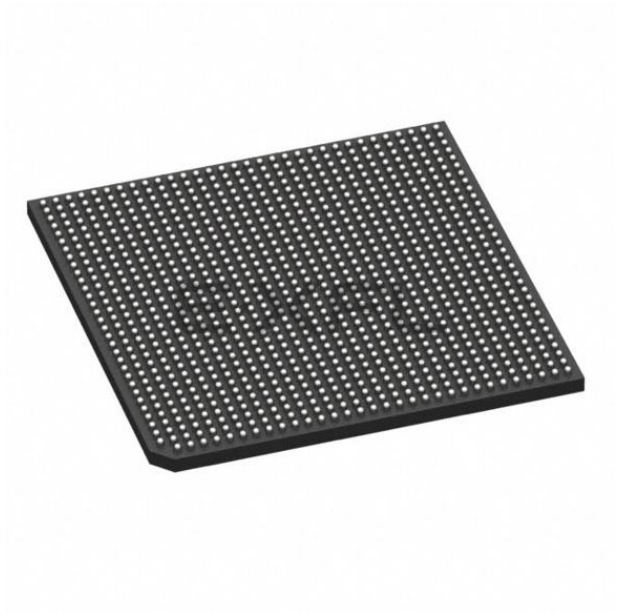

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Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	2 Core, 32-Bit
Speed	1.067GHz
Co-Processors/DSP	Signal Processing; SPE, Security; SEC
RAM Controllers	DDR2, DDR3
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (4)
SATA	-
USB	-
Voltage - I/O	1.5V, 1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	1023-BFBGA, FCBGA
Supplier Device Package	1023-FCBGA (33x33)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8572evtarlb

Table 14 provides the current draw characteristics for $MV_{REF}n$.

Table 14. Current Draw Characteristics for MV_{REF} *n*

Parameter / Condition		Symbol	Min	Max	Unit	Note
Current draw for $MV_{REF}n$	DDR2 SDRAM	$I_{MVREF}n$	—	1500	µA	1
	DDR3 SDRAM			1250		

1. The voltage regulator for $MV_{REF}n$ must be able to supply up to 1500 µA or 1250 uA current for DDR2 or DDR3, respectively.

6.2 DDR2 and DDR3 SDRAM Interface AC Electrical Characteristics

This section provides the AC electrical characteristics for the DDR SDRAM controller interface. The DDR controller supports both DDR2 and DDR3 memories. Note that although the minimum data rate for most off-the-shelf DDR3 DIMMs available is 800 MHz, JEDEC specification does allow the DDR3 to run at the data rate as low as 606 MHz. Unless otherwise specified, the AC timing specifications described in this section for DDR3 is applicable for data rate between 606 MHz and 800 MHz, as long as the DC and AC specifications of the DDR3 memory to be used are compliant to both JEDEC specifications as well as the specifications and requirements described in this MPC8572E hardware specifications document.

6.2.1 DDR2 and DDR3 SDRAM Interface Input AC Timing Specifications

Table 15, Table 16, and Table 17 provide the input AC timing specifications for the DDR controller when interfacing to DDR2 and DDR3 SDRAM.

Table 15. DDR2 SDRAM Interface Input AC Timing Specifications for 1.8-V Interface

At recommended operating conditions with GV_{DD} of 1.8 V ± 5%

Parameter		Symbol	Min	Max	Unit	Notes
AC input low voltage	>=667 MHz	V_{ILAC}	—	$MV_{REF}n$ – 0.20	V	—
	<= 533 MHz		—	$MV_{REF}n$ – 0.25		
AC input high voltage	>=667 MHz	V_{IHAC}	$MV_{REF}n$ + 0.20	—	V	—
—	<= 533 MHz		$MV_{REF}n$ + 0.25	—		

Table 16. DDR3 SDRAM Interface Input AC Timing Specifications for 1.5-V Interface

At recommended operating conditions with GV_{DD} of 1.5 V ± 5%. DDR3 data rate is between 606 MHz and 800 MHz.

Parameter	Symbol	Min	Max	Unit	Notes
AC input low voltage	V_{ILAC}	—	$MV_{REF}n$ – 0.175	V	—
AC input high voltage	V_{IHAC}	$MV_{REF}n$ + 0.175	—	V	—

Figure 17 shows the TBI receive the timing diagram.

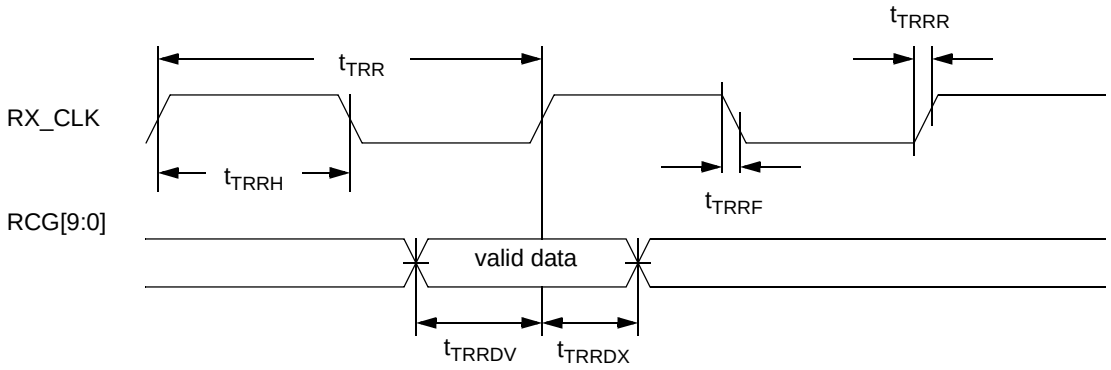

Figure 17. TBI Single-Clock Mode Receive AC Timing Diagram

8.2.6 RGMII and RTBI AC Timing Specifications

Table 34 presents the RGMII and RTBI AC timing specifications.

Table 34. RGMII and RTBI AC Timing Specifications

At recommended operating conditions with LV_{DD}/TV_{DD} of 2.5 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)	t_{SKRGT}	–500	0	500	ps
Data to clock input skew (at receiver) [2]	t_{SKRGT}	1.0	—	2.8	ns
Clock period [3]	t_{RGT}	7.2	8.0	8.8	ns
Duty cycle for 10BASE-T and 100BASE-TX [3, 4]	t_{RGTH}/t_{RGT}	40	50	60	%
Rise time (20%–80%)	t_{RGTR}	—	—	0.75	ns
Fall time (20%–80%)	t_{RGTF}	—	—	0.75	ns

Notes:

1. Note that, in general, the clock reference symbol representation for this section is based on the symbols RGT to represent RGMII and RTBI timing. For example, the subscript of t_{RGT} represents the TBI (T) receive (RX) clock. Note also that the notation for rise (R) and fall (F) times follows the clock symbol that is being represented. For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (RGT).
2. This implies that PC board design requires clocks to be routed such that an additional trace delay of greater than 1.5 ns will be added to the associated clock signal.
3. For 10 and 100 Mbps, t_{RGT} scales to 400 ns ± 40 ns and 40 ns ± 4 ns, respectively.
4. Duty cycle may be stretched/shrunk during speed changes or while transitioning to a received packet's clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned between.

8.3.3 SGMII Transmitter and Receiver DC Electrical Characteristics

Table 38 and Table 39 describe the SGMII SerDes transmitter and receiver AC-Coupled DC electrical characteristics. Transmitter DC characteristics are measured at the transmitter outputs (SD2_TX[n] and $\overline{\text{SD2\_TX[n]}}$) as depicted in Figure 23.

Table 38. SGMII DC Transmitter Electrical Characteristics

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Supply Voltage	XV_{DD_SRDS2}	1.045	1.1	1.155	V	—
Output high voltage	VOH	—	—	$XV_{DD\_SRDS2-Typ}/2 + \lvert\overline{V_{OD}}\rvert_{-max}/2$	mV	1
Output low voltage	VOL	$XV_{DD\_SRDS2-Typ}/2 - \lvert V_{OD}\rvert_{-max}/2$	—	—	mV	1
Output ringing	V_{RING}	—	—	10	%	—
Output differential voltage[2, 3, 5]	$\lvert V_{OD}\rvert$	359	550	791	mV	Equalization setting: 1.0x
		329	505	725		Equalization setting: 1.09x
		299	458	659		Equalization setting: 1.2x
		270	414	594		Equalization setting: 1.33x
		239	367	527		Equalization setting: 1.5x
		210	322	462		Equalization setting: 1.71x
		180	275	395		Equalization setting: 2.0x
Output offset voltage	V_{OS}	473	550	628	mV	1, 4
Output impedance (single-ended)	R_O	40	—	60	Ω	—
Mismatch in a pair	ΔR_O	—	—	10	%	—
Change in V_{OD} between “0” and “1”	$\Delta \lvert V_{OD}\rvert$	—	—	25	mV	—

Table 38. SGMII DC Transmitter Electrical Characteristics (continued)

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Change in V_{OS} between "0" and "1"	ΔV_{OS}	—	—	25	mV	—
Output current on short to GND	I_{SA}, I_{SB}	—	—	40	mA	—

Note:

1. This will not align to DC-coupled SGMII. $XV_{DD\_SRDS2\text{-}Typ}$=1.1 V.
2. $|V_{OD}| = |V_{SD2\_TXn} - V_{\overline{SD2\_TXn}}|$. $|V_{OD}|$ is also referred as output differential peak voltage. $V_{TX\text{-}DIFFp\text{-}p} = 2*|V_{OD}|$.
3. The $|V_{OD}|$ value shown in the table assumes the following transmit equalization setting in the XMITEQ**AB** (for SerDes 2 lanes A & B) or XMITEQ**EF** (for SerDes 2 lanes E & E) bit field of MPC8572E's SerDes 2 Control Register:
   - •The MSbit (bit 0) of the above bit field is set to zero (selecting the full $V_{DD\text{-}DIFF\text{-}p\text{-}p}$ amplitude - power up default);
   - •The LSbits (bit [1:3]) of the above bit field is set based on the equalization setting shown in table.
4. V_{OS} is also referred to as output common mode voltage.
   - 5.The $|V_{OD}|$ value shown in the Typ column is based on the condition of $XV_{DD\_SRDS2\text{-}Typ}$=1.1V, no common mode offset variation (V_{OS} =550mV), SerDes2 transmitter is terminated with 100-Ω differential load between SD2_TX[n] and $\overline{SD2\_TX}$[n].

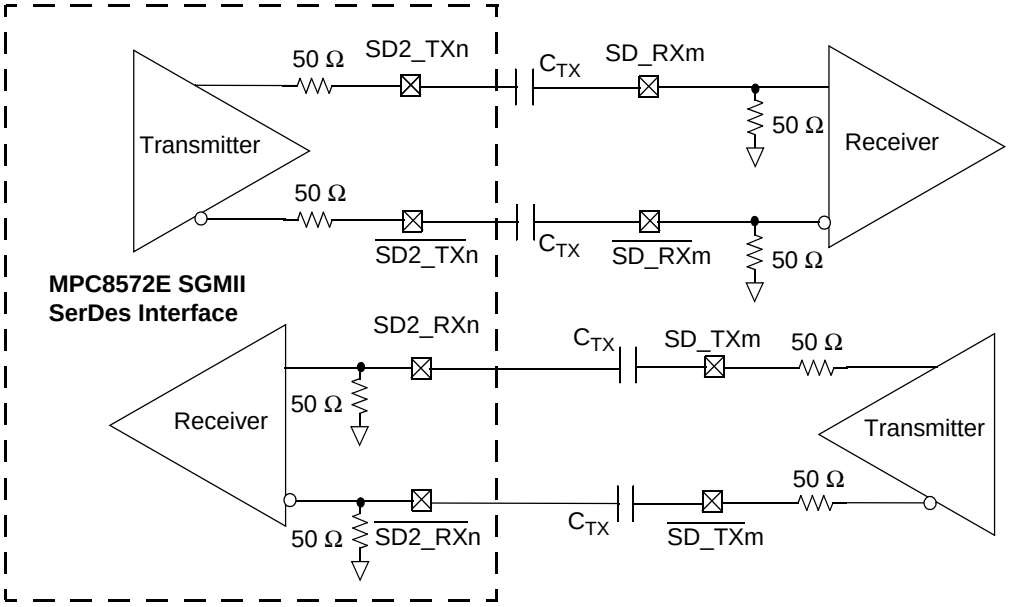

Figure 22. 4-Wire AC-Coupled SGMII Serial Link Connection Example

Table 49. Local Bus General Timing Parameters (BV_{DD} = 3.3 V DC)—PLL Enabled (continued)

At recommended operating conditions with BV_{DD} of 3.3 V ± 5%. (continued)

Parameter	Symbol [1]	Min	Max	Unit	Notes
Local bus clock to LALE assertion	$t_{LBKHOV4}$	—	2.3	ns	3
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKHOX1}$	0.7	—	ns	3
Output hold from local bus clock for LAD/LDP	$t_{LBKHOX2}$	0.7	—	ns	3
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$t_{LBKHOZ1}$	—	2.5	ns	5
Local bus clock to output high impedance for LAD/LDP	$t_{LBKHOZ2}$	—	2.5	ns	5

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(First two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(First two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one(1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to LSYNC_IN for PLL enabled and internal local bus clock for PLL bypass mode.
3. All signals are measured from $BV_{DD}/2$ of the rising edge of LSYNC_IN for PLL enabled or internal local bus clock for PLL bypass mode to $0.4 \times BV_{DD}$ of the signal in question for 3.3-V signaling levels.
4. Input timings are measured at the pin.
5. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
6. t_{LBOTOT} is a measurement of the minimum time between the negation of LALE and any change in LAD. t_{LBOTOT} is programmed with the LBCR[AHD] parameter.
7. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[n]. Skew measured between complementary signals at $BV_{DD}/2$.
8. Guaranteed by design.

Table 50 describes the general timing parameters of the local bus interface at BV_{DD} = 2.5 V DC.

Table 50. Local Bus General Timing Parameters (BV_{DD} = 2.5 V DC)—PLL Enabled

At recommended operating conditions with BV_{DD} of 2.5 V ± 5%

Parameter	Symbol [1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	6.67	12	ns	2
Local bus duty cycle	t_{LBKH}/t_{LBK}	43	57	%	—
LCLK[n] skew to LCLK[m] or LSYNC_OUT	$t_{LBKSKEW}$	—	150	ps	7, 8
Input setup to local bus clock (except $\overline{\text{LGTA}}$/LUPWAIT)	$t_{LBIVKH1}$	1.9	—	ns	3, 4
$\overline{\text{LGTA}}$/LUPWAIT input setup to local bus clock	$t_{LBIVKH2}$	1.8	—	ns	3, 4
Input hold from local bus clock (except $\overline{\text{LGTA}}$/LUPWAIT)	$t_{LBIXKH1}$	1.1	—	ns	3, 4
$\overline{\text{LGTA}}$/LUPWAIT input hold from local bus clock	$t_{LBIXKH2}$	1.1	—	ns	3, 4
LALE output negation to high impedance for LAD/LDP (LATCH hold time)	t_{LBOTOT}	1.5	—	ns	6

Table 50. Local Bus General Timing Parameters (BV_{DD} = 2.5 V DC)—PLL Enabled (continued)

At recommended operating conditions with BV_{DD} of 2.5 V ± 5% (continued)

Parameter	Symbol [1]	Min	Max	Unit	Notes
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKHOV1}$	—	2.4	ns	—
Local bus clock to data valid for LAD/LDP	$t_{LBKHOV2}$	—	2.5	ns	3
Local bus clock to address valid for LAD	$t_{LBKHOV3}$	—	2.4	ns	3
Local bus clock to LALE assertion	$t_{LBKHOV4}$	—	2.4	ns	3
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKHOX1}$	0.8	—	ns	3
Output hold from local bus clock for LAD/LDP	$t_{LBKHOX2}$	0.8	—	ns	3
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$t_{LBKHOZ1}$	—	2.6	ns	5
Local bus clock to output high impedance for LAD/LDP	$t_{LBKHOZ2}$	—	2.6	ns	5

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(First two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(First two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one(1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to LSYNC_IN for PLL enabled and internal local bus clock for PLL bypass mode.
3. All signals are measured from $BV_{DD}/2$ of the rising edge of LSYNC_IN for PLL enabled or internal local bus clock for PLL bypass mode to $0.4 \times BV_{DD}$ of the signal in question for 2.5-V signaling levels.
4. Input timings are measured at the pin.
5. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
6. t_{LBOTOT} is a measurement of the minimum time between the negation of LALE and any change in LAD. t_{LBOTOT} is programmed with the LBCR[AHD] parameter.
7. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[n]. Skew measured between complementary signals at $BV_{DD}/2$.
8. Guaranteed by design.

Table 51 describes the general timing parameters of the local bus interface at BV_{DD} = 1.8 V DC

Table 51. Local Bus General Timing Parameters (BV_{DD} = 1.8 V DC)—PLL Enabled

At recommended operating conditions with BV_{DD} of 1.8 V ± 5%

Parameter	Symbol [1]	Min	Max	Unit	Notes
Local bus cycle time	t_{LBK}	6.67	12	ns	2
Local bus duty cycle	t_{LBKH}/t_{LBK}	43	57	%	—
LCLK[n] skew to LCLK[m] or LSYNC_OUT	$t_{LBKSKEW}$	—	150	ps	7, 8
Input setup to local bus clock (except $\overline{\text{LGTA}}$/LUPWAIT)	$t_{LBIVKH1}$	2.4	—	ns	3, 4
$\overline{\text{LGTA}}$/LUPWAIT input setup to local bus clock	$t_{LBIVKH2}$	1.9	—	ns	3, 4
Input hold from local bus clock (except $\overline{\text{LGTA}}$/LUPWAIT)	$t_{LBIXKH1}$	1.1	—	ns	3, 4

Table 52. Local Bus General Timing Parameters—PLL Bypassed (continued)

At recommended operating conditions with BV_{DD} of 3.3 V ± 5%

Parameter	Symbol [1]	Min	Max	Unit	Notes
$\overline{\text{LGTA}}$/LUPWAIT input hold from local bus clock	$t_{LBIXKL2}$	-1.3	—	ns	4, 5
LALE output negation to high impedance for LAD/LDP (LATCH hold time)	t_{LBOTOT}	1.5	—	ns	6
Local bus clock to output valid (except LAD/LDP and LALE)	$t_{LBKLOV1}$	—	–0.3	ns	
Local bus clock to data valid for LAD/LDP	$t_{LBKLOV2}$	—	–0.1	ns	4
Local bus clock to address valid for LAD	$t_{LBKLOV3}$	—	0.0	ns	4
Local bus clock to LALE assertion	$t_{LBKLOV4}$	—	0.0	ns	4
Output hold from local bus clock (except LAD/LDP and LALE)	$t_{LBKLOX1}$	–3.3	—	ns	4
Output hold from local bus clock for LAD/LDP	$t_{LBKLOX2}$	–3.3	—	ns	4
Local bus clock to output high Impedance (except LAD/LDP and LALE)	$t_{LBKLOZ1}$	—	0.2	ns	7
Local bus clock to output high impedance for LAD/LDP	$t_{LBKLOZ2}$	—	0.2	ns	7

Notes:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(First two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(First two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one(1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to local bus clock for PLL bypass mode. Timings may be negative with respect to the local bus clock because the actual launch and capture of signals is done with the internal launch/capture clock, which precedes LCLK by t_{LBKHKT}.
3. Maximum possible clock skew between a clock LCLK[m] and a relative clock LCLK[n]. Skew measured between complementary signals at $BV_{DD}/2$.
4. All signals are measured from BVDD/2 of the rising edge of local bus clock for PLL bypass mode to 0.4 x BVDD of the signal in question for 3.3-V signaling levels.
5. Input timings are measured at the pin.
6. t_{LBOTOT} is a measurement of the minimum time between the negation of LALE and any change in LAD. t_{LBOTOT} is programmed with the LBCR[AHD] parameter.
7. For purposes of active/float timing measurements, the Hi-Z or off state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.

NOTE

In PLL bypass mode, LCLK[n] is the inverted version of the internal clock with the delay of t_{LBKHKT}. In this mode, signals are launched at the rising edge of the internal clock and are captured at the falling edge of the internal clock with the exception of $\overline{\text{LGTA}}$/LUPWAIT (which is captured on the rising edge of the internal clock).

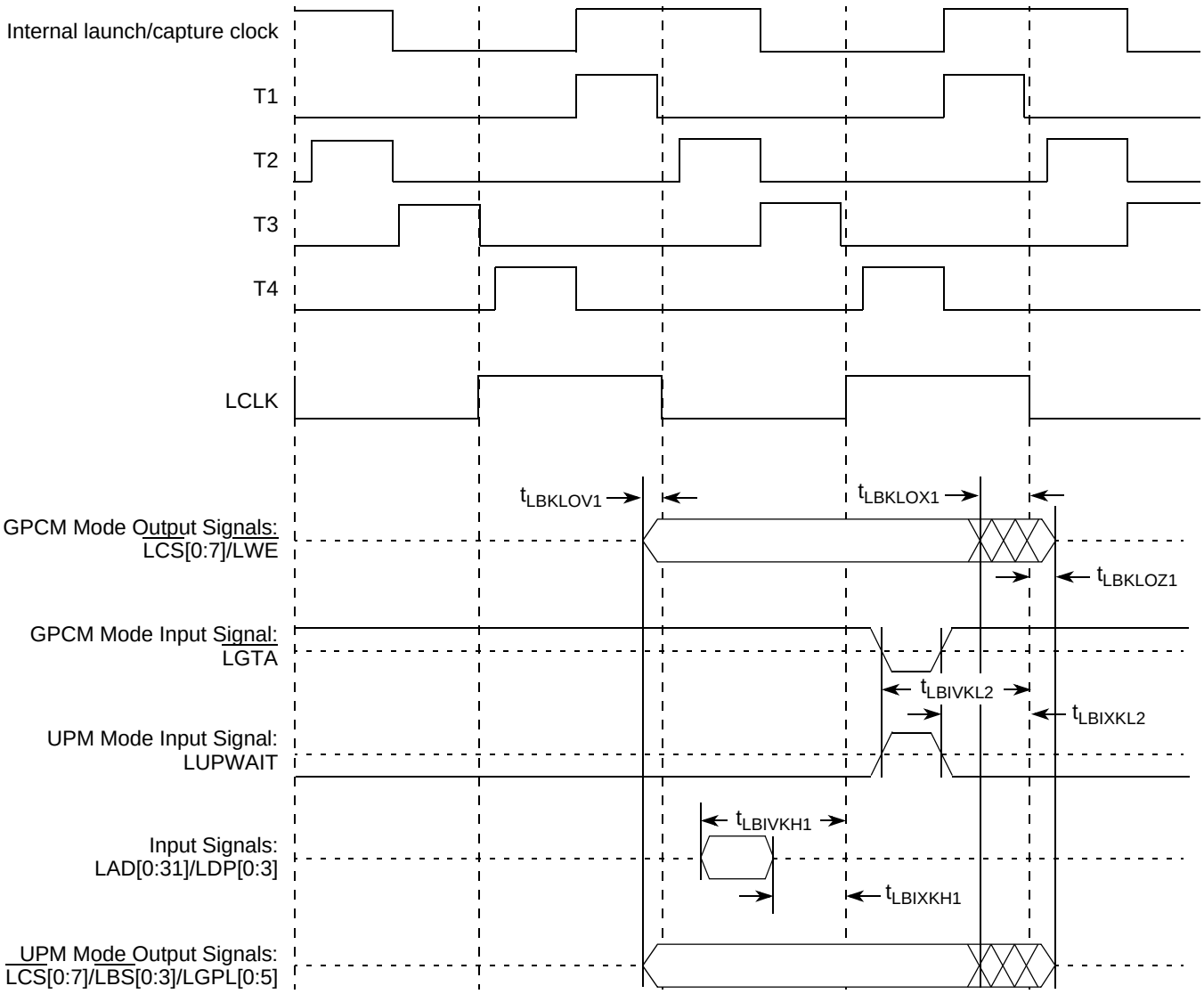

Figure 35. Local Bus Signals, GPCM/UPM Signals for LCCR[CLKDIV] = 8 or 16 (PLL Bypass Mode)

Table 55. I²C AC Electrical Specifications (continued)

At recommended operating conditions with OV_{DD} of 3.3 V ± 5%. All values refer to V_{IH} (min) and V_{IL} (max) levels (see Table 2).

Parameter	Symbol[1]	Min	Max	Unit
Capacitive load for each bus line	Cb	—	400	pF

Notes:

1. The symbols used for timing specifications herein follow the pattern $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{I2DVKH} symbolizes I²C timing (I2) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SXKL} symbolizes I²C timing (I2) for the time that the data with respect to the START condition (S) went invalid (X) relative to the t_{I2C} clock reference (K) going to the low (L) state or hold time. Also, t_{I2PVKH} symbolizes I²C timing (I2) for the time that the data with respect to the STOP condition (P) reaching the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time.
2. As a transmitter, the MPC8572E provides a delay time of at least 300 ns for the SDA signal (referred to the VIHmin of the SCL signal) to bridge the undefined region of the falling edge of SCL to avoid unintended generation of START or STOP condition. When the MPC8572E acts as the I2C bus master while transmitting, the MPC8572E drives both SCL and SDA. As long as the load on SCL and SDA are balanced, the MPC8572E would not cause unintended generation of START or STOP condition. Therefore, the 300 ns SDA output delay time is not a concern. If, under some rare condition, the 300 ns SDA output delay time is required for the MPC8572E as transmitter, application note AN2919 referred to in note 4 below is recommended.
3. The maximum t_{I2OVKL} has only to be met if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.
4. The requirements for I²C frequency calculation must be followed. Refer to Freescale application note AN2919, *Determining the I²C Frequency Divider Ratio for SCL*.

Figure 40 provides the AC test load for the I²C.

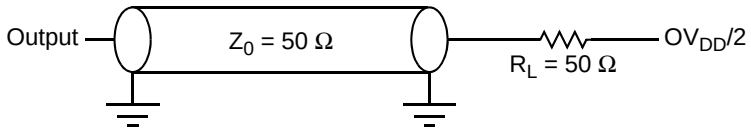

Figure 40. I²C AC Test Load

Figure 41 shows the AC timing diagram for the I²C bus.

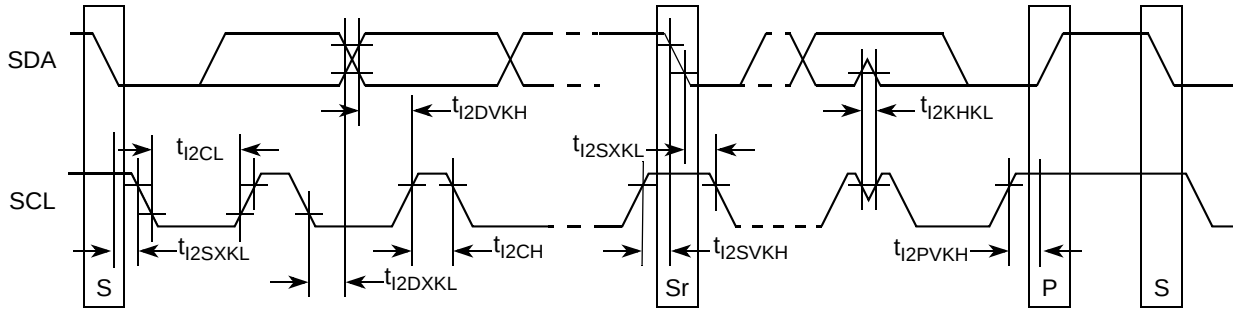

Figure 41. I²C Bus AC Timing Diagram

14 GPIO

This section describes the DC and AC electrical specifications for the GPIO interface of the MPC8572E.

14.1 GPIO DC Electrical Characteristics

Table 56 provides the DC electrical characteristics for the GPIO interface operating at BV_{DD} = 3.3 V DC.

Table 56. GPIO DC Electrical Characteristics (3.3 V DC)

Parameter	Symbol	Min	Max	Unit
Supply voltage 3.3V	BV_{DD}	3.13	3.47	V
High-level input voltage	V_{IH}	2	BV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.8	V
Input current (BV_{IN} [1] = 0 V or BV_{IN} = BV_{DD})	I_{IN}	—	±5	μA
High-level output voltage (BV_{DD} = min, I_{OH} = –2 mA)	V_{OH}	BV_{DD} – 0.2	—	V
Low-level output voltage (BV_{DD} = min, I_{OL} = 2 mA)	V_{OL}	—	0.2	V

Note:
1. Note that the symbol BV_{IN}, in this case, represents the BV_{IN} symbol referenced in Table 1.

Table 57 provides the DC electrical characteristics for the GPIO interface operating at BV_{DD} = 2.5 V DC.

Table 57. GPIO DC Electrical Characteristics (2.5 V DC)

Parameter	Symbol	Min	Max	Unit
Supply voltage 2.5V	BV_{DD}	2.37	2.63	V
High-level input voltage	V_{IH}	1.70	BV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.7	V
Input current (BV_{IN} [1] = 0 V or BV_{IN} = BV_{DD})	I_{IH}	—	10	μA
	I_{IL}		–15	
High-level output voltage (BV_{DD} = min, I_{OH} = –1 mA)	V_{OH}	2.0	BV_{DD} + 0.3	V
Low-level output voltage (BV_{DD} min, I_{OL} = 1 mA)	V_{OL}	GND – 0.3	0.4	V

Note:
1. The symbol BV_{IN}, in this case, represents the BV_{IN} symbol referenced in Table 1.

15 High-Speed Serial Interfaces (HSSI)

The MPC8572E features two Serializer/Deserializer (SerDes) interfaces to be used for high-speed serial interconnect applications. The SerDes1 interface can be used for PCI Express and/or Serial RapidIO data transfers. The SerDes2 is dedicated for SGMII application.

This section describes the common portion of SerDes DC electrical specifications, which is the DC requirement for SerDes Reference Clocks. The SerDes data lane's transmitter and receiver reference circuits are also shown.

15.1 Signal Terms Definition

The SerDes utilizes differential signaling to transfer data across the serial link. This section defines terms used in the description and specification of differential signals.

Figure 43 shows how the signals are defined. For illustration purpose, only one SerDes lane is used for description. The figure shows waveform for either a transmitter output (SDn_TX and $\overline{\text{SDn\_TX}}$) or a receiver input (SDn_RX and $\overline{\text{SDn\_RX}}$). Each signal swings between A Volts and B Volts where A > B.

Using this waveform, the definitions are as follows. To simplify illustration, the following definitions assume that the SerDes transmitter and receiver operate in a fully symmetrical differential signaling environment.

1. **Single-Ended Swing**

   The transmitter output signals and the receiver input signals SDn_TX, $\overline{\text{SDn\_TX}}$, SDn_RX and $\overline{\text{SDn\_RX}}$ each have a peak-to-peak swing of A - B Volts. This is also referred as each signal wire's Single-Ended Swing.

2. **Differential Output Voltage, V_{OD}** (or **Differential Output Swing**):

   The Differential Output Voltage (or Swing) of the transmitter, V_{OD}, is defined as the difference of the two complimentary output voltages: $V_{SDn\_TX} - V_{\overline{SDn\_TX}}$. The V_{OD} value can be either positive or negative.

3. **Differential Input Voltage, V_{ID}** (or **Differential Input Swing**):

   The Differential Input Voltage (or Swing) of the receiver, V_{ID}, is defined as the difference of the two complimentary input voltages: $V_{SDn\_RX} - V_{\overline{SDn\_RX}}$. The V_{ID} value can be either positive or negative.

4. **Differential Peak Voltage, V_{DIFFp}**

   The peak value of the differential transmitter output signal or the differential receiver input signal is defined as Differential Peak Voltage, $V_{DIFFp} = |A - B|$ Volts.

5. **Differential Peak-to-Peak, $V_{DIFFp-p}$**

   Because the differential output signal of the transmitter and the differential input signal of the receiver each range from A – B to –(A – B) Volts, the peak-to-peak value of the differential transmitter output signal or the differential receiver input signal is defined as Differential Peak-to-Peak Voltage, $V_{DIFFp-p} = 2*V_{DIFFp} = 2 * |(A - B)|$ Volts, which is twice of differential swing in amplitude, or twice of the differential peak. For example, the output differential peak-peak voltage can also be calculated as $V_{TX-DIFFp-p} = 2*|V_{OD}|$.

15.2.3 Interfacing With Other Differential Signaling Levels

- With on-chip termination to SGND_SRDS*n* (xcorevss), the differential reference clocks inputs are HCSL (High-Speed Current Steering Logic) compatible DC-coupled.
- Many other low voltage differential type outputs like LVDS (Low Voltage Differential Signaling) can be used but may need to be AC-coupled due to the limited common mode input range allowed (100 to 400 mV) for DC-coupled connection.
- LVPECL outputs can produce signal with too large amplitude and may need to be DC-biased at clock driver output first, then followed with series attenuation resistor to reduce the amplitude, additionally to AC-coupling.

NOTE

Figure 48 to Figure 51 below are for conceptual reference only. Due to the fact that clock driver chip's internal structure, output impedance and termination requirements are different between various clock driver chip manufacturers, it is very possible that the clock circuit reference designs provided by clock driver chip vendor are different from what is shown below. They might also vary from one vendor to the other. Therefore, Freescale Semiconductor can neither provide the optimal clock driver reference circuits, nor guarantee the correctness of the following clock driver connection reference circuits. The system designer is recommended to contact the selected clock driver chip vendor for the optimal reference circuits with the MPC8572E SerDes reference clock receiver requirement provided in this document.

Figure 49 shows the SerDes reference clock connection reference circuits for LVDS type clock driver. Because LVDS clock driver's common mode voltage is higher than the MPC8572E SerDes reference clock input's allowed range (100 to 400mV), AC-coupled connection scheme must be used. It assumes the LVDS output driver features 50-Ω termination resistor. It also assumes that the LVDS transmitter establishes its own common mode level without relying on the receiver or other external component.

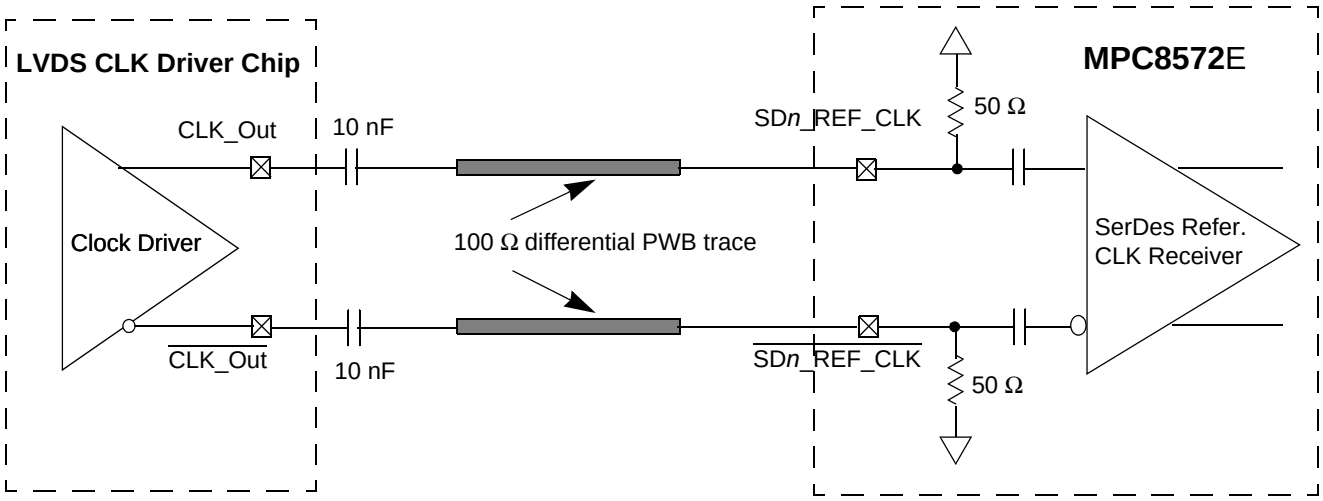

Figure 49. AC-Coupled Differential Connection with LVDS Clock Driver (Reference Only)

Figure 50 shows the SerDes reference clock connection reference circuits for LVPECL type clock driver. Because LVPECL driver's DC levels (both common mode voltages and output swing) are incompatible with MPC8572E SerDes reference clock input's DC requirement, AC-coupling must be used. Figure 50 assumes that the LVPECL clock driver's output impedance is 50Ω. R1 is used to DC-bias the LVPECL outputs prior to AC-coupling. Its value could be ranged from 140Ω to 240Ω depending on clock driver vendor's requirement. R2 is used together with the SerDes reference clock receiver's 50-Ω termination resistor to attenuate the LVPECL output's differential peak level such that it meets the MPC8572E SerDes reference clock's differential input amplitude requirement (between 200mV and 800mV differential peak). For example, if the LVPECL output's differential peak is 900mV and the desired SerDes reference clock input amplitude is selected as 600mV, the attenuation factor is 0.67, which requires R2 = 25Ω. Consult

Table 62. Differential Transmitter (TX) Output Specifications (continued)

Symbol	Parameter	Min	Nominal	Max	Units	Comments
$V_{TX-DC-CM}$	The TX DC Common Mode Voltage	0	—	3.6	V	The allowed DC Common Mode voltage under any conditions. See Note 6.
$I_{TX-SHORT}$	TX Short Circuit Current Limit		—	90	mA	The total current the Transmitter can provide when shorted to its ground
$T_{TX-IDLE-MIN}$	Minimum time spent in Electrical Idle	50	—	—	UI	Minimum time a Transmitter must be in Electrical Idle Utilized by the Receiver to start looking for an Electrical Idle Exit after successfully receiving an Electrical Idle ordered set
$T_{TX-IDLE-SET-TO-IDLE}$	Maximum time to transition to a valid Electrical idle after sending an Electrical Idle ordered set	—	—	20	UI	After sending an Electrical Idle ordered set, the Transmitter must meet all Electrical Idle Specifications within this time. This is considered a debounce time for the Transmitter to meet Electrical Idle after transitioning from L0.
$T_{TX-IDLE-TO-DIFF-DATA}$	Maximum time to transition to valid TX specifications after leaving an Electrical idle condition	—	—	20	UI	Maximum time to meet all TX specifications when transitioning from Electrical Idle to sending differential data. This is considered a debounce time for the TX to meet all TX specifications after leaving Electrical Idle
$RL_{TX-DIFF}$	Differential Return Loss	12	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4
RL_{TX-CM}	Common Mode Return Loss	6	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4
$Z_{TX-DIFF-DC}$	DC Differential TX Impedance	80	100	120	Ω	TX DC Differential mode Low Impedance
Z_{TX-DC}	Transmitter DC Impedance	40	—	—	Ω	Required TX D+ as well as D- DC Impedance during all states
$L_{TX-SKEW}$	Lane-to-Lane Output Skew	—	—	500 + 2 UI	ps	Static skew between any two Transmitter Lanes within a single Link
C_{TX}	AC Coupling Capacitor	75	—	200	nF	All Transmitters shall be AC coupled. The AC coupling is required either within the media or within the transmitting component itself. See Note 8.

Table 62. Differential Transmitter (TX) Output Specifications (continued)

Symbol	Parameter	Min	Nominal	Max	Units	Comments
$T_{crosslink}$	Crosslink Random Timeout	0	—	1	ms	This random timeout helps resolve conflicts in crosslink configuration by eventually resulting in only one Downstream and one Upstream Port. See Note 7.

Notes:
1. No test load is necessarily associated with this value.
2. Specified at the measurement point into a timing and voltage compliance test load as shown in Figure 57 and measured over any 250 consecutive TX UIs. (Also refer to the transmitter compliance eye diagram shown in Figure 55.)
3. A T_{TX-EYE} = 0.70 UI provides for a total sum of deterministic and random jitter budget of $T_{TX-JITTER-MAX}$ = 0.30 UI for the Transmitter collected over any 250 consecutive TX UIs. The $T_{TX-EYE-MEDIAN-to-MAX-JITTER}$ median is less than half of the total TX jitter budget collected over any 250 consecutive TX UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value.
4. The Transmitter input impedance shall result in a differential return loss greater than or equal to 12 dB and a common mode return loss greater than or equal to 6 dB over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements is 50 ohms to ground for both the D+ and D- line (that is, as measured by a Vector Network Analyzer with 50 ohm probes—see Figure 57). Note that the series capacitors C_{TX} is optional for the return loss measurement.
5. Measured between 20-80% at transmitter package pins into a test load as shown in Figure 57 for both V_{TX-D+} and V_{TX-D-}.
6. See Section 4.3.1.8 of the PCI Express Base Specifications Rev 1.0a.
7. See Section 4.2.6.3 of the PCI Express Base Specifications Rev 1.0a.
8. MPC8572E SerDes transmitter does not have C_{TX} built-in. An external AC Coupling capacitor is required.

16.4.2 Transmitter Compliance Eye Diagrams

The TX eye diagram in Figure 55 is specified using the passive compliance/test measurement load (see Figure 57) in place of any real PCI Express interconnect + RX component.

There are two eye diagrams that must be met for the transmitter. Both eye diagrams must be aligned in time using the jitter median to locate the center of the eye diagram. The different eye diagrams differ in voltage depending whether it is a transition bit or a de-emphasized bit. The exact reduced voltage level of the de-emphasized bit is always relative to the transition bit.

The eye diagram must be valid for any 250 consecutive UIs.

A recovered TX UI is calculated over 3500 consecutive unit intervals of sample data. The eye diagram is created using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the TX UI.

NOTE

It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function (that is, least squares and median deviation fits).

16.5.1 Compliance Test and Measurement Load

The AC timing and voltage parameters must be verified at the measurement point, as specified within 0.2 inches of the package pins, into a test/measurement load shown in Figure 57.

NOTE

The allowance of the measurement point to be within 0.2 inches of the package pins is meant to acknowledge that package/board routing may benefit from D+ and D– not being exactly matched in length at the package pin boundary.

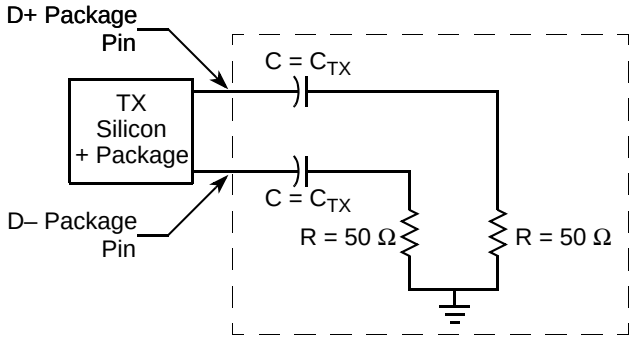

Figure 57. Compliance Test/Measurement Load

17 Serial RapidIO

This section describes the DC and AC electrical specifications for the RapidIO interface of the MPC8572E for the LP-Serial physical layer. The electrical specifications cover both single and multiple-lane links. Two transmitters (short run and long run) and a single receiver are specified for each of three baud rates, 1.25, 2.50, and 3.125 GBaud.

Two transmitter specifications allow for solutions ranging from simple board-to-board interconnect to driving two connectors across a backplane. A single receiver specification is given that accepts signals from both the short run and long run transmitter specifications.

The short run transmitter should be used mainly for chip-to-chip connections on either the same printed circuit board or across a single connector. This covers the case where connections are made to a mezzanine (daughter) card. The minimum swings of the short run specification reduce the overall power used by the transceivers.

The long run transmitter specifications use larger voltage swings that are capable of driving signals across backplanes. This allows a user to drive signals across two connectors and a backplane. The specifications allow a distance of at least 50 cm at all baud rates.

All unit intervals are specified with a tolerance of +/– 100 ppm. The worst case frequency difference between any transmit and receive clock is 200 ppm.

To ensure interoperability between drivers and receivers of different vendors and technologies, AC coupling at the receiver input must be used.

Table 76. MPC8572E Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
SD1_RX[7:0]	Receive Data (positive)	P32, N30, M32, L30, G30, F32, E30, D32	I	XV_{DD_SRDS1}	—
$\overline{\text{SD1\_RX}}$[7:0]	Receive Data (negative)	P31, N29, M31, L29, G29, F31, E29, D31	I	XV_{DD_SRDS1}	—
SD1_TX[7]	PCIe1 Tx Data Lane 7 / SRIO or PCIe2 Tx Data Lane 3 / PCIe3 TX Data Lane 1	M26	O	XV_{DD_SRDS1}	—
SD1_TX[6]	PCIe1 Tx Data Lane 6 / SRIO or PCIe2 Tx Data Lane 2 / PCIe3 TX Data Lane 0	L24	O	XV_{DD_SRDS1}	—
SD1_TX[5]	PCIe1 Tx Data Lane 5 / SRIO or PCIe2 Tx Data Lane 1	K26	O	XV_{DD_SRDS1}	—
SD1_TX[4]	PCIe1 Tx Data Lane 4 / SRIO or PCIe2 Tx Data Lane 0	J24	O	XV_{DD_SRDS1}	—
SD1_TX[3]	PCIe1 Tx Data Lane 3	G24	O	XV_{DD_SRDS1}	—
SD1_TX[2]	PCIe1 Tx Data Lane 2	F26	O	XV_{DD_SRDS1}	—
SD1_TX[1]	PCIe1 Tx Data Lane 1]	E24	O	XV_{DD_SRDS1}	—
SD1_TX[0]	PCIe1 Tx Data Lane 0	D26	O	XV_{DD_SRDS1}	—
$\overline{\text{SD1\_TX}}$[7:0]	Transmit Data (negative)	M27, L25, K27, J25, G25, F27, E25, D27	O	XV_{DD_SRDS1}	—
SD1_PLL_TPD	PLL Test Point Digital	J32	O	XV_{DD_SRDS1}	17
SD1_REF_CLK	PLL Reference Clock	H32	I	XV_{DD_SRDS1}	—
$\overline{\text{SD1\_REF\_CLK}}$	PLL Reference Clock Complement	H31	I	XV_{DD_SRDS1}	—
Reserved	—	C29, K32	—	—	26
Reserved	—	C30, K31	—	—	27
Reserved	—	C24, C25, H26, H27	—	—	28
Reserved	—	AL20, AL21	—	—	29
SerDes (x4) SGMII					
SD2_RX[3:0]	Receive Data (positive)	AK32, AJ30, AF30, AE32	I	XV_{DD_SRDS2}	—

Table 79. CCB Clock Ratio

Binary Value of LA[29:31] Signals	CCB:SYSCLK Ratio
000	4:1
001	5:1
010	6:1
011	8:1
100	10:1
101	12:1
110	Reserved
111	Reserved

19.3 e500 Core PLL Ratio

The clock speed for each e500 core can be configured differently, determined by the values of various signals at power up.

Table 80 describes the clock ratio between e500 Core0 and the e500 core complex bus (CCB). This ratio is determined by the binary value of LBCTL, LALE and LGPL2/$\overline{\text{LOE}}$/$\overline{\text{LFRE}}$ at power up, as shown in Table 80.

Table 80. e500 Core0 to CCB Clock Ratio

Binary Value of LBCTL, LALE, LGPL2/LOE/LFRE Signals	e500 Core0:CCB Clock Ratio	Binary Value of LBCTL, LALE, LGPL2/LOE/LFRE Signals	e500 Core0:CCB Clock Ratio
000	Reserved	100	2:1
001	Reserved	101	5:2 (2.5:1)
010	Reserved	110	3:1
011	3:2 (1.5:1)	111	7:2 (3.5:1)

Table 82. DDR Clock Ratio (continued)

Binary Value of TSEC_1588_CLK_OUT, TSEC_1588_PULSE_OUT1, TSEC_1588_PULSE_OUT2 Signals	DDR:DDRCLK Ratio
101	12:1
110	14:1
111	Synchronous mode

19.5 Frequency Options

19.5.1 Platform to Sysclk Frequency Options

Table 83 shows the expected frequency values for the platform frequency when using the specified CCB clock to SYSCLK ratio.

Table 83. Frequency Options for Platform Frequency

CCB to SYSCLK Ratio	SYSCLK (MHz)							
	33.33	41.66	50	66.66	83	100	111	133.33
	Platform /CCB Frequency (MHz)							
4						400	444	533
5					415	500	555	
6				400	498	600		
8			400	533				
10		417	500					
12	400	500	600					

19.5.2 Minimum Platform Frequency Requirements for High-Speed Interfaces

Section 4.4.3.6, “I/O Port Selection,” in the *MPC8572E PowerQUICC III Integrated Host Processor Family Reference Manual* describes various high-speed interface configuration options. Note that the CCB clock frequency must be considered for proper operation of such interfaces as described below.

For proper PCI Express operation, the CCB clock frequency must be greater than or equal to:

$$\frac{527\ \text{MHz} \times (\text{PCI Express link width})}{8}$$

See Section 21.1.3.2, “Link Width,” in the *MPC8572E PowerQUICC III Integrated Host Processor Family Reference Manual* for PCI Express interface width details. Note that the “PCI Express link width”

[3] Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by part number specifications may support other maximum core frequencies.

22.2 Part Marking

Parts are marked as the example shown in Figure 67.

FC-PBGA

Notes:
MMMMMM is the 6-digit mask number.
ATWLYYWW is the traceability code.
CCCCC is the country of assembly. This space is left blank if parts are assembled in the United States.

Figure 67. Part Marking for FC-PBGA Device

Table 89 explains line four of Figure 67.

Table 89. Meaning of Last Line of Part Marking

Digit	Description
A	Assembly Site E Oak Hill Q KLM
WL	Lot number
YY	Year assembled
WW	Work week assembled

23 Document Revision History

Table 90 provides a revision history for the MPC8572E hardware specification.

Table 90. Document Revision History

Rev. Number	Date	Substantive Change(s)
7	03/2016	• Updated Section 22.2, “Part Marking,” changed the five-digit mask number to six digits.