 3DES
 - AESU—Advanced Encryption Standard unit
 - Implements the Rijndael symmetric key cipher
 - ECB, CBC, CTR, CCM, GCM, CMAC, OFB-128, CFB-128, and LRW modes
 - 128-, 192-, and 256-bit key lengths
 - AFEU—ARC four execution unit
 - Implements a stream cipher compatible with the RC4 algorithm
 - 40- to 128-bit programmable key
 - MDEU—message digest execution unit
 - SHA-1 with 160-bit message digest
 - SHA-2 (SHA-256, SHA-384, SHA-512)
 - MD5 with 128-bit message digest
 - HMAC with all algorithms
 - KEU—Kasumi execution unit
 - Implements F8 algorithm for encryption and F9 algorithm for integrity checking
 - Also supports A5/3 and GEA-3 algorithms
 - RNG—random number generator
 - XOR engine for parity checking in RAID storage applications
 - CRC execution unit
 - CRC-32 and CRC-32C
- Pattern Matching Engine with DEFLATE decompression

Figure 1 shows the MPC8572E block diagram.

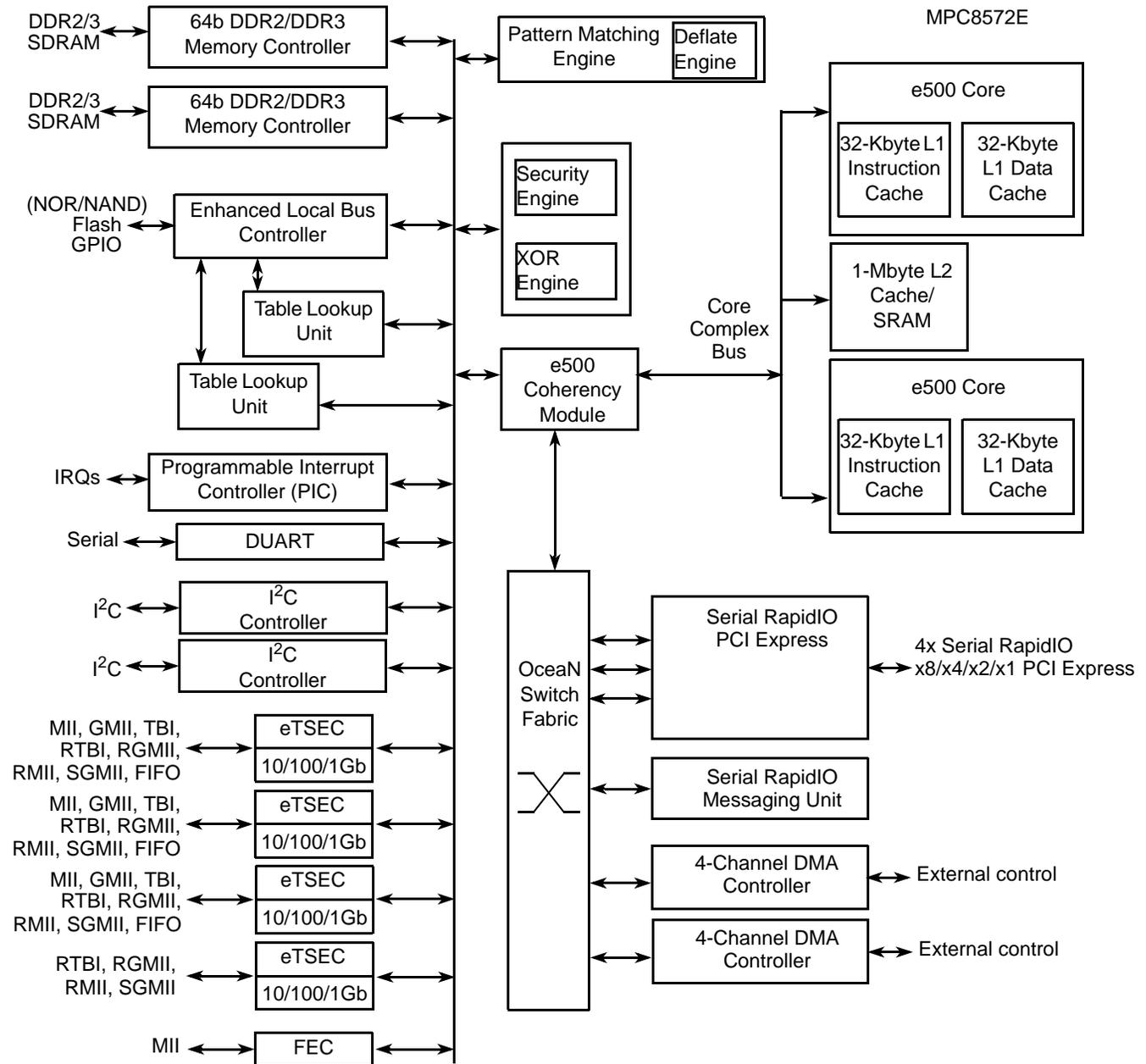


Figure 1. MPC8572E Block Diagram

2 Electrical Characteristics

This section provides the AC and DC electrical specifications for the MPC8572E. The MPC8572E is currently targeted to these specifications. Some of these specifications are independent of the I/O cell, but are included for a more complete reference. These are not purely I/O buffer design specifications.

2.1 Overall DC Electrical Characteristics

This section covers the ratings, conditions, and other characteristics.

2.1.1 Absolute Maximum Ratings

Table 1 provides the absolute maximum ratings.

Table 1. Absolute Maximum Ratings¹

Characteristic		Symbol	Range	Unit	Notes
Core supply voltage		V_{DD}	-0.3 to 1.21	V	—
PLL supply voltage		AV_{DD}	-0.3 to 1.21	V	—
Core power supply for SerDes transceivers		SV_{DD}	-0.3 to 1.21	V	—
Pad power supply for SerDes transceivers		XV_{DD}	-0.3 to 1.21	V	—
DDR SDRAM Controller I/O supply voltage	DDR2 SDRAM Interface	GV_{DD}	-0.3 to 1.98	V	—
	DDR3 SDRAM Interface	—	-0.3 to 1.65		—
Three-speed Ethernet I/O, FEC management interface, MII management voltage		LV_{DD} (for eTSEC1 and eTSEC2)	-0.3 to 3.63 -0.3 to 2.75	V	2
		TV_{DD} (for eTSEC3 and eTSEC4, FEC)	-0.3 to 3.63 -0.3 to 2.75	—	2
DUART, system control and power management, I ² C, and JTAG I/O voltage		OV_{DD}	-0.3 to 3.63	V	—
Local bus and GPIO I/O voltage		BV_{DD}	-0.3 to 3.63 -0.3 to 2.75 -0.3 to 1.98	V	—
Input voltage	DDR2 and DDR3 SDRAM interface signals	MV_{IN}	-0.3 to ($GV_{DD} + 0.3$)	V	3
	DDR2 and DDR3 SDRAM interface reference	MV_{REF}^n	-0.3 to ($GV_{DD}/2 + 0.3$)	V	—
	Three-speed Ethernet signals	LV_{IN} TV_{IN}	-0.3 to ($LV_{DD} + 0.3$) -0.3 to ($TV_{DD} + 0.3$)	V	3
	Local bus and GPIO signals	BV_{IN}	-0.3 to ($BV_{DD} + 0.3$)	—	—
	DUART, SYSCLK, system control and power management, I ² C, and JTAG signals	OV_{IN}	-0.3 to ($OV_{DD} + 0.3$)	V	3
Storage temperature range		T_{STG}	-55 to 150	°C	—

Notes:

- Functional operating conditions are given in Table 2. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.
- The 3.63V maximum is only supported when the port is configured in GMII, MII, RMII or TBI modes; otherwise the 2.75V maximum applies. See Section 8.2, “FIFO, GMII, MII, TBI, RGMII, RMII, and RTBI AC Timing Specifications,” for details on the recommended operating conditions per protocol.
- (M,L,O) V_{IN} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 2.

Figure 17 shows the TBI receive the timing diagram.

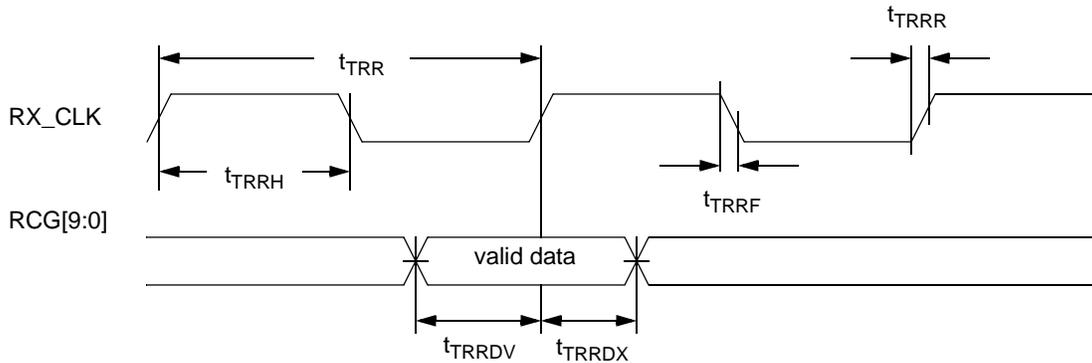


Figure 17. TBI Single-Clock Mode Receive AC Timing Diagram

8.2.6 RGMII and RTBI AC Timing Specifications

Table 34 presents the RGMII and RTBI AC timing specifications.

Table 34. RGMII and RTBI AC Timing Specifications

At recommended operating conditions with LV_{DD}/TV_{DD} of $2.5\text{ V} \pm 5\%$.

Parameter/Condition	Symbol ¹	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)	t_{SKRGT}	-500	0	500	ps
Data to clock input skew (at receiver) ²	t_{SKRGT}	1.0	—	2.8	ns
Clock period ³	t_{RGT}	7.2	8.0	8.8	ns
Duty cycle for 10BASE-T and 100BASE-TX ^{3, 4}	t_{RGTH}/t_{RGT}	40	50	60	%
Rise time (20%–80%)	t_{RGTR}	—	—	0.75	ns
Fall time (20%–80%)	t_{RGTF}	—	—	0.75	ns

Notes:

- Note that, in general, the clock reference symbol representation for this section is based on the symbols RGT to represent RGMII and RTBI timing. For example, the subscript of t_{RGT} represents the TBI (T) receive (RX) clock. Note also that the notation for rise (R) and fall (F) times follows the clock symbol that is being represented. For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (RGT).
- This implies that PC board design requires clocks to be routed such that an additional trace delay of greater than 1.5 ns will be added to the associated clock signal.
- For 10 and 100 Mbps, t_{RGT} scales to $400\text{ ns} \pm 40\text{ ns}$ and $40\text{ ns} \pm 4\text{ ns}$, respectively.
- Duty cycle may be stretched/shrunk during speed changes or while transitioning to a received packet's clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned between.

Figure 18 shows the RGMII and RTBI AC timing and multiplexing diagrams.

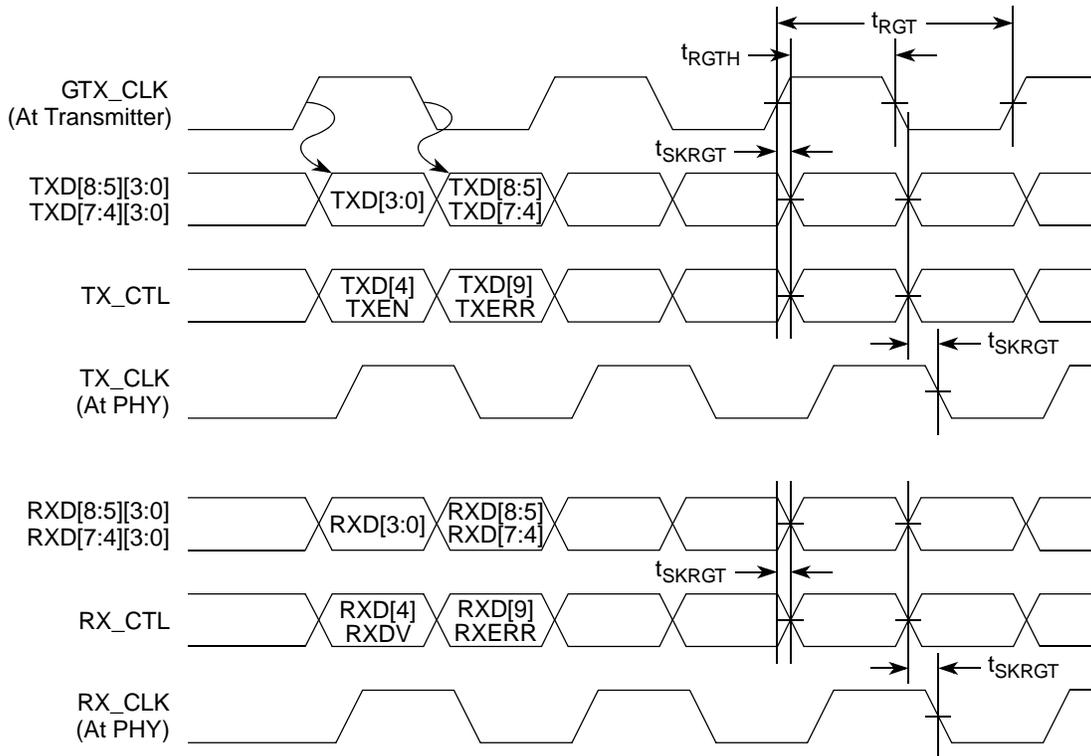


Figure 18. RGMII and RTBI AC Timing and Multiplexing Diagrams

8.2.7 RMII AC Timing Specifications

This section describes the RMII transmit and receive AC timing specifications.

8.2.7.1 RMII Transmit AC Timing Specifications

Table 35 shows the RMII transmit AC timing specifications.

Table 35. RMII Transmit AC Timing Specifications

At recommended operating conditions with LV_{DD}/TV_{DD} of 2.5/ 3.3 V ± 5%.

Parameter/Condition	Symbol ¹	Min	Typ	Max	Unit
TSECn_TX_CLK clock period	t _{RMT}	15.0	20.0	25.0	ns
TSECn_TX_CLK duty cycle	t _{RMTH}	35	50	65	%
TSECn_TX_CLK peak-to-peak jitter	t _{RMTJ}	—	—	250	ps
Rise time TSECn_TX_CLK (20%–80%)	t _{RMTR}	1.0	—	2.0	ns
Fall time TSECn_TX_CLK (80%–20%)	t _{RMTF}	1.0	—	2.0	ns

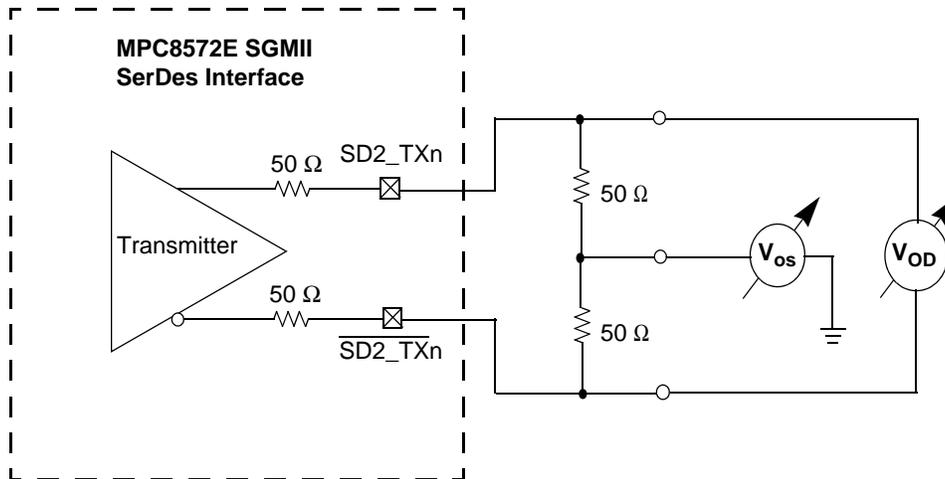


Figure 23. SGMII Transmitter DC Measurement Circuit

Table 39 lists the SGMII DC receiver electrical characteristics.

Table 39. SGMII DC Receiver Electrical Characteristics

Parameter	Symbol	Min	Typ	Max	Unit	Notes	
Supply Voltage	XV_{DD_SRDS2}	1.045	1.1	1.155	V	—	
DC Input voltage range	—	N/A			—	1	
Input differential voltage	LSTS = 0	$V_{RX_DIFFp-p}$	100	—	1200	mV	2, 4
	LSTS = 1		175	—			
Loss of signal threshold	LSTS = 0	VLOS	30	—	100	mV	3, 4
	LSTS = 1		65	—	175		
Input AC common mode voltage	V_{CM_ACp-p}	—	—	100	mV	5	
Receiver differential input impedance	Z_{RX_DIFF}	80	100	120	Ω	—	
Receiver common mode input impedance	Z_{RX_CM}	20	—	35	Ω	—	
Common mode input voltage	V_{CM}	—	$V_{xcorevss}$	—	V	6	

Note:

1. Input must be externally AC-coupled.
2. $V_{RX_DIFFp-p}$ is also referred to as peak to peak input differential voltage
3. The concept of this parameter is equivalent to the Electrical Idle Detect Threshold parameter in PCI Express. Refer to PCI Express Differential Receiver (RX) Input Specifications section for further explanation.
4. The LSTS shown in the table refers to the LSTSAB or LSTSEF bit field of MPC8572E's SerDes 2 Control Register.
5. V_{CM_ACp-p} is also referred to as peak to peak AC common mode voltage.
6. On-chip termination to SGND_SRDS2 (xcorevss).

8.3.4 SGMII AC Timing Specifications

This section describes the SGMII transmit and receive AC timing specifications. Transmitter and receiver characteristics are measured at the transmitter outputs ($\overline{SD2_TX[n]}$ and $\overline{SD2_TX}[n]$) or at the receiver inputs ($\overline{SD2_RX}[n]$ and $\overline{SD2_RX}[n]$) as depicted in Figure 25, respectively.

8.3.4.1 SGMII Transmit AC Timing Specifications

Table 40 provides the SGMII transmit AC timing targets. A source synchronous clock is not provided.

Table 40. SGMII Transmit AC Timing Specifications

At recommended operating conditions with $XV_{DD_SRDS2} = 1.1V \pm 5\%$.

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Deterministic Jitter	JD	—	—	0.17	UI p-p	—
Total Jitter	JT	—	—	0.35	UI p-p	—
Unit Interval	UI	799.92	800	800.08	ps	1
V_{OD} fall time (80%-20%)	t _{fall}	50	—	120	ps	—
V_{OD} rise time (20%-80%)	t _{rise}	50	—	120	ps	—

Notes:

- Each UI is 800 ps \pm 100 ppm.

8.3.4.2 SGMII Receive AC Timing Specifications

Table 41 provides the SGMII receive AC timing specifications. Source synchronous clocking is not supported. Clock is recovered from the data. Figure 24 shows the SGMII receiver input compliance mask eye diagram.

Table 41. SGMII Receive AC Timing Specifications

At recommended operating conditions with $XV_{DD_SRDS2} = 1.1V \pm 5\%$.

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Deterministic Jitter Tolerance	JD	0.37	—	—	UI p-p	1
Combined Deterministic and Random Jitter Tolerance	JDR	0.55	—	—	UI p-p	1
Sinusoidal Jitter Tolerance	JSIN	0.1	—	—	UI p-p	1
Total Jitter Tolerance	JT	0.65	—	—	UI p-p	1
Bit Error Ratio	BER	—	—	10^{-12}	—	—
Unit Interval	UI	799.92	800	800.08	ps	2
AC Coupling Capacitor	C _{TX}	5	—	200	nF	3

Notes:

- Measured at receiver.
- Each UI is 800 ps \pm 100 ppm.
- The external AC coupling capacitor is required. It is recommended to be placed near the device transmitter outputs.
- See *RapidIO 1x/4x LP Serial Physical Layer Specification* for interpretation of jitter specifications.

8.4 eTSEC IEEE Std 1588™ AC Specifications

Figure 26 shows the data and command output timing diagram.

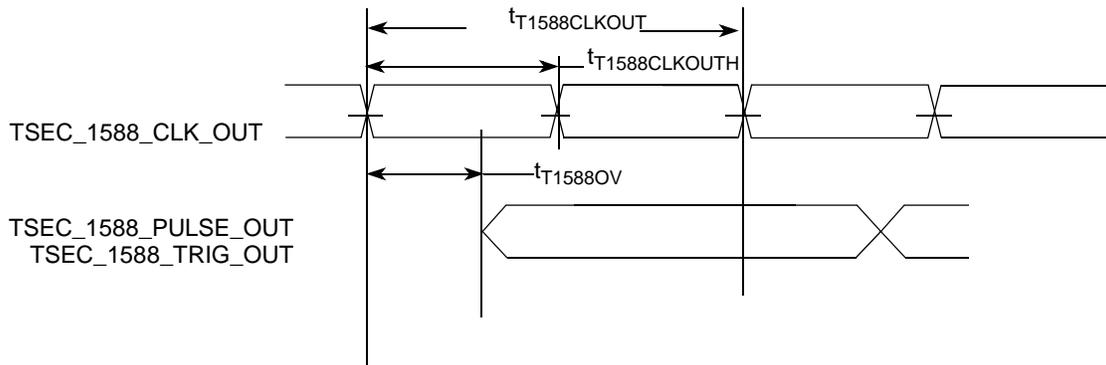


Figure 26. eTSEC IEEE 1588 Output AC Timing

¹ The output delay is count starting rising edge if $t_{T1588CLKOUT}$ is non-inverting. Otherwise, it is count starting falling edge.

Figure 27 shows the data and command input timing diagram.

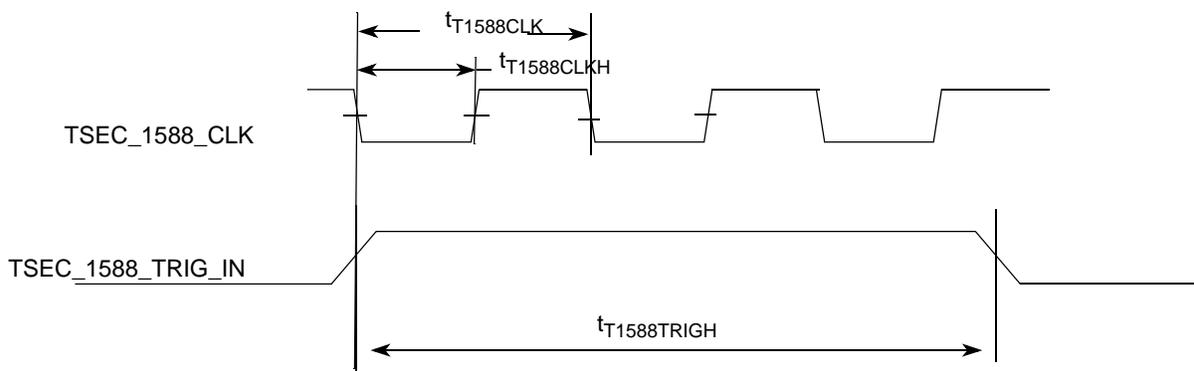


Figure 27. eTSEC IEEE 1588 Input AC timing

Table 42 provides the IEEE 1588 AC timing specifications.

Table 42. eTSEC IEEE 1588 AC Timing Specifications

At recommended operating conditions with LV_{DD}/TV_{DD} of $3.3\text{ V} \pm 5\%$ or $2.5\text{ V} \pm 5\%$

Parameter/Condition	Symbol	Min	Typ	Max	Unit	Note
TSEC_1588_CLK clock period	$t_{T1588CLK}$	3.3	—	$T_{TX_CLK}^{*9}$	ns	1
TSEC_1588_CLK duty cycle	$t_{T1588CLKH} / t_{T1588CLK}$	40	50	60	%	—
TSEC_1588_CLK peak-to-peak jitter	$t_{T1588CLKINJ}$	—	—	250	ps	—
Rise time eTSEC_1588_CLK (20%–80%)	$t_{T1588CLKINR}$	1.0	—	2.0	ns	—
Fall time eTSEC_1588_CLK (80%–20%)	$t_{T1588CLKINF}$	1.0	—	2.0	ns	—
TSEC_1588_CLK_OUT clock period	$t_{T1588CLKOUT}$	$2 * t_{T1588CLK}$	—	—	ns	—

10 Local Bus Controller (eLBC)

This section describes the DC and AC electrical specifications for the local bus interface of the MPC8572E.

10.1 Local Bus DC Electrical Characteristics

Table 46 provides the DC electrical characteristics for the local bus interface operating at $BV_{DD} = 3.3\text{ V}$ DC.

Table 46. Local Bus DC Electrical Characteristics (3.3 V DC)

Parameter	Symbol	Min	Max	Unit
Supply voltage 3.3V	BV_{DD}	3.13	3.47	V
High-level input voltage	V_{IH}	2	$BV_{DD} + 0.3$	V
Low-level input voltage	V_{IL}	-0.3	0.8	V
Input current ($BV_{IN}^1 = 0\text{ V}$ or $BV_{IN} = BV_{DD}$)	I_{IN}	—	± 5	μA
High-level output voltage ($BV_{DD} = \text{min}$, $I_{OH} = -2\text{ mA}$)	V_{OH}	$BV_{DD} - 0.2$	—	V
Low-level output voltage ($BV_{DD} = \text{min}$, $I_{OL} = 2\text{ mA}$)	V_{OL}	—	0.2	V

Note:

- Note that the symbol BV_{IN} , in this case, represents the BV_{IN} symbol referenced in Table 1.

Table 47 provides the DC electrical characteristics for the local bus interface operating at $BV_{DD} = 2.5\text{ V}$ DC.

Table 47. Local Bus DC Electrical Characteristics (2.5 V DC)

Parameter	Symbol	Min	Max	Unit
Supply voltage 2.5V	BV_{DD}	2.37	2.63	V
High-level input voltage	V_{IH}	1.70	$BV_{DD} + 0.3$	V
Low-level input voltage	V_{IL}	-0.3	0.7	V
Input current ($BV_{IN}^1 = 0\text{ V}$ or $BV_{IN} = BV_{DD}$)	I_{IH}	—	10	μA
	I_{IL}		-15	
High-level output voltage ($BV_{DD} = \text{min}$, $I_{OH} = -1\text{ mA}$)	V_{OH}	2.0	$BV_{DD} + 0.3$	V
Low-level output voltage ($BV_{DD} = \text{min}$, $I_{OL} = 1\text{ mA}$)	V_{OL}	$\text{GND} - 0.3$	0.4	V

Note:

- The symbol BV_{IN} , in this case, represents the BV_{IN} symbol referenced in Table 1.

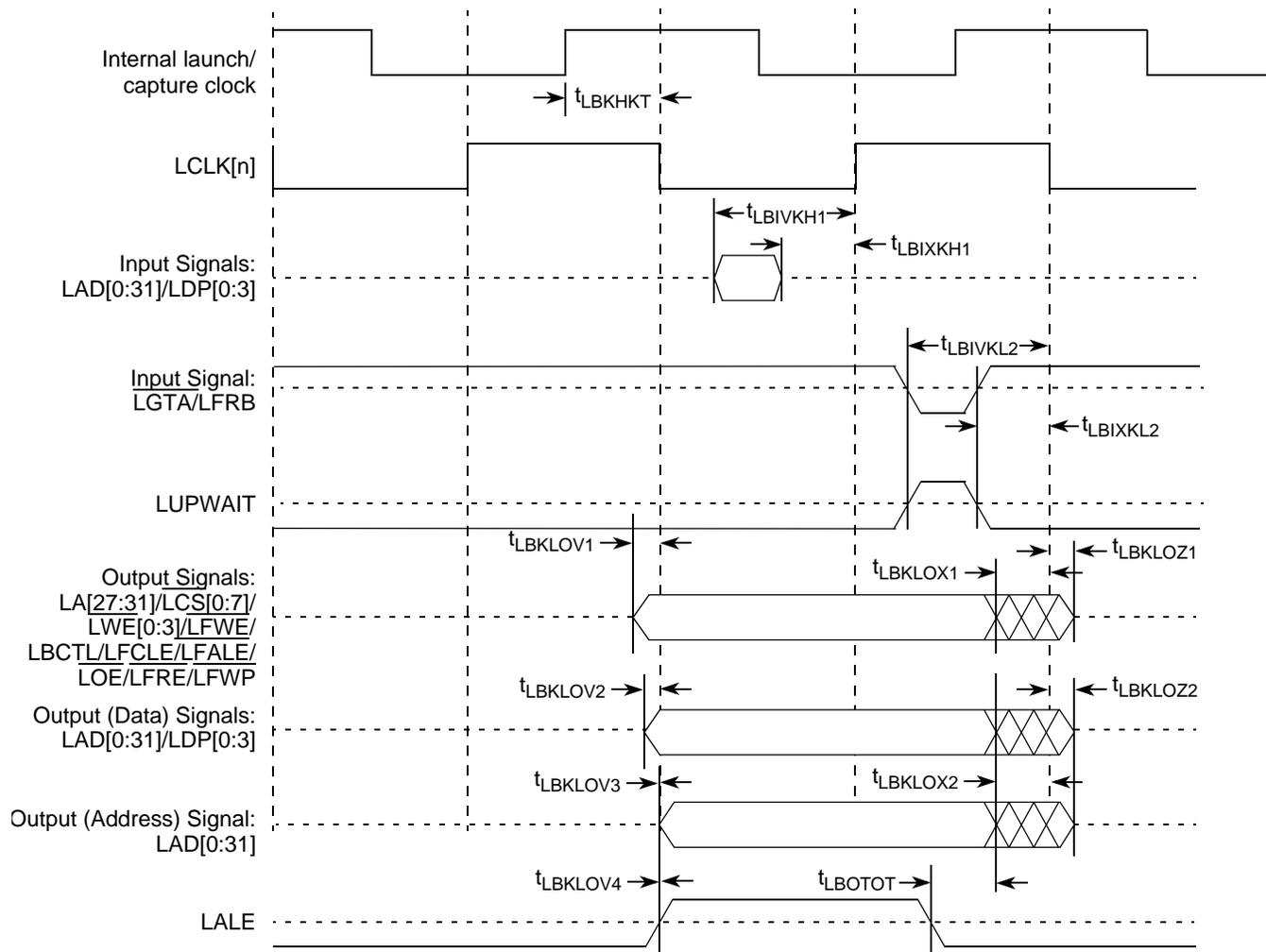


Figure 31. Local Bus Signals (PLL Bypass Mode)

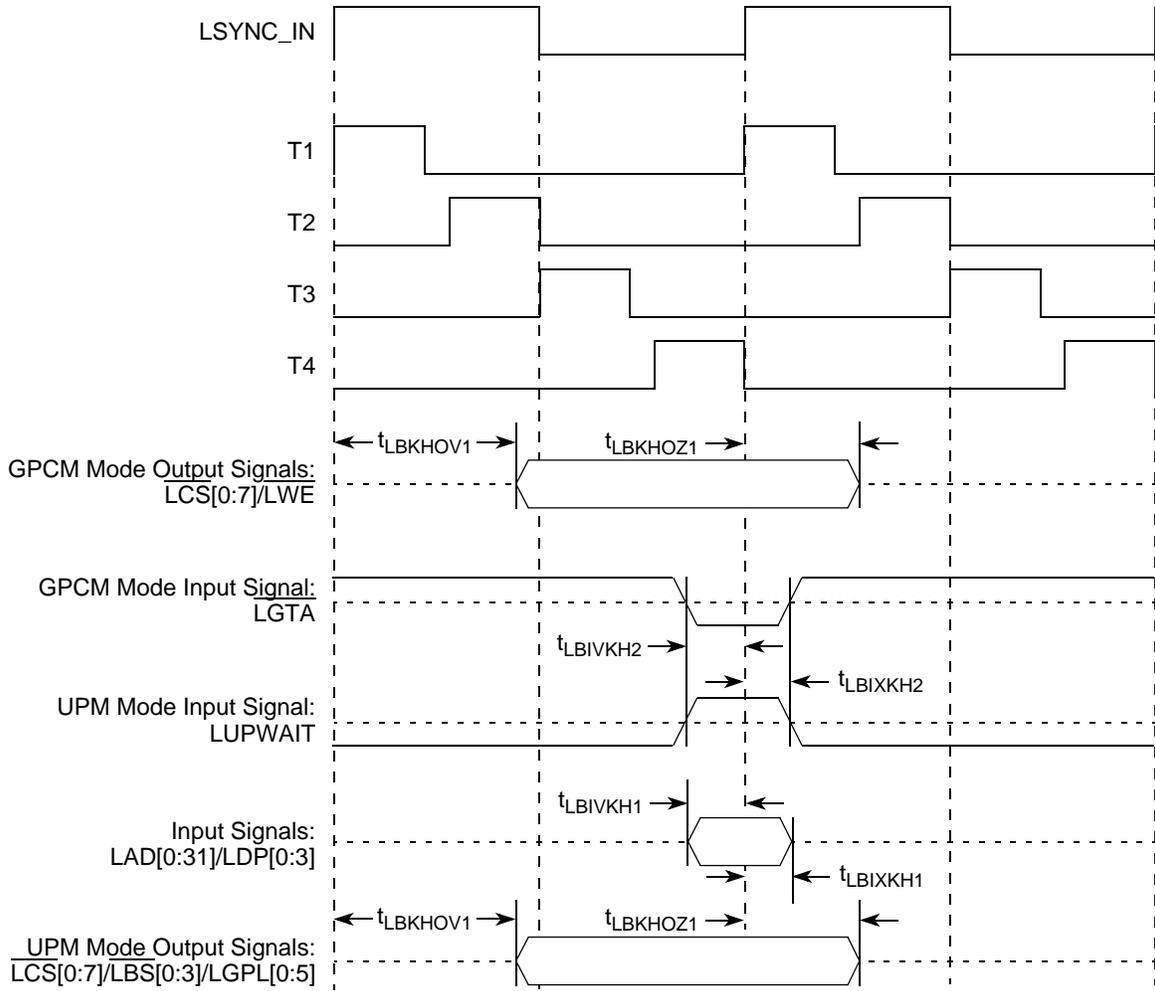


Figure 34. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 8 or 16 (PLL Enabled)

11 Programmable Interrupt Controller

In IRQ edge trigger mode, when an external interrupt signal is asserted (according to the programmed polarity), it must remain asserted for at least 3 system clocks (SYSCLK periods).

12 JTAG

This section describes the AC electrical specifications for the IEEE 1149.1 (JTAG) interface of the MPC8572E.

Table 53 provides the JTAG AC timing specifications as defined in Figure 37 through Figure 39.

Table 53. JTAG AC Timing Specifications (Independent of SYSCLK) ¹

At recommended operating conditions with OV_{DD} of $3.3\text{ V} \pm 5\%$.

Parameter	Symbol ²	Min	Max	Unit	Notes
JTAG external clock frequency of operation	f_{JTG}	0	33.3	MHz	—
JTAG external clock cycle time	t_{JTG}	30	—	ns	—
JTAG external clock pulse width measured at 1.4 V	t_{JKHKL}	15	—	ns	—
JTAG external clock rise and fall times	t_{JTGR} & t_{JTGF}	0	2	ns	6
\overline{TRST} assert time	t_{TRST}	25	—	ns	3
Input setup times:				ns	
Boundary-scan data	t_{JTDVKH}	4	—		4
TMS, TDI	t_{JTIVKH}	0	—		
Input hold times:				ns	
Boundary-scan data	t_{JTDXKH}	20	—		4
TMS, TDI	t_{JTIXKH}	25	—		
Valid times:				ns	
Boundary-scan data	t_{JTKLDV}	4	20		5
TDO	t_{JTKLOV}	4	25		
Output hold times:				ns	
Boundary-scan data	t_{JTKLDX}	30	—		5
TDO	t_{JTKLOX}	30	—		

Table 62. Differential Transmitter (TX) Output Specifications (continued)

Symbol	Parameter	Min	Nominal	Max	Units	Comments
$T_{\text{crosslink}}$	Crosslink Random Timeout	0	—	1	ms	This random timeout helps resolve conflicts in crosslink configuration by eventually resulting in only one Downstream and one Upstream Port. See Note 7.

Notes:

1. No test load is necessarily associated with this value.
2. Specified at the measurement point into a timing and voltage compliance test load as shown in [Figure 57](#) and measured over any 250 consecutive TX UIs. (Also refer to the transmitter compliance eye diagram shown in [Figure 55](#).)
3. A $T_{\text{TX-EYE}} = 0.70$ UI provides for a total sum of deterministic and random jitter budget of $T_{\text{TX-JITTER-MAX}} = 0.30$ UI for the Transmitter collected over any 250 consecutive TX UIs. The $T_{\text{TX-EYE-MEDIAN-to-MAX-JITTER}}$ median is less than half of the total TX jitter budget collected over any 250 consecutive TX UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value.
4. The Transmitter input impedance shall result in a differential return loss greater than or equal to 12 dB and a common mode return loss greater than or equal to 6 dB over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements is 50 ohms to ground for both the D+ and D- line (that is, as measured by a Vector Network Analyzer with 50 ohm probes—see [Figure 57](#)). Note that the series capacitors C_{TX} is optional for the return loss measurement.
5. Measured between 20-80% at transmitter package pins into a test load as shown in [Figure 57](#) for both $V_{\text{TX-D+}}$ and $V_{\text{TX-D-}}$.
6. See Section 4.3.1.8 of the PCI Express Base Specifications Rev 1.0a.
7. See Section 4.2.6.3 of the PCI Express Base Specifications Rev 1.0a.
8. MPC8572E SerDes transmitter does not have C_{TX} built-in. An external AC Coupling capacitor is required.

16.4.2 Transmitter Compliance Eye Diagrams

The TX eye diagram in [Figure 55](#) is specified using the passive compliance/test measurement load (see [Figure 57](#)) in place of any real PCI Express interconnect + RX component.

There are two eye diagrams that must be met for the transmitter. Both eye diagrams must be aligned in time using the jitter median to locate the center of the eye diagram. The different eye diagrams differ in voltage depending whether it is a transition bit or a de-emphasized bit. The exact reduced voltage level of the de-emphasized bit is always relative to the transition bit.

The eye diagram must be valid for any 250 consecutive UIs.

A recovered TX UI is calculated over 3500 consecutive unit intervals of sample data. The eye diagram is created using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the TX UI.

NOTE

It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function (that is, least squares and median deviation fits).

16.5 Receiver Compliance Eye Diagrams

The RX eye diagram in [Figure 56](#) is specified using the passive compliance/test measurement load (see [Figure 57](#)) in place of any real PCI Express RX component.

Note: In general, the minimum Receiver eye diagram measured with the compliance/test measurement load (see [Figure 57](#)) is larger than the minimum Receiver eye diagram measured over a range of systems at the input Receiver of any real PCI Express component. The degraded eye diagram at the input Receiver is due to traces internal to the package as well as silicon parasitic characteristics which cause the real PCI Express component to vary in impedance from the compliance/test measurement load. The input Receiver eye diagram is implementation specific and is not specified. RX component designer should provide additional margin to adequately compensate for the degraded minimum Receiver eye diagram (shown in [Figure 56](#)) expected at the input Receiver based on some adequate combination of system simulations and the Return Loss measured looking into the RX package and silicon. The RX eye diagram must be aligned in time using the jitter median to locate the center of the eye diagram.

The eye diagram must be valid for any 250 consecutive UIs.

A recovered TX UI is calculated over 3500 consecutive unit intervals of sample data. The eye diagram is created using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the TX UI.

NOTE

The reference impedance for return loss measurements is 50. to ground for both the D+ and D- line (that is, as measured by a Vector Network Analyzer with 50. probes—see [Figure 57](#)). Note that the series capacitors, CTX, are optional for the return loss measurement.

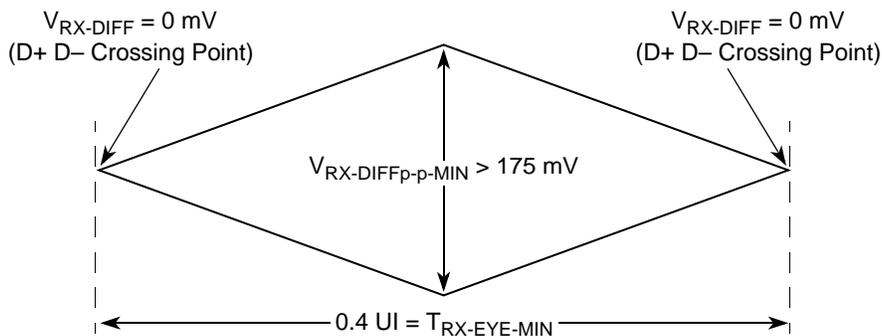


Figure 56. Minimum Receiver Eye Timing and Voltage Compliance Specification

17.5 Transmitter Specifications

LP-Serial transmitter electrical and timing specifications are stated in the text and tables of this section.

The differential return loss, S11, of the transmitter in each case shall be better than

- -10 dB for $(\text{Baud Frequency})/10 < \text{Freq}(f) < 625$ MHz, and
- -10 dB + $10\log(f/625 \text{ MHz})$ dB for $625 \text{ MHz} \leq \text{Freq}(f) \leq \text{Baud Frequency}$

The reference impedance for the differential return loss measurements is 100Ω resistive. Differential return loss includes contributions from on-chip circuitry, chip packaging and any off-chip components related to the driver. The output impedance requirement applies to all valid output levels.

It is recommended that the 20%-80% rise/fall time of the transmitter, as measured at the transmitter output, in each case have a minimum value 60 ps.

It is recommended that the timing skew at the output of an LP-Serial transmitter between the two signals that comprise a differential pair not exceed 25 ps at 1.25 GB, 20 ps at 2.50 GB and 15 ps at 3.125 GB.

Table 65. Short Run Transmitter AC Timing Specifications—1.25 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Output Voltage,	V_O	-0.40	2.30	Volts	Voltage relative to COMMON of either signal comprising a differential pair
Differential Output Voltage	V_{DIFFPP}	500	1000	mV p-p	—
Deterministic Jitter	J_D	—	0.17	UI p-p	—
Total Jitter	J_T	—	0.35	UI p-p	—
Multiple output skew	S_{MO}	—	1000	ps	Skew at the transmitter output between lanes of a multilane link
Unit Interval	UI	800	800	ps	+/- 100 ppm

Table 66. Short Run Transmitter AC Timing Specifications—2.5 GBaud

Characteristic	Symbol	Range		Unit	Notes
		Min	Max		
Output Voltage,	V_O	-0.40	2.30	Volts	Voltage relative to COMMON of either signal comprising a differential pair
Differential Output Voltage	V_{DIFFPP}	500	1000	mV p-p	—
Deterministic Jitter	J_D	—	0.17	UI p-p	—
Total Jitter	J_T	—	0.35	UI p-p	—

Table 76. MPC8572E Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
$\overline{\text{IRQ_OUT}}$	Interrupt Output	U24	O	OV _{DD}	2, 4
1588					
TSEC_1588_CLK	Clock In	AM22	I	LV _{DD}	—
TSEC_1588_TRIG_IN	Trigger In	AM23	I	LV _{DD}	—
TSEC_1588_TRIG_OUT	Trigger Out	AA23	O	LV _{DD}	5, 9
TSEC_1588_CLK_OUT	Clock Out	AC23	O	LV _{DD}	5, 9
TSEC_1588_PULSE_OUT1	Pulse Out1	AA22	O	LV _{DD}	5, 9
TSEC_1588_PULSE_OUT2	Pulse Out2	AB23	O	LV _{DD}	5, 9
Ethernet Management Interface 1					
EC1_MDC	Management Data Clock	AL30	O	LV _{DD}	5, 9
EC1_MDIO	Management Data In/Out	AM25	I/O	LV _{DD}	—
Ethernet Management Interface 3					
EC3_MDC	Management Data Clock	AF19	O	TV _{DD}	5, 9
EC3_MDIO	Management Data In/Out	AF18	I/O	TV _{DD}	—
Ethernet Management Interface 5					
EC5_MDC	Management Data Clock	AF14	O	TV _{DD}	21
EC5_MDIO	Management Data In/Out	AF15	I/O	TV _{DD}	—
Gigabit Ethernet Reference Clock					
EC_GTX_CLK125	Reference Clock	AM24	I	LV _{DD}	32
Three-Speed Ethernet Controller 1					
TSEC1_RXD[7:0]/FIFO1_RXD[7:0]	Receive Data	AM28, AL28, AM26, AK23, AM27, AK26, AL29, AM30	I	LV _{DD}	1
TSEC1_TXD[7:0]/FIFO1_TXD[7:0]	Transmit Data	AC20, AD20, AE22, AB22, AC22, AD21, AB21, AE21	O	LV _{DD}	1, 5, 9
TSEC1_COL/FIFO1_TX_FC	Collision Detect/Flow Control	AJ23	I	LV _{DD}	1
TSEC1_CRS/FIFO1_RX_FC	Carrier Sense/Flow Control	AM31	I/O	LV _{DD}	1, 16
TSEC1_GTX_CLK	Transmit Clock Out	AK27	O	LV _{DD}	
TSEC1_RX_CLK/FIFO1_RX_C LK	Receive Clock	AL25	I	LV _{DD}	1

Table 82. DDR Clock Ratio (continued)

Binary Value of TSEC_1588_CLK_OUT, TSEC_1588_PULSE_OUT1, TSEC_1588_PULSE_OUT2 Signals	DDR:DDRCLK Ratio
101	12:1
110	14:1
111	Synchronous mode

19.5 Frequency Options

19.5.1 Platform to Sysclk Frequency Options

Table 83 shows the expected frequency values for the platform frequency when using the specified CCB clock to SYSCLK ratio.

Table 83. Frequency Options for Platform Frequency

CCB to SYSCLK Ratio	SYSCLK (MHz)							
	33.33	41.66	50	66.66	83	100	111	133.33
	Platform /CCB Frequency (MHz)							
4						400	444	533
5					415	500	555	
6				400	498	600		
8			400	533				
10		417	500					
12	400	500	600					

19.5.2 Minimum Platform Frequency Requirements for High-Speed Interfaces

Section 4.4.3.6, “I/O Port Selection,” in the *MPC8572E PowerQUICC III Integrated Host Processor Family Reference Manual* describes various high-speed interface configuration options. Note that the CCB clock frequency must be considered for proper operation of such interfaces as described below.

For proper PCI Express operation, the CCB clock frequency must be greater than or equal to:

$$\frac{527 \text{ MHz} \times (\text{PCI Express link width})}{8}$$

See Section 21.1.3.2, “Link Width,” in the *MPC8572E PowerQUICC III Integrated Host Processor Family Reference Manual* for PCI Express interface width details. Note that the “PCI Express link width”

Thermal

$$V_f > 0.40 \text{ V}$$

$$V_f < 0.90 \text{ V}$$

Operating range 2–300 μA

Diode leakage < 10 nA @ 125°C

An approximate value of the ideality may be obtained by calibrating the device near the expected operating temperature.

Ideality factor is defined as the deviation from the ideal diode equation:

$$I_{fw} = I_s \left[e^{\frac{qV_f}{nKT}} - 1 \right]$$

Another useful equation is:

$$V_H - V_L = n \frac{KT}{q} \left[\ln \frac{I_H}{I_L} \right]$$

Where:

I_{fw} = Forward current

I_s = Saturation current

V_d = Voltage at diode

V_f = Voltage forward biased

V_H = Diode voltage while I_H is flowing

V_L = Diode voltage while I_L is flowing

I_H = Larger diode bias current

I_L = Smaller diode bias current

q = Charge of electron ($1.6 \times 10^{-19} \text{ C}$)

n = Ideality factor (normally 1.0)

K = Boltzman's constant ($1.38 \times 10^{-23} \text{ Joules/K}$)

T = Temperature (Kelvins)

The ratio of I_H to I_L is usually selected to be 10:1. The above simplifies to the following:

$$V_H - V_L = 1.986 \times 10^{-4} \times nT$$

Solving for T , the equation becomes:

$$nT = \frac{V_H - V_L}{1.986 \times 10^{-4}}$$

21.6 Pull-Up and Pull-Down Resistor Requirements

The MPC8572E requires weak pull-up resistors (2–10 kΩ is recommended) on open drain type pins including I²C pins and MPIC interrupt pins.

Correct operation of the JTAG interface requires configuration of a group of system control pins as demonstrated in Figure 66. Care must be taken to ensure that these pins are maintained at a valid deasserted state under normal operating conditions as most have asynchronous behavior and spurious assertion gives unpredictable results.

The following pins must NOT be pulled down during power-on reset: $\overline{\text{DMA_DACK}}[0:1]$, EC5_MDC, HRESET_REQ, TRIG_OUT/READY_P0/QUIESCE, MSRCID[2:4], MDVAL, and ASLEEP. The $\overline{\text{TEST_SEL}}$ pin must be set to a proper state during POR configuration. For more details, refer to the pinlist table of the individual device.

21.7 Output Buffer DC Impedance

The MPC8572E drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I²C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $\text{OV}_{\text{DD}}/2$ (see Figure 64). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $\text{OV}_{\text{DD}}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

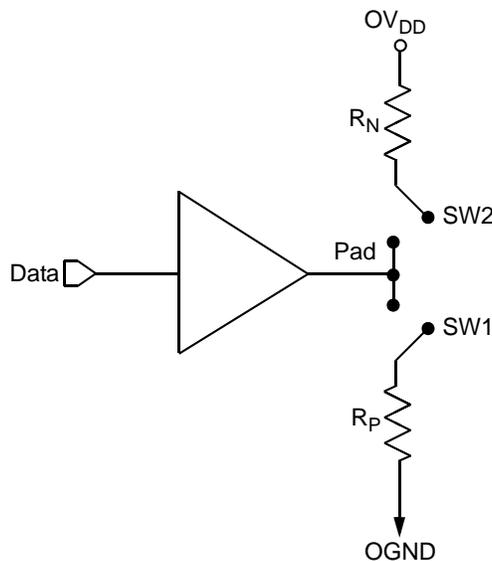


Figure 64. Driver Impedance Measurement

Figure 66. JTAG Interface Connection

21.10 Guidelines for High-Speed Interface Termination

21.10.1 SerDes 1 Interface Entirely Unused

If the high-speed SerDes 1 interface is not used at all, the unused pin should be terminated as described in this section.

The following pins must be left unconnected (float):

- SD1_TX[7:0]
- $\overline{\text{SD1_TX}}$ [7:0]
- Reserved pins C24, C25, H26, H27

The following pins must be connected to XGND_SRDS1:

- SD1_RX[7:0]
- $\overline{\text{SD1_RX}}$ [7:0]
- SD1_REF_CLK
- $\overline{\text{SD1_REF_CLK}}$

Pins K32 and C29 must be tied to $\text{XV}_{\text{DD_SRDS1}}$. Pins K31 and C30 must be tied to XGND_SRDS1 through a 300- Ω resistor.

The POR configuration pin `cfg_srds1_en` on TSEC2_TXD[5] can be used to power down SerDes 1 block for power saving. Note that both SVDD_SRDS1 and XVDD_SRDS1 must remain powered.

21.10.2 SerDes 1 Interface Partly Unused

If only part of the high speed SerDes 1 interface pins are used, the remaining high-speed serial I/O pins should be terminated as described in this section.

The following pins must be left unconnected (float) if not used:

- SD1_TX[7:0]
- $\overline{\text{SD1_TX}}$ [7:0]
- Reserved pins: C24, C25, H26, H27

The following pins must be connected to XGND_SRDS1 if not used:

- SD1_RX[7:0]
- $\overline{\text{SD1_RX}}$ [7:0]

Pins K32 and C29 must be tied to $\text{XV}_{\text{DD_SRDS1}}$. Pins K31 and C30 must be tied to XGND_SRDS1 through a 300- Ω resistor.